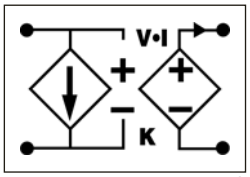
B048F096T24
B048F096M24

BCM™
Bus Converter

- 48 V to 9.6 V V•I Chip™ Bus Converter
- 240 Watt (360 Watt for 1 ms)
- High density – 813 W/in³
- Small footprint – 210 W/in²
- Low weight – 0.5 oz (15 g)
- ZVS / ZCS isolated Sine Amplitude Converter™
- Typical efficiency 96%
- 125°C operation (T_J)
- <1 µs transient response
- 3.5 million hours MTBF
- No output filtering required





V_{IN} = 38 - 55 V
V_{OUT} = 7.60 - 11.0 V
I_{OUT} = 25 A
K = 1/5
R_{OUT} = 12.0 mΩ max

Product Description

The V•I Chip™ bus converter is a high efficiency (>96%), narrow input range Sine Amplitude Converter™ (SAC™) operating from a 38 to 55 Vdc primary bus to deliver an isolated 7.60 V to 11.0 V secondary. The bus converter may be used to power non-isolated POL converters or as an independent 7.60 – 11.0 V source. Due to the fast response time and low noise of the bus converter, the need for limited life aluminum electrolytic or tantalum capacitors at the load is reduced—or eliminated—resulting in savings of board area, materials and total system cost.

The bus converter achieves a power density of 813 W/in³ in a V•I Chip package compatible with standard pick-and-place and surface mount assembly process. The V•I Chip package provides flexible thermal management through its low junction-to-board and junction-to-case thermal resistance. Owing to its high conversion efficiency and safe operating temperature range, the bus converter does not require a discrete heat sink in typical applications. Low junction-to-case and junction-to-lead thermal impedances assure low junction temperatures and long life in the harshest environments.

Absolute Maximum Ratings

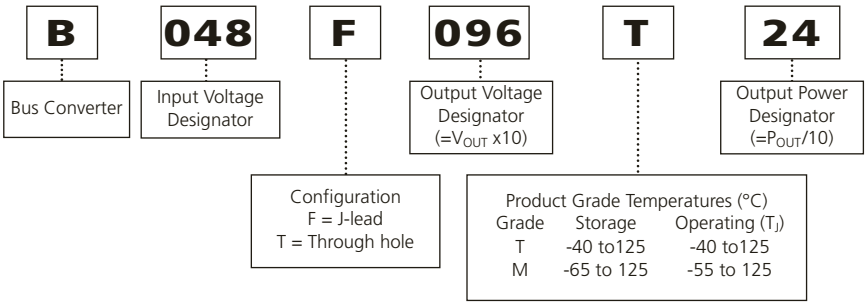
Parameter	Values	Unit	Notes
+In to -In	-1.0 to 60	Vdc	
	100	Vdc	For 100 ms
PC to -In	-0.3 to 7.0	Vdc	
+Out to -Out	-0.5 to 16	Vdc	
Isolation voltage	2,250	Vdc	Input to output
Output current	31.5	A	Continuous
Peak output current	37.5	A	For 1 ms
Output power	240	W	Continuous
Peak output power	360	W	For 1 ms
Case temperature during reflow [a]	225	°C	MSL 5
	245	°C	MSL 6, TOB = 4 hrs
Operating junction temperature [b]	-40 to 125	°C	T-Grade
	-55 to 125	°C	M-Grade
Storage temperature	-40 to 125	°C	T-Grade
	-65 to 125	°C	M-Grade

Notes:

[a] 245°C reflow capability applies to product with manufacturing date code 1001 and greater.

[b] The referenced junction is defined as the semiconductor having the highest temperature. This temperature is monitored by a shutdown comparator.

Part Numbering



Specifications

Input (Conditions are at 48 V_{IN}, full load, and 25°C ambient unless otherwise specified)

Parameter	Min	Typ	Max	Unit	Note
Input voltage range	38	48	55	Vdc	
Input dV/dt			1	V/µs	
Input undervoltage turn on			38.0	Vdc	
Input undervoltage turn off	32.0			Vdc	
Input overvoltage turn on	55.0			Vdc	
Input overvoltage turn off			60.0	Vdc	
Input quiescent current		2.5		mA	PC low
Inrush current overshoot		5.3		A	Using test circuit in Figure 20; See Figure 1
Input current			5.4	Adc	
Input reflected ripple current		140		mA p-p	Using test circuit in Figure 20; See Figure 4
No load power dissipation		3.1	4.1	W	
Internal input capacitance		4.0		µF	
Internal input inductance		5		nH	
Recommended external input capacitance		47		µF	200 nH maximum source inductance; See Figure 20

Input Waveforms

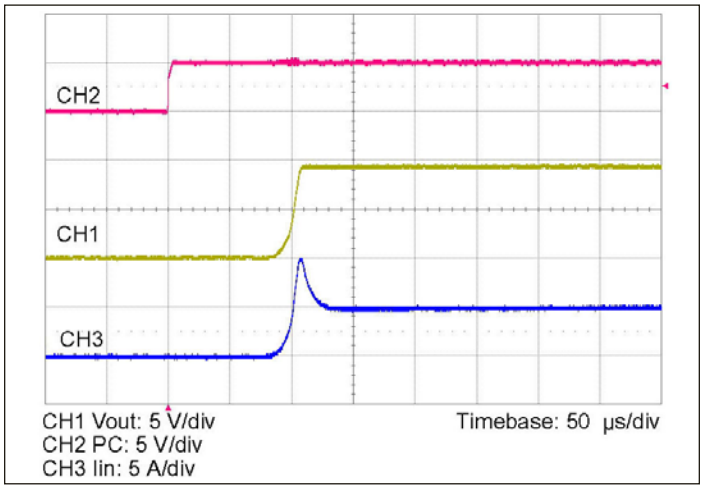


Figure 1 — *Inrush transient current at full load and 48 V_{IN} with PC enabled*

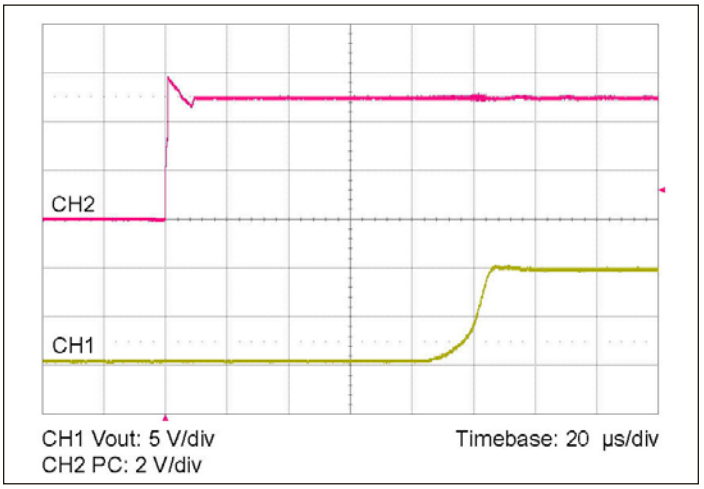


Figure 2 — *Output voltage turn on waveform with PC enabled at full load and 48 V_{IN}*

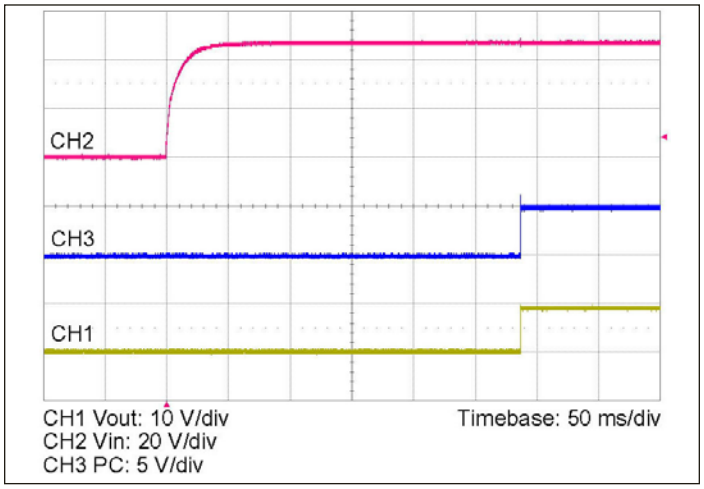


Figure 3 — *Output voltage turn on waveform with input turn on at full load and 48 V_{IN}*

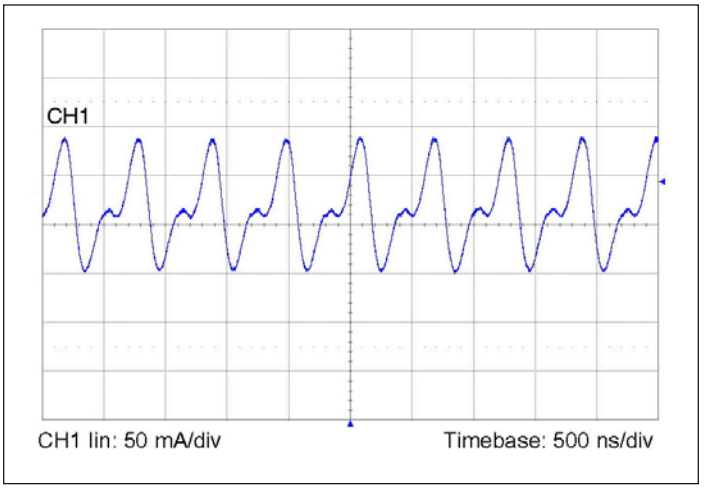


Figure 4 — *Input reflected ripple current at full load and 48 V_{IN}*

Output (Conditions are at 48 Vin, full load, and 25°C ambient unless otherwise specified)

Parameter	Min	Typ	Max	Unit	Note
Output voltage	7.60		11.0	Vdc	No load
	7.30		10.7	Vdc	Full load
Output power	0		240	W	40 - 55 V_{IN}
	0		227	W	38 - 55 V_{IN}
Rated DC current	0		31.5	Adc	$P_{OUT} \leq 240$ W
Peak repetitive power			360	W	Max pulse width 1ms, max duty cycle 10%, baseline power 50%
Current share accuracy		5	10	%	See Parallel Operation on Page 10
Efficiency					
Half load	95.5	96.2		%	See Figure 5
Full load	95.5	96.2		%	See Figure 5
Internal output inductance		1.6		nH	
Internal output capacitance		55		µF	Effective value
Load capacitance			1,600	µF	
Output overvoltage set point	11.0			Vdc	Module will shut down
Output ripple voltage					
No external bypass		176	200	mVp-p	See Figures 7 and 9
10 µF bypass capacitor		17		mVp-p	See Figure 8
Short circuit protection set point	33			Adc	Module will shut down
Average short circuit current		0.43		A	
Effective switching frequency	3.0	3.1	3.4	MHz	Fixed, 1.6 MHz per phase
Line regulation					
K	0.1980	1/5	0.2020		$V_{OUT} = K \bullet V_{IN}$ at no load
Load regulation					
R_{OUT}		8.9	12.0.	mΩ	
Transient response					
Voltage overshoot		92		mV	100% load step; See Figures 10 and 11
Response time		200		ns	See Figures 10 and 11
Recovery time		1		µs	See Figures 10 and 11
Output overshoot					
Input turn on		0		mV	No output filter; See Figure 3
PC enable		0		mV	No output filter; See Figure 2
Output turn on delay					
From application of power		288		ms	No output filter; See Figure 3
From release of PC pin		70		ms	No output filter

Output Waveforms

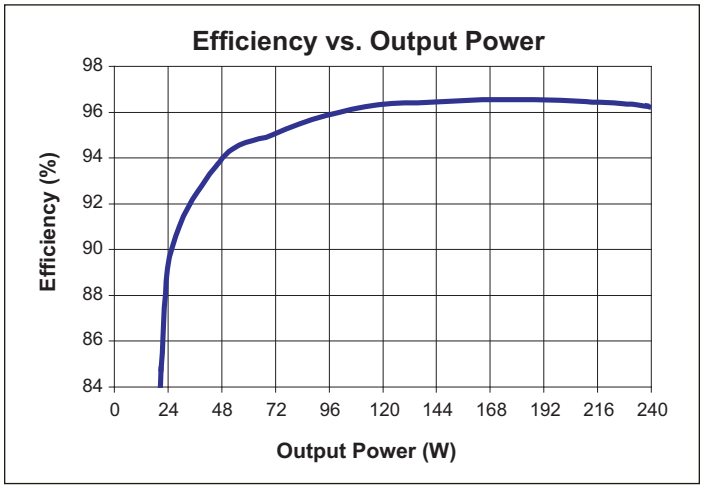


Figure 5 — *Efficiency vs. output power*

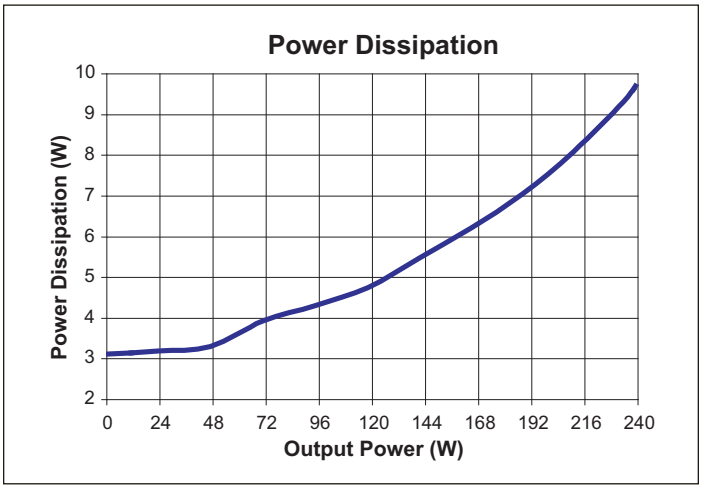


Figure 6 — *Power dissipation as a function of output power*

Output Waveforms

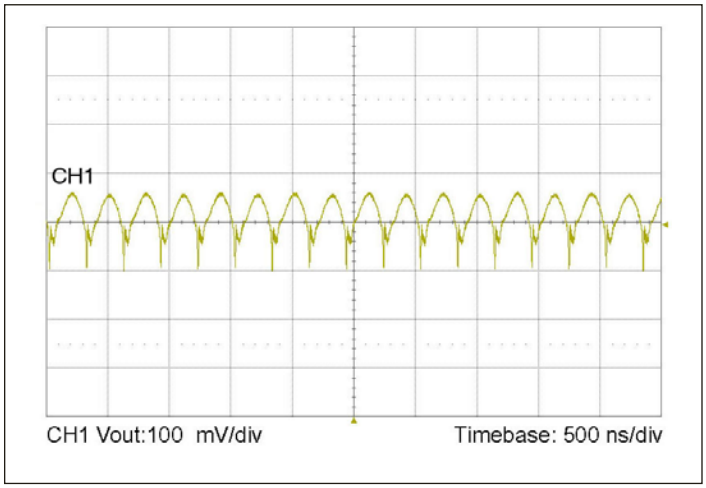


Figure 7 — *Output voltage ripple at full load and 48 V_{IN} without any external bypass capacitor.*

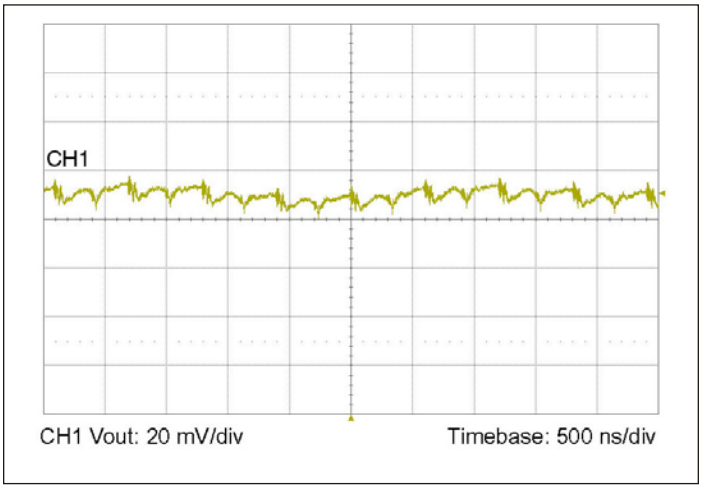


Figure 8 — *Output voltage ripple at full load and 48 V_{IN} with 10 µF ceramic external bypass capacitor and 20 nH of distribution inductance.*

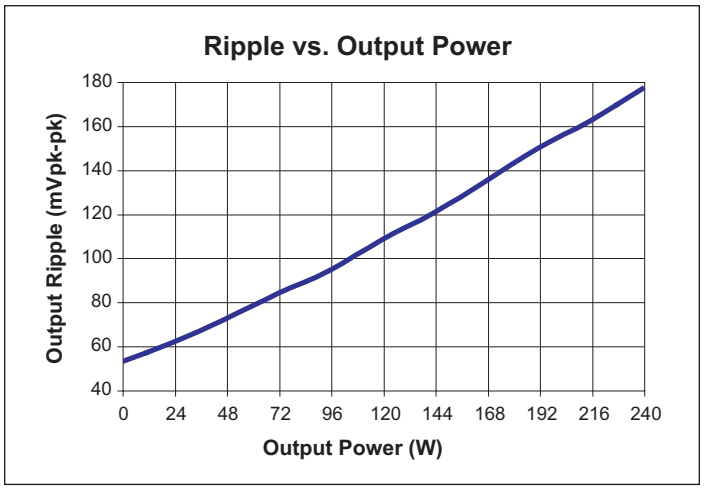


Figure 9 — *Output voltage ripple vs. output power at 48 V_{IN} without any external bypass capacitor.*

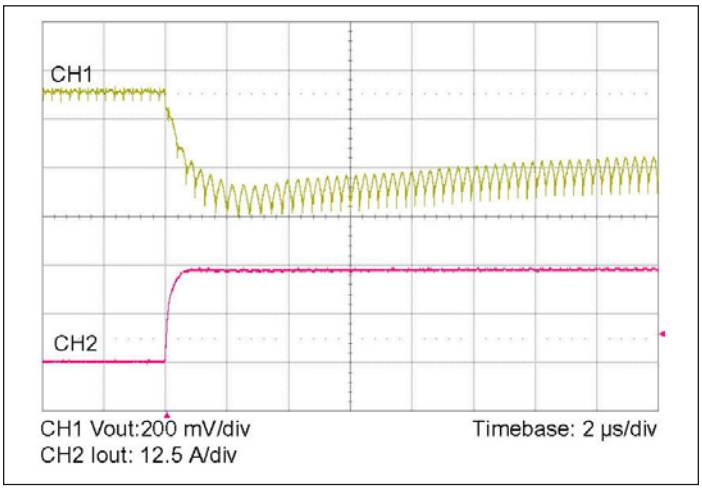


Figure 10 — *0 – 25 A load step with 100 µF input capacitor and no output capacitor.*

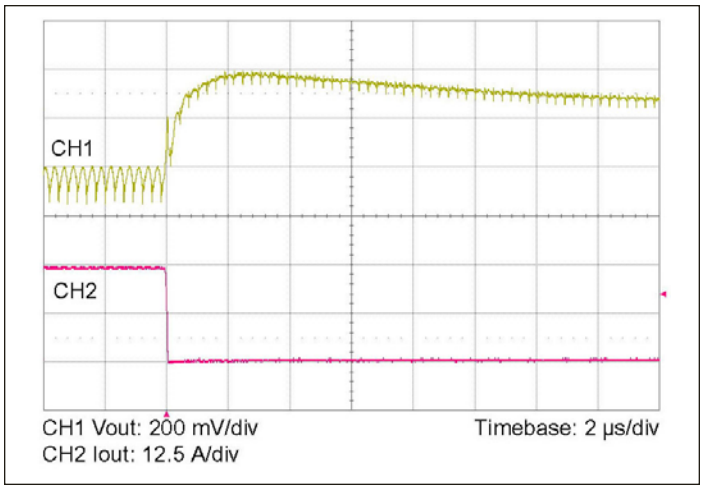


Figure 11 — *25 – 0 A load step with 100 µF input capacitor and no output capacitor.*

General

Parameter	Min	Typ	Max	Unit	Note
MTBF					
MIL-HDBK-217F		3.5		Mhrs	25°C, GB
Isolation specifications					
Voltage	2,250			Vdc	Input to output
Capacitance		3,000		pF	Input to output
Resistance	10			MΩ	Input to output
Agency approvals		cTÜVus			UL/CSA 60950-1, EN 60950-1
		CE Mark			Low voltage directive
		RoHS			
Mechanical					See mechanical drawings, Figures 15 – 18
Weight		0.53/15		oz/g	
Dimensions					
Length		1.28/32,5		in/mm	
Width		0.87/22		in/mm	
Height		0.265/6,73		in/mm	
Thermal					
Overtemperature shutdown	125	130	135	°C	Junction temperature
Thermal capacity		9.3		Ws/°C	
Junction-to-case thermal impedance ($R_{\theta JC}$)		1.1		°C/W	See Thermal Considerations on Page 10
Junction-to-board thermal impedance ($R_{\theta JB}$)		2.1		°C/W	

Auxiliary Pins (Conditions are at 48 Vin, full load, and 25°C ambient unless otherwise specified)

Parameter	Min	Typ	Max	Unit	Note
Primary control (PC)					
DC voltage	4.8	5.0	5.2	Vdc	
Module disable voltage	2.4	2.5		Vdc	
Module enable voltage		2.5	2.6	Vdc	
Current limit	2.4	2.5	2.9	mA	Source only
Enable delay time		70		ms	
Disable delay time		20		µs	See Figure 12, time from PC low to output low

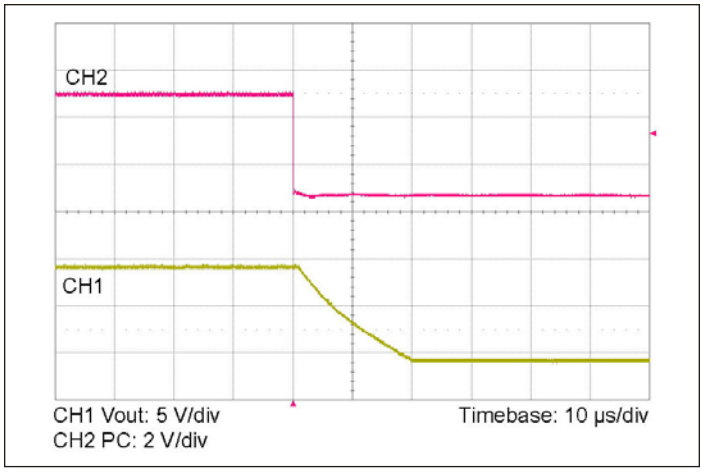


Figure 12 — *V_{OUT} at full load vs. PC disable*

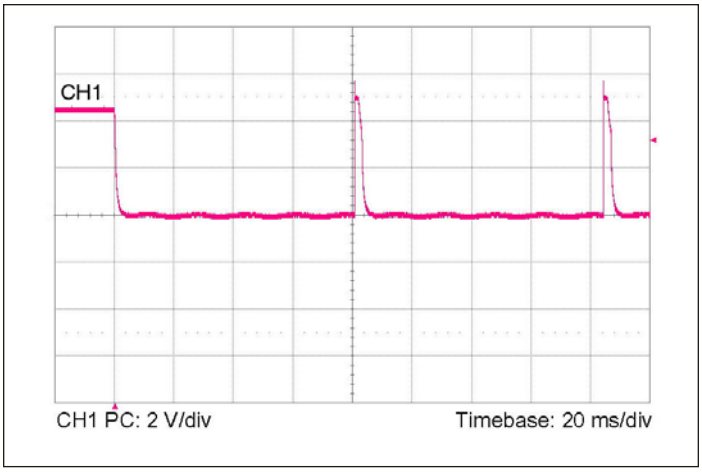


Figure 13 — *PC signal during fault*

Pin / Control Functions

+In / -In – DC Voltage Input Ports

The V•I Chip™ module input voltage range should not be exceeded. An internal undervoltage/overvoltage lockout function prevents operation outside of the normal operating input range. The BCM™ bus converter turns on within an input voltage window bounded by the "Input undervoltage turn on" and "Input overvoltage turn off" levels, as specified. The module may be protected against accidental application of a reverse input voltage by the addition of a rectifier in series with the positive input, or a reverse rectifier in shunt with the positive input located on the load side of the input fuse.

The connection of the module to its power source should be implemented with minimal distribution inductance. If the interconnect inductance exceeds 100 nH, the input should be bypassed with a RC damper to retain low source impedance and stable operation. With an interconnect inductance of 200 nH, the RC damper may be 47 µF in series with 0.3Ω. A single electrolytic or equivalent low-Q capacitor may be used in place of the series RC bypass.

PC – Primary Control

The Primary Control port is a multifunction node that provides the following functions:

Enable / Disable – If the PC port is left floating, the BCM™ module output is enabled. Once this port is pulled lower than 2.4 Vdc with respect to –In, the output is disabled. This action can be realized by employing a relay, opto-coupler, or open collector transistor. Refer to Figures 1-3, 12 and 13 for the typical enable/disable characteristics. This port should not be toggled at a rate higher than 1 Hz. The PC port should also not be driven by or pulled up to an external voltage source.

Primary Auxiliary Supply – The PC port can source up to 2.4 mA at 5.0 Vdc. The PC port should never be used to sink current.

Alarm – The module contains circuitry that monitors output overload, input overvoltage or undervoltage, and internal junction temperatures. In response to an abnormal condition in any of the monitored parameters, the PC port will toggle. Refer to Figure 13 for PC alarm characteristics.

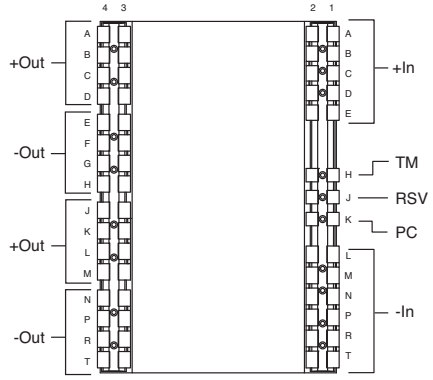
TM and RSV – Reserved for factory use.

+Out / -Out – DC Voltage Output Ports

Two sets of contacts are provided for the +Out port. They must be connected in parallel with low interconnect resistance. Similarly, two sets of contacts are provided for the –Out port. They must be connected in parallel with low interconnect resistance. Within the specified operating range, the average output voltage is defined by the Level 1 DC behavioral model of Figure 21. The current source capability of the module is rated in the specifications section of this document.

The low output impedance of the module reduces or eliminates the need for limited life aluminum electrolytic or tantalum capacitors at the input of POL converters.

Total load capacitance at the output of the modules should not exceed the specified maximum. Owing to the wide bandwidth and low output impedance of the module, low frequency bypass capacitance and significant energy storage may be more densely and efficiently provided by adding capacitance at the input of the BCM module.



Signal Name	Designation
+In	A1-E1, A2-E2
–In	L1-T1, L2-T2
TM	H1, H2
RSV	J1, J2
PC	K1, K2
+Out	A3-D3, A4-D4, J3-M3, J4-M4
–Out	E3-H3, E4-H4, N3-T3, N4-T4

Figure 14 — *BCM™ bus converter pin configuration*

Mechanical Drawings

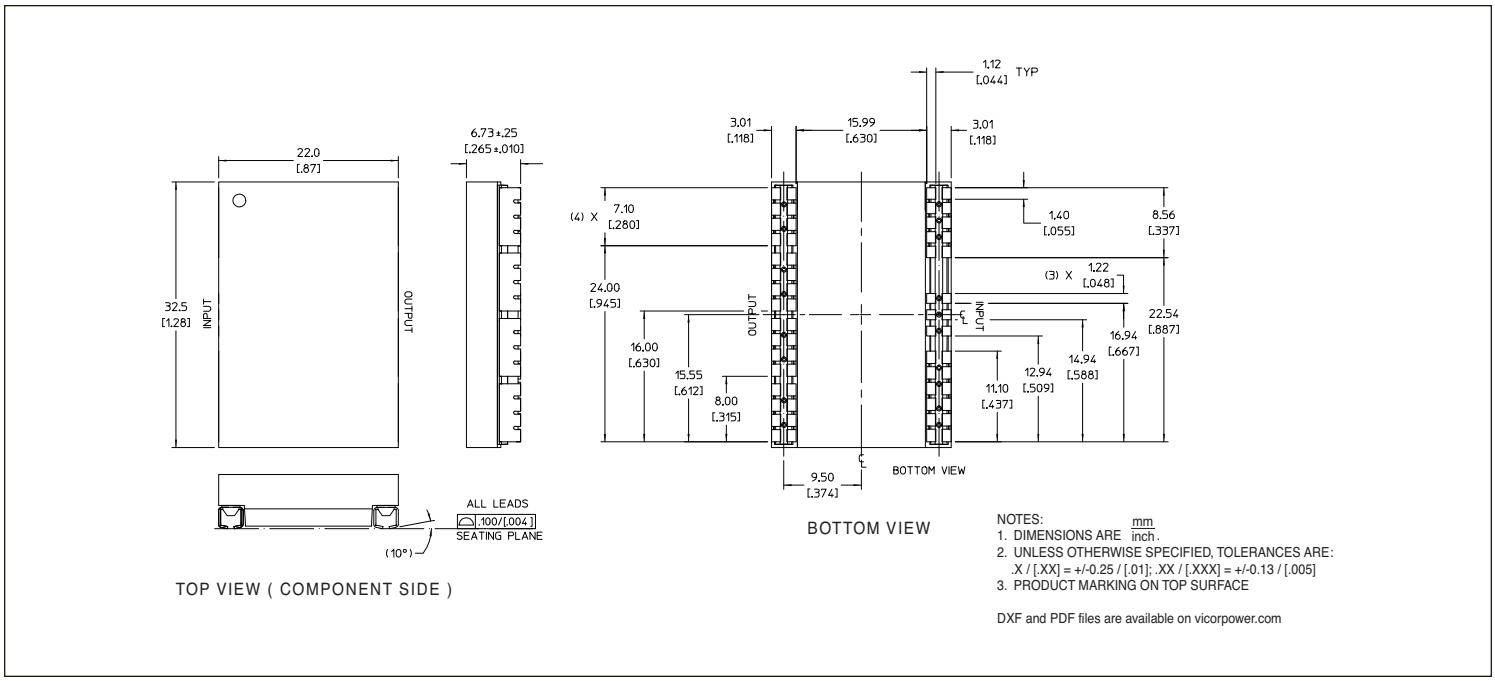


Figure 15 — *BCM™ module J-Lead mechanical outline; onboard mounting*

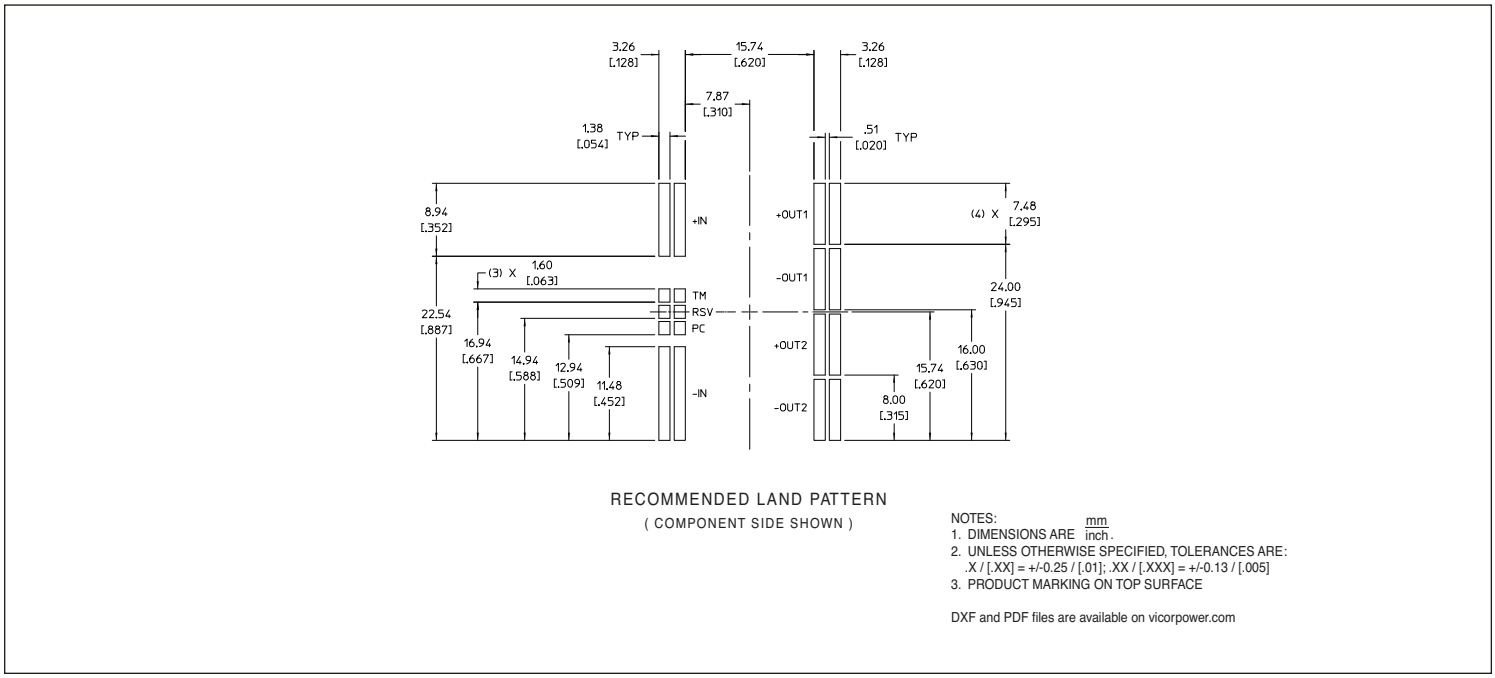


Figure 16 — *BCM™ module PCB land layout information*

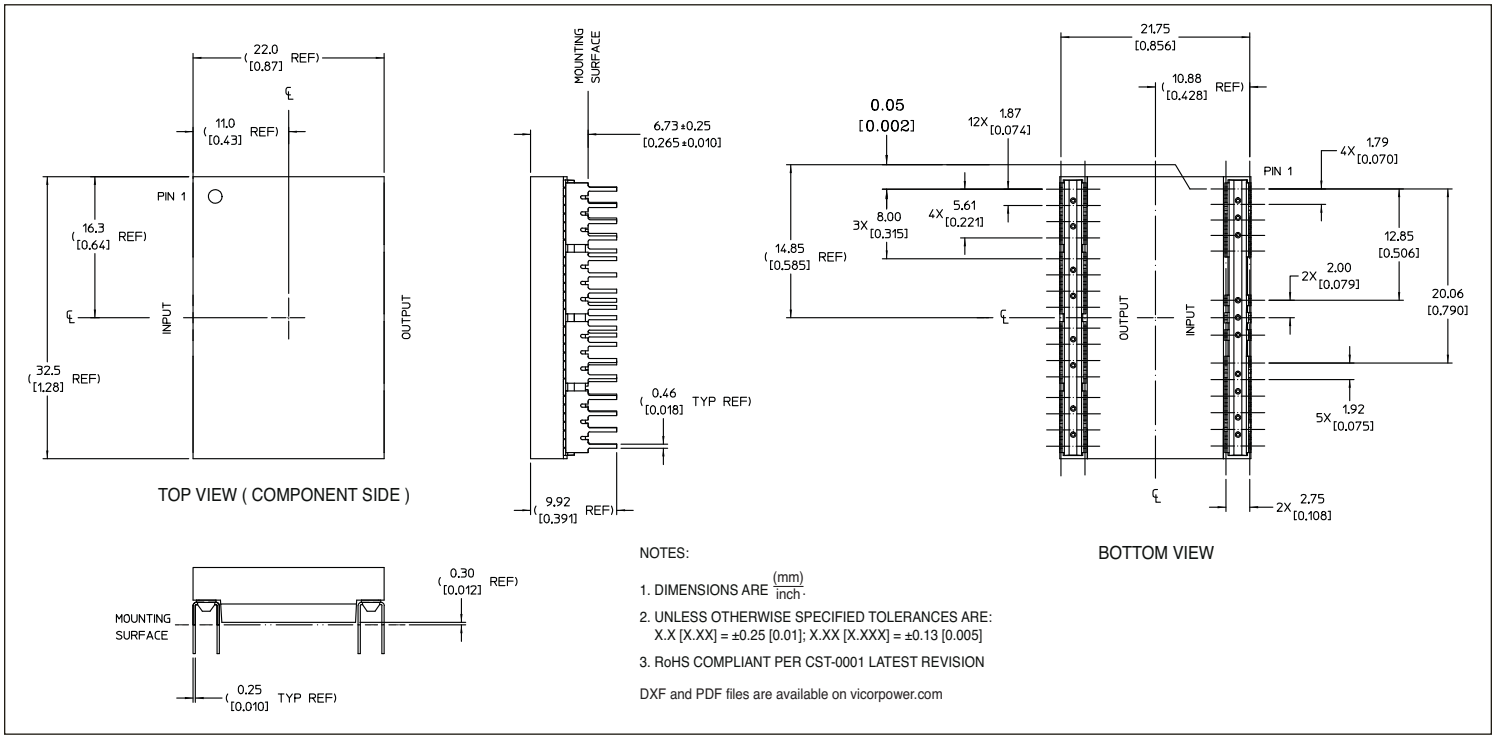


Figure 17 — *BCM™ through-hole module mechanical outline*

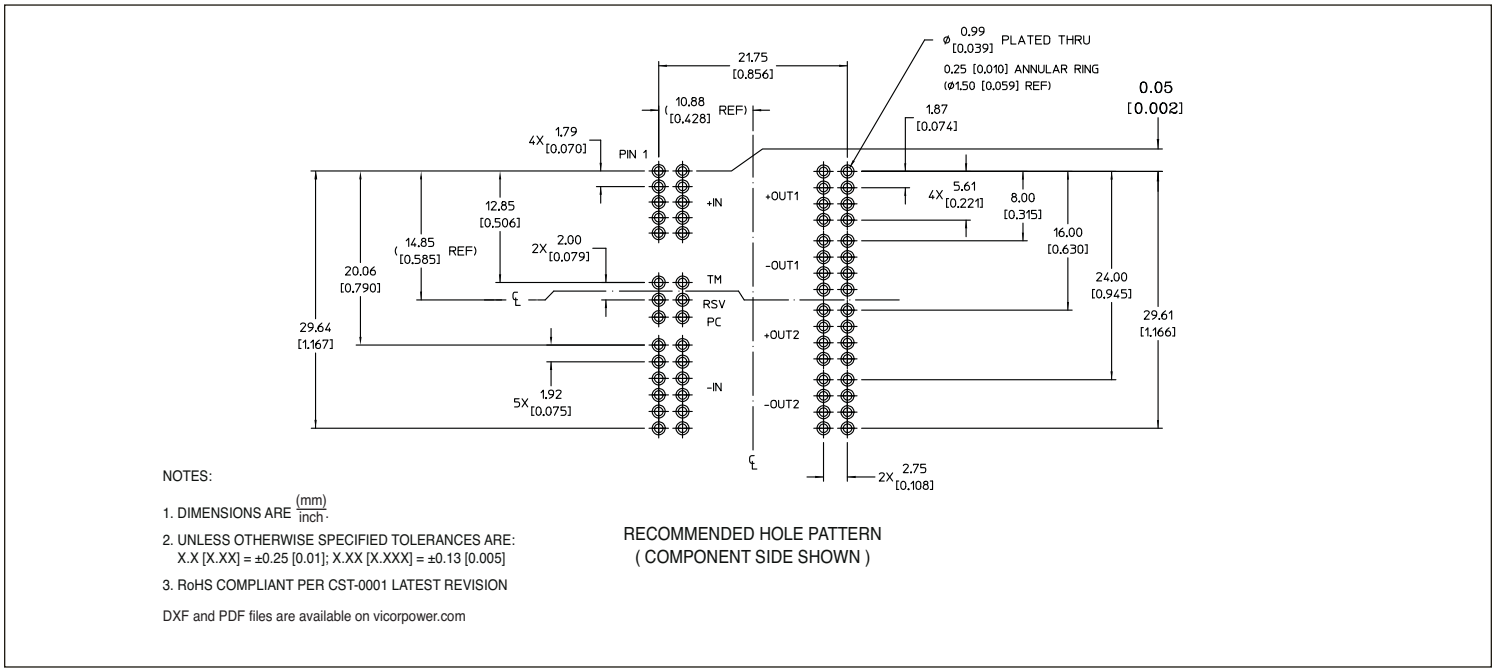
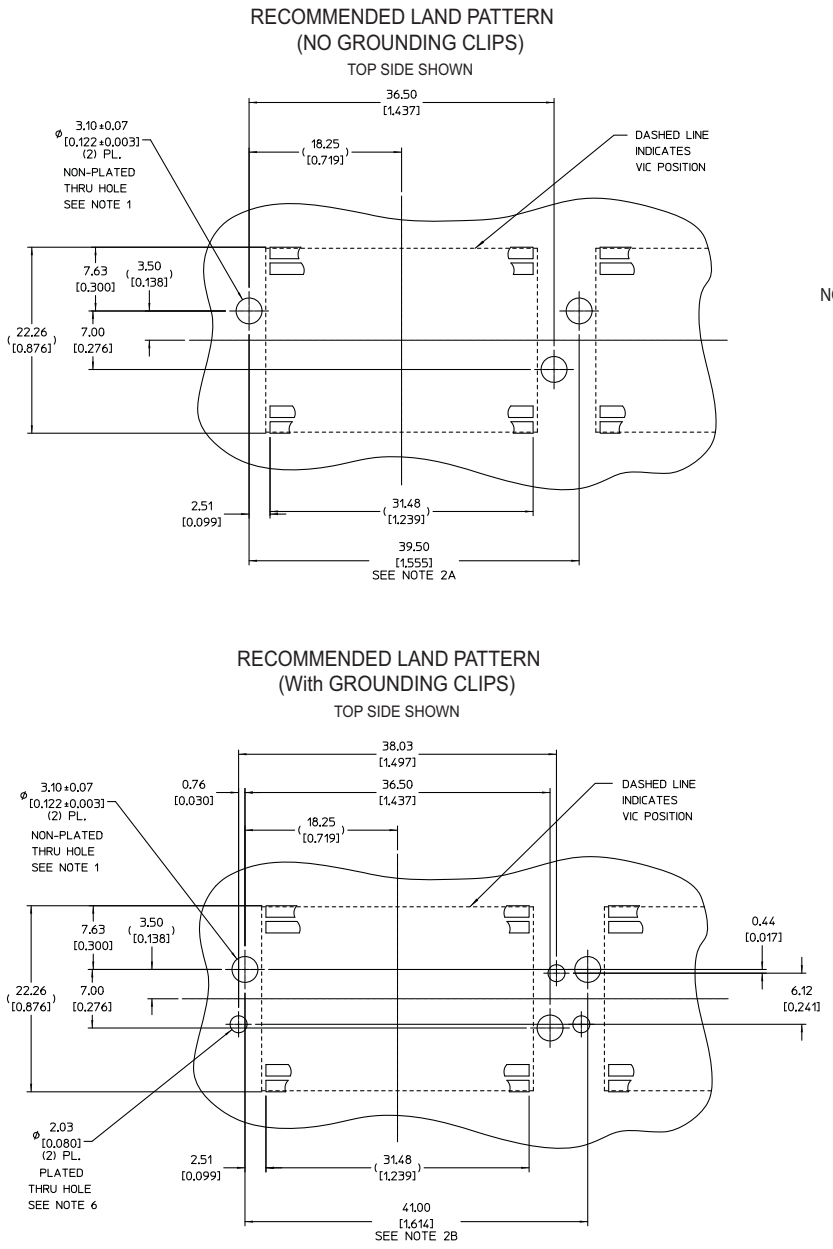


Figure 18 — *BCM™ through-hole module PCB layout information*



NOTES:

1. MAINTAIN 3.50 [0.138] DIA. KEEP-OUT ZONE FREE OF COPPER, ALL PCB LAYERS.

2. (A) MINIMUM RECOMMENDED PITCH IS 39.50 [1.555], THIS PROVIDES 7.00 [0.275] COMPONENT EDGE-TO-EDGE SPACING, AND 0.50 [0.020] CLEARANCE BETWEEN VICOR HEAT SINKS.

   (B) MINIMUM RECOMMENDED PITCH IS 41.00 [1.614], THIS PROVIDES 8.50 [0.334] COMPONENT EDGE-TO-EDGE SPACING, AND 2.00 [0.079] CLEARANCE BETWEEN VICOR HEAT SINKS.

3. V•I CHIP™ MODULE LAND PATTERN SHOWN FOR REFERENCE ONLY; ACTUAL LAND PATTERN MAY DIFFER. DIMENSIONS FROM EDGES OF LAND PATTERN TO PUSH-PIN HOLES WILL BE THE SAME FOR ALL FULL SIZE V•ICHIP PRODUCTS.

4. RoHS COMPLIANT PER CST-0001 LATEST REVISION.

5. UNLESS OTHERWISE SPECIFIED:
   DIMENSIONS ARE MM [INCH].
   TOLERANCES ARE:
   X.X [X.XX] = ±0.3 [0.01]
   X.XX [X.XXX] = ±0.13 [0.005]

6. PLATED THROUGH HOLES FOR GROUNDING CLIPS (33855) SHOWN FOR REFERENCE. HEAT SINK ORIENTATION AND DEVICE PITCH WILL DICTATE FINAL GROUNDING SOLUTION.

Figure 19 — *Hole location for push pin heat sink relative to V•I Chip™ module*

Parallel Operation

The BCM™ bus converter will inherently current share when operated in an array. Arrays may be used for higher power or redundancy in an application.

Current sharing accuracy is maximized when the source and load impedance presented to each bus converter within an array are equal. The recommended method to achieve matched impedances is to dedicate common copper planes within the PCB to deliver and return the current to the array, rather than rely upon traces of varying lengths. In typical applications the current being delivered to the load is larger than that sourced from the input, allowing traces to be utilized on the input side if necessary. The use of dedicated power planes is, however, preferable.

The bus converter power train and control architecture allow bi-directional power transfer, including reverse power processing from the module output to its input. Reverse power transfer is enabled if the module input is within its operating range and the module is otherwise enabled. The bus converter's ability to process power in reverse improves the module's transient response to an output load dump.

Thermal Considerations

V•I Chip™ products are multi-chip modules whose temperature distribution varies greatly for each part number as well as with the input/output conditions, thermal management and environmental conditions. Maintaining the top of the B048F096T24 case to less than 100°C will keep all junctions within the module below 125°C for most applications. The percent of total heat dissipated through the top surface versus through the J-lead is entirely dependent on the particular mechanical and thermal environment. The heat dissipated through the top surface is typically 60%. The heat dissipated through the J-lead onto the PCB board surface is typically 40%. Use 100% top surface dissipation when designing for a conservative cooling solution.
It is not recommended to use a module for an extended period of time at full load without proper heat sinking.

Input Impedance Recommendations

To take full advantage of the BCM™ bus converter capabilities, the impedance presented to its input terminals must be low from DC to approximately 5 MHz. The source should exhibit low inductance and should have a critically damped response. If the interconnect inductance is excessive, the module input pins should be bypassed with an RC damper (e.g., 47 µF in series with 0.3 ohm) to retain low source impedance and proper operation. Given the wide bandwidth of the module, the source response is generally the limiting factor in the overall system response.

Anomalies in the response of the source will appear at the output of the module multiplied by its K factor. The DC resistance of the source should be kept as low as possible to minimize voltage deviations. This is especially important if the module is operated near low or high line as the overvoltage/undervoltage detection circuitry could be activated.

Input Fuse Recommendations

V•I Chip modules are not internally fused in order to provide flexibility in configuring power systems. However, input line fusing of the modules must always be incorporated within the power system. A fast acting fuse should be placed in series with the +In port.

Application Notes

For application notes on soldering, thermal management, board layout, and system design click on the link below:

http://www.vicorpower.com/technical_library/application_information/chips/

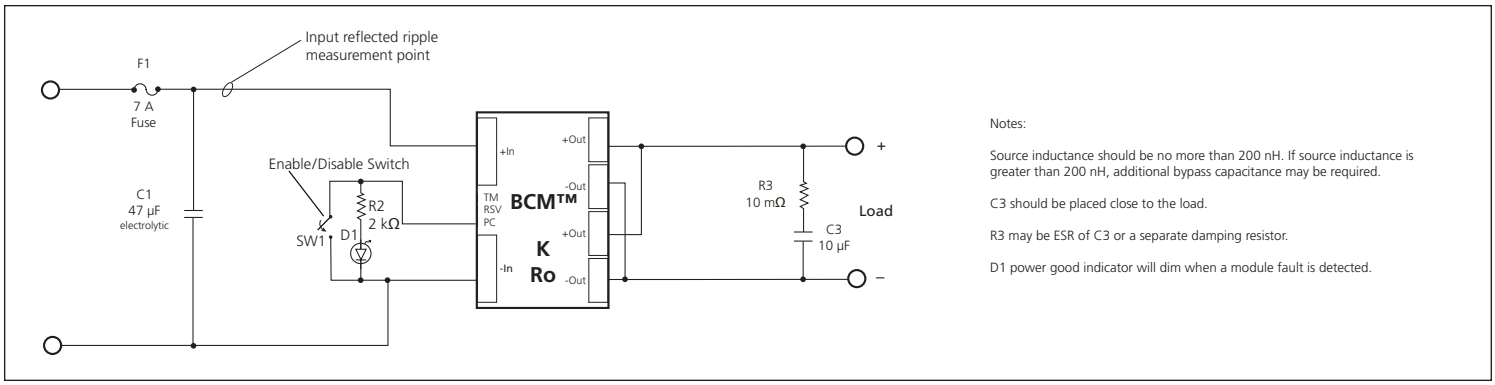


Figure 20 *— BCM™ module test circuit*

BCM™ Bus Converter Level 1 DC Behavioral Model for 48 V to 9.6 V, 240 W

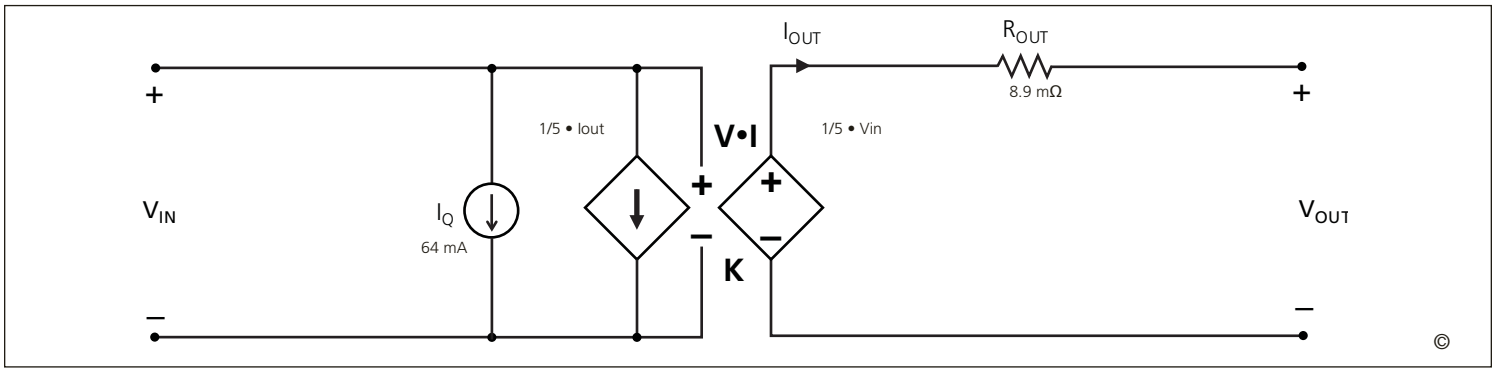


Figure 21 *— This model characterizes the DC operation of the bus converter, including the converter transfer function and its losses. The model enables estimates or simulations of output voltage as a function of input voltage and output load, as well as total converter power dissipation or heat generation.*

BCM™ Bus Converter Level 2 Transient Behavioral Model for 48 V to 9.6 V, 240 W

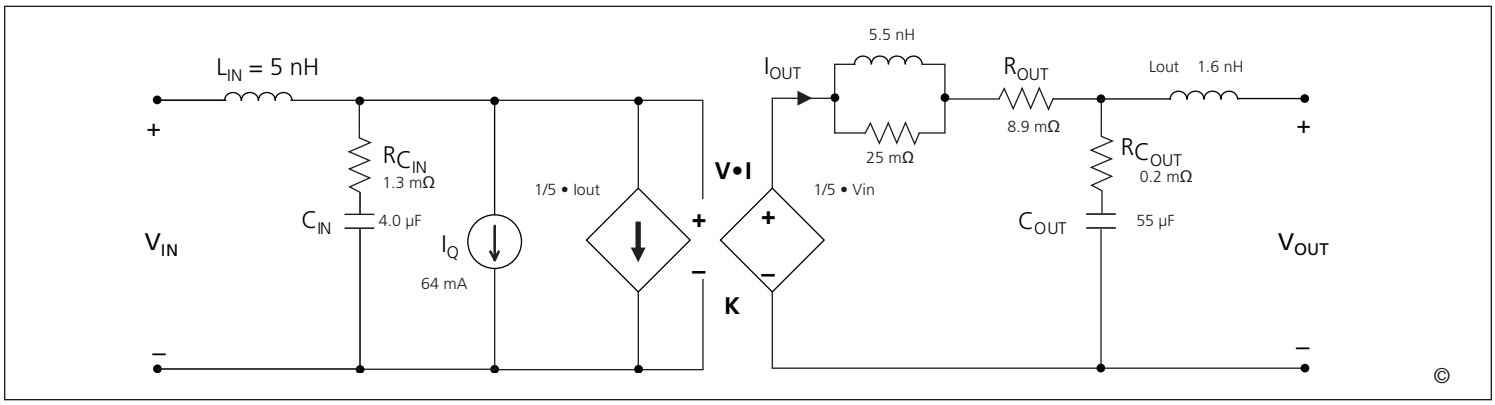


Figure 22 *— This model characterizes the AC operation of the bus converter including response to output load or input voltage transients or steady state modulations. The model enables estimates or simulations of input and output voltages under transient conditions, including response to a stepped load with or without external filtering elements.*

Warranty

Vicor products are guaranteed for two years from date of shipment against defects in material or workmanship when in normal use and service. This warranty does not extend to products subjected to misuse, accident, or improper application or maintenance. Vicor shall not be liable for collateral or consequential damage. This warranty is extended to the original purchaser only.

EXCEPT FOR THE FOREGOING EXPRESS WARRANTY, VICOR MAKES NO WARRANTY, EXPRESS OR IMPLIED, INCLUDING, BUT NOT LIMITED TO, THE WARRANTY OF MERCHANTABILITY OR FITNESS FOR A PARTICULAR PURPOSE.

Vicor will repair or replace defective products in accordance with its own best judgement. For service under this warranty, the buyer must contact Vicor to obtain a Return Material Authorization (RMA) number and shipping instructions. Products returned without prior authorization will be returned to the buyer. The buyer will pay all charges incurred in returning the product to the factory. Vicor will pay all reshipment charges if the product was defective within the terms of this warranty.

Information published by Vicor has been carefully checked and is believed to be accurate; however, no responsibility is assumed for inaccuracies. Vicor reserves the right to make changes to any products without further notice to improve reliability, function, or design. Vicor does not assume any liability arising out of the application or use of any product or circuit; neither does it convey any license under its patent rights nor the rights of others. Vicor general policy does not recommend the use of its components in life support applications wherein a failure or malfunction may directly threaten life or injury. Per Vicor Terms and Conditions of Sale, the user of Vicor components in life support applications assumes all risks of such use and indemnifies Vicor against all damages.

Vicor's comprehensive line of power solutions includes high density AC-DC and DC-DC modules and accessory components, fully configurable AC-DC and DC-DC power supplies, and complete custom power systems.

Information furnished by Vicor is believed to be accurate and reliable. However, no responsibility is assumed by Vicor for its use. Vicor components are not designed to be used in applications, such as life support systems, wherein a failure or malfunction could result in injury or death. All sales are subject to Vicor's Terms and Conditions of Sale, which are available upon request.

Specifications are subject to change without notice.

Intellectual Property Notice

Vicor and its subsidiaries own Intellectual Property (including issued U.S. and Foreign Patents and pending patent applications) relating to the products described in this data sheet. Interested parties should contact Vicor's Intellectual Property Department.

The products described on this data sheet are protected by the following U.S. Patents Numbers:
5,945,130; 6,403,009; 6,710,257; 6,911,848; 6,930,893; 6,934,166; 6,940,013; 6,969,909; 7,038,917; 7,166,898; 7,187,263; 7,361,844; D496,906; D505,114; D506,438; D509,472 and for use under 6,975,098 and 6,984,965

Vicor Corporation
25 Frontage Road
Andover, MA, USA 01810
Tel: 800-735-6200
Fax: 978-475-6715

email
Customer Service: custserv@vicorpower.com
Technical Support: apps@vicorpower.com