











### TMUX6111, TMUX6112, TMUX6113

SCDS383 - AUGUST 2018

# TMUX611x 36-V, Low-Capacitance, Low-Leakage-Current, Precision, Quad SPST Switches

#### 1 Features

Low On-Capacitance: 4.2 pFLow Input Leakage: 1 pA

Low Charge Injection: -0.6 pC

• Rail-to-Rail Operation

Wide Supply Range: ±5 V to ±16.5 V (dual), 10 V

to 16.5 V (single)

• Low On-Resistance: 125  $\Omega$ 

· Fast Switch Turn-On Time: 66 ns

 Break-Before-Make Switching Action for TMUX6113

EN Pin Connectable to V<sub>DD</sub>

Logic Levels: 2 V to V<sub>DD</sub>

Low Supply Current: 17 μA

ESD Protection HBM: 2000 V

 Industry-Standard TSSOP and smaller WQFN Package

### 2 Applications

- Factory Automation and Industrial Process Controls
- Programmable Logic Controllers (PLC)
- Analog Input Modules
- ATE Test Equipment
- Digital Multimeters
- · Battery Monitoring Systems

### 3 Description

TMUX6111. TMUX6112, and TMUX6113 metal-oxide devices are modern complementary semiconductor (CMOS) devices that have four independently selectable single-pole/single-throw (SPST) switches. The devices work well with dual supplies (±5 V to ±16.5 V), a single supply (10 V to 16.5 V), or unsymmetric supplies (such as  $V_{DD} = 12$ V,  $V_{SS} = -5$  V). All digital inputs have transistortransistor logic (TTL) compatible thresholds, ensuring both TTL and CMOS logic compatibility.

The switches are turned on with Logic 0 on the digital control inputs in the TMUX6111. Logic 1 is required to turn on switches in the TMUX6112. The TMUX6113 has two switches with similar digital control logic to the TMUX6111 while the logic is inverted on the other two switches. The TMUX6113 has break-before-make (BBM) switching behavior, allowing the device to be used in multiplexer applications.

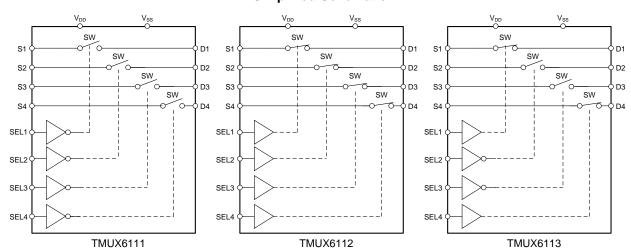
The TMUX611x devices are part of Texas Instruments Precision Switches and Multiplexers family. The devices have very low leakage current and charge injection, allowing them to be used in high-precision measurement applications. Low supply current of 17  $\mu$ A enables the device usage in portable applications.

#### **Device Information**(1)

PART NUMBER	PACKAGE	BODY SIZE (NOM)
TMUX6111	TSSOP (16)	5.00 mm × 4.40 mm
TMUX6112 TMUX6113	WQFN (16)	3.00 mm x 3.00 mm

 For all available packages, see the orderable addendum at the end of the data sheet.

#### Simplified Schematic





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### 4 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

DATE	REVISION	NOTES
August 2018	*	Initial release.

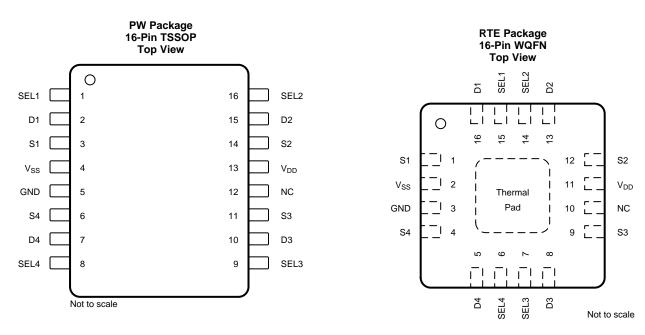


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## 5 Device Comparison Table

PRODUCT	DESCRIPTION
TMUX6111	36-V, Low Capacitance, Low-Leakage-Current, Precision, Quad SPST Switches (Normally Closed)
TMUX6112	36-V, Low Capacitance, Low-Leakage-Current, Precision, Quad SPST Switches (Normally Open)
TMUX6113	36-V, Low Capacitance, Low-Leakage-Current, Precision, Quad SPST Switches (Dual Open + Dual Closed)

# 6 Pin Configuration and Functions



### **Pin Functions**

	PIN		<b>-</b> 1/2-(1)	
NAME	TSSOP	WQFN	TYPE <sup>(1)</sup>	DESCRIPTION
SEL1	1	15	I	Logic control input 1.
D1	2	16	I/O	Drain pin 1. Can be an input or output.
S1	3	1	I/O	Source pin 1. Can be an input or output.
V <sub>SS</sub>	4	2	Р	Negative power supply. This pin is the most negative power-supply potential. In single-supply applications, this pin can be connected to ground. For reliable operation, connect a decoupling capacitor ranging from 0.1 $\mu F$ to 10 $\mu F$ between $V_{SS}$ and GND.
GND	5	3	Р	Ground (0 V) reference
S4	6	4	I/O	Source pin 4. Can be an input or output.
D4	7	5	I/O	Drain pin 4. Can be an input or output.
SEL4	8	6	I	Logic control input 4.
SEL3	9	7	1	Logic control input 3.
D3	10	8	I/O	Drain pin 3. Can be an input or output.
S3	11	9	I/O	Source pin 3. Can be an input or output.
NC	12	10	-	No internal connection.
V <sub>DD</sub>	13	11	Р	Positive power supply. This pin is the most positive power-supply potential. For reliable operation, connect a decoupling capacitor ranging from 0.1 $\mu$ F to 10 $\mu$ F between V <sub>DD</sub> and GND.
S2	14	12	I/O	Source pin 2. Can be an input or output.
D2	15	13	I/O	Drain pin 2. Can be an input or output.
SEL2	16	14	1	Logic control input 2.
_	-	EP	_	Exposed Pad. The exposed pad is electrically connected to $V_{SS}$ internally. Connect EP to $V_{SS}$ to achieve rated thermal and ESD performance.

(1) I = input, O = output, I/O = input and output, P = power



### 7 Specifications

#### 7.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)<sup>(1)</sup>

		MIN	MAX	UNIT
V <sub>DD</sub> to V <sub>SS</sub>			36	V
V <sub>DD</sub> to GND	Supply voltage	-0.3	18	V
V <sub>SS</sub> to GND		-18	0.3	V
V <sub>DIG</sub>	Digital input pin (SEL1, SEL2, SEL3, SEL4) voltage	GND -0.3	V <sub>DD</sub> +0.3	V
I <sub>DIG</sub>	Digital input pin (SEL1, SEL2, SEL3, SEL4) current	-30	30	mA
V <sub>ANA_IN</sub>	Analog input pin (Sx) voltage	V <sub>SS</sub> -0.3	V <sub>DD</sub> +0.3	V
I <sub>ANA_IN</sub>	Analog input pin (Sx) current	-30	30	mA
V <sub>ANA_OUT</sub>	Analog output pin (D) voltage	V <sub>SS</sub> -0.3	V <sub>DD</sub> +0.3	V
I <sub>ANA_OUT</sub>	Analog output pin (D) current	-30	30	mA
T <sub>A</sub>	Ambient temperature	<b>–</b> 55	150	°C
T <sub>J</sub>	Junction temperature		150	°C
T <sub>stg</sub>	Storage temperature	-65	150	°C

Stresses beyond those listed under Absolute Maximum Rating may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Condition. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

### 7.2 ESD Ratings

			VALUE	UNIT
V	Clastrostatic discharge	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001, all pins <sup>(1)</sup>	±2000	\/
V <sub>(ESD)</sub>	Electrostatic discharge	Charged device model (CDM), per JEDEC specification JESD22-C101, all pins (2)	±500	V

<sup>(1)</sup> JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

#### 7.3 Thermal Information

		TMUX6111/ TMU	TMUX6111/ TMUX6112/ TMUX6113			
	THERMAL METRIC	PW (TSSOP)	RTE (QFN)	UNIT		
		16 PINS	16 PINS			
$R_{\theta JA}$	Junction-to-ambient thermal resistance	111.0	TBD	°C/W		
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	41.7	TBD	°C/W		
$R_{\theta JB}$	Junction-to-board thermal resistance	57.2	TBD	°C/W		
$\Psi_{JT}$	Junction-to-top characterization parameter	4.1	TBD	°C/W		
$\Psi_{JB}$	Junction-to-board characterization parameter	56.6	TBD	°C/W		
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	N/A	TBD	°C/W		

### 7.4 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

		MIN	NOM MAX	UNIT
$V_{DD}$ to $V_{SS}$ <sup>(1)</sup>	Power supply voltage differential	10	33	٧
V <sub>DD</sub> to GND	Positive power supply voltage (singlle supply, V <sub>SS</sub> = 0 V)	10	16.5	٧
V <sub>DD</sub> to GND	Positive power supply voltage (dual supply)	5	16.5	V
V <sub>SS</sub> to GND	Negative power supply voltage (dual supply)	-5	-16.5	V
V <sub>S</sub> (2)	Source pins voltage	V <sub>SS</sub>	$V_{DD}$	V

<sup>(1)</sup>  $V_{DD}$  and  $V_{SS}$  can be any value as long as 10 V  $\leq$  ( $V_{DD} - V_{SS}$ )  $\leq$  33 V.

<sup>2)</sup> JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

<sup>(2)</sup> V<sub>S</sub> is the voltage on all the S pins.



### **Recommended Operating Conditions (continued)**

over operating free-air temperature range (unless otherwise noted)

		MIN	NOM MAX	UNIT
V <sub>D</sub> Drain p	n voltage	$V_{SS}$	$V_{DD}$	V
V <sub>DIG</sub> Digital i	nput pin (SEL1, SEL2, SEL3, SEL4) voltage	0	$V_{DD}$	V
I <sub>CH</sub> Channe	l current (T <sub>A</sub> = 25°C)	-25	25	V
T <sub>A</sub> Ambier	t temperature	-40	125	V

### 7.5 Electrical Characteristics (Dual Supplies: ±15 V)

at  $T_A = 25$ °C,  $V_{DD} = 15$  V, and  $V_{SS} = -15$  V (unless otherwise noted)

	PARAMETER	TEST CO	NDITIONS	MIN	TYP	MAX	UNIT
ANALOG	SWITCH			•			
V <sub>A</sub>	Analog signal range		$T_A = -40^{\circ}\text{C} \text{ to } +125^{\circ}\text{C}$	$V_{SS}$		$V_{DD}$	V
		$V_S = 0 \text{ V}, I_S = 1 \text{ mA}$			120	135	Ω
_					140	160	Ω
R <sub>ON</sub>	On-resistance	$V_S = \pm 10 \text{ V}, I_S = 1 \text{ mA}$	$T_A = -40$ °C to +85°C			210	Ω
			$T_A = -40^{\circ}\text{C} \text{ to } +125^{\circ}\text{C}$			245	Ω
					2.5	V <sub>DD</sub> 135 160 210	Ω
$\Delta R_{ON}$	On-resistance mismatch between channels	$V_S = \pm 10 \text{ V}, I_S = 1 \text{ mA}$	$T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$				Ω
	between channels		$T_A = -40^{\circ}\text{C to } +125^{\circ}\text{C}$			12	Ω
					23	33	Ω
R <sub>ON_FLAT</sub>	On-resistance flatness	$V_S = -10 \text{ V}, 0 \text{ V}, +10 \text{ V}, I_S$	$T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$			37	Ω
011_1 2111		= 1 mA	$T_A = -40^{\circ}\text{C to } +125^{\circ}\text{C}$			37	Ω
R <sub>ON_DRIFT</sub>	On-resistance drift	V <sub>S</sub> = 0 V			0.52		%/°C
ON_DIGIT I				-0.02	0.001	0.02	nA
I <sub>S(OFF)</sub>	Source off leakage current <sup>(1)</sup>	Switch state is off, $V_S = +10 \text{ V/} -10 \text{ V}$ , $V_D = -10 \text{ V/} + 10 \text{ V}$	$T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$	-0.13		0.05	nA
			$T_A = -40^{\circ}\text{C to } +125^{\circ}\text{C}$	-1			nA
		0 11 1 1 1 1 1 1 1 1	A	-0.05	0.005	0.05	nA
$I_{D(OFF)}$	Drain off leakage current(1)	Switch state is off, $V_S = +10 \text{ V/} -10 \text{ V}$ , $V_D = -10 \text{ V}$	$T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$	-0.14			nA
D(OIT)	S	V/ +10 V	$T_A = -40^{\circ}\text{C to } +125^{\circ}\text{C}$	-1		V <sub>DD</sub> 135 160 210 245 11 12 12 33 37 37 0.02 0.05 0.5 0.05 0.1 0.5 0.07 0.15 1 0.8	nA
		Outline state in an M	A	-0.07	0.008	160 210 245 11 12 13 37 37 0.02 0.05 0.5 0.05 0.1 0.5 0.07 0.15 1 0.8	nA
I <sub>D(ON)</sub>	Drain on leakage current	Switch state is on, $V_S = +10 \text{ V/} -10 \text{ V}$ , $V_D = -10 \text{ V}$	$T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$	-0.27		0.15	nA
D(ON)	3	V/ +10 V	$T_A = -40^{\circ}\text{C to } +125^{\circ}\text{C}$	-2		V <sub>DD</sub> 135 160 210 245 11 12 12 33 37 0.02 0.05 0.15 0.07 0.15 1 0.8	nA
DIGITAL II	NPUT (SELx pins)						
V <sub>IH</sub>	Logic voltage high			2			V
V <sub>IL</sub>	Logic voltage low					0.8	V
R <sub>PD(IN)</sub>	Pull-down resistance on SELx pins				6		МΩ
POWER S	UPPLY		I	1		ļ	
					17	21	μA
I <sub>DD</sub>	V <sub>DD</sub> supply current	$V_A = 0 \text{ V or } 3.3 \text{ V}, V_S = 0$	$T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$			0.1 0.5 008 0.07 0.15 1 0.8 6	μA
· <del>=</del>		V	$T_A = -40^{\circ}\text{C to } +125^{\circ}\text{C}$			22	μA
					8	10	<u>.</u> μΑ
I <sub>SS</sub>	V <sub>SS</sub> supply current	$V_A = 0 \text{ V or } 3.3 \text{ V, } V_S = 0$	$T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$			10	<u>.</u> μΑ
50	,,	V	$T_A = -40^{\circ}\text{C to } +125^{\circ}\text{C}$			11	μA

<sup>(1)</sup> When  $V_{\mbox{\scriptsize S}}$  is positive,  $V_{\mbox{\scriptsize D}}$  is negative, and vice versa.



### 7.6 Switching Characteristics (Dual Supplies: ±15 V)

at  $T_A = 25$ °C,  $V_{DD} = 15$  V, and  $V_{SS} = -15$  V (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
		$V_S = \pm 10 \text{ V}, R_L = 300 \Omega, C_L = 35 \text{ pF}$		66	77	ns
t <sub>ON</sub>	Enable turn-on time	$V_{S}$ = ±10 V, $R_{L}$ = 300 $\Omega$ , $C_{L}$ = 35 pF, $T_{A}$ = $-40^{\circ}C$ to +85°C			107	ns
		$V_{S}$ = ±10 V, $R_{L}$ = 300 $\Omega$ , $C_{L}$ = 35 pF, $T_{A}$ = $-40^{\circ}C$ to +125°C			117	ns
		$V_S = \pm 10 \text{ V}, R_L = 300 \Omega, C_L = 35 \text{ pF}$		56	68	ns
t <sub>OFF</sub>	Enable turn-off time	$\rm V_S=\pm 10~V,~R_L=300~\Omega$ , $\rm C_L=35~pF,~T_A=-40^{\circ}C$ to $+85^{\circ}C$			77	ns
		$\rm V_S=\pm 10~V,~R_L=300~\Omega$ , $\rm C_L=35~pF,~T_A=-40^{\circ}C$ to $\rm +125^{\circ}C$			81	ns
tBBM	Break-before-make time delay (TMUX6113 Only)	$\mbox{V}_{\mbox{\scriptsize S}}$ = 10 V, $\mbox{R}_{\mbox{\scriptsize L}}$ = 300 $\Omega$ , $\mbox{C}_{\mbox{\scriptsize L}}$ = 35 pF, $\mbox{T}_{\mbox{\scriptsize A}}$ = –40°C to +125°C	8	40		ns
0	Obana inication	$V_S = 0 \text{ V}, R_S = 0 \Omega, C_L = 1 \text{ nF}$		0.6		рС
QJ	Charge injection	$V_{S}$ = -15 V to 15 V, $R_{S}$ = 0 $\Omega$ , $C_{L}$ = 1 nF		1.3		рС
O <sub>ISO</sub>	Off-isolation	$R_L = 50 \Omega$ , $C_L = 5 pF$ , $f = 1 MHz$		-85		dB
		$R_L$ = 50 $\Omega$ , $C_L$ = 5 pF, f = 1 MHz, adjacent channel		-100		dB
X <sub>TALK</sub>	Channel-to-channel crosstalk	$R_{L}$ = 50 $\Omega$ , $C_{L}$ = 5 pF, f = 1 MHz, non-adjacentchannel		-115		dB
IL	Insertion loss	$R_L = 50 \Omega$ , $C_L = 5 pF$ , $f = 1 MHz$		-7.5		dB
ACPSRR	AC Power Supply Rejection Ratio	$R_L$ = 10 $k\Omega$ , $C_L$ = 5 pF, $V_{PP} {=}~0.62$ V on $V_{DD},$ f= 1 MHz		-59		dB
		$R_L$ = 10 $k\Omega$ , $C_L$ = 5 pF, $V_{PP}$ = 0.62 V on $V_{SS},f$ = 1 MHz		-59		dB
BW	-3dB Bandwidth	$R_L = 50 \Omega$ , $C_L = 5 pF$		800		MHz
THD	Total harmonic distortion + noise	$R_L = 10k \Omega$ , $C_L = 5 pF$ , $f = 20Hz$ to $20kHz$		0.08		%
C <sub>IN</sub>	Digital input capacitance	$V_{IN} = 0 \text{ V or } V_{DD}$		1.5		pF
0	Course off conscitones	V <sub>S</sub> = 0 V, f = 1 MHz (PW package)		1.9	3.0	pF
C <sub>S(OFF)</sub>	Source off-capacitance	V <sub>S</sub> = 0 V, f = 1 MHz (RTE package)		2.5	3.6	pF
C <sub>D(OFF)</sub>	Drain off-capacitance	V <sub>S</sub> = 0 V, f = 1 MHz		2.4	3.1	pF
C <sub>S(ON),</sub> C <sub>D(ON)</sub>	Source and drain on- capacitance	V <sub>S</sub> = 0 V, f = 1 MHz		4.2	6.0	pF

## 7.7 Electrical Characteristics (Single Supply: 12 V)

at  $T_A = 25^{\circ}$ C.  $V_{DD} = 12$  V. and  $V_{SS} = 0$  V (unless otherwise noted)

	PARAMETER	TEST CO	MIN	TYP	MAX	UNIT		
ANALOG SWITCH								
V <sub>A</sub>	Analog signal range	$T_A = -40^{\circ}\text{C to } +125^{\circ}\text{C}$	V <sub>SS</sub>		$V_{DD}$	V		
R <sub>ON</sub>					230	265	Ω	
	On-resistance	$V_S = 10 \text{ V}, I_S = 1 \text{ mA}$	$T_A = -40$ °C to +85°C			355	Ω	
			$T_A = -40^{\circ}\text{C to } +125^{\circ}\text{C}$			405	Ω	
ΔR <sub>ON</sub>	On-resistance mismatch between channels				5	18	Ω	
		$V_S = 10 \text{ V}, I_S = 1 \text{ mA}$	$T_A = -40$ °C to +85°C			18	Ω	
			$T_A = -40^{\circ}\text{C to } +125^{\circ}\text{C}$			19	Ω	
R <sub>ON_DRIFT</sub>	On-resistance drift	V <sub>S</sub> = 0 V			0.5		%/°C	
I <sub>S(OFF)</sub>				-0.02	0.001	0.02	nA	
	Source off leakage current <sup>(1)</sup>	Switch state is off, $V_S = 10 \text{ V}/1 \text{ V}$ , $V_D = 1 \text{ V}/10 \text{ V}$	$T_A = -40$ °C to +85°C	-0.1		0.05	nA	
		10 0/ 1 0, 00 = 1 0/ 10 0	$T_A = -40^{\circ}\text{C to } +125^{\circ}\text{C}$	-0.8		0.4	nA	

When  $V_S$  is positive,  $V_D$  is negative, and vice versa.



### Electrical Characteristics (Single Supply: 12 V) (continued)

at  $T_A = 25$ °C,  $V_{DD} = 12$  V, and  $V_{SS} = 0$  V (unless otherwise noted)

	PARAMETER	TEST CO	MIN	TYP	MAX	UNIT	
				-0.03	0.005	0.03	nA
$I_{D(OFF)}$	Drain off leakage current <sup>(1)</sup>	Switch state is off, $V_S = 10 \text{ V/ } 1 \text{ V}$ , $V_D = 1 \text{ V/ } 10 \text{ V}$	$T_A = -40$ °C to +85°C	-0.1		0.08	nA
		10 77 1 7, 70 - 1 77 10 7	$T_A = -40^{\circ}C \text{ to } +125^{\circ}C$	-0.8		0.4	nA
				-0.05	0.008	0.05	nA
$I_{D(ON)}$	Drain on leakage current	Drain on leakage current Switch state is on, $V_S = \frac{1}{100}$	$T_A = -40$ °C to +85°C	-0.2		0.15	nA
			$T_A = -40^{\circ}\text{C} \text{ to } +125^{\circ}\text{C}$	-1.6		0.8	nA
DIGITAL	INPUT (SELx pins)						
$V_{IH}$	Logic voltage high			2			V
V <sub>IL</sub>	Logic voltage low					8.0	V
R <sub>PD(EN)</sub>	Pull-down resistance on SELx pins					6	МΩ
POWER S	SUPPLY						
					13	16	μΑ
$I_{DD}$	V <sub>DD</sub> supply current	$V_A = 0 \text{ V or } 3.3 \text{ V}, V_S = 0$	$T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$			16	μΑ
		v	$T_A = -40^{\circ}\text{C to } +125^{\circ}\text{C}$			17	μΑ

### 7.8 Switching Characteristics (Single Supply: 12 V)

at  $T_A = 25$ °C,  $V_{DD} = 12$  V, and  $V_{SS} = 0$  V (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
		$V_S$ = 8 V, $R_L$ = 300 $\Omega$ , $C_L$ = 35 pF		72	84	ns
t <sub>ON</sub>	Enable turn-on time	$\mbox{V}_{\mbox{\scriptsize S}}=8$ V, $\mbox{R}_{\mbox{\scriptsize L}}=300~\Omega$ , $\mbox{C}_{\mbox{\scriptsize L}}=35$ pF, $\mbox{T}_{\mbox{\scriptsize A}}=-40\mbox{\sc oc}$ to +85°C			117	ns
		$\rm V_S = 8~V,~R_L = 300~\Omega$ , $\rm C_L = 35~pF,~T_A = -40^{\circ}C$ to $+125^{\circ}C$			128	ns
		$V_S$ = 8 V, $R_L$ = 300 $\Omega$ , $C_L$ = 35 pF		57	66	ns
t <sub>OFF</sub>	Enable turn-off time	$\mbox{V}_{\mbox{S}}=\mbox{8}$ V, $\mbox{R}_{\mbox{L}}=300~\Omega$ , $\mbox{C}_{\mbox{L}}=35~\mbox{pF},$ $\mbox{T}_{\mbox{A}}=-40\mbox{°C}$ to +85°C			78	ns
		$\mbox{V}_{\mbox{\scriptsize S}}=\mbox{8 V}, \mbox{ R}_{\mbox{\scriptsize L}}=\mbox{300 }\Omega$ , $\mbox{C}_{\mbox{\scriptsize L}}=\mbox{35 pF}, \mbox{ T}_{\mbox{\scriptsize A}}=-40^{\circ}\mbox{C}$ to +125°C			84	ns
t <sub>BBM</sub>	Break-before-make time delay (TMUX6113 only)	$\rm V_S = 8~V,~R_L = 300~\Omega$ , $\rm C_L = 35~pF,~T_A = -40^{\circ}C$ to $+125^{\circ}C$	17	47		ns
QJ	Change injection	$V_S = 6 \text{ V}, R_S = 0 \Omega, C_L = 1 \text{ nF}$		-0.7		рС
	Charge injection	$V_S$ = 0 V to 12 V, $R_S$ = 0 $\Omega$ , $C_L$ = 1 nF		-0.7		рС
O <sub>ISO</sub>	Off-isolation	$R_L$ = 50 $\Omega$ , $C_L$ = 5 pF, f = 1 MHz		-86		dB
		$R_L$ = 50 $\Omega$ , $C_L$ = 5 pF, f = 1 MHz, adjacent channel		-98		dB
X <sub>TALK</sub>	Channel-to-channel crosstalk	$R_{L}$ = 50 $\Omega$ , $C_{L}$ = 5 pF, f = 1 MHz, non-adjacent channel		-117		dB
IL	Insertion loss	$R_L = 50 \Omega$ , $C_L = 5 pF$ , $f = 1 MHz$		-15		dB
ACPSRR	AC Power Supply Rejection Ratio	$R_L$ = 10 k $\Omega$ , $C_L$ = 5 pF, $V_{PP}$ = 0.62 V, f= 1 MHz		<b>–</b> 59		dB
BW	-3dB Bandwidth	$R_L = 50 \Omega$ , $C_L = 5 pF$		750		MHz
C <sub>IN</sub>	Digital input capacitance	$V_{IN} = 0 \text{ V or } V_{DD}$		1.6		pF
C <sub>S(OFF)</sub>	Source off conneitance	V <sub>S</sub> = 6 V, f = 1 MHz (PW package)		2.2	3.1	pF
	Source off-capacitance	V <sub>S</sub> = 6 V, f = 1 MHz (RTE package)		2.9	4.0	pF
C <sub>D(OFF)</sub>	Drain off-capacitance	V <sub>S</sub> = 6 V, f = 1 MHz		2.8	3.5	pF
C <sub>S(ON)</sub> , C <sub>D(ON)</sub>	Source and drain on- capacitance	V <sub>S</sub> = 6 V, f = 1 MHz		4.6	6.3	pF



# 8 Parameter Measurement Information

#### 8.1 Truth Tables

Table 1, Table 2, Table 3and show the truth tables for the TMUX6111, TMUX6112, and TMUX6113, respectively.

Table 1. TMUX6111 Truth Table

SELx	STATE
0	All Switch ON
1	All Switch OFF

Table 2. TMUX6112 Truth Table

SELx	STATE
0	All Switch OFF
1	All Switch ON

Table 3. TUMUX6113 Truth Table

SELx	STATE
0	Switch 1, 4 OFF Switch 2, 3 ON
1	Switch 1, 4 ON Switch 2, 3 OFF



#### 9 Detailed Description

#### 9.1 Overview

The TMUX6111, TMUX6112, and TMUX6113 are 4-channel single-pole/ single-throw (SPDT) switches that supports dual supplies (±5 V to ±16.5 V) or single supply (10 V to 16.5 V) operation. Each channel of the switch is turned on or turned off based on the state of its corresponding SELx pin. The Functional Block Diagram section provides a top-level block diagram of the switches.

#### 9.1.1 On-Resistance

The on-resistance of the TMUX611x is the ohmic resistance across the source (Sx) and drain (Dx) pins of the device. The on-resistance varies with input voltage and supply voltage. The symbol  $R_{ON}$  is used to denote on-resistance. The measurement setup used to measure  $R_{ON}$  is shown in Figure 1. Voltage (V) and current ( $I_{CH}$ ) are measured using this setup, and  $R_{ON}$  is computed as shown in Equation 1:

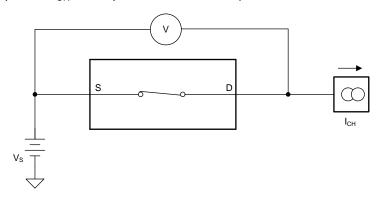


Figure 1. On-Resistance Measurement Setup

$$R_{ON} = V / I_{CH}$$
 (1)

#### 9.1.2 Off-Leakage Current

There are two types of leakage currents associated with a switch during the off state:

- 1. Source off-leakage current
- 2. Drain off-leakage current

Source leakage current is defined as the leakage current flowing into or out of the source pin when the switch is off. This current is denoted by the symbol  $I_{S(OFF)}$ .

Drain leakage current is defined as the leakage current flowing into or out of the drain pin when the switch is off. This current is denoted by the symbol  $I_{D(OFF)}$ .

The setup used to measure both off-leakage currents is shown in Figure 2

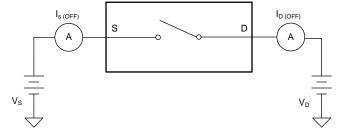


Figure 2. Off-Leakage Measurement Setup



#### 9.1.3 On-Leakage Current

On-leakage current is defined as the leakage current that flows into or out of the drain pin when the switch is in the on state. The source pin is left floating during the measurement. Figure 3 shows the circuit used for measuring the on-leakage current, denoted by  $I_{D(ON)}$ .

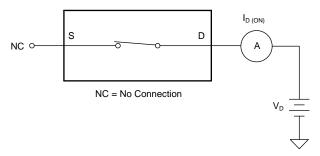


Figure 3. On-Leakage Measurement Setup

#### 9.1.4 Break-Before-Make Delay

The break-before-make delay is a safety feature of the TMUX6113 switch. The TMUX6113's ON switches first break the connection before the OFF switches make connection. The time delay between the *break* and the *make* is known as break-before-make delay. Figure 4 shows the setup used to measure break-before-make delay, denoted by the symbol t<sub>BBM</sub>.

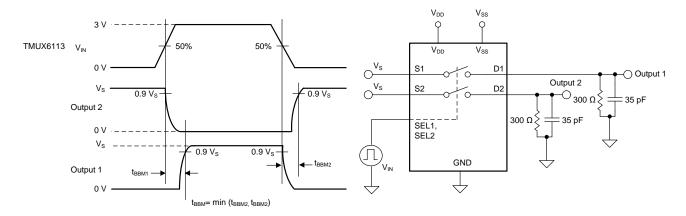


Figure 4. Break-Before-Make Delay Measurement Setup

#### 9.1.5 Turn-On and Turn-Off Time

Turn-on time is defined as the time taken by the output of the TMUX611x to rise to a 90% final value after the SELx signal has risen (for NC switches) or fallen (for NO switches) to a 50% final value. Figure 5 shows the setup used to measure turn-on time. Turn-on time is denoted by the symbol  $t_{\rm ON}$ .

Turn off time is defined as the time taken by the output of the TMUX611x to fall to a 10% initial value after the SELx signal has fallen (for NC switches) or risen (for NO switches) to a 50% initial value. Figure 5 shows the setup used to measure turn-off time. Turn-off time is denoted by the symbol t<sub>OFF</sub>.



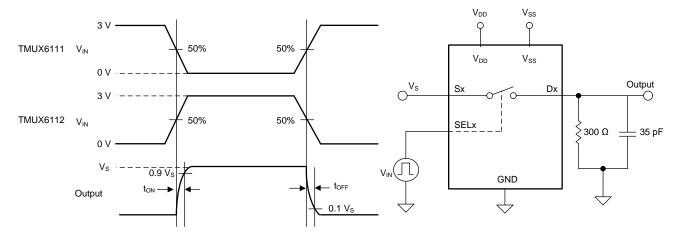


Figure 5. Turn-On and Turn-Off Time Measurement Setup

#### 9.1.6 Charge Injection

The TMUX611x have a simple transmission-gate topology. Any mismatch in capacitance between the NMOS and PMOS transistors results in a charge injected into the drain or source during the falling or rising edge of the gate signal. The amount of charge injected into the source or drain of the device is known as charge injection, and is denoted by the symbol  $Q_{\text{INJ}}$ . Figure 6 shows the setup used to measure charge injection.

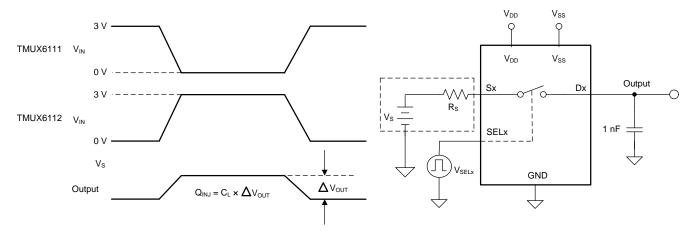


Figure 6. Charge-Injection Measurement Setup

#### 9.1.7 Off Isolation

Off isolation is defined as the voltage at the drain pin (Dx) of the TMUX611x when a 1- $V_{RMS}$  signal is applied to the source pin (Sx) of an OFF switch. Figure 7 shows the setup used to measure off isolation. Use Equation 2 to compute off isolation.



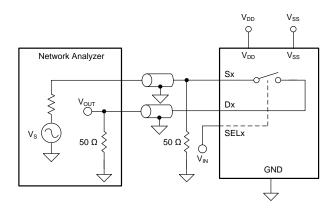


Figure 7. Off Isolation Measurement Setup

Off Isolation = 
$$20 \cdot \text{Log}\left(\frac{V_{\text{OUT}}}{V_{\text{S}}}\right)$$
 (2)

#### 9.1.8 Channel-to-Channel Crosstalk

Channel-to-channel crosstalk is defined as the voltage at the source pin (Sx) of an off-channel, when a  $1-V_{RMS}$  signal is applied at the source pin of an on-channel. Figure 8 shows the setup used to measure, and Equation 3 is the equation used to compute, channel-to-channel crosstalk.

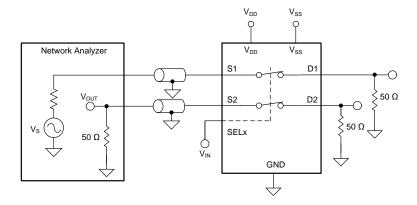


Figure 8. Channel-to-Channel Crosstalk Measurement Setup

Channel-to-Channel Crosstalk = 
$$20 \cdot Log\left(\frac{V_{OUT}}{V_{S}}\right)$$
 (3)

#### 9.1.9 Bandwidth

Bandwidth is defined as the range of frequencies that are attenuated by < 3 dB when the input is applied to the source pin (Sx) of an on-channel, and the output is measured at the drain pin (Dx) of the TMUX611x. Figure 9 shows the setup used to measure bandwidth of the switch. Use Equation 4 to compute the attenuation.



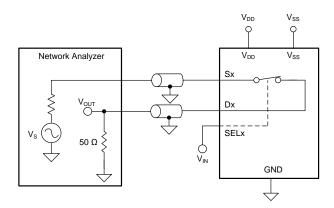


Figure 9. Bandwidth Measurement Setup

Attenuation = 
$$20 \cdot \text{Log}\left(\frac{V_2}{V_1}\right)$$

#### 9.1.10 THD + Noise

The total harmonic distortion (THD) of a signal is a measurement of the harmonic distortion, and is defined as the ratio of the sum of the powers of all harmonic components to the power of the fundamental frequency at the mux output. The on-resistance of the TMUX611x varies with the amplitude of the input signal and results in distortion when the drain pin is connected to a low-impedance load. Total harmonic distortion plus noise is denoted as THD+N. Figure 10 shows the setup used to measure THD+N of the TMUX611x.

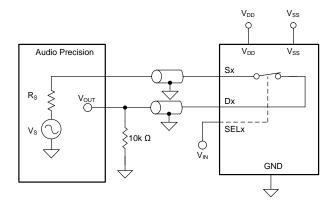
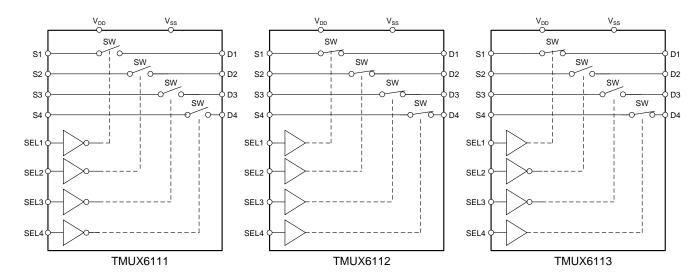


Figure 10. THD+N Measurement Setup



#### 9.2 Functional Block Diagram



### 9.3 Feature Description

### 9.3.1 Ultra-low Leakage Current

The TMUX611x provide extremely low on- and off-leakage currents. The devices are capable of switching signals from high source-impedance inputs into a high input-impedance op amp with minimal offset error because of the ultralow leakage currents. Figure 11 shows typical leakage currents of the devices versus temperature.



Figure 11. Leakage Current vs Temperature

#### 9.3.2 Ultra-low Charge Injection

The TMUX611x are implemented with simple transmission gate topology, as shown in Figure 12. Any mismatch in the stray capacitance associated with the NMOS and PMOS causes an output level change whenever the switch is opened or closed. The devices utilize special charge-injection cancellation circuitry that reduces the source (Sx)-to-drain (Dx) charge injection to as low as -0.6 pC at  $V_S = 0$  V, as shown in Figure 13.



#### **Feature Description (continued)**

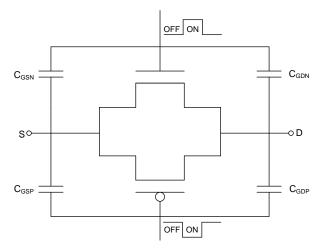


Figure 12. Transmission Gate Topology



Figure 13. Source-to-Drain Charge Injection vs Source or Drain Voltage

#### 9.3.3 Bidirectional and Rail-to-Rail Operation

The TMUX611x conducts equally well from source (Sx) to drain (Dx) or from drain (Dx) to source (Sx). Each channel of the switches has very similar characteristics in both directions. The input signal to the devices swings from  $V_{SS}$  to  $V_{DD}$  without any significant degradation in performance. The on-resistance of these devices varies with input signal.

#### 9.4 Device Functional Modes

Each channel of the TMUX611x is turned on or turned off based on the state of its corresponding SELx pin. The SELx pins are weakly pulled-down through an internal 6M ohm resistor, allowing the switches to stay in a determined state when power is applies to the devices. The SELx pins can be connected to  $V_{DD}$ .



### 10 Application and Implementation

#### NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

### 10.1 Application Information

The TMUX611x family offers outstanding input/output leakage currents and ultralow charge injection. These devices operate up to 33 (dual supply) or 16.5V (single supply), and offer true rail-to-rail input and output. The on-capacitance of the TMUX611x is very low. These features makes the TMUX611x a family of precision, robust, high-performance analog multiplexer for high-voltage, industrial applications.

#### 10.2 Typical Application

One useful application to take advantage of TMUX611x family's precision performance is the sample and hold circuit. A sample and hold circuit can be useful for an analog to digital converter (ADC) to sample a varying input voltage with improved reliability and stability. It can also be used to store the output samples from a single digital-to-analog converter (DAC) in a multi-output application. A simple sample and hold circuit can be realized using an analog switch like one of the TMUX611x analog switches.

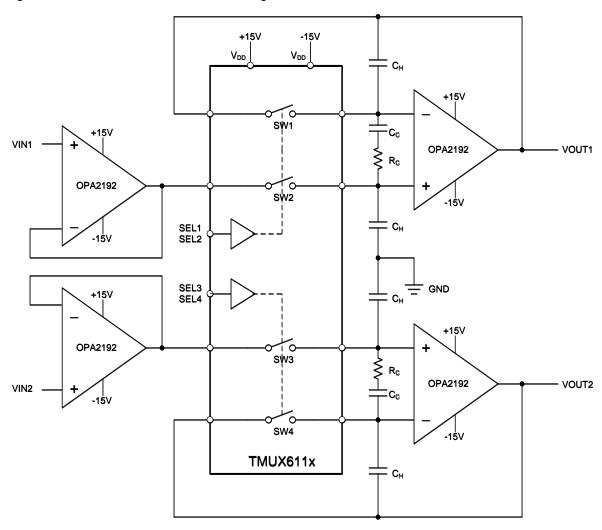


Figure 14. A 2-output Sample and Hold Circuit Realized Using the TMUX611x Analog Switch



#### **Typical Application (continued)**

#### 10.2.1 Design Requirements

The purpose of this precision design is to implement an optimized 2-output sample and hold circuit using a 4-channel SPST switch. The sample and hold circuit needs to be capable of supporting high voltage output swing up to ± 15V with minimized pedestal error and fast settling time. The overall system block diagram is illustrated in Figure 14.

### 10.2.2 Detailed Design Procedure

The TMUX611x switch is used in conjunction with the voltage holding capacitors (C<sub>H</sub>) to implement the sample and hold circuit. The basic operation is:

- 1. When the switch (SW2 or SW3) is closed, it samples the input voltage and charges the holding capacitors (C<sub>H</sub>) to the input voltages values.
- 2. When the switch (SW2 or SW3) is open, the holding capacitors (C<sub>H</sub>) holds its previous value, maintaining stable voltage at the amplifier output (V<sub>OUT</sub>).

Ideally, the switch delivers only the input signals to the holding capacitors. However, when the switch gets toggled, some amount of charge also gets transferred to the switch output in the form of charge injection, resulting slight sampling error. The TMUX611x switch family has excellent charge injection performance of only 0.6 pC, making it an ideal choice for this implementation to minimize sampling error.

Due to switch and capacitor leakage current, the voltage on the hold capacitors droops with time. The TMUX611x minimizes the droops due to its ultra-low leakage performance. At 25°C, the TMUX611x has extremely tiny leakage current at 1pA typical and 20pA max.

The TMUX611x device is also selected due to its high voltage capability. The device supports up to  $\pm$  16.5 V dual supply operation, making it an ideal solution in this high voltage sample and hold application.

A second switch SW1 (or SW4) is also included to operate in parallel with SW2 (or SW3) to reduce pedestal error during switch toggling. Because both switches are driven at the same potential, they act as common-mode signal to the op-amp, thereby minimizing the charge injection effects caused by the switch toggling action. Compensation network consisting of  $R_C$  and  $C_C$  is also added to further reduce the pedestal error, whiling reducing the hold-time glitch and improving the settling time of the circuit.



### 11 Power Supply Recommendations

The TMUX611x operates across a wide supply range of  $\pm 5$  V to  $\pm 16.5$  V (10 V to 16.5 V in single-supply mode). They also perform well with asymmetrical supplies such as  $V_{DD}=12$  V and  $V_{SS}=-5$  V. For improved supply noise immunity, use a supply decoupling capacitor ranging from 0.1  $\mu$ F to 10  $\mu$ F at both the  $V_{DD}$  and  $V_{SS}$  pins to ground. Always ensure the ground (GND) connection is established before supplies are ramped. As a best practice, it is recommended to ramp  $V_{SS}$  first before  $V_{DD}$  in dual or asymmetrical supply applications.

The on-resistance of the devices varies with supply voltage, as illustrated in Figure 15



Figure 15. On-Resistance Variation With Supply and Input Voltage



### 12 Layout

#### 12.1 Layout Guidelines

Figure 16 illustrates an example of a PCB layout with the TMUX6112PW. Some key considerations are:

- Decouple the V<sub>DD</sub> and V<sub>SS</sub> pins with a 0.1-μF capacitor, placed as close to the pin as possible. Make sure
  that the capacitor voltage rating is sufficient for the V<sub>DD</sub> and V<sub>SS</sub> supplies.
- · Keep the input lines as short as possible.
- Use a solid ground plane to help distribute heat and reduce electromagnetic interference (EMI) noise pickup.
- Do not run sensitive analog traces in parallel with digital traces. Avoid crossing digital and analog traces if possible, and only make perpendicular crossings when necessary.

#### 12.2 Layout Example

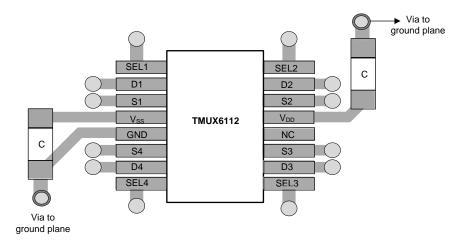


Figure 16. TMUX611x Layout Example



### 13 Device and Documentation Support

#### 13.1 Documentation Support

#### 13.1.1 Related Documentation

- ADS8664 12-Bit, 500-kSPS, 4- and 8-Channel, Single-Supply, SAR ADCs with Bipolar Input Ranges (SBAS492)
- OPA140 High-Precision, Low-Noise, Rail-to-Rail Output, 11-MHz JFET Op Amp (SBOS498)
- OPA192 36-V, Precision, Rail-to-Rail Input/Output, Low Offset Voltage, Low Input Bias Current Op Amp with e-Trim™ (SBOS620)

#### 13.2 Related Links

The table below lists quick access links. Categories include technical documents, support and community resources, tools and software, and quick access to order now.

Table 4. Related Links

PARTS	PRODUCT FOLDER	ORDER NOW	TECHNICAL DOCUMENTS	TOOLS & SOFTWARE	SUPPORT & COMMUNITY	
TMUX6111	Click here	Click here	Click here	Click here	Click here	
TMUX6112	Click here	Click here	Click here	Click here	Click here	
TMUX6113	Click here	Click here	Click here	Click here	Click here	

#### 13.3 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

#### 13.4 Community Resources

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

TI E2E™ Online Community TI's Engineer-to-Engineer (E2E) Community. Created to foster collaboration among engineers. At e2e.ti.com, you can ask questions, share knowledge, explore ideas and help solve problems with fellow engineers.

**Design Support** *TI's Design Support* Quickly find helpful E2E forums along with design support tools and contact information for technical support.

#### 13.5 Trademarks

E2E is a trademark of Texas Instruments.

#### 13.6 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

#### 13.7 Glossary

SLYZ022 — TI Glossary.

This glossary lists and explains terms, acronyms, and definitions.



### 14 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.





28-Aug-2018

#### PACKAGING INFORMATION

Orderable Device	Status	Package Type	_	Pins	Package	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Device Marking	Samples
	(1)		Drawing		Qty	(2)	(6)	(3)		(4/5)	
PTMUX6112PWR	ACTIVE	TSSOP	PW	16	2000	TBD	Call TI	Call TI	-40 to 125		Samples
TMUX6111PWR	PREVIEW	TSSOP	PW	16	2000	TBD	Call TI	Call TI	-40 to 125		
TMUX6111RTER	PREVIEW	WQFN	RTE	16	3000	TBD	Call TI	Call TI	-40 to 125		
TMUX6112PWR	PREVIEW	TSSOP	PW	16	2000	TBD	Call TI	Call TI	-40 to 125		
TMUX6112RTER	PREVIEW	WQFN	RTE	16	3000	TBD	Call TI	Call TI	-40 to 125		
TMUX6113PWR	PREVIEW	TSSOP	PW	16	2000	TBD	Call TI	Call TI	-40 to 125		
TMUX6113RTER	PREVIEW	WQFN	RTE	16	3000	TBD	Call TI	Call TI	-40 to 125		

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) **RoHS**: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

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Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

- (3) MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead/Ball Finish Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

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### PACKAGE OPTION ADDENDUM

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PW (R-PDSO-G16)

### PLASTIC SMALL OUTLINE



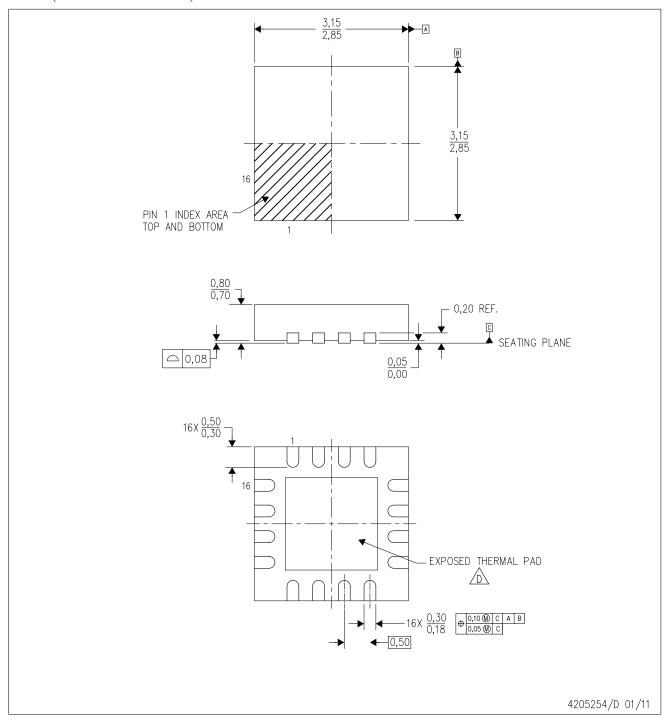
NOTES:

- A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M—1994.
- B. This drawing is subject to change without notice.
- Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0,15 each side.
- Body width does not include interlead flash. Interlead flash shall not exceed 0,25 each side.
- E. Falls within JEDEC MO-153



# RTE (S-PWQFN-N16)

## PLASTIC QUAD FLATPACK NO-LEAD



NOTES:

- A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M—1994.
- B. This drawing is subject to change without notice.
- C. Quad Flatpack, No-leads (QFN) package configuration.
- The package thermal pad must be soldered to the board for thermal and mechanical performance. See the Product Data Sheet for details regarding the exposed thermal pad dimensions.
- E. Falls within JEDEC MO-220.



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