MP28119

Ultra-small
1.7MHz, 600mA, Low-Voltage
Synchronous Step-Down Converter

The Future of Analog IC Technology

DESCRIPTION

The MP28119 is a 1.7MHz constant frequency, current mode, ultra-small PWM step-down converter. The device integrates a main switch and a synchronous rectifier for high efficiency and eliminates the needs for an external schottky diode. It is ideal for powering portable equipment that runs from a single cell Lithium-Ion (Li+) battery. The MP28119 can supply 600mA of load current from a 2.5V to 6V input voltage.

The output voltages of the MP28119EG-1.0 and MP28119EG-1.2 are fixed at 1.0V and 1.2V respectively.

The MP28119 can also run at 100% duty cycle for low dropout applications.

The MP28119 has an output discharge function that controls the discharging of the output capacitor.

The MP28119 is available in an ultra-small 2x2 QFN package.

FEATURES

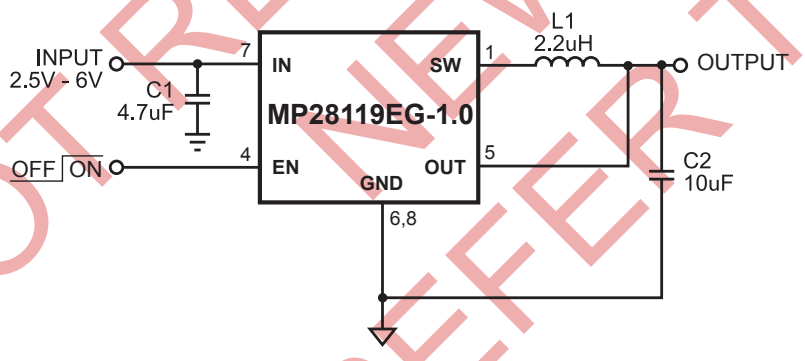
- High Efficiency: Up to 89%
- 1.7MHz Constant Switching Frequency
- 600mA Available Load Current
- 2.5V to 6V Input Voltage Range
- 100% Duty Cycle in Dropout
- Current Mode Control
- Short Circuit Protection
- Thermal Fault Protection
- Output Discharge Function
- <0.1μA Shutdown Current
- 1.0V and 1.2V Fixed Out-put Versions
- Ultra-Small 2x2 QFN Package

APPLICATIONS

- Cellular and Smart Phones
- Microprocessors and DSP Core Supplies
- PDAs
- MP3 Players
- Digital Still and Video Cameras
- Portable Instruments

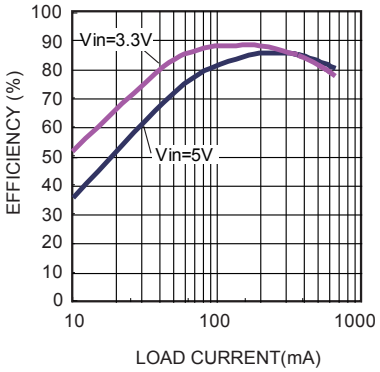
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TYPICAL APPLICATION



Efficiency vs. Load Current

V_{OUT}=1V, L=2.2uH



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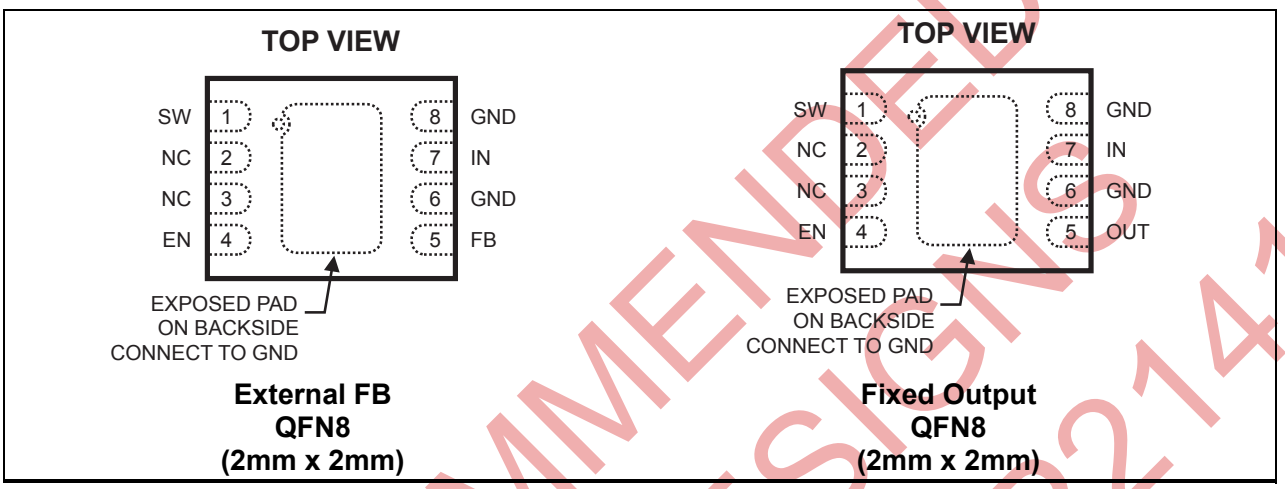
ORDERING INFORMATION

Part Number	V_{OUT}	Package	Temperature	Top Marking
MP28119EG-1.0	1.0V	2x2 QFN8	-20°C to +85°C	9K
MP28119EG-1.2*	1.2V			
MP28119EG*	ADJ			4Q

* Contact factory for availability.

** MP28119EG-1.0 is fixed to 1V output, and MP28119EG-1.2 is fixed to 1.2V output. For RoHS Compliant Packaging, add suffix - LF (e.g. MP28119EG-LF); For Tape and Reel, add suffix -Z (e.g. MP28119EG-LF-Z).

PACKAGE REFERENCE



ABSOLUTE MAXIMUM RATINGS (1)

V_{IN} to GND ..... –0.3V to +6.5V
V_{SW} to GND ..... –0.3V to V_{IN} +0.3V
V_{FB}, V_{EN} to GND ..... –0.3V to +6.5V
Junction Temperature ..... +150°C
Continuous Power Dissipation (T_A = +25°C) (2) 1.5W
Lead Temperature ..... +260°C
Storage Temperature ..... –65°C to +150°C

Recommended Operating Conditions (3)

Supply Voltage V_{IN} ..... 2.5V to 6V
Output Voltage V_{OUT} ..... 0.6V to 6V
Operating Temperature ..... –20°C to +85°C

Thermal Resistance (4)

	θ_{JA}	θ_{JC}
2x2 QFN8	80	16 °C/W

Notes:

1) Exceeding these ratings may damage the device.
2) The maximum allowable power dissipation is a function of the maximum junction temperature. T_J(MAX) the junction-to-ambient thermal resistance θ_{JA} and the ambient temperature T_A. The maximum allowable power dissipation at any ambient temperature is calculated using: $P_D(MAX)=(T_J(MAX)-T_A)/\ \theta_{JA}$. Exceeding the maximum allowable power dissipation will cause excessive die temperature, and the regulator will go into thermal shutdown. Internal thermal shutdown circuitry protects the device from permanent damage.
3) The device is not guaranteed to function outside of its operating conditions.
4) Measured on JESD51-7 4-layer board.

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ELECTRICAL CHARACTERISTICS (5) *(continued)*

V_{IN} = V_{EN} = 3.6V, T_A = +25°C, unless otherwise noted.

Parameter	Symbol	Condition	Min	Typ	Max	Units
Supply Current		V_{EN} = V_{IN}, V_{FB} = 0.51V		400	600	μA
Shutdown Current		V_{EN} = 0V, V_{IN} = 6V		0.01	1	μA
IN Under Voltage Lockout Threshold	UVLO	Rising Edge	2.15	2.3	2.45	V
IN Under Voltage Lockout Hysteresis				80		mV
Discharge MOSFET On Resistance				100		Ω
Regulated FB Voltage	VFB	T_A = +25°C, MP28119EG	0.588	0.600	0.612	V
		–20°C ≤ T_A ≤ +85°C	0.582	0.600	0.618	
FB Input Bias Current		V_{FB} = 0.65V, MP28119EG	–50	0.5	+50	nA
Regulated Output Voltage		MP28119EG-1.0 I_{OUT} = 50mA;–20°C ≤ T_A ≤ +85°C	0.97	1.000	1.03	V
		MP28119EG-1.2 I_{OUT} = 50mA;–20°C ≤ T_A ≤ +85°C	1.164	1.200	1.236	
PFET On Resistance	Rds(on)1	I_{SW} = 100mA		0.42		Ω
NFET On Resistance	Rds(on)2	I_{SW} = –100mA		0.28		Ω
SW Leakage Current		V_{EN} = 0V,V_{IN} = 6V, V_{SW} = 0V or 6V	–1		+1	μA
PFET Current Limit		Duty Cycle = 100%, Current Pulse Width < 1ms	0.8	1.6	2.4	A
Oscillator Frequency	Focs			1.7		MHz
Minimum On Time	Ton			100		ns
Thermal Shutdown Trip Threshold				145		°C
EN Turn-onTrip Threshold		–20°C ≤ T_A ≤ +85°C			1.5	V
EN Turn-off Trip Threshold		–20°C ≤ T_A ≤ +85°C	0.3			V
EN Input Current		V_{EN} = 0V to 6V	–1		+1	μA

Notes:

5) 100% production test at +25°C. Typical and temperature specifications are guaranteed by design and characterization.

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PIN FUNCTIONS

Pin #	Name	Description
1	SW	Power Switch Output. Inductor connection to drains of the internal PFET and NFET switches.
2, 3	NC	No Connection
4	EN	Regulator Enable Control Input. Drive EN above 1.5V to turn on the MP28119. Drive EN below 0.3V to turn it off (shutdown current < 0.1μA).
5	FB	Feedback Input (MP28119EG). Connect FB to the center point of the external resistor divider. The feedback threshold voltage is 0.6V.
	OUT	Output Voltage Sense Input (MP28119EG-1.0 and MP28119EG-1.2). An internal resistor divider is connected to this pin to set the proper output voltage.
6, 8	GND, Exposed Pad	Ground. Connect exposed pad to GND plane for optimal thermal performance.
7	IN	Supply Input. Bypass to GND with a 2.2μF or greater ceramic capacitor.

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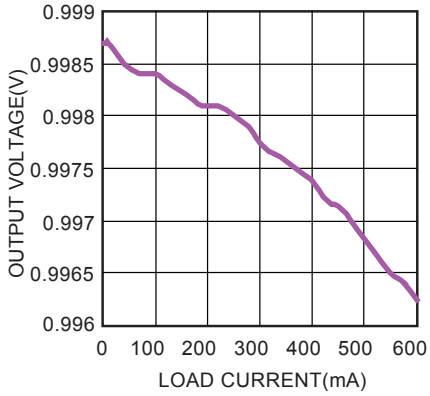
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TYPICAL PERFORMANCE CHARACTERISTICS

Vin=5V, Vout=1V, L1=2.2uH, C1=4.7uF, C2=10uF, T_A=25°C, tested on MP28119EG-1.0, unless otherwise noted.

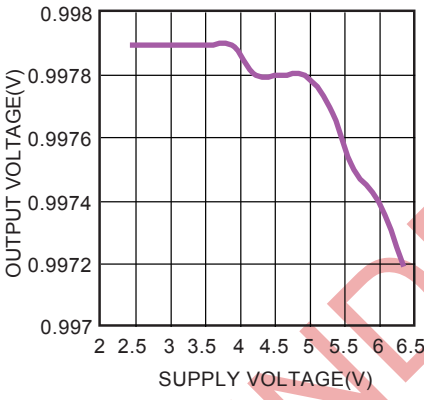
Output Voltage vs. Load Current

V_{IN}=5V



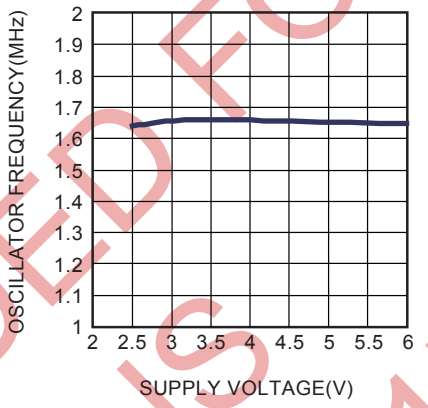
Output Voltage vs. Supply Voltage

I_{OUT}=300mA



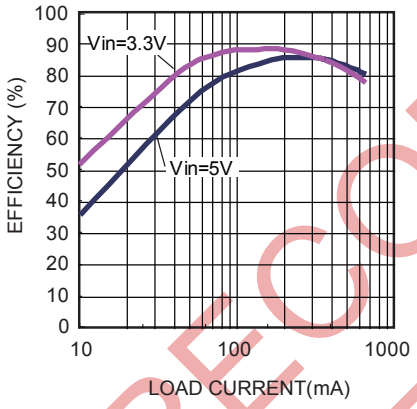
Oscillator Frequency vs. Supply Voltage

I_{OUT}=300mA



Efficiency vs. Load Current

V_{OUT}=1V, L=2.2uH



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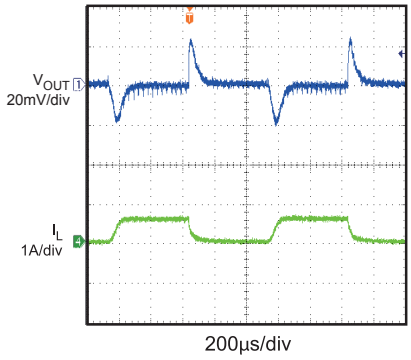
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TYPICAL PERFORMANCE CHARACTERISTICS *(continued)*

Vin=5V, Vout=1V, L1=2.2uH, C1=4.7uF, C2=10uF, T_A=25°C, tested on MP28119EG-1.0, unless otherwise noted.

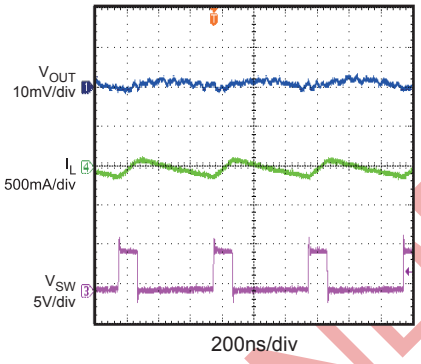
Load Transient

V_{IN}=5V, I_{OUT}=0mA->600mA Step
V_{OUT}=1V



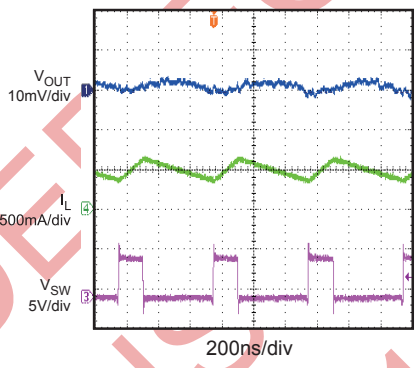
Light Load Operation

V_{IN}=5V, V_{OUT}=1V, I_{OUT}=0A



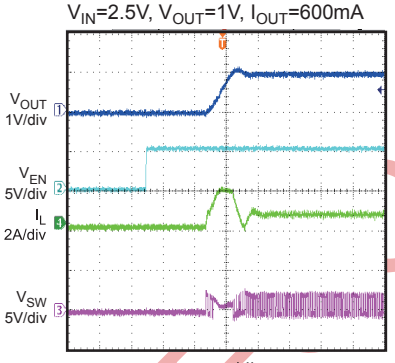
Heavy Load Operation

V_{IN}=5V, V_{OUT}=1V, I_{OUT}=600mA



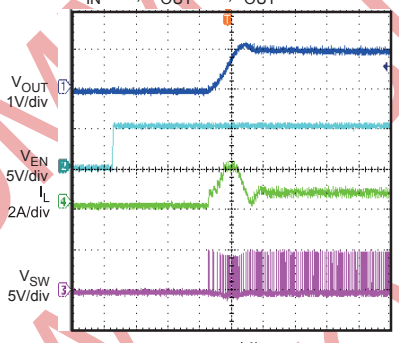
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V_{IN}=2.5V, V_{OUT}=1V, I_{OUT}=600mA



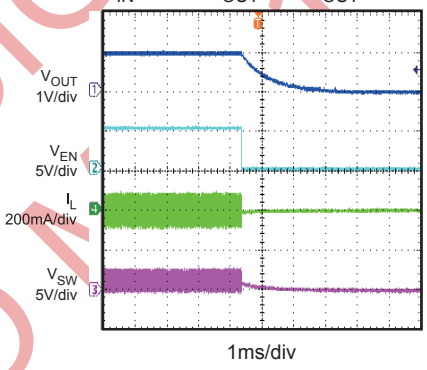
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V_{IN}=5V, V_{OUT}=1V, I_{OUT}=600mA



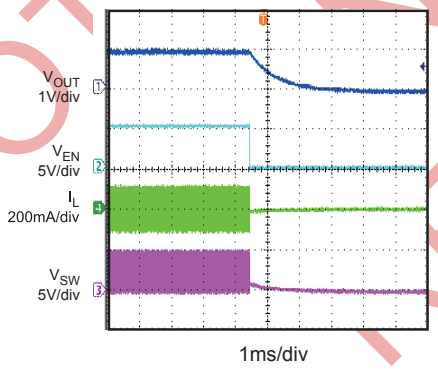
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V_{IN} = 2.5V, V_{OUT} = 1V, I_{OUT} = 0mA



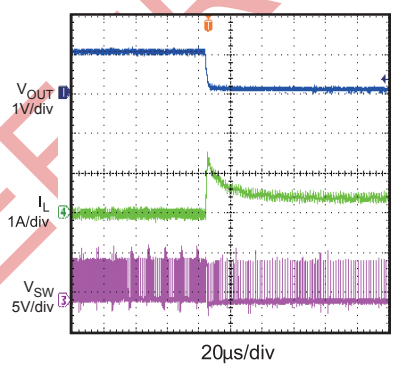
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V_{IN} = 5V, V_{OUT} = 1V, I_{OUT} = 0mA



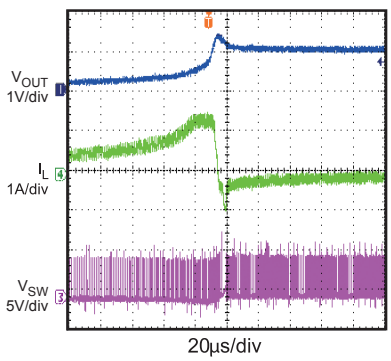
Short Circuit Protection

V_{IN}=5V, I_{OUT}=0mA



Short Circuit Recovery

V_{IN}=5V, I_{OUT}=0mA



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OPERATION

The MP28119 is a constant frequency current mode PWM step-down converter. The MP28119 is optimized for low voltage, Li-Ion battery powered applications where high efficiency and small size are critical. The MP28119 uses an external resistor divider to set the output voltage from 0.6V to 6V. The device integrates both a main switch and synchronous rectifier, which provides high efficiency and eliminates an external Schottky diode. The MP28119 can achieve 100% duty cycle. The duty cycle D of a step-down converter is defined as:

$$D = T_{ON} \times f_{OSC} \times 100\% \approx \frac{V_{OUT}}{V_{IN}} \times 100\%$$

Where T_{ON} is the main switch on time, f_{OSC} is the oscillator frequency (1.7MHz), V_{OUT} is the output voltage and V_{IN} is the input voltage.

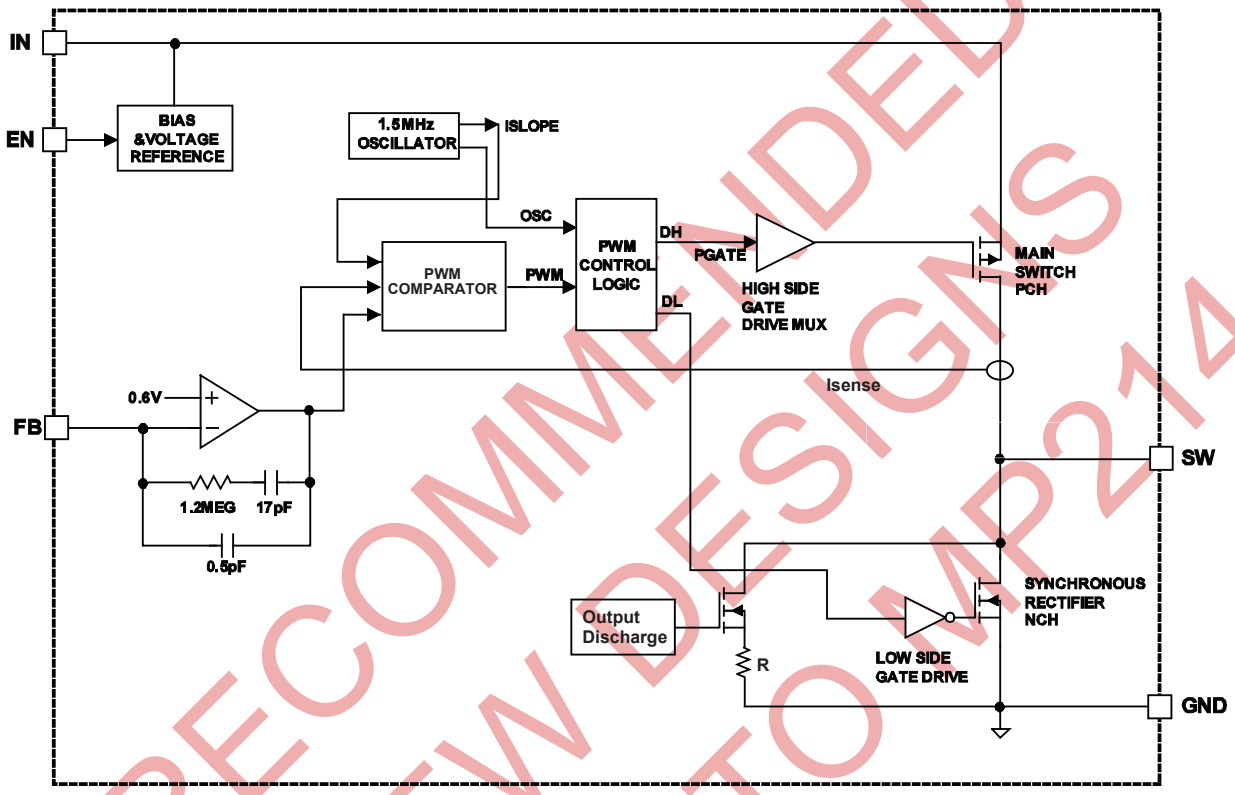


Figure 1—Functional Block Diagram (MP28119EG)

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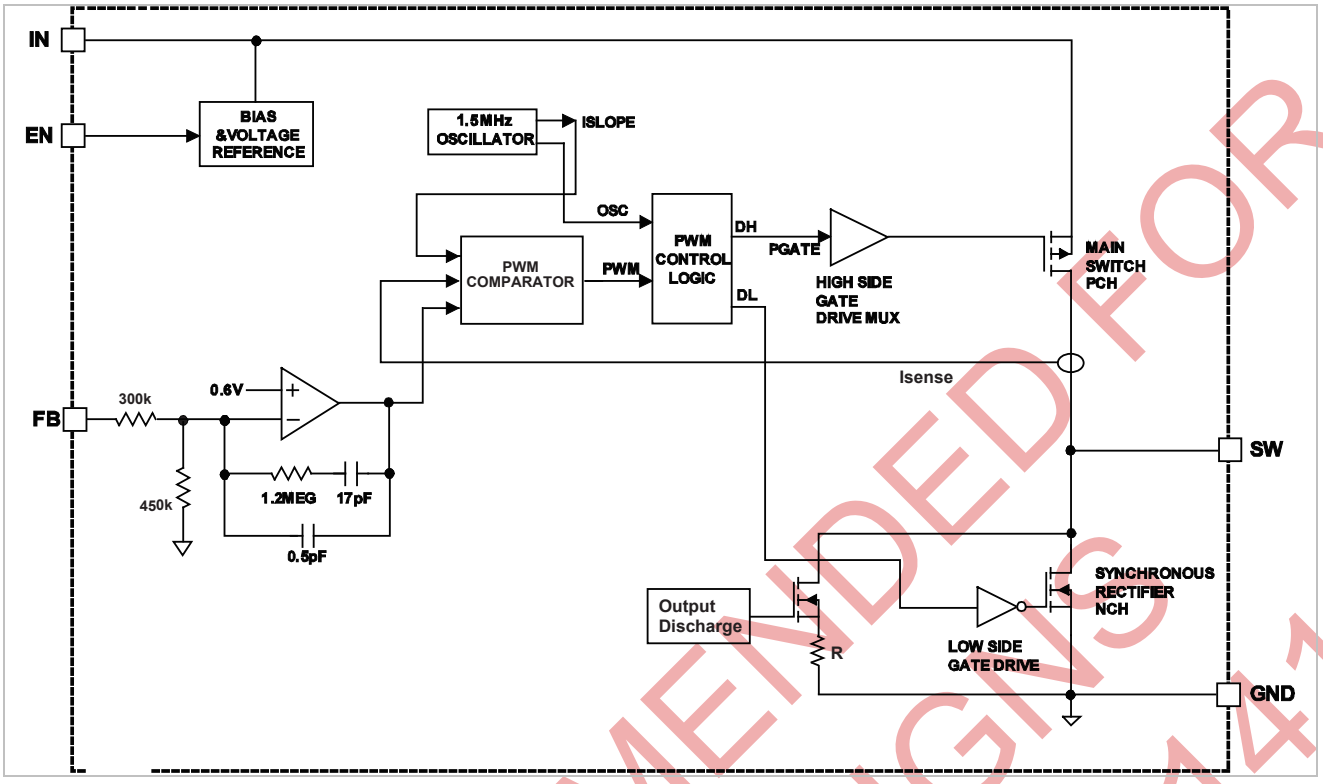


Figure 2—Functional Block Diagram
(MP28119EG-1.0)

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Current Mode PWM Control

Slope compensated current mode PWM control provides stable switching and cycle-by-cycle current limit for superior load and line response and protection of the internal main switch and synchronous rectifier. The MP28119 switches at a constant frequency (1.7MHz) and regulates the output voltage. During each cycle the PWM comparator modulates the power transferred to the load by changing the inductor peak current based on the feedback error voltage. During normal operation, the main switch is turned on for a certain time to ramp the inductor current at each rising edge of the internal oscillator, and switched off when the peak inductor current is above the error voltage. When the main switch is off, the synchronous rectifier will be turned on immediately and stay on until either the next cycle starts.

Dropout Operation

The MP28119 allows the main switch to remain on for more than one switching cycle and increases the duty cycle while the input voltage is dropping close to the output voltage. When the duty cycle reaches 100%, the main switch is held on continuously to deliver current to the output up to the PFET current limit. The output voltage then is the input voltage minus the voltage drop across the main switch and the inductor.

Short Circuit Protection

The MP28119 has short circuit protection. When the output is shorted to ground, the oscillator frequency is reduced to prevent the inductor current from increasing beyond the PFET current limit. The PFET current limit is also reduced to lower the short circuit current. The frequency and current limit will return to the normal values once the short circuit condition is removed and the feedback voltage reaches 0.6V.

Maximum Load current

The MP28119 can operate down to 2.5V input voltage, however the maximum load current decreases at lower input due to large IR drop on the main switch and synchronous rectifier. The slope compensation signal reduces the peak inductor current as a function of the duty cycle to prevent sub-harmonic oscillations at duty cycles greater than 50%. Conversely the current limit increases as the duty cycle decreases.

Output Discharge Function

When the device is disabled, the part goes into output discharge mode automatically and its internal discharge MOSFET provides a resistive discharge path for the output capacitor. This function is only suitable for discharge output capacitor for the limited time. If the part is used in a pre-bias application, the external power supply providing the pre-bias should supply larger current than the discharge MOSFET sinks.

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APPLICATION INFORMATION

Output Voltage Setting (MP28119EG)

The external resistor divider sets the output voltage (see Figure 3). The feedback resistor R1 also sets the feedback loop bandwidth with the internal compensation capacitor (see Figure 1).

Choose R1 around 300kΩ for optimal transient response. R2 is then given by:

$$R2 = \frac{R1}{\frac{V_{OUT}}{0.6V} - 1}$$

Table 1—Resistor Selection vs. Output Voltage Setting

V_{OUT}	R1	R2
1.2V	300kΩ (1%)	300kΩ (1%)
1.5V	300kΩ (1%)	200kΩ (1%)
1.8V	300kΩ (1%)	150kΩ (1%)
2.5V	300kΩ (1%)	95.3kΩ (1%)

For the fixed output versions (MP28119EG-1.0 and MP28119EG-1.2), R1 is shorted and R2 is open.

Input Capacitor Selection

The input capacitor reduces the surge current drawn from the input and switching noise from the device. The input capacitor impedance at the switching frequency shall be less than input source impedance to prevent high frequency switching current passing to the input. Ceramic capacitors with X5R or X7R dielectrics are highly recommended because of their low ESR and small temperature coefficients. For most applications, a 4.7μF capacitor is sufficient.

Inductor Selection

A 1μH to 10μH inductor with DC current rating at least 25% higher than the maximum load current is recommended for most applications. For best efficiency, the inductor DC resistance shall be <200mΩ. For most designs, the inductance value can be derived from the following equation:

$$L = \frac{V_{OUT} \times (V_{IN} - V_{OUT})}{V_{IN} \times \Delta I_L \times f_{OSC}}$$

Where ΔI_L is the inductor ripple current. Choose inductor ripple current approximately 30% of the maximum load current, 600mA.

The maximum inductor peak current is:

$$I_{L(MAX)} = I_{LOAD} + \frac{\Delta I_L}{2}$$

Under light load conditions below 100mA, larger inductance is recommended for improved efficiency.

Output Capacitor Selection

The output capacitor keeps output voltage ripple small and ensures regulation loop stable. The output capacitor impedance shall be low at the switching frequency. Ceramic capacitors with X5R or X7R dielectrics are recommended. The output ripple ΔV_{OUT} is approximately:

$$\Delta V_{OUT} \le \frac{V_{OUT} \times (V_{IN} - V_{OUT})}{V_{IN} \times f_{OSC} \times L} \times \left(ESR + \frac{1}{8 \times f_{OSC} \times C3} \right)$$

PCB Layout Guide

PCB layout is very important to achieve stable operation. It is highly recommended to duplicate EVB layout for optimum performance.

If change is necessary, please follow these guidelines and take Figure 3 for reference.

1) Keep the path of switching current short and minimize the loop area formed by Input cap, high-side MOSFET and low-side MOSFET.

2) Bypass ceramic capacitors are suggested to be put close to the Vin Pin.

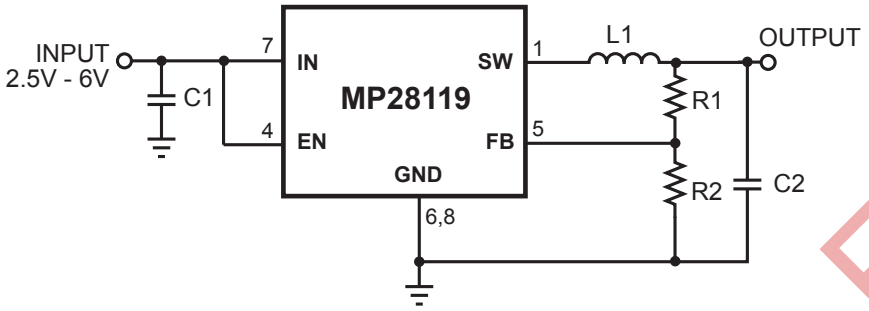
3) Place the external feedback resistors as close to the FB pin as possible.

4) Route SW away from sensitive analog areas such as FB.

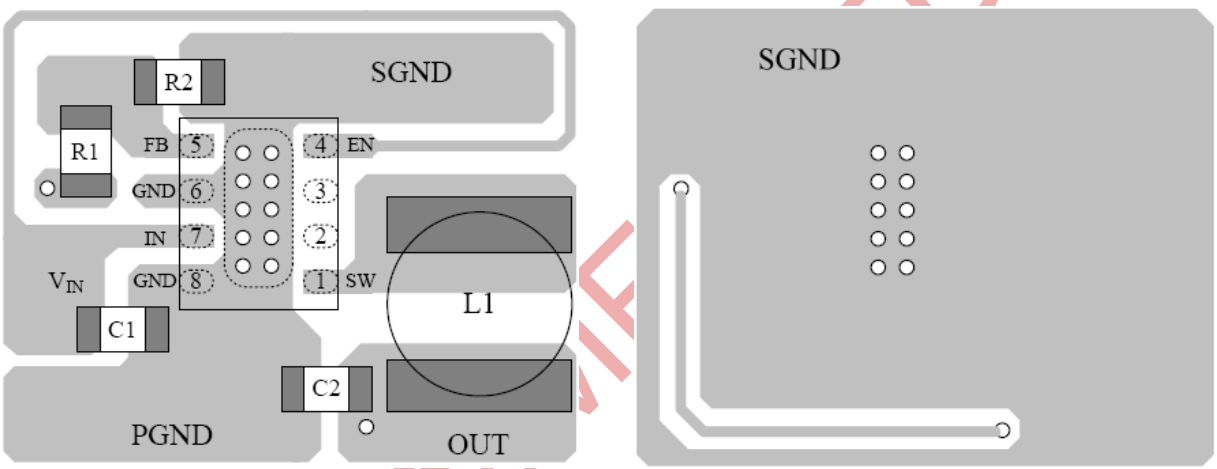
5) Connect IN, SW, and especially GND respectively to a large copper area to cool the chip to improve thermal performance and long-term reliability.

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MP28119 Typical Application Circuit



TOP Layer **Bottom Layer**

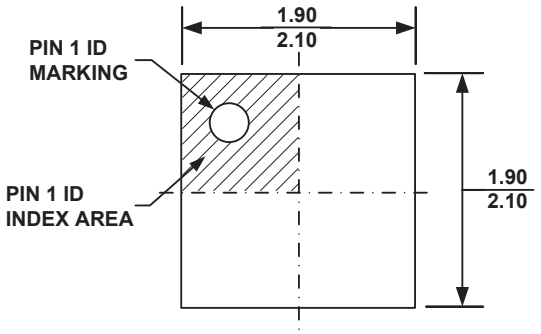
Figure 3—MP28119 Typical Application Circuit and PCB Layout Guide

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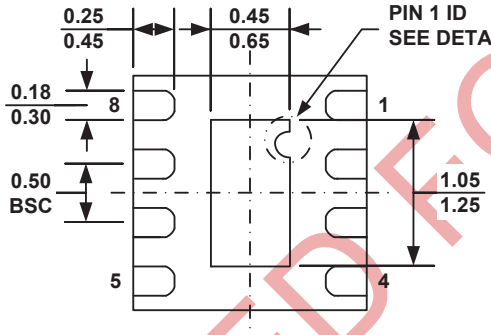
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PACKAGE INFORMATION

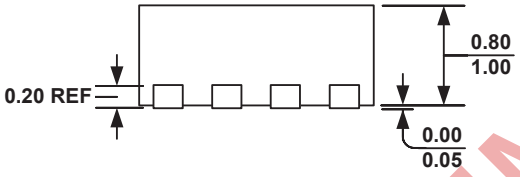
2MM X 2MM QFN8



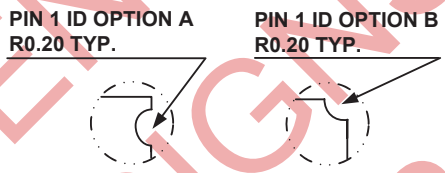
TOP VIEW



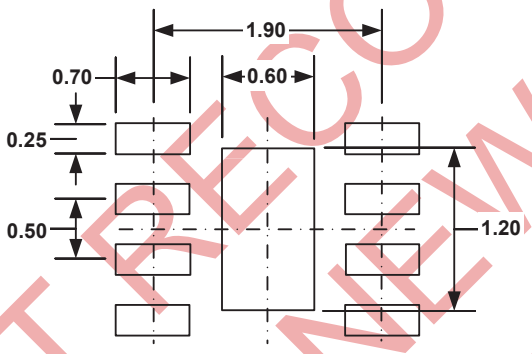
BOTTOM VIEW



SIDE VIEW



DETAIL A



RECOMMENDED LAND PATTERN

NOTE:

1) ALL DIMENSIONS ARE IN MILLIMETERS.
2) EXPOSED PADDLE SIZE DOES NOT INCLUDE MOLD FLASH.
3) LEAD COPLANARITY SHALL BE 0.10 MILLIMETER MAX.
4) DRAWING CONFORMS TO JEDEC MO-229, VARIATION VCCD-3.
5) DRAWING IS NOT TO SCALE.

NOT RECOMMENDED FOR NEW DESIGNS REFER TO MP2141N

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