



dBCool® Remote Thermal Controller and Voltage Monitor

ADT7468

FEATURES

- Monitors up to 5 voltages
- Controls and monitors up to 4 fans
- High and low frequency fan drive signal
- 1 on-chip and 2 remote temperature sensors
- Series resistance cancellation on the remote channel
- Extended temperature measurement range, up to 191°C
- Dynamic T_{MIN} control mode intelligently optimizes system acoustics
- Automatic fan speed control mode manages system cooling based on measured temperature
- Enhanced acoustic mode dramatically reduces user perception of changing fan speeds
- Thermal protection feature via THERM output
- Monitors performance impact of Intel® Pentium® 4 processor
- Thermal control circuit via THERM input
- 2-wire, 3-wire, and 4-wire fan speed measurement
- Limit comparison of all monitored values
- Meets SMBus 2.0 electrical specifications (fully SMBus 1.1-compliant)

GENERAL DESCRIPTION

The ADT7468 dBCool controller is a thermal monitor and multiple PWM fan controller for noise sensitive or power sensitive applications requiring active system cooling. The ADT7468 can drive a fan using either a low or high frequency drive signal, monitor the temperature of up to two remote sensor diodes plus its own internal temperature, and measure and control the speed of up to four fans, so that they operate at the lowest possible speed for minimum acoustic noise.

The automatic fan speed control loop optimizes fan speed for a given temperature. A unique, dynamic T_{MIN} control mode enables the system thermals/acoustics to be intelligently managed. The effectiveness of the system's thermal solution can be monitored using the THERM input. The ADT7468 also provides critical thermal protection to the system using the bidirectional THERM pin as an output to prevent system or component overheating.

FUNCTIONAL BLOCK DIAGRAM

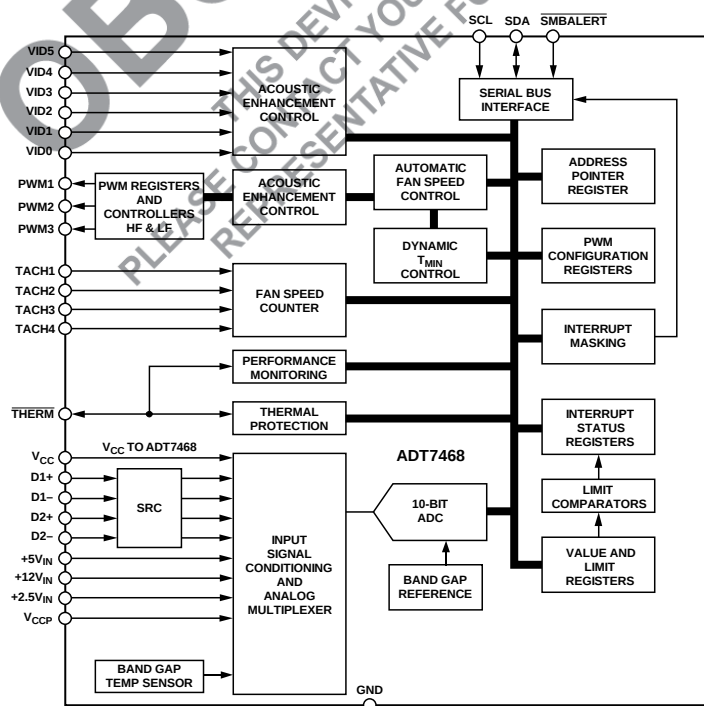


Figure 1.

TABLE OF CONTENTS

Specifications.....	3	PWM Logic State.....	33
Absolute Maximum Ratings.....	5	Fan Speed Control.....	34
ESD Caution.....	5	Miscellaneous Functions	35
Pin Configuration and Function Descriptions.....	6	Operating from 3.3 V Standby.....	35
Typical Performance Characteristics	8	XNOR Tree Test Mode.....	35
Product Description	10	Power-On Default	35
Comparison between ADT7463 and ADT7468	10	Automatic Fan Control Overview.....	37
Recommended Implementation.....	11	Dynamic T_{MIN} Control Mode.....	38
Serial Bus Interface	12	Programming the Automatic Fan Speed Control Loop	40
Write Operations	13	Step 1: Hardware Configuration	40
Read Operations	14	Step 2: Configuring the Mux	43
SMBus Timeout	14	Step 3: T_{MIN} Settings for Thermal Calibration Channels.....	45
VID Code Monitoring	15	Step 4: PWM_{MIN} for Each PWM (Fan) Output	46
VID Code Registers.....	15	Step 5: PWM_{MAX} for PWM (Fan) Outputs.....	46
Analog-to-Digital Converter	16	Step 6: T_{RANGE} for Temperature Channels.....	47
Voltage Measurement Input.....	16	Step 7: T_{THERM} for Temperature Channels	50
Additional ADC Functions for Voltage Measurements	16	Step 8: T_{HYST} for Temperature Channels	51
Temperature Measurement	18	Step 9: Operating Points for Temperature Channels	52
Additional ADC Functions for		Step 10: High and Low Limits for Temperature Channels ...	53
Temperature Measurement	21	Step 11: Monitoring $THERM$	56
Limits, Status Registers, and Interrupts.....	22	Step 12: Ramp Rate for Acoustic Enhancement.....	57
Limit Values.....	22	Enhancing System Acoustics.....	59
Status Registers	23	Acoustic Enhancement Mode Overview	59
Interrupts	23	Register Tables.....	61
Active Cooling	28	ADT7468 Programming Block Diagram	80
Driving the Fan Using PWM Control	28	Outline Dimensions	81
Laying Out 2-Wire and 3-Wire Fans	30	Ordering Guide	81
Fan Speed Measurement.....	31		
Fan Spin-Up	33		

REVISION HISTORY

SPECIFICATIONS

$T_A = T_{MIN}$ to T_{MAX} , $V_{CC} = V_{MIN}$ to V_{MAX} , unless otherwise noted.

All voltages are measured with respect to GND, unless otherwise specified. Typicals are at $T_A = 25^\circ\text{C}$ and represent most likely parametric norms. Logic inputs accept input high voltages up to V_{MAX} even when device is operating down to V_{MIN} . Timing specifications are tested at logic levels of $V_{IL} = 0.8\text{ V}$ for a falling edge and $V_{IH} = 2.0\text{ V}$ for a rising edge. SMBus timing specifications are guaranteed by design and are not production tested.

Table 1.

Parameter	Min	Typ	Max	Unit	Test Conditions/Comments
POWER SUPPLY					
Supply Voltage	3.0	3.3	5.5	V	
Supply Current, I_{CC}			3 20	mA μA	Interface inactive, ADC active Standby mode
TEMP-TO-DIGITAL CONVERTER					
Local Sensor Accuracy			± 1.5	$^\circ\text{C}$	$0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$
	-3.5		+2	$^\circ\text{C}$	$-40^\circ\text{C} \leq T_A \leq +100^\circ\text{C}$
	-4		+2	$^\circ\text{C}$	$-40^\circ\text{C} \leq T_A \leq +120^\circ\text{C}$
Resolution		0.25		$^\circ\text{C}$	
Remote Diode Sensor Accuracy			± 1.5	$^\circ\text{C}$	$0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$; $0^\circ\text{C} \leq T_D \leq 120^\circ\text{C}$
	-3.5		+2	$^\circ\text{C}$	$-40^\circ\text{C} \leq T_A \leq +100^\circ\text{C}$; $0^\circ\text{C} \leq T_D \leq +120^\circ\text{C}$
	-4.5		+2	$^\circ\text{C}$	$-40^\circ\text{C} \leq T_A \leq +120^\circ\text{C}$; $0^\circ\text{C} \leq T_D \leq +120^\circ\text{C}$
Resolution		0.25		$^\circ\text{C}$	
Remote Sensor Source Current		6		μA	First current
		36		μA	Second current
		96		μA	Third current
ANALOG-TO-DIGITAL CONVERTER (INCLUDING MUX AND ATTENUATORS)					
Total Unadjusted Error (TUE)			± 2	%	For 12 V and 5 V channels
			± 1.5	%	For all other channels
Differential Nonlinearity (DNL)			± 1	LSB	8 bits
Power Supply Sensitivity		± 0.1		%/V	
Conversion Time (Voltage Input)		11		ms	Averaging enabled
Conversion Time (Local Temperature)		12		ms	Averaging enabled
Conversion Time (Remote Temperature)		38		ms	Averaging enabled
Total Monitoring Cycle Time		145		ms	Averaging enabled
Total Monitoring Cycle Time		19		ms	Averaging disabled
Input Resistance	40	70	100	k Ω	For V_{CC} channel
	80	140	200	k Ω	For all other channels
FAN RPM-TO-DIGITAL CONVERTER					
Accuracy			± 5	%	$0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$, 3.3 V
			± 7	%	$-40^\circ\text{C} \leq T_A \leq +120^\circ\text{C}$, 3.3 V
			± 10	%	$-40^\circ\text{C} \leq T_A \leq +120^\circ\text{C}$, 5.5 V
Full-Scale Count			65,535		
Nominal Input RPM		109		RPM	Fan count = 0xBFFF
		329		RPM	Fan count = 0x3FFF
		5000		RPM	Fan count = 0x0438
		10000		RPM	Fan count = 0x021C
Internal Clock Frequency	85.5	90	94.5	kHz	$0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$, $V_{CC} = 3.3\text{ V}$
	83.7	90	96.3	kHz	$-40^\circ\text{C} \leq T_A \leq +120^\circ\text{C}$, $V_{CC} = 3.3\text{ V}$
	81	90	99	kHz	$-40^\circ\text{C} \leq T_A \leq +120^\circ\text{C}$, $V_{CC} = 5.5\text{ V}$

ADT7468

Parameter	Min	Typ	Max	Unit	Test Conditions/Comments
OPEN-DRAIN DIGITAL OUTPUTS, PWM1 TO PWM3, XTO					
Current Sink, I_{OL}			8.0	mA	
Output Low Voltage, V_{OL}			0.4	V	$I_{OUT} = -8.0 \text{ mA}$, $V_{CC} = +3.3 \text{ V}$
High Level Output Current, I_{OH}		0.1	1.0	μA	$V_{OUT} = V_{CC}$
OPEN-DRAIN SERIAL DATA BUS OUTPUT (SDA)					
Output Low Voltage, V_{OL}			0.4	V	$I_{OUT} = -4.0 \text{ mA}$, $V_{CC} = +3.3 \text{ V}$
High Level Output Current, I_{OH}		0.1	1.0	μA	$V_{OUT} = V_{CC}$
SMBUS DIGITAL INPUTS (SCL, SDA)					
Input High Voltage, V_{IH}	2.0			V	
Input Low Voltage, V_{IL}			0.4	V	
Hysteresis		500		mV	
DIGITAL INPUT LOGIC LEVELS (TACH INPUTS)					
Input High Voltage, V_{IH}	2.0		5.5	V	Maximum input voltage
Input Low Voltage, V_{IL}	-0.3		0.8	V	Minimum input voltage
Hysteresis		0.5		V p-p	
DIGITAL INPUT LOGIC LEVELS (THERM) ADTL+					
Input High Voltage, V_{IH}		$0.75 \times V_{CCP}$		V	
Input Low Voltage, V_{IL}			0.4	V	
DIGITAL INPUT CURRENT					
Input High Current, I_{IH}	-1			μA	$V_{IN} = V_{CC}$
Input Low Current, I_{IL}			1	μA	$V_{IN} = 0$
Input Capacitance, C_{IN}		5		pF	
SERIAL BUS TIMING					See Figure 2
Clock Frequency, f_{SCLK}	10		400	kHz	
Glitch Immunity, t_{SW}			50	ns	
Bus Free Time, t_{BUF}	4.7			μs	
Start Setup Time, $t_{SU;STA}$	4.7			μs	
Start Hold Time, $t_{HD;STA}$	4.0			μs	
SCL Low Time, t_{LOW}	4.7			μs	
SCL High Time, t_{HIGH}	4.0		50	μs	
SCL, SDA Rise Time, t_R			1000	ns	
SCL, SDA Fall Time, t_F			300	μs	
Data Setup Time, $t_{SU;DAT}$	250			ns	
Data Hold Time, $t_{HD;DAT}$	300			ns	
Detect Clock Low Timeout, $t_{TIMEOUT}$	15		35	ms	Can be optionally disabled

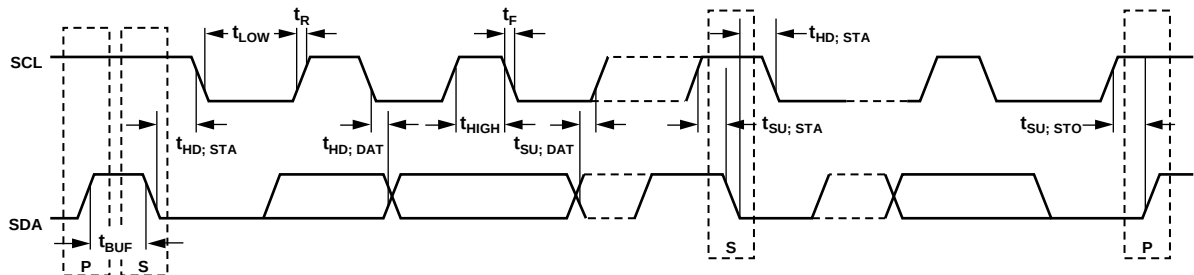


Figure 2. Serial Bus Timing Diagram

ABSOLUTE MAXIMUM RATINGS

Table 2.

Parameter	Rating
Positive Supply Voltage (V_{CC})	5.5 V
Maximum Voltage on 12 V_{IN} Pin	20 V
Voltage on Any Input or Output Pin	−0.3 V to +6.5 V
Input Current at Any Pin	±5 mA
Package Input Current	±20 mA
Maximum Junction Temperature (T_{JMAX})	150°C
Storage Temperature Range	−65°C to +150°C
Lead Temperature, Soldering	
IR Reflow Peak Temperature	220°C
For Pb-free	260°C
Lead Temperature (Soldering 10 sec)	300°C
ESD Rating	1000 V

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Thermal Characteristics

24-lead QSOP package:

$$\theta_{JA} = 150^{\circ}\text{C/W}$$

$$\theta_{JC} = 39^{\circ}\text{C/W}$$

ESD CAUTION

ESD (electrostatic discharge) sensitive device. Electrostatic charges as high as 4000 V readily accumulate on the human body and test equipment and can discharge without detection. Although this product features proprietary ESD protection circuitry, permanent damage may occur on devices subjected to high energy electrostatic discharges. Therefore, proper ESD precautions are recommended to avoid performance degradation or loss of functionality.



PIN CONFIGURATION AND FUNCTION DESCRIPTIONS

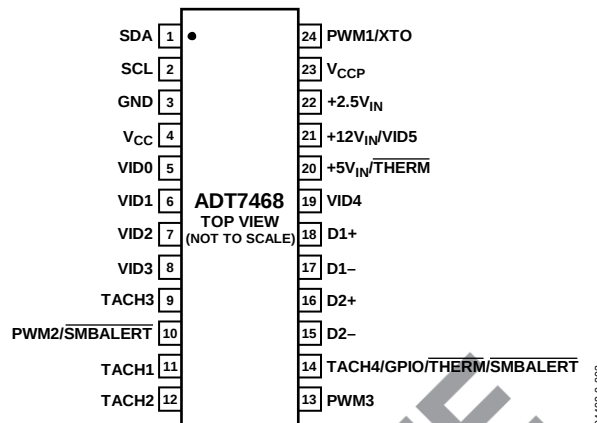


Figure 3. Pin Configuration

Table 3. Pin Function Descriptions

Pin No.	Mnemonic	Description
1	SDA	Digital I/O (Open Drain). SMBus bidirectional serial data. Requires pull-up resistor.
2	SCL	Digital Input (Open Drain). SMBus serial clock input. Requires pull-up resistor.
3	GND	Ground Pin.
4	V _{CC}	Power Supply. Can be powered by 3.3 V standby, if monitoring in low power states is required. V _{CC} is also monitored through this pin. The ADT7468 can also be powered from a 5 V supply. Setting Bit 7 of Configuration Register 1 (Reg. 0x40) rescales the V _{CC} input attenuators to correctly measure a 5 V supply.
5	VID0	Digital Input (Open Drain). Voltage supply readouts from CPU. This value is read into the VID register (Reg. 0x43).
6	VID1	Digital Input (Open Drain). Voltage supply readouts from CPU. This value is read into the VID register (Reg. 0x43).
7	VID2	Digital Input (Open Drain). Voltage supply readouts from CPU. This value is read into the VID register (Reg. 0x43).
8	VID3	Digital Input (Open Drain). Voltage supply readouts from CPU. This value is read into the VID register (Reg. 0x43).
9	TACH3	Digital Input (Open Drain). Fan tachometer input to measure speed of Fan 3. Can be reconfigured as an analog input (AIN3) to measure the speed of 2-wire fans.
10	PWM2	Digital Output (Open Drain). Requires 10 kΩ typical pull-up. Pulse width modulated output to control Fan 2 speed. Can be configured as a high or low frequency drive.
	<u>SMBALERT</u>	Digital Output (Open Drain). This pin can be reconfigured as an <u>SMBALERT</u> interrupt output to signal out-of-limit conditions.
11	TACH1	Digital Input (Open Drain). Fan tachometer input to measure speed of Fan 1. Can be reconfigured as an analog input (AIN1) to measure the speed of 2-wire fans.
12	TACH2	Digital Input (Open Drain). Fan tachometer input to measure speed of Fan 2. Can be reconfigured as an analog input (AIN2) to measure the speed of 2-wire fans.
13	PWM3	Digital I/O (Open Drain). Pulse width modulated output to control speed of Fan 3 and Fan 4. Requires 10 kΩ typical pull-up. Can be configured as a high or low frequency drive.
14	TACH4	Digital Input (Open Drain). Fan tachometer input to measure speed of Fan 4. Can be reconfigured as an analog input (AIN4) to measure the speed of 2-wire fans.
	<u>GPIO</u>	General Purpose Open Drain Digital I/O.
	<u>THERM</u>	Alternatively, the pin can be reconfigured as a bidirectional <u>THERM</u> pin. Can be used to time and monitor assertions on the <u>THERM</u> input. For example, this pin can be connected to the <u>PROCHOT</u> output of an Intel Pentium 4 processor or to the output of a trip point temperature sensor. This pin can also be used as an output to signal overtemperature conditions.
	<u>SMBALERT</u>	Digital Output (Open Drain). This pin can be reconfigured as an <u>SMBALERT</u> interrupt output to signal out-of-limit conditions.
15	D2-	Cathode Connection to Second Thermal Diode.

Pin No.	Mnemonic	Description
16	D2+	Anode Connection to Second Thermal Diode.
17	D1-	Cathode Connection to First Thermal Diode.
18	D1+	Anode Connection to First Thermal Diode.
19	VID4	Digital Input (Open Drain). Voltage supply readouts from CPU. This value is read into the VID register (Reg. 0x43).
20	+5V _{IN} THERM	Analog Input. Monitors +5 V power supply. Alternatively, this pin can be reconfigured as a bidirectional THERM pin. Can be used to time and monitor assertions on the THERM input. For example, it can be connected to the PROCHOT output of an Intel Pentium 4 processor or to the output of a trip point temperature sensor. This pin can also be used as an output to signal overtemperature conditions.
21	+12V _{IN} VID5	Analog Input. Monitors +12 V power supply. Digital Input (Open Drain). Voltage supply readouts from CPU. This value is read into the VID register (Reg. 0x43). Supports VRM10 solutions.
22	+2.5V _{IN}	Analog Input. Monitors +2.5 V supply, typically a chipset voltage.
23	V _{CCP}	Analog Input. Monitors processor core voltage (0 V to 3 V).
24	PWM1 XTO	Digital Output (Open Drain). Pulse width modulated output to control Fan 1 speed. Requires 10 kΩ typical pull-up. Also functions as the output from the XOR tree in XOR test mode.

OBSOLETE
 THIS DEVICE IS OBSOLETE
 PLEASE CONTACT YOUR ON SEMICONDUCTOR
 REPRESENTATIVE FOR INFORMATION

TYPICAL PERFORMANCE CHARACTERISTICS

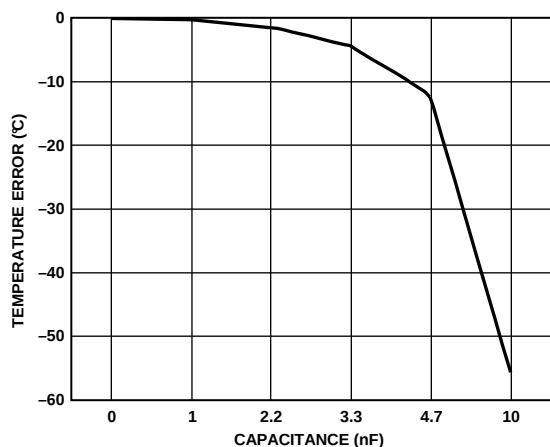


Figure 4. Temperature Error vs. Capacitance between D+ and D-

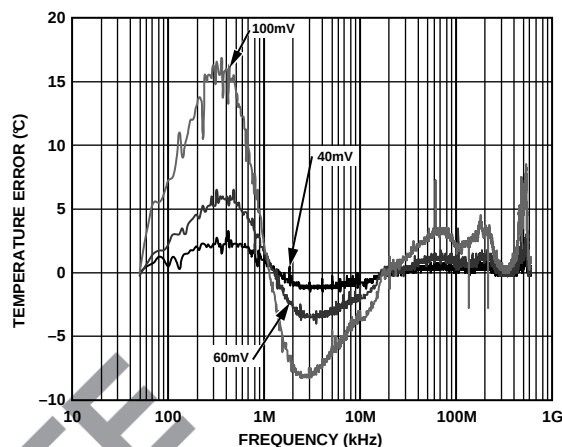


Figure 7. Remote Temperature Error vs. Common Mode Noise Frequency

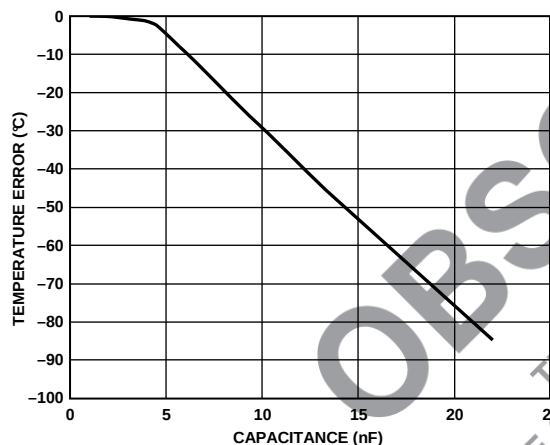


Figure 5. External Temperature Error vs. D+/D- Capacitance

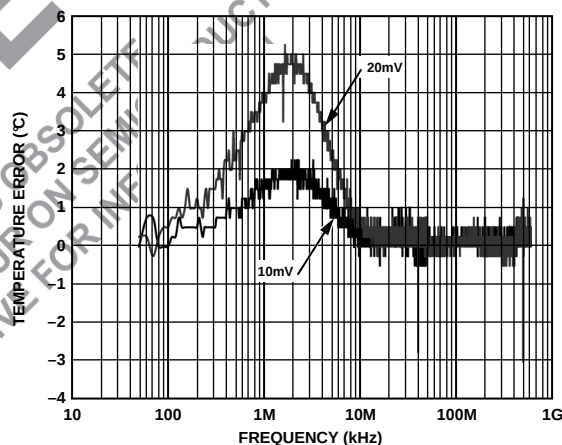


Figure 8. Remote Temperature Error vs. Differential Mode Noise Frequency

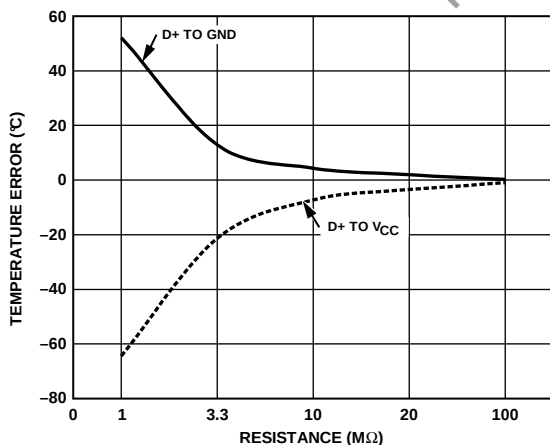
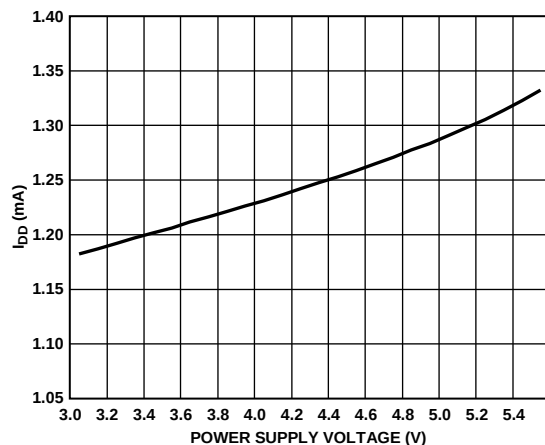


Figure 6. Remote Temperature Error vs. PCB Resistance

Figure 9. Normal I_{DD} vs. Power Supply

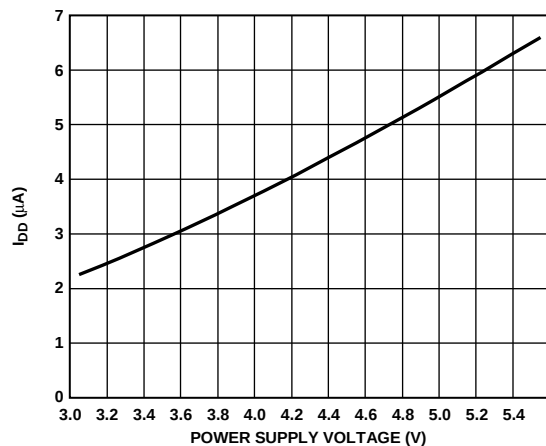


Figure 10. Shutdown I_{DD} vs. Power Supply

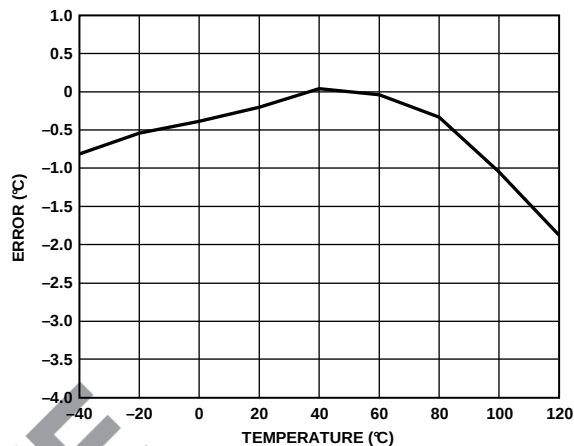


Figure 13. Internal Temperature Error vs. ADT7468 Temperature

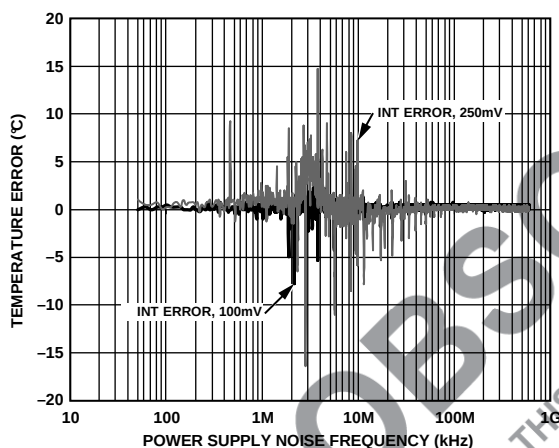


Figure 11. Internal Temperature Error vs. Power Supply

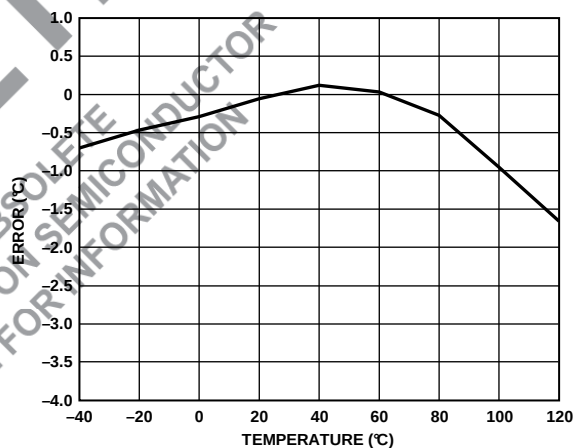


Figure 14. Remote Temperature Error vs. ADT7468 Temperature

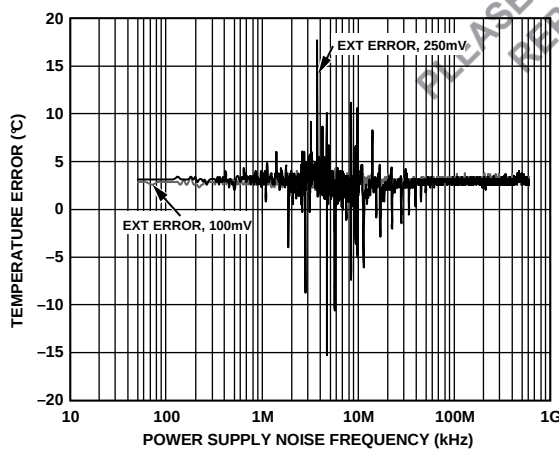


Figure 12. Remote Temperature Error vs. Power Supply Noise Frequency

PRODUCT DESCRIPTION

The ADT7468 is a complete thermal monitor and multiple fan controller for any system requiring thermal monitoring and cooling. The device communicates with the system via a serial system management bus. The serial bus controller has a serial data line for reading and writing addresses and data (Pin 1), and an input line for the serial clock (Pin 2). All control and programming functions for the ADT7468 are performed over the serial bus. In addition, a pin can be reconfigured as an SMBALERT output to signal out-of-limit conditions.

COMPARISON BETWEEN ADT7463 AND ADT7468

The ADT7468 is an upgrade to the ADT7463. The ADT7468 and ADT7463 are almost pin and register map compatible. The ADT7468 and ADT7463 have the following differences:

1. On the ADT7468, the PWM drive signals can be configured as either high frequency or low frequency drives. The low frequency option is programmable between 10 Hz and 100 Hz. The high frequency option is 22.5 kHz. On the ADT7463, only the low frequency option is available.
2. Once V_{CC} is powered up, monitoring of temperature and fan speeds is enabled on the ADT7468 when V_{CCP} is powered up. If V_{CCP} is never powered up, this is enabled when the first SMBus transaction with the ADT7468 is completed. On the ADT7463, the STRT bit in Configuration Register 1 must be set to enable monitoring.
3. The fans are switched off by default upon power-up. On the ADT7463, the fans run at full speed on power-up.

Fail-safe cooling is provided such that when the measured temperature exceeds the THERM limit (100°C), the fans run at full speed.

Fail-safe cooling is also provided 4.6 secs after V_{CCP} is powered-up (see Figure 48). The fans operate at full speed if the ADT7468 has not been addressed via the SMBus within 4.6 secs of when the V_{CCP} is powered up. This protects the system in the event that the SMBus fails. The ADT7468 can be programmed at any time, and it behaves as programmed. If V_{CCP} is never powered-up, fail-safe cooling is effectively disabled. If V_{CCP} is disabled, writing to the ADT7468 at any time causes the ADT7468 to operate normally.
4. Series resistance cancellation (SRC) is provided on the remote temperature channels on the ADT7468, but not on the ADT7463. SRC automatically cancels linear offset introduced by a series resistance between the thermal diode and the sensor.
5. The ADT7468 has an extended temperature measurement range. The measurement range goes from -64°C to +191°C.

On the ADT7463, the measurement range is from -127°C to +127°C. This means that the ADT7468 can measure higher temperatures. The ADT7468 also includes the ADT7463 temperature range; the temperature measurement range can be switched by setting Bit 0 of Configuration Register 5.

6. The ADT7468 maximum fan speed (% duty cycle) in the automatic fan speed control loop can be programmed. The maximum fan speed is 100% duty cycle on the ADT7463 and is not programmable.
7. The offset register in the ADT7468 is programmable up to $\pm 64^\circ\text{C}$ with 0.50°C resolution. The offset register of the ADT7463 is programmable up to $\pm 32^\circ\text{C}$ with 0.25°C resolution.
8. V_{CCP} is monitored on Pin 23 of the ADT7468 and can be used to set the threshold for THERM (PROCHOT) (2/3 of V_{CCP}). The threshold for THERM (PROCHOT) is set at $V_{IH} = 1.7\text{ V}$ and $V_{IL} = 0.8\text{ V}$ on the ADT7463.
9. On the ADT7463, Pin 22 can be reconfigured as SMBus ALERT. This is not available on the ADT7468; instead, SMBALERT can be enabled on Pin 14.
10. A GPIO can also be made available on Pin 14 on the ADT7468. This is not available on the ADT7463. Set the GPIO polarity and direction in Configuration Register 5. The GPIO status bit is Bit 5 of Status Register 2 (shared with TACH4 and THERM, because only one can be enabled at a time).
11. The ADT7463 has three possible SMBus addresses, which are selectable using the address select and address enable pins. The ADT7468 has one SMBus address available at Address 0x2E.

Due to the inclusion of extra functionality, the register map has changed, including an additional configuration register: Configuration Register 5 at Address 0x7C.

Configuration Register 5

Bit 0: If Bit 0 is set to 1, in terms of temperature the ADT7468 is backward-compatible with the ADT7463. Measurements including the T_{MIN} calibration circuit, and fan control work in the range of -127°C to +127°C. Also, care should be taken in reprogramming the temperature limits (T_{MIN} , operating point, and THERM limits) to their desired twos complement value, because the power-on default for them is at Offset 64. The extended temperature range is -64°C to 191°C. The default is 1, which is in the -64°C to +191°C temperature range.

Bit 1 = 0 is the high frequency (22.5 kHz) fan drive signal.

Bit 1 = 1 switches the fan drive to low frequency PWM, programmable between 10 Hz and 100 Hz, the same as the ADT7463. The default = 0 = HF PWM.

Bit 2 sets the direction for the GPIO: 0 = input, 1 = output.

Bit 3 sets the GPIO polarity: 0 = active low, 1 = active high.

Setting the Functionality of Pin 14

Pin 14 on the ADT7468 has four possible functions: SMBALERT, THERM, GPIO, and TACH4. The user chooses the required functionality by setting Bit 0 and Bit 1 of Configuration Register 4 at Address 0x7D.

Table 4. Pin 14 Settings

Bit 0	Bit 1	Function
0	0	TACH4
0	1	THERM
1	0	SMBALERT
1	1	GPIO

RECOMMENDED IMPLEMENTATION

Configuring the ADT7468 as in Figure 15 allows the system designer to use the following features:

- Two PWM outputs for fan control of up to three fans (the front and rear chassis fans are connected in parallel).
- Three TACH fan speed measurement inputs.
- V_{CC} measured internally through Pin 3.
- CPU temperature measured using Remote 1 temperature channel.
- Ambient temperature measured through Remote 2 temperature channel.
- Bidirectional THERM pin. This feature allows Intel Pentium 4 PROCHOT monitoring and can function as an overtemperature THERM output. Alternatively, it can be programmed as an SMBALERT system interrupt output.

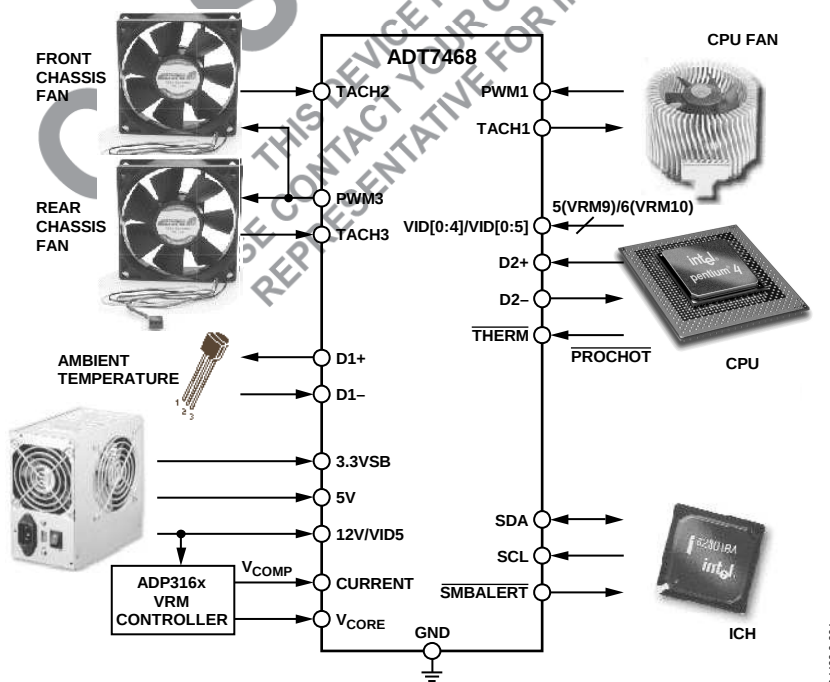


Figure 15. ADT7468 Configuration

SERIAL BUS INTERFACE

On PCs and servers, control of the ADT7468 is carried out using the serial system management bus (SMBus). The ADT7468 is connected to this bus as a slave device, under the control of a master controller, which is usually (but not necessarily) the ICH.

The ADT7468 has a fixed 7-bit serial bus address of 0101110 or 0x2E. The read/write bit must be added to get the 8-bit address (01011100 or 0x5C). Data is sent over the serial bus in sequences of nine clock pulses: eight bits of data followed by an acknowledge bit from the slave device. Transitions on the data line must occur during the low period of the clock signal and remain stable during the high period, because a low-to-high transition might be interpreted as a stop signal when the clock is high. The number of data bytes that can be transmitted over the serial bus in a single read or write operation is limited only by what the master and slave devices can handle.

When all data bytes have been read or written, stop conditions are established. In write mode, the master pulls the data line high during the 10th clock pulse to assert a stop condition. In read mode, the master device overrides the acknowledge bit by pulling the data line high during the low period before the ninth clock pulse. This is known as a no acknowledge. The master then takes the data line low during the low period before the 10th clock pulse, and then high during the 10th clock pulse to assert a stop condition.

Any number of bytes of data can be transferred over the serial bus in one operation, but it is not possible to mix read and write in one operation, because the type of operation is determined at the beginning and cannot subsequently be changed without starting a new operation.

In the ADT7468, write operations contain either one or two bytes, and read operations contain one byte and perform the

following functions. To write data to one of the device data registers or read data from it, the address pointer register must be set so that the correct data register is addressed, then data can be written into that register or read from it. The first byte of a write operation always contains an address that is stored in the address pointer register. If data is to be written to the device, then the write operation contains a second data byte that is written to the register selected by the address pointer register. This write operation is illustrated in Figure 16. The device address is sent over the bus, and then R/W is set to 0. This is followed by two data bytes. The first data byte is the address of the internal data register to be written to, which is stored in the address pointer register. The second data byte is the data to be written to the internal data register.

When reading data from a register, there are two possibilities:

- If the ADT7468's address pointer register value is unknown or not the desired value, it must be set to the correct value before data can be read from the desired data register. This is done by performing a write to the ADT7468, but only the data byte containing the register address is sent, since no data is written to the register. This is shown in Figure 17.

A read operation is then performed consisting of the serial bus address, R/W, bit set to 1, followed by the data byte read from the data register. This is shown in Figure 18.

- If the address pointer register is known to be at the desired address, data can be read from the corresponding data register without first writing to the address pointer register, as shown in Figure 18

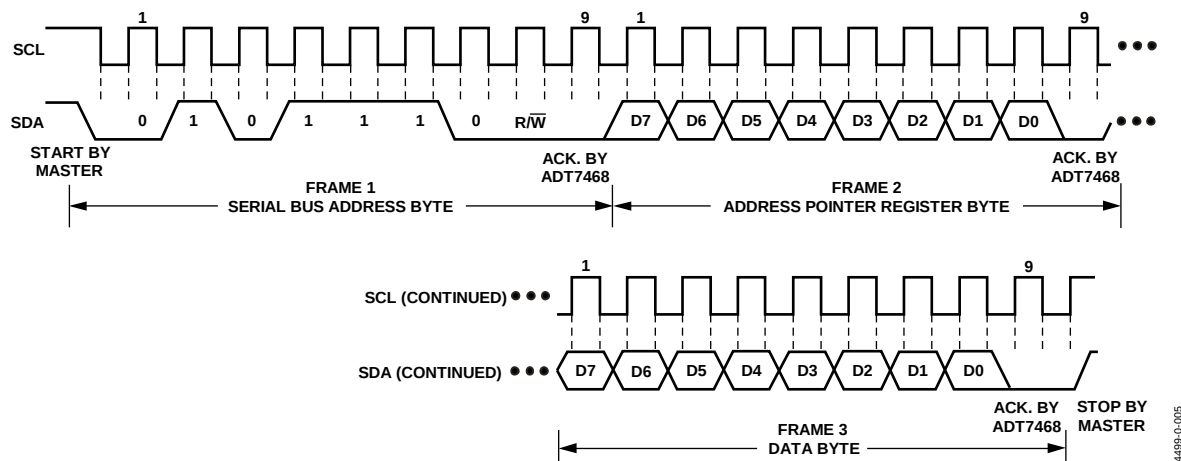


Figure 16. Writing a Register Address to the Address Pointer Register, then Writing Data to the Selected Register

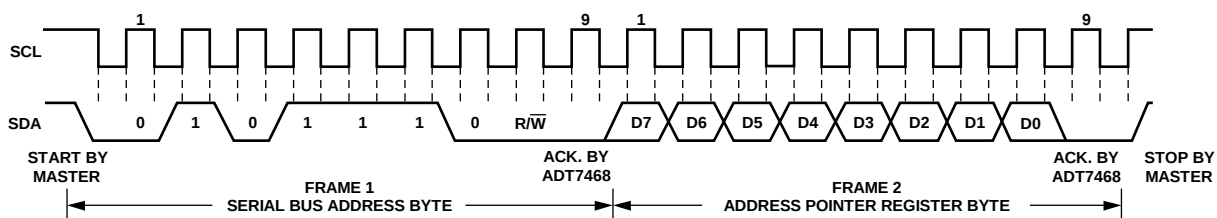


Figure 17. Writing to the Address Pointer Register Only

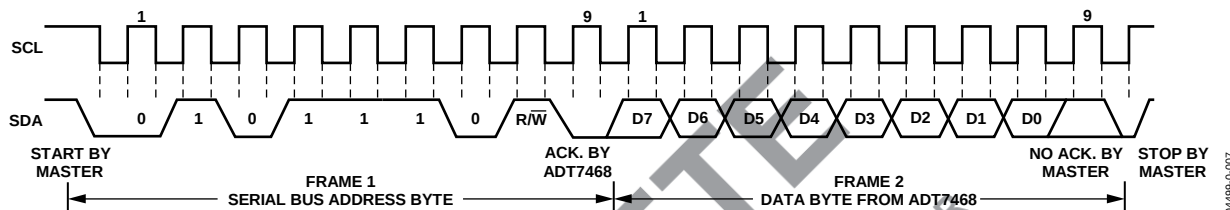


Figure 18. Reading Data from a Previously Selected Register

It is possible to read a data byte from a data register without first writing to the address pointer register, if the address pointer register is already at the correct value. However, it is not possible to write data to a register without writing to the address pointer register, because the first data byte of a write is always written to the address pointer register.

In addition to supporting the send byte and receive byte protocols, the ADT7468 also supports the read byte protocol. (See Intel's *System Management Bus Specifications Rev. 2* for more information.)

If several read or write operations must be performed in succession, the master can send a repeat start condition instead of a stop condition to begin a new operation.

WRITE OPERATIONS

The SMBus specification defines several protocols for different types of read and write operations. The ones used in the ADT7468 are discussed below. The following abbreviations are used in the diagrams:

S: START
P: STOP
R: READ
W: WRITE
A: ACKNOWLEDGE
 $\bar{A}$: NO ACKNOWLEDGE

The ADT7468 uses the following SMBus write protocols.

Send Byte

In this operation, the master device sends a single command byte to a slave device as follows:

1. The master device asserts a start condition on SDA.

2. The master sends the 7-bit slave address followed by the write bit (low).
3. The addressed slave device asserts ACK on SDA.
4. The master sends a command code.
5. The slave asserts ACK on SDA.
6. The master asserts a stop condition on SDA and the transaction ends.

For the ADT7468, the send byte protocol is used to write a register address to RAM for a subsequent single byte read from the same address. This operation is illustrated in Figure 19.

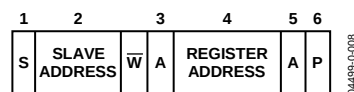


Figure 19. Setting a Register Address for a Subsequent Read

If the master is required to read data from the register immediately after setting up the address, it can assert a repeat start condition immediately after the final ACK and carry out a single byte read without asserting an intermediate stop condition.

Write Byte

In this operation, the master device sends a command byte and one data byte to the slave device, as follows:

1. The master device asserts a start condition on SDA.
2. The master sends the 7-bit slave address followed by the write bit (low).
3. The addressed slave device asserts ACK on SDA.
4. The master sends a command code.
5. The slave asserts ACK on SDA.

6. The master sends a data byte.
7. The slave asserts ACK on SDA.
8. The master asserts a stop condition on SDA to end the transaction.

This operation is illustrated in Figure 20.

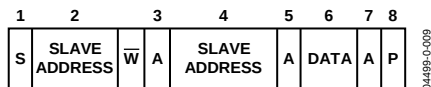


Figure 20. Single Byte Write to a Register

READ OPERATIONS

The ADT7468 uses the following SMBus read protocols.

Receive Byte

This operation is useful when repeatedly reading a single register. The register address must have been set up previously. In this operation, the master device receives a single byte from a slave device as follows:

1. The master device asserts a start condition on SDA.
2. The master sends the 7-bit slave address followed by the read bit (high).
3. The addressed slave device asserts ACK on SDA.
4. The master receives a data byte.
5. The master asserts NO ACK on SDA.
6. The master asserts a stop condition on SDA and the transaction ends.

In the ADT7468, the receive byte protocol is used to read a single byte of data from a register whose address has previously been set by a send byte or write byte operation. This operation is illustrated in Figure 21.

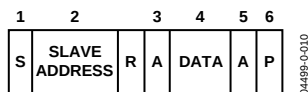


Figure 21. Single Byte Read from a Register

Alert Response Address

Alert response address (ARA) is a feature of SMBus devices that allows an interrupting device to identify itself to the host when multiple devices exist on the same bus.

The SMBALERT output can be used as either an interrupt output or an SMBALERT. One or more outputs can be connected to a common SMBALERT line connected to the master. If a device's SMBALERT line goes low, the following procedure occurs:

1. SMBALERT is pulled low.
2. The master initiates a read operation and sends the alert response address (ARA = 0001 100). This is a general call address that must not be used as a specific device address.
3. The device whose SMBALERT output is low responds to the alert response address, and the master reads its device address. The address of the device is now known and can be interrogated in the usual way.
4. If more than one device's SMBALERT output is low, the one with the lowest device address has priority in accordance with normal SMBus arbitration.
5. Once the ADT7468 has responded to the alert response address, the master must read the status registers and the SMBALERT is cleared only if the error condition is absent.

SMBUS TIMEOUT

The ADT7468 includes an SMBus timeout feature. If there is no SMBus activity for 35 ms, the ADT7468 assumes that the bus is locked and releases the bus. This prevents the device from locking or holding the SMBus expecting data. Some SMBus controllers cannot handle the SMBus timeout feature, so it can be disabled.

Configuration Register 1 (Reg. 0x40)

<6> TODIS = 0, SMBus timeout enabled (default).

<6> TODIS = 1, SMBus timeout disabled.

VID CODE MONITORING

The ADT7468 has five dedicated voltage ID (VID code) inputs. These are digital inputs that can be read back through the VID register (Reg. 0x43) to determine the processor voltage required or being used in the system. Five VID code inputs support VRM9.x solutions. In addition, Pin 21 (12 V input) can be reconfigured as a sixth VID input to satisfy future VRM requirements.

VID CODE REGISTERS

VID Code Register 0x43

<0> = VID0, reflects the logic state of Pin 5.

<1> = VID1, reflects the logic state of Pin 6.

<2> = VID2, reflects the logic state of Pin 7.

<3> = VID3, reflects the logic state of Pin 8.

<4> = VID4, reflects the logic state of Pin 19.

<5> = VID5, reconfigurable 12 V input. This bit reads 0 when Pin 21 is configured as the 12 V input. This bit reflects the logic state of Pin 21 when the pin is configured as VID5.

<6> THLD = 0, VID switching threshold = 1 V,
 $V_{OL} < 0.8 \text{ V}$, $V_{IH} > 1.7 \text{ V}$, $V_{MAX} = 3.3 \text{ V}$

THLD = 1, VID switching threshold = 0.6 V,
 $V_{OL} < 0.4 \text{ V}$, $V_{IH} > 0.8 \text{ V}$, $V_{MAX} = 3.3 \text{ V}$

<7> VIDSEL = 0, Pin 21 functions as a 12 V measurement input. Software can read this bit to determine that there are five VID inputs being monitored. Bit 5 of Register 0x43 (VID5) always reads back 0. Bit 0 of Status Register 2 (Reg. 0x42) reflects 12 V out-of-limit measurements.

VIDSEL = 1, Pin 21 functions as the sixth VID code input (VID5). Software can read this bit to determine that there are six VID inputs being monitored. Bit 5 of Register 0x43 reflects the logic state of Pin 21. Bit 0 of Status Register 2 (Reg. 0x42) reflects VID code changes.

VID Code Input Threshold Voltage

The switching threshold for the VID code inputs is approximately 1 V. To enable future compatibility, it is possible to reduce the VID code input threshold to 0.6 V. Bit 6 (THLD) of the VID register (Reg. 0x43) controls the VID input threshold voltage.

Reconfiguring Pin 21 as VID5 Input

Pin 21 can be reconfigured as a sixth VID code input (VID5) for VRM10-compatible systems. Because the pin is configured as VID5, it is not possible to monitor a 12 V supply.

Bit 7 of the VID register (Reg. 0x43) determines the function of Pin 21. System or BIOS software can read the state of Bit 7 to determine whether the system is designed to monitor 12 V or is monitoring a sixth VID input.

Status Register 2 (Reg. 0x42)

<0> 12 V/VC = 0, if Pin 21 is configured as VID5, then Logic 0 denotes no change in VID code within the last 11 μs .

<0> 12 V/VC = 1, if Pin 21 is configured as VID5, then Logic 1 means that a change has occurred on the VID code inputs within the last 11 μs . An SMBALERT is generated, if this function is enabled.

VID Code Change Detect Function

The ADT7468 has a VID code change detect function. When Pin 21 is configured as the VID5 input, VID code changes can be detected and reported back by the ADT7468. Bit 0 of Status Register 2 (Reg. 0x42) is the 12 V/VC bit and denotes a VID change when set. The VID code change bit is set when the logic states on the VID inputs are different than they were 11 μs previously. The change of VID code can be used to generate an SMBALERT interrupt. If an SMBALERT interrupt is not required, Bit 0 of Interrupt Mask Register 2 (Reg. 0x75), when set, prevents SMBALERTs from occurring on VID code changes.

ANALOG-TO-DIGITAL CONVERTER

All analog inputs are multiplexed into the on-chip, successive approximation, analog-to-digital converter, which has a resolution of 10 bits. The basic input range is 0 V to 2.25 V, but the inputs have built-in attenuators to allow measurement of 2.5 V, 3.3 V, 5 V, 12 V, and the processor core voltage V_{CCP} without any external components. To allow for the tolerance of these supply voltages, the ADC produces an output of 3/4 full scale (decimal 768 or 300 hex) for the nominal input voltage and therefore has adequate headroom to cope with overvoltages.

VOLTAGE MEASUREMENT INPUT

The ADT7468 has four external voltage measurement channels and can measure its own supply voltage, V_{CC} . Pins 20 to 23 can measure 5 V, 12 V, 2.5 V supplies, and the processor core voltage V_{CCP} (0 V to 3 V input). The V_{CC} supply voltage measurement is carried out through the V_{CC} pin (Pin 4). Setting Bit 7 of Configuration Register 1 (Reg. 0x40) allows a 5 V supply to power the ADT7468 and be measured without overranging the V_{CC} measurement channel. The 2.5 V input can be used to monitor a chipset supply voltage in computer systems.

Input Circuitry

The internal structure for the analog inputs is shown in Figure 22. The input circuit consists of an input protection diode, an attenuator, and a capacitor to form a first-order low-pass filter that gives input immunity to high frequency noise.

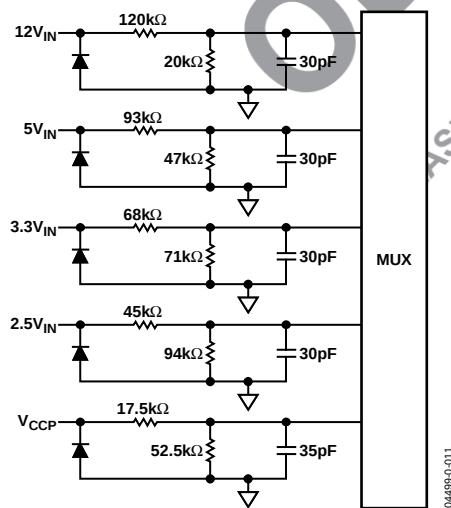


Figure 22. Structure of Analog Inputs

Voltage Measurement Registers

Reg. 0x20, 2.5 V reading = 0x00 default

Reg. 0x21, V_{CCP} reading = 0x00 default

Reg. 0x22, V_{CC} reading = 0x00 default

Reg. 0x23, 5 V reading = 0x00 default

Reg. 0x24, 12 V reading = 0x00 default

Voltage Limit Registers

Associated with each voltage measurement channel is a high and low limit register. Exceeding the programmed high or low limit causes the appropriate status bit to be set. Exceeding either limit can also generate SMBALERT interrupts.

Reg. 0x44, 2.5 V low limit = 0x00 default

Reg. 0x45, 2.5 V high limit = 0xFF default

Reg. 0x46, V_{CCP} low limit = 0x00 default

Reg. 0x47, V_{CCP} high limit = 0xFF default

Reg. 0x48, V_{CC} low limit = 0x00 default

Reg. 0x49, V_{CC} high limit = 0xFF default

Reg. 0x4A, 5 V low limit = 0x00 default

Reg. 0x4B, 5 V high limit = 0xFF default

Reg. 0x4C, 12 V Low Limit = 0x00 default

Reg. 0x4D, 12 V High Limit = 0xFF default

Table 6 shows the input ranges of the analog inputs and output codes of the 10-bit ADC.

When the ADC is running, it samples and converts a voltage input in 0.7 ms and averages 16 conversions to reduce noise; a measurement takes nominally 11 ms.

ADDITIONAL ADC FUNCTIONS FOR VOLTAGE MEASUREMENTS

A number of other functions are available on the ADT7468 to offer the system designer increased flexibility.

Turn-Off Averaging

For each voltage measurement read from a value register, 16 readings have actually been made internally and the results averaged before being placed into the value register. For instance, where faster conversions are needed, setting Bit 4 of Configuration Register 2 (Reg. 0x73) turns averaging off. This effectively gives a reading 16 times faster (0.7 ms), but the reading may be noisier.

Bypass Voltage Input Attenuator

Setting Bit 5 of Configuration Register 2 (Reg. 0x73) removes the attenuation circuitry from the 2.5 V, V_{CCP} , V_{CC} , 5 V, and 12 V inputs, which allows the user to directly connect external sensors or rescale the analog voltage measurement inputs for other applications. The input range of the ADC without the attenuators is 0 V to 2.25 V.

Single-Channel ADC Conversion

Setting Bit 6 of Configuration Register 2 (Reg. 0x73) places the ADT7468 into single-channel ADC conversion mode. In this mode, the ADT7468 can be made to read a single voltage channel only. If the internal ADT7468 clock is used, the selected input is read every 0.7 ms. The appropriate ADC channel is selected by writing to Bits <7:5> of the TACH1 minimum high byte register (0x55).

Configuration Register 2 (Reg. 0x73)

<4> = 1, averaging off.

<5> = 1, bypass input attenuators.

<6> = 1, single-channel convert mode.

TACH1 Minimum High Byte (Reg. 0x55)

<7:5> Selects ADC channel for single-channel convert mode.

Table 5. Programming the Single Channel ADC Function

Bits <7:5> Reg. 0x55	Channel Selected
000	2.5 V
001	V _{CCP}
010	V _{CC}
011	5 V
100	12 V
101	Remote 1 temperature
110	Local temperature
111	Remote 2 temperature

Table 6. 10-Bit A/D Output Code vs. V_{IN}

Input Voltage					A/D Output	
12 V _{IN}	5 V _{IN}	V _{CC} (3.3 V _{IN}) ¹	2.5 V _{IN}	V _{CCP}	Decimal	Binary (10 Bits)
<0.0156	<0.0065	<0.0042	<0.0032	<0.00293	0	00000000 00
0.0156–0.0312	0.0065–0.0130	0.0042–0.0085	0.0032–0.0065	0.00293–0.0058	1	00000000 01
0.0312–0.0469	0.0130–0.0195	0.0085–0.0128	0.0065–0.0097	0.0058–0.0087	2	00000000 10
0.0469–0.0625	0.0195–0.0260	0.0128–0.0171	0.0097–0.0130	0.0087–0.0117	3	00000000 11
0.0625–0.0781	0.0260–0.0325	0.0171–0.0214	0.0130–0.0162	0.0117–0.0146	4	00000001 00
0.0781–0.0937	0.0325–0.0390	0.0214–0.0257	0.0162–0.0195	0.0146–0.0175	5	00000001 01
0.0937–0.1093	0.0390–0.0455	0.0257–0.0300	0.0195–0.0227	0.0175–0.0205	6	00000001 10
0.1093–0.1250	0.0455–0.0521	0.0300–0.0343	0.0227–0.0260	0.0205–0.0234	7	00000001 11
0.1250–0.1406	0.0521–0.0586	0.0343–0.0386	0.0260–0.0292	0.0234–0.0263	8	00000010 00
4.0000–4.0156	1.6675–1.6740	1.1000–1.1042	0.8325–0.8357	0.7500–0.7529	256 (1/4 scale)	01000000 00
8.0000–8.0156	3.3300–3.3415	2.2000–2.2042	1.6650–1.6682	1.5000–1.5029	512 (1/2 scale)	10000000 00
12.0000–12.0156	5.0025–5.0090	3.3000–3.3042	2.4975–2.5007	2.2500–2.2529	768 (3/4 scale)	11000000 00
15.8281–15.8437	6.5983–6.6048	4.3527–4.3570	3.2942–3.2974	2.9677–2.9707	1013	11111101 01
15.8437–15.8593	6.6048–6.6113	4.3570–4.3613	3.2974–3.3007	2.9707–2.9736	1014	11111101 10
15.8593–15.8750	6.6113–6.6178	4.3613–4.3656	3.3007–3.3039	2.9736–2.9765	1015	11111101 11
15.8750–15.8906	6.6178–6.6244	4.3656–4.3699	3.3039–3.3072	2.9765–2.9794	1016	11111110 00
15.8906–15.9062	6.6244–6.6309	4.3699–4.3742	3.3072–3.3104	2.9794–2.9824	1017	11111110 01
15.9062–15.9218	6.6309–6.6374	4.3742–4.3785	3.3104–3.3137	2.9824–2.9853	1018	11111110 10
15.9218–15.9375	6.6374–6.6390	4.3785–4.3828	3.3137–3.3169	2.9853–2.9882	1019	11111110 11
15.9375–15.9531	6.6439–6.6504	4.3828–4.3871	3.3169–3.3202	2.9882–2.9912	1020	11111111 00
15.9531–15.9687	6.6504–6.6569	4.3871–4.3914	3.3202–3.3234	2.9912–2.9941	1021	11111111 01
15.9687–15.9843	6.6569–6.6634	4.3914–4.3957	3.3234–3.3267	2.9941–2.9970	1022	11111111 10
>15.9843	>6.6634	>4.3957	>3.3267	>2.9970	1023	11111111 11

¹ The V_{CC} output codes listed assume that V_{CC} is 3.3 V. If V_{CC} input is reconfigured for 5 V operation (by setting Bit 7 of Configuration Register 1), then the V_{CC} output codes are the same as for the 5 V_{IN} column.

TEMPERATURE MEASUREMENT

A simple method of measuring temperature is to exploit the negative temperature coefficient of a diode, measuring the base-emitter voltage (V_{BE}) of a transistor, operated at constant current. Unfortunately, this technique requires calibration to null the effect of the absolute value of V_{BE} , which varies from device to device.

The technique used in the ADT7468 is to measure the change in V_{BE} when the device is operated at three currents. Previous devices have used only two operating currents, but the use of a third current allows automatic cancellation of resistances in series with the external temperature sensor.

Figure 24 shows the input signal conditioning used to measure the output of an external temperature sensor. This figure shows the external sensor as a substrate transistor, but it could equally be a discrete transistor. If a discrete transistor is used, the collector is not grounded and should be linked to the base. To prevent ground noise from interfering with the measurement, the more negative terminal of the sensor is not referenced to ground, but is biased above ground by an internal diode at the D- input. C1 can optionally be added as a noise filter (the recommended maximum value is 1000 pF). However, a better option in noisy environments is to add a filter, as described in the Noise Filtering section.

Local Temperature Measurement

The ADT7468 contains an on-chip band gap temperature sensor whose output is digitized by the on-chip 10-bit ADC. The 8-bit MSB temperature data is stored in the local temperature register (Address 0x26). Because both positive and negative temperatures can be measured, the temperature data is stored in Offset 64 format or twos complement format, as shown in Table 7 and Table 8. Theoretically, the temperature sensor and ADC can measure temperatures from -128°C to $+127^{\circ}\text{C}$ (or -61°C to $+191^{\circ}\text{C}$ in the extended temperature range) with a resolution of 0.25°C . However, this exceeds the operating temperature range of the device, so local temperature measurements outside the ADT7468 operating temperature range are not possible.

Remote Temperature Measurement

The ADT7468 can measure the temperature of two remote diode sensors or diode-connected transistors connected to Pins 17 and 18, or Pins 15 and 16.

The forward voltage of a diode or diode-connected transistor operated at a constant current exhibits a negative temperature coefficient of about $-2 \text{ mV}/^{\circ}\text{C}$. Unfortunately, the absolute value of V_{BE} varies from device to device and individual calibration is required, and therefore the technique is unsuitable

for mass production. The technique used in the ADT7468 is to measure the change in V_{BE} when the device is operated at three currents.

This is given by

$$\Delta V_{BE} = KT / q \times \ln(N)$$

where:

K is Boltzmann's constant.

q is the charge on the carrier.

T is the absolute temperature in Kelvin.

N is the ratio of the two currents.

Figure 23 shows the input signal conditioning used to measure the output of a remote temperature sensor. This figure shows the external sensor as a substrate transistor, provided for temperature monitoring on some microprocessors. It could also be a discrete transistor such as a 2N3904/2N3906.

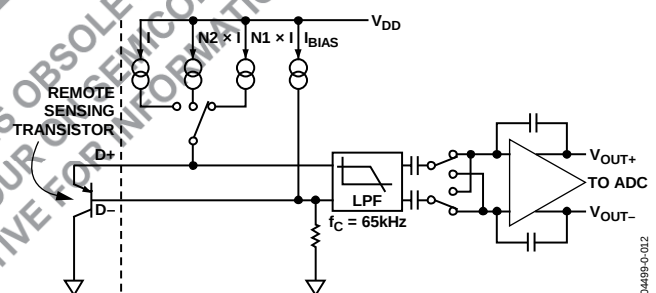


Figure 23. Signal Conditioning for Remote Diode Temperature Sensors

If a discrete transistor is used, the collector is not grounded and should be linked to the base. If a PNP transistor is used, the base is connected to the D- input and the emitter is connected to the D+ input. If an NPN transistor is used, the emitter is connected to the D- input and the base to the D+ input. Figure 25 and Figure 26 show how to connect the ADT7468 to an NPN or PNP transistor for temperature measurement. To prevent ground noise from interfering with the measurement, the more negative terminal of the sensor is not referenced to ground, but is biased above ground by an internal diode at the D- input.

To measure ΔV_{BE} , the operating current through the sensor is switched among three related currents. Shown in Figure 23, $N1 \times I$ and $N2 \times I$ are different multiples of the current I . The currents through the temperature diode are switched between I and $N1 \times I$, giving ΔV_{BE1} , and then between I and $N2 \times I$, giving ΔV_{BE2} . The temperature can then be calculated using the two ΔV_{BE} measurements. This method can also cancel the effect of any series resistance on the temperature measurement.

The resulting ΔV_{BE} waveforms are passed through a 65 kHz low-pass filter to remove noise and then sent to a chopper-stabilized amplifier that amplifies and rectifies the waveform to

produce a dc voltage proportional to ΔV_{BE} . The ADC digitizes this voltage, and a temperature measurement is produced. To reduce the effects of noise, digital filtering is performed by averaging the results of 16 measurement cycles.

The results of remote temperature measurements are stored in 10-bit, twos complement format, as illustrated in Table 7. The extra resolution for the temperature measurements is held in the Extended Resolution Register 2 (Reg. 0x77). This gives temperature readings with a resolution of 0.25°C.

Noise Filtering

For temperature sensors operating in noisy environments, previous practice was to place a capacitor across the D+ and D– pins to decrease the effects of noise. However, large capacitances affect the accuracy of the temperature measurement, leading to a recommended maximum capacitor value of 1000 pF. A capacitor of this value reduces the noise, but does not eliminate it, making use of the sensor difficult in a very noisy environment.

The ADT7468 has a major advantage over other devices for eliminating the effects of noise on the external sensor. Using the series resistance cancellation feature, a filter can be constructed between the external temperature sensor and the part. The effect of any filter resistance seen in series with the remote sensor is automatically canceled from the temperature result.

The construction of a filter allows the ADT7468 and the remote temperature sensor to operate in noisy environments. Figure 24 shows a low-pass R-C-R filter, with the following values:

$$R = 100\ \Omega, C = 1\ \text{nF}.$$

This filtering reduces both common-mode noise and differential noise.

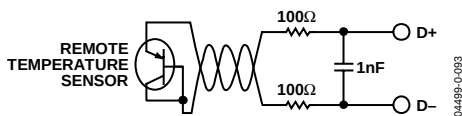


Figure 24. Filter Between Remote Sensor and ADT7468

Series Resistance Cancellation

Parasitic resistance to the ADT7468 D+ and D– inputs (seen in series with the remote diode) is caused by a variety of factors, including PCB track resistance and track length. This series resistance appears as a temperature offset in the remote sensor's temperature measurement. This error typically causes a 0.5°C offset per Ω of parasitic resistance in series with the remote diode.

The ADT7468 automatically cancels out the effect of this series resistance on the temperature reading, giving a more accurate result, without the need for user characterization of this resistance. The ADT7468 is designed to automatically cancel,

typically, up to 3 k Ω of resistance. By using an advanced temperature measurement method, this is transparent to the user. This feature allows resistances to be added to the sensor path to produce a filter, allowing the part to be used in noisy environments. See the Noise Filtering section for details.

Factors Affecting Diode Accuracy

Remote Sensing Diode

The ADT7468 is designed to work with either substrate transistors built into processors or with discrete transistors. Substrate transistors are generally PNP types with the collector connected to the substrate. Discrete types can be either PNP or NPN transistors connected as a diode (base-shortened to the collector). If an NPN transistor is used, the collector and base are connected to D+, and the emitter is connected to D–. If a PNP transistor is used, the collector and base are connected to D– and the emitter is connected to D+.

To reduce the error due to variations in both substrate and discrete transistors, a number of factors should be taken into consideration:

- The ideality factor, n_f , of the transistor is a measure of the deviation of the thermal diode from ideal behavior. The ADT7468 is trimmed for an n_f value of 1.008. Use the following equation to calculate the error introduced at a temperature T (°C), when using a transistor whose n_f does not equal 1.008. See the processor data sheet for the n_f values.

$$\Delta T = (n_f - 1.008) \times (273.15\ \text{K} + T)$$

To correct for this error, the user can write the ΔT value to the offset register. The ADT7468 then automatically adds or subtracts it from the temperature measurement.

- Some CPU manufacturers specify the high and low current levels of the substrate transistors. The high current level of the ADT7468, I_{HIGH} , is 96 μA and the low level current, I_{LOW} , is 6 μA . If the ADT7468 current levels do not match the current levels specified by the CPU manufacturer, it might be necessary to remove an offset. The CPU's data sheet should advise whether this offset needs to be removed and how to calculate it. This offset can be programmed to the offset register. It is important to note that, if more than one offset must be considered, the algebraic sum of these offsets must be programmed to the offset register.

If a discrete transistor is used with the ADT7468, the best accuracy is obtained by choosing devices according to the following criteria:

ADT7468

- Base-emitter voltage greater than 0.25 V at 6 μ A, with the highest operating temperature.
- Base-emitter voltage less than 0.95 V at 100 μ A, with the lowest operating temperature.
- Base resistance less than 100 Ω .
- Small variation in h_{FE} (from 50 to 150), which indicates tight control of V_{BE} characteristics.

Transistors, such as 2N3904, 2N3906, or equivalents in SOT-23 packages, are suitable devices to use.

Table 7. Temperature Data Format

Temperature	Digital Output (10-Bit) ¹
-128°C	1000 0000 00
-125°C	1000 0011 00
-100°C	1001 1100 00
-75°C	1011 0101 00
-50°C	1100 1110 00
-25°C	1110 0111 00
-10°C	1111 0110 00
0°C	0000 0000 00
10.25°C	0000 1010 01
25.5°C	0001 1001 10
50.75°C	0011 0010 11
75°C	0100 1011 00
100°C	0110 0100 00
125°C	0111 1101 00
127°C	0111 1111 00

¹ Bold numbers denote 2 LSB of measurement in Extended Resolution Register 2 (Reg. 0x77) with 0.25°C resolution.

Table 8. Extended Range, Temperature Data Format

Temperature	Digital Output (10-Bit) ¹
-64°C	0000 0000 00
-1°C	0011 1111 00
0°C	0100 0000 00
1°C	0100 0001 00
10°C	0100 1010 00
25°C	0101 1001 00
50°C	0111 0010 00
75°C	1000 1001 00
100°C	1010 0100 00
125°C	1011 1101 00
191°C	1111 1111 00

¹ Bold numbers denote 2 LSB of measurement in Extended Resolution Register 2 (Reg. 0x77) with 0.25°C resolution.

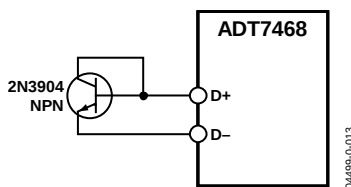


Figure 25. Measuring Temperature Using an NPN Transistor

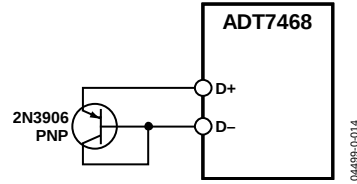


Figure 26. Measuring Temperature Using a PNP Transistor

Nulling Temperature Errors

As CPUs are developed that run faster, it is getting more difficult to avoid high frequency clocks when routing the D+/D- traces around a system board. Even when recommended layout guidelines are followed, some temperature errors may still be attributed to noise coupled onto the D+/D- lines. Constant high frequency noise usually attenuates or increases temperature measurements by a linear, constant value.

The ADT7468 has temperature offset registers at Addresses 0x70 and 0x72 for the Remote 1 and Remote 2 temperature channels. By doing a one-time calibration of the system, the user can determine the offset caused by system board noise and null it using the offset registers. The offset registers automatically add an Offset 64/twos complement 8-bit reading to every temperature measurement. The LSBs add 0.5°C offset to the temperature reading; therefore, the 8-bit register effectively allows temperature offsets up to $\pm 64^\circ\text{C}$ with a resolution of 0.5°C. This ensures that the readings in the temperature measurement registers are as accurate as possible.

Temperature Offset Registers

Reg. 0x70, Remote 1 temperature offset = 0x00 (0°C default)

Reg. 0x71, Local temperature offset = 0x00 (0°C default)

Reg. 0x72, Remote 2 temperature offset = 0x00 (0°C default)

ADT7463/ADT7468 Backwards Compatible Mode

By setting Bit 1 of Configuration Register 5 (0x7C), all temperature measurements are stored in the zone temperature value registers (0x25, 0x26, and 0x27) in twos complement in the range of -64°C to $+127^\circ\text{C}$ (the ADT7468 makes calculations based on the Offset 64 extended range and clamps the results, if necessary.) The temperature limits must be reprogrammed in twos complement. If a twos complement temperature below -63°C is entered, the temperature is clamped to -63°C . In this mode, the diode fault condition remains at $-128^\circ\text{C} = 1000\ 0000$, while in the extended temperature range (-64°C to $+191^\circ\text{C}$) the fault condition is represented by $-64^\circ\text{C} = 0000\ 0000$.

Temperature Measurement Registers

Reg. 0x25, Remote 1 temperature

Reg. 0x26, Local temperature

Reg. 0x27, Remote 2 temperature

Reg. 0x77, Extended Resolution 2 = 0x00 default

<7:6> TDM2, Remote 2 temperature LSBs

<5:4> LTMP, local temperature LSBs

<3:2> TDM1, Remote 1 temperature LSBs

Temperature Measurement Limit Registers

Each temperature measurement channel is associated with high and low limit registers. Exceeding the programmed high or low limit causes the appropriate status bit to be set. Exceeding either limit can also generate SMBALERT interrupts.

Reg. 0x4E, Remote 1 temperature low limit = 0x01 default

Reg. 0x4F, Remote 1 temperature high limit = 0x7F default

Reg. 0x50, local temperature low limit = 0x01 default

Reg. 0x51, local temperature high limit = 0x7F default

Reg. 0x52, Remote 2 temperature low limit = 0x01 default

Reg. 0x53, Remote 2 temperature high limit = 0x7F default

Reading Temperature from the ADT7468

It is important to note that temperature can be read from the ADT7468 as an 8-bit value (with 1°C resolution) or as a 10-bit value (with 0.25°C resolution). If only 1°C resolution is required, the temperature readings can be read back at any time and in no particular order.

If the 10-bit measurement is required, this involves a 2-register read for each measurement. The extended resolution register (Reg. 0x77) should be read first. This causes all temperature reading registers to be frozen until all temperature reading registers have been read from. This prevents an MSB reading from being updated while its two LSBs are being read and vice versa.

ADDITIONAL ADC FUNCTIONS FOR TEMPERATURE MEASUREMENT

A number of other functions are available on the ADT7468 to offer the system designer increased flexibility.

Turn-Off Averaging

For each temperature measurement read from a value register, 16 readings are made internally and the results are averaged before being placed into the value register. Sometimes it is necessary to take a very fast measurement. Setting Bit 4 of Configuration Register 2 (Reg. 0x73) turns averaging off.

Table 9. Conversion Time with Averaging Disabled

Channel	Measurement Time
Voltage Channels	0.7 ms
Remote Temperature 1	7 ms
Remote Temperature 2	7 ms
Local Temperature	1.3 ms

Table 10. Conversion Time with Averaging Enabled

Channel	Measurement Time
Voltage Channels	11 ms
Remote Temperature	39 ms
Local Temperature	12 ms

Single-Channel ADC Conversions

Setting Bit 6 of Configuration Register 2 (Reg. 0x73) places the ADT7468 into single-channel ADC conversion mode. In this mode, the ADT7468 can be made to read a single temperature channel only. The appropriate ADC channel is selected by writing to Bits <7:5> of the TACH1 minimum high byte register (0x55).

Table 11. Channel Selection

Bits <7:5> Reg. 0x55	Channel Selected
101	Remote 1 temperature
110	Local temperature
111	Remote 2 temperature

Configuration Register 2 (Reg. 0x73)

<4> = 1, averaging off.

<6> = 1, single-channel convert mode.

TACH1 Minimum High Byte (Reg. 0x55)

<7:5> selects ADC channel for single-channel convert mode.

Overtemperature Events

Overtemperature events on any of the temperature channels can be detected and dealt with automatically in automatic fan speed control mode. Register 0x6A to Register 0x6C are the THERM temperature limits. When a temperature exceeds its THERM temperature limit, all PWM outputs run at the maximum PWM duty cycle (Reg. 0x38, Reg. 0x39, and Reg. 0x3A). This effectively runs the fans at the fastest allowed speed. The fans stay running at this speed until the temperature drops below THERM minus hysteresis. (This can be disabled by setting the boost bit in Configuration Register 3, Bit 2, Reg. 0x78.) The hysteresis value for that THERM temperature limit is the value programmed into Reg. 0x6D and Reg. 0x6E (hysteresis registers). The default hysteresis value is 4°C.

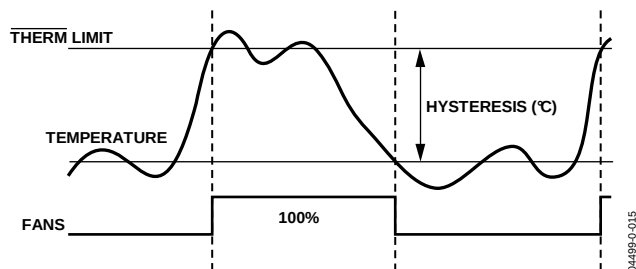


Figure 27. THERM Temperature Limit Operation

LIMITS, STATUS REGISTERS, AND INTERRUPTS

LIMIT VALUES

Each measurement channel on the ADT7468 is associated with high and low limits. These can form the basis of system status monitoring; a status bit can be set for any out-of-limit condition and detected by polling the device. Alternatively, SMBALERT interrupts can be generated to flag a processor or micro-controller of out-of-limit conditions.

8-Bit Limits

The following is a list of 8-bit limits on the ADT7468.

Voltage Limit Registers

Reg. 0x44, 2.5 V low limit = 0x00 default, Register 0x44 2.5 V low limit = 0x00 default

Reg. 0x45, 2.5 V high limit = 0xFF default

Reg. 0x46, V_{CCP} low limit = 0x00 default

Reg. 0x47, V_{CCP} high limit = 0xFF default

Reg. 0x48, V_{CC} low limit = 0x00 default

Reg. 0x49, V_{CC} high limit = 0xFF default

Reg. 0x4A, 5 V low limit = 0x00 default

Reg. 0x4B, 5 V high limit = 0xFF default

Reg. 0x4C, 12 V low limit = 0x00 default

Reg. 0x4D, 12 V high limit = 0xFF default

Reg. 0x46, V_{CCP} low limit = 0x00 default

Reg. 0x47, V_{CCP} high limit = 0xFF default

Reg. 0x48, V_{CC} low limit = 0x00 default

Reg. 0x49, V_{CC} high limit = 0xFF default

Temperature Limit Registers

Reg. 0x4E, Remote 1 temperature low limit = 0x01 default

Reg. 0x4F, Remote 1 temperature high limit = 0x7F default

Reg. 0x6A, Remote 1 THERM limit = 0x64 default

Reg. 0x50, local temperature low limit = 0x01 default

Reg. 0x51, local temperature high limit = 0x7F default

Reg. 0x6B, local THERM limit = 0x64 default

Reg. 0x52, Remote 2 temperature low limit = 0x01 default

Reg. 0x53, Remote 2 Temperature high limit = 0x7F default

Reg. 0x6C, Remote 2 THERM limit = 0x64 default

THERM Limit Register

Reg. 0x7A, THERM limit = 0x00 default

16-Bit Limits

The fan TACH measurements are 16-bit results. The fan TACH limits are also 16 bits, consisting of a high byte and low byte. Because slow or stalled fans are normally the only conditions of interest, only high limits exist for fan TACHs. Because the fan TACH period is measured, exceeding the limit indicates a slow or stalled fan.

Fan Limit Registers

Reg. 0x54, TACH1 minimum low byte = 0x00 default

Reg. 0x55, TACH1 minimum high byte = 0x00 default

Reg. 0x56, TACH2 minimum low byte = 0x00 default

Reg. 0x57, TACH2 minimum high byte = 0x00 default

Reg. 0x58, TACH3 minimum low byte = 0x00 default

Reg. 0x59, TACH3 minimum high byte = 0x00 default

Reg. 0x5A, TACH4 minimum low byte = 0x00 default

Reg. 0x5B, TACH4 minimum high byte = 0x00 default

Out-of-Limit Comparisons

Once all limits have been programmed, the ADT7468 can be enabled for monitoring. The ADT7468 measures all voltage and temperature measurements in a round-robin format and sets the appropriate status bit for out-of-limit conditions. TACH measurements are not part of this round-robin cycle. Comparisons are done differently depending on whether the measured value is compared to a high or low limit.

High limit: > comparison performed

Low limit: $\leq$ comparison performed

Voltage and temperature channels use a window comparator for error detecting and therefore have high and low limits. Fan speed measurements use only a low limit, which is needed only in manual fan control mode.

Analog Monitoring Cycle Time

The analog monitoring cycle begins when a 1 is written to the start bit (Bit 0) of Configuration Register 1 (Reg. 0x40). By default, the ADT7463 powers up with this bit set. The ADC measures each analog input and as each measurement is completed, the result is automatically stored in the appropriate value register. This round-robin monitoring cycle continues unless disabled by writing a 0 to Bit 0 of Configuration Register 1.

Since the ADC normally runs freely in this manner, the time taken to monitor all the analog inputs is usually not of interest, because the most recently measured value of any input can be read at any time.

For applications where the monitoring cycle time is important, it can be calculated easily.

The measured channels are

- Four dedicated supply voltage inputs
- Supply voltage (V_{CC} pin)
- Local temperature
- Two remote temperatures

As mentioned previously, the ADC performs round-robin conversions and takes 11 ms for each voltage measurement, 12 ms for a local temperature reading, and 39 ms for each remote temperature reading. The total monitoring cycle time for averaged voltage and temperature monitoring is, therefore, nominally

$$(5 \times 11) + 12 + (2 \times 39) = 145 \text{ ms}$$

Fan TACH measurements are made in parallel and are not synchronized with the analog measurements in any way.

STATUS REGISTERS

The results of limit comparisons are stored in Status Registers 1 and 2. The status register bit for each channel reflects the status of the last measurement and limit comparison on that channel. If a measurement is within limits, the corresponding status register bit is cleared to 0. If the measurement is out-of-limit, the corresponding status register bit is set to 1.

The state of the various measurement channels can be polled by reading the status registers over the serial bus. In Bit 7 (OOL) of Status Register 1 (Reg. 0x41), 1 means that an out-of-limit event has been flagged in Status Register 2. This means that the user also needs to read Status Register 2. Alternatively, Pin 10 or Pin 14 can be configured as an SMBALERT output. This hard interrupt automatically notifies the system supervisor of an out-of-limit condition. Reading the status registers clears the appropriate status bit as long as the error condition that caused the interrupt has cleared. Status register bits are sticky. The status bits are referred to as sticky, because they remain set until read by software. Whenever a status bit is set, indicating an out-of-limit condition, it remains set even if the event that caused it has ceased (until read). The only way to clear the status bit is to read the status register after the event has ceased. Interrupt status mask registers (0x74, and 0x75) allow individual interrupt sources to be masked from causing an SMBALERT. However, if one of these masked interrupt sources goes out-of-limit, its associated status bit is set in the interrupt status registers.

Status Register 1 (Reg. 0x41)

Bit 7 (OOL) = 1, denotes a bit in Status Register 2 is set and Status Register 2 should be read.

Bit 6 (R2T) = 1, Remote 2 temperature high or low limit has been exceeded.

Bit 5 (LT) = 1, local temperature high or low limit has been exceeded.

Bit 4 (R1T) = 1, Remote 1 temperature high or low limit has been exceeded.

Bit 3 (5 V) = 1, 5 V high or low limit has been exceeded.

Bit 2 (V_{CC}) = 1, V_{CC} high or low limit has been exceeded.

Bit 1 (V_{CCP}) = 1, V_{CCP} high or low limit has been exceeded.

Bit 0 (2.5 V) = 1, 2.5 V high or low limit has been exceeded.

Status Register 2 (Reg. 0x42)

Bit 7 (D2) = 1, indicates an open or short on D2+/D2- inputs.

Bit 6 (D1) = 1, indicates an open or short on D1+/D1- inputs.

Bit 5 (F4P) = 1, indicates Fan 4 has dropped below minimum speed. Alternatively, indicates that the THERM limit has been exceeded, if the THERM function is used.

Bit 4 (FAN3) = 1, indicates Fan 3 has dropped below minimum speed.

Bit 3 (FAN2) = 1, indicates Fan 2 has dropped below minimum speed.

Bit 2 (FAN1) = 1, indicates Fan 1 has dropped below minimum speed.

Bit 1 (OVT) = 1, indicates a THERM overtemperature limit has been exceeded.

Bit 0 (12V/VC) = 1, indicates a 12 V high or low limit has been exceeded. If the VID code change function is used, this bit indicates a change in VID code on the VID0 to VID5 inputs.

INTERRUPTS

SMBALERT Interrupt Behavior

The ADT7468 can be polled for status, or an SMBALERT interrupt can be generated for out-of-limit conditions. It is important to note how the SMBALERT output and status bits behave when writing interrupt handler software.

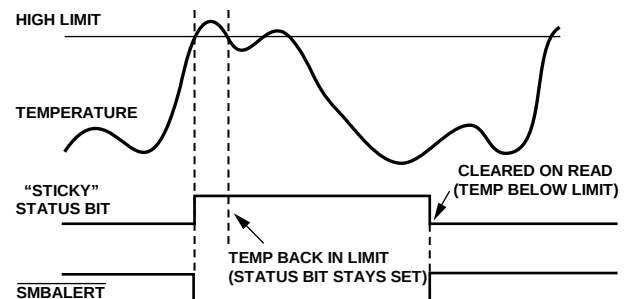


Figure 28. SMBALERT and Status Bit Behavior

Figure 28 shows how the SMBALERT output and sticky status bits behave. Once a limit is exceeded, the corresponding status bit is set to 1. The status bit remains set until the error condition subsides and the status register is read. This ensures that an out-of-limit event cannot be missed, if software is polling the device periodically. Note that the SMBALERT output remains low for the entire duration that a reading is out-of-limit and until the status register has been read. This has implications on how software handles the interrupt.

Handling SMBALERT Interrupts

To prevent the system from being tied up servicing interrupts, it is recommended to handle the SMBALERT interrupt as follows:

1. Detect the SMBALERT assertion.
2. Enter the interrupt handler.
3. Read the status registers to identify the interrupt source.
4. Mask the interrupt source by setting the appropriate mask bit in the interrupt mask registers (Reg. 0x74 and Reg. 0x75).
5. Take the appropriate action for a given interrupt source.
6. Exit the interrupt handler.
7. Periodically poll the status registers. If the interrupt status bit has cleared, reset the corresponding interrupt mask bit to 0. This causes the SMBALERT output and status bits to behave as shown in Figure 29.

Masking Interrupt Sources

Interrupt Mask Registers 1 and 2 are located at Addresses 0x74 and 0x75. These allow individual interrupt sources to be masked out to prevent SMBALERT interrupts. Note that masking an interrupt source prevents only the SMBALERT output from being asserted; the appropriate status bit is set normally.

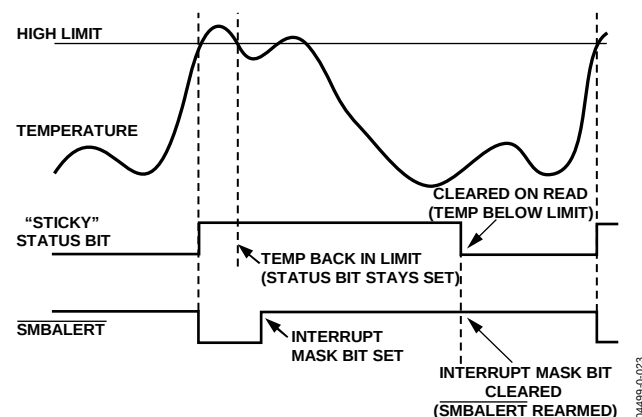


Figure 29. How Masking the Interrupt Source Affects SMBALERT Output

Interrupt Mask Register 1 (Reg. 0x74)

Bit 7 (OOL) = 1, masks SMBALERT for any alert condition flagged in Status Register 2.

Bit 6 (R2T) = 1, masks SMBALERT for Remote 2 temperature.

Bit 5 (LT) = 1, masks SMBALERT for local temperature.

Bit 4 (R1T) = 1, masks SMBALERT for Remote 1 temperature.

Bit 3 (5 V) = 1, masks SMBALERT for 5 V channel.

Bit 2 (V_{CC}) = 1, masks SMBALERT for V_{CC} channel.

Bit 0 (V_{CCP}) = 1, masks SMBALERT for V_{CCP} channel.

Interrupt Mask Register 2 (Reg. 0x75)

Bit 7 (D2) = 1, masks SMBALERT for Diode 2 errors.

Bit 6 (D1) = 1, masks SMBALERT for Diode 1 errors.

Bit 5 (FAN4) = 1, masks SMBALERT for Fan 4 failure.

If the TACH4 pin is being used as the THERM input, this bit masks SMBALERT for a THERM event.

Bit 4 (FAN3) = 1, masks SMBALERT for Fan 3.

Bit 3 (FAN2) = 1, masks SMBALERT for Fan 2.

Bit 2 (FAN1) = 1, masks SMBALERT for Fan 1.

Bit 1 (OVT) = 1, masks SMBALERT for overtemperature (exceeding THERM temperature limits).

Bit 0 (12V/VC) = 1, masks SMBALERT for 12 V channel or for a VID code change, depending on the function used.

Enabling the SMBALERT Interrupt Output

The SMBALERT interrupt function is disabled by default.

Pin 10 or Pin 14 can be reconfigured as an SMBALERT output to signal out-of-limit conditions.

Table 12. Configuring Pin 10 as SMBALERT Output

Register	Bit Setting
Configuration Register 3 (Reg. 0x78)	<0> Pin 10 = alert <1> Pin 10 = PWM2

Assigning THERM Functionality to a Pin

Pin 14 on the ADT7468 has four possible functions: SMBALERT, THERM, GPIO, and TACH4. The user chooses the required functionality by setting Bit 0 and Bit 1 of Configuration Register 4 at Address 0x7D.

If $\overline{\text{THERM}}$ is enabled (Bit 1, Configuration Register 3 at Address 0x78):

- Pin 20 becomes $\overline{\text{THERM}}$.
- If Pin 14 is configured as $\overline{\text{THERM}}$ (Bit 0 and Bit 1 of Configuration Register 4 at Address 0x7D), then $\overline{\text{THERM}}$ is enabled on this pin.

If $\overline{\text{THERM}}$ is not enabled:

- Pin 20 becomes a 5 V measurement input.
- If Pin 14 is configured as $\overline{\text{THERM}}$, then $\overline{\text{THERM}}$ is disabled on this Pin.

Table 13. Configuring Pin 14

Bit 0	Bit 1	Function
0	0	TACH4
0	1	$\overline{\text{THERM}}$
1	0	SMBALERT
1	1	GPIO

$\overline{\text{THERM}}$ as an Input

When $\overline{\text{THERM}}$ is configured as an input, the user can time assertions on the $\overline{\text{THERM}}$ pin. This can be useful for connecting to the PROCHOT output of a CPU to gauge system performance.

The user can also set up the ADT7468 to run the fans at 100% whenever the $\overline{\text{THERM}}$ pin is driven low externally by setting the boost bit (Bit 2) in Configuration Register 3 (Address 0x78) to 1. Note that to set this up, the fan must be already running, for example, in manual mode when the current duty cycle is above 0x00, or in automatic mode when the temperature is above T_{MIN} . If the temperature is below T_{MIN} or if the duty cycle in manual mode is set to 0x00, then pulling the $\overline{\text{THERM}}$ low externally has no effect. See Figure 30 for more information.

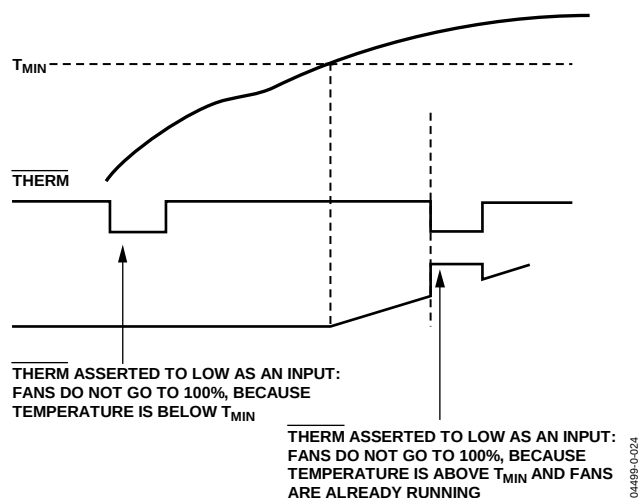


Figure 30. Asserting $\overline{\text{THERM}}$ Low as an Input in Automatic Fan Speed Control Mode

$\overline{\text{THERM}}$ Timer

The ADT7468 has an internal timer to measure $\overline{\text{THERM}}$ assertion time. The $\overline{\text{THERM}}$ input can be connected to the PROCHOT output of a Pentium 4 CPU to measure system performance or be connected to the output of a trip point temperature sensor, to name a couple of functions of the timer.

The timer is started on the assertion of the ADT7468's $\overline{\text{THERM}}$ input and stopped when $\overline{\text{THERM}}$ is unasserted. The timer counts $\overline{\text{THERM}}$ times cumulatively, that is, the timer resumes counting on the next $\overline{\text{THERM}}$ assertion. The $\overline{\text{THERM}}$ timer continues to accumulate $\overline{\text{THERM}}$ assertion times until the timer is read (it is cleared on read) or until it reaches full scale. If the counter reaches full scale, it stops at that reading until cleared.

The 8-bit $\overline{\text{THERM}}$ timer register (Reg. 0x79) is designed such that Bit 0 is set to 1 on the first $\overline{\text{THERM}}$ assertion. Once the cumulative $\overline{\text{THERM}}$ assertion time has exceeded 45.52 ms, Bit 1 of the $\overline{\text{THERM}}$ timer is set and Bit 0 now becomes the LSB of the timer with a resolution of 22.76 ms (see Figure 31).

When using the $\overline{\text{THERM}}$ timer, be aware of the following after a $\overline{\text{THERM}}$ timer read (Reg. 0x79):

1. The contents of the timer are cleared on read.
2. The F4P bit (Bit 5) of Status Register 2 needs to be cleared (assuming that the $\overline{\text{THERM}}$ timer limit has been exceeded).

If the $\overline{\text{THERM}}$ timer is read during a $\overline{\text{THERM}}$ assertion, then the following happens:

1. The contents of the timer are cleared.
2. Bit 0 of the $\overline{\text{THERM}}$ timer is set to 1 (because a $\overline{\text{THERM}}$ assertion is occurring).
3. The $\overline{\text{THERM}}$ timer increments from 0.
4. If the $\overline{\text{THERM}}$ timer limit (Reg. 0x7A) = 0x00, then the F4P bit is set.

Generating SMBALERT Interrupts from THERM Timer Events

The ADT7468 can generate SMBALERTs when a programmable THERM timer limit has been exceeded. This allows the system designer to ignore brief, infrequent THERM assertions while capturing longer THERM timer events. Register 0x7A is the THERM timer limit register. This 8-bit register allows a limit from 0 secs (first THERM assertion) to 5.825 secs to be set before an SMBALERT is generated. The THERM timer value is compared with the contents of the THERM timer limit register. If the THERM timer value exceeds the THERM timer limit value, then the F4P bit (Bit 5) of Status Register 2 is set and an SMBALERT is generated. Note that the F4P bit (Bit 5) of Mask Register 2 (Reg. 0x75) masks out SMBALERTs if this bit is set to 1, although the F4P bit of Interrupt Status Register 2 stays set if the THERM timer limit is exceeded.

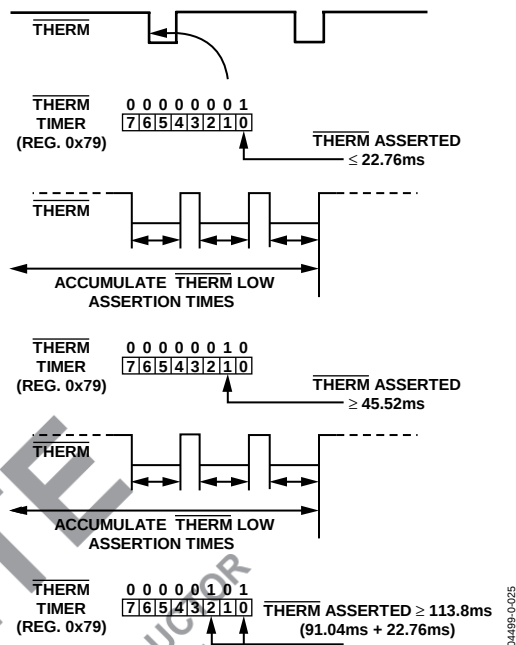


Figure 31. Understanding the THERM Timer

Figure 32 is a functional block diagram of the THERM timer, limit, and associated circuitry. Writing a value of 0x00 to the THERM timer limit register (Reg. 0x7A) causes SMBALERT to be generated on the first THERM assertion. A THERM timer limit value of 0x01 generates an SMBALERT, once cumulative THERM assertions exceed 45.52 ms.

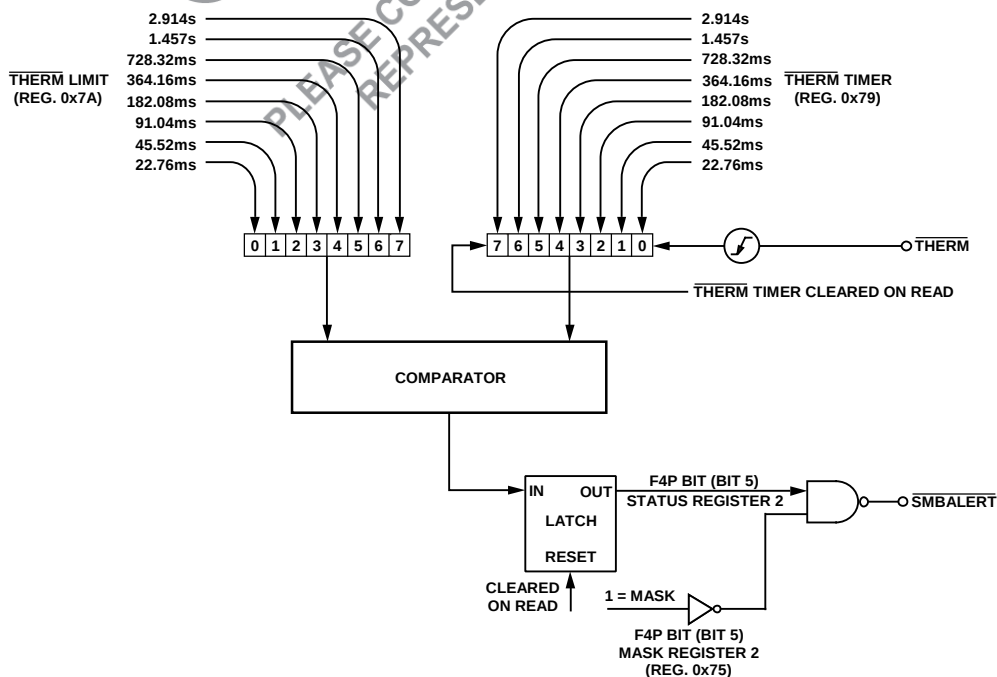


Figure 32. Functional Block Diagram of THERM Monitoring Circuitry

Configuring the Relevant THERM Behavior

1. Configure the desired pin as the THERM timer input.

Setting Bit 1 (THERM timer enable) of Configuration Register 3 (Reg. 0x78) enables the THERM timer monitoring functionality. This is disabled on Pins 14 and 20 by default.

Setting Bits 0 and 1 (PIN14FUNC) of Configuration Register 4 (Reg. 0x7D) enables THERM timer output functionality on Pin 20 (Bit 1 of Configuration Register 3, THERM, must also be set). Pin 14 can also be used as TACH4.

2. Select the desired fan behavior for THERM timer events.

Assuming that the fans are running, setting Bit 2 (BOOST bit) of Configuration Register 3 (Reg. 0x78) causes all fans to run at 100% duty cycle whenever THERM is asserted. This allows fail-safe system cooling. If this bit is 0, the fans run at their current settings and are not affected by THERM events. If the fans are not already running when THERM is asserted, the fans do not run to full speed.

3. Select whether THERM timer events should generate SMBALERT interrupts.

Bit 5 (F4P) of Mask Register 2 (Reg. 0x75), when set, masks out SMBALERTs when the THERM timer limit value is exceeded. This bit should be cleared, if SMBALERTs based on THERM events are required.

4. Select a suitable THERM limit value.

This value determines whether an SMBALERT is generated on the first THERM assertion, or only if a cumulative THERM assertion time limit is exceeded. A value of 0x00 causes an SMBALERT to be generated on the first THERM assertion.

5. Select a THERM monitoring time.

This value specifies how often OS or BIOS level software checks the THERM timer. For example, BIOS could read the THERM timer once an hour to determine the cumulative THERM assertion time. If, for example, the total THERM assertion time is <22.76 ms in Hour 1, >182.08 ms in Hour 2, and >5.825 secs in Hour 3, this can indicate that system performance is degrading significantly, because THERM is asserting more frequently on an hourly basis.

Alternatively, OS or BIOS level software can timestamp

when the system is powered on. If an SMBALERT is generated due to the THERM timer limit being exceeded, another timestamp can be taken. The difference in time can be calculated for a fixed THERM timer limit time. For example, if it takes one week for a THERM timer limit of 2.914 secs to be exceeded and the next time it takes only one hour, then this is an indication of a serious degradation in system performance.

Configuring the THERM Pin as an Output

In addition to monitoring THERM as an input, the ADT7468 can optionally drive THERM low as an output. In cases where PROCHOT is bidirectional, THERM can be used to throttle the processor by asserting PROCHOT. The user can preprogram system-critical thermal limits. If the temperature exceeds a thermal limit by 0.25°C, THERM asserts low. If the temperature is still above the thermal limit on the next monitoring cycle, THERM stays low. THERM remains asserted low until the temperature is equal to or below the thermal limit. Because the temperature for that channel is measured only once for every monitoring cycle, after THERM asserts it is guaranteed to remain low for at least one monitoring cycle.

The THERM pin can be configured to assert low, if the Remote 1, local, or Remote 2 THERM temperature limits are exceeded by 0.25°C. The THERM temperature limit registers are at Registers 0x6A, 0x6B, and 0x6C, respectively. Setting Bit 3 of Registers 0x5F, 0x60, and 0x61 enables the THERM output feature for the Remote 1, local, and Remote 2 temperature channels, respectively. Figure 33 shows how the THERM pin asserts low as an output in the event of a critical overtemperature.

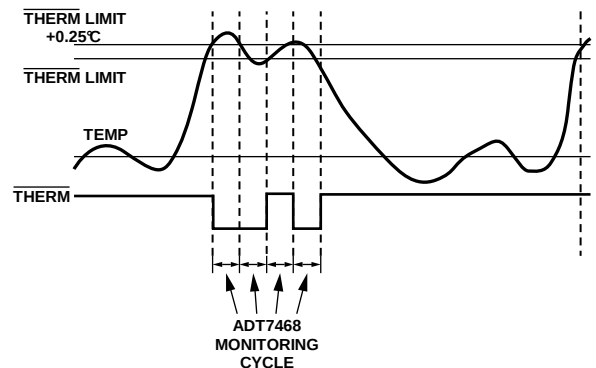


Figure 33. Asserting THERM as an Output, Based on Tripping THERM Limits

An alternative method of disabling THERM is to program the THERM temperature limit to -64°C or less in Offset 64 mode, or -128°C or less in twos complement mode; that is, for THERM temperature limit values less than -63°C or -128°C, respectively, THERM is disabled.

ACTIVE COOLING

DRIVING THE FAN USING PWM CONTROL

The ADT7468 uses pulse-width modulation (PWM) to control fan speed. This relies on varying the duty cycle (or on/off ratio) of a square wave applied to the fan to vary the fan speed. The external circuitry required to drive a fan using PWM control is extremely simple. For 4-wire fans, the PWM drive might need only a pull-up resistor. In many cases, the 4-wire fan PWM input has a built-in pull-up resistor.

The ADT7468 PWM frequency can be set to a selection of low frequencies or a single high PWM frequency. The low frequency options are usually used for 2-wire and 3-wire fans, while the high frequency option is usually used with 4-wire fans.

For 2-wire or 3-wire fans, a single N-channel MOSFET is the only drive device required. The specifications of the MOSFET depend on the maximum current required by the fan being driven and the input capacitance of the FET. Because a 10 k Ω (or greater) resistor must be used as a PWM pull-up, an FET with large input capacitance can cause the PWM output to become distorted and adversely affect the fan control range. This is a requirement only when using high frequency PWM mode.

Typical notebook fans draw a nominal 170 mA, and so SOT devices can be used where board space is a concern. In desktops, fans can typically draw 250 mA to 300 mA each. If you drive several fans in parallel from a single PWM output or drive larger server fans, the MOSFET must handle the higher current requirements. The only other stipulation is that the MOSFET should have a gate voltage drive, $V_{GS} > 3.3$ V, for direct interfacing to the PWM_OUT pin. V_{GS} can be greater than 3.3 V as long as the pull-up on the gate is tied to 5 V. The MOSFET should also have a low on resistance to ensure that there is not significant voltage drop across the FET, which would reduce the voltage applied across the fan and, therefore, the maximum operating speed of the fan.

Figure 34 shows how to drive a 3-wire fan using PWM control.

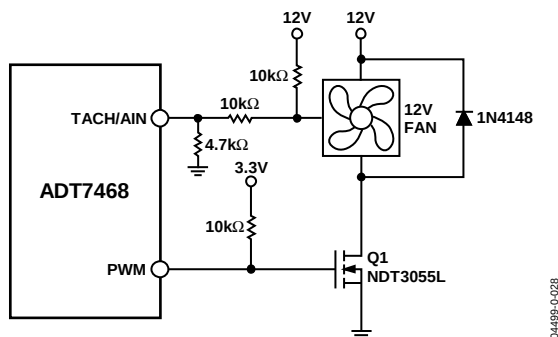


Figure 34. Driving a 3-Wire Fan Using an N-Channel MOSFET

Figure 34 uses a 10 k Ω pull-up resistor for the TACH signal. This assumes that the TACH signal is an open-collector from the fan. In all cases, the TACH signal from the fan must be kept below 5 V maximum to prevent damaging the ADT7468. If in doubt as to whether the fan has an open-collector or totem-pole TACH output, use one of the input signal conditioning circuits shown in the Fan Speed Measurement section.

Figure 35 shows a fan drive circuit using an NPN transistor, such as a general-purpose MMBT2222. While these devices are inexpensive, they tend to have much lower current handling capabilities and higher on resistance than MOSFETs. When choosing a transistor, care should be taken to ensure that it meets the fan's current requirements. Ensure that the base resistor is chosen such that the transistor is saturated when the fan is powered on.

Because 4-wire fans are powered continuously, the fan speed is not switched on or off as with previous PWM driven/powered fans. This enables it to perform better than 3-wire fans, especially for high frequency applications.

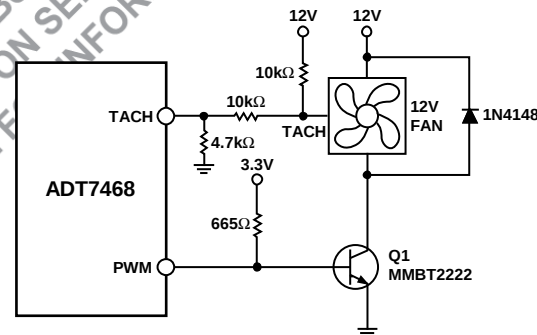


Figure 35. Driving a 3-Wire Fan Using an NPN Transistor

Figure 36 shows a typical drive circuit for 4-wire fans.

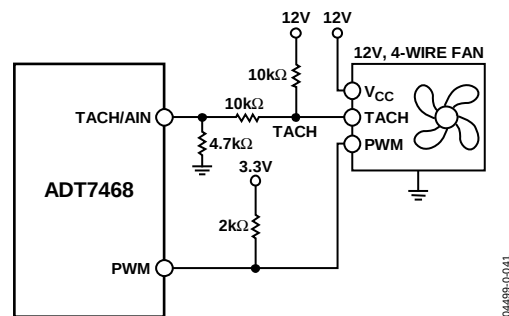


Figure 36. Driving a 4-Wire Fan

The ADT7468 has four TACH inputs available for fan speed measurement, but only three PWM drive outputs. If a fourth fan is being used in the system, it should be driven from the PWM3 output in parallel with the third fan. Figure 37 shows how to drive two fans in parallel using low cost NPN transistors. Figure 38 shows the equivalent circuit using a MOSFET.

Because the MOSFET can handle up to 3.5 A, a second fan can be connected directly in parallel with the first. Care should be taken in designing drive circuits with transistors and FETs to ensure that the PWM pins are not required to source current and that they sink less than the 5 mA maximum current specified in the data sheet.

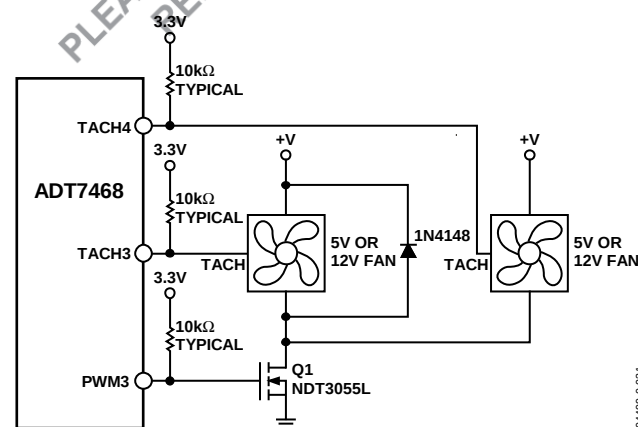
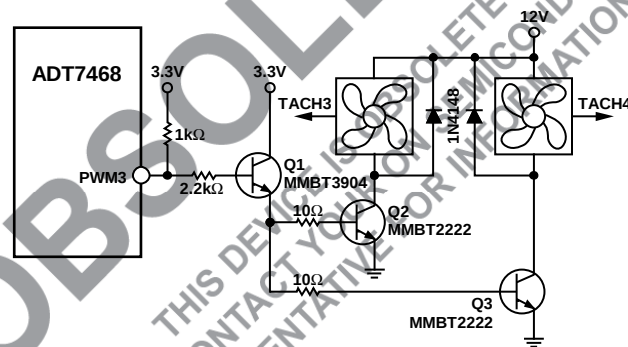
Driving up to Three Fans from PWM3

TACH measurements for fans are synchronized to particular PWM channels; for example, TACH1 is synchronized to PWM1. TACH3 and TACH4 are both synchronized to PWM3, and therefore PWM3 can drive two fans. Alternatively, PWM3 can be programmed to synchronize TACH2, TACH3, and TACH4 to the PWM3 output. This allows PWM3 to drive two or three fans. In this case, the drive circuitry is as shown in Figure 37 and Figure 38. The SYNC bit in Register 0x62 enables this function.

Synchronization is not required in high frequency mode when used with 4-wire fans.

<4> (SYNC) Enhance Acoustics Register 1 (Reg. 0x62)

SYNC = 1, synchronizes TACH2, TACH3, and TACH4 to PWM3.



If the fan output has a resistive pull-up to 12 V (or other voltage greater than 5 V) then the fan output can be clamped with a Zener diode, as shown in Figure 43. The Zener diode voltage should be chosen so that it is greater than V_{IH} of the TACH input but less than 5 V, allowing for the voltage tolerance of the Zener. A value of between 3 V and 5 V is suitable.

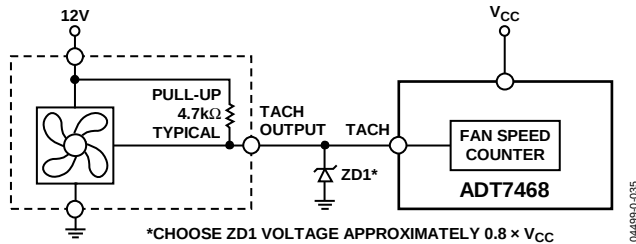


Figure 43. Fan with TACH Pull-Up Greater than 5 V, for example, 12 V), Clamped with Zener Diode

If the fan has a strong pull-up ($<1\text{ k}\Omega$) to 12 V or a totem-pole output, then a series resistor can be added to limit the Zener current, as shown in Figure 44.

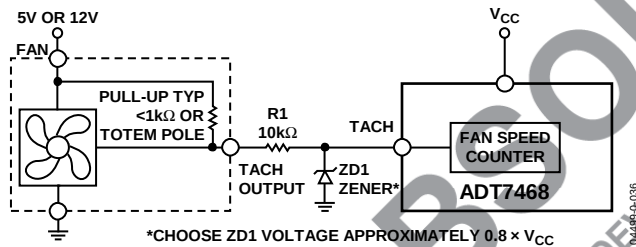


Figure 44. Fan with Strong TACH Pull-Up Greater than V_{CC} or Totem-Pole Output, Clamped with Zener and Resistor

Alternatively, a resistive attenuator can be used, as shown in Figure 45. R1 and R2 should be chosen such that

$$2\text{ V} < V_{PULL-UP} \times R2 / (R_{PULL-UP} + R1 + R2) < 5\text{ V}$$

The fan inputs have an input resistance of nominally 160 kΩ to ground, which should be taken into account when calculating resistor values.

With a pull-up voltage of 12 V and pull-up resistor less than 1 kΩ, suitable values for R1 and R2 would be 100 kΩ and 47 kΩ, respectively. This gives a high input voltage of 3.83 V.

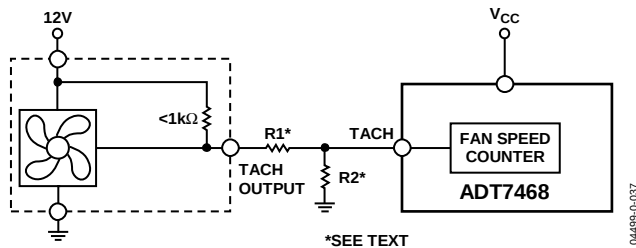


Figure 45. Fan with Strong TACH Pull-Up Greater than V_{CC} or Totem-Pole Output, Attenuated with R1/R2

FAN SPEED MEASUREMENT

The fan counter does not count the fan TACH output pulses directly, because the fan speed could be less than 1,000 RPM and it would take several seconds to accumulate a reasonably large and accurate count. Instead, the period of the fan revolution is measured by gating an on-chip 90 kHz oscillator into the input of a 16-bit counter for N periods of the fan TACH output (see Figure 46), so the accumulated count is actually proportional to the fan tachometer period and inversely proportional to the fan speed.

N , the number of pulses counted, is determined by the settings of Register 0x7B (TACH pulses per revolution register). This register contains two bits for each fan, allowing one, two (default), three, or four TACH pulses to be counted.

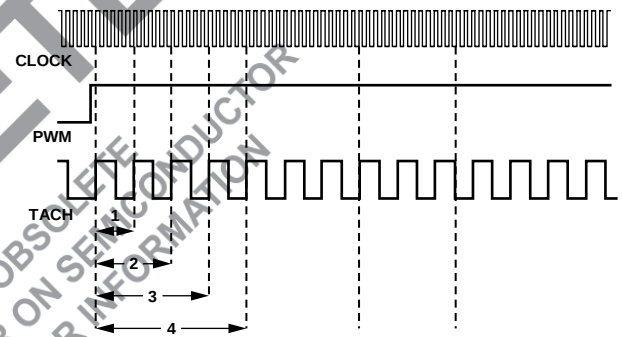


Figure 46. Fan Speed Measurement

Fan Speed Measurement Registers

The fan tachometer readings are 16-bit values consisting of a 2-byte read from the ADT7468.

Reg. 0x28, TACH1 low byte = 0x00 default

Reg. 0x29, TACH1 high byte = 0x00 default

Reg. 0x2A, TACH2 low byte = 0x00 default

Reg. 0x2B, TACH2 high byte = 0x00 default

Reg. 0x2C, TACH3 low byte = 0x00 default

Reg. 0x2D, TACH3 high byte = 0x00 default

Reg. 0x2E, TACH4 low byte = 0x00 default

Reg. 0x2F, TACH4 high byte = 0x00 default

Reading Fan Speed from the ADT7468

The measurement of fan speeds involves a 2-register read for each measurement. The low byte should be read first. This causes the high byte to be frozen until both high and low byte registers have been read, preventing erroneous TACH readings.

The fan tachometer reading registers report back the number of 11.11 μ s period clocks (90 kHz oscillator) gated to the fan speed counter, from the rising edge of the first fan TACH pulse to the rising edge of the third fan TACH pulse (assuming two pulses per revolution are being counted). Because the device is essentially measuring the fan TACH period, the higher the count value, the slower the fan is actually running. A 16-bit fan tachometer reading of 0xFFFF indicates either that the fan has stalled or is running very slowly (<100 RPM).

High limit: >comparison performed

Because the actual fan TACH period is being measured, falling below a fan TACH limit by 1 sets the appropriate status bit and can be used to generate an SMBALERT.

Fan TACH Limit Registers

The fan TACH limit registers are 16-bit values consisting of two bytes.

Reg. 0x54, TACH1 minimum low byte = 0xFF default

Reg. 0x55, TACH1 minimum high byte = 0xFF default

Reg. 0x56, TACH2 minimum low byte = 0xFF default

Reg. 0x57, TACH2 minimum high byte = 0xFF default

Reg. 0x58, TACH3 minimum low byte = 0xFF default

Reg. 0x59, TACH3 minimum high byte = 0xFF default

Reg. 0x5A, TACH4 minimum low byte = 0xFF default

Reg. 0x5B, TACH4 minimum high byte = 0xFF default

Fan Speed Measurement Rate

The fan TACH readings are normally updated once every second.

The fast bit (Bit 3) of Configuration Register 3 (Reg. 0x78), when set, updates the fan TACH readings every 250 ms.

If any of the fans are not being driven by a PWM channel but are powered directly from 5 V or 12 V, their associated dc bit in Configuration Register 3 should be set. This allows TACH readings to be taken on a continuous basis for fans connected directly to a dc source. For optimal results, the associated dc bit should always be set when using 4-wire fans.

Calculating Fan Speed

Assuming a fan with a two pulses per revolution (and two pulses per revolution being measured), fan speed is calculated by

$$\text{Fan Speed (RPM)} = (90,000 \times 60) / \text{Fan TACH Reading}$$

where the Fan TACH reading is the 16-bit fan tachometer reading.

Example:

TACH1 high byte (Reg. 0x29) = 0x17

TACH1 low byte (Reg. 0x28) = 0xFF

What is Fan 1 speed in RPM?

$$\text{Fan 1 TACH Reading} = 0x17FF = 6143 \text{ (decimal)}$$

$$\text{RPM} = (f \times 60) / \text{Fan 1 TACH Reading}$$

$$\text{RPM} = (90000 \times 60) / 6143$$

$$\text{Fan Speed} = 879 \text{ RPM}$$

Fan Pulses per Revolution

Different fan models can output either 1, 2, 3, or 4 TACH pulses per revolution. Once the number of fan TACH pulses has been determined, it can be programmed into the fan pulses per revolution register (Reg. 0x7B) for each fan. Alternatively, this register can be used to determine the number of pulses per revolution output by a given fan. By plotting fan speed measurements at 100% speed with different pulses per revolution setting, the smoothest graph with the lowest ripple determines the correct pulses per revolution value.

Fan Pulses per Revolution Register

<1:0> Fan 1 default = 2 pulses per revolution

<3:2> Fan 2 default = 2 pulses per revolution

<5:4> Fan 3 default = 2 pulses per revolution

<7:6> Fan 4 default = 2 pulses per revolution

00 = 1 pulse per revolution

01 = 2 pulses per revolution

10 = 3 pulses per revolution

11 = 4 pulses per revolution

2-Wire Fan Speed Measurements (Low Frequency Mode Only)

The ADT7468 is capable of measuring the speed of 2-wire fans, that is, fans without TACH outputs. To do this, the fan must be interfaced as shown in the Driving 2-Wire Fans section. In this case, the TACH inputs should be reprogrammed as analog inputs, AIN.

Configuration Register 2 (Reg. 0x73)

Bit 3 (AIN4) = 1, Pin 14 is reconfigured to measure the speed of a 2-wire fan using an external sensing resistor and coupling capacitor.

Bit 2 (AIN3) = 1, Pin 9 is reconfigured to measure the speed of a 2-wire fan using an external sensing resistor and coupling capacitor.

Bit 1 (AIN2) = 1, Pin 12 is reconfigured to measure the speed of a 2-wire fan using an external sensing resistor and coupling capacitor.

Bit 0 (AIN1) = 1, Pin 11 is reconfigured to measure the speed of a 2-wire fan using an external sensing resistor and coupling capacitor.

AIN Switching Threshold

Having configured the TACH inputs as AIN inputs for 2-wire measurements, a user can select the sensing threshold for the AIN signal.

Configuration Register 4 (Reg. 0x7D)

<3:2> AINL, input threshold for 2-wire fan speed measurements.

- 00 = ± 20 mV
- 01 = ± 40 mV
- 10 = ± 80 mV
- 11 = ± 130 mV

FAN SPIN-UP

The ADT7468 has a unique fan spin-up function. It spins the fan at 100% PWM duty cycle until two TACH pulses are detected on the TACH input. Once two TACH pulses have been detected, the PWM duty cycle goes to the expected running value, for example, 33%. The advantage is that fans have different spin-up characteristics and take different times to overcome inertia. The ADT7468 runs the fans just fast enough to overcome inertia and is quieter during spin-up than fans programmed to spin up for a given spin-up time.

Fan Start-Up Timeout

To prevent the generation of false interrupts as a fan spins up (because it is below running speed), the ADT7468 includes a fan start-up timeout function. During this time, the ADT7468 looks for two TACH pulses. If two TACH pulses are not detected, then an interrupt is generated.

Using Configuration Register 4 (0x40) Bit 5 (FSPDIS), this functionality can be changed (see the Disabling Fan Start-Up Timeout section).

PWM1 Configuration (Reg. 0x5C)

<2:0> spin, startup timeout for PWM1.

- 000 = No startup timeout
- 001 = 100 ms
- 010 = 250 ms default
- 011 = 400 ms
- 100 = 667 ms
- 101 = 1 sec
- 110 = 2 sec
- 111 = 4 sec

PWM2 Configuration (Reg. 0x5D)

<2:0> spin, startup timeout for PWM2.

- 000 = No startup timeout
- 001 = 100 ms
- 010 = 250 ms default
- 011 = 400 ms
- 100 = 667 ms
- 101 = 1 sec
- 110 = 2 sec
- 111 = 4 sec

PWM3 Configuration (Reg. 0x5E)

<2:0> spin, start-up timeout for PWM3.

- 000 = No startup timeout
- 001 = 100 ms
- 010 = 250 ms default
- 011 = 400 ms
- 100 = 667 ms
- 101 = 1 sec
- 110 = 2 sec
- 111 = 4 sec

Disabling Fan Start-Up Timeout

Although fan start-up times make fan spin-ups much quieter than fixed-time spin-up times, the option exists to use fixed spin-up times. Setting Bit 5 (FSPDIS) to 1 in Configuration Register 1 (Reg. 0x40) disables the spin-up for two TACH pulses. Instead, the fan spins up for the fixed time as selected in Reg. 0x5C to Reg. 0x5E.

PWM LOGIC STATE

The PWM outputs can be programmed high for 100% duty cycle (noninverted) or low for 100% duty cycle (inverted).

PWM1 Configuration (Reg. 0x5C)

<4> INV

- 0 = Logic high for 100% PWM duty cycle
- 1 = Logic low for 100% PWM duty cycle

PWM2 Configuration (Reg. 0x5D)

<4> INV

- 0 = Logic high for 100% PWM duty cycle
- 1 = Logic low for 100% PWM duty cycle

PWM3 Configuration (Reg. 0x5E)

<4> INV

- 0 = Logic high for 100% PWM duty cycle
- 1 = Logic low for 100% PWM duty cycle

Low Frequency Mode PWM Drive Frequency

The PWM drive frequency can be adjusted for the application. Register 0x5F to Register 0x61 configure the PWM frequency for PWM1 to PWM3, respectively. In high frequency mode, the PWM drive frequency is always 22.5 kHz and cannot be changed.

PWM1 Frequency Registers (Reg. 0x5F to Reg. 0x61)

<2:0> frequency

000 = 11.0 Hz

001 = 14.7 Hz

010 = 22.1 Hz

011 = 29.4 Hz

100 = 35.3 Hz default

101 = 44.1 Hz

110 = 58.8 Hz

111 = 88.2 Hz

FAN SPEED CONTROL

The ADT7468 controls fan speed using two modes: automatic and manual.

In automatic fan speed control mode, fan speed automatically varies with temperature, without CPU intervention, once initial parameters are set up. The advantage is that, if the CPU hangs, the user is guaranteed that the system is protected from overheating. The automatic fan speed control incorporates a feature called dynamic T_{MIN} calibration, which reduces the design effort required to program the automatic fan speed control loop. See the Programming the Automatic Fan Speed Control Loop section.

In manual fan speed control mode, the ADT7468 allows the duty cycle of any PWM output to be manually adjusted. This can be useful if the user wants to change fan speed in the software or adjust PWM duty cycle output for test purposes. Bits <7:5> of Reg. 0x5C to Reg. 0x5E (PWM Configuration) control the behavior of each PWM output.

PWM Configuration Register (Reg. 0x5C to Reg. 0x5E)

<7:5> BHVR

111 = manual mode

Once under manual control, each PWM output can be manually updated by writing to Reg. 0x30 to Reg. 0x32 (PWMx current duty cycle registers).

Programming the PWM Current Duty Cycle Registers

The PWM current duty cycle registers are 8-bit registers that allow the PWM duty cycle for each output to be set anywhere from 0% to 100% in steps of 0.39%. The value to be programmed into the PWM_{MIN} register is given by

$$Value \text{ (decimal)} = PWM_{MIN} / 0.39$$

Example 1: For a PWM duty cycle of 50%,

$$Value \text{ (decimal)} = 50 / 0.39 = 128 \text{ (decimal)}$$

$$Value = 128 \text{ (decimal) or } 0x80 \text{ (hex)}$$

Example 2: For a PWM duty cycle of 33%,

$$Value \text{ (decimal)} = 33 / 0.39 = 85 \text{ (decimal)}$$

$$Value = 85 \text{ (decimal) or } 0x54 \text{ (hex)}$$

PWM Duty Cycle Registers

Reg. 0x30 PWM1 duty cycle = 0x00 (0% default)

Reg. 0x31 PWM2 duty cycle = 0x00 (0% default)

Reg. 0x32 PWM3 duty cycle = 0x00 (0% default)

By reading the PWMx current duty cycle registers, the user can keep track of the current duty cycle on each PWM output, even when the fans are running in automatic fan speed control mode or acoustic enhancement mode. See the Programming the Automatic Fan Speed Control Loop section for details.

MISCELLANEOUS FUNCTIONS

OPERATING FROM 3.3 V STANDBY

The ADT7468 has been specifically designed to operate from a 3.3 V STBY supply. In computers that support S3 and S5 states, the core voltage of the processor is lowered in these states. If using the dynamic T_{MIN} mode, lowering the core voltage of the processor changes the CPU temperature and changes the dynamics of the system under dynamic T_{MIN} control. Likewise, when monitoring THERM, the THERM timer should be disabled during these states.

Dynamic T_{MIN} Control Register 1 (Reg. 0x36)

<1> $V_{CCPLO} = 1$

When the power is supplied from 3.3 V STBY and the V_{CCP} voltage drops below the V_{CCP} low limit, the following occurs:

1. Status Bit 1 (V_{CCP}) in Status Register 1 is set.
2. $\overline{SMBALERT}$ is generated, if enabled.
3. $\overline{THERM}$ monitoring is disabled. The THERM timer should hold its value prior to the S3 or S5 state.
4. Dynamic T_{MIN} control is disabled. This prevents T_{MIN} from being adjusted due to an S3 or S5 state.
5. The ADT7468 is prevented from entering the shutdown state.

Once the core voltage, V_{CCP} , goes above the V_{CCP} low limit, everything is re-enabled and the system resumes normal operation.

XNOR TREE TEST MODE

The ADT7468 includes an XNOR tree test mode. This mode is useful for in-circuit test equipment at board-level testing. By applying a stimulus to the pins included in the XNOR tree, it is possible to detect opens or shorts on the system board.

The XNOR tree test is invoked by setting Bit 0 (XEN) of the XNOR tree test enable register (Reg. 0x6F).

Figure 47 shows the signals that are exercised in the XNOR tree test mode.

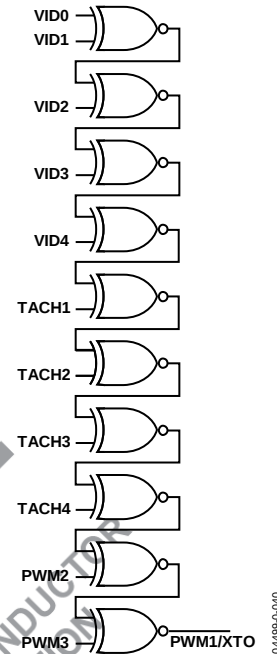


Figure 47. XNOR Tree Test

POWER-ON DEFAULT

When the ADT7468 is powered up, it polls the V_{CCP} input.

If V_{CCP} stays below 0.75 V (the system CPU power rail is not powered-up), then the ADT7468 assumes the functionality of the default registers after the ADT7468 is addressed via any valid SMBus transaction.

If V_{CCP} goes high (the system processor power rail is powered up), then a fail-safe timer begins to count down. If the ADT7468 is not addressed by any valid SMBus transaction before the fail-safe timeout (4.6 s) lapses, then the ADT7468 drives the fans to full speed. If the ADT7468 is addressed by a valid SMBus transaction after this point, the fans stop, and the ADT7468 assumes its default settings and begins normal operation.

If V_{CCP} goes high (the system processor power rail is powered up), then a fail-safe timer begins to count down. If the ADT7468 has been addressed by a valid SMBus transaction before the fail-safe timeout (4.6 sec) lapsed, then the ADT7468 operates normally, assuming the functionality of all the default registers. See the flow chart in Figure 48.

ADT7468

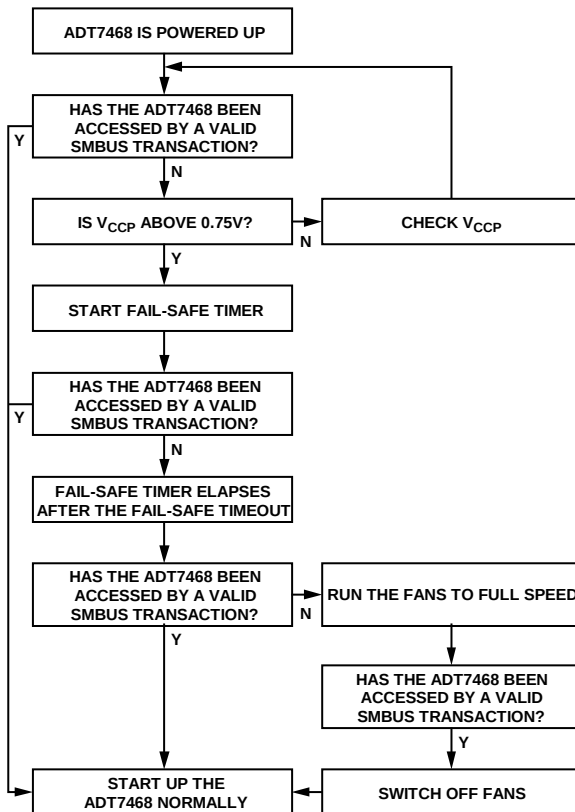


Figure 48. Power-On Flow Chart

AUTOMATIC FAN CONTROL OVERVIEW

The ADT7468 can automatically control the speed of fans based upon the measured temperature. This is done independently of CPU intervention once initial parameters are set up.

The ADT7468 has a local temperature sensor and two remote temperature channels that can be connected to a CPU on-chip thermal diode (available on Intel Pentium class and other CPUs). These three temperature channels can be used as the basis for automatic fan speed control to drive fans using pulse-width modulation (PWM).

Automatic fan speed control reduces acoustic noise by optimizing fan speed according to accurately measured temperature. Reducing fan speed can also decrease system current consumption. The automatic fan speed control mode is very flexible, owing to the number of programmable parameters, including T_{MIN} and T_{RANGE} . The T_{MIN} and T_{RANGE} values for a temperature channel and, therefore, for a given fan are critical, because they define the thermal characteristics of the system. The thermal validation of the system is one of the most important steps in the design process, so these values should be selected carefully.

Figure 49 gives a top level overview of the automatic fan control circuitry on the ADT7468. From a systems-level perspective, up to three system temperatures can be monitored and used to control three PWM outputs. The three PWM outputs can be used to control up to four fans. The ADT7468 allows the speed of four fans to be monitored. Each temperature channel has a thermal calibration block, allowing the designer to individually configure the thermal characteristics of each temperature channel. For example, one can decide to run the CPU fan when the CPU temperature increases above 60°C and a chassis fan when the local temperature increases above 45°C. At this stage, the designer has not assigned these thermal calibration settings to a particular fan drive (PWM) channel. The right side of Figure 49 shows controls that are fan-specific. The designer has individual control over parameters such as minimum PWM duty cycle, fan speed failure thresholds, and even ramp control of the PWM outputs. Automatic fan control, then, ultimately allows graceful fan speed changes that are less perceptible to the system user.

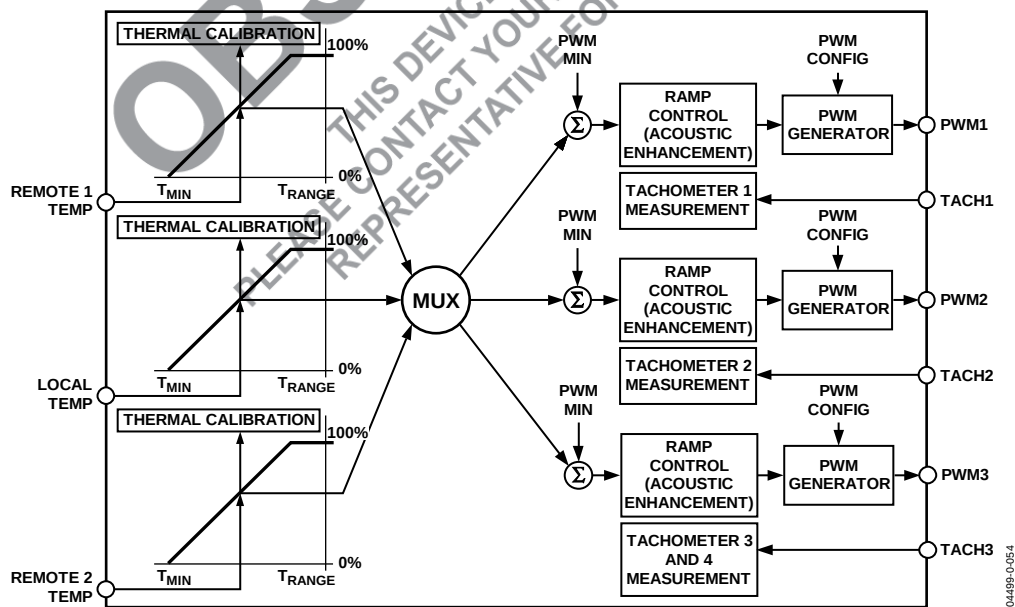


Figure 49. Automatic Fan Control Block Diagram

DYNAMIC T_{MIN} CONTROL MODE

In addition to the automatic fan speed control mode described in the Automatic Fan Control Overview section, the ADT7468 has a mode that extends the basic automatic fan speed control loop. Dynamic T_{MIN} control allows the ADT7468 to intelligently adapt the system's cooling solution for best system performance or lowest possible system acoustics, depending on user or design requirements. Use of dynamic T_{MIN} control alleviates the need to design for worst-case conditions and significantly reduces system design and validation time.

Designing for Worst-Case Conditions

System design must always allow for worst-case conditions. In PC design, the worst-case conditions include, but are not limited to the following:

- **Worst-Case Altitude**

A computer can be operated at different altitudes. The altitude affects the relative air density, which alters the effectiveness of the fan cooling solution. For example, comparing 40°C air temperature at 10,000 ft. to 20°C air temperature at sea level, relative air density is increased by 40%. This means that the fan can spin 40% slower and make less noise at sea level than at 10,000 ft., while keeping the system at the same temperature at both locations.

- **Worst-Case Fan**

Due to manufacturing tolerances, fan speeds in RPM are normally quoted with a tolerance of $\pm 20\%$. The designer needs to assume that the fan RPM can be 20% below tolerance. This translates to reduced system airflow and elevated system temperature. Note that fans 20% out of tolerance can negatively impact system acoustics, because they run faster and generate more noise.

- **Worst-Case Chassis Airflow**

The same motherboard can be used in a number of different chassis configurations. The design of the chassis and the physical location of fans and components determine the system thermal characteristics. Moreover, for a given chassis, the addition of add-in cards, cables, or other system configuration options can alter the system airflow and reduce the effectiveness of the system cooling solution. The cooling solution can also be inadvertently altered by the end user. For example, placing a computer against a wall can block the air ducts and reduce system airflow.

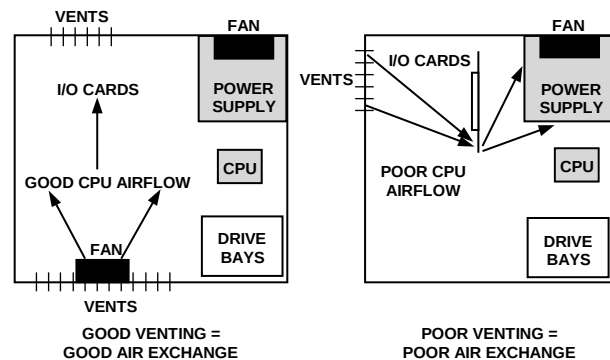


Figure 50. Chassis Airflow Issues

- **Worst-Case Processor Power Consumption**

This data sheet maximum does not necessarily reflect the true processor power consumption. Designing for worst-case CPU power consumption can result in a processor becoming overcooled (generating excess system noise).

- **Worst-Case Peripheral Power Consumption**

The tendency is to design to data sheet maximums for peripheral components (again overcooling the system).

- **Worst-Case Assembly**

Every system manufactured is unique because of manufacturing variations. Heat sinks may be loose fitting or slightly misaligned. Too much or too little thermal grease might be used, or variations in application pressure for thermal interface material could affect the efficiency of the thermal solution. Accounting for manufacturing variations in every system is difficult; therefore, the system must be designed for the worst case.

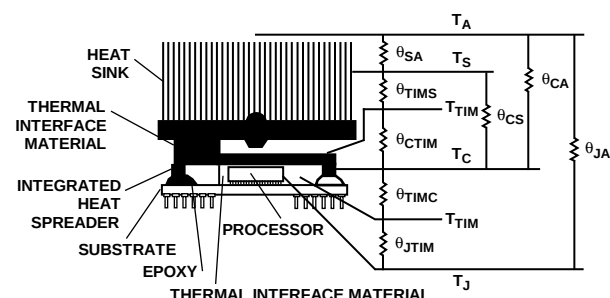


Figure 51. Thermal Model

Although a design usually accounts for worst-case conditions in all these cases, the actual system is almost never operated at worst-case conditions. The alternative to designing for the worst case is to use the dynamic T_{MIN} control function.

Dynamic T_{MIN} Control Overview

Dynamic T_{MIN} control mode builds upon the basic automatic fan control loop by adjusting the T_{MIN} value based on system performance and measured temperature. This is important, because, instead of designing for the worst case, the system thermals can be defined as operating zones. ADT7468 can self-adjust its fan control loop to maintain either an operating zone temperature or a system target temperature. For example, one can specify that the ambient temperature in a system should be maintained at 50°C. If the temperature is below 50°C, the fans might not need to run or might run very slowly. If the temperature is higher than 50°C, the fans need to throttle up.

The challenge presented by any thermal design is finding the right settings to suit the system's fan control solution. This can involve designing for the worst case, followed by weeks of system thermal characterization, and finally fan acoustic optimization (for psycho-acoustic reasons). Getting the most benefit from the automatic fan control mode involves characterizing the system to find the best T_{MIN} and T_{RANGE} settings for the control loop, and the best PWM_{MIN} value for the quietest fan speed setting. Using the ADT7468's dynamic T_{MIN} control mode, however, shortens the characterization time and alleviates tweaking the control loop settings, because the device can self-adjust during system operation.

Dynamic T_{MIN} control mode is operated by specifying the operating zone temperatures required for the system. Associated with this control mode are three operating point registers, one for each temperature channel. This allows the system thermal solution to be broken down into distinct thermal zones. For example, CPU operating temperature is 70°C, VRM operating temperature is 80°C, and ambient operating temperature is 50°C. The ADT7468 dynamically alters the control solution to maintain each zone temperature as closely as possible to its target operating point.

Operating Point Registers

Reg. 0x33, Remote 1 operating point = 0xA4 (100°C default)

Reg. 0x34, local operating point = 0xA4 (100°C default)

Reg. 0x35, Remote 2 operating point = 0xA4 (100°C default)

Figure 52 shows an overview of the parameters that affect the operation of the dynamic T_{MIN} control loop.

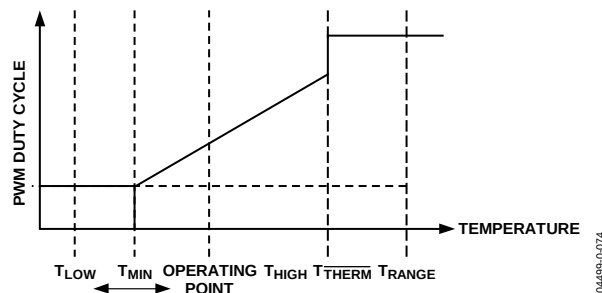


Figure 52. Dynamic T_{MIN} Control Loop

Table 14 provides a brief description of each parameter.

Table 14. T_{MIN} Control Loop Parameters

Parameter	Description
T_{LOW}	If the temperature drops below the T_{LOW} limit, an error flag is set in a status register and an SMBALERT interrupt can be generated.
T_{HIGH}	If the temperature exceeds the T_{HIGH} limit, an error flag is set in a status register and an SMBALERT interrupt can be generated.
T_{MIN}	The temperature at which the fan turns on under automatic fan speed control.
Operating point	The target temperature for a particular temperature zone. The ADT7468 attempts to maintain system temperature at about the operating point by adjusting the T_{MIN} parameter of the control loop.
T_{THERM}	If the temperature exceeds this critical limit, the fans can be run at 100% for maximum cooling.
T_{RANGE}	Programs the PWM duty cycle vs. temperature control slope.

Dynamic T_{MIN} Control Programming

Because the dynamic T_{MIN} control mode is a basic extension of the automatic fan control mode, program the automatic fan control mode parameters first, as described in Step 1 to Step 8, then proceed with dynamic T_{MIN} control mode programming.

PROGRAMMING THE AUTOMATIC FAN SPEED CONTROL LOOP

Note that to more efficiently understand the automatic fan speed control loop, it is strongly recommended to use the ADT7468 evaluation board and software while reading this section.

This section provides the system designer with an understanding of the automatic fan control loop, and provides step-by-step guidance for effectively evaluating and selecting critical system parameters. To optimize the system characteristics, the designer needs to give some thought to system configuration, including the number of fans, where they are located, and what temperatures are being measured in the particular system.

The mechanical or thermal engineer who is tasked with the system thermal characterization should also be involved at the beginning of the process.

STEP 1: HARDWARE CONFIGURATION

During system design, the motherboard sensing and control capabilities should be addressed early in the design stages. Decisions about how these capabilities are used should involve the system thermal/mechanical engineer. Ask the following questions:

1. What ADT7468 functionality will be used?
 - PWM2 or SMBALERT?
 - TACH4 fan speed measurement or overtemperature THERM function?

- 5 V voltage monitoring or overtemperature THERM function?
- 12 V voltage monitoring or VID5 input?

The ADT7468 offers multifunctional pins that can be reconfigured to suit different system requirements and physical layouts. These multifunction pins are software programmable.

2. How many fans will be supported in system, three or four? This influences the choice of whether to use the TACH4 pin or to reconfigure it for the THERM function.
3. Is the CPU fan to be controlled using the ADT7468 or will it run at full speed 100% of the time?

If it runs at 100%, this frees up a PWM output, but the system is louder.

4. Where will the ADT7468 be physically located in the system?

This influences the assignment of the temperature measurement channels to particular system thermal zones. For example, locating the ADT7468 close to the VRM controller circuitry allows the VRM temperature to be monitored using the local temperature channel.

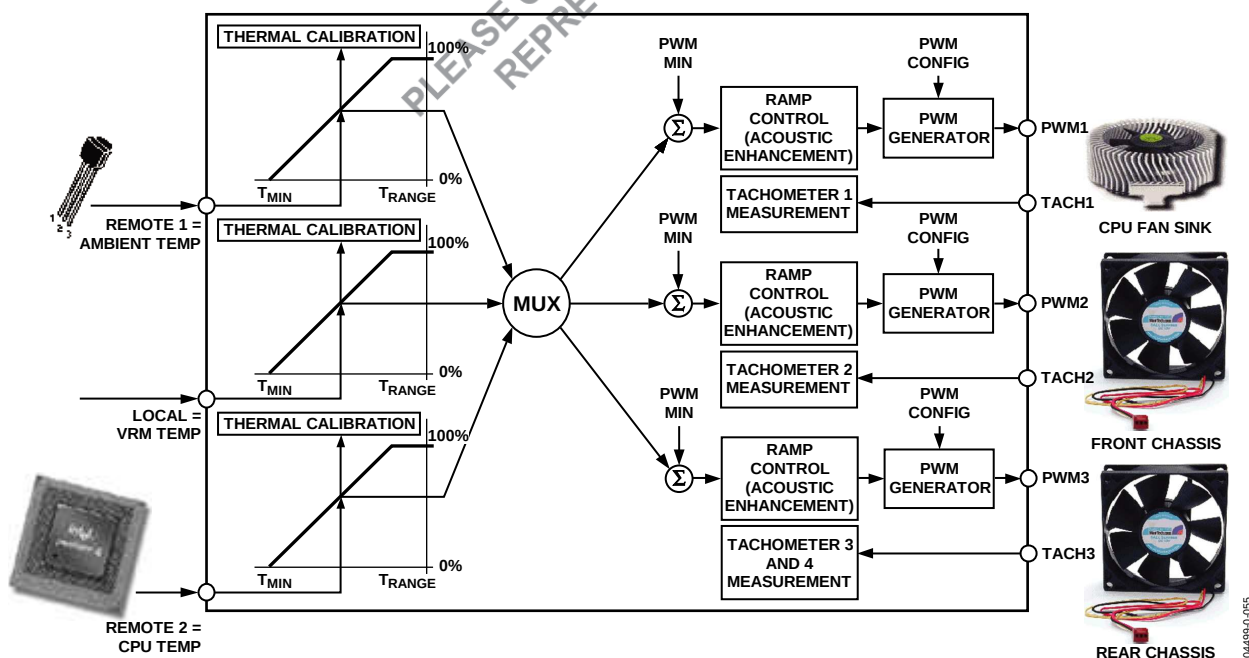


Figure 53. Hardware Configuration Example

Recommended Implementation 1

Configuring the ADT7468 as shown in Figure 54 provides the system designer with the following features:

1. Six VID inputs (VID0 to VID5) for VRM10 support.
2. Two PWM outputs for fan control of up to three fans. (The front and rear chassis fans are connected in parallel.)
3. Three TACH fan speed measurement inputs.
4. V_{CC} measured internally through Pin 4.
5. CPU core voltage measurement (V_{CORE}).
6. 2.5 V measurement input used to monitor CPU current (connected to V_{COMP} output of ADP316x VRM controller). This is used to determine CPU power consumption.
7. 5 V measurement input.
8. VRM temperature using local temperature sensor.
9. CPU temperature measured using the Remote 1 temperature channel.
10. Ambient temperature measured through the Remote 2 temperature channel.
11. If not using VID5, this pin can be reconfigured as the 12 V monitoring input.
12. Bidirectional THERM pin allows the monitoring of PROCHOT output from an Intel P4 processor, for example, or can be used as an overtemperature THERM output.
13. SMBALERT system interrupt output.

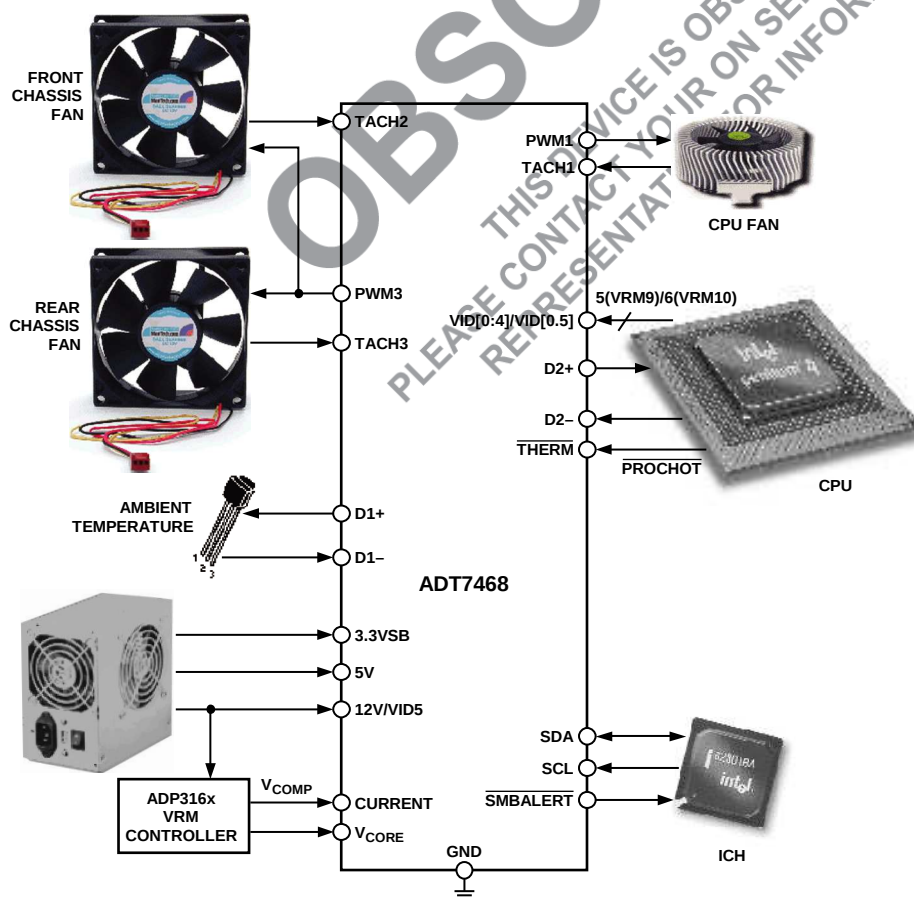


Figure 54. Recommended Implementation 1

04499-0-056

STEP 2: CONFIGURING THE MUX

After the system hardware configuration is determined, the fans can be assigned to particular temperature channels. Not only can fans be assigned to individual channels, but the behavior of the fans is also configurable. For example, fans can run under automatic fan control, they can run manually (under software control), or they can run at the fastest speed calculated by multiple temperature channels. The mux is the bridge between temperature measurement channels and the three PWM outputs.

Bits <7:5> (BHVR) of Registers 0x5C, 0x5D, and 0x5E (PWM configuration registers) control the behavior of the fans connected to the PWM1, PWM2, and PWM3 outputs. The values selected for these bits determine how the mux connects a temperature measurement channel to a PWM output.

Automatic Fan Control MUX Options

<7:5> (BHVR), Registers 0x5C, 0x5D, 0x5E.

000 = Remote 1 temperature controls PWMx

001 = local temperature controls PWMx

010 = Remote 2 temperature controls PWMx

101 = Fastest speed calculated by local and Remote 2 temperature controls PWMx

110 = Fastest speed calculated by all three temperature channels controls PWMx

The fastest speed calculated options pertain to controlling one PWM output based on multiple temperature channels. The thermal characteristics of the three temperature zones can be set to drive a single fan. An example would be the fan turning on when Remote 1 temperature exceeds 60°C or if the local temperature exceeds 45°C.

Other MUX Options

<7:5> (BHVR), Registers 0x5C, 0x5D, 0x5E.

011 = PWMx runs full speed

100 = PWMx disabled (default)

111 = manual mode. PWMx is runner under software control. In this mode, PWM duty cycle registers (Registers 0x30 to 0x32) are writable and control the PWM outputs.

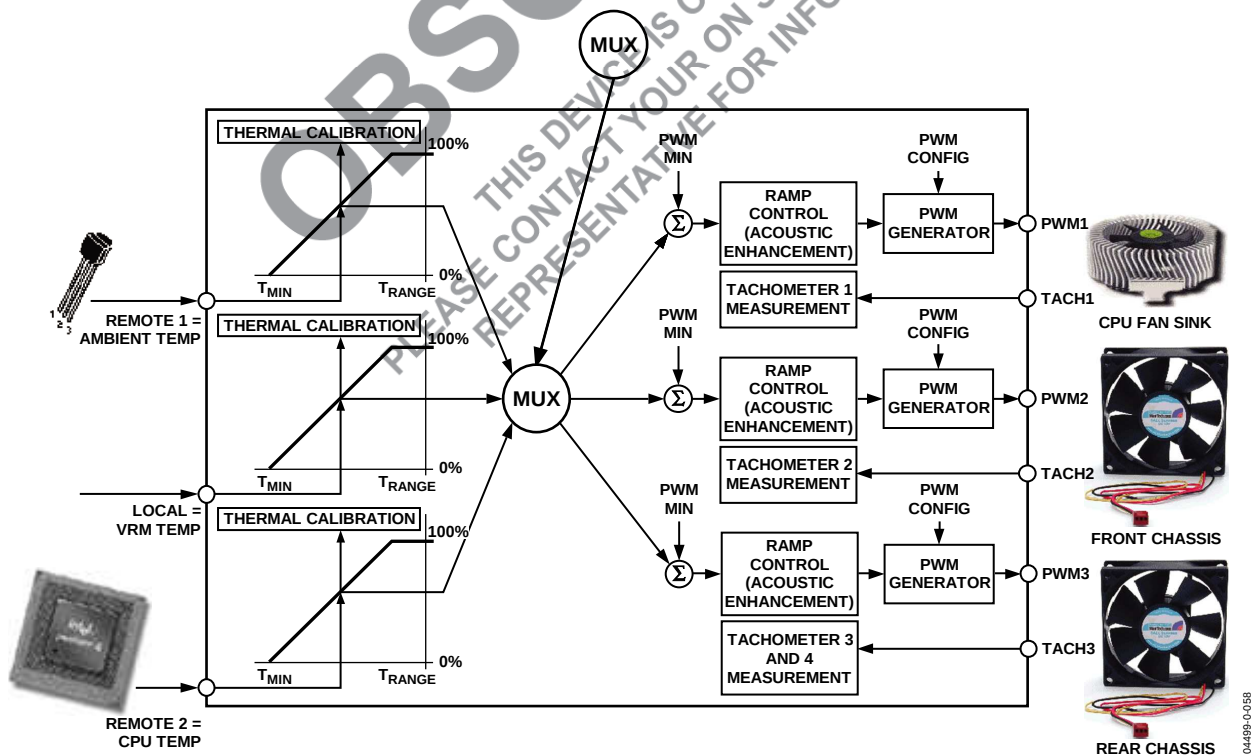


Figure 56. Assigning Temperature Channels to Fan Channels

Mux Configuration Example

This is an example of how to configure the mux in a system using the ADT7468 to control three fans. The CPU fan sink is controlled by PWM1, the front chassis fan is controlled by PWM2, and the rear chassis fan is controlled by PWM3. The MUX is configured for the following fan control behavior:

- PWM1 (CPU fan sink) is controlled by the fastest speed calculated by the local (VRM temperature) and Remote 2 (processor) temperature. In this case, the CPU fan sink is also being used to cool the VRM.
- PWM2 (front chassis fan) is controlled by the Remote 1 temperature (ambient).
- PWM3 (rear chassis fan) is controlled by the Remote 1 temperature (ambient).

Example Mux Settings

<7:5> (BHVR), PWM1 Configuration Register 0x5C.

101 = Fastest speed calculated by local and Remote 2 temperature controls PWM1

<7:5> (BHVR), PWM2 Configuration Register 0x5D.

000 = Remote 1 temperature controls PWM2

<7:5> (BHVR), PWM3 Configuration Register 0x5E.

000 = Remote 1 temperature controls PWM3

These settings configure the mux, as shown in Figure 57.

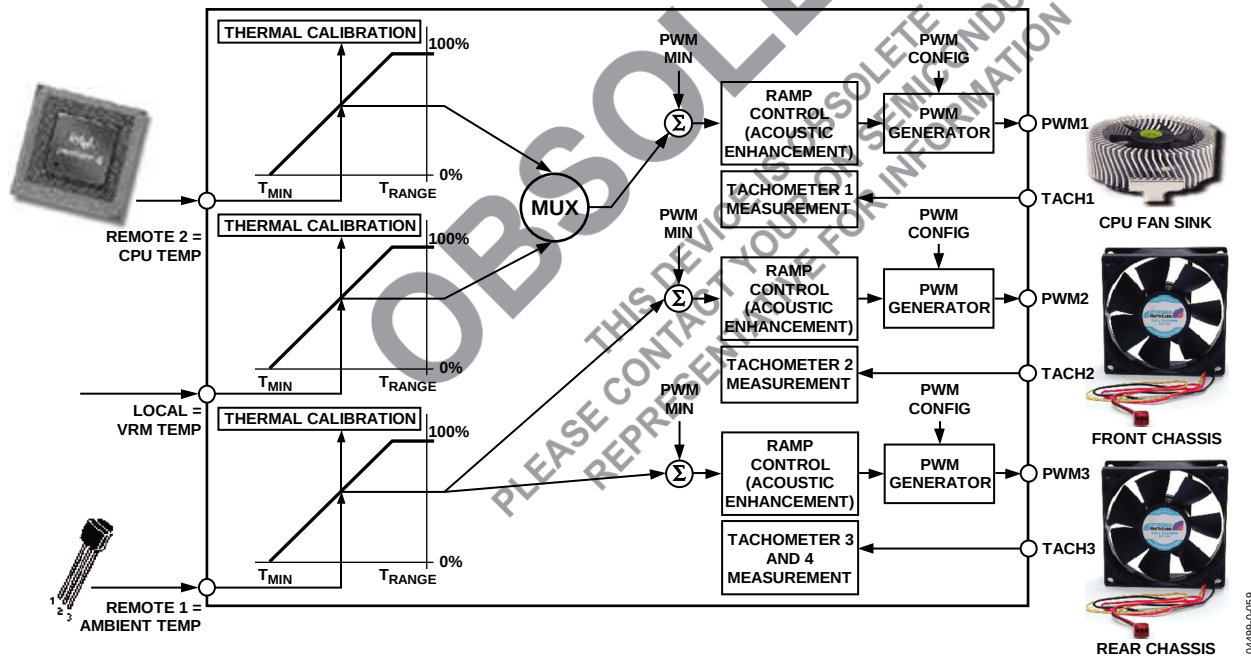


Figure 57. Mux Configuration Example

STEP 3: T_{MIN} SETTINGS FOR THERMAL CALIBRATION CHANNELS

T_{MIN} is the temperature at which the fans start to turn on under automatic fan control. The speed at which the fan runs at T_{MIN} is programmed later. The T_{MIN} values chosen are temperature channel specific, for example, 25°C for ambient channel, 30°C for VRM temperature, and 40°C for processor temperature.

T_{MIN} is an 8-bit value, either twos complement or Offset 64, that can be programmed in 1°C increments. There is a T_{MIN} register associated with each temperature measurement channel: Remote 1 local, and Remote 2 temperature. Once the T_{MIN} value is exceeded, the fan turns on and runs at the minimum PWM duty cycle. The fan turns off once the temperature has dropped below $T_{MIN} - T_{HYST}$.

To overcome fan inertia, the fan is spun up until two valid TACH rising edges are counted. See the Fan Start-Up Timeout section for more details. In some cases, primarily for psycho-acoustic reasons, it is desirable that the fan never switch off below T_{MIN} . Bits <7:5> of Enhanced Acoustics Register 1 (Reg. 0x62), when set, keep the fans running at the PWM minimum duty cycle, if the temperature should fall below T_{MIN} .

T_{MIN} Registers

Reg. 0x67, Remote 1 temperature $T_{MIN} = 0x9A$ (90°C)

Reg. 0x68, Local temperature $T_{MIN} = 0x9A$ (90°C)

Reg. 0x69, Remote 2 temperature $T_{MIN} = 0x9A$ (90°C)

Enhance Acoustics Register 1 (Reg. 0x62)

Bit 7 (MIN3) = 0, PWM3 is off (0% PWM duty cycle) when temperature is below $T_{MIN} - T_{HYST}$.

Bit 7 (MIN3) = 1, PWM3 runs at PWM3 minimum duty cycle below $T_{MIN} - T_{HYST}$.

Bit 6 (MIN2) = 0, PWM2 is off (0% PWM duty cycle) when temperature is below $T_{MIN} - T_{HYST}$.

Bit 6 (MIN2) = 1, PWM2 runs at PWM2 minimum duty cycle below $T_{MIN} - T_{HYST}$.

Bit 5 (MIN1) = 0, PWM1 is off (0% PWM duty cycle) when temperature is below $T_{MIN} - T_{HYST}$.

Bit 5 (MIN1) = 1, PWM1 runs at PWM1 minimum duty cycle below $T_{MIN} - T_{HYST}$.

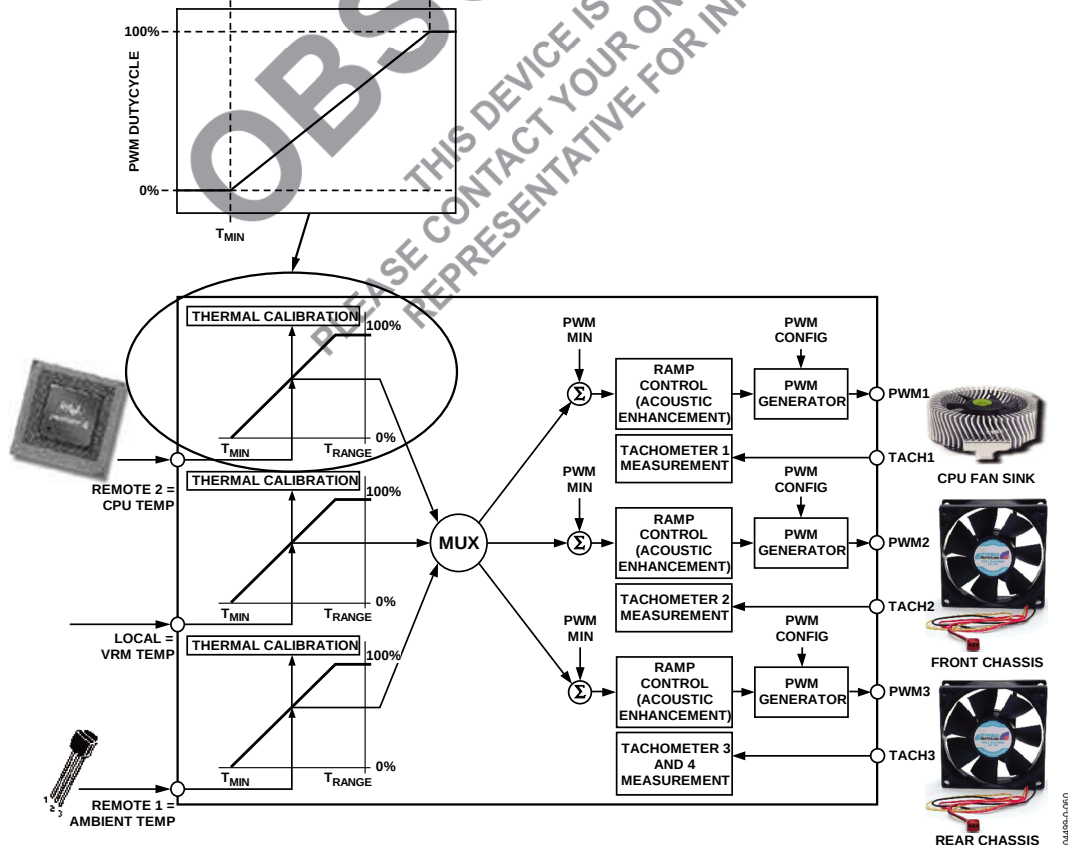


Figure 58. Understanding the T_{MIN} Parameter

STEP 4: PWM_{MIN} FOR EACH PWM (FAN) OUTPUT

PWM_{MIN} is the minimum PWM duty cycle at which each fan in the system runs. It is also the start speed for each fan under automatic fan control once the temperature rises above T_{MIN}. For maximum system acoustic benefit, PWM_{MIN} should be as low as possible. Depending on the fan used, the PWM_{MIN} setting is usually in the 20% to 33% duty cycle range. This value can be found through fan validation.

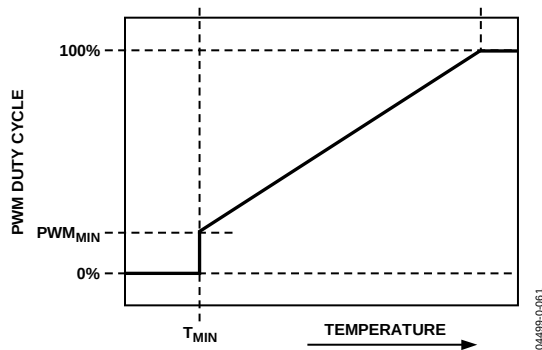


Figure 59. PWM_{MIN} Determines Minimum PWM Duty Cycle

More than one PWM output can be controlled from a single temperature measurement channel. For example, Remote 1 temperature can control PWM1 and PWM2 outputs. If two different fans are used on PWM1 and PWM2, then the fan characteristics can be set up differently. As a result, Fan 1 driven by PWM1 can have a different PWM_{MIN} value than that of Fan 2 connected to PWM2. Figure 60 illustrates this as PWM1_{MIN} (front fan) is turned on at a minimum duty cycle of 20%, while PWM2_{MIN} (rear fan) turns on at a minimum of 40% duty cycle. Note, however, that both fans turn on at exactly the same temperature, defined by T_{MIN}.

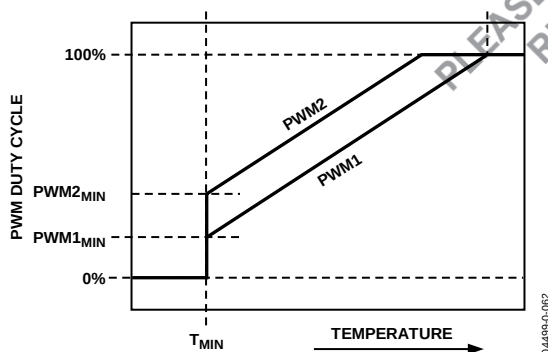


Figure 60. Operating Two Different Fans from a Single Temperature Channel

Programming the PWM_{MIN} Registers

The PWM_{MIN} registers are 8-bit registers that allow the minimum PWM duty cycle for each output to be configured anywhere from 0% to 100%. This allows the minimum PWM duty cycle to be set in steps of 0.39%.

The value to be programmed into the PWM_{MIN} register is given by

$$\text{Value (decimal)} = \text{PWM}_{\text{MIN}} / 0.39$$

Example 1: For a minimum PWM duty cycle of 50%,

$$\text{Value (decimal)} = 50 / 0.39 = 128 \text{ (decimal)}$$

$$\text{Value} = 128 \text{ (decimal) or } 80 \text{ (hex)}$$

Example 2: For a minimum PWM duty cycle of 33%,

$$\text{Value (decimal)} = 33 / 0.39 = 85 \text{ (decimal)}$$

$$\text{Value} = 85 \text{ (decimal) or } 54 \text{ (hex)}$$

PWM_{MIN} Registers

Reg. 0x64, PWM1 minimum duty cycle = 0x80 (50% default)

Reg. 0x65, PWM2 minimum duty cycle = 0x80 (50% default)

Reg. 0x66, PWM3 minimum duty cycle = 0x80 (50% default)

Note on Fan Speed and PWM Duty Cycle

The PWM duty cycle does not directly correlate to fan speed in RPM. Running a fan at 33% PWM duty cycle does not equate to running the fan at 33% speed. Driving a fan at 33% PWM duty cycle actually runs the fan at closer to 50% of its full speed. This is because fan speed in %RPM generally relates to the square root of PWM duty cycle. Given a PWM square wave as the drive signal, fan speed in RPM approximates to

$$\% \text{ fanspeed} = \sqrt{\text{PWM duty cycle} \times 10}$$

STEP 5: PWM_{MAX} FOR PWM (FAN) OUTPUTS

PWM_{MAX} is the maximum duty cycle that each fan in the system runs at under the automatic fan speed control loop. For maximum system acoustic benefit, PWM_{MAX} should be as low as possible, but should be capable of maintaining the processor temperature limit at an acceptable level. If the THERM temperature limit is exceeded, the fans are still boosted to 100% for fail-safe cooling.

There is a PWM_{MAX} limit for each fan channel. The default value of this register is 0xFF and so has no effect unless it is programmed.

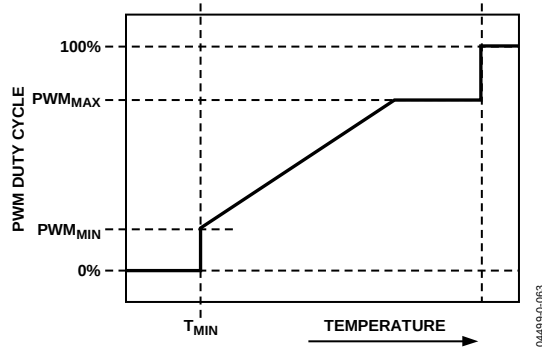


Figure 61. PWM_{MAX} Determines Maximum PWM Duty Cycle Below the T_{HERM} Temperature Limit

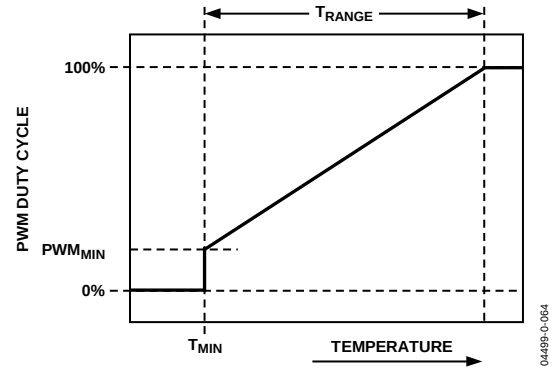


Figure 62. T_{RANGE} Parameter Affects Cooling Slope

Programming the PWM_{MAX} Registers

The PWM_{MAX} registers are 8-bit registers that allow the maximum PWM duty cycle for each output to be configured anywhere from 0% to 100%. This allows the maximum PWM duty cycle to be set in steps of 0.39%.

The value to be programmed into the PWM_{MAX} register is given by

$$\text{Value (decimal)} = PWM_{MAX}/0.39$$

Example 1: For a maximum PWM duty cycle of 50%,

$$\text{Value (decimal)} = 50/0.39 = 128 \text{ (decimal)}$$

$$\text{Value} = 128 \text{ (decimal) or } 80 \text{ (hex)}$$

Example 2: For a minimum PWM duty cycle of 75%,

$$\text{Value (decimal)} = 75/0.39 = 85 \text{ (decimal)}$$

$$\text{Value} = 192 \text{ (decimal) or } C0 \text{ (hex)}$$

PWM_{MAX} Registers

Reg. 0x38, PWM1 Maximum Duty Cycle = 0xFF (100% default)

Reg. 0x39, PWM2 Maximum Duty Cycle = 0xFF (100% default)

Reg. 0x3A, PWM3 Maximum Duty Cycle = 0xFF (100% default)

See the Note on Fan Speed and PWM Duty Cycle.

STEP 6: T_{RANGE} for Temperature Channels

T_{RANGE} is the range of temperature over which automatic fan control occurs once the programmed T_{MIN} temperature has been exceeded. T_{RANGE} is a temperature slope, not an arbitrary value; that is, a T_{RANGE} of 40°C holds true only for $PWM_{MIN} = 33\%$. If PWM_{MIN} is increased or decreased, the effective T_{RANGE} changes.

The T_{RANGE} or fan control slope is determined by the following procedure:

1. Determine the maximum operating temperature for that channel (for example, 70°C).
2. Determine experimentally the fan speed (PWM duty cycle value) that does not exceed the temperature at the worst-case operating points. (For example, 70°C is reached when the fans are running at 50% PWM duty cycle.)
3. Determine the slope of the required control loop to meet these requirements.
4. Using the ADT7468 evaluation software, you can graphically program and visualize this functionality. Ask your local Analog Devices representative for details.

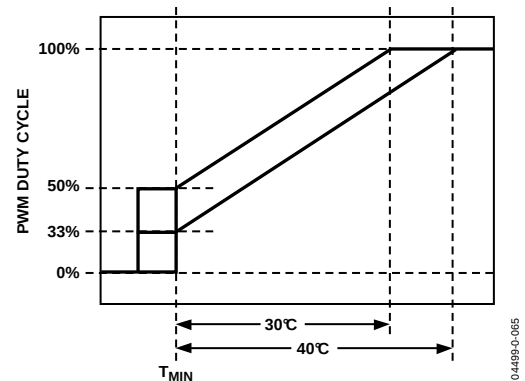


Figure 63. Adjusting PWM_{MIN} Affects T_{RANGE}

T_{RANGE} is implemented as a slope, which means that as PWM_{MIN} is changed, T_{RANGE} changes, but the actual slope remains the same. The higher the PWM_{MIN} value, the smaller the effective T_{RANGE} , that is, the fan reaches full speed (100%) at a lower temperature.

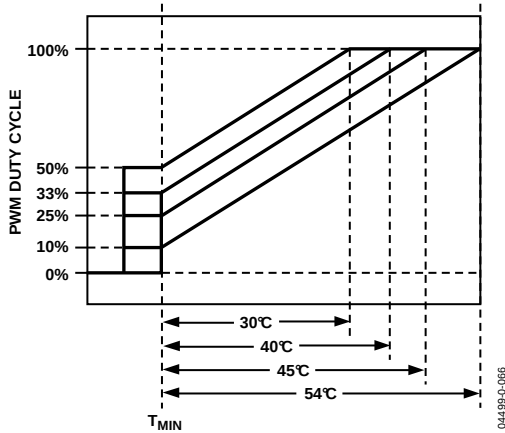


Figure 64. Increasing PWM_{MIN} Changes Effective T_{RANGE}

For a given T_{RANGE} value, the temperature at which the fan runs at full speed for different PWM_{MIN} values can be easily calculated:

$$T_{MAX} = T_{MIN} + (Max\ DC - Min\ DC) \times T_{RANGE} / 170$$

where:

T_{MAX} is the temperature at which the fan runs full speed.

T_{MIN} is the temperature at which the fan turns on.

Max DC is the maximum duty cycle (100%) = 255 decimal.

Min DC is equal to PWM_{MIN}.

T_{RANGE} is the duty PWM duty cycle vs. temperature slope.

Example: Calculate T, given that T_{MIN} = 30°C, T_{RANGE} = 40°C, and PWM_{MIN} = 10% duty cycle = 26 (decimal).

$$T_{MAX} = T_{MIN} + (Max\ DC - Min\ DC) \times T_{RANGE} / 170$$

$$T_{MAX} = 30^{\circ}C + (100\% - 10\%) \times 40^{\circ}C / 170$$

$$T_{MAX} = 30^{\circ}C + (255 - 26) \times 40^{\circ}C / 170$$

$$T_{MAX} = 84^{\circ}C\ (effective\ T_{RANGE} = 54^{\circ}C)$$

Example: Calculate T_{MAX}, given that T_{MIN} = 30°C, T_{RANGE} = 40°C, and PWM_{MIN} = 25% duty cycle = 64 (decimal).

$$T_{MAX} = T_{MIN} + (Max\ DC - Min\ DC) \times T_{RANGE} / 170$$

$$T_{MAX} = 30^{\circ}C + (100\% - 25\%) \times 40^{\circ}C / 170$$

$$T_{MAX} = 30^{\circ}C + (255 - 64) \times 40^{\circ}C / 170$$

$$T_{MAX} = 75^{\circ}C\ (effective\ T_{RANGE} = 45^{\circ}C)$$

Example: Calculate T_{MAX}, given that T_{MIN} = 30°C, T_{RANGE} = 40°C, and PWM_{MIN} = 33% duty cycle = 85 (decimal).

$$T_{MAX} = T_{MIN} + (Max\ DC - Min\ DC) \times T_{RANGE} / 170$$

$$T_{MAX} = 30^{\circ}C + (100\% - 33\%) \times 40^{\circ}C / 170$$

$$T_{MAX} = 30^{\circ}C + (255 - 85) \times 40^{\circ}C / 170$$

$$T_{MAX} = 70^{\circ}C\ (effective\ T_{RANGE} = 40^{\circ}C)$$

Example: Calculate T_{MAX}, given that T_{MIN} = 30°C, T_{RANGE} = 40°C, and PWM_{MIN} = 50% duty cycle = 128 (decimal).

$$T_{MAX} = T_{MIN} + (Max\ DC - Min\ DC) \times T_{RANGE} / 170$$

$$T_{MAX} = 30^{\circ}C + (100\% - 50\%) \times 40^{\circ}C / 170$$

$$T_{MAX} = 30^{\circ}C + (255 - 128) \times 40^{\circ}C / 170$$

$$T_{MAX} = 60^{\circ}C\ (effective\ T_{RANGE} = 30^{\circ}C)$$

Selecting a T_{RANGE} Slope

The T_{RANGE} value can be selected for each temperature channel: Remote 1, local, and Remote 2 temperature. Bits <7:4> (T_{RANGE}) of Registers 0x5F to 0x61 define the T_{RANGE} value for each temperature channel.

Table 15. Selecting a T_{RANGE} Value

Bits <7:4> ¹	T _{RANGE} (°C)
0000	2
0001	2.5
0010	3.33
0011	4
0100	5
0101	6.67
0110	8
0111	10
1000	13.33
1001	16
1010	20
1011	26.67
1100	32 (default)
1101	40
1110	53.33
1111	80

¹ Register 0x5F configures Remote 1 T_{RANGE}. Register 0x60 configures Local T_{RANGE}. Register 0x61 configures Remote 2 T_{RANGE}.

Summary of T_{RANGE} Function

When using the automatic fan control function, the temperature at which the fan reaches full speed can be calculated by

$$T_{MAX} = T_{MIN} + T_{RANGE} \quad (1)$$

Equation 1 holds true only when PWM_{MIN} is equal to 33% PWM duty cycle.

Increasing or decreasing PWM_{MIN} changes the effective T_{RANGE}, although the fan control still follows the same PWM duty cycle to temperature slope. The effective T_{RANGE} for different PWM_{MIN} values can be calculated using Equation 2:

$$T_{MAX} = T_{MIN} + (Max\ DC - Min\ DC) \times T_{RANGE}/170 \quad (2)$$

where:

$(Max\ DC - Min\ DC) \times T_{RANGE}/170$ is the effective T_{RANGE} value.

See the Note on Fan Speed and PWM Duty Cycle.

Figure 65 shows PWM duty cycle versus temperature for each T_{RANGE} setting. The lower graph shows how each T_{RANGE} setting affects fan speed versus temperature. As can be seen from the graph, the effect on fan speed is nonlinear.

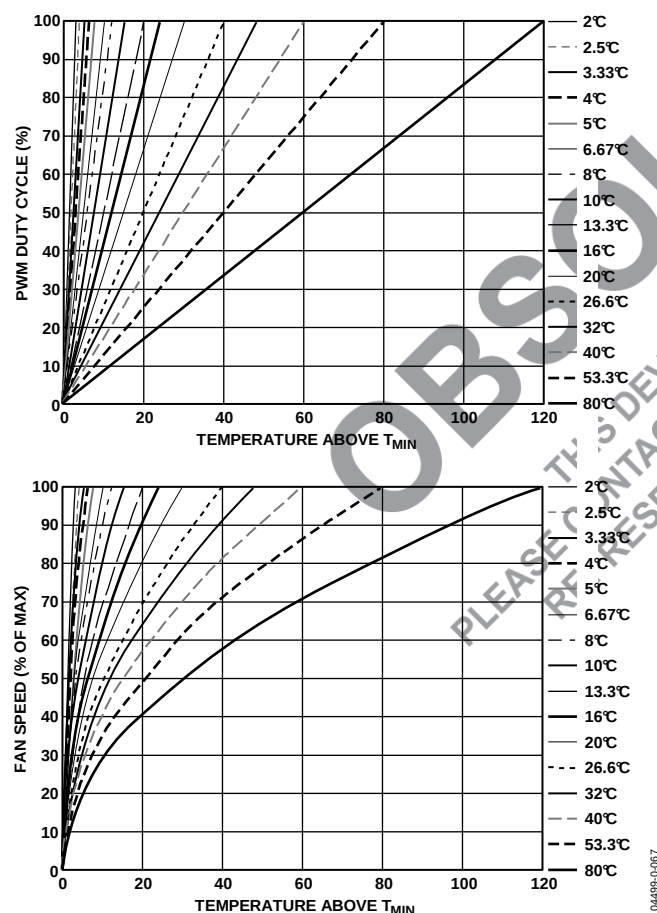


Figure 65. T_{RANGE} Vs. Actual Fan Speed Profile

The graphs in Figure 65 assume that the fan starts from 0% PWM duty cycle. Clearly, the minimum PWM duty cycle, PWM_{MIN}, needs to be factored in to see how the loop actually performs in the system. Figure 66 shows how T_{RANGE} is affected when the PWM_{MIN} value is set to 20%. It can be seen that the

fan actually runs at about 45% fan speed when the temperature exceeds T_{MIN}.

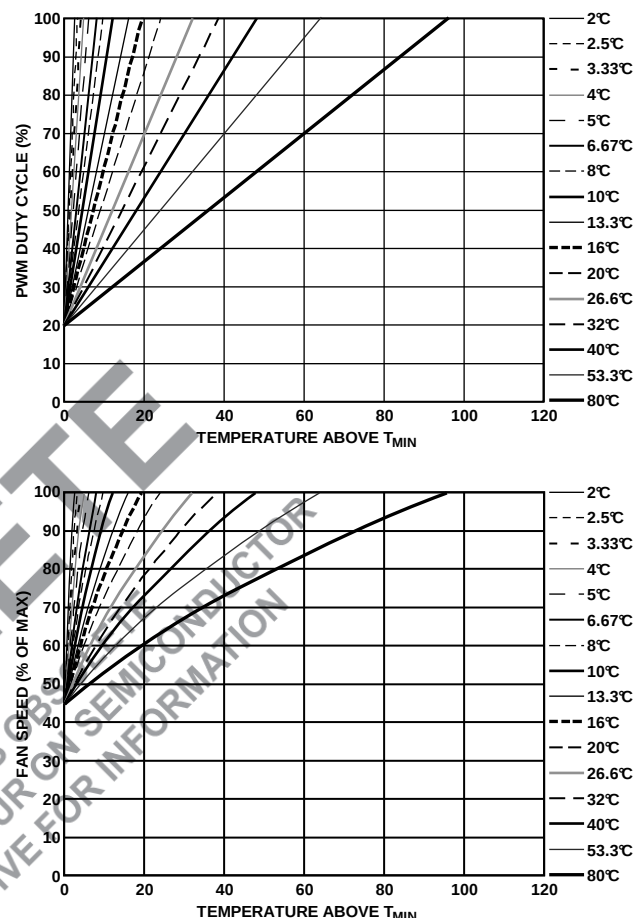


Figure 66. T_{RANGE} and % Fan Speed Slopes with PWM_{MIN} = 20%

Example: Determining T_{RANGE} for Each Temperature Channel

The following example shows how the different T_{MIN} and T_{RANGE} settings can be applied to three different thermal zones. In this example, the following T_{RANGE} values apply:

T_{RANGE} = 80°C for ambient temperature

T_{RANGE} = 53.3°C for CPU temperature

T_{RANGE} = 40°C for VRM temperature

This example uses the MUX configuration described in Step 2, with the ADT7468 connected as shown in Figure 57. Both CPU temperature and VRM temperature drive the CPU fan connected to PWM1. Ambient temperature drives the front chassis fan and rear chassis fan connected to PWM2 and PWM3. The front chassis fan is configured to run at PWM_{MIN} = 20%. The rear chassis fan is configured to run at PWM_{MIN} = 30%. The CPU fan is configured to run at PWM_{MIN} = 10%.

Note on 4-Wire Fans

The control range for 4-wire fans is much wider than that of 2-wire or 3-wire fans. In many cases, 4-wire fans can start with a PWM drive of as little as 20%.

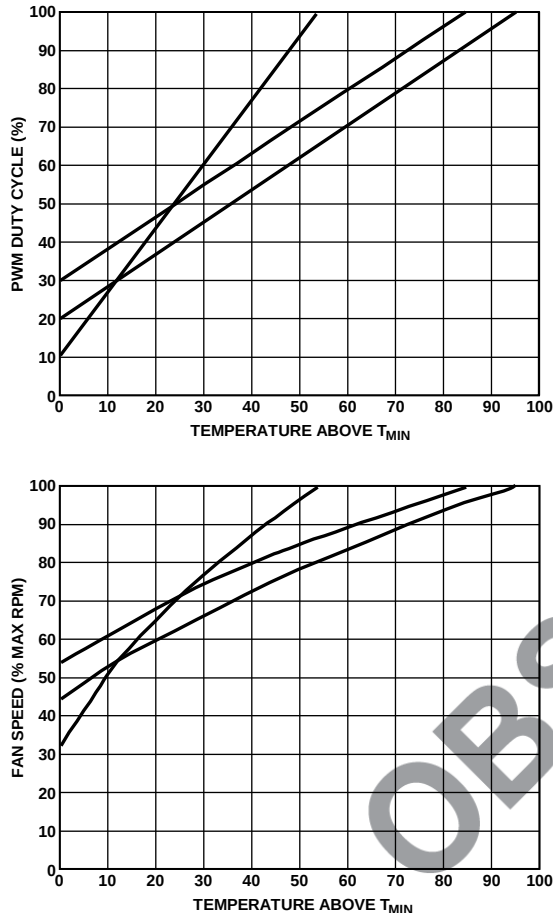


Figure 67. T_{RANGE} and % Fan Speed Slopes for VRM, Ambient, and CPU Temperature Channels

STEP 7: T_{THERM} FOR TEMPERATURE CHANNELS

T_{THERM} is the absolute maximum temperature allowed on a temperature channel. Above this temperature, a component such as the CPU or VRM might be operating beyond its safe operating limit. When the temperature measured exceeds T_{THERM} , all fans are driven at 100% PWM duty cycle (full speed) to provide critical system cooling.

The fans remain running at 100% until the temperature drops below T_{THERM} minus hysteresis, where hysteresis is the number programmed into the hysteresis Registers 0x6D and 0x6E. The default hysteresis value is 4°C.

The T_{THERM} limit should be considered the maximum worst-case operating temperature of the system. Because exceeding any T_{THERM} limit runs all fans at 100%, it has very negative acoustic effects. Ultimately, this limit should be set up as a fail-safe, and one should ensure that it is not exceeded under normal system operating conditions.

Note that the T_{THERM} limits are nonmaskable and affect the fan speed regardless of how automatic fan control settings are configured. This allows some flexibility, because a T_{RANGE} value can be selected based on its slope, while a hard limit (such as 70°C), can be programmed as T_{MAX} (the temperature at which the fan reaches full speed) by setting T_{THERM} to that limit (for example, 70°C).

THERM Registers

Reg. 0x6A, Remote 1 THERM limit = 0xA4 (100°C default)

Reg. 0x6B, Local THERM limit = 0xA4 (100°C default)

Reg. 0x6C, Remote 2 THERM limit = 0xA4 (100°C default)

Hysteresis Registers

Reg. 0x6D, Remote 1, local hysteresis register

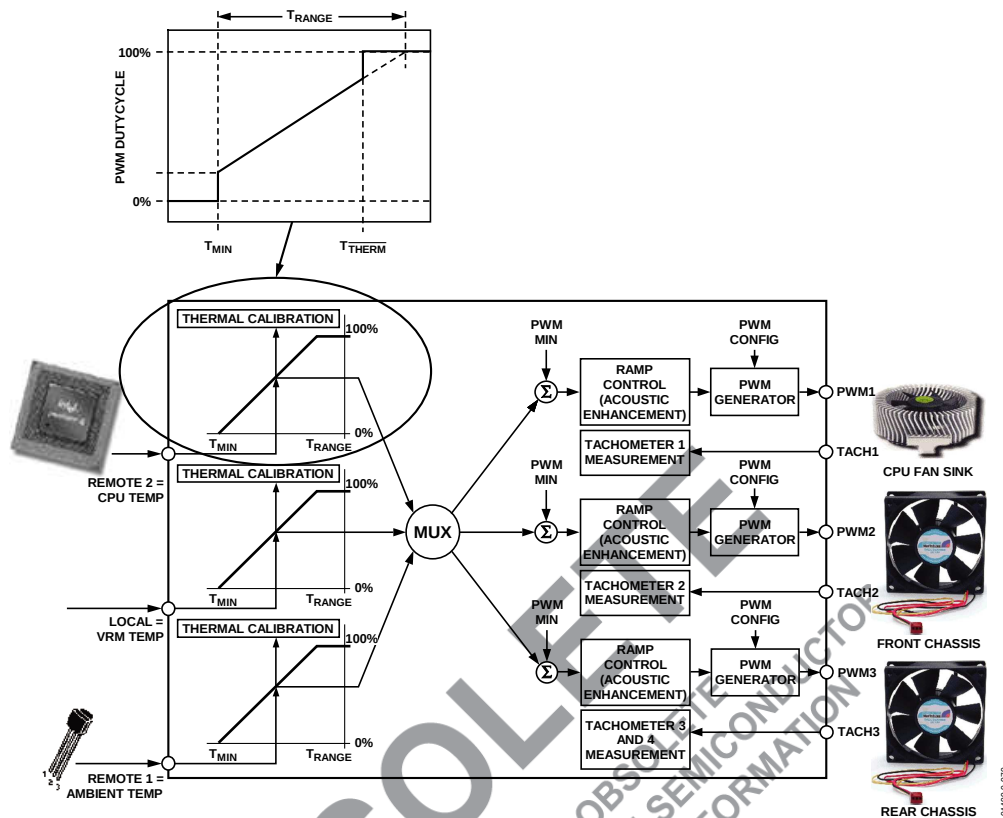
<7:4>, Remote 1 temperature hysteresis (4°C default)

<3:0>, Local temperature hysteresis (4°C default)

Reg. 0x6E, Remote 2 Temperature Hysteresis Register

<7:4>, Remote 2 temperature hysteresis (4°C default)

Because each hysteresis setting is four bits, hysteresis values are programmable from 1°C to 15°C. It is recommended that hysteresis values never be programmed to 0°C, because this disables hysteresis. In effect, this would cause the fans to cycle between normal speed and 100% speed, creating unsettling acoustic noise.

Figure 68. How T_{THERM} Relates to Automatic Fan Control

STEP 8: T_{HYST} FOR TEMPERATURE CHANNELS

T_{HYST} is the amount of extra cooling a fan provides after the temperature measured has dropped back below T_{MIN} before the fan turns off. The premise for temperature hysteresis (T_{HYST}) is that, without it, the fan would merely chatter or cycle on and off regularly whenever temperature is hovering at about the T_{MIN} setting.

The T_{HYST} value chosen determines the amount of time needed for the system to cool down or heat up as the fan is turning on and off. Values of hysteresis are programmable in the range 1°C to 15°C. Larger values of T_{HYST} prevent the fans from chattering on and off. The T_{HYST} default value is set at 4°C.

The T_{HYST} setting applies not only to the temperature hysteresis for fan on/off, but the same setting is used for the T_{THERM} hysteresis value, described in Step 6. Therefore, programming Registers 0x6D and 0x6E sets the hysteresis for both fan on/off and the T_{THERM} function.

Hysteresis Registers

Reg. 0x6D, Remote 1, local hysteresis register

<7:4>, Remote 1 temperature hysteresis (4°C default).

<3:0>, local temperature hysteresis (4°C default).

Reg. 0x6E, Remote 2 temperature hysteresis register

<7:4>, Remote 2 temperature hysteresis (4°C default).

In some applications, it is required that fans not turn off below T_{MIN} , but remain running at PWM_{MIN} . Bits <7:5> of Enhanced Acoustics Register 1 (Reg. 0x62) allow the fans to be turned off or to be kept spinning below T_{MIN} . If the fans are always on, the T_{HYST} value has no effect on the fan when the temperature drops below T_{MIN} .

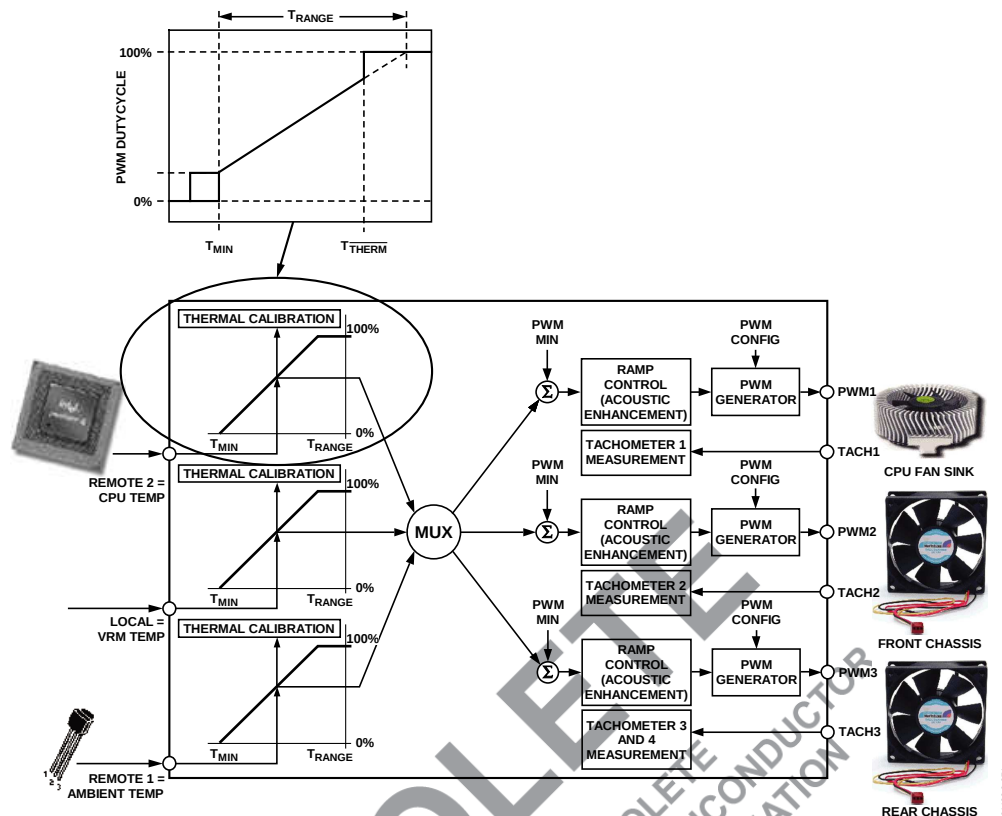


Figure 69. The T_{HYST} Value Applies to Fan On/Off Hysteresis and T_{THERM} Hysteresis

Enhance Acoustics Register 1 (Reg. 0x62)

Bit 7 (MIN3) = 0, PWM3 is off (0% PWM duty cycle) when the temperature is below $T_{MIN} - T_{HYST}$.

Bit 7 (MIN3) = 1, PWM3 runs at PWM3 minimum duty cycle below $T_{MIN} - T_{HYST}$.

Bit 6 (MIN2) = 0, PWM2 is off (0% PWM duty cycle) when the temperature is below $T_{MIN} - T_{HYST}$.

Bit 6 (MIN2) = 1, PWM2 runs at PWM2 minimum duty cycle below $T_{MIN} - T_{HYST}$.

Bit 5 (MIN1) = 0, PWM1 is off (0% PWM duty cycle) when the temperature is below $T_{MIN} - T_{HYST}$.

Bit 5 (MIN1) = 1, PWM1 runs at PWM1 minimum duty cycle below $T_{MIN} - T_{HYST}$.

STEP 9: OPERATING POINTS FOR TEMPERATURE CHANNELS

The operating point for each temperature channel is the optimal temperature for that thermal zone. The hotter each zone is allowed to be, the quieter the system, because the fans

are not required to run as fast. The ADT7468 increases or decreases fan speeds as necessary to maintain the operating point temperature, allowing for system-to-system variation and removing the need for worst-case design. If a sensible operating point value is chosen, any T_{MIN} value can be selected in the system characterization. If the T_{MIN} value is too low, the fans run sooner than required, and the temperature is below the operating point. In response, the ADT7468 increases T_{MIN} to keep the fans off longer and to allow the temperature zone to get closer to the operating point. Likewise, too high a T_{MIN} value causes the operating point to be exceeded, and in turn, the ADT7468 reduces T_{MIN} to turn the fans on sooner to cool the system.

Programming Operating Point Registers

There are three operating point registers, one for each temperature channel. These 8-bit registers allow the operating point temperatures to be programmed with 1°C resolution.

Operating Point Registers

Reg. 0x33, Remote 1 operating point = 0xA4 (100°C default)

Reg. 0x34, local operating point = 0xA4 (100°C default)

Reg. 0x35, Remote 2 operating point = 0xA4 (100°C default)

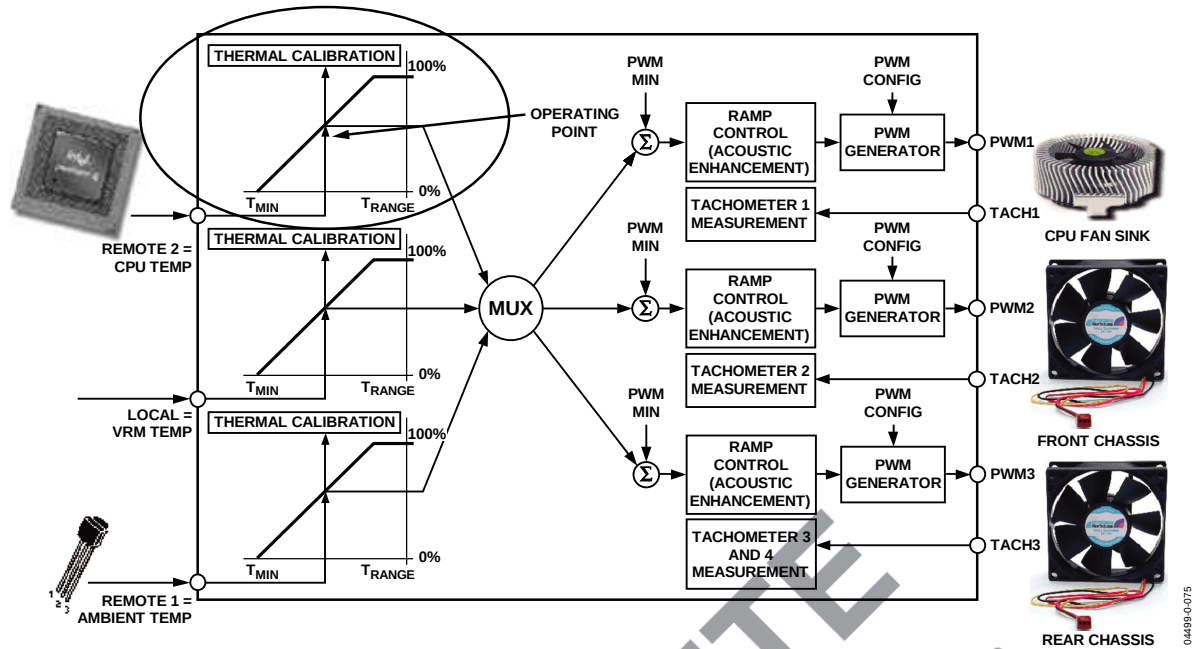


Figure 70. Operating Point Value Dynamically Adjusts Automatic Fan Control Settings

STEP 10: HIGH AND LOW LIMITS FOR TEMPERATURE CHANNELS

The low limit defines the temperature at which the T_{MIN} value starts to be increased, if temperature falls below this value. This has the net effect of reducing the fan speed, allowing the system to get hotter. An interrupt can be generated when the temperature drops below the low limit.

The high limit defines the temperature at which the T_{MIN} value starts to be reduced, if temperature increases above this value. This has the net effect of increasing fan speed to cool down the system. An interrupt can be generated when the temperature rises above the high limit.

Programming High and Low Limits

There are six limit registers; a high limit and low limit are associated with each temperature channel. These 8-bit registers allow the high and low limit temperatures to be programmed with 1°C resolution.

Temperature Limit Registers

- Reg. 0x4E, Remote 1 temperature low limit = 0x01
- Reg. 0x4F, Remote 1 temperature high limit = 0x7F
- Reg. 0x50, Local temperature low limit = 0x01
- Reg. 0x51, Local temperature high limit = 0x7F
- Reg. 0x52, Remote 2 temperature low limit = 0x01
- Reg. 0x53, Remote 2 temperature high limit = 0x7F

How Dynamic T_{MIN} Control Works

The basic premise is as follows:

1. Set the target temperature for the temperature zone, which could be, for example, the Remote 1 thermal diode. This value is programmed to the Remote 1 operating temperature register.
2. As the temperature in that zone (Remote 1 temperature) rises toward and exceeds the operating point temperature, T_{MIN} is reduced and the fan speed increases.
3. As the temperature drops below the operating point temperature, T_{MIN} is increased and the fan speed is reduced.

However, the loop operation is not as simple as described in these steps. A number of conditions govern the situations in which T_{MIN} can increase or decrease.

Short Cycle and Long Cycle

The ADT7468 implements two loops: a short cycle and a long cycle. The short cycle takes place every n monitoring cycles. The long cycle takes place every $2n$ monitoring cycles. The value of n is programmable for each temperature channel. The bits are located at the following register locations:

Remote 1 = CYR1 = Bits <2:0> of calibration control Register 2 (Address = 0x37).

Local = CYL = Bits <5:3> of Calibration Control Register 2
(Address = 0x37).

Remote 2 = CYR2 = Bits <7:6> of Calibration Control Register 2
and Bit 0 of Calibration Control Register 1 (Address = 0x36).

Table 16. Cycle Bit Assignments

Code	Short Cycle	Duration (sec)	Long Cycle	Duration (sec)
000	8 cycles	1	16 cycles	2
001	16 cycles	2	32 cycles	4
010	32 cycles	4	64 cycles	8
011	64 cycles	8	128 cycles	16
100	128 cycles	16	256 cycles	32
101	256 cycles	32	512 cycles	64
110	512 cycles	64	1024 cycles	128
111	1024 cycles	128	2048 cycles	256

Care should be taken when choosing the cycle time. A long cycle time means that T_{MIN} is updated less often. If your system has very fast temperature transients, the dynamic T_{MIN} control loop will always be lagging. If you choose a cycle time that is too fast, the full benefit of changing T_{MIN} might not be realized and would need to change again on the next cycle. In effect, it is overshooting. It is necessary to carry out some calibration to identify the most suitable response time.

Figure 71 shows the steps taken during the short cycle.

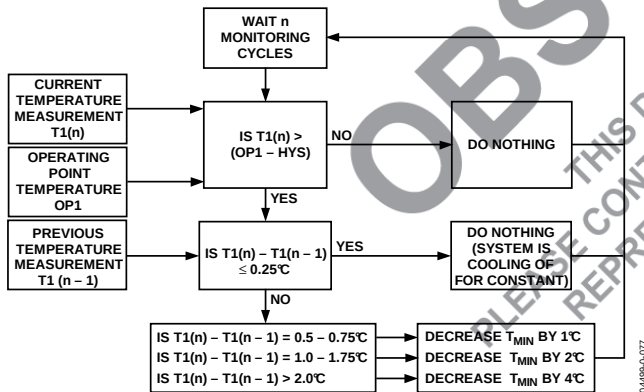


Figure 71. Short Cycle Steps

Figure 72 shows the steps taken during the long cycle.

The following examples illustrate some of the circumstances that might cause T_{MIN} to increase, decrease, or stay the same.

Normal Operation—No T_{MIN} Adjustment

1. If measured temperature never exceeds the programmed operating point minus the hysteresis temperature, then T_{MIN} is not adjusted, that is, remains at its current setting.
2. If measured temperature never drops below the low temperature limit, then T_{MIN} is not adjusted.

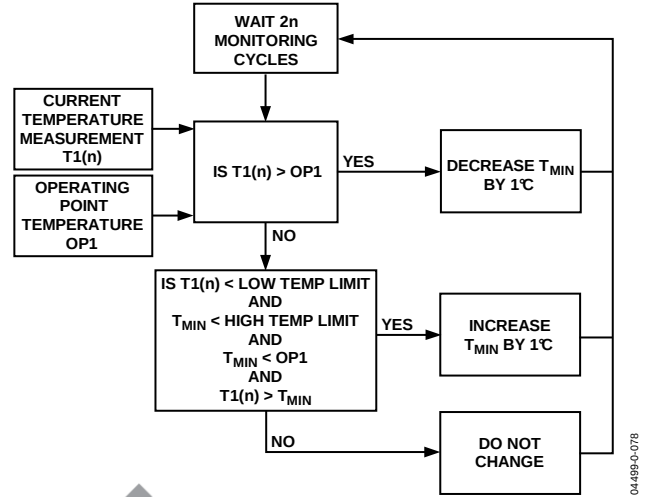


Figure 72. Long Cycle Steps

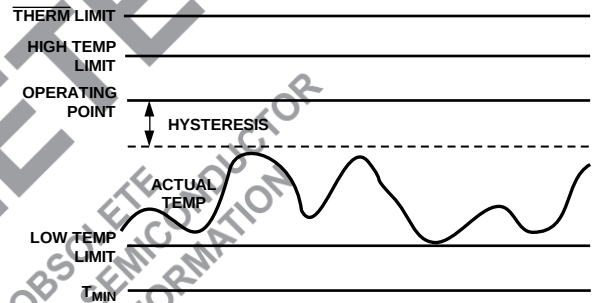


Figure 73. Temperature Between Operating Point and Low Temperature Limit

Because neither the operating point minus the hysteresis temperature nor the low temperature limit has been exceeded, the T_{MIN} value is not adjusted, and the fan runs at a speed determined by the fixed T_{MIN} and T_{RANGE} values defined in the automatic fan speed control mode.

Operating Point Exceeded— T_{MIN} Reduced

When the measured temperature is below the operating point temperature minus the hysteresis, T_{MIN} remains the same.

Once the temperature exceeds the operating temperature minus the hysteresis ($OP - Hyst$), T_{MIN} starts to decrease. This occurs during the short cycle (see Figure 71). The rate at which T_{MIN} decreases depends on the programmed value of n . It also depends on how much the temperature has increased between this monitoring cycle and the last monitoring cycle, that is, if the temperature has increased by 1°C , then T_{MIN} is reduced by 2°C . Decreasing T_{MIN} has the effect of increasing the fan speed, thus providing more cooling to the system.

If the temperature is slowly increasing only in the range ($OP - Hyst$), that is, $\leq 0.25^\circ\text{C}$ per short monitoring cycle, then T_{MIN} does not decrease. This allows small changes in temperature in the desired operating zone without changing T_{MIN} . The long cycle makes no change to T_{MIN} in the temperature range ($OP - Hyst$), because the temperature has not exceeded the operating temperature.

Once the temperature exceeds the operating temperature, the long cycle causes T_{MIN} to be reduced by 1°C every long cycle while the temperature remains above the operating temperature. This takes place in addition to the decrease in T_{MIN} that would occur due to the short cycle. In Figure 74, because the temperature is increasing at a rate $\leq 0.25^{\circ}\text{C}$ per short cycle, no reduction in T_{MIN} takes place during the short cycle.

Once the temperature has fallen below the operating temperature, T_{MIN} stays the same. Even when the temperature starts to increase slowly, T_{MIN} stays the same, because the temperature increases at a rate $\leq 0.25^{\circ}\text{C}$ per cycle.

Increasing the T_{MIN} Cycle

When the temperature drops below the low temperature limit, T_{MIN} can increase in the long cycle. Increasing T_{MIN} has the effect of running the fan slower and, therefore, quieter. The long cycle diagram in Figure 74 shows the conditions that need to be true for T_{MIN} to increase. Here is a quick summary of those conditions and the reasons they need to be true.

T_{MIN} can increase if

- The measured temperature has fallen below the low temperature limit. This means the user must choose the low limit carefully. It should not be so low that the temperature never falls below it, because T_{MIN} would never increase and the fans would run faster than necessary.
- T_{MIN} is below the high temperature limit. T_{MIN} is never allowed to increase above the high temperature limit. As a result, the high limit should be carefully chosen, because it determines how high T_{MIN} can go.
- T_{MIN} is below the operating point temperature. T_{MIN} should never be allowed to increase above the operating point temperature, because the fans would not switch on until the temperature rose above the operating point.
- The temperature is above T_{MIN} . The dynamic T_{MIN} control is turned off below T_{MIN} .

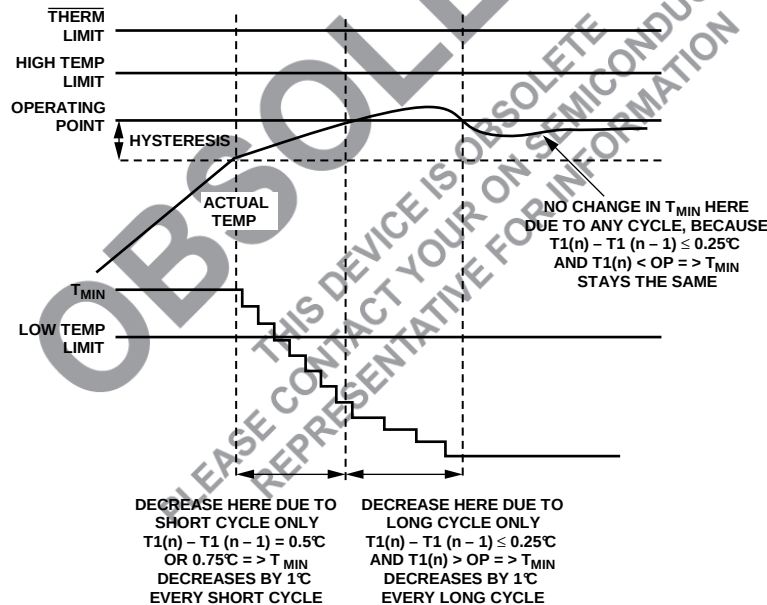


Figure 74. Effect of Exceeding Operating Point Minus Hysteresis Temperature

Figure 75 shows how T_{MIN} increases when the current temperature is above T_{MIN} and below the low temperature limit, and T_{MIN} is below the high temperature limit and below the operating point. Once the temperature rises above the low temperature limit, T_{MIN} stays the same.

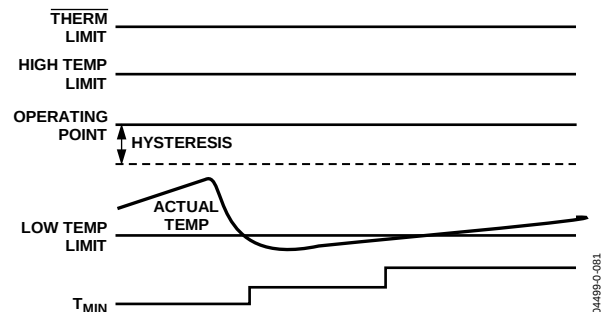


Figure 75. Increasing T_{MIN} for Quieter Operation

Preventing T_{MIN} from Reaching Full Scale

Because T_{MIN} is dynamically adjusted, it is undesirable for T_{MIN} to reach full scale (127°C), because the fan would never switch on. As a result, T_{MIN} is allowed to vary only within a specified range:

1. The lowest possible value for T_{MIN} is -127°C (twos complement mode) or -64°C (Offset 64 mode).
2. T_{MIN} cannot exceed the high temperature limit.
3. If the temperature is below T_{MIN} , the fan is switched off or is running at minimum speed, and dynamic T_{MIN} control is disabled.

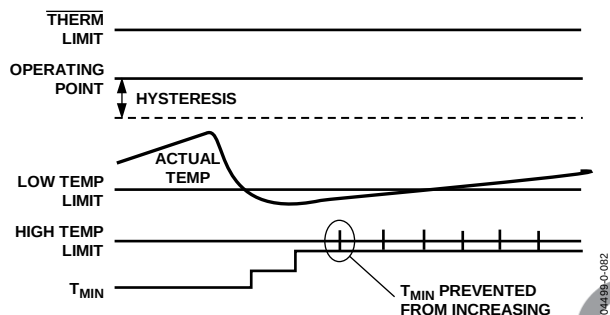


Figure 76. T_{MIN} Adjustments Limited by the High Temperature Limit

STEP 11: MONITORING THERM

Using the operating point limit ensures that the dynamic T_{MIN} control mode is operating in the best possible acoustic position while ensuring that the temperature never exceeds the maximum operating temperature. Using the operating point limit allows T_{MIN} to be independent of system level issues because of its self-corrective nature. In PC design, the operating point for the chassis is usually the worst-case internal chassis temperature.

The optimal operating point for the processor is determined by monitoring the thermal monitor in the Intel Pentium 4 processor. To do this, the PROCHOT output of the Pentium 4 is connected to the THERM input of the ADT7468.

The operating point for the processor can be determined by allowing the current temperature to be copied to the operating point register when the PROCHOT output pulls the THERM input low on the ADT7468. This gives the maximum temperature at which the Pentium 4 can run before clock modulation occurs.

Enabling the THERM Trip Point as the Operating Point

Bits <4:2> of dynamic T_{MIN} control Register 1 (Reg. 0x36) enable/disable THERM monitoring to program the operating point.

Dynamic T_{MIN} Control Register 1 (0x36)

<2> PHTR2 = 1, copies the Remote 2 current temperature to the Remote 2 operating point register, if THERM is asserted.

The operating point contains the temperature at which THERM is asserted. This allows the system to run as quietly as possible without affecting system performance.

PHTR2 = 0, ignores any THERM assertions. The Remote 2 operating point register reflects its programmed value.

<3> PHTL = 1, copies the local current temperature to the local temperature operating point register, if THERM is asserted. The operating point contains the temperature at which THERM is asserted. This allows the system to run as quietly as possible without affecting system performance.

PHTL = 0, ignores any THERM assertions. The local temperature operating point register reflects its programmed value.

<4> PHTR1 = 1, copies the Remote 1 current temperature to the Remote 1 operating point register, if THERM is asserted. The operating point contains the temperature at which THERM is asserted. This allows the system to run as quietly as possible without affecting system performance.

PHTR1 = 0, ignores any THERM assertions. The Remote 1 operating point register reflects its programmed value.

Enabling Dynamic T_{MIN} Control Mode

Bits <7:5> of dynamic T_{MIN} control Register 1 (Reg. 0x36) enable/disable dynamic T_{MIN} control on the temperature channels.

Dynamic T_{MIN} Control Register 1 (0x36)

<5> R2T = 1, enables dynamic T_{MIN} control on the Remote 2 temperature channel. The chosen T_{MIN} value is dynamically adjusted based on the current temperature, operating point, and high and low limits for this zone.

R2T = 0, disables dynamic T_{MIN} control. The T_{MIN} value chosen is not adjusted and the channel behaves as described in the Automatic Fan Control Overview section.

<6> LT = 1, enables dynamic T_{MIN} control on the local temperature channel. The chosen T_{MIN} value is dynamically adjusted based on the current temperature, operating point, and high and low limits for this zone.

LT = 0, disables dynamic T_{MIN} control. The T_{MIN} value chosen is not adjusted and the channel behaves as described in the Automatic Fan Control Overview section.

<7> R1T = 1, enables dynamic T_{MIN} control on the Remote 1 temperature channel. The chosen T_{MIN} value is dynamically adjusted based on the current temperature, operating point, and high and low limits for this zone.

R1T = 0, disables dynamic T_{MIN} control. The T_{MIN} value chosen is not adjusted and the channel behaves as described in the Automatic Fan Control Overview section.

STEP 12: RAMP RATE FOR ACOUSTIC ENHANCEMENT

The optimal ramp rate for acoustic enhancement can be found through system characterization after the thermal optimization has been finished. The effect of each ramp rate should be logged, if possible, to determine the best setting for a given solution.

Enhanced Acoustics Register 1 (Reg. 0x62)

<2:0> ACOU, selects the ramp rate for PWM1.

- 000 = 1 time slot = 35 sec
- 001 = 2 time slots = 17.6 sec
- 010 = 3 time slots = 11.8 sec
- 011 = 5 time slots = 7 sec
- 100 = 8 time slots = 4.4 sec
- 101 = 12 time slots = 3 sec
- 110 = 24 time slots = 1.6 sec
- 111 = 48 time slots = 0.8 sec

Enhance Acoustics Register 2 (Reg. 0x63)

<2:0> ACOU3, selects the ramp rate for PWM3.

- 000 = 1 time slot = 35 sec
- 001 = 2 time slots = 17.6 sec
- 010 = 3 time slots = 11.8 sec
- 011 = 5 time slots = 7 sec
- 100 = 8 time slots = 4.4 sec
- 101 = 12 time slots = 3 sec
- 110 = 24 time slots = 1.6 sec
- 111 = 48 time slots = 0.8 sec

<6:4> ACOU2, selects the ramp rate for PWM2.

- 000 = 1 time slot = 35 sec
- 001 = 2 time slots = 17.6 sec
- 010 = 3 time slots = 11.8 sec
- 011 = 5 time slots = 7 sec
- 100 = 8 time slots = 4.4 sec
- 101 = 12 time slots = 3 sec
- 110 = 24 time slots = 1.6 sec
- 111 = 48 time slots = 0.8 sec

Another way to view the ramp rates is to measure the time it takes for the PWM output to ramp up from 0% to 100% duty cycle for an instantaneous change in temperature. This can be tested by putting the ADT7468 into manual mode and changing the PWM output from 0% to 100% PWM duty cycle. The PWM output takes 35 secs to reach 100%, when a ramp rate of one time slot is selected.

Figure 77 shows remote temperature plotted against PWM duty cycle for enhanced acoustics mode. The ramp rate is set to 48, which corresponds to the fastest ramp rate. Assume that a new temperature reading is available every 115 ms. With these

settings, it takes approximately 0.76 secs to go from 33% duty cycle to 100% duty cycle (full speed). Even though the temperature increases very rapidly, the fan ramps up to full speed gradually.

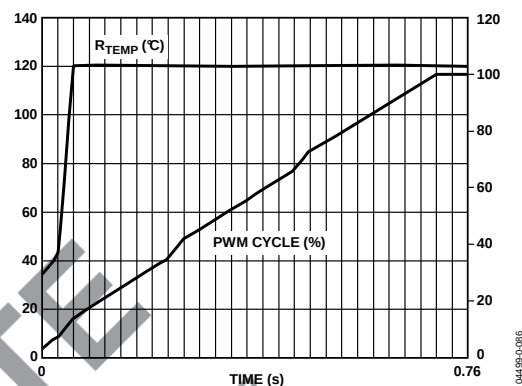


Figure 77. Enhanced Acoustics Mode with Ramp Rate = 48

Figure 78 shows how changing the ramp rate from 48 to 8 affects the control loop. The overall response of the fan is slower. Because the ramp rate is reduced, it takes longer for the fan to achieve full running speed. In this case, it takes approximately 4.4 sec for the fan to reach full speed.

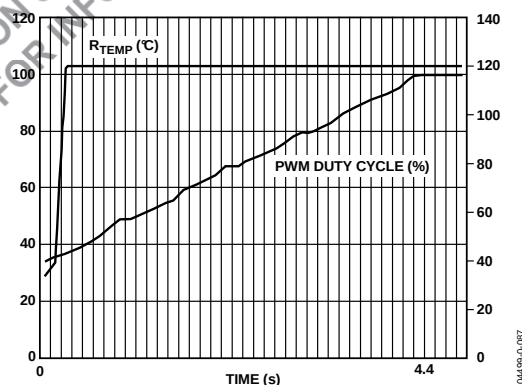


Figure 78. Enhanced Acoustics Mode with Ramp Rate = 8

Figure 79 shows the PWM output response for a ramp rate of 2. In this instance, the fan took about 17.6 sec to reach full running speed.

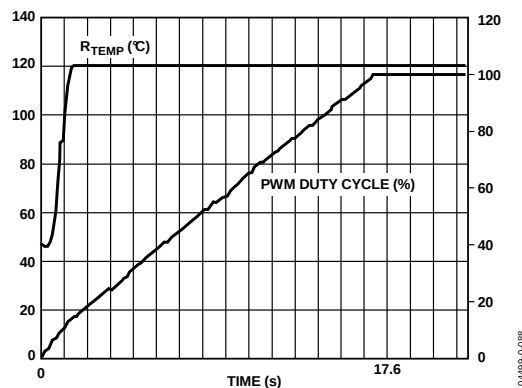


Figure 79. Enhanced Acoustics Mode with Ramp Rate = 2

Figure 80 shows how the control loop reacts to temperature with the slowest ramp rate. The ramp rate is set to 1, while all other control parameters remain the same. With the slowest ramp rate selected, it takes 35 sec for the fan to reach full speed.

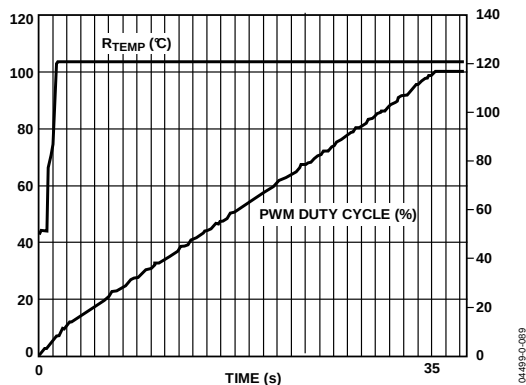


Figure 80. Enhanced Acoustics Mode with Ramp Rate = 1

As Figure 77 to Figure 80 show, the rate at which the fan reacts to temperature change is dependent on the ramp rate selected in the enhanced acoustics registers. The higher the ramp rate, the faster the fan reaches the newly calculated fan speed.

Figure 81 shows the behavior of the PWM output as temperature varies. As the temperature increases, the fan speed ramps up. Small drops in temperature do not affect the ramp-up function, because the newly calculated fan speed is still higher than the previous PWM value. Enhanced acoustics mode allows the PWM output to be made less sensitive to temperature variations. This is dependent on the ramp rate selected and programmed into the enhanced acoustics registers.

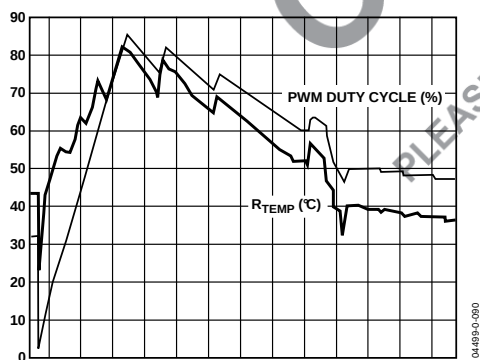


Figure 81. How Fan Reacts to Temperature Variation in Enhanced Acoustics Mode

Slower Ramp Rates

The ADT7468 can be programmed for much longer ramp times by slowing the ramp rates. Each ramp rate can be slowed by a factor of 4.

PWM1 Configuration Register (Reg. 0x5C)

<3> slow, 1 slows the ramp rate for PWM1 by 4.

PWM2 Configuration Register (Reg. 0x5D)

<3> slow, 1 slows the ramp rate for PWM2 by 4.

PWM3 Configuration Register (Reg. 0x5E)

<3> slow, 1 slows the ramp rate for PWM3 by 4.

The following sections list the ramp-up times when the slow bit is set for each PWM output.

Enhanced Acoustics Register 1 (Reg. 0x62)

<2:0> ACOU, selects the ramp rate for PWM1.

000 = 140 sec

001 = 70.4 sec

010 = 47.2 sec

011 = 28 sec

100 = 17.6 sec

101 = 12 sec

110 = 6.4 sec

111 = 3.2 sec

Enhance Acoustics Register 2 (Reg. 0x63)

<2:0> ACOU3, selects the ramp rate for PWM3.

000 = 140 sec

001 = 70.4 sec

010 = 47.2 sec

011 = 28 sec

100 = 17.6 sec

101 = 12 sec

110 = 6.4 sec

111 = 3.2 sec

<6:4> ACOU2, selects the ramp rate for PWM2.

000 = 140 sec

001 = 70.4 sec

010 = 47.2 sec

011 = 28 sec

100 = 17.6 sec

101 = 12 sec

110 = 6.4 sec

111 = 3.2 sec

ENHANCING SYSTEM ACOUSTICS

Automatic fan speed control mode reacts instantaneously to changes in temperature, that is, the PWM duty cycle responds immediately to temperature change. Any impulses in temperature can cause an impulse in fan noise. For psycho-acoustic reasons, the ADT7468 can prevent the PWM output from reacting instantaneously to temperature changes. Enhanced acoustic mode controls the maximum change in PWM duty cycle at a given time. The objective is to prevent the fan from cycling up and down, annoying the user.

ACOUSTIC ENHANCEMENT MODE OVERVIEW

Figure 82 gives a top-level overview of the automatic fan control circuitry on the ADT7468 and shows where acoustic enhancement fits in. Acoustic enhancement is intended as a postdesign tweak made by a system or mechanical engineer evaluating best settings for the system. Having determined the optimal settings for the thermal solution, the engineer can adjust the system acoustics. The goal is to implement a system that is acoustically acceptable without causing user annoyance due to fan cycling. It is important to realize that although a system might pass an acoustic noise requirement specification (for example, 36 dB), if the fan is annoying, it fails the consumer test.

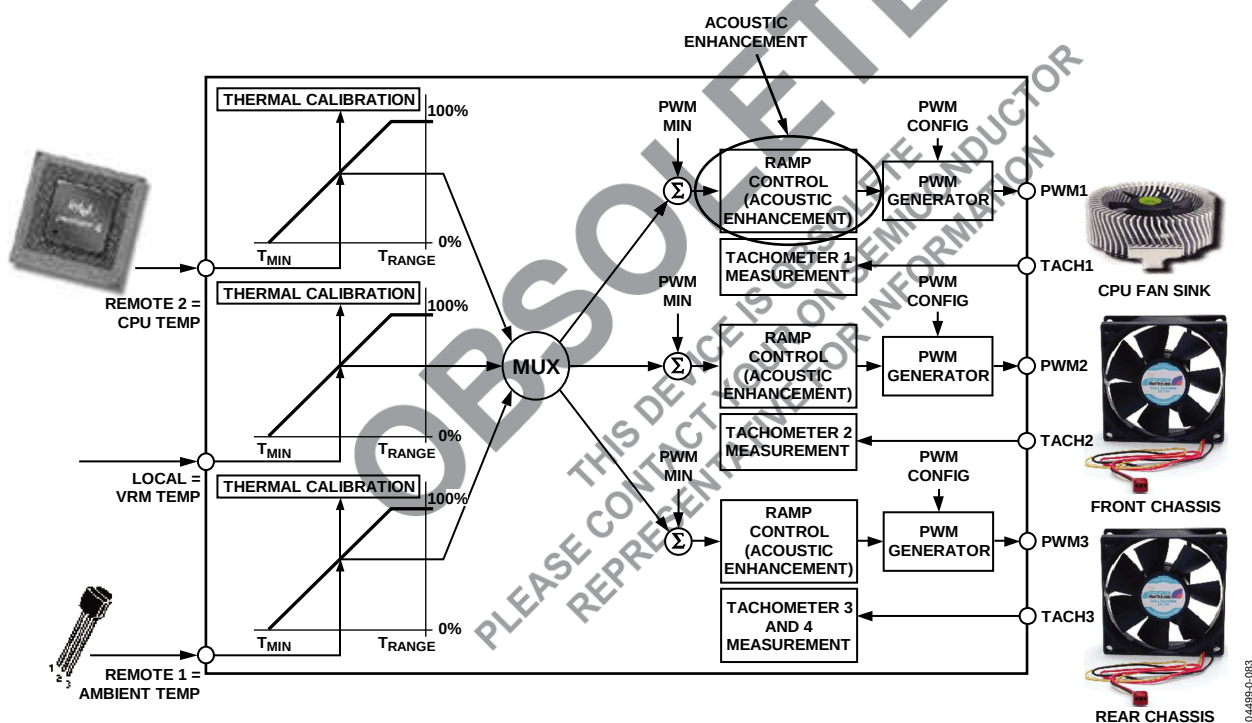


Figure 82. Acoustic Enhancement Smoothes Fan Speed Variations under Automatic Fan Speed Control

Approaches to System Acoustic Enhancement

There are two different approaches to implementing system acoustic enhancement: temperature-centric and fan-centric. The ADT7468 uses the fan-centric approach.

The temperature-centric approach involves smoothing transient temperatures as they are measured by a temperature source (for example, Remote 1 temperature). The temperature values used to calculate the PWM duty cycle values are smoothed, reducing fan speed variation. However, this approach causes an inherent delay in updating fan speed and causes the thermal characteristics of the system to change. It also causes the system fans to stay on longer than necessary, because the fan's reaction is merely delayed. The user has no control over noise from different fans driven by the same temperature source. Consider, for example, a system in which control of a CPU cooler fan (on PWM1) and a chassis fan (on PWM2) use Remote 1 temperature. Because the Remote 1 temperature is smoothed, both fans are updated at exactly the same rate. If the chassis fan is much louder than the CPU fan, there is no way to improve its acoustics without changing the thermal solution of the CPU cooling fan.

The fan-centric approach to system acoustic enhancement controls the PWM duty cycle, driving the fan at a fixed rate (for example, 6%). Each time the PWM duty cycle is updated, it is incremented by a fixed 6%. As a result, the fan ramps smoothly to its newly calculated speed. If the temperature starts to drop, the PWM duty cycle immediately decreases by 6% at every update. Therefore, the fan ramps smoothly up or down without inherent system delay. Consider, for example, controlling the same CPU cooler fan (on PWM1) and chassis fan (on PWM2) using Remote 1 temperature. The T_{MIN} and T_{RANGE} settings have already been defined in automatic fan speed control mode, that is, thermal characterization of the control loop has been optimized. Now the chassis fan is noisier than the CPU cooling fan. Using the fan-centric approach, PWM2 can be placed into acoustic enhancement mode independently of PWM1. The acoustics of the chassis fan can, therefore, be adjusted without affecting the acoustic behavior of the CPU cooling fan, although both fans are controlled by Remote 1 temperature.

Enabling Acoustic Enhancement for Each PWM Output

Enhance Acoustics Register 1 (Reg. 0x62)

<3> = 1, enables acoustic enhancement on PWM1 output

Enhance Acoustics Register 2 (Reg. 0x63)

<7> = 1, enables acoustic enhancement on PWM2 output

<3> = 1, enables acoustic enhancement on PWM3 output

Effect of Ramp Rate on Enhanced Acoustics Mode

The PWM signal driving the fan has a period, T , given by the PWM drive frequency, f , because $T = 1/f$. For a given PWM

period, T , the PWM period is subdivided into 255 equal time slots. One time slot corresponds to the smallest possible increment in the PWM duty cycle. A PWM signal of 33% duty cycle is, therefore, high for $1/3 \times 255$ time slots and low for $2/3 \times 255$ time slots. Therefore, a 33% PWM duty cycle corresponds to a signal that is high for 85 time slots and low for 170 time slots.

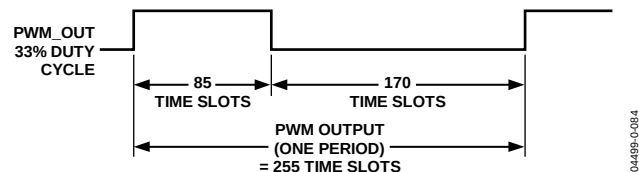


Figure 83. 33% PWM Duty Cycle Represented in Time Slots

The ramp rates in the enhanced acoustics mode are selectable from the values 1, 2, 3, 5, 8, 12, 24, and 48. The ramp rates are discrete time slots. For example, if the ramp rate is 8, then eight time slots are added to the PWM high duty cycle each time the PWM duty cycle needs to be increased. If the PWM duty cycle value needs to be decreased, it is decreased by eight time slots. Figure 84 shows how the enhanced acoustics mode algorithm operates.

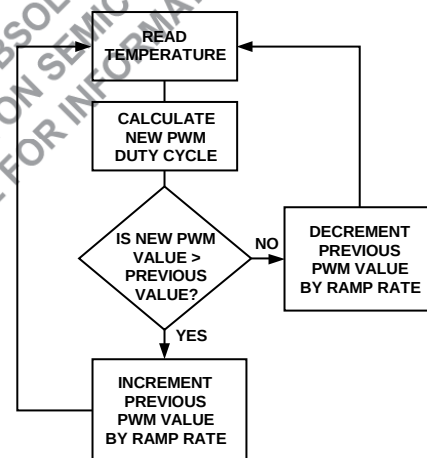


Figure 84. Enhanced Acoustics Algorithm

The enhanced acoustics mode algorithm calculates a new PWM duty cycle based on the temperature measured. If the new PWM duty cycle value is greater than the previous PWM value, then the previous PWM duty cycle value is incremented by either 1, 2, 3, 5, 8, 12, 24, or 48 time slots, depending on the settings of the enhance acoustics registers. If the new PWM duty cycle value is less than the previous PWM value, then the previous PWM duty cycle is decremented by 1, 2, 3, 5, 8, 12, 24, or 48 time slots. Each time the PWM duty cycle is incremented or decremented, its value is stored as the previous PWM duty cycle for the next comparison.

A ramp rate of 1 corresponds to one time slot, which is $1/255$ of the PWM period. In enhanced acoustics mode, incrementing or decrementing by 1 changes the PWM output by $1/255 \times 100\%$.

REGISTER TABLES

Table 17. Register Map

Address	R/W	Description	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Default	Lockable
0x20	R	2.5 V measurement	9	8	7	6	5	4	3	2	0x00	No
0x21	R	V _{CCP} measurement	9	8	7	6	5	4	3	2	0x00	No
0x22	R	V _{CC} measurement	9	8	7	6	5	4	3	2	0x00	No
0x23	R	5 V measurement	9	8	7	6	5	4	3	2	0x00	No
0x24	R	12 V measurement	9	8	7	6	5	4	3	2	0x00	No
0x25	R	Remote 1 Temperature	9	8	7	6	5	4	3	2	0x01	No
0x26	R	Local Temperature	9	8	7	6	5	4	3	2	0x01	No
0x27	R	Remote 2 Temperature	9	8	7	6	5	4	3	2	0x01	No
0x28	R	TACH1 Low Byte	7	6	5	4	3	2	1	0	0x00	No
0x29	R	TACH1 High Byte	15	14	13	12	11	10	9	8	0x00	No
0x2A	R	TACH2 Low Byte	7	6	5	4	3	2	1	0	0x00	No
0x2B	R	TACH2 High Byte	15	14	13	12	11	10	9	8	0x00	No
0x2C	R	TACH3 Low Byte	7	6	5	4	3	2	1	0	0x00	No
0x2D	R	TACH3 High Byte	15	14	13	12	11	10	9	8	0x00	No
0x2E	R	TACH4 Low Byte	7	6	5	4	3	2	1	0	0x00	No
0x2F	R	TACH4 High Byte	15	14	13	12	11	10	9	8	0x00	No
0x30	R/W	PWM1 Current Duty Cycle	7	6	5	4	3	2	1	0	0x00	No
0x31	R/W	PWM2 Current Duty Cycle	7	6	5	4	3	2	1	0	0x00	No
0x32	R/W	PWM3 Current Duty Cycle	7	6	5	4	3	2	1	0	0x00	No
0x33	R/W	Remote 1 Operating Point	7	6	5	4	3	2	1	0	0xA4	Yes
0x34	R/W	Local Temp Operating Point	7	6	5	4	3	2	1	0	0xA4	Yes
0x35	R/W	Remote 2 Operating Point	7	6	5	4	3	2	1	0	0xA4	Yes
0x36	R/W	Dynamic T _{MIN} Control Reg. 1	R2T	LT	R1T	PHTR2	PHTL	PHTR1	V _{CCP} LO	CYR2	0x00	Yes
0x37	R/W	Dynamic T _{MIN} Control Reg. 2	CYR2	CYR2	CYL	CYL	CYL	CYR1	CYR1	CYR1	0x00	Yes
0x38	R/W	Max PWM 1 Duty Cycle	7	6	5	4	3	2	1	0	0xFF	Yes
0x39	R/W	Max PWM 2 Duty Cycle	7	6	5	4	3	2	1	0	0xFF	Yes
0x3A	R/W	Max PWM 3 Duty Cycle	7	6	5	4	3	2	1	0	0xFF	Yes
0x3D	R	Device ID Register	7	6	5	4	3	2	1	0	0x68	No
0x3E	R	Company ID Number	7	6	5	4	3	2	1	0	0x41	No
0x3F	R	Revision Number	VER	VER	VER	VER	STP	STP	STP	STP	0x71/0x72	No
0x40	R/W	Configuration 1	V _{CC} = 5 V	TODIS	FSPDIS	Vxl	FSPD	RDY	LOCK	STRT	0x01	Yes
0x41	R	Interrupt Status 1	OOL	R2T	LT	R1T	5 V/THERM Timer Limit	V _{CC}	V _{CCP}	2.5 V	0x00	No
0x42	R	Interrupt Status 2	D2 FAULT	D1 FAULT	FAN4/THERM	FAN3	FAN2	FAN1	THERM Temp Limit	12 V/VC	0x00	No
0x43	R/W	VID Config	VIDSEL	THLD	VID 5	VID 4	VID 3	VID 2	VID 1	VID 0	0x00	No
0x44	R/W	2.5 V Low Limit	7	6	5	4	3	2	1	0	0x00	No
0x45	R/W	2.5 V High Limit	7	6	5	4	3	2	1	0	0xFF	No
0x46	R/W	V _{CCP} Low Limit	7	6	5	4	3	2	1	0	0x00	No
0x47	R/W	V _{CCP} High Limit	7	6	5	4	3	2	1	0	0xFF	No

ADT7468

Address	R/W	Description	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Default	Lockable
0x48	R/W	V _{CC} Low Limit	7	6	5	4	3	2	1	0	0x00	No
0x49	R/W	V _{CC} High Limit	7	6	5	4	3	2	1	0	0xFF	No
0x4A	R/W	5 V Low Limit	7	6	5	4	3	2	1	0	0x00	No
0x4B	R/W	5 V High Limit	7	6	5	4	3	2	1	0	0xFF	No
0x4C	R/W	12 V Low Limit	7	6	5	4	3	2	1	0	0x00	No
0x4D	R/W	12 V High Limit	7	6	5	4	3	2	1	0	0xFF	No
0x4E	R/W	Remote 1 Temp Low Limit	7	6	5	4	3	2	1	0	0x01	No
0x4F	R/W	Remote 1 Temp High Limit	7	6	5	4	3	2	1	0	0x7F	No
0x50	R/W	Local Temp Low Limit	7	6	5	4	3	2	1	0	0x01	No
0x51	R/W	Local Temp High Limit	7	6	5	4	3	2	1	0	0x7F	No
0x52	R/W	Remote 2 Temp Low Limit	7	6	5	4	3	2	1	0	0x01	No
0x53	R/W	Remote 2 Temp High Limit	7	6	5	4	3	2	1	0	0x7F	No
0x54	R/W	TACH1 Minimum Low Byte	7	6	5	4	3	2	1	0	0xFF	No
0x55	R/W	TACH1 Minimum High Byte	15	14	13	12	11	10	9	8	0xFF	No
0x56	R/W	TACH2 Minimum Low Byte	7	6	5	4	3	2	1	0	0xFF	No
0x57	R/W	TACH2 Minimum High Byte	15	14	13	12	11	10	9	8	0xFF	No
0x58	R/W	TACH3 Minimum Low Byte	7	6	5	4	3	2	1	0	0xFF	No
0x59	R/W	TACH3 Minimum High Byte	15	14	13	12	11	10	9	8	0xFF	No
0x5A	R/W	TACH4 Minimum Low Byte	7	6	5	4	3	2	1	0	0xFF	No
0x5B	R/W	TACH4 Minimum High Byte	15	14	13	12	11	10	9	8	0xFF	No
0x5C	R/W	PWM1 Configuration Register	BHVR	BHVR	BHVR	INV	SLOW	SPIN	SPIN	SPIN	0x82	Yes
0x5D	R/W	PWM2 Configuration Register	BHVR	BHVR	BHVR	INV	SLOW	SPIN	SPIN	SPIN	0x82	Yes
0x5E	R/W	PWM3 Configuration Register	BHVR	BHVR	BHVR	INV	SLOW	SPIN	SPIN	SPIN	0x82	Yes
0x5F	R/W	Remote 1 T _{RANGE} /PWM 1 Frequency	RANGE	RANGE	RANGE	RANGE	THRM	FREQ	FREQ	FREQ	0XC4	Yes
0x60	R/W	Local T _{RANGE} /PWM 2 Frequency	RANGE	RANGE	RANGE	RANGE	THRM	FREQ	FREQ	FREQ	0XC4	Yes
0x61	R/W	Remote 2 T _{RANGE} /PWM3 Frequency	RANGE	RANGE	RANGE	RANGE	THRM	FREQ	FREQ	FREQ	0XC4	Yes
0x62	R/W	Enhance Acoustics Reg. 1	MIN3	MIN2	MIN1	SYNC	EN1	ACOU	ACOU	ACOU	0X00	Yes
0x63	R/W	Enhance Acoustics Reg. 2	EN2	ACOU2	ACOU2	ACOU2	EN3	ACOU3	ACOU3	ACOU3	0X00	Yes
0x64	R/W	PWM1 Min Duty Cycle	7	6	5	4	3	2	1	0	0X80	Yes
0x65	R/W	PWM2 Min Duty Cycle	7	6	5	4	3	2	1	0	0X80	Yes
0x66	R/W	PWM3 Min Duty Cycle	7	6	5	4	3	2	1	0	0X80	Yes
0x67	R/W	Remote 1 Temp T _{MIN}	7	6	5	4	3	2	1	0	0X9A	Yes
0x68	R/W	Local Temp T _{MIN}	7	6	5	4	3	2	1	0	0X9A	Yes
0x69	R/W	Remote 2 Temp T _{MIN}	7	6	5	4	3	2	1	0	0X9A	Yes

Address	R/W	Description	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Default	Lockable
0x6A	R/W	Remote 1 THERM Temp Limit	7	6	5	4	3	2	1	0	0xA4	Yes
0x6B	R/W	Local THERM Temp Limit	7	6	5	4	3	2	1	0	0xA4	Yes
0x6C	R/W	Remote 2 THERM Temp Limit	7	6	5	4	3	2	1	0	0xA4	Yes
0x6D	R/W	Remote 1 and Local Temp/T _{MIN} Hysteresis	HYSR1	HYSR1	HYSR1	HYSR1	HYSL	HYSL	HYSL	HYSL	0X44	Yes
0x6E	R/W	Remote 2 Temp/T _{MIN} Hysteresis	HYSR2	HYSR2	HYSR2	HYRS	RES	RES	RES	RES	0X40	Yes
0x6F	R/W	XNOR Tree Test Enable	RES	RES	RES	RES	RES	RES	RES	XEN	0X00	Yes
0x70	R/W	Remote 1 Temperature Offset	7	6	5	4	3	2	1	0	0X00	Yes
0x71	R/W	Local Temperature Offset	7	6	5	4	3	2	1	0	0X00	Yes
0x72	R/W	Remote 2 Temperature Offset	7	6	5	4	3	2	1	0	0X00	Yes
0x73	R/W	Configuration Register 2	SHDN	CONV	ATTN	AVG	AIN4	AIN3	AIN2	AIN1	0X00	Yes
0x74	R/W	Interrupt Mask 1 Register	OOL	R2T	LT	RIT	5 V	V _{CC}	V _{CCP}	2.5 V	0X00	No
0x75	R/W	Interrupt Mask 2 Register	D2	D1	F4P	FAN3	FAN2	FAN1	OVT	12 V / VC	0X00	No
0x76	R/W	Extended Resolution 1	5 V	5 V	V _{CC}	V _{CC}	V _{CCP}	V _{CCP}	2.5 V	2.5 V	0X00	No
0x77	R/W	Extended Resolution 2	TDM2	TDM2	LTMP	LTMP	TDM1	TDM1	12 V	12 V	0X00	No
0x78	R/W	Configuration Register 3	DC4	DC3	DC2	DC1	FAST	BOOST	THERM	ALERT Enable	0X00	Yes
0x79	R	THERM Timer Status Register	TMR	TMR	TMR	TMR	TMR	TMR	TMR	ASRT/TMRO	0X00	No
0x7A	R/W	THERM Timer Limit Register	LIMIT	LIMIT	LIMIT	LIMIT	LIMIT	LIMIT	LIMIT	LIMIT	0X00	No
0x7B	R/W	TACH Pulses per Revolution	FAN4	FAN4	FAN3	FAN3	FAN2	FAN2	FAN1	FAN1	0X55	No
0x7C	R/W	Configuration Register 5	RES	RES	RES	RES	GPIOP	GPIOD	LF/HF	TWOS COMPL	0X00	Yes
0x7D	R/W	Configuration Register 4	BpAtt 12 V	BpAtt 5 V	BpAtt V _{CCP}	BpAtt 2.5 V	AINL	AINL	Pin 14 Func	Pin 14 Func	0X00	Yes
0x7E	R	Test Register 1	DO NOT WRITE TO THESE REGISTERS								0X00	Yes
0x7F	R	Test Register 2	DO NOT WRITE TO THESE REGISTERS								0X00	Yes

Table 18. Voltage Reading Registers (Power-On Default = 0x00)¹

Register Address	R/W	Description
0x20	Read only	Reflects the voltage measurement at the 2.5 V input on Pin 22 (8 MSBs of reading).
0x21	Read only	Reflects the voltage measurement ² at the V _{CCP} input on Pin 23 (8 MSBs of reading).
0x22	Read only	Reflects the voltage measurement ³ at the V _{CC} input on Pin 4 (8 MSBs of reading).
0x23	Read only	Reflects the voltage measurement at the 5 V input on Pin 20 (8 MSBs of reading).
0x24	Read only	Reflects the voltage measurement at the 12 V input on Pin 21 (8 MSBs of reading).

¹ If the extended resolution bits of these readings are also being read, the extended resolution registers (Reg. 0x76, 0x77) must be read first. Once the extended resolution registers have been read, the associated MSB reading registers are frozen until read. Both the extended resolution registers and the MSB registers are frozen.

² If V_{CCP}Low (Bit 1 of the Dynamic T_{MIN} Control Register 1, 0x36) is set, V_{CCP} can control the sleep state of the ADT7468.

³ V_{CC} (Pin 4) is the supply voltage for the ADT7468.

Table 19. Temperature Reading Registers (Power-On Default = 0x01)^{1, 2, 3}

Register Address	R/W	Description
0x25	Read only	Remote 1 temperature reading ^{3, 4} (8 MSBs of reading).
0x26	Read only	Local temperature reading (8 MSBs of reading).
0x27	Read only	Remote 2 temperature reading ^{3, 4} (8 MSBs of reading).

¹ If the extended resolution bits of these readings are also being read, the extended resolution registers (Reg. 0x76, 0x77) must be read first. Once the extended resolution registers have been read, all associated MSB reading registers are frozen until read. Both the extended resolution registers and the MSB registers are frozen.

² These temperature readings can be in twos complement or Offset 64 format; this interpretation is determined by Bit 0 of Configuration Register 5 (0x7C).

³ In twos complement mode, a temperature reading of –128°C (0x80) indicates a diode fault (open or short) on that channel.

⁴ In Offset 64 mode, a temperature reading of –64°C (0x00) indicates a diode fault (open or short) on that channel.

Table 20. Fan Tachometer Reading Registers (Power-On Default = 0x00)¹

Register Address	R/W	Description
0x28	Read only	TACH1 low byte
0x29	Read only	TACH1 high byte
0x2A	Read only	TACH2 low byte
0x2B	Read only	TACH2 high byte
0x2C	Read only	TACH3 low byte
0x2D	Read only	TACH3 high byte
0x2E	Read only	TACH4 low byte
0x2F	Read only	TACH4 high byte

¹ These registers count the number of 11.11 μs periods (based on an internal 90 kHz clock) that occur between a number of consecutive fan TACH pulses (default = 2).

The number of TACH pulses used to count can be changed using the fan pulses per revolution register (Reg. 0x7B). This allows the fan speed to be accurately measured. Because a valid fan tachometer reading requires that two bytes are read, the low byte *must* be read first. Both the low and high bytes are then frozen until read. At power-on, these registers contain 0x0000 until such time as the first valid fan TACH measurement is read into these registers. This prevents false interrupts from occurring while the fans are spinning up. A count of 0xFFFF indicates that a fan is one of the following:

- Stalled or blocked (object jamming the fan).
- Failed (internal circuitry destroyed).
- Not populated. (The ADT7468 expects to see a fan connected to each TACH. If a fan is not connected to that TACH, its TACH minimum high and low bytes should be set to 0xFFFF.)
- Alternate function, for example, TACH4 reconfigured as THERM pin.
- 2-wire instead of 3-wire fan.

Table 21. Current PWM Duty Cycle Registers (Power-On Default = 0x00)¹

Register Address	R/W	Description
0x30	Read/write	PWM1 current duty cycle (0% to 100% duty cycle = 0x00 to 0xFF).
0x31	Read/write	PWM2 current duty cycle (0% to 100% duty cycle = 0x00 to 0xFF).
0x32	Read/write	PWM3 current duty cycle (0% to 100% duty cycle = 0x00 to 0xFF).

¹ These registers reflect the PWM duty cycle driving each fan at any given time. When in automatic fan speed control mode, the ADT7468 reports the PWM duty cycles back through these registers. The PWM duty cycle values vary according to temperature in automatic fan speed control mode. During fan startup, these registers report back 0x00. In software mode, the PWM duty cycle outputs can be set to any duty cycle value by writing to these registers.

Table 22. Operating Point Registers (Power-On Default = 0x64)^{1, 2, 3}

Register Address	R/W ³	Description
0x33	Read/write	Remote 1 operating point register (default = 100°C).
0x34	Read/write	Local temperature operating point register (default = 100°C).
0x35	Read/write	Remote 2 operating point register (default = 100°C).

¹ These registers set the target operating point for each temperature channel when the dynamic T_{MIN} control feature is enabled.

² The fans being controlled are adjusted to maintain temperature about an operating point.

³ These registers become read-only when the Configuration Register 1 lock bit is set to 1. Any subsequent attempts to write to these registers fail.

Table 23. Register 0x36—Dynamic T_{MIN} Control Register 1 (Power-On Default = 0x00)¹

Bit	Name	R/W	Description
<0>	CYR2	Read/write	MSB of 3-bit remote 2 cycle value. The other two bits of the code reside in Dynamic T _{MIN} Control Register 2 (Reg. 0x37). These three bits define the delay time between making subsequent T _{MIN} adjustments in the control loop, in terms of the number of monitoring cycles. The system has associated thermal time constants that need to be found to optimize the response of fans and the control loop.
<1>	V _{CCP} LO	Read/write	V _{CCP} LO = 1. When the power is supplied from 3.3 V STANDBY and the core voltage (V _{CCP}) drops below its V _{CCP} low limit value (Reg. 0x46), the following occurs: • Status Bit 1 in Status Register 1 is set. • SMBALERT is generated, if enabled. • PROCHOT monitoring is disabled. • Dynamic T_{MIN} control is disabled. • The device is prevented from entering shutdown. • Everything is re-enabled once V_{CCP} increases above the V_{CCP} low limit.
<2>	PHTR1	Read/write	PHTR1 = 1 copies the Remote 1 current temperature to the Remote 1 operating point register, if THERM is asserted. The operating point contains the temperature at which THERM is asserted, which allows the system to run as quietly as possible without affecting system performance. PHTR1 = 0 ignores any THERM assertions on the THERM pin. The Remote 1 operating point register reflects its programmed value.
<3>	PHTL	Read/write	PHTL = 1 copies the local channel's current temperature to the local operating point register, if THERM is asserted. The operating point contains the temperature at which THERM is asserted, which allows the system to run as quietly as possible without affecting system performance. PHTL = 0 ignores any THERM assertions on the THERM pin. The local operating temperature point register reflects its programmed value.
<4>	PHTR2	Read/write	PHTR2 = 1 copies the Remote 2 current temperature to the Remote 2 operating point register, if THERM is asserted. The operating point contains the temperature at which THERM is asserted. This allows the system to run as quietly as possible without affecting system performance. PHTR2 = 0 ignores any THERM assertions on the THERM pin. The Remote 2 operating point register reflects its programmed value.
<5>	R1T	Read/write	R1T = 1 enables dynamic T _{MIN} control on the Remote 1 temperature channel. The chosen T _{MIN} value is dynamically adjusted based on the current temperature, operating point, and high and low limits for this zone. R1T = 0 disables dynamic T _{MIN} control. The T _{MIN} value chosen is not adjusted, and the channel behaves as described in the Fan Speed Control section.
<6>	LT	Read/write	LT=1 enables dynamic T _{MIN} control on the local temperature channel. The chosen T _{MIN} value is dynamically adjusted based on the current temperature, operating point, and high and low limits for this zone. LT = 0 disables dynamic T _{MIN} control. The T _{MIN} value chosen is not adjusted, and the channel behaves as described in the Fan Speed Control section.
<7>	R2T	Read/write	R2T = 1 enables dynamic T _{MIN} control on the Remote 2 temperature channel. The chosen T _{MIN} value is dynamically adjusted based on the current temperature, operating point, and high and low limits for this zone. R2T = 0 disables dynamic T _{MIN} control. The T _{MIN} value chosen is not adjusted, and the channel behaves as described in the Fan Speed Control section.

¹This register becomes read-only when the Configuration Register 1 lock bit is set to 1. Any subsequent attempts to write to this register fail.

Table 24. Register 0x37—Dynamic T_{MIN} Control Register 2 (Power-On Default = 0x00)¹

Bit	Name	R/W	Description
<2:0>	CYR1	Read/write	3-Bit Remote 1 Cycle Value. These three bits define the delay time between making subsequent T _{MIN} adjustments in the control loop for the Remote 1 channel, in terms of number of monitoring cycles. The system has associated thermal time constants that need to be found to optimize the response of fans and the control loop.
		Bits	Decrease Cycle Increase Cycle
		000	8 cycles (1 sec) 16 cycles (2 sec)
		001	16 cycles (2 sec) 32 cycles (4 sec)
		010	32 cycles (4 sec) 64 cycles (8 sec)
		011	64 cycles (8 sec) 128 cycles (16 sec)
		100	128 cycles (16 sec) 256 cycles (32 sec)
		101	256 cycles (32 sec) 512 cycles (64 sec)
		110	512 cycles (64 sec) 1024 cycles (128 sec)
		111	1024 cycles (128 sec) 2048 cycles (256 sec)
<5:3>	CYL	Read/write	3-bit local temperature cycle value. These three bits define the delay time between making subsequent T _{MIN} adjustments in the control loop for the local temperature channel, in terms of number of monitoring cycles. The system has associated thermal time constants that need to be found to optimize the response of fans and the control loop.
		Bits	Decrease Cycle Increase Cycle
		000	8 cycles (1 sec) 16 cycles (2 sec)
		001	16 cycles (2 sec) 32 cycles (4 sec)
		010	32 cycles (4 sec) 64 cycles (8 sec)
		011	64 cycles (8 sec) 128 cycles (16 sec)
		100	128 cycles (16 sec) 256 cycles (32 sec)
		101	256 cycles (32 sec) 512 cycles (64 sec)
		110	512 cycles (64 sec) 1024 cycles (128 sec)
		111	1024 cycles (128 sec) 2048 cycles (256 sec)
<7:6>	CYR2	Read/write	2 LSBs of 3-bit Remote 2 cycle value. The MSB of the 3-bit code resides in Dynamic T _{MIN} Control Register 1 (Reg. 0x36). These three bits define the delay time between making subsequent T _{MIN} adjustments in the control loop for the Remote 2 channel, in terms of number of monitoring cycles. The system has associated thermal time constants that need to be found to optimize the response of fans and the control loop.
		Bits	Decrease Cycle Increase Cycle
		000	8 cycles (1 sec) 16 cycles (2 sec)
		001	16 cycles (2 sec) 32 cycles (4 sec)
		010	32 cycles (4 sec) 64 cycles (8 sec)
		011	64 cycles (8 sec) 128 cycles (16 sec)
		100	128 cycles (16 sec) 256 cycles (32 sec)
		101	256 cycles (32 sec) 512 cycles (64 sec)
		110	512 cycles (64 sec) 1024 cycles (128 sec)
		111	1024 cycles (128 sec) 2048 cycles (256 sec)

¹This register becomes read-only when the Configuration Register 1 lock bit is set to 1. Any subsequent attempts to write to this register fail.

Table 25. Maximim PWM Duty Cycle (Power-On Default = 0xFF)^{1, 2}

Register Address	R/W ²	Description
0x38	Read/write	Maximum duty cycle for PWM1 output, default = 100% (0xFF).
0x39	Read/write	Maximum duty cycle for PWM2 output, default = 100% (0xFF).
0x3A	Read/write	Maximum duty cycle for PWM3 output, default = 100% (0xFF).

¹These registers set the maximum PWM duty cycle of the PWM output.

²This register becomes read-only when the Configuration Register 1 lock bit is set to 1. Any subsequent attempts to write to this register fail.

Table 26. Register 0x40—Configuration Register 1 (Power-On Default = 0x01)

Bit	Name	R/W	Description
<0>	STRT	Read/write	Logic 1 enables monitoring and PWM control outputs based on the limit settings programmed. Logic 0 disables monitoring and PWM control based on the default power-up limit settings. Note that the limit values programmed are preserved even if a Logic 0 is written to this bit and the default settings are enabled. This bit becomes read-only and cannot be changed once Bit 1 (LOCK bit) has been written. All limit registers should be programmed by BIOS before setting this bit to 1. (Lockable.)
<1>	LOCK	Write once	Logic 1 locks all limit values to their current settings. Once this bit is set, all lockable registers become read-only and cannot be modified until the ADT7468 is powered down and powered up again. This prevents rogue programs such as viruses from modifying critical system limit settings. (Lockable.)
<2>	RDY	Read-only	This bit is set to 1 by the ADT7468 to indicate only that the device is fully powered-up and ready to begin system monitoring.
<3>	FSPD	Read/write	When set to 1, this bit runs all fans at full speed. Power-on default = 0. This bit never gets locked.
<4>	Vxl	Read/write	BIOS should set this bit to a 1 when the ADT7468 is configured to measure current from an ADI ADOPT™ VRM controller and to measure the CPU's core voltage. This bit allows monitoring software to display CPU watts usage. (lockable.)
<5>	FSPDIS	Read/write	Logic 1 disables fan spin-up for two TACH pulses. Instead, the PWM outputs go high for the entire fan spin-up timeout selected.
<6>	TODIS	Read/write	When this bit is set to 1, the SMBus timeout feature is enabled. This allows the ADT7468 to be used with SMBus controllers that cannot handle SMBus timeouts. (lockable.)
<7>	V _{CC}	Read/write	When this bit is set to 1, the ADT7468 rescales its V _{CC} pin to measure 5 V supply. If this bit is 0, the ADT7468 measures V _{CC} as a 3.3 V supply. (lockable.)

Table 27. Register 0x41—Interrupt Status Register 1 (Power-On Default = 0x00)

Bit	Name	R/W	Description
<0>	2.5 V	Read only	2.5 V = 1 indicates that the 2.5 V high or low limit has been exceeded. This bit is cleared on a read of the status register only if the error condition has subsided.
<1>	V _{CCP}	Read only	V _{CCP} = 1 indicates that the V _{CCP} high or low limit has been exceeded. This bit is cleared on a read of the status register only if the error condition has subsided.
<2>	V _{CC}	Read only	V _{CC} = 1 indicates that the V _{CC} high or low limit has been exceeded. This bit is cleared on a read of the status register only if the error condition has subsided.
<3>	5 V/ THERM timer	Read only	A one indicates the 5 V high or low limit has been exceeded. This bit is cleared on a read of the status register only if the error condition has subsided. If Pin 20 is configured as THERM, this bit is asserted when the timer limit has been exceeded.
<4>	R1T	Read only	R1T = 1 indicates that the Remote 1 low or high temperature has been exceeded. This bit is cleared on a read of the status register only if the error condition has subsided.
<5>	LT	Read only	LT = 1 indicates that the local low or high temperature has been exceeded. This bit is cleared on a read of the status register only if the error condition has subsided.
<6>	R2T	Read only	R2T = 1 indicates that the Remote 2 low or high temperature has been exceeded. This bit is cleared on a read of the status register only if the error condition has subsided.
<7>	OOL	Read only	OOL = 1 indicates that an out-of-limit event has been latched in Status Register 2. This bit is a logical OR of all status bits in Status Register 2. Software can test this bit in isolation to determine whether any of the voltage, temperature, or fan speed readings represented by Status Register 2 are out-of-limit, which saves the need to read Status Register 2 during every interrupt or polling cycle.

Table 28. Register 0x42—Interrupt Status Register 2 (Power-On Default = 0x00)

Bit	Name	R/W	Description
<0>	12V/VC	Read only	A one indicates that the 12 V high or low limit has been exceeded. This bit is cleared on a read of the status register only if the error condition has subsided. If Pin 21 is configured as VID5, this bit is the VID change bit. This bit is set when the levels on VID0 to VID5 are different than they were 11 μ s previously. This pin can be used to generate an <u>SMBALERT</u> whenever the VID code changes.
<1>	OVT	Read only	OVT = 1 indicates that one of the <u>THERM</u> overtemperature limits has been exceeded. This bit is cleared on a read of the status register when the temperature drops below <u>THERM</u> – T_{HYST} .
<2>	FAN1	Read only	FAN1 = 1 indicates that Fan 1 has dropped below minimum speed or has stalled. This bit is not set when the PWM 1 output is off.
<3>	FAN2	Read only	FAN2 = 1 indicates that Fan 2 has dropped below minimum speed or has stalled. This bit is not set when the PWM 2 output is off.
<4>	FAN3	Read only	FAN3 = 1 indicates that Fan 3 has dropped below minimum speed or has stalled. This bit is not set when the PWM 3 output is off.
<5>	F4P	Read only Read/write Read only	When Pin 14 is programmed as a TACH4 input, F4P = 1 indicates that Fan 4 has dropped below minimum speed or has stalled. This bit is <i>not</i> set when the PWM3 output is off. When Pin 14 is programmed as a GPIO output, writing to this bit determines the logic output of the GPIO. If Pin 14 is configured as the <u>THERM</u> timer input for <u>THERM</u> monitoring, then this bit is set when the <u>THERM</u> assertion time exceeds the limit programmed in the <u>THERM</u> limit register (Reg. 0x7A).
<6>	D1	Read only	D1 = 1 indicates either an open or short circuit on the Thermal Diode 1 inputs.
<7>	D2	Read only	D2 = 1 indicates either an open or short circuit on the Thermal Diode 2 inputs.

Table 29. Register 0x43-VID Register (Power-On Default = 0x00)

Bit	Name	R/W	Description
<4:0>	VID[4:0]	Read-only	The VID[4:0] inputs from the CPU indicate the expected processor core voltage. On power-up, these bits reflect the state of the VID pins, even if monitoring is not enabled.
<5>	VID5	Read-only	Reads VID5 from the CPU when Bit 7 = 1. If Bit 7 = 0, then the VID5 bit always reads back 0 (power-on default).
<6>	THLD	Read/write	Selects the input switching threshold for the VID inputs. THLD = 0 selects a threshold of 1 V ($V_{OL} < 0.8$ V, $V_{IH} > 1.7$ V). THLD = 1 lowers the switching threshold to 0.6 V ($V_{OL} < 0.4$ V, $V_{IH} > 0.8$ V).
<7>	VIDSEL	Read/write	VIDSEL = 0 configures Pin 21 as the 12 V measurement input (default).

Table 30. Voltage Limit Registers¹

Register Address	R/W	Description ²	Power-On Default
0x44	Read/write	2.5 V	0x00
0x45	Read/write	2.5 V	0xFF
0x46	Read/write	V_{CCP} low limit	0x00
0x47	Read/write	V_{CCP} high limit	0xFF
0x48	Read/write	V_{CC} low limit	0x00
0x49	Read/write	V_{CC} high limit	0xFF
0x4A	Read/write	5 V low limit	0x00
0x4B	Read/write	5 V high limit	0xFF
0x4C	Read/write	12 V low limit	0x00
0x4D	Read/write	12 V high limit	0xFF

¹ Setting the Configuration Register 1 lock bit has no effect on these registers.² High Limits: An interrupt is generated when a value exceeds its high limit (>comparison). Low limits: An interrupt is generated when a value is equal to or below its low limit ($\leq$ comparison).

Table 31. Temperature Limit Registers¹

Register Address	R/W	Description ²	Power-On Default
0x4E	Read/write	Remote 1 temperature low limit	0x81
0x4F	Read/write	Remote 1 temperature high limit	0x7F
0x50	Read/write	Local temperature low limit	0x81
0x51	Read/write	Local temperature high limit	0x7F
0x52	Read/write	Remote 2 temperature low limit	0x81
0x53	Read/write	Remote 2 temperature high limit	0x7F

¹ Exceeding any of these temperature limits by 1°C causes the appropriate status bit to be set in the interrupt status register. Setting the Configuration Register 1 lock bit has no effect on these registers.

² High Limits: An interrupt is generated when a value exceeds its high limit (> comparison). Low Limits: An interrupt is generated when a value is equal to or below its low limit (≤ comparison).

Table 32. Fan Tachometer Limit Registers¹

Register Address	R/W	Description	Power-On Default
0x54	Read/write	TACH1 minimum low byte	0xFF
0x55	Read/write	TACH1 minimum high byte/Single channel ADC channel select	0xFF
0x56	Read/write	TACH2 minimum low byte	0xFF
0x57	Read/write	TACH2 minimum high byte	0xFF
0x58	Read/write	TACH3 minimum low byte	0xFF
0x59	Read/write	TACH3 minimum high byte	0xFF
0x5A	Read/write	TACH4 minimum low byte	0xFF
0x5B	Read/write	TACH4 minimum high byte	0xFF

¹ Exceeding any of the TACH limit registers by 1 indicates that the fan is running too slowly or has stalled. The appropriate status bit is set in Interrupt Status Register 2 to indicate the fan failure. Setting the Configuration Register 1 lock bit has no effect on these registers.

Table 33. Register 0x55—TACH 1 Minimum High Byte (Power-On Default = 0xFF)

Bits	Name	R/W	Description
<4:0>	Reserved	Read only	These bits are reserved when Bit 6 of Configuration 2 Register (0x73) is set (single-channel ADC mode). Otherwise, these bits represent Bits <4:0> of the TACH1 minimum high byte.
<7:5>	SCADC	Read/write	When Bit 6 of Config 2 Register (0x73) is set (single channel ADC mode), these bits are used to select the only channel from which the ADC makes measurements. Otherwise, these bits represent Bits <7:5> of the TACH1 minimum high byte.

Table 34. PWM Configuration Registers

Register Address		R/W ¹	Description	Power-On Default
0x5C		Read/write	PWM1 configuration.	0x82
0x5D		Read/write	PWM2 configuration.	0x82
0x5E		Read/write	PWM3 configuration.	0x82
Bit	Name	R/W	Description	
<2:0>	SPIN	Read/write	These bits control the startup timeout for PWMx. The PWM output stays high until two valid TACH rising edges are seen from the fan. If there is not a valid TACH signal during the fan TACH measurement directly after the fan startup timeout period, then the TACH measurement reads 0xFFFF and Status Register 2 reflects the fan fault. If the TACH minimum high and low bytes contain 0xFFFF or 0x0000, then the Status Register 2 bit is not set, even if the fan has not started. 000 = No startup timeout 001 = 100 ms 010 = 250 ms (default) 011 = 400 ms 100 = 667 ms 101 = 1 sec 110 = 2 secs 111 = 4 secs	
<3>	SLOW	Read/write	SLOW = 1 makes the ramp rates for acoustic enhancement four times longer.	
<4>	INV	Read/write	This bit inverts the PWM output. The default is 0, which corresponds to a logic high output for 100% duty cycle. Setting this bit to 1 inverts the PWM output, so 100% duty cycle corresponds to a logic low output.	
<7:5>	BHVR	Read/write	These bits assign each fan to a particular temperature sensor for localized cooling. 000 = Remote 1 temperature controls PWMx (automatic fan control mode). 001 = local temperature controls PWMx (automatic fan control mode). 010 = Remote 2 temperature controls PWMx (automatic fan control mode). 011 = PWMx runs full speed. 100 = PWMx disabled (default). 101 = fastest speed calculated by local and Remote 2 temperature controls PWMx. 110 = fastest speed calculated by all three temperature channel controls PWMx. 111 = manual mode. PWM duty cycle registers (Reg. 0x30 to Reg. 0x32) become writable.	

¹ These registers become read-only when the Configuration Register 1 lock bit is set to 1. Any subsequent attempts to write to these registers fail.

Table 35. TEMP T_{RANGE}/PWM Frequency Registers

Register Address		R/W ¹	Description	Power-On Default
0x5F		Read/write	Remote 1 T _{RANGE} /PWM1 frequency.	0xC4
0x60		Read/write	Local temperature T _{RANGE} /PWM2 frequency.	0xC4
0x61		Read/write	Remote 2 T _{RANGE} /PWM3 frequency.	0xC4
Bit	Name	R/W	Description	
<2:0>	FREQ	Read/write	These bits control the PWMx frequency. 000 = 11.0 Hz 001 = 14.7 Hz 010 = 22.1 Hz 011 = 29.4 Hz 100 = 35.3 Hz (default) 101 = 44.1 Hz 110 = 58.8 Hz 111 = 88.2 Hz	
<3>	THRM	Read/write	THRM = 1 causes the <u>THERM</u> pin (Pin 14/20) to assert low as an output when this temperature channel's <u>THERM</u> limit has been exceeded by 0.25°C. The <u>THERM</u> pin remains asserted until the temperature is equal to or below the <u>THERM</u> limit. The minimum time that <u>THERM</u> asserts is one monitoring cycle. This allows clock modulation of devices that incorporate this feature. THRM = 0 makes the <u>THERM</u> pin act as an input only, for example, for Pentium 4 PROCHOT monitoring, when Pin 14/20 is configured as <u>THERM</u> .	
<7:4>	RANGE	Read/write	These bits determine the PWM duty cycle vs. the temperature slope for automatic fan control. 0000 = 2°C 0001 = 2.5°C 0010 = 3.33°C 0011 = 4°C 0100 = 5°C 0101 = 6.67°C 0110 = 8°C 0111 = 10°C 1000 = 13.33°C 1001 = 16°C 1010 = 20°C 1011 = 26.67°C 1100 = 32°C (Default) 1101 = 40°C 1110 = 53.33°C 1111 = 80°C	

¹These registers become read-only when the Configuration Register 1 lock bit is set. Any further attempts to write to these registers have no effect.

Table 36. Register 0x62—Enhanced Acoustics Register 1 (Power-On Default = 0x00)

Bit	Name	R/W ¹	Description	
<2:0>	ACOU	Read/write	These bits select the ramp rate applied to the PWM1 output. Instead of PWM1 jumping instantaneously to its newly calculated speed, PWM1 ramps gracefully at the rate determined by these bits. This feature enhances the acoustics of the fan being driven by the PWM1 output.	
			Time Slot Increase	Time for 33% to 100%
			000 = 1	35 sec
			001 = 2	17.6 sec
			010 = 3	11.8 sec
			011 = 4	7 sec
			100 = 8	4.4 sec
			101 = 12	3 sec
110 = 24	1.6 sec			
111 = 48	0.8 sec			
<3>	EN1	Read/write	When this bit is 1, acoustic enhancement is enabled on PWM1 output.	
<4>	SYNC	Read/write	SYNC = 1 synchronizes fan speed measurements on TACH2, TACH3, and TACH4 to PWM3. This allows up to three fans to be driven from PWM3 output and their speeds to be measured. SYNC = 0 synchronizes only TACH3 and TACH4 to PWM3 output.	
<5>	MIN1	Read/write	When the ADT7468 is in automatic fan control mode, this bit defines whether PWM1 is off (0% duty cycle) or at PWM1 minimum duty cycle when the controlling temperature is below its T _{MIN} – hysteresis value. 0 = 0% duty cycle below T _{MIN} – hysteresis. 1 = PWM1 minimum duty cycle below T _{MIN} – hysteresis.	
<6>	MIN2	Read/write	When the ADT7468 is in automatic fan speed control mode, this bit defines whether PWM2 is off (0% duty cycle) or at PWM2 minimum duty cycle when the controlling temperature is below its T _{MIN} – hysteresis value. 0 = 0% duty cycle below T _{MIN} – hysteresis. 1 = PWM 2 minimum duty cycle below T _{MIN} – hysteresis.	
<7>	MIN3	Read/write	When the ADT7468 is in automatic fan speed control mode, this bit defines whether PWM3 is off (0% duty cycle) or at PWM3 minimum duty cycle when the controlling temperature is below its T _{MIN} – hysteresis value. 0 = 0% duty cycle below T _{MIN} – hysteresis. 1 = PWM3 minimum duty cycle below T _{MIN} – hysteresis.	

¹This register becomes read-only when the Configuration Register 1 lock bit is set to 1. Any further attempts to write to this register have no effect.

Table 37. Register 0x63 – Enhanced Acoustics Register 2 (Power-On Default = 0x00)

Bit	Name	R/W ¹	Description	
<2:0>	ACOU3	Read/write	These bits select the ramp rate applied to the PWM3 output. Instead of PWM3 jumping instantaneously to its newly calculated speed, PWM3 ramps gracefully at the rate determined by these bits. This effect enhances the acoustics of the fan being driven by the PWM3 output.	
			Time Slot Increase	Time for 33% to 100%
			000 = 1	35 sec
			001 = 2	17.6 sec
			010 = 3	11.8 sec
			011 = 5	7 sec
			100 = 8	4.4 sec
			101 = 12	3 sec
			110 = 24	1.6 sec
111 = 48	0.8 sec			
< 3 >	EN3	Read/write	When this bit is 1, acoustic enhancement is enabled on PWM3 output.	
<6:4>	ACOU2	Read/write	These bits select the ramp rate applied to the PWM2 output. Instead of PWM2 jumping instantaneously to its newly calculated speed, PWM2 ramps gracefully at the rate determined by these bits. This effect enhances the acoustics of the fans being driven by the PWM2 output.	
			Time Slot Increase	Time for 33% to 100%
			000 = 1	35 sec
			001 = 2	17.6 sec
			010 = 3	11.8 sec
			011 = 5	7 sec
			100 = 8	4.4 sec
			101 = 12	3 sec
			110 = 24	1.6 sec
<7>	EN2	Read/write	When this bit is 1, acoustic enhancement is enabled on PWM2 output.	

¹This register becomes read-only when the Configuration Register 1 lock bit is set to 1. Any further attempts to write to this register have no effect.

Table 38. PWM Minimum Duty Cycle Registers

Register Address	R/W ¹	Description	Power-On Default
0x64	Read/write	PWM1 minimum duty cycle.	0x80 (50% duty cycle)
0x65	Read/write	PWM2 minimum duty cycle.	0x80 (50% duty cycle)
0x66	Read/write	PWM3 minimum duty cycle.	0x80 (50% duty cycle)
Bit	Name	R/W ¹	Description
<7:0>	PWM duty cycle	Read/write	These bits define the PWM _{MIN} duty cycle for PWMx.
			0x00 = 0% duty cycle (fan off).
			0x40 = 25% duty cycle.
			0x80 = 50% duty cycle.
			0xFF = 100% duty cycle (fan full speed).

¹These registers become read-only when the ADT7468 is in automatic fan control mode.

Table 39. T_{MIN} Registers¹

Register Address	R/W ²	Description	Power-On Default
0x67	Read/write	Remote 1 temperature T _{MIN} .	0x5A (90°C)
0x68	Read/write	Local temperature T _{MIN} .	0x5A (90°C)
0x69	Read/write	Remote 2 temperature T _{MIN} .	0x5A (90°C)

¹These are the T_{MIN} registers for each temperature channel. When the temperature measured exceeds T_{MIN}, the appropriate fan runs at minimum speed and increases with temperature according to T_{RANGE}.

²These registers become read-only when the Configuration Register 1 lock bit is set. Any further attempts to write to these registers have no effect.

Table 40. THERM Limit Registers¹

Register Address	R/W ²	Description	Power-On Default
0x6A	Read/write	Remote 1 THERM limit.	0x64 (100°C)
0x6B	Read/write	Local THERM limit.	0x64 (100°C)
0x6C	Read/write	Remote 2 THERM limit.	0x64 (100°C)

¹ If any temperature measured exceeds its THERM limit, all PWM outputs drive their fans at 100% duty cycle. This is a fail-safe mechanism incorporated to cool the system in the event of a critical overtemperature. It also ensures some level of cooling in the event that software or hardware locks up. If set to 0x80, this feature is disabled. The PWM output remains at 100% until the temperature drops below THERM limit – hysteresis. If the THERM pin is programmed as an output, then exceeding these limits by 0.25°C can cause the THERM pin to assert low as an output.

² These registers become read-only when the Configuration Register 1 lock bit is set to 1. Any further attempts to write to these registers have no effect.

Table 41. Temperature/T_{MIN} Hysteresis Registers¹

Register Address	R/W ²	Description	Power-On Default
0x6D <3:0>	Read/write HYSL	Remote 1 and local temperature hysteresis. Local temperature hysteresis. 0°C to 15°C of hysteresis can be applied to the local temperature AFC and dynamic T _{MIN} control loops.	0x44
<7:4>	HYSR1	Remote 1 temperature hysteresis. 0°C to 15°C of hysteresis can be applied to the Remote 1 temperature AFC and dynamic T _{MIN} control loops.	
0x6E <7:4>	Read/write HYSR2	Remote 2 temperature hysteresis. Local temperature hysteresis. 0°C to 15°C of hysteresis can be applied to the local temperature AFC and dynamic T _{MIN} control loops.	0x40

¹ Each 4-bit value controls the amount of temperature hysteresis applied to a particular temperature channel. Once the temperature for that channel falls below its T_{MIN} value, the fan remains running at PWM_{MIN} duty cycle until the temperature = T_{MIN} – hysteresis. Up to 15°C of hysteresis can be assigned to any temperature channel. The hysteresis value chosen also applies to that temperature channel, if its THERM limit is exceeded. The PWM output being controlled goes to 100%, if the THERM limit is exceeded and remains at 100% until the temperature drops below THERM – hysteresis. For acoustic reasons, it is recommended that the hysteresis value not be programmed less than 4°C. Setting the hysteresis value lower than 4°C causes the fan to switch on and off regularly when the temperature is close to T_{MIN}.

² These registers become read-only when the Configuration Register 1 lock bit is set to 1. Any further attempts to write to these registers have no effect.

Table 42. XNOR Tree Test Enable

Register Address	R/W ¹	Description	Power-On Default
0x6F <0>	Read/write XEN	XNOR tree test enable register. If the XEN bit is set to 1, the device enters the XNOR tree test mode. Clearing the bit removes the device from the XNOR tree test mode.	0x00
<7:1>	Reserved	Unused. Do not write to these bits.	

¹ This register becomes read-only when the Configuration Register 1 lock bit is set to 1. Any further attempts to write to this register have no effect.

Table 43. Remote 1 Temperature Offset

Register Address	R/W ¹	Description	Power-On Default
0x70 <7:0>	Read/write Read/write	Remote 1 temperature offset. Allows a twos complement offset value to be automatically added to or subtracted from the Remote 1 temperature reading. This is to compensate for any inherent system offsets such as PCB trace resistance. LSB value = 0.5°C.	0x00

¹ This register becomes read-only when the Configuration Register 1 lock bit is set to 1. Any further attempts to write to this register have no effect.

Table 44. Local Temperature Offset

Register Address	R/W ¹	Description	Power-On Default
0x71 <7:0>	Read/write Read/write	Local temperature offset. Allows a twos complement offset value to be automatically added to or subtracted from the local temperature reading. LSB value = 0.5°C.	0x00

¹ This register becomes read-only when the Configuration Register 1 lock bit is set to 1. Any further attempts to write to this register have no effect.

Table 45. Remote 2 Temperature Offset

Register Address	R/W ¹	Description	Power-On Default
0x72 <7:0>	Read/write Read/write	Remote 2 temperature offset. Allows a twos complement offset value to be automatically added to or subtracted from the Remote 2 temperature reading. This is to compensate for any inherent system offsets such as PCB trace resistance. LSB value = 0.5°C.	0x00

¹ This register becomes read-only when the Configuration Register 1 lock bit is set to 1. Any further attempts to write to this register have no effect.

Table 46. Register 0x73—Configuration Register 2 (Power-On Default = 0x00)

Bit	Name	R/W ¹	Description
0	AIN1	Read/write	AIN1 = 0, speed of 3-wire fans measured using the TACH output from the fan. AIN1 = 1, Pin 6 is reconfigured to measure the speed of 2-wire fans using an external sensing resistor and coupling capacitor. AIN voltage threshold is set via Configuration Register 4 (Reg. 0x7D). Only relevant in low frequency mode.
1	AIN2	Read/write	AIN2 = 0, speed of 3-wire fans measured using the TACH output from the fan. AIN2 = 1, Pin 7 is reconfigured to measure the speed of 2-wire fans using an external sensing resistor and coupling capacitor. AIN voltage threshold is set via Configuration Register 4 (Reg. 0x7D). Only relevant in low frequency mode.
2	AIN3	Read/write	AIN3 = 0, speed of 3-wire fans measured using the TACH output from the fan. AIN3 = 1, Pin 4 is reconfigured to measure the speed of 2-wire fans using an external sensing resistor and coupling capacitor. AIN voltage threshold is set via Configuration Register 4 (Reg. 0x7D). Only relevant in low frequency mode.
3	AIN4	Read/write	AIN4 = 0, speed of 3-wire fans measured using the TACH output from the fan. AIN4 = 1, Pin 9 is reconfigured to measure the speed of 2-wire fans using an external sensing resistor and coupling capacitor. AIN voltage threshold is set via Configuration Register 4 (Reg. 0x7D). Only relevant in low frequency mode.
4	AVG	Read/write	AVG = 1, averaging on the temperature and voltage measurements is turned off. This allows measurements on each channel to be made much faster.
5	ATTN	Read/write	ATTN = 1, the ADT7468 removes the attenuators from the 2.5 V, V _{CCP} , 5 V, and 12 V inputs. These inputs can be used for other functions such as connecting up external sensors. It is possible to remove attenuators from individual channels using Bits <7:4> of Configuration Register 4 (0x7D).
6	CONV	Read/write	CONV = 1, the ADT7468 is put into a single-channel ADC conversion mode. In this mode, the ADT7468 can be made to read continuously from one input only, for example, Remote 1 temperature. The appropriate ADC channel is selected by writing to Bits <7:5> of TACH1 minimum high byte register (0x55).
			Bits <7:5> Reg. 0x55
			000 2.5 V
			001 V _{CCP}
			010 V _{CC} (3.3 V)
			011 5 V
			100 12 V
			101 Remote 1 temperature
			110 Local temperature
			111 Remote 2 temperature
7	SHDN	Read/write	SHDN = 1, ADT7468 goes into shutdown mode. All PWM outputs assert low (or high depending on state of INV bit) to switch off all fans. The PWM current duty cycle registers read 0x00 to indicate that the fans are not being driven.

¹ This register becomes read-only when the Configuration Register 1 lock bit is set to 1. Any further attempts to write to this register have no effect.

Table 47. Register 0x74—Interrupt Mask Register 1 (Power-On Default <7:0> = 0x00)

Bit	Name	R/W	Description
0	2.5V	Read/write	2.5 V = 1, masks <u>SMBALERT</u> for out-of-limit conditions on the 2.5 V channel.
1	V _{CCP}	Read/write	V _{CCP} = 1, masks <u>SMBALERT</u> for out-of-limit conditions on the V _{CCP} channel.
2	V _{CC}	Read/write	V _{CC} = 1, masks <u>SMBALERT</u> for out-of-limit conditions on the V _{CC} channel.
3	5V	Read/write	5 V = 1, masks <u>SMBALERT</u> for out-of-limit conditions on the 5 V channel.
4	RIT	Read/write	RIT = 1, masks <u>SMBALERT</u> for out-of-limit conditions on the Remote 1 temperature channel.
5	LT	Read/write	LT = 1, masks <u>SMBALERT</u> for out-of-limit conditions on the local temperature channel.
6	R2T	Read/write	R2T = 1, masks <u>SMBALERT</u> for out-of-limit conditions on the Remote 2 temperature channel.
7	OOL	Read/write	OOL = 1, masks <u>SMBALERT</u> for any out-of-limit condition in Status Register 2.

Table 48. Register 0x75—Interrupt Mask Register 2 (Power-On Default <7:0> = 0x00)

Bit	Name	R/W	Description
0	12V/VC	Read/write	When Pin 21 is configured as a 12 V input, 12V/VC = 1 masks <u>SMBALERT</u> for out-of-limit conditions on the 12 V channel. When Pin 21 is programmed as VID5, this bit masks an <u>SMBALERT</u> , if the VID5 VID code bit changes.
1	OVT	Read only	OVT = 1, masks <u>SMBALERT</u> for overtemperature <u>THERM</u> conditions.
2	FAN1	Read/write	FAN1 = 1, masks <u>SMBALERT</u> for a Fan 1 fault.
3	FAN2	Read/write	FAN2 = 1, masks <u>SMBALERT</u> for a Fan 2 fault.
4	FAN3	Read/write	FAN3 = 1, masks <u>SMBALERT</u> for a Fan 3 fault.
5	F4P	Read/write	If Pin 14 is configured as TACH 4, F4P = 1 masks <u>SMBALERT</u> for a Fan 4 fault. If Pin 14 is configured as <u>THERM</u> , F4P = 1 masks <u>SMBALERT</u> for an exceeded <u>THERM</u> timer limit. If Pin 14 is configured as GPIO, F4P = 1 masks <u>SMBALERT</u> when GPIO is an input and GPIO is asserted.
6	D1	Read/write	D1 = 1 masks <u>SMBALERT</u> for a diode open or short on a Remote 1 channel.
7	D2	Read/write	D2 = 1 masks <u>SMBALERT</u> for a diode open or short on a Remote 2 channel.

Table 49. Register 0x76—Extended Resolution Register 1¹

Bit	Name	R/W	Description
<1:0>	2.5V	Read only	2.5 V LSBs. Holds the 2 LSBs of the 10-bit 2.5 V measurement.
<3:2>	V _{CCP}	Read only	V _{CCP} LSBs. Holds the 2 LSBs of the 10-bit V _{CCP} measurement.
<5:4>	V _{CC}	Read only	V _{CC} LSBs. Holds the 2 LSBs of the 10-bit V _{CC} measurement.
<7:6>	5 V	Read only	5 V LSBs. Holds the 2 LSBs of the 10-bit 5 V measurement.

¹ If this register is read, this register and the registers holding the MSB of each reading are frozen until read.

Table 50. Register 0x77—Extended Resolution Register 2¹

Bit	Name	R/W	Description
<1:0>	12V	Read only	12 V LSBs. Holds the 2 LSBs of the 10-bit 12 V measurement.
<3:2>	TDM1	Read only	Remote 1 temperature LSBs. Holds the 2 LSBs of the 10-bit Remote 1 temperature measurement.
<5:4>	LTMP	Read only	Local temperature LSBs. Holds the 2 LSBs of the 10-bit local temperature measurement.
<7:6>	TDM2	Read only	Remote 2 temperature LSBs. Holds the 2 LSBs of the 10-bit Remote 2 temperature measurement.

¹ If this register is read, this register and the registers holding the MSB of each reading are frozen until read.

Table 51. Register 0x78—Configuration Register 3 (Power-On Default = 0x00)

Bit	Name	R/W ¹	Description
<0>	ALERT	Read/write	ALERT = 1, Pin 10 (PWM2/SMBALERT) is configured as an SMBALERT interrupt output to indicate out-of-limit error conditions. ALERT = 0, Pin 10 (PWM2/SMBALERT) is configured as the PWM2 output.
<1>	THERM	Read/write	THERM Enable = 1 enables THERM functionality on Pin 20 and Pin 14, if Pin 14 is configured as THERM, determined by Bits 0 and 1 (PIN14FUNC) of Configuration Register 4. When THERM is asserted, if the fans are running and the boost bit is set, the fans run at full speed. Alternatively, THERM can be programmed so that a timer is triggered to time how long THERM has been asserted.
<2>	BOOST	Read/write	When THERM is an input and BOOST = 1, assertion of THERM causes all fans to run at the maximum programmed duty cycle for fail-safe cooling.
<3>	FAST	Read/write	FAST = 1, enables fast TACH measurements on all channels. This increases the TACH measurement rate from once per second to once every 250 ms (4 ×).
<4>	DC1	Read/write	DC1 = 1, enables TACH measurements to be continuously made on TACH1. Fans must be driven by dc. Setting this bit prevents pulse stretching, because it is not required for dc-driven motors.
<5>	DC2	Read/write	DC2 = 1, enables TACH measurements to be continuously made on TACH2. Fans must be driven by dc. Setting this bit prevents pulse stretching, because it is not required for dc-driven motors.
<6>	DC3	Read/write	DC3 = 1, enables TACH measurements to be continuously made on TACH3. Setting this bit prevents pulse stretching, because it is not required for dc-driven motors.
<7>	DC4	Read/write	DC4 = 1, enables TACH measurements to be continuously made on TACH4. Setting this bit prevents pulse stretching, because it is not required for dc-driven motors.

¹This register becomes read-only when the Configuration Register 1 lock bit is set to 1. Any further attempts to write to this register have no effect.

Table 52. Register 0x79—THERM Timer Status Register (Power-On Default = 0x00)

Bit	Name	R/W	Description
<7:1>	TMR	Read only	Times how long THERM input is asserted. These seven bits read zero until the THERM assertion time exceeds 45.52 ms.
<0>	ASRT/ TMR0	Read only	This bit is set high on the assertion of the THERM input and is cleared on read. If the THERM assertion time exceeds 45.52 ms, this bit is set and becomes the LSB of the 8-bit TMR reading. This allows THERM assertion times from 45.52 ms to 5.82 secs to be reported back with a resolution of 22.76 ms.

Table 53. Register 0x7A—THERM Timer Limit Register (Power-On Default = 0x00)

Bit	Name	R/W	Description
<7:0>	LIMIT	Read/write	Sets maximum THERM assertion length allowed before an interrupt is generated. This is an 8-bit limit with a resolution of 22.76 ms allowing THERM assertion limits of 45.52 ms to 5.82 secs to be programmed. If the THERM assertion time exceeds this limit, Bit 5 (F4P) of Interrupt Status Register 2 (Reg. 0x42) is set. If the limit value is 0x00, then an interrupt is generated immediately on the assertion of the THERM input.

Table 54. Register 0x7B—TACH Pulses per Revolution Register (Power-On Default = 0x55)

Bit	Name	R/W	Description
<1:0>	FAN1	Read/write	Sets number of pulses to be counted when measuring Fan 1 speed. Can be used to determine fan pulses per revolution for unknown fan type.
			Pulses Counted
			00 = 1
			01 = 2 (default)
<3:2>	FAN2	Read/write	Sets number of pulses to be counted when measuring Fan 2 speed. Can be used to determine fan pulses per revolution for unknown fan type.
			Pulses Counted
			00 = 1
			01 = 2 (default)
<5:4>	FAN3	Read/write	Sets number of pulses to be counted when measuring Fan 3 speed. Can be used to determine fan pulses per revolution for unknown fan type.
			Pulses Counted
			00 = 1
			01 = 2 (default)
<7:6>	FAN4	Read/write	Sets number of pulses to be counted when measuring Fan 4 speed. Can be used to determine fan pulses per revolution for unknown fan type.
			Pulses Counted
			00 = 1
			01 = 2 (default)

Table 55. Register 0x7C—Configuration Register 5 (Power-On Default = 0x00)

Bit	Name	R/W ¹	Description
<0>	2sC	Read/write	2sC = 1, sets the temperature range to twos complement temperature range. 2sC = 0, changes the temperature range to Offset 64. When this bit is changed, the ADT7468 interprets all relevant temperature register values as defined by this bit.
<1>	HF/LF		Sets the PWM drive frequency to high frequency mode (0) or low frequency mode (1).
<2>	GPIOD		GPIO direction. When GPIO function is enabled, this determines whether the GPIO is an input (0) or an output (1).
<3>	GPIOP		GPIO polarity. When the GPIO function is enabled and is programmed as an output, this bit determines whether the GPIO is active low (0) or high (1).
<4:7>	RES		Unused.

¹This register becomes read-only when the Configuration Register 1 lock bit is set to 1. Any further attempts to write to this register have no effect.

Table 56. Register 0x7D—Configuration Register 4 (Power-On Default = 0x00)

Bit	Name	R/W ¹	Description
<1:0>	PIN14FUNC	Read/write	These bits set the functionality of Pin 14: 00 = TACH4 (default) 01 = Bidirectional $\overline{\text{THERM}}$ 10 = $\overline{\text{SMBALERT}}$ 11 = GPIO
<3:2>	AINL	Read/write	These two bits define the input threshold for 2-wire fan speed measurements (low frequency mode only): 00 = ± 20 mV 01 = ± 40 mV 10 = ± 80 mV 11 = ± 130 mV
<4>	BpAtt2.5V		Bypass 2.5 V attenuator. When set, the measurement scale for this channel changes from 0 V (0x00) to 2.2965 V (0xFF).
<5>	BpAttV _{CCP}		Bypass V _{CCP} attenuator. When set, the measurement scale for this channel changes from 0 V (0x00) to 2.2965 V (0xFF).
<6>	BpAtt5V		Bypass 5 V attenuator. When set, the measurement scale for this channel changes from 0 V (0x00) to 2.2965 V (0xFF).
<7>	BpAtt12V		Bypass 12 V attenuator. When set, the measurement scale for this channel changes from 0 V (0x00) to 2.2965 V (0xFF).

¹This register becomes read-only when the Configuration Register 1 lock bit is set to 1. Any further attempts to write to this register have no effect.

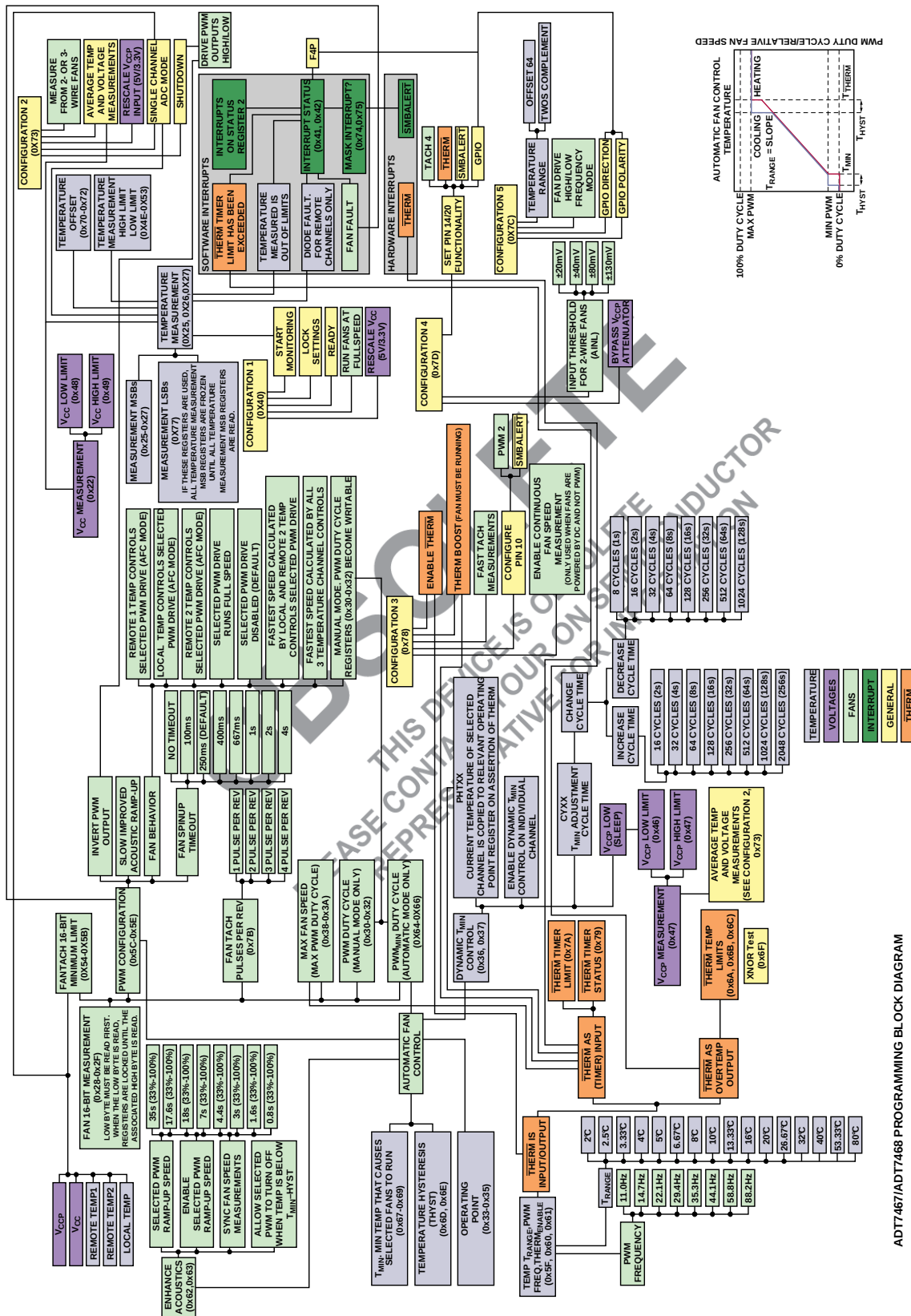
Table 57. Register 0x7E—Manufacturer's Test Register 1 (Power-On Default = 0x00)

Bit	Name	R/W	Description
<7:0>	Reserved	Read only	Manufacturer's test register. These bits are reserved for manufacturer's test purposes and should <i>not</i> be written to under normal operation.

Table 58. Register 0x7F—Manufacturer's Test Register 2 (Power-On Default = 0x00)

Bit	Name	R/W	Description
<7:0>	Reserved	Read only	Manufacturer's test register. These bits are reserved for manufacturer's test purposes and should <i>not</i> be written to under normal operation.

PROGRAMMING BLOCK DIAGRAM



ADT7468/ADT7468 PROGRAMMING BLOCK DIAGRAM

04069-1.5-044

Figure 85.

OUTLINE DIMENSIONS

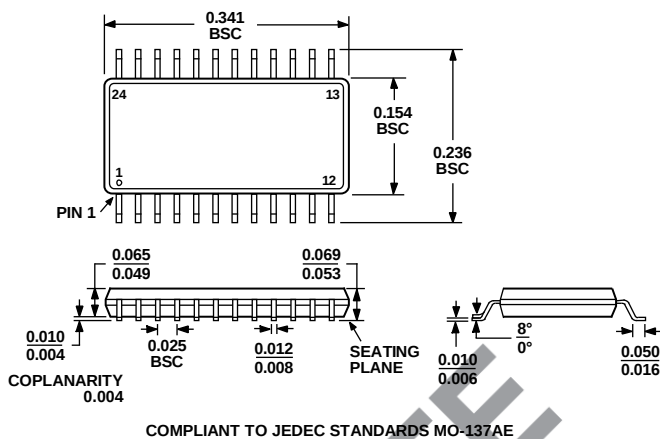


Figure 86. 24-Lead Shrink Small Outline Package [QSOP]
(RQ-24)
Dimensions shown in inches

ORDERING GUIDE

Model	Temperature Range	Package Description	Package Option
ADT7468ARQ	-40°C to +120°C	24-Lead QSOP	RQ-24
ADT7468ARQ-REEL	-40°C to +120°C	24-Lead QSOP	RQ-24
ADT7468ARQ-REEL7	-40°C to +120°C	24-Lead QSOP	RQ-24
ADT7468ARQZ ¹	-40°C to +120°C	24-Lead QSOP	RQ-24
ADT7468ARQZ-REEL ¹	-40°C to +120°C	24-Lead QSOP	RQ-24
ADT7468ARQZ-RL7 ¹	-40°C to +120°C	24-Lead QSOP	RQ-24
EVAL-ADT7468EB		Evaluation Board	

¹ Z = Pb-free part.

ON Semiconductor and  are registered trademarks of Semiconductor Components Industries, LLC (SCILLC). SCILLC reserves the right to make changes without further notice to any products herein. SCILLC makes no warranty, representation or guarantee regarding the suitability of its products for any particular purpose, nor does SCILLC assume any liability arising out of the application or use of any product or circuit, and specifically disclaims any and all liability, including without limitation special, consequential or incidental damages. "Typical" parameters which may be provided in SCILLC data sheets and/or specifications can and do vary in different applications and actual performance may vary over time. All operating parameters, including "Typicals" must be validated for each customer application by customer's technical experts. SCILLC does not convey any license under its patent rights nor the rights of others. SCILLC products are not designed, intended, or authorized for use as components in systems intended for surgical implant into the body, or other applications intended to support or sustain life, or for any other application in which the failure of the SCILLC product could create a situation where personal injury or death may occur. Should Buyer purchase or use SCILLC products for any such unintended or unauthorized application, Buyer shall indemnify and hold SCILLC and its officers, employees, subsidiaries, affiliates, and distributors harmless against all claims, costs, damages, and expenses, and reasonable attorney fees arising out of, directly or indirectly, any claim of personal injury or death associated with such unintended or unauthorized use, even if such claim alleges that SCILLC was negligent regarding the design or manufacture of the part. SCILLC is an Equal Opportunity/Affirmative Action Employer. This literature is subject to all applicable copyright laws and is not for resale in any manner.

PUBLICATION ORDERING INFORMATION

LITERATURE FULFILLMENT:

Literature Distribution Center for ON Semiconductor
P.O. Box 5163, Denver, Colorado 80217 USA
Phone: 303-675-2175 or 800-344-3860 Toll Free USA/Canada
Fax: 303-675-2176 or 800-344-3867 Toll Free USA/Canada
Email: orderlit@onsemi.com

N. American Technical Support: 800-282-9855

Toll Free USA/Canada

Europe, Middle East and Africa Technical Support:

Phone: 421 33 790 2910

Japan Customer Focus Center

Phone: 81-3-5773-3850

ON Semiconductor Website: www.onsemi.com

Order Literature: <http://www.onsemi.com/orderlit>

For additional information, please contact your local Sales Representative