

SCAN921025H and SCAN921226H High Temperature 20-80 MHz 10 Bit Bus LVDS SerDes with IEEE 1149.1 (JTAG) and at-speed BIST

Check for Samples: [SCAN921025H](#), [SCAN921226H](#)

FEATURES

- High Temperature Operation to 125°C
- IEEE 1149.1 (JTAG) Compliant and At-Speed BIST Test Mode
- Clock Recovery from PLL Lock to Random Data Patterns
- Ensured Transition Every Data Transfer Cycle
- Chipset (Tx + Rx) Power Consumption < 600 mW (Typ) @ 80 MHz
- Single Differential Pair Eliminates Multi-Channel Skew
- 800 Mbps Serial Bus LVDS Data Rate (at 80 MHz Clock)
- 10-bit Parallel Interface for 1 Byte Data Plus 2 Control Bits
- Synchronization Mode and LOCK Indicator
- Programmable Edge Trigger on Clock
- High Impedance on Receiver Inputs When Power is Off
- Bus LVDS Serial Output Rated for 27Ω Load
- Small 49-Lead NFBGA Package

APPLICATIONS

- Automotive
- Industrial
- Military/Aerospace

DESCRIPTION

The SCAN921025H transforms a 10-bit wide parallel LVCMOS/LVTTL data bus into a single high speed Bus LVDS serial data stream with embedded clock. The SCAN921226H receives the Bus LVDS serial data stream and transforms it back into a 10-bit wide parallel data bus and recovers parallel clock.

Both devices are compliant with IEEE 1149.1 Standard for Boundary Scan Test. IEEE 1149.1 features provide the design or test engineer access via a standard Test Access Port (TAP) to the backplane or cable interconnects and the ability to verify differential signal integrity. The pair of devices also features an at-speed BIST mode which allows the interconnects between the Serializer and Deserializer to be verified at-speed.

The SCAN921025H transmits data over backplanes or cable. The single differential pair data path makes PCB design easier. In addition, the reduced cable, PCB trace count, and connector size tremendously reduce cost. Since one output transmits clock and data bits serially, it eliminates clock-to-data and data-to-data skew. The powerdown pin saves power by reducing supply current when not using either device. Upon power up of the Serializer, you can choose to activate synchronization mode or allow the Deserializer to use the synchronization-to-random-data feature. By using the synchronization mode, the Deserializer will establish lock to a signal within specified lock times. In addition, the embedded clock ensures a transition on the bus every 12-bit cycle. This eliminates transmission errors due to charged cable conditions. Furthermore, you may put the SCAN921025H output pins into tri-state to achieve a high impedance state. The PLL can lock to frequencies between 20 MHz and 80 MHz.



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Block Diagram

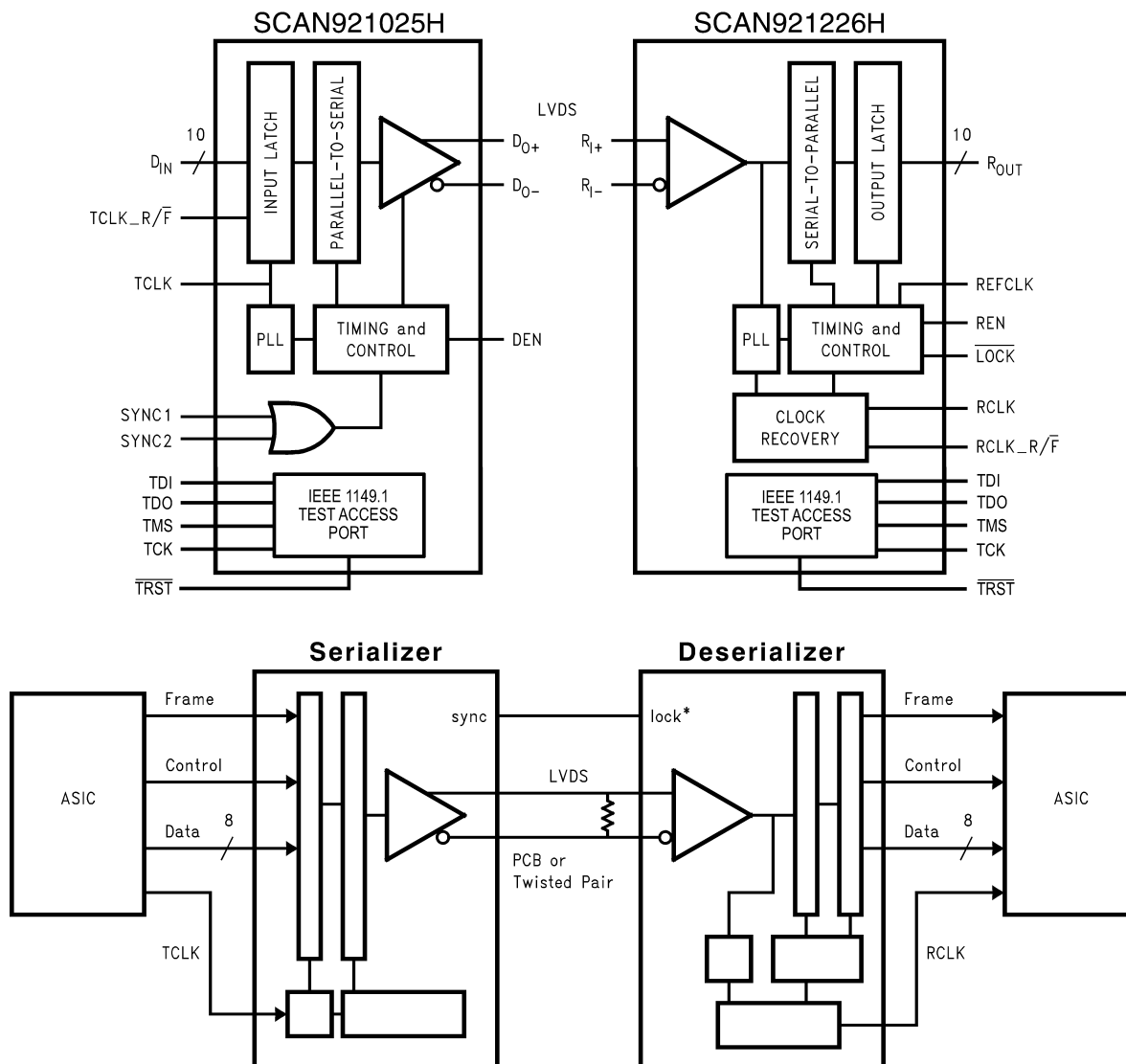


Figure 1. Application

Functional Description

The SCAN921025H and SCAN921226H are a 10-bit Serializer and Deserializer chipset designed to transmit data over differential backplanes at clock speeds from 20 to 80 MHz. The chipset is also capable of driving data over Unshielded Twisted Pair (UTP) cable.

The chipset has three active states of operation: Initialization, Data Transfer, and Resynchronization; and two passive states: Powerdown and Tri-state. In addition to the active and passive states, there are also test modes for JTAG access and at-speed BIST.

The following sections describe each operation and passive state and the test modes.

Initialization

Initialization of both devices must occur before data transmission begins. Initialization refers to synchronization of the Serializer and Deserializer PLL's to local clocks, which may be the same or separate. Afterwards, synchronization of the Deserializer to Serializer occurs.

Step 1: When you apply V_{CC} to both Serializer and/or Deserializer, the respective outputs enter tri-state, and on-chip power-on circuitry disables internal circuitry. When V_{CC} reaches V_{CCOK} (2.5V) the PLL in each device begins locking to a local clock. For the Serializer, the local clock is the transmit clock (TCLK) provided by the source ASIC or other device. For the Deserializer, you must apply a local clock to the REFCLK pin.

The Serializer outputs remain in tri-state while the PLL locks to the TCLK. After locking to TCLK, the Serializer is now ready to send data or SYNC patterns, depending on the levels of the SYNC1 and SYNC2 inputs or a data stream. The SYNC pattern sent by the Serializer consists of six ones and six zeros switching at the input clock rate.

Note that the Deserializer $\overline{LOCK}$ output will remain high while its PLL locks to the incoming data or to SYNC patterns on the input.

Step 2: The Deserializer PLL must synchronize to the Serializer to complete initialization. The Deserializer will lock to non-repetitive data patterns. However, the transmission of SYNC patterns enables the Deserializer to lock to the Serializer signal within a specified time. See [Figure 17](#).

The user's application determines control of the SYNC1 and SYNC 2 pins. One recommendation is a direct feedback loop from the $\overline{LOCK}$ pin. Under all circumstances, the Serializer stops sending SYNC patterns after both SYNC inputs return low.

When the Deserializer detects edge transitions at the Bus LVDS input, it will attempt to lock to the embedded clock information. When the Deserializer locks to the Bus LVDS clock, the $\overline{LOCK}$ output will go low. When $\overline{LOCK}$ is low, the Deserializer outputs represent incoming Bus LVDS data.

Data Transfer

After initialization, the Serializer will accept data from inputs DIN0–DIN9. The Serializer uses the TCLK input to latch incoming Data. The TCLK_R/F pin selects which edge the Serializer uses to strobe incoming data. TCLK_R/F high selects the rising edge for clocking data and low selects the falling edge. If either of the SYNC inputs is high for 5*TCLK cycles, the data at DIN0–DIN9 is ignored regardless of clock edge.

After determining which clock edge to use, a start and stop bit, appended internally, frame the data bits in the register. The start bit is always high and the stop bit is always low. The start and stop bits function as the embedded clock bits in the serial stream.

The Serializer transmits serialized data and clock bits (10+2 bits) from the serial data output ($DO\pm$) at 12 times the TCLK frequency. For example, if TCLK is 80 MHz, the serial rate is $80 \times 12 = 960$ Mega-bits-per-second. Since only 10 bits are from input data, the serial "payload" rate is 10 times the TCLK frequency. For instance, if TCLK = 80 MHz, the payload data rate is $80 \times 10 = 800$ Mbps. The data source provides TCLK and must be in the range of 20 MHz to 80 MHz nominal.

The Serializer outputs ($DO\pm$) can drive a point-to-point connection or in limited multi-point or multi-drop backplanes. The outputs transmit data when the enable pin (DEN) is high, $\overline{PWRDN}$ = high, and SYNC1 and SYNC2 are low. When DEN is driven low, the Serializer output pins will enter tri-state.

When the Deserializer synchronizes to the Serializer, the $\overline{\text{LOCK}}$ pin is low. The Deserializer locks to the embedded clock and uses it to recover the serialized data. ROUT data is valid when $\overline{\text{LOCK}}$ is low. Otherwise ROUT0–ROUT9 is invalid.

The ROUT0–ROUT9 pins use the RCLK pin as the reference to data. The polarity of the RCLK edge is controlled by the RCLK_R/F input. See [Figure 14](#).

ROUT(0-9), $\overline{\text{LOCK}}$ and RCLK outputs will drive a maximum of three CMOS input gates (15 pF load) with a 80 MHz clock.

Resynchronization

When the Deserializer PLL locks to the embedded clock edge, the Deserializer $\overline{\text{LOCK}}$ pin asserts a low. If the Deserializer loses lock, the $\overline{\text{LOCK}}$ pin output will go high and the outputs (including RCLK) will enter tri-state.

The user's system monitors the $\overline{\text{LOCK}}$ pin to detect a loss of synchronization. Upon detection, the system can arrange to pulse the Serializer SYNC1 or SYNC2 pin to resynchronize. Multiple resynchronization approaches are possible. One recommendation is to provide a feedback loop using the $\overline{\text{LOCK}}$ pin itself to control the sync request of the Serializer (SYNC1 or SYNC2). Dual SYNC pins are provided for multiple control in a multi-drop application. Sending sync patterns for resynchronization is desirable when lock times within a specific time are critical. However, the Deserializer can lock to random data, which is discussed in the next section.

Random Lock Initialization and Resynchronization

The initialization and resynchronization methods described in their respective sections are the fastest ways to establish the link between the Serializer and Deserializer. However, the SCAN921226H can attain lock to a data stream without requiring the Serializer to send special SYNC patterns. This allows the SCAN921226H to operate in "open-loop" applications. Equally important is the Deserializer's ability to support hot insertion into a running backplane. In the open loop or hot insertion case, we assume the data stream is essentially random. Therefore, because lock time varies due to data stream characteristics, we cannot possibly predict exact lock time. However, please see [Table 1](#) for some general random lock times under specific conditions. The primary constraint on the "random" lock time is the initial phase relation between the incoming data and the REFCLK when the Deserializer powers up. As described in the next paragraph, the data contained in the data stream can also affect lock time.

If a specific pattern is repetitive, the Deserializer could enter "false lock" - falsely recognizing the data pattern as the clocking bits. We refer to such a pattern as a repetitive multi-transition, RMT. This occurs when more than one Low-High transition takes place in a clock cycle over multiple cycles. This occurs when any bit, except DIN 9, is held at a low state and the adjacent bit is held high, creating a 0-1 transition. In the worst case, the Deserializer could become locked to the data pattern rather than the clock. Circuitry within the SCAN921226H can detect that the possibility of "false lock" exists. The circuitry accomplishes this by detecting more than one potential position for clocking bits. Upon detection, the circuitry will prevent the $\overline{\text{LOCK}}$ output from becoming active until the potential "false lock" pattern changes. The false lock detect circuitry expects the data will eventually change, causing the Deserializer to lose lock to the data pattern and then continue searching for clock bits in the serial data stream. Graphical representations of RMT are shown in [Figure 2](#). Please note that RMT only applies to bits DIN0–DIN8.

Powerdown

When no data transfer occurs, you can use the Powerdown state. The Serializer and Deserializer use the Powerdown state, a low power sleep mode, to reduce power consumption. The Deserializer enters Powerdown when you drive $\overline{\text{PWRDN}}$ and REN low. The Serializer enters Powerdown when you drive $\overline{\text{PWRDN}}$ low. In Powerdown, the PLL stops and the outputs enter tri-state, which disables load current and reduces supply current to the milliampere range. To exit Powerdown, you must drive the $\overline{\text{PWRDN}}$ pin high.

Before valid data exchanges between the Serializer and Deserializer, you must reinitialize and resynchronize the devices to each other. Initialization of the Serializer takes 510 TCLK cycles. The Deserializer will initialize and assert $\overline{\text{LOCK}}$ high until lock to the Bus LVDS clock occurs.

Tri-state

The Serializer enters tri-state when the DEN pin is driven low. This puts both driver output pins (DO+ and DO-) into tri-state. When you drive DEN high, the Serializer returns to the previous state, as long as all other control pins remain static (SYNC1, SYNC2, PWRDN, TCLK_R/F).

When you drive the REN pin low, the Deserializer enters tri-state. Consequently, the receiver output pins (ROUT0–ROUT9) and RCLK will enter tri-state. The LOCK output remains active, reflecting the state of the PLL.

Table 1.

Random Lock Times for the SCAN921226H ⁽¹⁾		
	80 MHz	Units
Maximum	18	μs
Mean	3.0	μs
Minimum	0.43	μs
Conditions:	PRBS 2 ¹⁵ , V _{CC} = 3.3V	

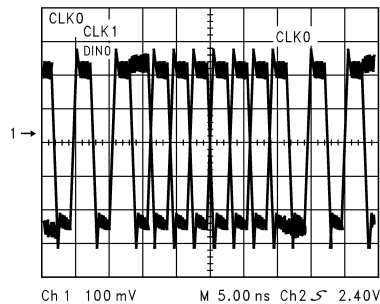
(1) Difference in lock times are due to different starting points in the data pattern with multiple parts.

Test Modes

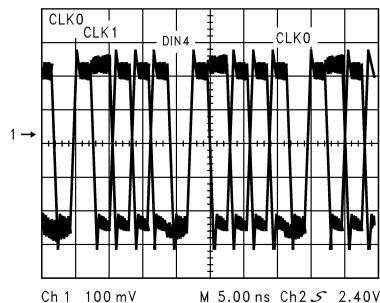
In addition to the IEEE 1149.1 test access to the digital TTL pins, the SCAN921025H and SCAN921226H have two instructions to test the LVDS interconnects. The first is EXTEST. This is implemented at LVDS levels and is only intended as a go no-go test (e.g. missing cables). The second method is the RUNBIST instruction. It is an "at-system-speed" interconnect test. It is executed in approximately 28mS with a system clock speed of 80MHz. There are two bits in the RX BIST data register for notification of PASS/FAIL and TEST_COMPLETE. Pass indicates that the BER (Bit-Error-Rate) is better than 10⁻⁷.

An important detail is that once both devices have the RUNBIST instruction loaded into their respective instruction registers, both devices must move into the RTI state within 4K system clocks (At a SCLK of 66Mhz and TCK of 1MHz this allows for 66 TCK cycles). This is not a concern when both devices are on the same scan chain or LSP, however, it can be a problem with some multi-drop devices. This test mode has been simulated and verified using TI's SCANSTA111.

DIN0 Held Low-DIN1 Held High Creates an RMT Pattern



DIN4 Held Low-DIN5 Held High Creates an RMT Pattern



DIN8 Held Low-DIN9 Held High Creates an RMT Pattern

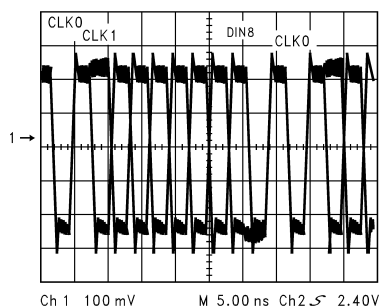


Figure 2. RMT Patterns Seen on the Bus LVDS Serial Output



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

Absolute Maximum Ratings⁽¹⁾

Supply Voltage (V_{CC})		-0.3V to +4V
LVC MOS/LVTTL Input Voltage		-0.3V to ($V_{CC} + 0.3V$)
LVC MOS/LVTTL Output Voltage		-0.3V to ($V_{CC} + 0.3V$)
Bus LVDS Receiver Input Voltage		-0.3V to +3.9V
Bus LVDS Driver Output Voltage		-0.3V to +3.9V
Bus LVDS Output Short Circuit Duration		10mS
Junction Temperature		+150°C
Storage Temperature		-65°C to +150°C
Lead Temperature	(Soldering, 4 seconds)	+220°C
Maximum Package Power Dissipation Capacity @ 25°C Package:	49L NFBGA	1.47 W
Package Derating:	49L NFBGA	11.8 mW/°C above +25°C
θ_{ja}		85°C/W
ESD Rating	HBM	>2kV
	MM	> 250V

(1) Absolute Maximum Ratings are those values beyond which the safety of the device cannot be ensured. They are not meant to imply that the devices should be operated at these limits. The table of Electrical Characteristics specifies conditions of device operation.

Recommended Operating Conditions

	Min	Nom	Max	Units
Supply Voltage (V_{CC})	3.0	3.3	3.6	V
Operating Free Air Temperature (T_A)	-40	+25	+125	°C
Receiver Input Range	0		2.4	V
Supply Noise Voltage (V_{CC})			100	mV _{P-P}

Electrical Characteristics

Over recommended operating supply and temperature ranges unless otherwise specified.

Symbol	Parameter	Conditions	Min	Typ	Max	Units	
SERIALIZER LVCMOS/LVTTL DC SPECIFICATIONS (apply to DIN0-9, TCLK, $\overline{\text{PWRDN}}$, TCLK_R/ $\overline{\text{F}}$, SYNC1, SYNC2, DEN)							
V _{IH}	High Level Input Voltage		2.0		V _{CC}	V	
V _{IL}	Low Level Input Voltage		GND		0.8	V	
V _{CL}	Input Clamp Voltage	I _{CL} = -18 mA		-0.86	-1.5	V	
I _{IN}	Input Current	V _{IN} = 0V or 3.6V	-10	±1	+10	μA	
DESERIALIZER LVCMOS/LVTTL DC SPECIFICATIONS (apply to pins $\overline{\text{PWRDN}}$, RCLK_R/ $\overline{\text{F}}$, REN, REFCLK = inputs; apply to pins ROUT, RCLK, $\overline{\text{LOCK}}$ = outputs)							
V _{IH}	High Level Input Voltage		2.0		V _{CC}	V	
V _{IL}	Low Level Input Voltage		GND		0.8	V	
V _{CL}	Input Clamp Voltage	I _{CL} = -18 mA		-0.62	-1.5	V	
I _{IN}	Input Current	V _{IN} = 0V or 3.6V	-10	±1	+15	μA	
V _{OH}	High Level Output Voltage	I _{OH} = -9 mA	2.2	3.0	V _{CC}	V	
V _{OL}	Low Level Output Voltage	I _{OL} = 9 mA	GND	0.25	0.5	V	
I _{OS}	Output Short Circuit Current	V _{OUT} = 0V	-15	-47	-85	mA	
I _{OZ}	Tri-state Output Current	$\overline{\text{PWRDN}}$ or REN = 0.8V, V _{OUT} = 0V or V _{CC}	-10	±0.1	+10	μA	
SERIALIZER Bus LVDS DC SPECIFICATIONS (apply to pins DO+ and DO-)							
V _{OD}	Output Differential Voltage (DO+)-(DO-)	RL = 27Ω, Figure 18	200	290		mV	
ΔV _{OD}	Output Differential Voltage Unbalance				35	mV	
V _{OS}	Offset Voltage		1.05	1.1	1.3	V	
ΔV _{OS}	Offset Voltage Unbalance			4.8	35	mV	
I _{OS}	Output Short Circuit Current	DO = 0V, DIN = High, $\overline{\text{PWRDN}}$ and DEN = 2.4V		-56	-90	mA	
I _{OZ}	Tri-state Output Current	$\overline{\text{PWRDN}}$ or DEN = 0.8V, DO = 0V or V _{CC}	-10	±1	+10	μA	
I _{OX}	Power-Off Output Current	V _{CC} = 0V, DO=0V or 3.6V	-20	±1	+30	μA	
DESERIALIZER Bus LVDS DC SPECIFICATIONS (apply to pins RI+ and RI-)							
V _{TH}	Differential Threshold High Voltage	V _{CM} = +1.1V		+6	+50	mV	
V _{TL}	Differential Threshold Low Voltage		-50	-12		mV	
I _{IN}	Input Current	V _{IN} = +2.4V, V _{CC} = 3.6V or 0V	-10	±1	+10	μA	
		V _{IN} = 0V, V _{CC} = 3.6V or 0V	-10	±0.05	+10	μA	
SERIALIZER SUPPLY CURRENT (apply to pins DVCC and AVCC)							
I _{CCD}	Serializer Supply Current	RL = 27Ω	f = 20 MHz		45	60	mA
	Worst Case	Figure 3	f = 80 MHz		90	105	mA
I _{CCXD}	Serializer Supply Current Powerdown	$\overline{\text{PWRDN}}$ = 0.8V, f = 80MHz			0.2	2.0	mA
DESERIALIZER SUPPLY CURRENT (apply to pins DVCC and AVCC)							
I _{CCR}	Deserializer Supply Current	C _L = 15 pF	f = 20 MHz		50	75	mA
	Worst Case	Figure 4	f = 80 MHz		100	120	mA
I _{CCXR}	Deserializer Supply Current Powerdown	$\overline{\text{PWRDN}}$ = 0.8V, REN = 0.8V			0.36	1.0	mA
SCAN CIRCUITRY DC SPECIFICATIONS, SERIALIZER AND DESERIALIZER (applies to SCAN pins as noted)							
V _{IH}	High Level Input Voltage	$\frac{\text{V}_{\text{CC}}}{\text{TRST}}$ = 3.0 to 3.6V, pins TCK, TMS, TDI, and		2.0		V _{CC}	V
V _{IL}	Low Level Input Voltage	$\frac{\text{V}_{\text{CC}}}{\text{TRST}}$ = 3.0 to 3.6V, pins TCK, TMS, TDI, and		GND		0.8	V
V _{CL}	Input Clamp Voltage	$\text{V}_{\text{CC}} = 3.0\text{V}$, I _{CL} = -18 mA, pins TCK, TMS, TDI, and $\frac{\text{TRST}}{\text{TRST}}$			-0.85	-1.5	V
I _{IH}	Input Current	$\text{V}_{\text{CC}} = 3.6\text{V}$, V _{IN} = 3.6V, pins TCK, TMS, TDI, and $\frac{\text{TRST}}{\text{TRST}}$			1	+10	μA
I _{IL}	Input Current	$\text{V}_{\text{CC}} = 3.6\text{V}$, V _{IN} = 0.0V, TCK Input		-10	-1		μA

Electrical Characteristics (continued)

Over recommended operating supply and temperature ranges unless otherwise specified.

Symbol	Parameter	Conditions	Min	Typ	Max	Units
I_{ILR}	Input Current	$V_{CC} = 3.6V$, $V_{IN} = 0V$, pins TMS, TDI, and TRST	-20	-10		μA
V_{OH}	High Level Output Voltage	$V_{CC} = 3.0V$, $I_{OH} = -12\text{ mA}$, TDO output	2.2	2.6		V
V_{OL}	Low Level Output Voltage	$V_{CC} = 3.0V$, $I_{OL} = 12\text{ mA}$, TDO output		0.3	0.5	V
I_{OS}	Output Short Circuit Current	$V_{CC} = 3.6V$, $V_{OUT} = 0.0V$, TDO output	-15	-90	-120	mA
I_{OZ}	Tri-state Output Current	PWRDN or REN = 0.8V, $V_{OUT} = 0V$ or VCC	-10	0	+10	μA

Serializer Timing Requirements for TCLK

Over recommended operating supply and temperature ranges unless otherwise specified.

Symbol	Parameter	Conditions	Min	Typ	Max	Units
t_{TCP}	Transmit Clock Period		12.5	T	50.0	ns
t_{TCIH}	Transmit Clock High Time		0.4T	0.5T	0.6T	ns
t_{TCIL}	Transmit Clock Low Time		0.4T	0.5T	0.6T	ns
t_{CLKT}	TCLK Input Transition Time			3	6	ns
t_{JIT}	TCLK Input Jitter				150	ps (RMS)

Serializer Switching Characteristics

Over recommended operating supply and temperature ranges unless otherwise specified.

Symbol	Parameter	Conditions	Min	Typ	Max	Units
t_{LLHT}	Bus LVDS Low-to-High Transition Time	$R_L = 27\Omega$ $C_L = 10\text{pF}$ to GND Figure 5 ⁽¹⁾		0.2	0.4	ns
t_{LHLT}	Bus LVDS High-to-Low Transition Time			0.25	0.4	ns
t_{DIS}	DIN (0-9) Setup to TCLK	$R_L = 27\Omega$, $C_L = 10\text{pF}$ to GND Figure 8	0			ns
t_{DIH}	DIN (0-9) Hold from TCLK		4.0			ns
t_{HZD}	DO $\pm$ HIGH to Tri-state Delay	$R_L = 27\Omega$, $C_L = 10\text{pF}$ to GND Figure 9 ⁽²⁾		3	10	ns
t_{LZD}	DO $\pm$ LOW to Tri-state Delay			3	10	ns
t_{ZHD}	DO $\pm$ Tri-state to HIGH Delay			5	10	ns
t_{ZLD}	DO $\pm$ Tri-state to LOW Delay			6.5	10	ns
t_{SPW}	SYNC Pulse Width	$R_L = 27\Omega$ Figure 11	$5 \cdot t_{TCP}$			ns
t_{PLD}	Serializer PLL Lock Time		$510 \cdot t_{TCP}$		$513 \cdot t_{TCP}$	ns
t_{SD}	Serializer Delay	$R_L = 27\Omega$, Figure 12	$t_{TCP} + 1.0$	$t_{TCP} + 2.5$	$t_{TCP} + 3.5$	ns
t_{DJIT}	Deterministic Jitter	$R_L = 27\Omega$, $C_L = 10\text{pF}$ to GND ⁽³⁾	20MHz	-330	140	ps
			80MHz	-130	+60	ps
t_{RJIT}	Random Jitter			6	10	ps (RMS)

(1) t_{LLHT} and t_{LHLT} specifications are Guaranteed By Design (GBD) using statistical analysis.

(2) Because the Serializer is in tri-state mode, the Deserializer will lose PLL lock and have to resynchronize before data transfer.

(3) t_{DJIT} specifications are Guaranteed By Design using statistical analysis.

Deserializer Timing Requirements for REFCLK

Over recommended operating supply and temperature ranges unless otherwise specified.

Symbol	Parameter	Conditions	Min	Typ	Max	Units
t_{RCP}	REFCLK Period		12.5	T	50.0	ns
t_{RFDC}	REFCLK Duty Cycle		30	50	70	%
t_{RCP} / t_{TCP}	Ratio of REFCLK to TCLK		95	1	105	
t_{RFTT}	REFCLK Transition Time			3	6	ns

Deserializer Switching Characteristics

Over recommended operating supply and temperature ranges unless otherwise specified.

Symbol	Parameter	Conditions	Pin/Freq.	Min	Typ	Max	Units
t _{RCP}	Receiver out Clock Period	t _{RCP} = t _{TCP} Figure 12	RCLK	12.5		50.0	ns
t _{CLH}	CMOS/TTL Low-to-High Transition Time	CL = 15 pF Figure 6	Rout(0-9), LOCK, RCLK		1.2	4	ns
t _{CHL}	CMOS/TTL High-to-Low Transition Time				1.1	4	ns
t _{DD}	Deserializer Delay Figure 13	All Temp, All Freq		1.75*t _{RCP} +1.25	1.75*t _{RCP} +5.0	1.75*t _{RCP} +8.5	ns
		Room Temp, 3.3V	20MHz	1.75*t _{RCP} +2.25	1.75*t _{RCP} +5.0	1.75*t _{RCP} +8.0	ns
		Room Temp, 3.3V	80MHz	1.75*t _{RCP} +2.25	1.75*t _{RCP} +5.0	1.75*t _{RCP} +8.0	ns
t _{ROS}	ROUT Data Valid before RCLK	Figure 14	RCLK 20MHz	0.4*t _{RCP}	0.5*t _{RCP}		ns
			RCLK 80MHz	0.35*t _{RCP}	0.5*t _{RCP}		ns
t _{ROH}	ROUT Data Valid after RCLK	Figure 14	20MHz	-0.4*t _{RCP}	-0.5*t _{RCP}		ns
			80MHz	-0.35*t _{RCP}	-0.5*t _{RCP}		ns
t _{RDC}	RCLK Duty Cycle			45	50	55	%
t _{HZR}	HIGH to Tri-state Delay	Figure 15	Rout(0-9)		2.8	10	ns
t _{LZR}	LOW to Tri-state Delay				2.8	10	ns
t _{ZHR}	Tri-state to HIGH Delay				4.2	10	ns
t _{ZLR}	Tri-state to LOW Delay				4.2	10	ns
t _{DSR1}	Deserializer PLL Lock Time from PWRDWN (with SYNCPAT)	Figure 16 Figure 17 ⁽¹⁾	20MHz		1.7	3.5	μs
			80MHz		1.0	2.5	μs
t _{DSR2}	Deserializer PLL Lock time from SYNCPAT		20MHz	0.65	1.5	μs	
			80MHz	0.29	0.8	μs	
t _{ZHLK}	Tri-state to HIGH Delay (power-up)		$\overline{\text{LOCK}}$		3.7	12	ns
t _{RNMI-R}	Ideal Noise Margin Right Figure 21		VCC = 3.15 to 3.6V	80MHz			+335
		VCC = 3.0V				+215	
			20MHz			+1	ns
t _{RNMI-L}	Ideal Noise Margin Left Figure 21	VCC = 3.15 to 3.6V	80MHz	-395			ps
		VCC = 3.0V		-520			
			20MHz	-1			ns

- (1) For the purpose of specifying deserializer PLL performance, t_{DSR1} and t_{DSR2} are specified with the REFCLK running and stable, and with specific conditions for the incoming data stream (SYNCPATS). It is recommended that the deserializer be initialized using either t_{DSR1} timing or t_{DSR2} timing. t_{DSR1} is the time required for the deserializer to indicate lock upon power-up or when leaving the power-down mode. Synchronization patterns should be sent to the device before initiating either condition. t_{DSR2} is the time required to indicate lock for the powered-up and enabled deserializer when the input (RI+ and RI-) conditions change from not receiving data to receiving synchronization patterns (SYNCPATS).

SCAN Circuitry Timing Requirements

Symbol	Parameter	Conditions	Min	Typ	Max	Units
f_{MAX}	Maximum TCK Clock Frequency	$R_L = 500\Omega$, $C_L = 35\text{ pF}$	25.0	50.0		MHz
t_S	TDI to TCK, H or L		1.0			ns
t_H	TDI to TCK, H or L		2.0			ns
t_S	TMS to TCK, H or L		2.5			ns
t_H	TMS to TCK, H or L		1.5			ns
t_W	TCK Pulse Width, H or L		10.0			ns
t_W	$\overline{\text{TRST}}$ Pulse Width, L		2.5			ns
t_{REC}	Recovery Time, $\overline{\text{TRST}}$ to TCK		2.0			ns

AC Timing Diagrams and Test Circuits

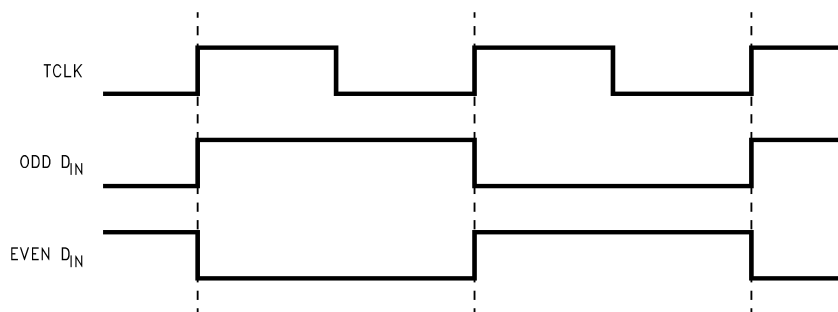


Figure 3. "Worst Case" Serializer ICC Test Pattern

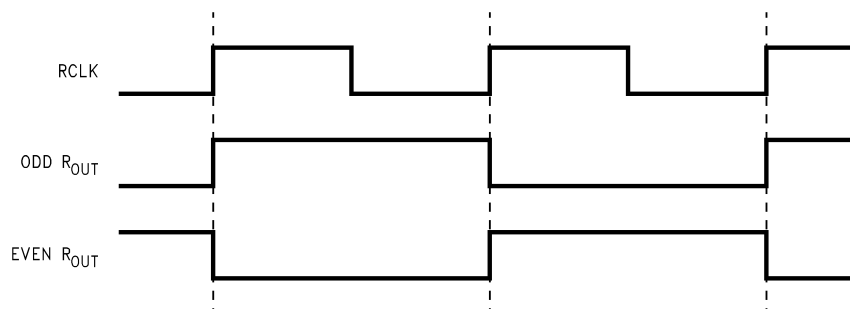


Figure 4. "Worst Case" Deserializer ICC Test Pattern

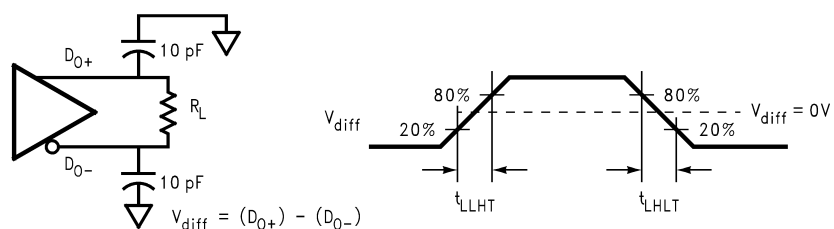


Figure 5. Serializer Bus LVDS Output Load and Transition Times

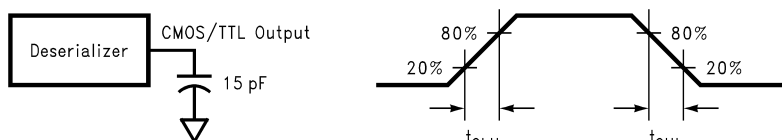


Figure 6. Deserializer CMOS/TTL Output Load and Transition Times

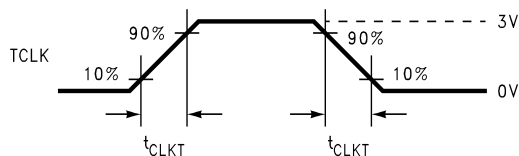
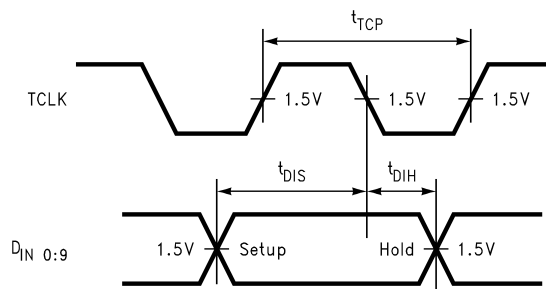


Figure 7. Serializer Input Clock Transition Time



Timing shown for TCLK_R/ $\bar{F}$ = LOW

Figure 8. Serializer Setup/Hold Times

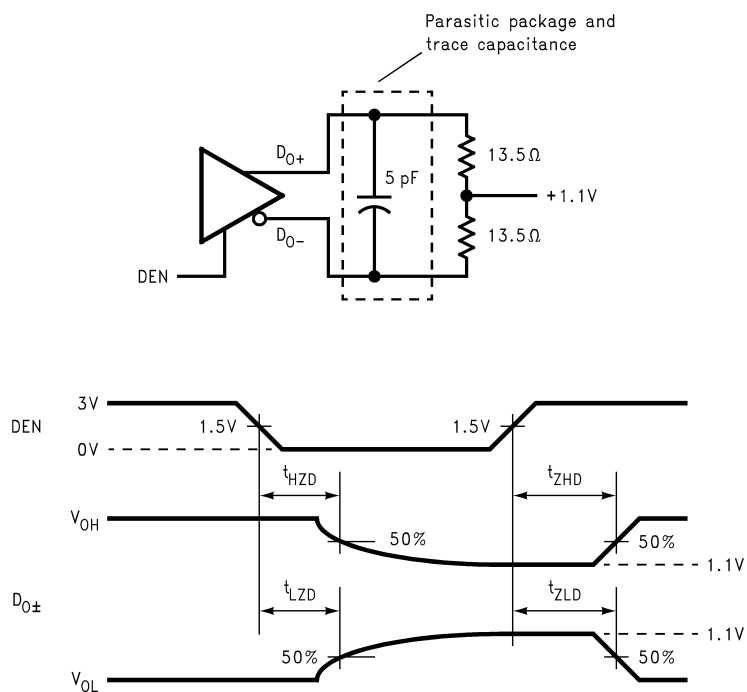


Figure 9. Serializer Tri-state Test Circuit and Timing

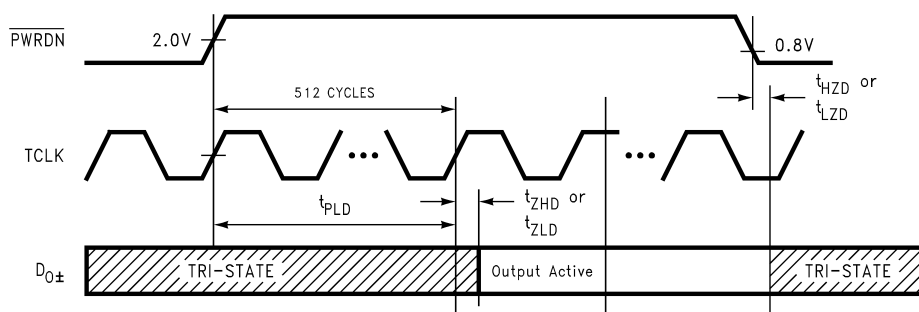


Figure 10. Serializer PLL Lock Time, and PWRDN Tri-state Delays

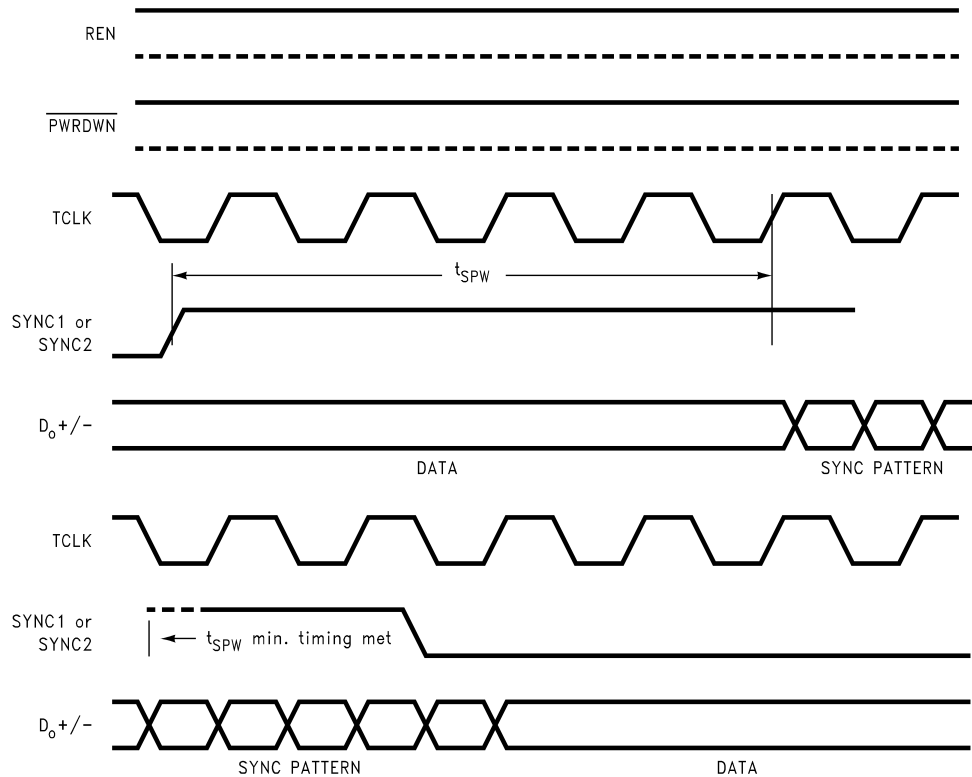


Figure 11. SYNC Timing Delays

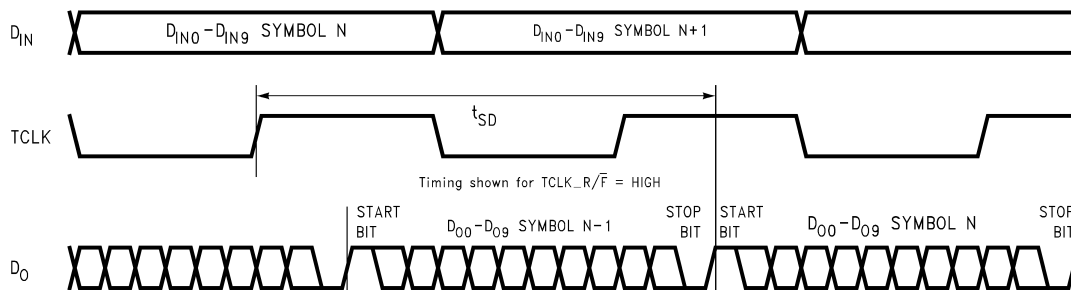


Figure 12. Serializer Delay

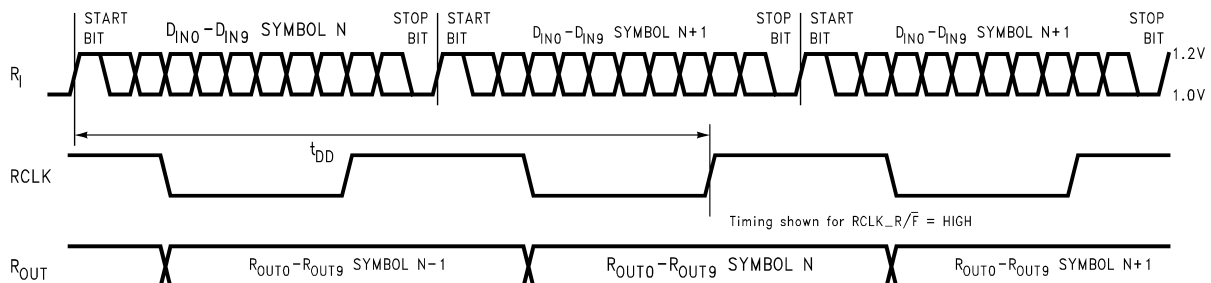
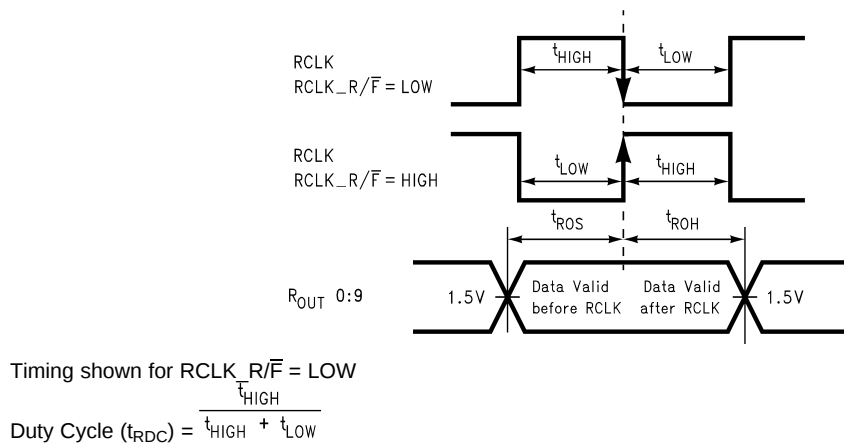
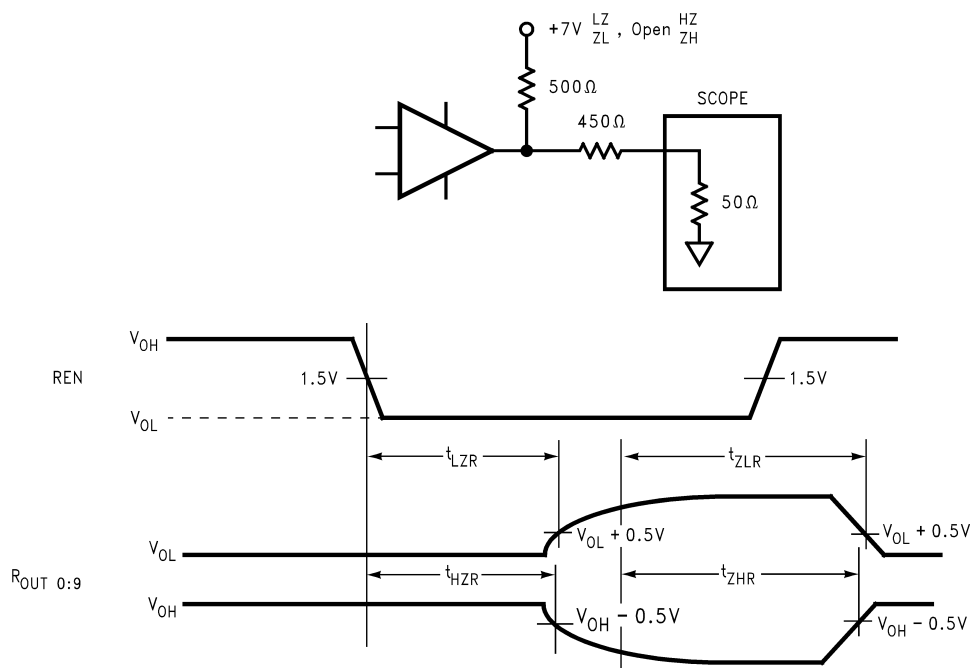


Figure 13. Deserializer Delay

**Figure 14. Deserializer Data Valid Out Times****Figure 15. Deserializer Tri-state Test Circuit and Timing**

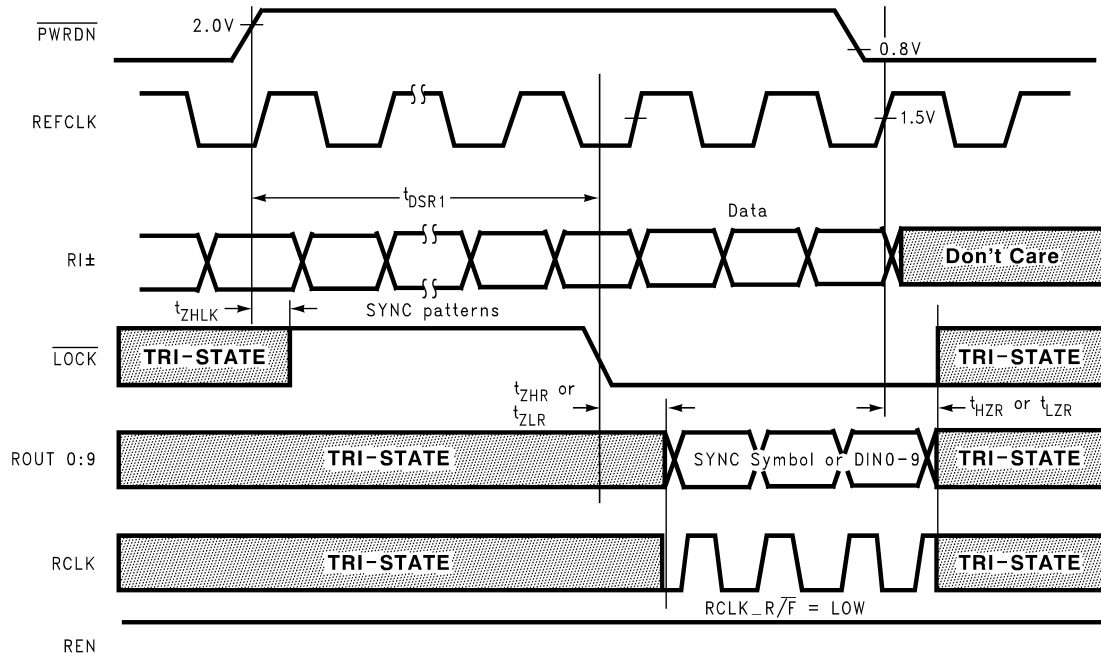


Figure 16. Deserializer PLL Lock Times and PWRDN Tri-state Delays

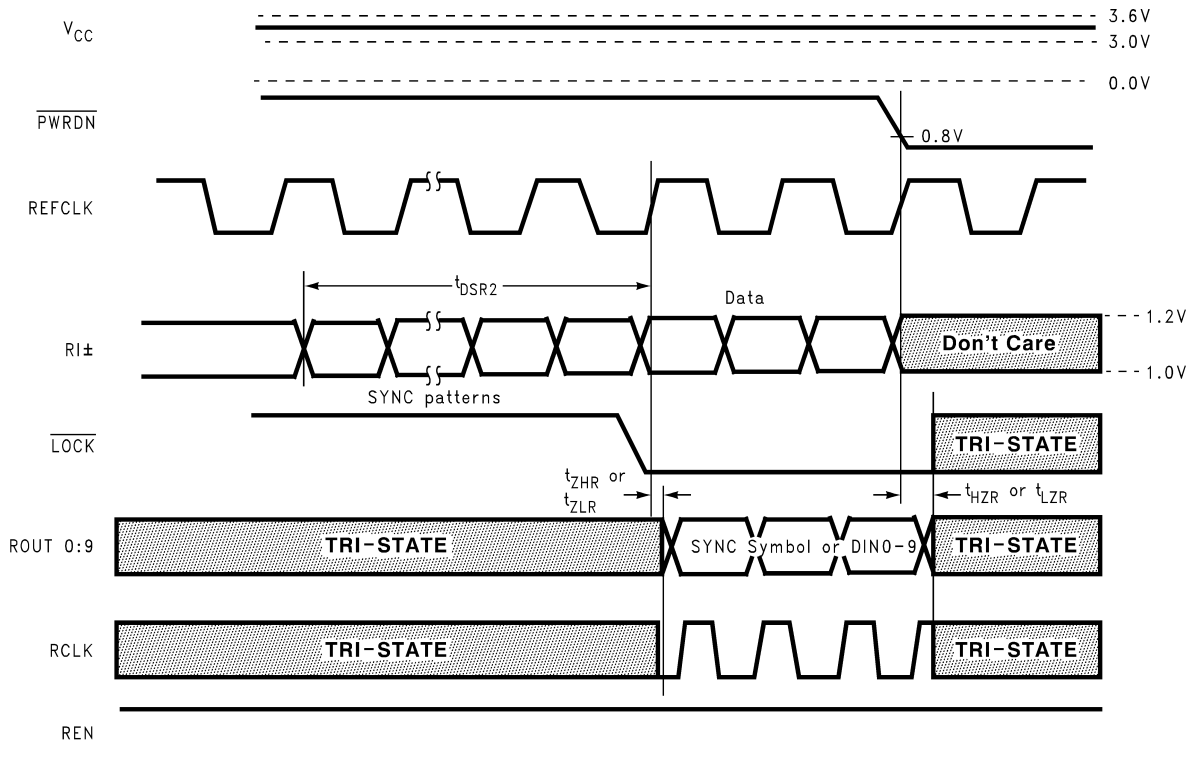
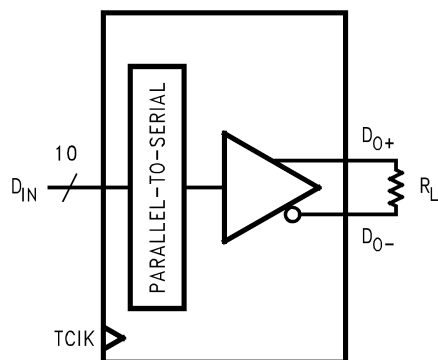


Figure 17. Deserializer PLL Lock Time from SyncPAT



$$V_{OD} = (DO^+) - (DO^-).$$

Differential output signal is shown as (DO+)-(DO-), device in Data Transfer mode.

Figure 18. V_{OD} Diagram

APPLICATION INFORMATION

USING THE SCAN921025H AND SCAN921226H

The Serializer and Deserializer chipset is an easy to use transmitter and receiver pair that sends 10 bits of parallel LVTTTL data over a serial Bus LVDS link up to 800 Mbps. An on-board PLL serializes the input data and embeds two clock bits within the data stream. The Deserializer uses a separate reference clock (REFCLK) and an onboard PLL to extract the clock information from the incoming data stream and then deserialize the data. The Deserializer monitors the incoming clock information, determines lock status, and asserts the $\overline{\text{LOCK}}$ output high when loss of lock occurs.

POWER CONSIDERATIONS

An all CMOS design of the Serializer and Deserializer makes them inherently low power devices. In addition, the constant current source nature of the Bus LVDS outputs minimizes the slope of the speed vs. I_{CC} curve of conventional CMOS designs.

POWERING UP THE DESERIALIZER

The SCAN921226H can be powered up at any time by following the proper sequence. The REFCLK input can be running before the Deserializer powers up, and it must be running in order for the Deserializer to lock to incoming data. The Deserializer outputs will remain in tri-state until the Deserializer detects data transmission at its inputs and locks to the incoming data stream.

TRANSMITTING DATA

Once you power up the Serializer and Deserializer, they must be phase locked to each other to transmit data. Phase locking occurs when the Deserializer locks to incoming data or when the Serializer sends patterns. The Serializer sends SYNC patterns whenever the SYNC1 or SYNC2 inputs are high. The $\overline{\text{LOCK}}$ output of the Deserializer remains high until it has locked to the incoming data stream. Connecting the $\overline{\text{LOCK}}$ output of the Deserializer to one of the SYNC inputs of the Serializer will ensure that enough SYNC patterns are sent to achieve Deserializer lock.

The Deserializer can also lock to incoming data by simply powering up the device and allowing the “random lock” circuitry to find and lock to the data stream.

While the Deserializer $\overline{\text{LOCK}}$ output is low, data at the Deserializer outputs (ROUT0-9) is valid, except for the specific case of loss of lock during transmission which is further discussed in the [RECOVERING FROM LOCK LOSS](#) section below.

NOISE MARGIN

The Deserializer noise margin is the amount of input jitter (phase noise) that the Deserializer can tolerate and still reliably receive data. Various environmental and systematic factors include:

Serializer: TCLK jitter, V_{CC} noise (noise bandwidth and out-of-band noise)

Media: ISI, Large V_{CM} shifts

Deserializer: V_{CC} noise

RECOVERING FROM LOCK LOSS

In the case where the Deserializer loses lock during data transmission, up to 3 cycles of data that were previously received can be invalid. This is due to the delay in the lock detection circuit. The lock detect circuit requires that invalid clock information be received 4 times in a row to indicate loss of lock. Since clock information has been lost, it is possible that data was also lost during these cycles. Therefore, after the Deserializer relocks to the incoming data stream and the Deserializer $\overline{\text{LOCK}}$ pin goes low, at least three previous data cycles should be suspect for bit errors.

The Deserializer can relock to the incoming data stream by making the Serializer resend SYNC patterns, as described above, or by random locking, which can take more time, depending on the data patterns being received.

HOT INSERTION

All the BLVDS devices are hot pluggable if you follow a few rules. When inserting, ensure the Ground pin(s) makes contact first, then the VCC pin(s), and then the I/O pins. When removing, the I/O pins should be unplugged first, then the VCC, then the Ground. Random lock hot insertion is illustrated in [Figure 22](#).

PCB CONSIDERATIONS

The Bus LVDS Serializer and Deserializer should be placed as close to the edge connector as possible. In multiple Deserializer applications, the distance from the Deserializer to the slot connector appears as a stub to the Serializer driving the backplane traces. Longer stubs lower the impedance of the bus, increase the load on the Serializer, and lower the threshold margin at the Deserializers. Deserializer devices should be placed much less than one inch from slot connectors. Because transition times are very fast on the Serializer Bus LVDS outputs, reducing stub lengths as much as possible is the best method to ensure signal integrity.

TRANSMISSION MEDIA

The Serializer and Deserializer can also be used in point-to-point configuration of a backplane, through a PCB trace, or through twisted pair cable. In point-to-point configuration, the transmission media need only be terminated at the receiver end. Please note that in point-to-point configuration, the potential of offsetting the ground levels of the Serializer vs. the Deserializer must be considered. Also, Bus LVDS provides a $\pm 1.2\text{V}$ common mode range at the receiver inputs.

FAILSAFE BIASING FOR THE SCAN921226H

The SCAN921226H has an improved input threshold sensitivity of $\pm 50\text{mV}$ versus $\pm 100\text{mV}$ for the DS92LV1210 or DS92LV1212. This allows for greater differential noise margin in the SCAN921226H. However, in cases where the receiver input is not being actively driven, the increased sensitivity of the SCAN921226H can pickup noise as a signal and cause unintentional locking. For example, this can occur when the input cable is disconnected.

External resistors can be added to the receiver circuit board to prevent noise pick-up. Typically, the non-inverting receiver input is pulled up and the inverting receiver input is pulled down by high value resistors. The pull-up and pull-down resistors (R_1 and R_2) provide a current path through the termination resistor (R_L) which biases the receiver inputs when they are not connected to an active driver. The value of the pull-up and pull-down resistors should be chosen so that enough current is drawn to provide a $+15\text{mV}$ drop across the termination resistor. Please see [Figure 19](#) for the Failsafe Biasing Setup.

USING t_{DJIT} AND t_{RNM} TO VALIDATE SIGNAL QUALITY

The parameter t_{RNM} is calculated by first measuring how much of the ideal bit the receiver needs to ensure correct sampling. After determining this amount, what remains of the ideal bit that is available for external sources of noise is called t_{RNM} . t_{RNM} includes transmitter jitter.

Please refer to [Figure 20](#) and [Figure 21](#) for a graphic representation of t_{DJIT} and t_{RNM} . Also, for a more detailed explanation of t_{RNM} , please see the Application Note titled 'How to Validate BLVDS SER/DES Signal Integrity Using an Eye Mask' (SNLA053).

The vertical limits of the mask are determined by the SCAN921226H receiver input threshold of $\pm 50\text{mV}$.

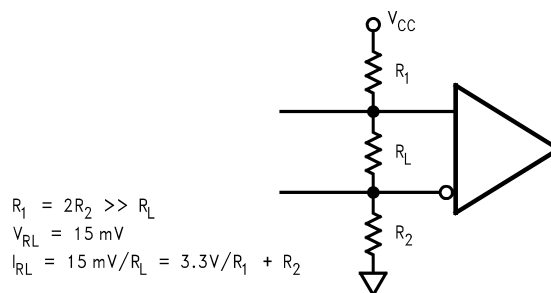


Figure 19. Failsafe Biasing Setup

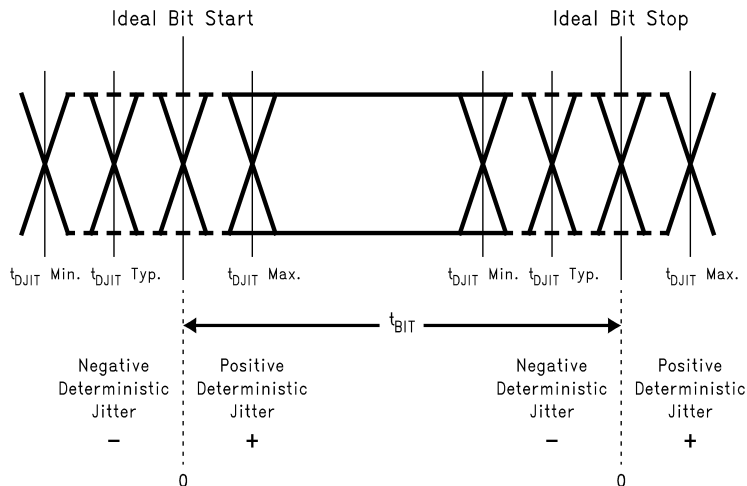
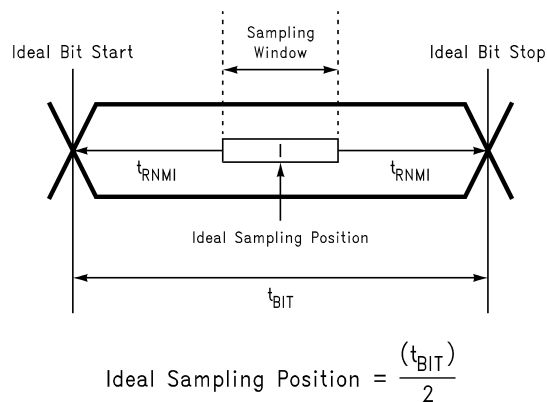


Figure 20. Deterministic Jitter and Ideal Bit Position



$t_{\text{RNMI-L}}$ is the ideal noise margin on the left of the figure, it is a negative value to indicate early with respect to ideal.
 $t_{\text{RNMI-R}}$ is the ideal noise margin on the right of the above figure, it is a positive value to indicate late with respect to ideal.

Figure 21. Ideal Deserializer Noise Margin (t_{RNMI}) and Sampling Window

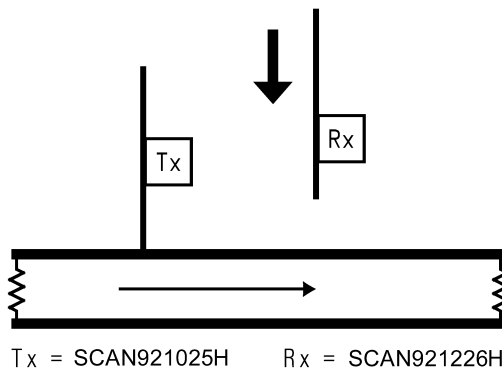


Figure 22. Random Lock Hot Insertion

Pin Diagrams

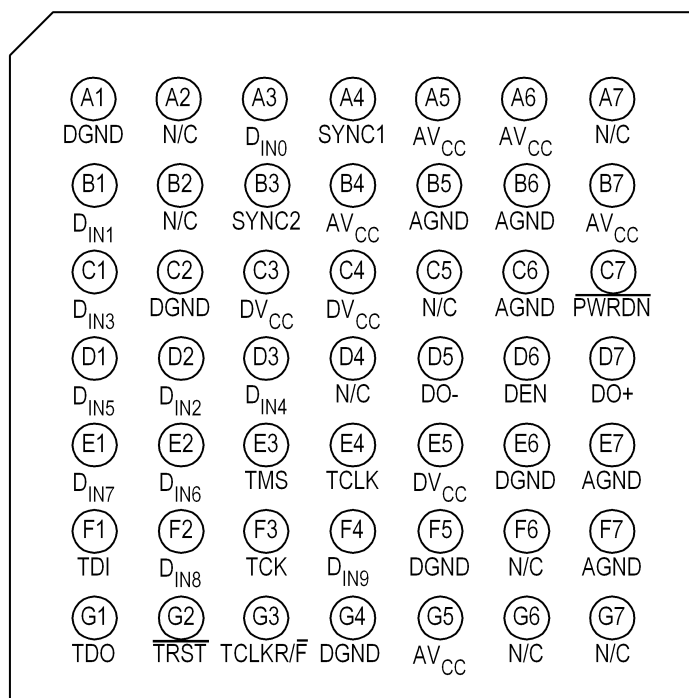


Figure 23. SCAN921025HSM - Serializer (Top View)

Figure 24.

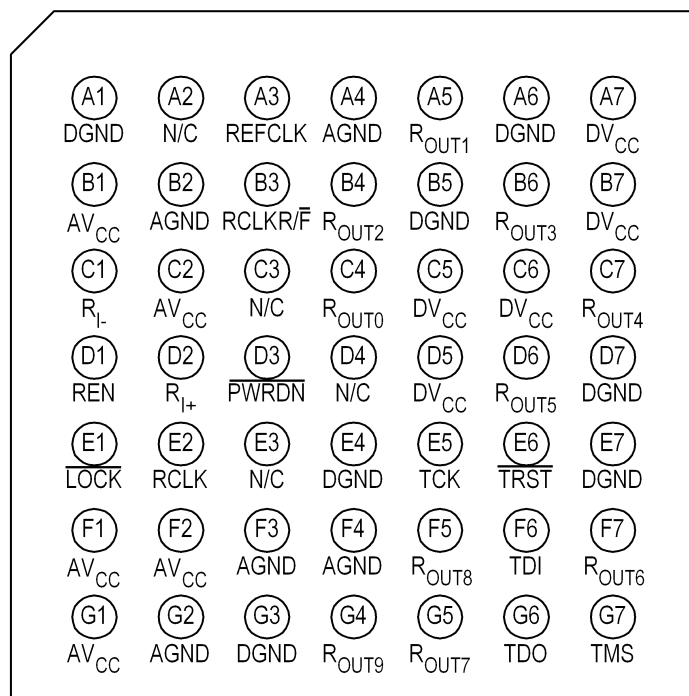


Figure 25. SCAN921226HSM - Deserializer (Top View)

Serializer Pin Descriptions

Pin Name	I/O	Ball Id.	Description
DIN	I	A3, B1, C1, D1, D2, D3, E1, E2, F2, F4	Data Input. LVTTTL levels inputs. Data on these pins are loaded into a 10-bit input register.
TCLKR/ $\overline{F}$	I	G3	Transmit Clock Rising/Falling strobe select. LVTTTL level input. Selects TCLK active edge for strobing of DIN data. High selects rising edge. Low selects falling edge.
DO+	O	D7	+ Serial Data Output. Non-inverting Bus LVDS differential output.
DO–	O	D5	– Serial Data Output. Inverting Bus LVDS differential output.
DEN	I	D6	Serial Data Output Enable. LVTTTL level input. A low puts the Bus LVDS outputs in tri-state.
$\overline{PWRDN}$	I	C7	Powerdown. LVTTTL level input. $\overline{PWRDN}$ driven low shuts down the PLL and tri-states outputs putting the device into a low power sleep mode.
TCLK	I	E4	Transmit Clock. LVTTTL level input. Input for 20MHz – 80MHz system clock.
SYNC	I	A4, B3	Assertion of SYNC (high) for at least 1024 synchronization symbols to be transmitted on the Bus LVDS serial output. Synchronization symbols continue to be sent if SYNC continues to be asserted. TTL level input. The two SYNC pins are ORed.
DVCC	I	C3, C4, E5	Digital Circuit power supply.
DGND	I	A1, C2, F5, E6, G4	Digital Circuit ground.
AVCC	I	A5, A6, B4, B7, G5	Analog power supply (PLL and Analog Circuits).
AGND	I	B5, B6, C6, E7, F7	Analog ground (PLL and Analog Circuits).
TDI	I	F1	Test Data Input to support IEEE 1149.1. There is an internal pullup resistor that defaults this input to high per IEEE 1149.1.
TDO	O	G1	Test Data Output to support IEEE 1149.1
TMS	I	E3	Test Mode Select Input to support IEEE 1149.1. There is an internal pullup resistor that defaults this input to high per IEEE 1149.1.
TCK	I	F3	Test Clock Input to support IEEE 1149.1
$\overline{TRST}$	I	G2	Test Reset Input to support IEEE 1149.1. There is an internal pullup resistor that defaults this input to high per IEEE 1149.1.
N/C	N/A	A2, A7, B2, C5, D4, F6, G6, G7	Leave open circuit, do not connect

Deserializer Pin Descriptions

Pin Name	I/O	Ball Id.	Description
ROUT	O	A5, B4, B6, C4, C7, D6, F5, F7, G4, G5	Data Output. ± 9 mA CMOS level outputs.
RCLKR/ $\overline{F}$	I	B3	Recovered Clock Rising/Falling strobe select. TTL level input. Selects RCLK active edge for strobing of ROUT data. High selects rising edge. Low selects falling edge.
RI+	I	D2	+ Serial Data Input. Non-inverting Bus LVDS differential input.
RI-	I	C1	- Serial Data Input. Inverting Bus LVDS differential input.
$\overline{PWRDN}$	I	D3	Powerdown. TTL level input. $\overline{PWRDN}$ driven low shuts down the PLL and tri-states outputs putting the device into a low power sleep mode.
$\overline{LOCK}$	O	E1	$\overline{LOCK}$ goes low when the Deserializer PLL locks onto the embedded clock edge. CMOS level output. Totem pole output structure, does not directly support wired OR connections.
RCLK	O	E2	Recovered Clock. Parallel data rate clock recovered from embedded clock. Used to strobe ROUT, CMOS level output.
REN	I	D1	Output Enable. TTL level input. When driven low, tri-states ROUT0–ROUT9 and RCLK.
DVCC	I	A7, B7, C5, C6, D5	Digital Circuit power supply.
DGND	I	A1, A6, B5, D7, E4, E7, G3	Digital Circuit ground.
AVCC	I	B1, C2, F1, F2, G1	Analog power supply (PLL and Analog Circuits).
AGND	I	A4, B2, F3, F4, G2	Analog ground (PLL and Analog Circuits).
REFCLK	I	A3	Use this pin to supply a REFCLK signal for the internal PLL frequency.
TDI	I	F6	Test Data Input to support IEEE 1149.1. There is an internal pullup resistor that defaults this input to high per IEEE 1149.1.
TDO	O	G6	Test Data Output to support IEEE 1149.1
TMS	I	G7	Test Mode Select Input to support IEEE 1149.1. There is an internal pullup resistor that defaults this input to high per IEEE 1149.1.
TCK	I	E5	Test Clock Input to support IEEE 1149.1
$\overline{TRST}$	I	E6	Test Reset Input to support IEEE 1149.1. There is an internal pullup resistor that defaults this input to high per IEEE 1149.1.
N/C	N/A	A2, C3, D4, E3	Leave open circuit, do not connect

Deserializer Truth Table

INPUTS		OUTPUTS		
$\overline{PWRDN}$	REN	ROUT [0:9] ⁽¹⁾	$\overline{LOCK}$ ⁽²⁾	RCLK ⁽³⁾⁽¹⁾
H ⁽⁴⁾	H	Z	H	Z
H	H	Active	L	Active
L	X	Z	Z	Z
H	L	Z	Active	Z

(1) ROUT and RCLK are tri-stated when $\overline{LOCK}$ is asserted High.

(2) $\overline{LOCK}$ Active indicates the $\overline{LOCK}$ output will reflect the state of the Deserializer with regard to the selected data stream.

(3) RCLK Active indicates the RCLK will be running if the Deserializer is locked. The Timing of RCLK with respect to ROUT is determined by RCLK_R/ $\overline{F}$.

(4) During Power-up.

REVISION HISTORY

Changes from Revision B (May 2013) to Revision C	Page
• Changed layout of National Data Sheet to TI format.	22

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
SCAN921025HSM	NRND	NFBGA	NZA	49	416	TBD	Call TI	Call TI	-40 to 125	SCAN921025 HSM	
SCAN921025HSM/NOPB	ACTIVE	NFBGA	NZA	49	416	Green (RoHS & no Sb/Br)	SNAGCU	Level-4-260C-72 HR	-40 to 125	SCAN921025 HSM	Samples
SCAN921025HSMX/NOPB	ACTIVE	NFBGA	NZA	49	2000	Green (RoHS & no Sb/Br)	SNAGCU	Level-4-260C-72 HR	-40 to 125	SCAN921025 HSM	Samples
SCAN921226HSM	NRND	NFBGA	NZA	49	416	TBD	Call TI	Call TI	-40 to 125	SCAN921226 HSM	
SCAN921226HSM/NOPB	ACTIVE	NFBGA	NZA	49	416	Green (RoHS & no Sb/Br)	SNAGCU	Level-4-260C-72 HR	-40 to 125	SCAN921226 HSM	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSELETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

⁽⁶⁾ Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

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TAPE AND REEL INFORMATION


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
SCAN921025HSMX/NOPB	NFBGA	NZA	49	2000	330.0	16.4	7.3	7.3	2.1	12.0	16.0	Q1

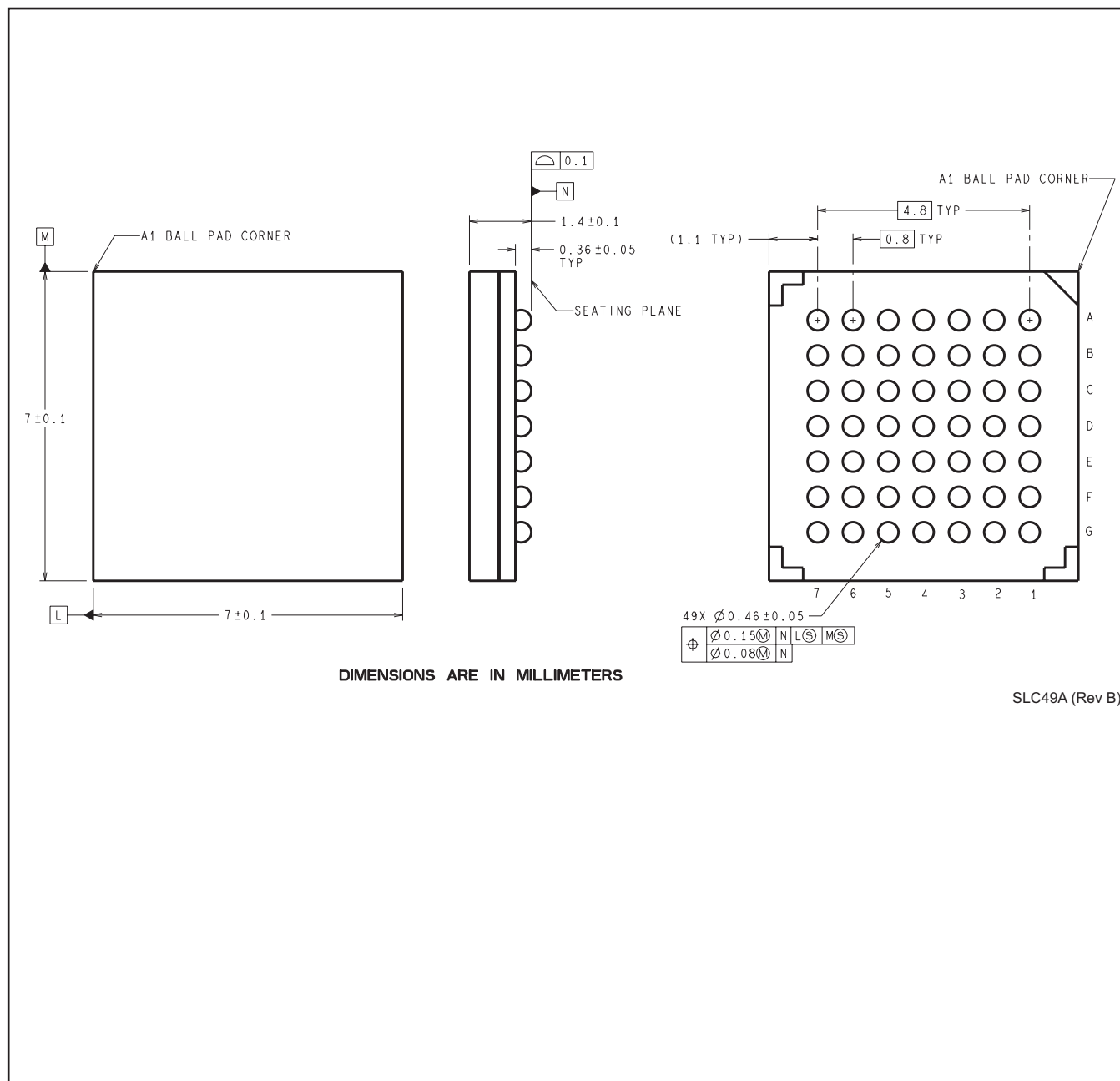
TAPE AND REEL BOX DIMENSIONS



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
SCAN921025HSMX/NOPB	NFBGA	NZA	49	2000	367.0	367.0	38.0

NZA0049A



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