

DATA SHEET

TDA5636; TDA5637 9 V VHF, hyperband and UHF mixers- oscillators for TV and VCR 3-band tuners

Product specification
Supersedes data of 1995 Mar 21
File under Integrated Circuits, IC02

1996 Jun 10

9 V VHF, hyperband and UHF mixers-oscillators for TV and VCR 3-band tuners

TDA5636; TDA5637

FEATURES

- Balanced mixer with a common emitter input for band A (single input)
- 2-pin oscillator for bands A and B
- Balanced mixer with a common base input for bands B and C (balanced input)
- 4-pin oscillator for band C
- Local oscillator buffer output for external prescaler
- SAW filter preamplifier with a low output impedance to drive a 75 Ω load
- Band gap voltage stabilizer for oscillator stability
- Electronic band switch
- External IF filter between the mixer output and the IF amplifier input
- Pin to pin compatible with TDA5636B; TDA5637B family (same function with symmetrical IF output).

APPLICATIONS

- 3-band all-channel TV and VCR tuners
- Any standard.

GENERAL DESCRIPTION

The TDA5636 and TDA5637 are monolithic integrated circuits that perform the mixer/oscillator functions for bands A, B and C in TV and VCR tuners. These low-power mixer/oscillators require a power supply of 9 V and are available in a very small package.

The devices give the designer the capability to design an economical and physically small 3-band tuner.

They are suitable for European standards, as illustrated in Fig.17, with the following RF bands:

- 48.25 to 168.25 MHz
- 175.25 to 447.25 MHz
- 455.25 to 855.25 MHz.

With an appropriate tuned circuit, they are also suitable for NTSC all-channel tuners (USA and Japan). The tuner development time can be drastically reduced by using these devices.

These circuits belong to the TDA5636B/TDA5637B family which has exactly the same function with an IF amplifier having a symmetrical IF output to drive a SAW filter directly. It is possible to build tuners with either an asymmetrical or a symmetrical IF output with one main tuner lay-out.

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QUICK REFERENCE DATA

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
V_P	supply voltage		–	9.0	–	V
I_P	supply current	band A	–	45	–	mA
		band B	–	41	–	mA
		band C	–	44	–	mA
f_{RF}	frequency range	RF input; band A; note 1	45	–	180	MHz
		RF input; band B; note 1	160	–	470	MHz
		RF input; band C; note 1	430	–	860	MHz
G_v	voltage gain	band A; $R_L = 75 \Omega$	–	20	–	dB
		band B; $R_L = 75 \Omega$	–	31	–	dB
		band C; $R_L = 75 \Omega$	–	31	–	dB
NF	noise figure	band A; $R_L = 75 \Omega$	–	7.5	–	dB
		band B; $R_L = 75 \Omega$	–	6	–	dB
		band C; $R_L = 75 \Omega$	–	7	–	dB
V_o	output voltage to get 1% cross modulation in channel	band A; $R_L = 75 \Omega$	–	110	–	dB μ V
		band B; $R_L = 75 \Omega$	–	110	–	dB μ V
		band C; $R_L = 75 \Omega$	–	110	–	dB μ V

Note

- The limits are related to the tank circuits used in Fig.17 and the intermediate frequency. Frequency bands may be adjusted by the choice of external components.

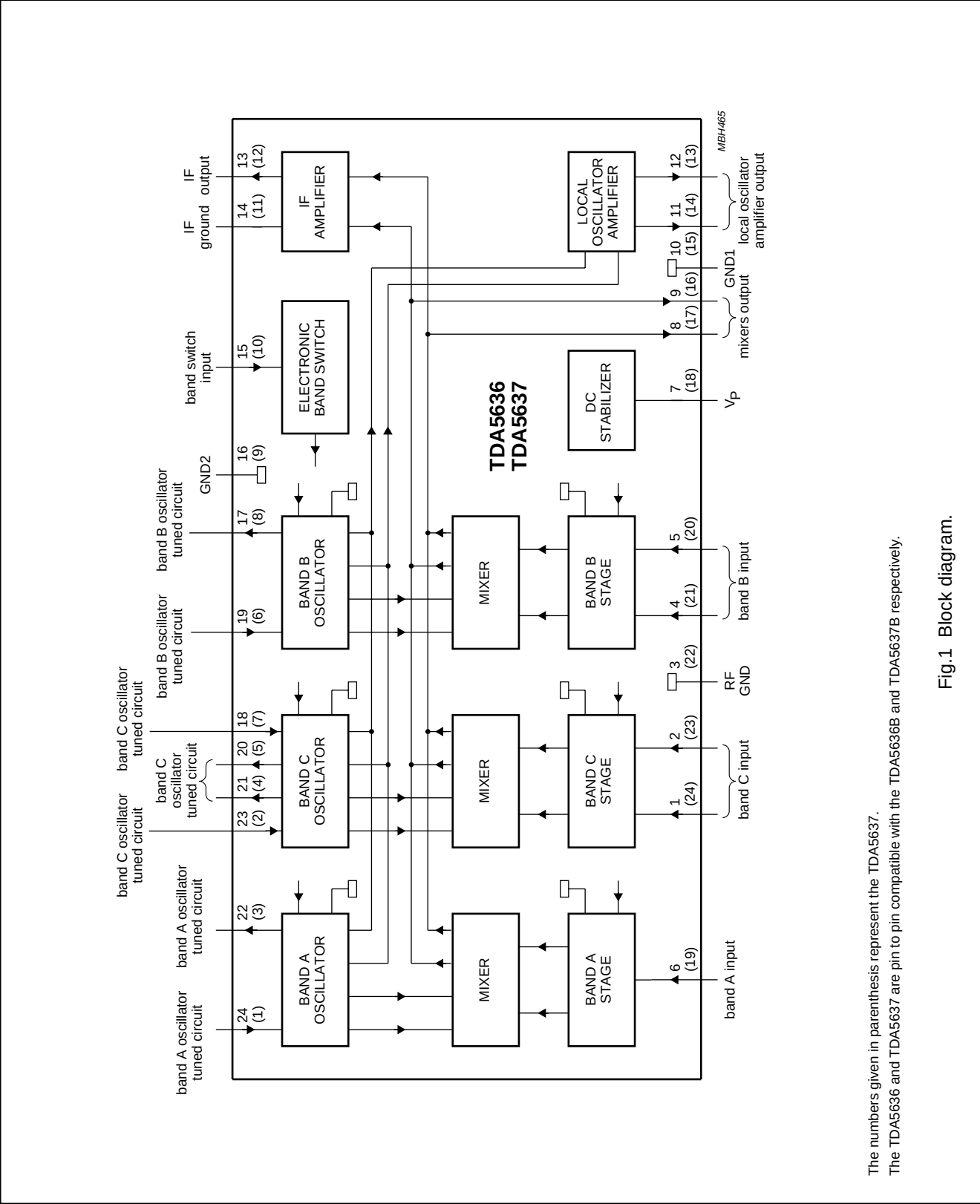
ORDERING INFORMATION

TYPE NUMBER	PACKAGE		
	NAME	DESCRIPTION	VERSION
TDA5636T	SO24	plastic small outline package; 24 leads; body width 7.5 mm	SOT137-1
TDA5636M	SSOP24	plastic shrink small outline package; 24 leads; body width 5.3 mm	SOT340-1
TDA5637T	SO24	plastic small outline package; 24 leads; body width 7.5 mm	SOT137-1
TDA5637M	SSOP24	plastic shrink small outline package; 24 leads; body width 5.3 mm	SOT340-1

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BLOCK DIAGRAM



The numbers given in parenthesis represent the TDA5637.
The TDA5636 and TDA5637 are pin compatible with the TDA5636B and TDA5637B respectively.

Fig.1 Block diagram.

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PINNING

SYMBOL	PIN		DESCRIPTION
	TDA5636	TDA5637	
CIN1	1	24	band C input 1
CIN2	2	23	band C input 2
RFGND	3	22	ground for RF inputs
BIN1	4	21	band B input 1
BIN2	5	20	band B input 2
AIN	6	19	band A input
V _P	7	18	supply voltage
MIXOUT1	8	17	mixers output 1
MIXOUT2	9	16	mixers output 2
GND1	10	15	ground 1 (0 V)
LOOUT1	11	14	local oscillator amplifier output 1
LOOUT2	12	13	local oscillator amplifier output 2
IFOUT	13	12	IF amplifier output
IFGND	14	11	IF amplifier ground
BS	15	10	electronic band switch input
GND2	16	9	ground 2 (0 V)
BOSCOC	17	8	band B oscillator output collector
COSCIB1	18	7	band C oscillator input base 1
BOSCIB	19	6	band B oscillator input base
COSCOC1	20	5	band C oscillator output collector 1
COSCOC2	21	4	band C oscillator output collector 2
AOSCOC	22	3	band A oscillator output collector
COSCIB2	23	2	band C oscillator input base 2
AOSCIB	24	1	band A oscillator input base

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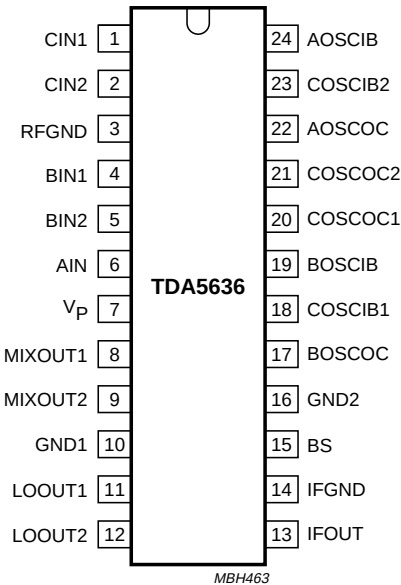


Fig.2 Pin configuration (TDA5636).

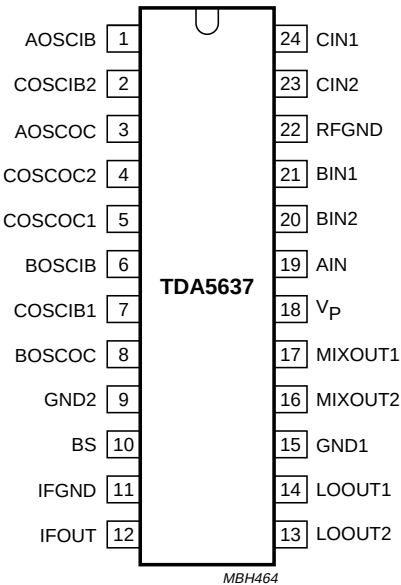


Fig.3 Pin configuration (TDA5637).

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LIMITING VALUES

In accordance with the Absolute Maximum Rating System (IEC 134).

SYMBOL	PARAMETER	MIN.	MAX.	UNIT
V_P	supply voltage	-0.3	+10.5	V
V_{SW}	switching voltage	0	10.5	V
I_O	output current of each pin to ground	-	-10	mA
$t_{sc(max)}$	maximum short-circuit time (all pins)	-	10	s
T_{stg}	IC storage temperature	-55	+150	°C
T_{amb}	operating ambient temperature	-10	+80	°C
T_j	junction temperature	-	+150	°C

THERMAL CHARACTERISTICS

SYMBOL	PARAMETER	TYPICAL VALUE	UNIT
$R_{th\ j-a}$	thermal resistance from junction to ambient in free air		
	SSOP	120	K/W
	SO	75	K/W

HANDLING

Human Body Model

For TDA5636 GND (10, 16), RFGND (3), IFGND (14) and V_P (7) are separate.

For TDA5637 GND (9, 15), RFGND (22), IFGND (11) and V_P (18) are separate.

All pins withstand 2000 V in accordance with the "UZW-BO/FQ-A302" specification equivalent to the "MIL-STD-883C" category B (2000 V).

Machine Model

All pins withstand 200 V in accordance with the "UZW-BO/FQ-B302 (issue date: Nov 6th, 1990)" specification.

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CHARACTERISTICS

$V_P = 9\text{ V}$; $T_{\text{amb}} = 25\text{ °C}$; measured in circuit of Fig.17; unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Supply						
V _P	supply voltage		8.1	9.0	9.9	V
I _P	supply current	band A	–	45	50.5	mA
		band B	–	41	46	mA
		band C	–	44	49	mA
V _{SW}	switching voltage	band A	0	–	1.1	V
		band B	1.6	–	2.4	V
		band C	3.0	–	V _P	V
I _{SW}	switching current	band A	–	–	2	μA
		band B	–	–	5	μA
		band C; V _{SW(C)} = 5 V	–	–	10	μA
Band A mixer (including IF amplifier)						
f _{RF}	frequency range	note 1	45	–	180	MHz
G _V	voltage gain	R _L = 75 Ω; f _{RF} = 50 MHz; see Fig.4	17.5	20.0	22.5	dB
		R _L = 75 Ω; f _{RF} = 180 MHz; see Fig.4	17.5	20.0	22.5	dB
NF	noise figure	R _L = 75 Ω; f _{RF} = 50 MHz; see Figs 5 and 6.	–	7.5	9.5	dB
		R _L = 75 Ω; f _{RF} = 180 MHz; see Figs 5 and 6.	–	7.5	9.5	dB
V _O	output voltage	1% cross-modulation in channel; R _L = 75 Ω; f _{RF} = 50 MHz; see Fig.7	107	110	–	dBμV
		1% cross-modulation in channel; R _L = 75 Ω; f _{RF} = 180 MHz; see Fig.7	107	110	–	dBμV
V _i	input voltage	10 kHz pulling in channel; f _{RF} = 180 MHz; note 6	–	100	–	dBμV
g _{os}	optimum source conductance	f _{RF} = 50 MHz	–	0.5	–	mS
		f _{RF} = 180 MHz	–	1	–	mS
Y _i	input admittance	see Fig.12	–	–	–	mS
C _i	input capacitance	f _{RF} = 50 to 180 MHz; see Fig.12	–	2	–	pF
Band A oscillator						
f _{OSC}	frequency range	note 2	80	–	216	MHz
f _{shift}	frequency shift	ΔV _p = 10%; note 3	–	–	200	kHz
f _{drift}	frequency drift	ΔT = 25 °C with no compensation; NP0 capacitors; note 4	–	–	600	kHz
		5 s to 15 min after switching on; note 5	–	–	200	kHz

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Band B mixer (including IF amplifier)						
f _{RF}	frequency range	note 1	160	–	470	MHz
G _V	voltage gain	R _L = 75 Ω; f _{RF} = 170 MHz; see Fig.8	28	31	34	dB
		R _L = 75 Ω; f _{RF} = 470 MHz; see Fig.8	28	31	34	dB
NF	noise figure (not corrected for image)	R _L = 75 Ω; f _{RF} = 170 MHz; see Fig.9	–	6.0	8.0	dB
		R _L = 75 Ω; f _{RF} = 470 MHz; see Fig.9	–	7.0	9.0	dB
V _O	output voltage	1% cross-modulation in channel; R _L = 75 Ω; f _{RF} = 170 MHz; see Fig.10	107	110	–	dBμV
		1% cross-modulation in channel; R _L = 75 Ω; f _{RF} = 470 MHz; see Fig.10	107	110	–	dBμV
V _i	input voltage	10 kHz pulling in channel; f _{RF} = 470 MHz; note 6	–	91	–	dBμV
	input voltage	N + 5 – 1 MHz pulling; f _{RF} = 430 MHz; see Fig.11	–	77	–	dBμV
Z _i	input impedance (R _s + jL _s ω)	f _{RF} = 170 to 470 MHz; see Fig.13	–	30	–	Ω
		f _{RF} = 170 to 470 MHz; see Fig.13	–	10	–	nH
Band B oscillator						
f _{OSC}	frequency range	note 2	200	–	500	MHz
f _{shift}	frequency shift	ΔV _P = 10%; note 3	–	–	400	kHz
f _{drift}	frequency drift	ΔT = 25 °C with no compensation; NP0 capacitors; note 4	–	–	2	MHz
		5 s to 15 min after switching on; note 5	–	–	300	kHz
Band C mixer (including IF amplifier)						
f _{RF}	frequency range	note 1	430	–	860	MHz
G _V	voltage gain	R _L = 75 Ω; f _{RF} = 430 MHz; see Fig.8	28	31	34	dB
		R _L = 75 Ω; f _{RF} = 860 MHz; see Fig.8	28	31	34	dB
NF	noise figure (not corrected for image)	R _L = 75 Ω; f _{RF} = 430 MHz; see Fig.9	–	7.0	9.0	dB
		R _L = 75 Ω; f _{RF} = 860 MHz; see Fig.9	–	8.0	10.0	dB
V _O	output voltage	1% cross-modulation in channel; R _L = 75 Ω; f _{RF} = 430 MHz; see Fig.10	107	110	–	dBμV
		1% cross-modulation in channel; R _L = 75 Ω; f _{RF} = 860 MHz; see Fig.10	107	110	–	dBμV
V _i	input voltage	10 kHz pulling in channel; f _{RF} = 860 MHz; note 6	–	93	–	dBμV
	input voltage	N + 5 – 1 MHz pulling; f _{RF} = 820 MHz; see Fig.11	–	79	–	dBμV
Z _i	input impedance (R _s + jL _s ω)	f _{RF} = 430 to 860 MHz; see Fig.14	–	40	–	Ω
		f _{RF} = 430 to 860 MHz; see Fig.14	–	10	–	nH

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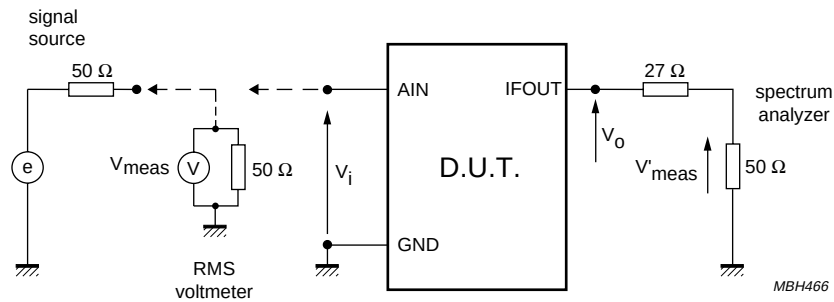
SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Band C oscillator						
f _{osc}	frequency range	note 2	470	–	900	MHz
f _{shift}	frequency shift	ΔV _p = 10%; note 3	–	–	500	kHz
f _{drift}	frequency drift	ΔT = 25 °C with compensation; note 4	–	–	1400	kHz
		5 s to 15 min after switching on; note 5	–	–	400	kHz
LO output						
S ₂₂	output reflection coefficient	see Fig.16	–	–	–	
Y _O	output admittance (Y _p + jωC _p)	see Fig.16	–	400	–	Ω
		see Fig.16	–	0.7	–	pF
V _o	output voltage	R _L = 50 Ω	83	91	100	dBμV
SRF	spurious signal on LO output with respect to LO output signal	R _L = 50 Ω; note 7	–	–	–10	dBc
HLO	LO signal harmonics with respect to LO signal	R _L = 50 Ω	–	–	–10	dBc
IF amplifier						
S ₂₂	output reflection coefficient	magnitude; through 1 nF; see Fig.15	–	–14.5	–	dB
		phase; through 1 nF; see Fig.15	–	5.5	–	deg.
Z _o	output impedance (R _s + jωL _s)	R _s ; through 1 nF	–	73	–	Ω
		L _s ; through 1 nF	–	12	–	nH

Notes

- The RF frequency range is defined by the oscillator frequency range and the intermediate frequency.
- Limits are related to the tank circuits used in Fig.17; frequency bands may be adjusted by the choice of external components.
- The frequency shift is defined as the change in oscillator frequency when the supply voltage varies from $V_P = 9$ to 8.1 V or from $V_P = 9$ to 9.9 V.
- The frequency drift is defined as the change in oscillator frequency when the ambient temperature varies from $T_{\text{amb}} = 25$ to 0 $^\circ\text{C}$ or from $T_{\text{amb}} = 25$ to 50 $^\circ\text{C}$.
- Switching on drift is defined as the change in oscillator frequency between 5 s and 15 min after switching on.
- The input level causing 10 kHz frequency detuning at the LO output; $f_{\text{OSC}} = f_{\text{RF}} + 38.9\text{ MHz}$.
- SRF: spurious signal on LO with respect to LO output signal:
 RF level = 120 $\text{dB}\mu\text{V}$ at $f_{\text{RF}} < 180\text{ MHz}$
 RF level = 107.5 $\text{dB}\mu\text{V}$ at $f_{\text{RF}} = 180$ to 225 MHz
 RF level = 97 $\text{dB}\mu\text{V}$ at $f_{\text{RF}} = 225$ to 860 MHz.

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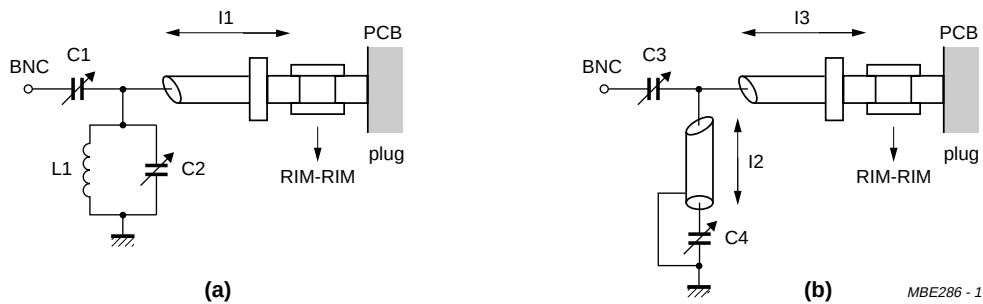


$$Z_i(\text{AIN}) \gg 50 \, \Omega \Rightarrow V_i = 2 \times V_{\text{meas}}.$$

$$V_o = V'_{\text{meas}} \times \left(\frac{50 + 27}{50} \right).$$

$$G_V = 20 \log \frac{V_o}{V_i}.$$

Fig.4 Band A gain measurement.

**(a) For $f_{RF} = 50 \text{ MHz}$:**

mixer A frequency response measured = 57 MHz, loss = 0 dB.
image suppression = 16 dB.

C1 = 9 pF.

C2 = 15 pF.

L1 = 7 turns (diameter. = 5.5 mm, wire diameter. = 0.5 mm).

l1 = semi rigid cable (RIM): 5 cm long

(semi rigid cable (RIM); 33 dB/100 m; 50 Ω; 96 pF/m).

(b) For $f_{RF} = 150 \text{ MHz}$:

mixer A frequency response measured = 150.3 MHz, loss = 1.3 dB.
image suppression = 13 dB.

C3 = 5 pF.

C4 = 25 pF.

l2 = semi rigid cable (RIM): 30 cm long.

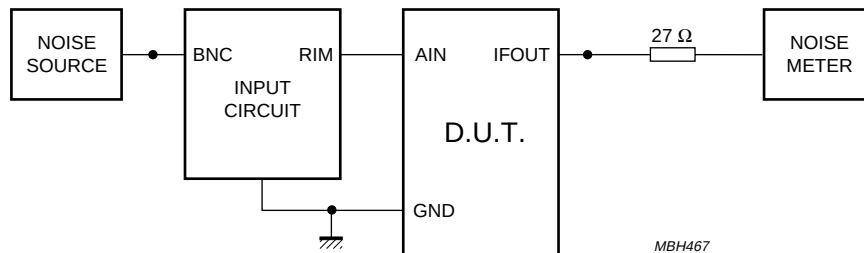
l3 = semi rigid cable (RIM): 5 cm long

(semi rigid cable (RIM); 33 dB/100 m; 50 Ω; 96 pF/m).

Fig.5 Input circuit for optimum noise figure in band A.

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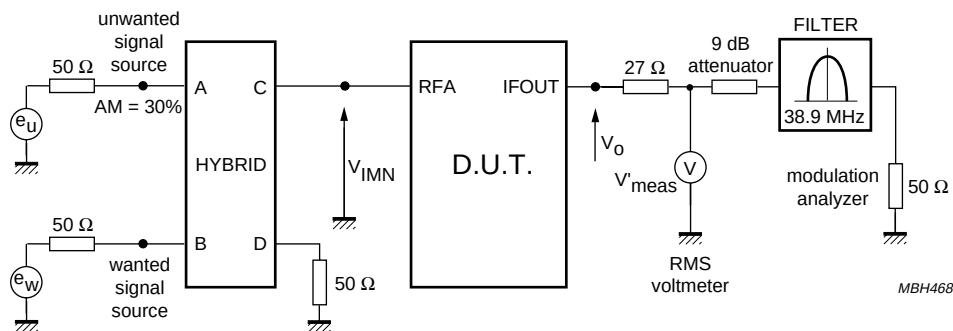
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Input circuit as shown in Fig.5.

$$NF = NF_{\text{meas}} - \text{loss}_{\text{(input circuit)}} \text{ dB.}$$

Fig.6 Noise figure measurement in band A.



Wanted output signal at f_{RFW} ; $V_{\text{ow}} = 100 \text{ dB}\mu\text{V}$.

We measure the level of the unwanted signal V_{ou} causing 0.3 % AM modulation in the wanted output signal; $V_{\text{ou}} = V'_{\text{meas}} \times \left(\frac{50 + 27}{50} \right)$.

$f_{\text{RFW}} = 50 \text{ MHz}$, $f_{\text{RFU}} = 55.5 \text{ MHz}$, $f_{\text{OSC}} = 88.9 \text{ MHz}$.

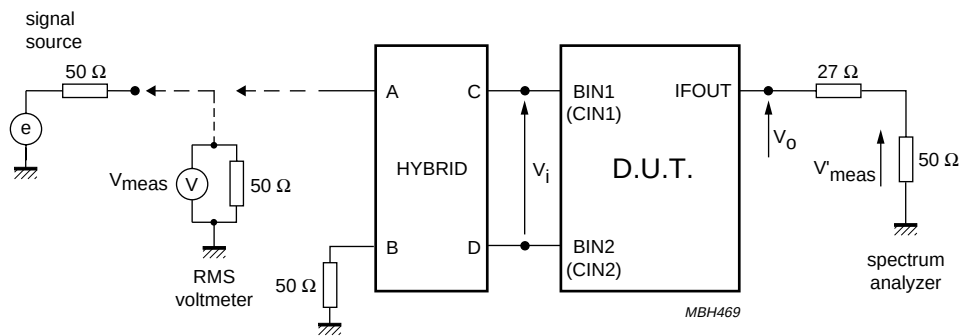
$f_{\text{RFW}} = 180 \text{ MHz}$, $f_{\text{RFU}} = 185.5 \text{ MHz}$, $f_{\text{OSC}} = 218.9 \text{ MHz}$.

Filter characteristics: $f_c = 38.9 \text{ MHz}$, $f_{-3\text{dB BW}} = 1.2 \text{ MHz}$, $f_{-30\text{dB BW}} = 2.64 \text{ MHz}$.

Fig.7 Cross modulation measurement in band A.

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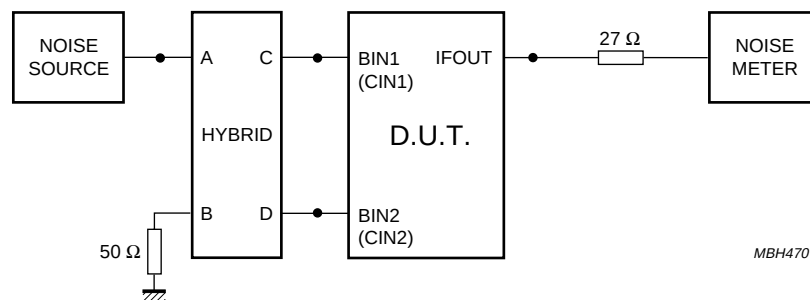
$$\text{loss}_{(\text{hybrid})} = 1 \text{ dB.}$$

$$V_i = V_{\text{meas}} - \text{loss}_{(\text{hybrid})}.$$

$$V_o = V'_{\text{meas}} \times \left(\frac{50 + 27}{50} \right).$$

$$\text{Voltage gain for band B and C} = 20 \log \frac{V_o}{V_i}.$$

Fig.8 Gain measurement in bands B and C.



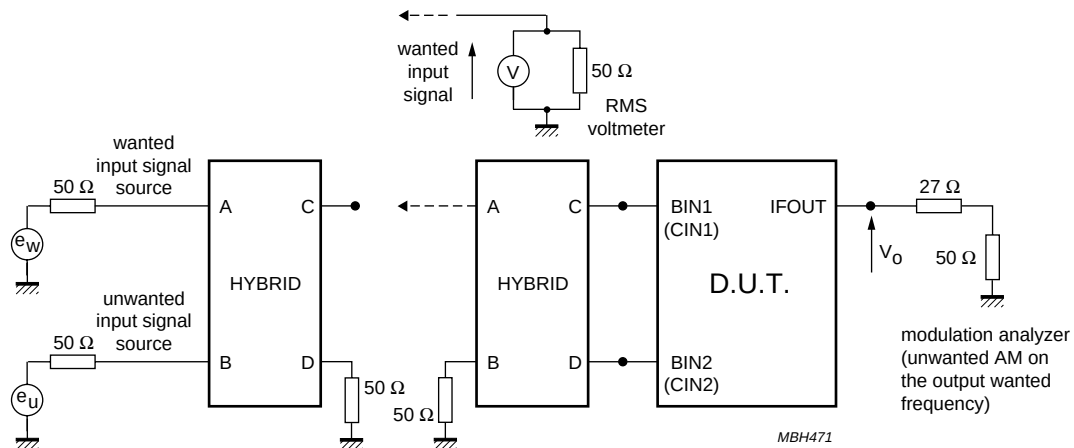
$$\text{loss}_{(\text{hybrid})} = 1 \text{ dB.}$$

$$\text{NF} = \text{NF}_{\text{meas}} - \text{loss}_{(\text{hybrid})} \text{ dB.}$$

Fig.9 Noise figure measurement in bands B and C.

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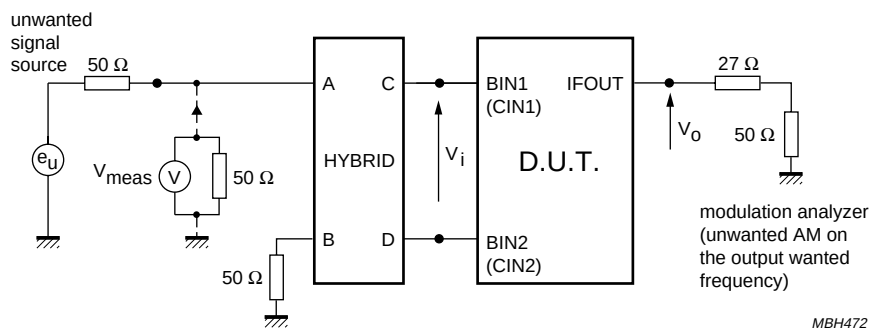
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Wanted input signal = 70 dBμV; unwanted input signal modulated with 30% AM at 100 kHz; V_{ou} = unwanted output signal V_{out} when the wanted output signal is modulated with 0.3% AM.

$f_{UNWANTED} = f_{WANTED} + 5.5 \text{ MHz}$; $f_{OSC} = f_{WANTED} + 38.9 \text{ MHz}$.

Fig.10 Cross modulation measurement in bands B and C.



$\text{loss}_{(\text{hybrid})} = 1 \text{ dB}$.

In band B: $f_{RFW} = 391 \text{ MHz}$ (in band C: $f_{RFW} = 781 \text{ MHz}$). These wanted signals are not used during the measurement.

In band B: $f_{OSC} = 429.9 \text{ MHz}$ (in band C: $f_{OSC} = 819.9 \text{ MHz}$).

In band B: $f_{RFU} = 430 \text{ MHz} = f_{RFW} + 5 \times 8 \text{ MHz} - 1 \text{ MHz}$. (in band C: $f_{RFU} = 820 \text{ MHz} = f_{RFW} + 5 \times 8 \text{ MHz} - 1 \text{ MHz}$).

We measure the level of the unwanted signal V_{iu} causing fm sidebands 30 dB below the oscillator carrier at the LO output.

$V_{iu} = V_{meas} - \text{loss}_{(\text{hybrid})}$.

Fig.11 N + 5 -1 MHz pulling measurement in bands B and C.

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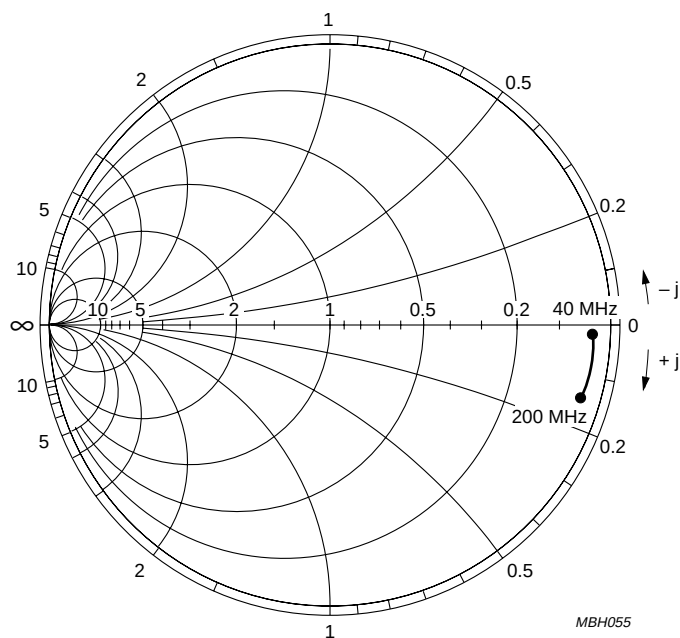


Fig.12 Input admittance (S_{11}) of the band A mixer input (40 to 200 MHz); $Y_0 = 20 \text{ mS}$.

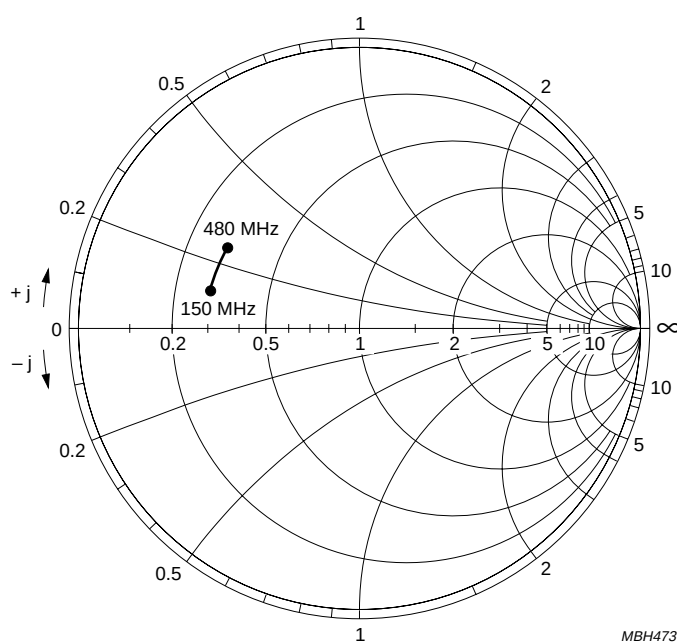


Fig.13 Input impedance (S_{11}) of the band B mixer input (150 to 480 MHz); $Z_0 = 100 \Omega$.

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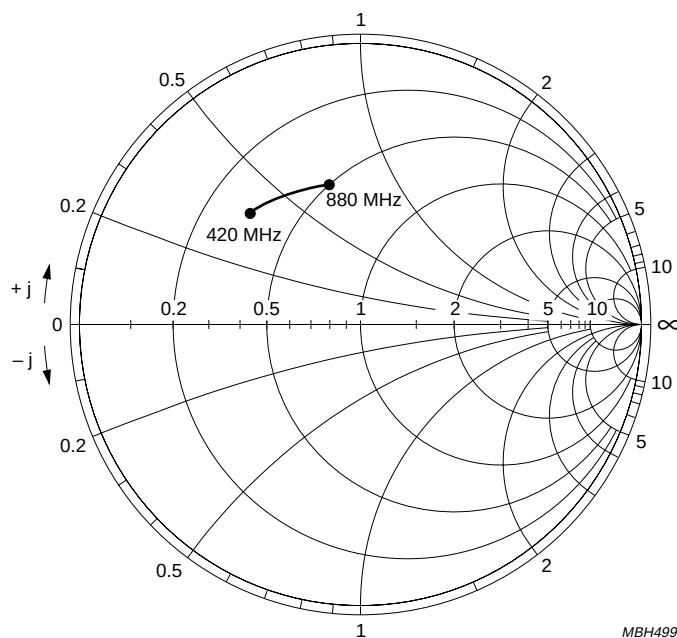


Fig.14 Input impedance (S_{11}) of the band C mixer input (420 to 880 MHz); $Z_0 = 100 \Omega$.

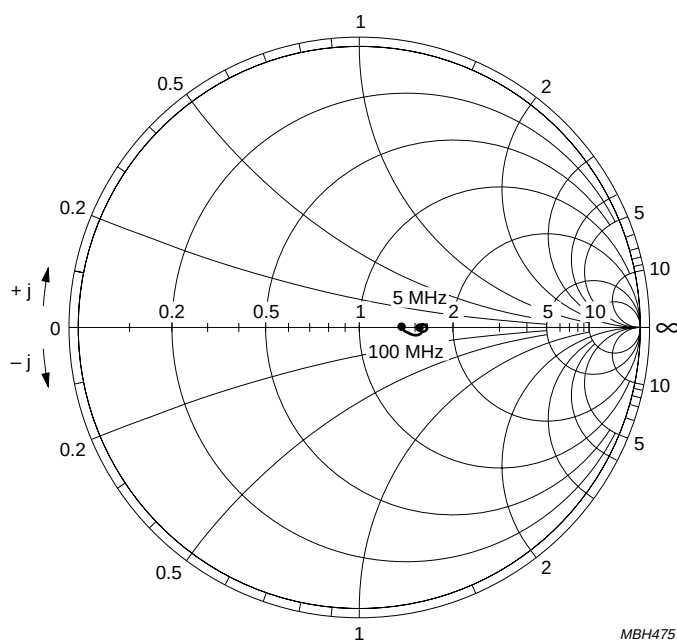


Fig.15 Output impedance (S_{22}) of the IF amplifier (20 to 60 MHz); $Z_0 = 50 \Omega$.

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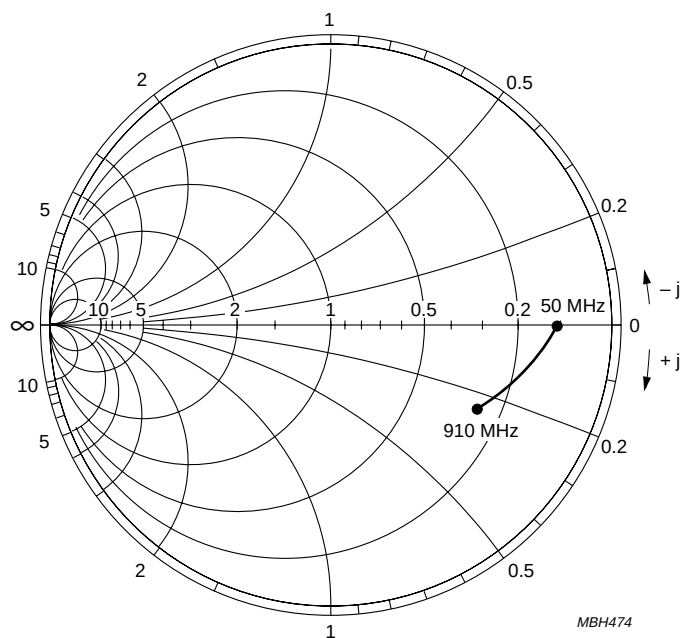
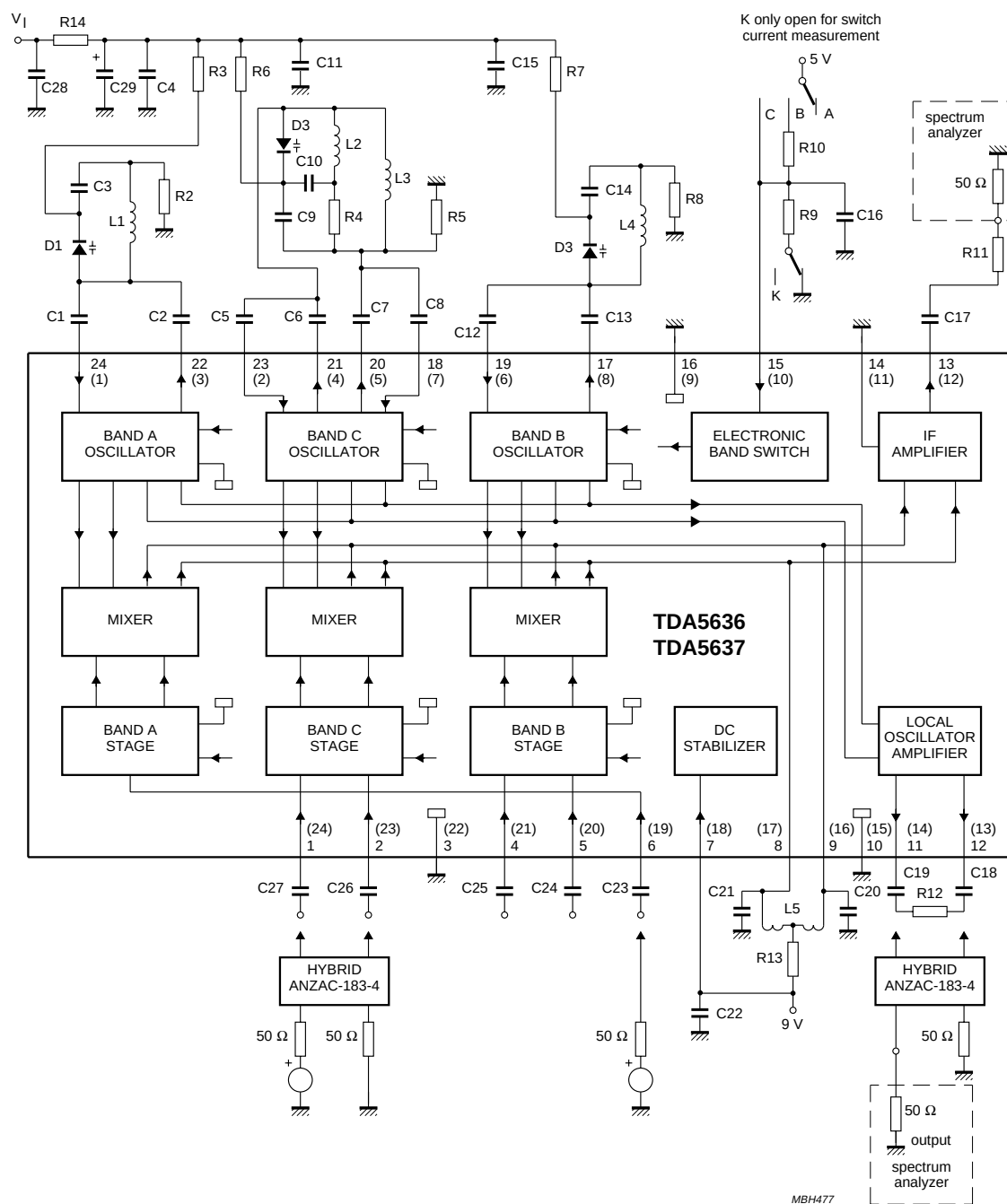


Fig.16 Output admittance (S_{22}) of the LO amplifier (50 to 910 MHz); $Y_o = 20 \text{ mS}$.

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APPLICATION INFORMATION



The numbers in parenthesis represent the TDA5637.

Fig.17 Measurement circuit.

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Application diagram component values

Table 1 Capacitors
(all SMD and NP0 except C5 to C9 and C29)

NUMBER	VALUE
C1	2 pF
C2	2 pF
C3	82 pF
C4	2.2 nF
C5	1 pF (N750)
C6	1 pF (N750)
C7	1 pF (N750)
C8	1 pF (N750)
C9	6 pF (N470)
C10	100 pF
C11	2.2 nF
C12	2 pF
C13	4 pF
C14	150 pF
C15	2.2 nF
C16	1.2 nF
C17	1 nF
C18	1 nF
C19	1 nF
C20	12 pF
C21	12 pF
C22	22 nF
C23	1 nF
C24	1 nF
C25	1 nF
C26	1 nF
C27	1 nF
C28	2.2 nF
C29	1 μ F (40 V electrolytic capacitor)

Table 2 Resistors (all SMD)

NUMBER	VALUE
R2	22 Ω
R3	47 k Ω
R4	2.2 k Ω
R5	22 k Ω
R6	47 k Ω
R7	47 k Ω
R8	12 Ω
R9	15 k Ω
R10	33 k Ω
R11	27 Ω
R12	100 Ω
R13	150 Ω
R14	47 k Ω

Table 3 Diodes, coils and transformers

NUMBER	VALUE
Diodes	
D1	BB132
D2	BB134
D3	BB146
Coils⁽¹⁾	
L1	8 turns ($\varnothing$ 3 mm)
L2	2 turns ($\varnothing$ 2.5 mm)
L3	3 turns ($\varnothing$ 2.5 mm)
L4	2 turns ($\varnothing$ 4 mm)
Transformer⁽²⁾	
L5	2 $\times$ 6 turns

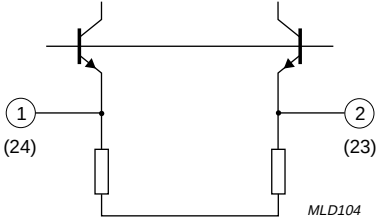
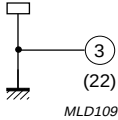
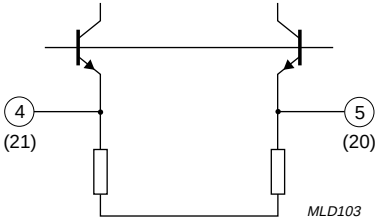
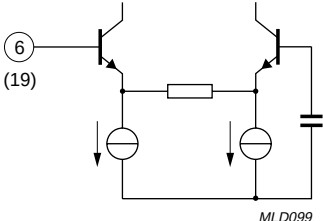
Notes

- Wire size for L1 to L4 is 0.4 mm.
- Coil type: TOKO 7KN; material: 113 KN, screw core: 03-0093, pot core: 04-0026.

9 V VHF, hyperband and UHF mixers-oscillators for TV and VCR 3-band tuners

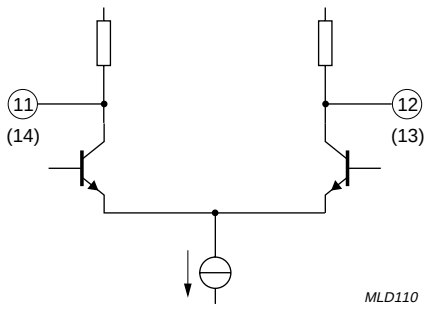
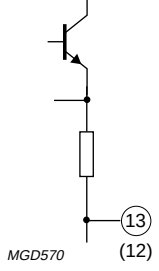
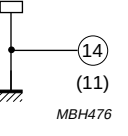
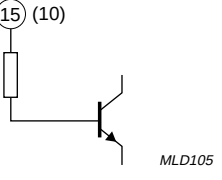
TDA5636; TDA5637

INTERNAL PIN CONFIGURATION

SYMBOL	PIN		DESCRIPTION	AVERAGE DC VOLTAGE ⁽¹⁾ IN (V)		
	TDA5636B	TDA5637B		BAND A	BAND B	BAND C
CIN1	1	24		0	0	2.2
CIN2	2	23		0	0	2.2
RFGND	3	22		0	0	0
BIN1	4	21		0	2.2	0
BIN2	5	20		0	2.2	0
AIN	6	19		2.2	1.2	1.2
V _P	7	18	supply voltage	9.0	9.0	9.0
MIXOUT1	8	17		8.4	8.4	8.4
MIXOUT2	9	16		8.4	8.4	8.4

9 V VHF, hyperband and UHF mixers-oscillators for TV and VCR 3-band tuners

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SYMBOL	PIN		DESCRIPTION	AVERAGE DC VOLTAGE ⁽¹⁾ IN (V)		
	TDA5636B	TDA5637B		BAND A	BAND B	BAND C
GND1	10	15	 MLD106	0	0	0
GND2	16	9		0	0	0
LOOUT1	11	14	 MLD110	5.6	5.6	5.6
LOOUT2	12	13		5.6	5.6	5.6
IFOUT	13	12	 MGD570	4.4	4.4	4.4
IFGND	14	11	 MBH476	0.0	0.0	0.0
BS	15	10	 MLD105	V _{SW(A)}	V _{SW(B)}	V _{SW(C)}

9 V VHF, hyperband and UHF mixers-oscillators for TV and VCR 3-band tuners

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SYMBOL	PIN		DESCRIPTION	AVERAGE DC VOLTAGE ⁽¹⁾ IN (V)		
	TDA5636B	TDA5637B		BAND A	BAND B	BAND C
BOSCOC	17	8	 MLD113	5.8	3.4	5.8
BOSCIB	19	6		1.2	2.3	1.2
COSCIB1	18	7	 MLD112	1.4	1.4	2.3
COSCOC1	20	5		5.8	5.8	4.2
COSCOC2	21	4		5.8	5.8	4.2
COSCIB2	23	2		1.4	1.4	2.3
AOSCOC	22	3	 MLD111	3.8	5.8	5.8
AOSCIB	24	1		2.1	1.0	1.0

Note

1. Average DC voltage measured in circuit of Fig.17.

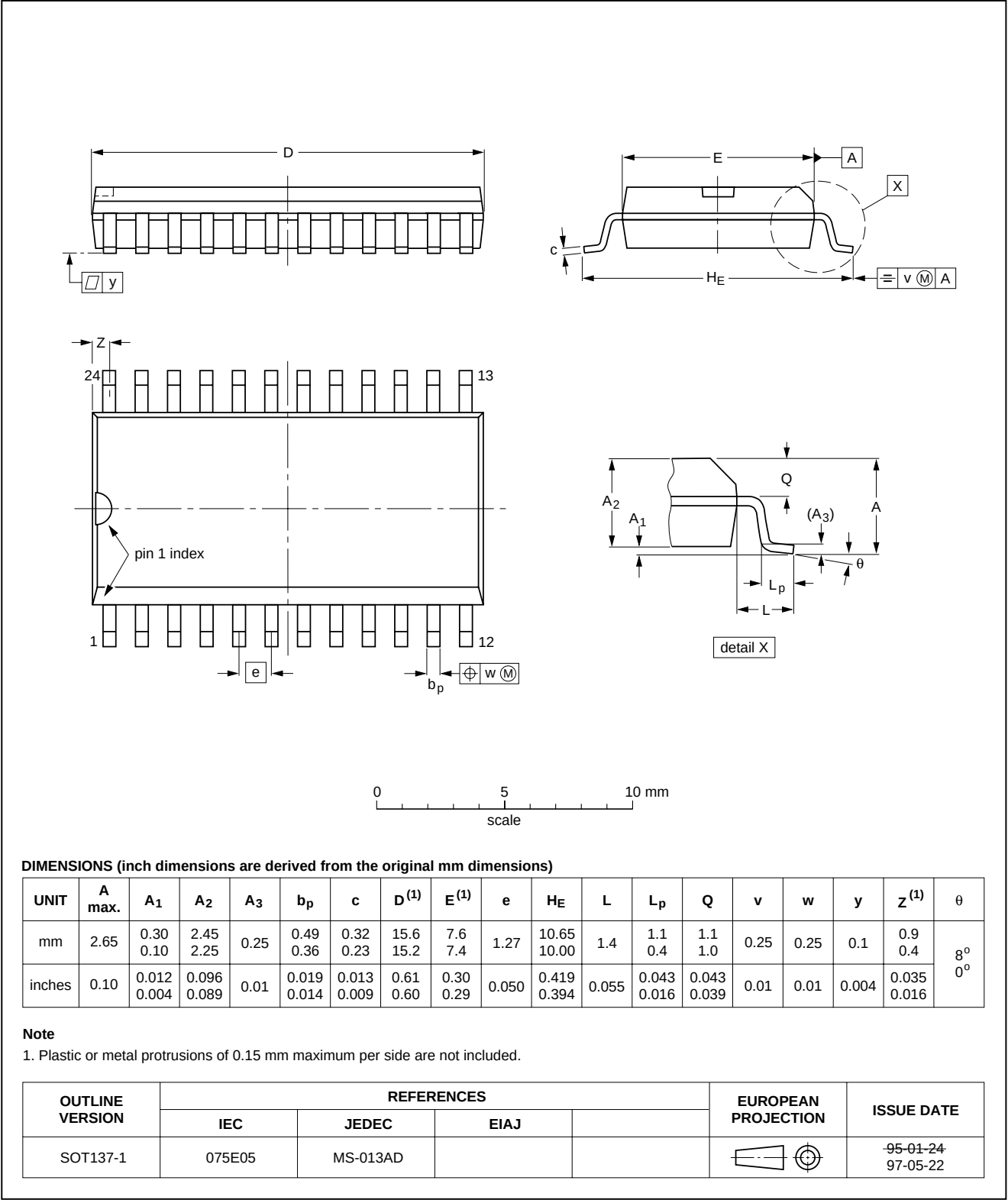
9 V VHF, hyperband and UHF mixers-oscillators for TV and VCR 3-band tuners

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PACKAGE OUTLINES

SO24: plastic small outline package; 24 leads; body width 7.5 mm

SOT137-1

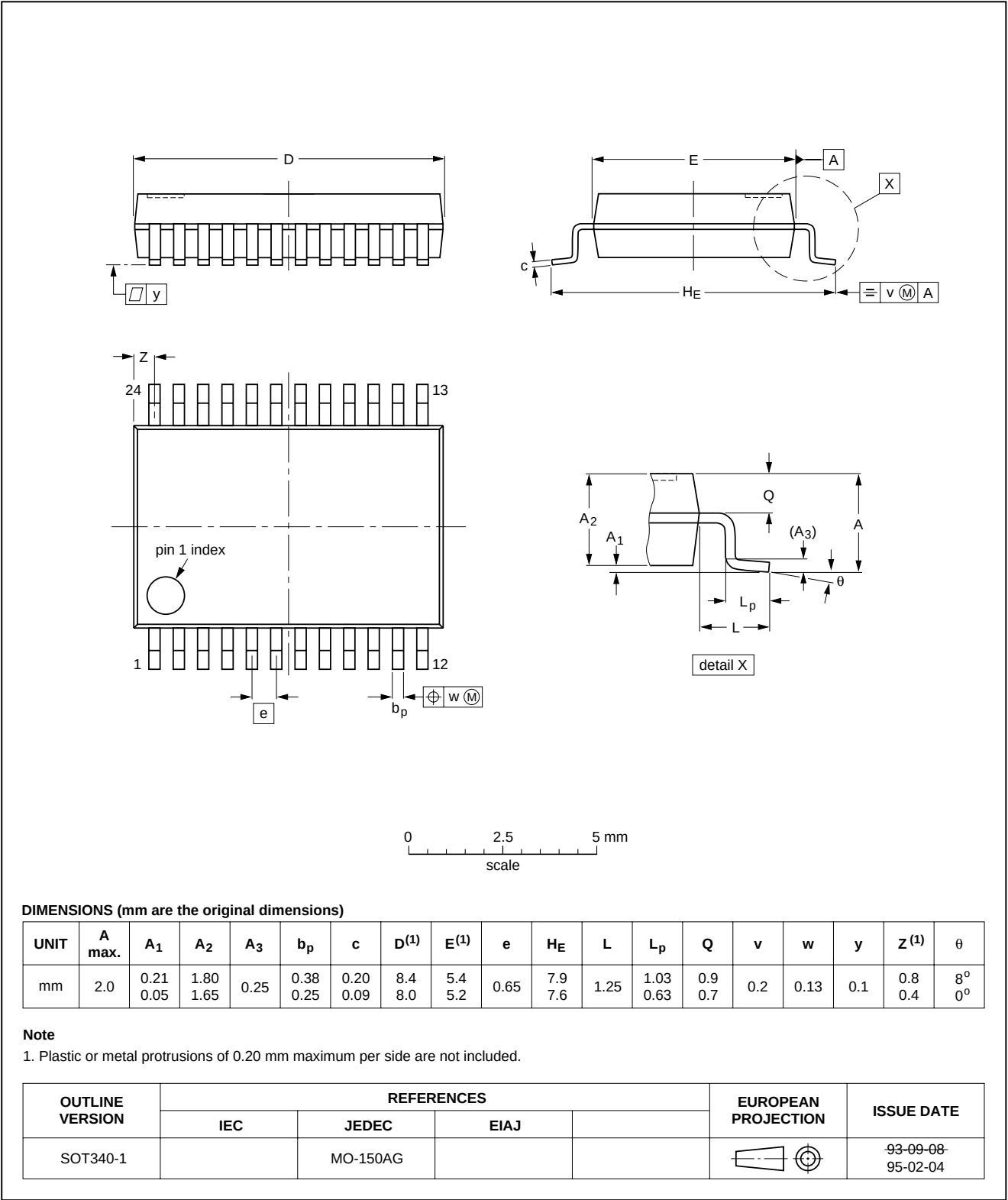


9 V VHF, hyperband and UHF mixers-oscillators for TV and VCR 3-band tuners

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SSOP24: plastic shrink small outline package; 24 leads; body width 5.3 mm

SOT340-1



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SOLDERING

Introduction

There is no soldering method that is ideal for all IC packages. Wave soldering is often preferred when through-hole and surface mounted components are mixed on one printed-circuit board. However, wave soldering is not always suitable for surface mounted ICs, or for printed-circuits with high population densities. In these situations reflow soldering is often used.

This text gives a very brief insight to a complex technology. A more in-depth account of soldering ICs can be found in our *"IC Package Databook"* (order code 9398 652 90011).

Reflow soldering

Reflow soldering techniques are suitable for all SO and SSOP packages.

Reflow soldering requires solder paste (a suspension of fine solder particles, flux and binding agent) to be applied to the printed-circuit board by screen printing, stencilling or pressure-syringe dispensing before package placement.

Several techniques exist for reflowing; for example, thermal conduction by heated belt. Dwell times vary between 50 and 300 seconds depending on heating method. Typical reflow temperatures range from 215 to 250 °C.

Preheating is necessary to dry the paste and evaporate the binding agent. Preheating duration: 45 minutes at 45 °C.

Wave soldering

SO

Wave soldering techniques can be used for all SO packages if the following conditions are observed:

- A double-wave (a turbulent wave with high upward pressure followed by a smooth laminar wave) soldering technique should be used.
- The longitudinal axis of the package footprint must be parallel to the solder flow.
- The package footprint must incorporate solder thieves at the downstream end.

SSOP

Wave soldering is **not** recommended for SSOP packages. This is because of the likelihood of solder bridging due to closely-spaced leads and the possibility of incomplete solder penetration in multi-lead devices.

If wave soldering cannot be avoided, the following conditions must be observed:

- A double-wave (a turbulent wave with high upward pressure followed by a smooth laminar wave) soldering technique should be used.
- The longitudinal axis of the package footprint must be parallel to the solder flow and must incorporate solder thieves at the downstream end.

Even with these conditions, only consider wave soldering SSOP packages that have a body width of 4.4 mm, that is SSOP16 (SOT369-1) or SSOP20 (SOT266-1).

METHOD (SO AND SSOP)

During placement and before soldering, the package must be fixed with a droplet of adhesive. The adhesive can be applied by screen printing, pin transfer or syringe dispensing. The package can be soldered after the adhesive is cured.

Maximum permissible solder temperature is 260 °C, and maximum duration of package immersion in solder is 10 seconds, if cooled to less than 150 °C within 6 seconds. Typical dwell time is 4 seconds at 250 °C.

A mildly-activated flux will eliminate the need for removal of corrosive residues in most applications.

Repairing soldered joints

Fix the component by first soldering two diagonally-opposite end leads. Use only a low voltage soldering iron (less than 24 V) applied to the flat part of the lead. Contact time must be limited to 10 seconds at up to 300 °C. When using a dedicated tool, all other leads can be soldered in one operation within 2 to 5 seconds between 270 and 320 °C.

**9 V VHF, hyperband and UHF mixers-
oscillators for TV and VCR 3-band tuners**

TDA5636; TDA5637**DEFINITIONS**

Data sheet status	
Objective specification	This data sheet contains target or goal specifications for product development.
Preliminary specification	This data sheet contains preliminary data; supplementary data may be published later.
Product specification	This data sheet contains final product specifications.
Limiting values	
Limiting values given are in accordance with the Absolute Maximum Rating System (IEC 134). Stress above one or more of the limiting values may cause permanent damage to the device. These are stress ratings only and operation of the device at these or at any other conditions above those given in the Characteristics sections of the specification is not implied. Exposure to limiting values for extended periods may affect device reliability.	
Application information	
Where application information is given, it is advisory and does not form part of the specification.	

LIFE SUPPORT APPLICATIONS

These products are not designed for use in life support appliances, devices, or systems where malfunction of these products can reasonably be expected to result in personal injury. Philips customers using or selling these products for use in such applications do so at their own risk and agree to fully indemnify Philips for any damages resulting from such improper use or sale.

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oscillators for TV and VCR 3-band tuners

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