LT3045EDD
20V, 500mA, Ultralow Noise Ultrahigh PSRR RF LDO Regulator

DESCRIPTION

Demo circuit DC2491A features the LT®3045EDD, a 500mA, ultralow noise and ultrahigh power supply rejection ratio (PSRR) low dropout (LDO) regulator with programmable current limit.

DC2491A operates over an input voltage range of 3.8V to 20V. The LT3045 delivers a maximum output current of 500mA. In addition to featuring ultralow noise and ultrahigh PSRR, the regulator offers programmable power good functionality as well as programmable current limit. Current monitoring is also achieved by sensing the ILIM pin voltage.

Built-in protection includes reverse battery protection, reverse current protection, internal current limit with foldback, and thermal limit with hysteresis.

The LT3045 data sheet gives a complete description of the device, operation and applications information. The data sheet must be read in conjunction with this demo manual for DC2491A. The LT3045EDD is assembled in a 10-lead (3mm × 3mm) plastic DFN package with an exposed pad on the bottom-side of the IC. Proper board layout is essential for maximum thermal performance.

Design files for this circuit board are available at http://www.linear.com/demo/DC2491A

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PERFORMANCE SUMMARY

Specifications are at T_A = 25°C

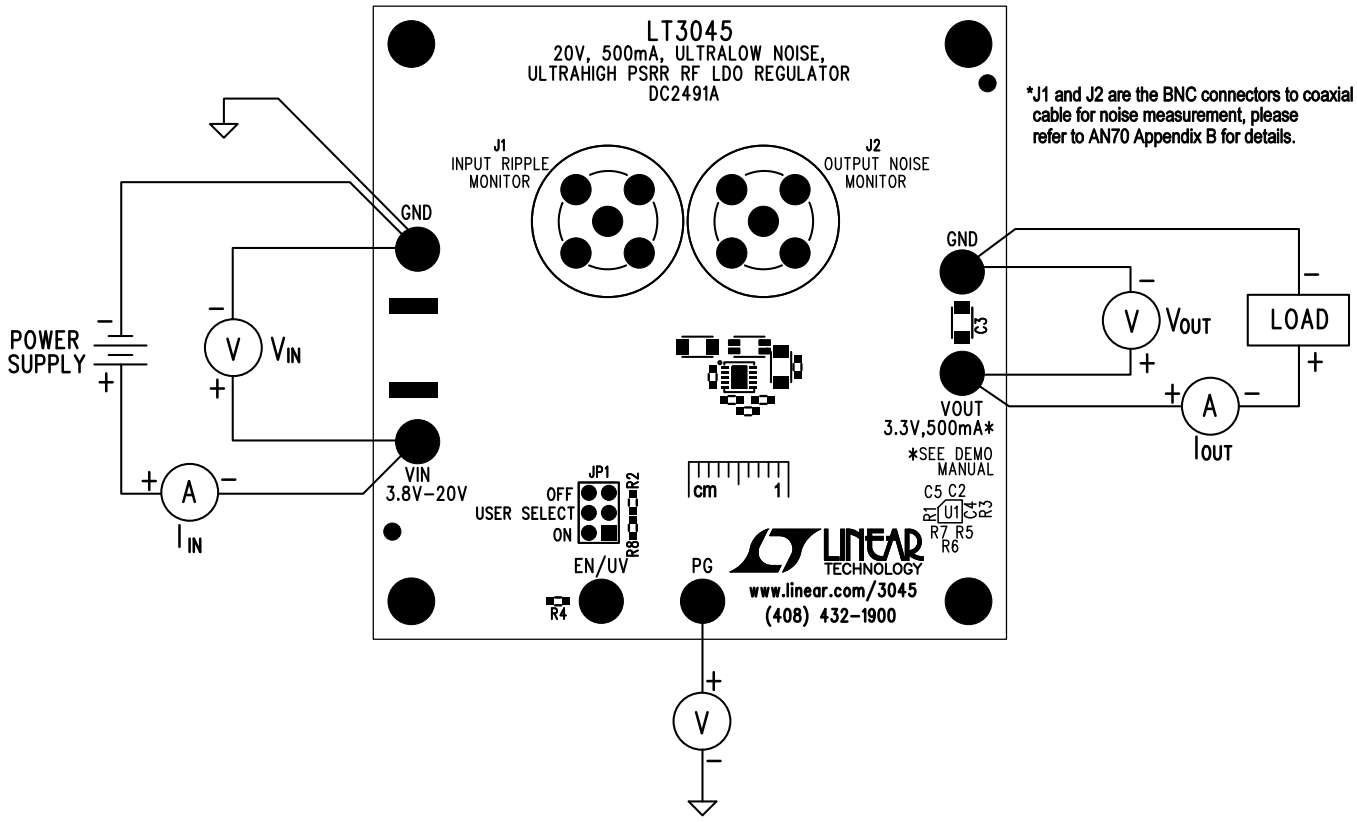
PARAMETER	CONDITIONS	MIN	TYP	MAX
Input Voltage Range (V_{IN})	I_{OUT} = 150mA, V_{OUT} = 3.3V	3.8V		20V
Input Voltage Range (V_{IN})	I_{OUT} = 500mA, V_{OUT} = 3.3V	3.8V		8.3V*
Output Voltage (V_{OUT})	V_{IN} = 5V, I_{OUT} = 500mA	3.2V	3.3V	3.4V
Shutdown Input Current (I_{IN})	JP1 = OFF, V_{IN} = 5V		0.1µA	

*The maximum input voltage for 500mA load current is set by the 60°C temperature rise of LT3045 on the demo circuit. Higher input voltage can be reached if larger copper area or force-air cooling is applied. The output current is also limited by the differential of input and output voltage, please refer the data sheet for details.

QUICK START PROCEDURE

DC2491A is easy to set up to evaluate the performance of the LT3045EDD. Refer to Figure 1 for proper measurement equipment setup and follow the procedure below.

1. Connect load between V_{OUT} and GND terminals.
2. With power off, connect the input power supply to the V_{IN} and GND terminals.
3. Make sure the shunt of JP1 is at ON option.
4. Turn the input power supply on and make sure the voltage is between 3.8V and 20V.
5. Refer to Application Notes AN70 and AN159 for measuring output noise and PSRR.
6. With JP1 at USER SELECT option, R6 and R7 can be used to set an accurate undervoltage lockout (UVLO) threshold.



*The maximum output current will be limited by internal current limit based on differential voltage of input and output voltage, please refer to the data sheet.

Figure 1. Test Procedure Setup Drawing for DC2491A

PCB LAYOUT

Best PSRR Performance: PCB Layout for Input Trace

For applications utilizing the LT3045 for post-regulating switching converters, placing a capacitor directly at the LT3045 input results in AC current (at the switching frequency) to flow near the LT3045. Without careful attention to PCB layout, this relatively high frequency switching current generates an electromagnetic field (EMF) that couples to the LT3045 output, thereby degrading its effective PSRR. While highly dependent on the PCB, the switching preregulator, and the input capacitor size, among other factors, the PSRR degradation can easily be 30dB at 1MHz. This degradation is present even if the LT3045 is de-soldered from the board, because it effectively degrades the PSRR of the PC board itself. While negligible for conventional low PSRR LDOs, LT3045's ultrahigh PSRR requires careful attention to higher order parasitics in order to realize the full performance offered by the regulator.

The LT3045 demo board alleviates this degradation in PSRR by using a specialized layout technique. On layer 3, the input trace (V_{IN}) is highlighted in red, with the return path (GND) highlighted on the bottom layer together with input capacitor C1. When an AC voltage is applied to the input of the board, AC current flows on this path, thus generating EMF. This EMF couples to output capacitor C2 and related traces, making the PSRR appear worse than it actually is. **With the input trace directly above the return path, the EMFs are in opposite directions, and consequently cancel each other out. Making sure these traces exactly overlap each other maximizes the cancellation effect and thus provides the maximum PSRR offered by the regulator.**

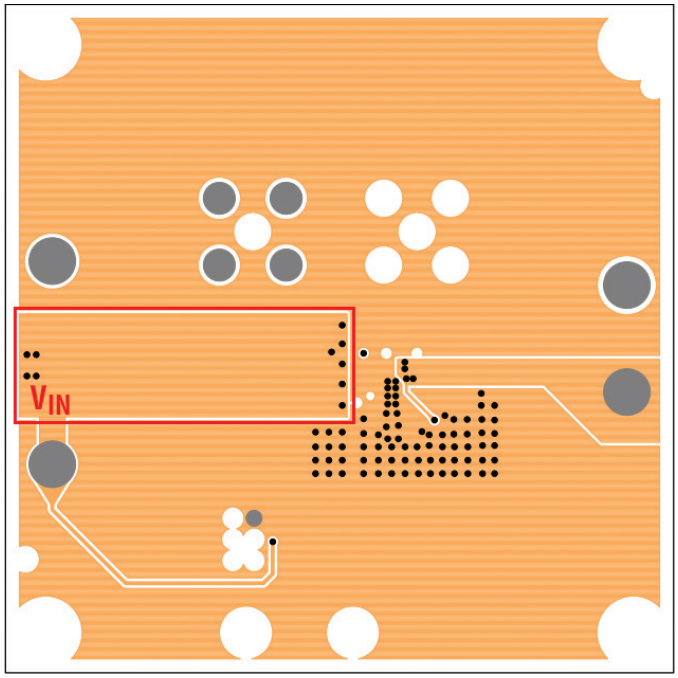


Figure 2. Layer 3 of DC2491A

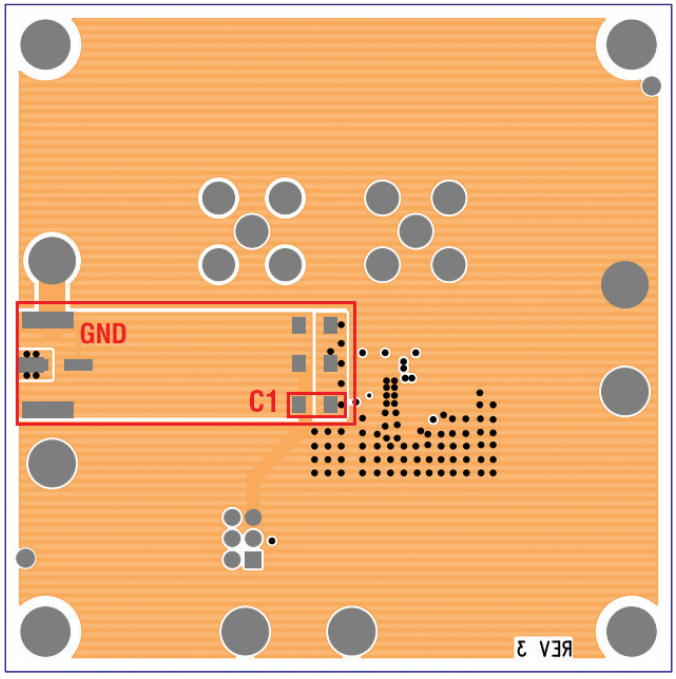


Figure 3. Bottom Layer of DC2491A

PCB LAYOUT

Best AC Performance: PCB Layout for Output Capacitor C2

For ultrahigh PSRR performance, the LT3045 bandwidth is made quite high (~1MHz), making it very close to the output capacitor's self-resonance frequency (~1.6MHz). Therefore, it is very important to avoid adding extra impedance (ESL and ESR) outside the feedback loop. To that end, minimize the effects of PCB trace and solder inductance by Kelvin connecting OUTS and SET pin capacitor (C_{SET}) GND directly to output capacitor (C2) terminals using split capacitor techniques. Pad 4 connects to the OUTS pin and Pad 1 connects to the GND side of the SET pin capacitor. With only small AC current flowing through these connections, the impact of solder joint/PCB trace inductance on stability is eliminated. While the LT3045 is robust enough not to oscillate if the recommended layout is not followed, phase/gain margin and PSRR will degrade.

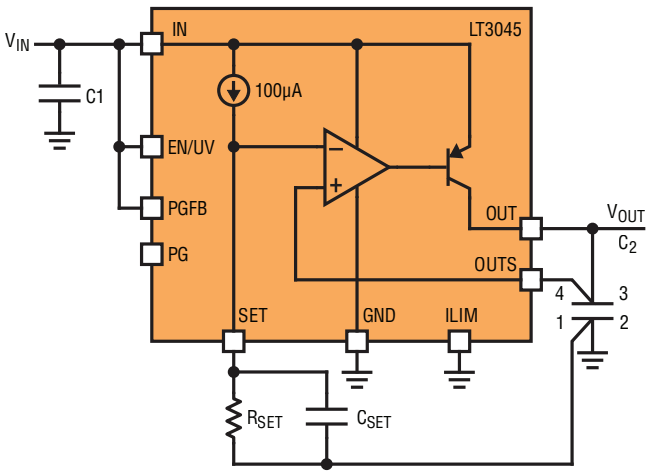


Figure 4. C2 and C_{SET} Connections for Best Performance

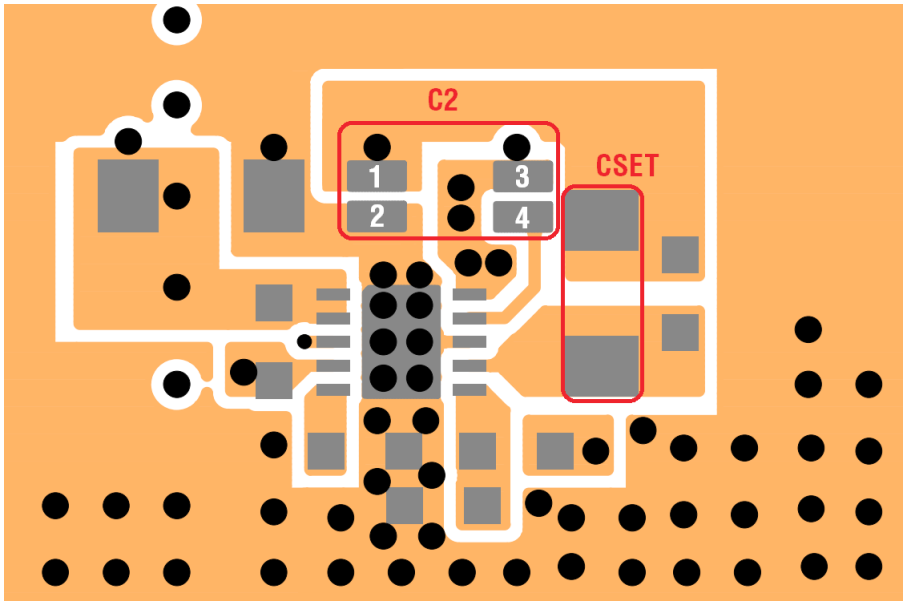
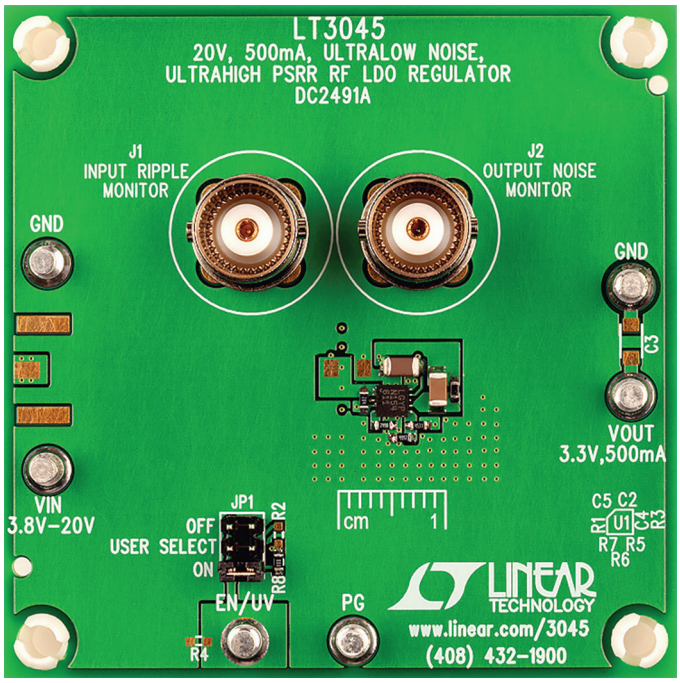


Figure 5. Split Pads for C2 on Top Layer of DC2491A

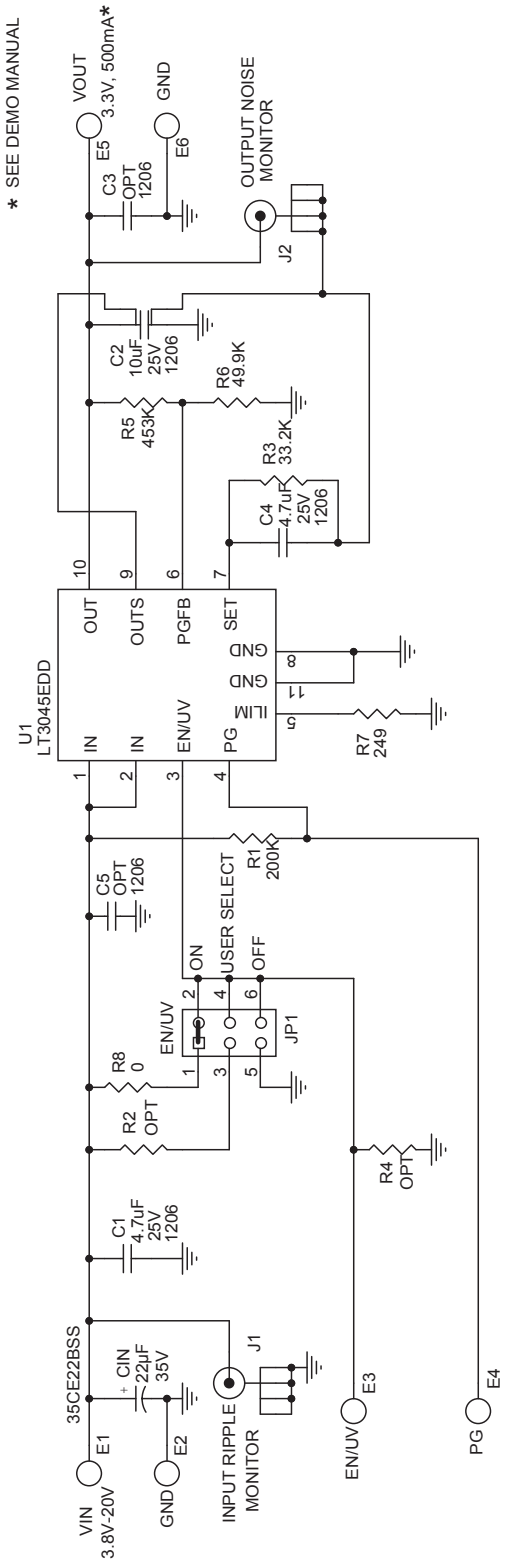
BOARD PHOTO


LT3045
20V, 500mA, ULTRALOW NOISE,
ULTRAHIGH PSRR RF LDO REGULATOR
DC2491A
J1
INPUT RIPPLE
MONITOR
J2
OUTPUT NOISE
MONITOR
GND
GND
C3
VOUT
3.3V,500mA
VIN
3.8V–20V
JP1
OFF
USER SELECT
ON
R2
R8
EN/UV
R4
PG
C5 C2
R1
U1
C4
R3
R7 R5
R6
cm
1
LINEAR
TECHNOLOGY
www.linear.com/3045
(408) 432-1900

PARTS LIST

ITEM	QTY	REFERENCE	PART DESCRIPTION	MANUFACTURER/PART NUMBER
Required Circuit Components				
1	1	CIN	CAP., ALUM, 22µF, 35V, 5mm × 5.4mm	SUN ELECTRONIC INDUSTRIES CORP., 35CE22BSS
2	2	C1, C4	CAP., X7R, 4.7µF, 25V, 10% 1206	MURATA, GRM31CR71E475KA88L
3	1	C2	CAP., X5R, 10µF, 25V, 10% 1206	MURATA, GJ831CR61E106KE83L
4	1	R1	RES., CHIP, 200k, 1/16W, 5% 0603	VISHAY, CRCW0603200KJNED
5	1	R3	RES., CHIP, 33.2k, 1/8W, 1% 0603	VISHAY, CRCW060333K2FKEA
6	1	R5	RES., CHIP, 453k, 1/8W, 1% 0603	VISHAY, CRCW0603453KFKEA
7	1	R6	RES., CHIP, 49.9k, 1/8W, 1% 0603	VISHAY, CRCW060349K9FKEA
8	1	R7	RES., CHIP, 249Ω, 1/16W, 1% 0603	VISHAY, CRCW0603249RFKEA
9	1	U1	IC, LT3045EDD DFN 3mm × 3mm	LINEAR TECH., LT3045EDD#PBF
Optional Electronic Components				
1	0	C3, C5 (OPT)	CAP., 1206	
2	0	R2, R4 (OPT)	RES., 0603	
3	1	R8	RES., CHIP, 0Ω, 1/16W, 5% 0603	VISHAY, CRCW06030000Z0EA
Hardware				
1	6	E1 – E6	TESTPOINT, TURRET, 0.094" PBF	MILL-MAX, 2501-2-00-80-00-00-07-0
2	1	JP1	HEADER 3-PIN 0.079" DOUBLE ROW	WURTH ELEKTRONIK, 62000621121
3	1	XJP1	SHUNT, 0.079" CENTER	WURTH ELEKTRONIK, 60800213421
4	2	J1, J2	CONN, BNC, 5 PINS	CONNEX, 112404
5	4	MH1 – MH4	STAND-OFF, NYLON 6.4mm	WURTH ELEKTRONIK, 702931000

SCHEMATIC DIAGRAM



NOTE: UNLESS OTHERWISE SPECIFIED

1. ALL RESISTORS ARE 0603.

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