

bq2403x (bqTiny-III™) 1.5-A Single-Chip Li-Ion and Li-Pol Charge Management IC EVM

This user's guide describes the bq2403x (bqTiny-III™) Evaluation Module. The EVM provides a convenient method for evaluating the performance of a charge management and system power solution for portable applications using the bq2403x product family. A completely designed and tested module is presented. The charger is designed to deliver up to 1.5 A of continuous current to the system or charger for one-cell Li-ion or Li-polymer applications (see the data sheet for correct device, x) using a dc power supply. The charger is programmed from the factory to deliver 1 A of charging current.

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1 Introduction

The bqTiny-III powers the system while independently charging the battery. This feature reduces the charge and discharge cycles on the battery, allows for proper charge termination, and allows the system to run with an absent or defective battery pack. This feature also allows for the system to instantaneously turn on from an external power source even when using a deeply discharged battery pack.

The bqTiny-III automatically selects the USB port or the ac adapter as the power source for the system. In the USB configuration, the host can select from the two preset input maximum rates of 100 mA and 500 mA. The bqTiny-III dynamically adjusts the USB charge rate based on the system load to stay within the 100-mA or 500-mA maximum rates. The AC pin can be programmed to perform like a USB input by pulling the PSEL pin low. An external resistor, RSET1, sets the magnitude of the charge current. If the charge current exceeds the available input current, the voltage on the OUT pin drops to the $DPPM_{OUT}$ threshold or the battery voltage, whichever is higher. The charging current is reduced to what current is available ($I_{BAT} = I_{IN} - I_{OUT}$).

The bqTiny-III charges the battery in three phases: conditioning, constant current, and constant voltage. Charge is terminated based on minimum current. A resistor-programmable charge timer provides a backup safety for charge termination. The bqTiny-III automatically re-starts the charge if the battery voltage falls below an internal threshold. The bqTiny-III automatically enters sleep mode when both supplies are removed (a drop to the battery voltage).

2 Considerations When Testing and Using bq2403x ICs

Consider the following noteworthy items while testing and using the bq2403x ICs.

There are three ICs (bq24030/1/2/5) to select from. The significant difference is the OUT pin voltage. The bq24030/1 has a LDO 6-V regulator connected the OUT pin. The intent on this charger is to provide a solution for using an inexpensive unregulated adaptor to power the charger. When unloaded, a 5-VDC adapter peak charges to ~8 V, but when loaded it quickly drops closer to 5 V. Since this is a linear regulator/charger, the higher the input voltage the lower the current level has to be to not exceed the power rating of the IC. The bq24032 has a LDO 4.4-V regulator connected to the OUT pin. The bq24035 shuts down if the input exceeds ~6.4 VDC.

The three potential sources to power the system (V_{OUT}) are: AC (AC-to-DC adapter), USB port, and battery. The IC is designed to power the system continuously. The battery, in most cases, is the last line of backup. If the AC or USB inputs are not available (or disabled), the battery connects to the system.

In thermal regulation condition ($T_J=125^{\circ}\text{C}$ —not a first-choice design mode of operation), the charge current is reduced to the battery, and the system still gets its power from the input. The battery supplement is still available in thermal regulation if the V_{OUT} falls to V_{BAT} . In thermal cutoff ($\sim 155^{\circ}\text{C}$), the input sources are disconnected, but the internal battery FET connects the battery to V_{OUT} .

The battery FET only opens (when needed) if a short on the V_{OUT} pulls more than 4 A of current, or any condition has V_{OUT} less than 1 Vdc (considered a short-circuit condition). In the short-circuit condition are two types of *pullups*: a 500- Ω resistor from each input to V_{OUT} and a 10-mA current source from the battery to the V_{OUT} . The system load has to be reduced ($>200\ \Omega$) on the output to allow V_{OUT} to rise above 1 Vdc. If the voltage on the DPPM pin is held below 1 V, then the short-circuit feature is disabled. Therefore, placing a small capacitor (~1000 pF) across the DPPM resistor delays the short-circuit protection on input powerup by a few microseconds. Typically, the system does not have much of a load below 1 V; so, powering up during a potential short-circuit condition usually is not a problem. V_{OUT} is always powered if there is any source voltage; so, dropping below 1 V is not a typical mode.

Another feature that protects system integrity is dynamic power path management (DPPM). The voltage on the DPPM pin ($DPPM_{IN}$) times a scaling factor of ~1.15 is the $DPPM_{OUT}$ voltage. The $DPPM_{OUT}$ voltage is the critical voltage, determined by the designer, where battery charging current is reduced to keep the system voltage (V_{OUT}) from further decay. A special feature to keep in mind is that when in DPPM mode the internal oscillator timer is slowed proportionally to how much the programmed charger current is reduced. This allows the timers (safety and others) to be appropriately adjusted during operation. Therefore, when performing any test where time is measured, keep in mind this adjustment factor.

Another critical feature is power handoff. The power handoff is initiated autonomously or by request. The PSEL (High/Low) sets which input source takes priority (AC or USB). This handoff happens immediately on request. The CE pin (going high) immediately enables the chip; disabling it (going low) delays handoff for 5 ms. For autonomous power selection (e.g., the selected source is lost), the IC switches sources when the Power Good (PG) status indicates the primary selected source is no longer good. PG is defined as $>(V_{BAT} + 80 \text{ mV})$. This means that if the battery is dead (missing, or discharged below a useable system voltage), the IC switches over to the other available source when the V_{OUT} reaches the *dead battery* voltage. This design feature prevents cycling between a stronger current-limiting source and the USB source. In most situations, if AC power is available (prior to losing it), the battery probably would not be discharged.

3 Performance Specification Summary

Table 1 summarizes the performance specifications of the EVM.

Table 1. Performance Specification Summary For bq24030/1/2/5 EVMs

SPECIFICATION	TEST CONDITIONS	MIN	TYP	MAX	UNITS
Input DC Voltage, $V_{I(AC)}$		4.8 V	5.0	6.5	Volts
Input DC USB Voltage, $V_{I(USB)}$			5.0		
Battery Charge Current, $I_{O(CHG)}$			1.0	2.0+	Amperes
Power Dissipation, bq2403x IC, 1 Cell	$P_{diss} = (V_{in} - V_{out})I_{out} + (V_{in} - V_{bat})I_{bat} + (V_{in} - V_{ldo})I_{LDO}$			see ⁽¹⁾	Watts

⁽¹⁾ The HPA073 (bq2403x) thermal design is optimized (8+ vias, 0.031-inch PWB, 2 oz. copper) to give $\theta_{JA} \sim 27^\circ\text{C/W}$.

4 Test Summary

This section covers the setup and tests performed in evaluating the EVM.

4.1 Equipment

- Power supply ($+5.25 \pm 0.25 \text{ VDC}$), current limit set to $2.0 \text{ A} \pm 0.2 \text{ A}$ for AC input to the UUT
- USB high-power port (500 mA) and cable (J1 is an alternate USB connection)
- Three Fluke 75 DMMs (equivalent or better)
- Oscilloscope, Model TDS220 (equivalent or better)

4.2 Equipment Setup

- Preset the UUT power supply voltage and current prior to connection to UUT; turn off the power supply and connect the supply to J2-AC/GND (+ to AC and – to GND).
- Connect a 10- Ω load to J7-OUT/GND.
- Connect a 1k- Ω load to J5-LDO/GND.
- Connect a fully discharged ($< 2.8\text{-VDC}$) single-cell Li-ion or Li-polymer battery to J8-BAT+/BAT–.
- Connect the DMMs as shown in Figure 1.

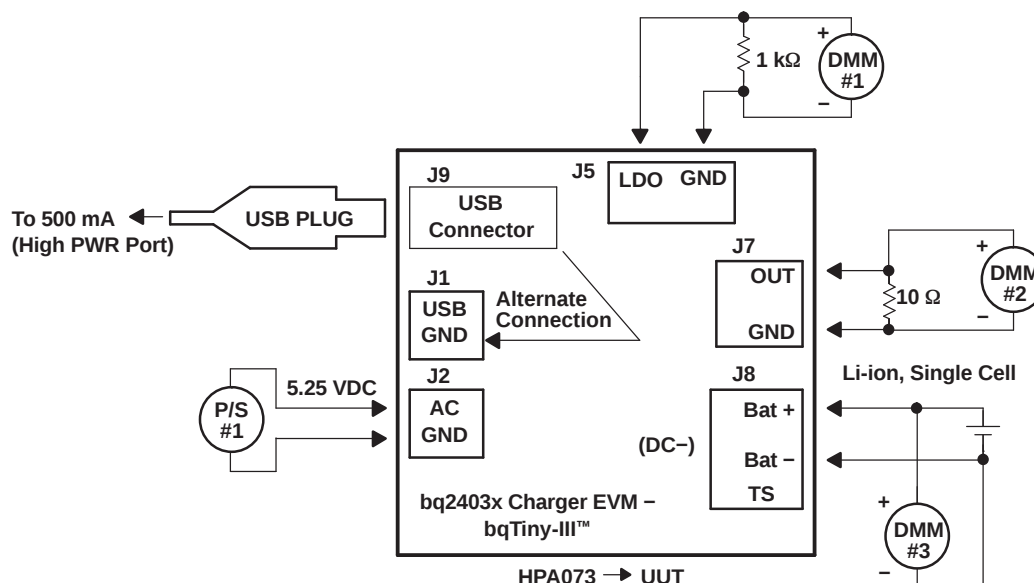


Figure 1. Test Diagram

4.3 Test Procedure

1. Verify that the equipment is set up according to Equipment Setup section.
2. Set jumpers on the UUT as follows: JMP1-0.5; JMP2-AC; JMP3-EN; set JMP4 through JMP7 to LED.
3. Adjust **R_{DPPM}** until TP1 is **31 kΩ ± 0.1 kΩ** with respect to GND, and adjust **R_{TMR}** until TP2 is 50 kΩ with respect to GND.
4. Verify that **V_{OUT}** is approximately equal to **V_{BAT}** (if **V_{out}** < 1.1 V, the output is in short-circuit mode. To get out of this mode, momentarily disconnect the 10-Ω load, or touch a 1-μF capacitor between the DPPM pin and ground).
5. Power up the **+5.25-VDC Supply** to the UUT.
6. Verify **V_{BAT}** is between 2.4 VDC and 3.0 VDC, and the charger is in pre-charge state: LEDs STAT1 (D2), STAT2 (D3), and ACPG (D5) are on.
If **V_{BAT}** is above the low-voltage threshold (**V_{LOWV}** ~3 V), then the IC is in fast-charge mode {STAT2 (D3) is off (High)}. If the IC is in fast charge, skip step 10.
7. Verify **I_{BAT}** is ~0.1 A (**I_{BAT}** = **I_{AC}** - (**V_{OUT}** / **R_{OUT}**) - 0.01 A)
8. Verify **V_{OUT}** is between 4.3 VDC and 4.5 VDC for the bq24032 IC. The bq24030/1/5 switches the input to the output for **V_{AC}** less than 6 V. The bq24030/1 regulates **V_{OUT}** to 6 V for larger inputs, and the bq24035 turns off the charging and output for an AC input above 6 VDC.
9. Verify **V_{LDO}** is between 3.2 VDC and 3.4 VDC.
10. Allow the battery to charge until **V_{BAT}** is between 3.2 VDC and 4.0 VDC. The charger should deliver the programmed constant current to the battery unless the input cannot source the required current.
11. Verify **D3** (STAT2) has turned off.
12. Verify **I_{BAT}** is ~1.0 A (for a 10-kΩ resistor on **ISET1**, **I_{BAT}** = **I_{AC}** - (**V_{OUT}** / **R_{OUT}**) - 0.01 A).
13. Verify **I_{AC}** is ~1.5 A (for 10-Ω **OUT** load and 10 kΩ on **ISET1**).
14. Apply a short between J3-4 (CE) and J3-3 (GND) on the UUT. This overrides the JMP3 100-kΩ pullup, disables the charging, puts the IC in low power mode and connects the battery to the OUT pin. Note that if CE is floated (JMP3 is removed and J3-4 connection is removed) the IC may bounce between the charging and disabled states. Verify on the scope that **V_{OUT}** does not drop out. Note that the transition between power sources is implemented by break-before-make switching and requires the capacitance on **V_{OUT}** to be able to hold up the system voltage for at least 50 μs.
15. Verify **D2** (STAT1) has turned off.
16. Verify **I_{AC}** drops below 10 mA (should be < 200 μA into the IC if ACPG LED (current) JMP6 is removed).

17. Verify V_{OUT} is within -50 mV of V_{BAT} .
18. Remove short between J3-4 and J3-3 on UUT. Verify on the scope that V_{OUT} does not drop out. Verify **D2** (STAT1) has turned on, charging has resumed and V_{OUT} is powered from the input.
19. Disconnect the **+5.25-VDC Input Supply** from the UUT AC input. Verify on the scope that V_{OUT} does not drop out. Verify V_{OUT} is within -50 mV of V_{BAT} and **D2** (STAT1) and **D5** (ACPG) LEDs turn off. This demonstrates battery power backup for loss of AC adapter.
20. Reapply the +5.25-VDC supply to the UUT AC input. Verify on the scope that V_{OUT} does not drop out. Verify **D2** (STAT1) and **D5** (ACPG) LEDs turn on.
21. Adjust R15 until the voltage on TP1 is $\sim 3.50\text{ VDC}$ (V_{BAT} should be less than 3.9 V for this demonstration; otherwise, discharge battery).
22. Reduce the current limit on the input supply to $\sim 1\text{ A}$ (going to AC pin on UUT) and verify on the scope that V_{OUT} has dropped to the VDPPM level of $\sim 4.0\text{ V}$ $\{(3.5\text{ V at TP1}) \times 1.15 = 4\text{ V}\}$. Note that the current into the battery is $\sim 600\text{ mA}$ (1-A input minus 400 mA to the system), which has been reduced to keep the output from falling below the programmed DPPM OUTPUT threshold of 4 V . This demonstrates DPPM operation (charging current to the battery is reduced if output drops to the DPPM OUTPUT voltage threshold attempting to keep the output voltage from dropping further).
23. Further reduce the input's current limit to 250 mA . Verify on the scope that V_{OUT} does not drop out. Verify that V_{OUT} drops just below V_{BAT} ($< 50\text{ mV}$). Because the available input current is less than the system **OUT** load, reducing the battery charging current to zero is still not enough reduction in load to keep the output from dropping. Once the output drops below $\sim 50\text{ mV}$, the internal battery FET turns on and allows the battery to source the OUT pin system load. This demonstrates battery supplement mode.
24. Return the current limit of the +5.25-V supply to $\sim 2\text{ A}$. Verify V_{OUT} returns to Vreg or Vin (see Step 8 of this test procedure).
25. Set JMP2 (PSEL) to USB (PSEL = low). Verify that the input current (AC) drops to between 400 mA and 500 mA . The programmed charge current of $\sim 1\text{ A}$ and the system load of $10\text{ }\Omega$ exceeds the USB 0.5-A limit; therefore, V_{OUT} drops until the DPPM OUTPUT voltage threshold, or battery voltage, is reached (which ever is higher). If the DPPM OUTPUT threshold is larger, the charging current is reduced to keep the output voltage from dropping further. If the battery voltage is higher, the battery supplements the current to keep the output from dropping too much (50 mV to 200 mV) below the battery voltage. Note that setting PSEL to low (USB mode; PSEL high is AC mode) selects the USB input as the primary source. If the USB source is not present, and the ac source is present, the IC uses the AC input source as if it were a USB input. This feature is useful if only one input power connector is desired, and two sources (USB and AC adaptor) are available. A keyed cable or a u-controller would set the PSEL pin for the available source. Note that the system would ideally go to a lower power mode prior to selecting USB operation to avoid pulling down V_{OUT} .
26. Plug in a USB cable from a high-power port (500 mA) into the UUT (or supply 5 VDC to J1). Verify that the USB input now supplies the input current instead of the AC (J2) input. This demonstrates that JMP2 (PSEL) defines the priority of the inputs. If PSEL = Low (USB priority), then the USB input is used first, if available, and if not it switches to AC power at USB-current levels.
27. Verify that **D4** (USBPG) turns on.
28. Set JMP2 (PSEL) to AC, and verify that the AC supply is providing $\sim 1.5\text{ A}$ of current (~ 0.44 to the load and 1 A to the battery plus miscellaneous).
29. Remove the ac-input supply and verify that the USB source is supplying between 400 mA and 500 mA of current to the input. The output should have dropped to the DPPM OUTPUT threshold or battery voltage (whichever is higher).
30. Verify that **D5** (ACPG) turns off.
31. Reapply the AC-input source and verify that the AC source is now providing the $\sim 1.5\text{ A}$ as before.
32. Verify that **D5** (ACPG) turns on.
33. Set JMP2 (PSEL) to USB, and verify that the USB source is now providing between 400 mA and 500 mA of current.
34. Set JMP1 to 0.1 on the UUT. Verify that the input current has dropped below 100 mA and V_{OUT} has dropped slightly below V_{BAT} . In this test, the system load is $10\text{ }\Omega$, which would result in the output dropping to 1 V at 100 mA if there were no other source to help out. As the output voltage drops to the DPPM OUTPUT threshold, the charging current is reduced to zero, but V_{OUT} continues to drop until it reaches the battery voltage. The internal battery FET turns on to supplement the OUTPUT. This

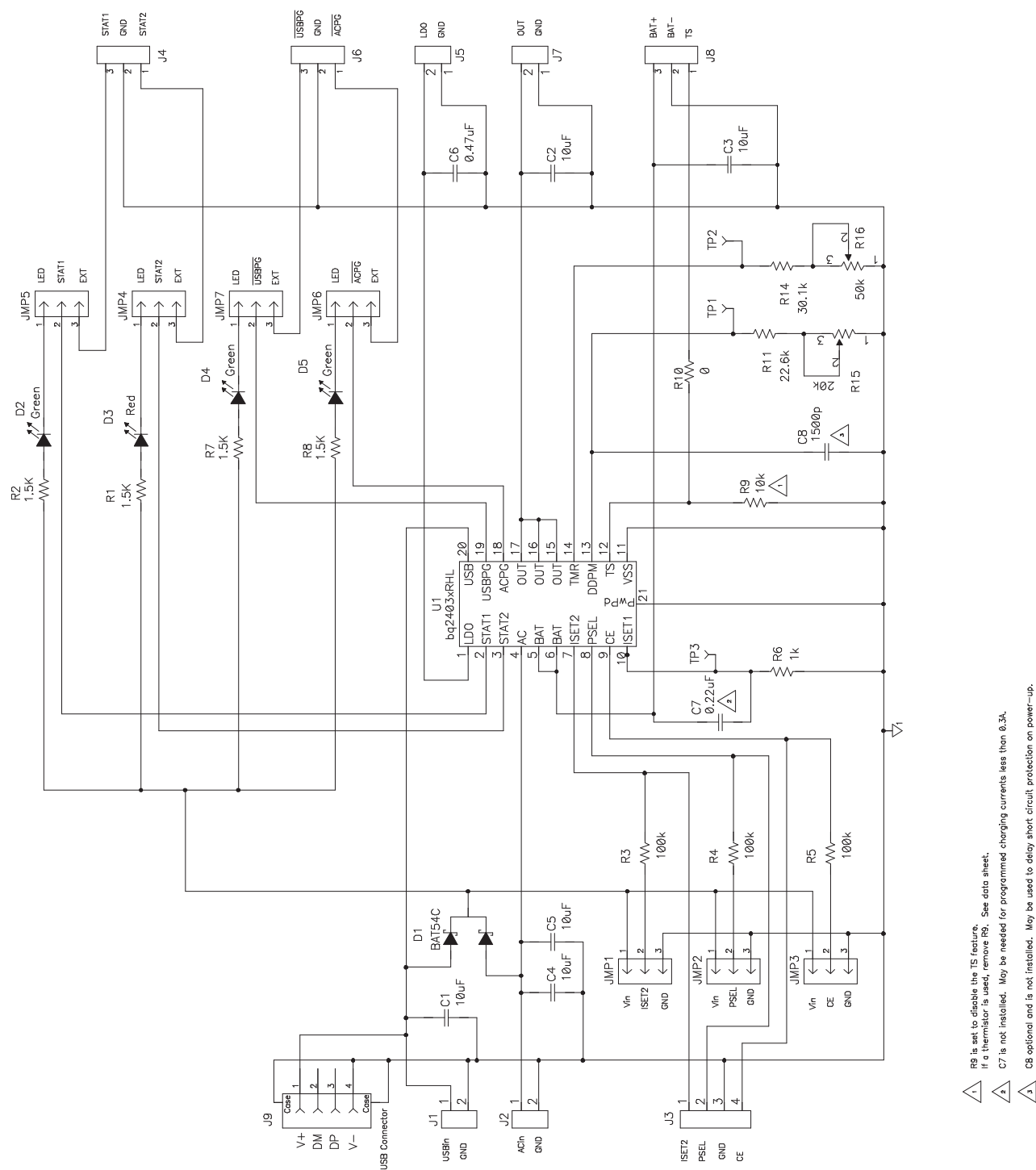
Test Summary

demonstrates battery supplement mode because the system load exceeds the available input current.

35. Disconnect the USB source and verify that the AC source takes over in USB mode at the 100-mA charge level.
36. Verify that **D4** (USBPG) turns off.
37. Set JMP2 to AC (PSEL = High). Skip to next step for HPA073-1/3 (bq24030/1/5 ICs). Verify that the (AC) input current is ~1 A. Verify that IBAT is reduced to half the programmed level, ~0.5 A. This is the AC half-charge mode and is implemented on bq24032 when ISET2 is low (0.1 A) and J2-PSEL is AC (High).
38. Set JMP1 to 0.5. Continue to let the battery charge. Note that once the battery voltage reaches regulation (~4.20 VDC for bq24030/2/5 and 4.1 VDC for bq24031) the charging current tapers off.
39. Verify that the charging terminates when the battery current tapers to C/10 or 100 mA (1 A/10, programmed charge current divided by 10). Verify D2 (STAT1) turns off (High) and D3 (STAT2) turns on (Low).
40. If a load is applied across the battery such that the battery is discharged to ~100 mV below the regulation voltage, the charger starts a new charging cycle.

This concludes the evaluation of the bq2403x EVM. Several more features implemented in the IC are not demonstrated in this user's guide. See the data sheet to learn more about thermal regulation, thermal cutoff, USB boot up, and short-circuit protection.

5 Schematic



- △ R9 is set to disable the TS feature. If a thermistor is used, remove R9. See data sheet.
- △ C7 is not installed. May be needed for programmed charging currents less than 0.3A.
- △ CB optional and is not installed. May be used to delay short circuit protection on power-up.

6 Physical Layouts

This section contains the board layout and assembly drawings for the EVM.

6.1 Board Layout

Figure 3 shows the top assembly view of the EVM. Figure 4 shows the top etch layer of the EVM. Figure 5 shows the board second etch layer of the EVM. Figure 6 shows the board third etch layer of the EVM. Figure 7 shows the bottom etch layer of the EVM.

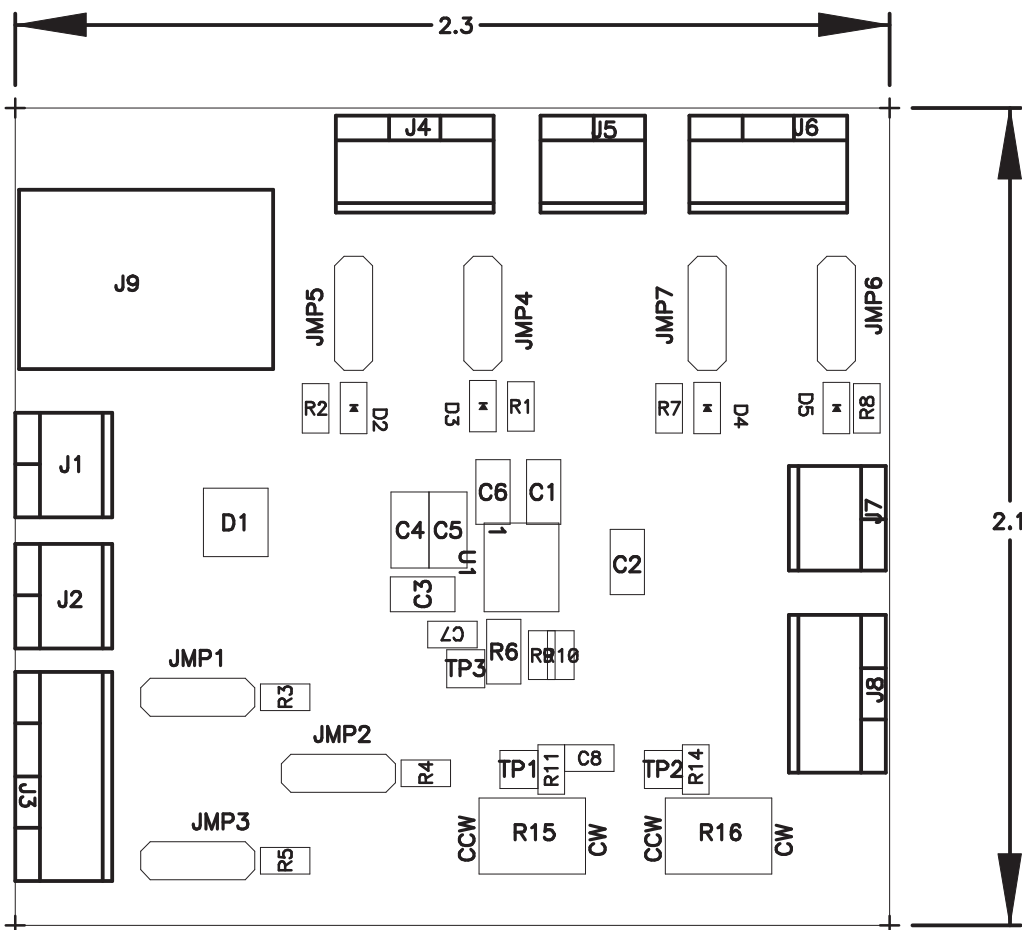


Figure 3. Top Assembly View

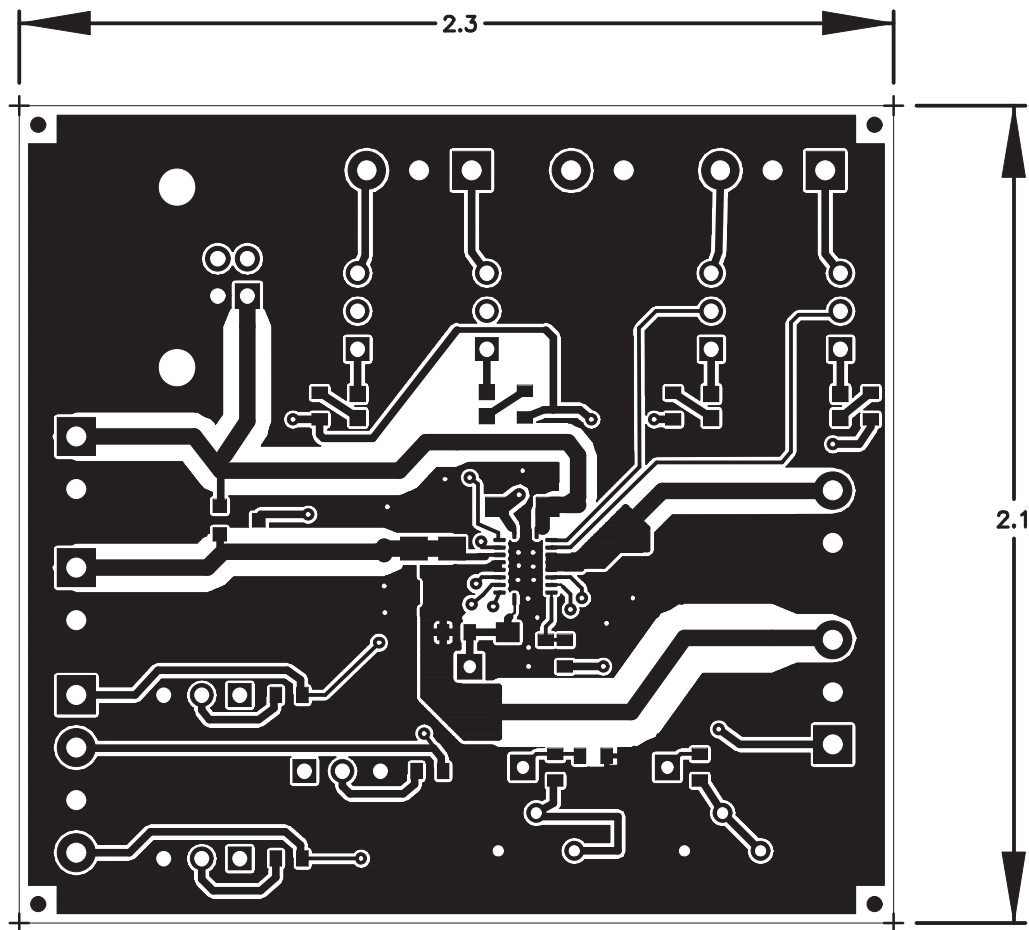


Figure 4. Board Layout – Top Etch Layer

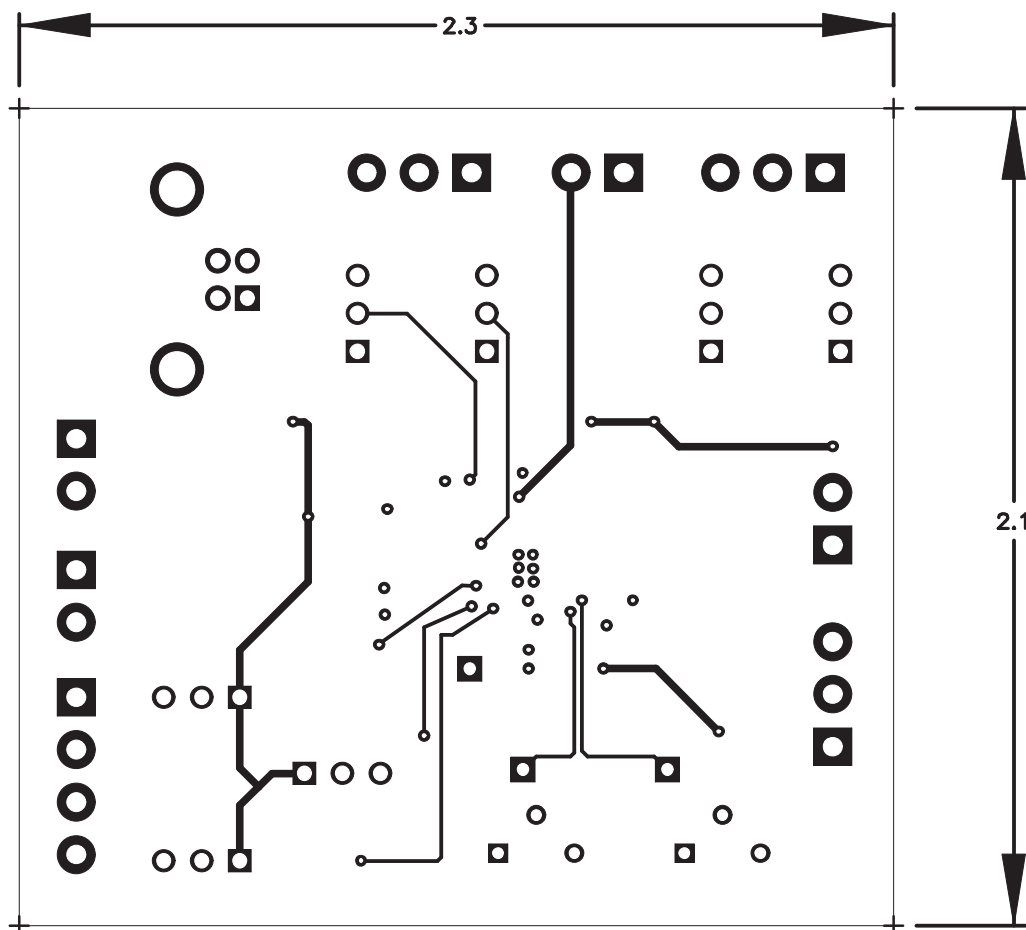


Figure 5. Board Layout – Second Etch Layer

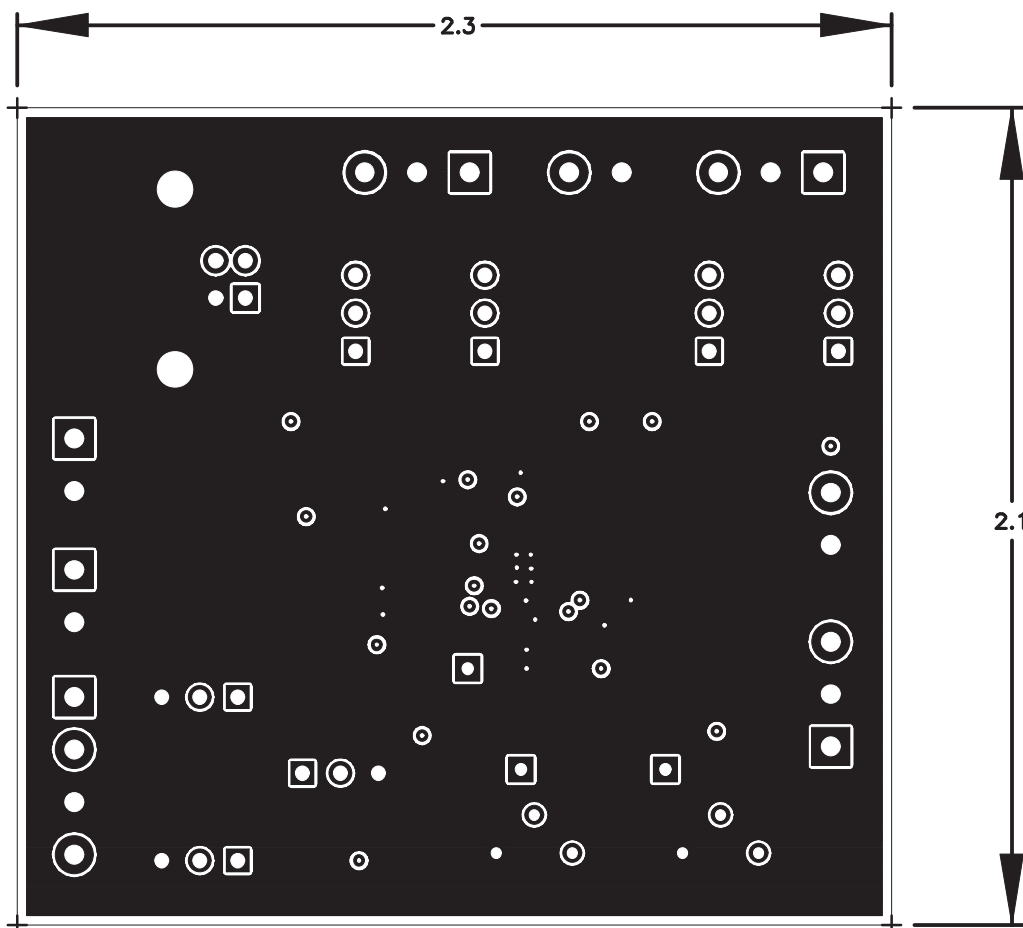


Figure 6. Board Layout – Third Etch Layer

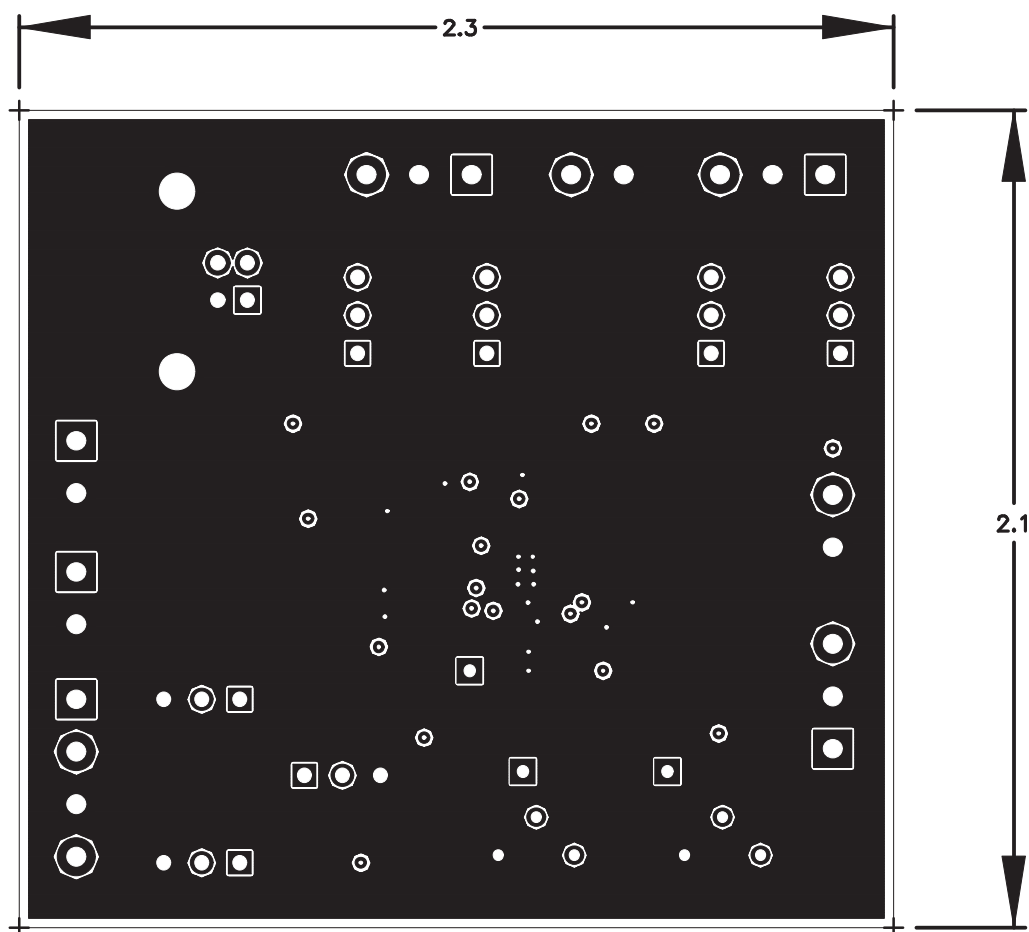


Figure 7. Board Layout – Bottom Etch Layer

7 Bill of Materials

Table 2. Bill of Materials⁽¹⁾⁽²⁾⁽³⁾⁽⁴⁾

COUNT				REFDES	DESCRIPTION	SIZE	MFR	PART NUMBER
bq24030-001	bq24032-002	bq24035-003	bq24031-004					
3	3	3	3	C1, C2, C3	Capacitor, ceramic, 10-μF, 6.3-V, X5R, 20%	805	Panasonic	ECJ-2FB0J106M
2	2	2	2	C4, C5	Capacitor, ceramic, 10-μF, 25-V, X5R, 20%	1206	Panasonic	ECJ-3YB1E106M
1	1	1	1	C6	Capacitor, ceramic, 0.47-μF, 16-V, X7R, 10%	805	Panasonic	ECJ-2YB1C474K
0	0	0	0	C7	Capacitor, ceramic, 0.22-μF, 10-V, X5R, 10%	603	Panasonic	ECJ-1VB1A224K
0	0	0	0	C8	Capacitor, ceramic, xxx-μF, 10-V, X5R, 10%	603	Panasonic	ECJ-1VB1C103K
1	1	1	1	D1	Diode, dual Schottky, 200-mA, 30-V	SOT23	Vishay-Liteon	BAT54C
3	3	3	3	D2, D4, D5	Diode, LED, green, 2.1-V, 20-mA, 6-mcd	603	Liteon	160-1183-1-ND
1	1	1	1	D3	Diode, LED, red, 1.8-V, 20-mA, 20-mcd	603	Liteon	160-1181-1-ND
4	4	4	4	J1, J2, J5, J7	Terminal block, 2-pin, 6-A, 3.5 mm	0.27 x 0.25	OST	ED1514
1	1	1	1	J3	Terminal block, 4-pin, 6-A, 3.5 mm	0.55 x 0.25	OST	ED1516
3	3	3	3	J4, J6, J8	Terminal block, 3-pin, 6-A, 3.5 mm	0.41 x 0.25	OST	ED1515
1	1	1	1	J9	Connector, USB upstream (Type B)	0.47 x 0.67	Molex	67068-1000
7	7	7	7	JMP1, JMP2, JMP3, JMP4, JMP5, JMP6, JMP7	Header, 3-pin, 100-mil spacing, (36-pin strip)	0.10 x 3	Sullins	PTC36SAAN
4	4	4	4	R1, R2, R7, R8	Resistor, Chip, 1.5-kΩ, 1/16-W, 1%	603	Std	Std
1	1	1	1	R10	Resistor, Chip, 0-Ω, 1/16-W, 1%	603	Std	Std
1	1	1	1	R11	Resistor, Chip, 22.6-kΩ, 1/16-W, 1%	603	Std	Std
1	1	1	1	R14	Resistor, Chip, 30.1-kΩ, 1/16-W, 1%	603	Std	Std
1	1	1	1	R15	Potentiometer, 20-kΩ, 1/4 inch Cermet, 12-turn, top-adjust	0.25 × 0.17	Bourns	3266W-203
1	1	1	1	R16	Potentiometer, 50-kΩ, 1/4 inch Cermet, 12-turn, top-adjust	0.25 × 0.17	Bourns	3266W-503
3	3	3	3	R3, R4, R5	Resistor, chip, 100-kΩ, 1/16-W, 1%	603	Std	Std
1	1	1	1	R6	Resistor, chip, 1-kΩ, 1/10W, 1%	805	Std	Std
1	1	1	1	R9	Resistor, chip, 10-kΩ, 1/16-W, 1%	603	Std	Std
3	3	3	3	TP1, TP2, TP3	Test point, 0.032-inch hole		None	Void
1			1	U1	IC, single chip charge and power path management	QFN	TI	bq24030RHL
	1			U1	IC, single chip charge and power path management	QFN	TI	bq24032ARHL
		1		U1	IC, single chip charge and power path management	QFN	TI	bq24035RHL
1	1	1	1	--	PCB, 2-inch x 2-inch x 0.31-inch		Any	HPA073

- (1) These assemblies are ESD sensitive, ESD precautions shall be observed.
- (2) These assemblies must be clean and free from flux and all contaminants. Use of no-clean flux is not acceptable.
- (3) These assemblies must comply with workmanship standards IPC-A-610 Class 2.
- (4) Reference designators marked with an asterisk (***) cannot be substituted. All other components can be substituted with equivalent MFR's components.

8 References

1. [SLUS618](#), bq2403x Datasheet

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EVM WARNINGS AND RESTRICTIONS

It is important to operate this EVM within the input voltage range of 0 V to 6.5 V and the output voltage range of 0 V to 6.5 V.

Exceeding the specified input range may cause unexpected operation and/or irreversible damage to the EVM. If there are questions concerning the input range, please contact a TI field representative prior to connecting the input power.

Applying loads outside of the specified output range may result in unintended operation and/or possible permanent damage to the EVM. Please consult the EVM User's Guide prior to connecting any load to the EVM output. If there is uncertainty as to the load specification, please contact a TI field representative.

During normal operation, some circuit components may have case temperatures greater than 85°C. The EVM is designed to operate properly with certain components above 60°C as long as the input and output ranges are maintained. These components include but are not limited to linear regulators, switching transistors, pass transistors, and current sense resistors. These types of devices can be identified using the EVM schematic located in the EVM User's Guide. When placing measurement probes near these devices during operation, please be aware that these devices may be very warm to the touch.

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