CS8420

Digital Audio Sample Rate Converter

Features

- Complete IEC60958, AES3, S/PDIF, EIAJ CP1201 compatible transceiver with asynchronous sample rate converter
- Flexible 3-wire serial digital i/o ports
- 8 kHz to 108 kHz sample rate range
- 1:3 and 3:1 maximum input to output sample rate ratio
- 128 dB dynamic range
- -117 dB THD+N at 1 kHz
- Excellent performance at almost a 1:1 ratio
- Excellent clock jitter rejection
- 24 bit i/o words
- Pin and micro-controller read/write access to Channel Status and User Data
- Micro-controller and stand-alone modes

General Description

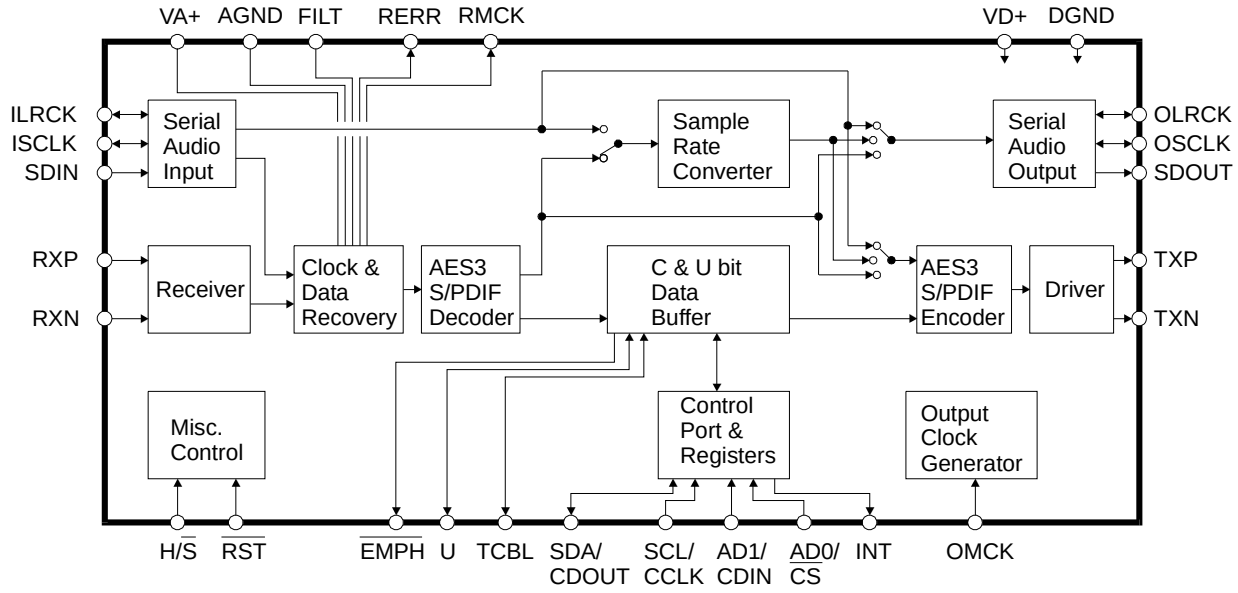
The CS8420 is a stereo digital audio sample rate converter (SRC) with AES3 type and serial digital audio inputs, AES3 type and serial digital audio outputs, along with comprehensive control ability via a 4-wire microcontroller port. Channel status and user data can be assembled in block sized buffers, making read/modify/write cycles easy.

Digital audio inputs and outputs may be 24, 20 or 16 bits. The input data can be completely asynchronous to the output data, with the output data being synchronous to an external system clock.

Target applications include CD-R, DAT, MD, DVD and VTR equipment, mixing consoles, digital audio transmission equipment, high quality D/A and A/D converters, effects processors and computer audio systems.

ORDERING INFO

CS8420-CS 28-pin SOIC, -10 to +70°C temp. range
CDB8420 Evaluation Board



Preliminary Product Information | This document contains information for a new product.
Cirrus Logic reserves the right to modify this product without notice.

CIRRUS LOGIC®
P.O. Box 17847, Austin, Texas 78760
(512) 445 7222 FAX: (512) 445 7581
http://www.cirrus.com

Copyright © Cirrus Logic, Inc. 1999
(All Rights Reserved)

TABLE OF CONTENTS

1. CHARACTERISTICS/SPECIFICATIONS 5
PERFORMANCE SPECIFICATIONS 5
DIGITAL FILTER CHARACTERISTICS 5
POWER AND THERMAL CHARACTERISTICS 5
DIGITAL CHARACTERISTICS 6
SWITCHING CHARACTERISTICS 6
SWITCHING CHARACTERISTICS - SERIAL AUDIO PORTS 7
SWITCHING CHARACTERISTICS - CONTROL PORT - SPI MODE 8
SWITCHING CHARACTERISTICS - CONTROL PORT - I^2C® MODE 9
2. TYPICAL CONNECTION DIAGRAM 10
3. GENERAL DESCRIPTION 11
4. DATA I/O FLOW AND CLOCKING OPTIONS 12
5. SAMPLE RATE CONVERTER (SRC) 15
5.1 Dither 15
5.2 SRC Locking, Varispeed and the Sample Rate Ratio Register 15
6. THREE-WIRE SERIAL AUDIO PORTS 16
7. AES3 TRANSMITTER AND RECEIVER 19
7.1 AES3 Receiver 19
7.1.1 PLL, Jitter Attenuation, and Varispeed 19
8. OMCK OUT ON RMCK 21
9. PLL EXTERNAL COMPONENTS 21
9.1 Error Reporting and Hold Function 21
9.2 Channel Status Data Handling 21
9.3 User Data Handling 22
9.4 Non-Audio Auto Detection 22
9.5 AES3 Transmitter 23
9.5.1 Transmitted Frame and Channel Status Boundary Timing 23
9.5.2 TXN and TXP Drivers 23
9.6 Mono Mode Operation 24
10. CONTROL PORT DESCRIPTION AND TIMING 26
10.1 SPI Mode 26
10.2 I^2C Mode 27
10.3 Interrupts 27
11. CONTROL PORT REGISTER BIT DEFINITIONS 28
12. SYSTEM AND APPLICATIONS ISSUES 44
12.1 Reset, Power Down and Start-up Options 44
12.2 ID Code and Revision Code 44

Contacting Cirrus Logic Support

For a complete listing of Direct Sales, Distributor, and Sales Representative contacts, visit the Cirrus Logic web site at:
http://www.cirrus.com/corporate/contacts/

Preliminary product information describes products which are in production, but for which full characterization data is not yet available. Advance product information describes products which are in development and subject to development changes. Cirrus Logic, Inc. has made best efforts to ensure that the information contained in this document is accurate and reliable. However, the information is subject to change without notice and is provided "AS IS" without warranty of any kind (express or implied). No responsibility is assumed by Cirrus Logic, Inc. for the use of this information, nor for infringements of patents or other rights of third parties. This document is the property of Cirrus Logic, Inc. and implies no license under patents, copyrights, trademarks, or trade secrets. No part of this publication may be copied, reproduced, stored in a retrieval system, or transmitted, in any form or by any means (electronic, mechanical, photographic, or otherwise) without the prior written consent of Cirrus Logic, Inc. Items from any Cirrus Logic website or disk may be printed for use by the user. However, no part of the printout or electronic files may be copied, reproduced, stored in a retrieval system, or transmitted, in any form or by any means (electronic, mechanical, photographic, or otherwise) without the prior written consent of Cirrus Logic, Inc.Furthermore, no part of this publication may be used as a basis for manufacture or sale of any items without the prior written consent of Cirrus Logic, Inc. The names of products of Cirrus Logic, Inc. or other vendors and suppliers appearing in this document may be trademarks or service marks of their respective owners which may be registered in some jurisdictions. A list of Cirrus Logic, Inc. trademarks and service marks can be found at http://www.cirrus.com.

12.3 Power Supply, Grounding, and PCB layout 44
12.4 Synchronization of Multiple CS8420s 45
12.5 Extended Range Sample Rate Conversion 45
13. SOFTWARE MODE - PIN DESCRIPTION 46
14. HARDWARE MODES 49
14.1 Overall Description 49
14.1.1 Hardware Mode Definitions 49
14.1.2 Serial Audio Port Formats 49
14.2 Hardware Mode 1 Description 50
14.2.1 Pin Description - Hardware Mode 1 51
14.3 Hardware Mode 2 Description 53
14.3.1 Pin Description - Hardware Mode 2 54
14.4 Hardware Mode 3 Description 56
14.4.1 Pin Description - Hardware Mode 3 58
14.5 Hardware Mode 4 Description 60
14.5.1 Pin Description - Hardware Mode 4 62
14.6 Hardware Mode 5 Description 64
14.6.1 Pin Description - Hardware Mode 5 65
14.7 Hardware Mode 6 Description 67
14.7.1 Pin Description - Hardware Mode 6 68
15. APPENDIX A: EXTERNAL AES3/SPDIF/IEC60958 TRANSMITTER AND RECEIVER COMPONENTS 70
15.1 AES3 Transmitter External Components 70
15.2 AES3 Receiver External Components 70
15.3 Isolating Transformer Requirements 71
16. APPENDIX B: CHANNEL STATUS AND USER DATA BUFFER MANAGEMENT 72
16.1 AES3 Channel Status(C) Bit Management 72
16.1.1 Manually accessing the E buffer 72
16.1.2 Reserving the first 5 bytes in the E buffer 74
16.1.3 Serial Copy Management System (SCMS) 74
16.1.4 Channel Status Data E Buffer Access 74
16.1.5 One Byte mode 74
16.1.6 Two Byte mode 74
16.2 AES3 User (U) Bit Management 75
16.2.1 Mode 1: Transmit All Zeros 75
16.2.2 Mode 2: Block Mode 75
16.2.3 IEC60958 Recommended U Data Format For Consumer Applications 75
16.2.4 Mode (3): Reserved 76
16.2.5 Mode (4): IEC Consumer B 76
17. PARAMETER DEFINITIONS 77
18. PACKAGE DIMENSIONS 78

LIST OF FIGURES

Figure 1. Audio Ports Master Mode Timing 7
Figure 2. Audio Ports Slave Mode and Data I/O Timing 7
Figure 3. SPI Mode Timing 8
Figure 4. I^2C Mode Timing 9
Figure 5. Recommended Connection Diagram for Software Mode 10
Figure 6. Software Mode Audio Data Flow Switching Options 12
Figure 7. Serial Audio Input, using PLL, SRC enabled 13
Figure 8. Serial Audio Input, No PLL, SRC enabled 13
Figure 9. AES3 Input, SRC enabled 13
Figure 10. Serial Audio Input, AES3 Input Clock Source, SRC Enabled 13

Figure 11. Serial Audio Input, SRC Output clocked by AES3 Recovered Clock 14
Figure 12. AES3 Input, SRC to Serial Audio Output, Serial Audio Input to AES3 Out 14
Figure 13. AES3 Input to Serial Audio Output, Serial Audio Input to AES3 Out, no SRC 14
Figure 14. AES3 Input to Serial Audio Output Only 14
Figure 15. Input Serial Port to AES3 Transmitter 14
Figure 16. Serial Audio Input Example Formats 17
Figure 17. Serial Audio Output Example Formats 18
Figure 18. Jitter Attenuation Characteristics of PLL with “slow” Filter Components 20
Figure 19. Jitter Attenuation Characteristics of PLL with “medium” Filter Components 20
Figure 20. Jitter Attenuation Characteristics of PLL with “fast” Filter Components 20
Figure 21. AES3 Receiver Timing for C & U pin output data 22
Figure 22. AES3 Transmitter Timing for C, U and V pin input data 24
Figure 23. Mono Mode Operation Compared to Normal Stereo Operation 25
Figure 24. Control Port Timing in SPI Mode 26
Figure 25. Control Port Timing in I^2C Mode 27
Figure 26. Hardware Mode 1 - Default Data Flow, AES3 Input 50
Figure 27. Hardware Mode 2 - Default Data Flow, Serial Audio Input 53
Figure 28. Hardware Mode 3 - Transceive Data Flow, with SRC 56
Figure 29. Hardware Mode 4 - Transceive Data Flow, without SRC 60
Figure 30. Hardware Mode 5 - AES3 Receiver Only 64
Figure 31. Hardware Mode 6 - AES3 Transmitter Only 67
Figure 32. Professional Output Circuit 70
Figure 33. Consumer Output Circuit 70
Figure 34. TTL/CMOS Output Circuit 70
Figure 35. Professional Input Circuit 71
Figure 36. Transformerless Professional Input Circuit 71
Figure 37. Consumer Input Circuit 71
Figure 38. TTL/CMOS Input Circuit 71
Figure 39. Channel Status Data Buffer Structure 72
Figure 40. Channel Status Block Handling When Fso is Not Equal to Fsi 73
Figure 41. Flowchart for Reading the E Buffer 73
Figure 42. Flowchart for Writing the E Buffer 73

LIST OF TABLES

Table 1. PLL External Component Values 21
Table 2. Summary of all Bits in the Control Register Map 29
Table 3. Hardware Mode Definitions 49
Table 4. Serial Audio Output Formats Available in Hardware Mode 49
Table 5. Serial Audio Input Formats Available in Hardware Mode 49
Table 6. Hardware Mode 1 Start-up Options 50
Table 7. HW Mode 2A COPY/C and ORIG/U Pin Function 53
Table 8. HW Mode 2 Serial Audio Port Format Selection 53
Table 9. Hardware Mode 2 Start-up Options 53
Table 10. Hardware Mode 3 Start-up Options 57
Table 11. Hardware Mode 4 Start-up Options 61
Table 12. Hardware Mode 5 Start-up Options 64
Table 13. HW 6C COPY/C and ORIG pin function 67
Table 14. HW 6 Serial Audio Port Format Selection 67

1. CHARACTERISTICS/SPECIFICATIONS

PERFORMANCE SPECIFICATIONS (T_A = 25 °C; VA+ = VD+ = 5V ±5%)

Parameter*	Symbol	Min	Typ	Max	Units
Dynamic Range		120	128	-	dB
Input Sample Rate (serial input port)	Fsi	8	-	108	kHz
Output Sample Rate	Fso	8	-	108	kHz
Output to Input Sample Rate Ratio		0.33	-	3	
Total Harmonic Distortion + Noise	THD+N				
1 kHz, -1dBFS, 0.33 < Fso/Fsi < 1.7		-	-	-117	dB
1 kHz, -1dBFS, 0.33 < Fso/Fsi < 3		-	-	-112	dB
10 kHz, -1dBFS, 0.33 < Fso/Fsi < 1.7		-	-	-110	dB
10 kHz, -1dBFS, 0.33 < Fso/Fsi < 3		-	-	-107	dB
Peak idle channel noise component		-	-	-140	dBFS
Input Jitter Tolerance of SRC		-	-	TBD	ns
Resolution		16	-	24	bits
Gain Error		-0.12	-	0	dB

DIGITAL FILTER CHARACTERISTICS (T_A = 25 °C; VA+ = VD+ = 5V ±5%)

Parameter*	Symbol	Min	Typ	Max	Units
Passband Upsampling		0	-	0.4535*Fsi	Hz
Downsampling		0	-	0.4535*Fso	Hz
Passband Ripple		-	-	±0.007	dB
Stopband (Downsampling)		0.5465*Fso	-	Fsi/2	Hz
Stopband Attenuation		110	-	-	dB
Group Delay (Note 1)	t_{gd}	-	-	1.75	ms
Group Delay Variation vs. Frequency	Δt_{gd}	-	-	0.0	µs
Interchannel Phase Deviation		-	-	0.0	°

Notes: 1. The value shown is for Fsi = Fso = 48 kHz. The group delay scales with input and output sample rate according to the following formula: t_{gd} = 41/Fsi + 43/Fso

POWER AND THERMAL CHARACTERISTICS (AGND, DGND = 0V, all voltages with respect to ground)

Parameter	Symbol	Min	Typ	Max	Units
Power Supply Voltage	VD+,VA+	4.75	5.0	5.25	V
Power Consumption at 96 kHz Fso and Fsi		-	660	TBD	mW
Power Consumption at 48 kHz Fso and Fsi		-	350	TBD	mW
Supply Current at 96 kHz Fso and Fsi VA+		-	7.0	TBD	mA
VD+		-	125	TBD	mA
Supply Current in power down ($\overline{RST}$ high, VD+ & VA+)		-	0.5	-	mA
Ambient Operating Temperature (Note 2)	T_A	-10	25	70	°C
Junction Temperature	T_J	-	-	135	°C
Junction to Ambient thermal impedance (28 pin SOIC)	θ_{JA}	-	65	-	°C/W

Notes: 2. '-CS' parts are specified to operate over -10°C to 70 °C but are tested at 25 °C only.

* Parameter Definitions are given at the end of this data sheet

ABSOLUTE MAXIMUM RATINGS (AGND, DGND = 0V, all voltages with respect to ground)

Parameter	Symbol	Min	Max	Units
Power Supply Voltage	VD+,VA+	-	6.0	V
Input Current, Any Pin Except Supply, RXP, RXN (Note 3)	I_{in}	-	±10	mA
Input Current, RXP, RXN	I_{in}	±0.25	±TBD	mA
Input Voltage	V_{in}	-0.3	(VD+) + 0.3	V
Ambient Operating Temperature (power applied)	T_A	-55	125	°C
Storage Temperature	T_{stg}	-65	150	°C

Notes: 3. Transient currents of up to 100mA will not cause SCR latch-up.

DIGITAL CHARACTERISTICS (T_A = 25 °C; VA+ = VD+ = 5V ±5%)

Parameter	Symbol	Min	Typ	Max	Units
High-Level Input Voltage, except RXP, RXN	V_{IH}	2.0	-	(VD+) + 0.3	V
Low-Level Input Voltage, except RXP, RXN	V_{IL}	-0.3	-	0.8	V
Low-Level Output Voltage, (Io=-20uA), except TXP, TXN	V_{OL}	-	-	0.4	V
High-Level Output Voltage, (Io=20uA), except TXP, TXN	V_{OH}	(VD+) - 1	-	-	V
Input Leakage Current	I_{in}	-	±10	±15	µA
Differential Input Voltage, RXP to RXN	V_{TH}	200	-	-	mV
Output High Voltage, TXP, TXN (I_{OH} = -21mA)		(VD+) - 0.7	(VD+) - 0.4	-	V
Output Low Voltage, TXP, TXN (I_{OL} = 21mA)		-	0.4	0.7	V

SWITCHING CHARACTERISTICS (T_A = 25 °C; VA+ = VD+ = 5V ±5%, Inputs: Logic 0 = 0V, Logic 1 = VD+; C_L = 20 pF)

Parameter	Symbol	Min	Typ	Max	Units
$\overline{\text{RST}}$ pin Low Pulse Width		200	-	-	µs
OMCK Frequency for OMCK = 512*Fso		4.096	-	55.3	MHz
OMCK Low and High Width for OMCK = 512*Fso		8.2	-	-	ns
OMCK Frequency for OMCK = 384*Fso		3.072	-	41.5	MHz
OMCK Low and High Width for OMCK = 384*Fso		12.3	-	-	ns
OMCK Frequency for OMCK = 256*Fso		2.048	-	27.7	MHz
OMCK Low and High Width for OMCK = 256*Fso		16.4	-	-	ns
PLL Clock Recovery Sample Rate Range		8.0	-	108.0	kHz
RMCK output jitter		-	200	-	ps RMS
RMCK output duty cycle		40	50	60	%
RMCK Input Frequency (Note 4)		2.048	-	27.7	MHz
RMCK Input Low and High Width (Note 4)		16.4	-	-	ns
AES3 Transmitter Output Jitter		-	-	1	ns

Notes: 4. PLL is bypassed, clock is input to the RMCK pin. The value given is guaranteed to work, with an external RMCK applied the part will actually work at much lower frequencies.

SWITCHING CHARACTERISTICS - SERIAL AUDIO PORTS (T_A = 25 °C; VA+ = VD+ = 5V ±5%, Inputs: Logic 0 = 0V, Logic 1 = VD+; C_L = 20 pF)

Parameter		Symbol	Min	Typ	Max	Units
OSCLK Active Edge to SDOUT Output Valid	(Note 5)	t_{dpd}	-	-	20	ns
SDIN Setup TIme Before ISCLK Active Edge	(Note 5)	t_{ds}	20	-	-	ns
SDIN Hold Time After ISCLK Active Edge	(Note 5)	t_{dh}	20	-	-	ns
Master Mode						
O/RMCK to I/OSCLK active edge delay	(Note 5)	t_{smd}	0	-	10	ns
O/RMCK to I/OLRCK delay	(Note 6)	t_{lmd}	0	-	10	ns
I/OSCLK and I/OLRCK Duty Cycle			-	50	-	%
Slave Mode						
I/OSCLK Period		t_{sckw}	36	-	-	ns
I/OSCLK Input Low Width		t_{sckl}	14	-	-	ns
I/OSCLK Input High Width		t_{sckh}	14	-	-	ns
I/OSCLK Active Edge to I/OLRCK Edge	(Note 5,6,7)	t_{lrckd}	20	-	-	ns
I/OLRCK Edge Setup Before I/OSCLK Active Edge	(Note 5,6,8)	t_{lrcks}	20	-	-	ns

Notes:
5. The active edges of ISCLK and OSCLK are programmable.
6. The polarity of ILRCK and OLRCK is programmable.
7. This delay is to prevent the previous I/OSCLK edge from being interpreted as the first one after I/OLRCK has changed.
8. This setup time ensures that this I/OSCLK edge is interpreted as the first one after I/OLRCK has changed.

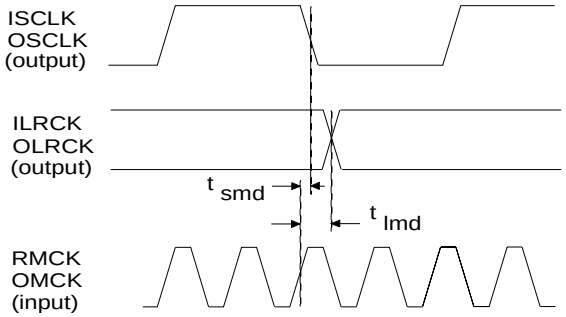


Figure 1. Audio Ports Master Mode Timing

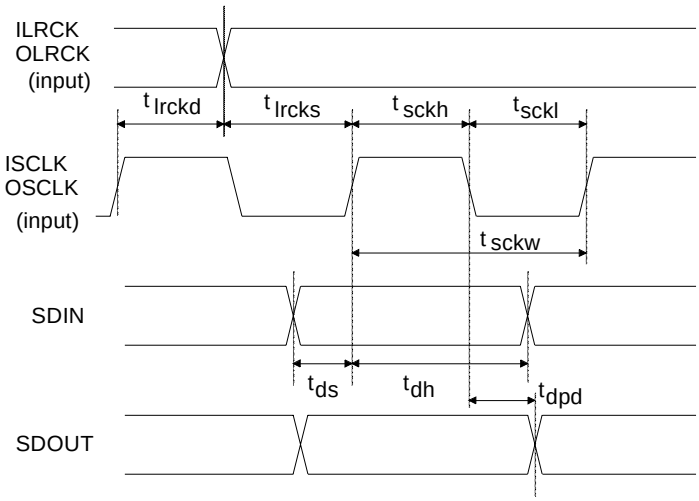


Figure 2. Audio Ports Slave Mode and Data I/O Timing

SWITCHING CHARACTERISTICS - CONTROL PORT - SPI MODE (T_A = 25 °C;

VA+ = VD+ = 5V ±5%, Inputs: Logic 0 = 0V, Logic 1 = VD+; C_L = 20 pF)

Parameter		Symbol	Min	Typ	Max	Units
CCLK Clock Frequency	(Note 9)	f_{sck}	0	-	6.0	MHz
$\overline{CS}$ High Time Between Transmissions		t_{csh}	1.0	-	-	µs
$\overline{CS}$ Falling to CCLK Edge		t_{css}	20	-	-	ns
CCLK Low Time		t_{scl}	66	-	-	ns
CCLK High Time		t_{sch}	66	-	-	ns
CDIN to CCLK Rising Setup Time		t_{dsu}	40	-	-	ns
CCLK Rising to DATA Hold Time	(Note 10)	t_{dh}	15	-	-	ns
CCLK Falling to CDOUT Stable		t_{pd}	-	-	45	ns
Rise Time of CDOUT		t_{r1}	-	-	25	ns
Fall Time of CDOUT		t_{f1}	-	-	25	ns
Rise Time of CCLK and CDIN	(Note 11)	t_{r2}	-	-	100	ns
Fall Time of CCLK and CDIN	(Note 11)	t_{f2}	-	-	100	ns

Notes:
9. If Fso or Fsi is lower than 46.875 kHz, then maximum CCLK frequency should be less than 128Fso and less than 128Fsi. This is dictated by the timing requirements necessary to access the Channel Status and User Bit buffer memory. Access to the control register file can be carried out at the full 6 MHz rate. The minimum allowable input sample rate is 8 kHz, so choosing CCLK of less than or equal to 1.024 MHz should be safe for all possible conditions
10. Data must be held for sufficient time to bridge the transition time of CCLK.
11. For f_{sck} <1 MHz.

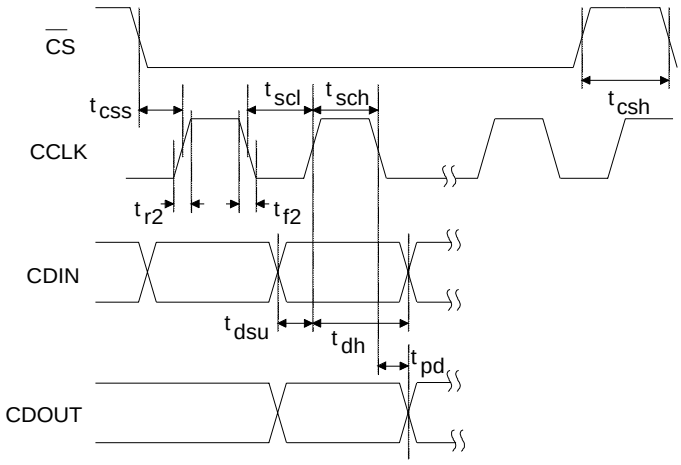


Figure 3. SPI Mode Timing

SWITCHING CHARACTERISTICS - CONTROL PORT - I²C® MODE (Note 12, T_A = 25 °C; VA+ = VD+ = 5V ±5%, Inputs: Logic 0 = 0V, Logic 1 = VD+; C_L = 20 pF)

Parameter	Symbol	Min	Typ	Max	Units
SCL Clock Frequency	f_{scl}	-	-	100	kHz
Bus Free Time Between Transmissions	t_{buf}	4.7	-	-	µs
Start Condition Hold Time (prior to first clock pulse)	t_{hdst}	4.0	-	-	µs
Clock Low Time	t_{low}	4.7	-	-	µs
Clock High Time	t_{high}	4.0	-	-	µs
Setup Time for Repeated Start Condition	t_{sust}	4.7	-	-	µs
SDA Hold Time from SCL Falling (Note 13)	t_{hdd}	0	-	-	µs
SDA Setup Time to SCL Rising	t_{sud}	250	-	-	ns
Rise Time of Both SDA and SCL Lines	t_r	-	-	1	µs
Fall Time of Both SDA and SCL Lines	t_f	-	-	300	ns
Setup Time for Stop Condition	t_{susp}	4.7	-	-	µs

Notes: 12. I²C is a registered trademark of Philips Semiconductors.

13. Data must be held for sufficient time to bridge the 300ns transition time of SCL.

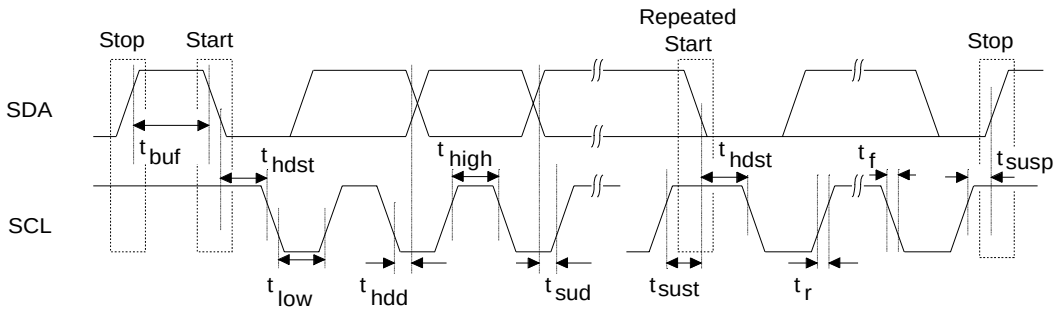
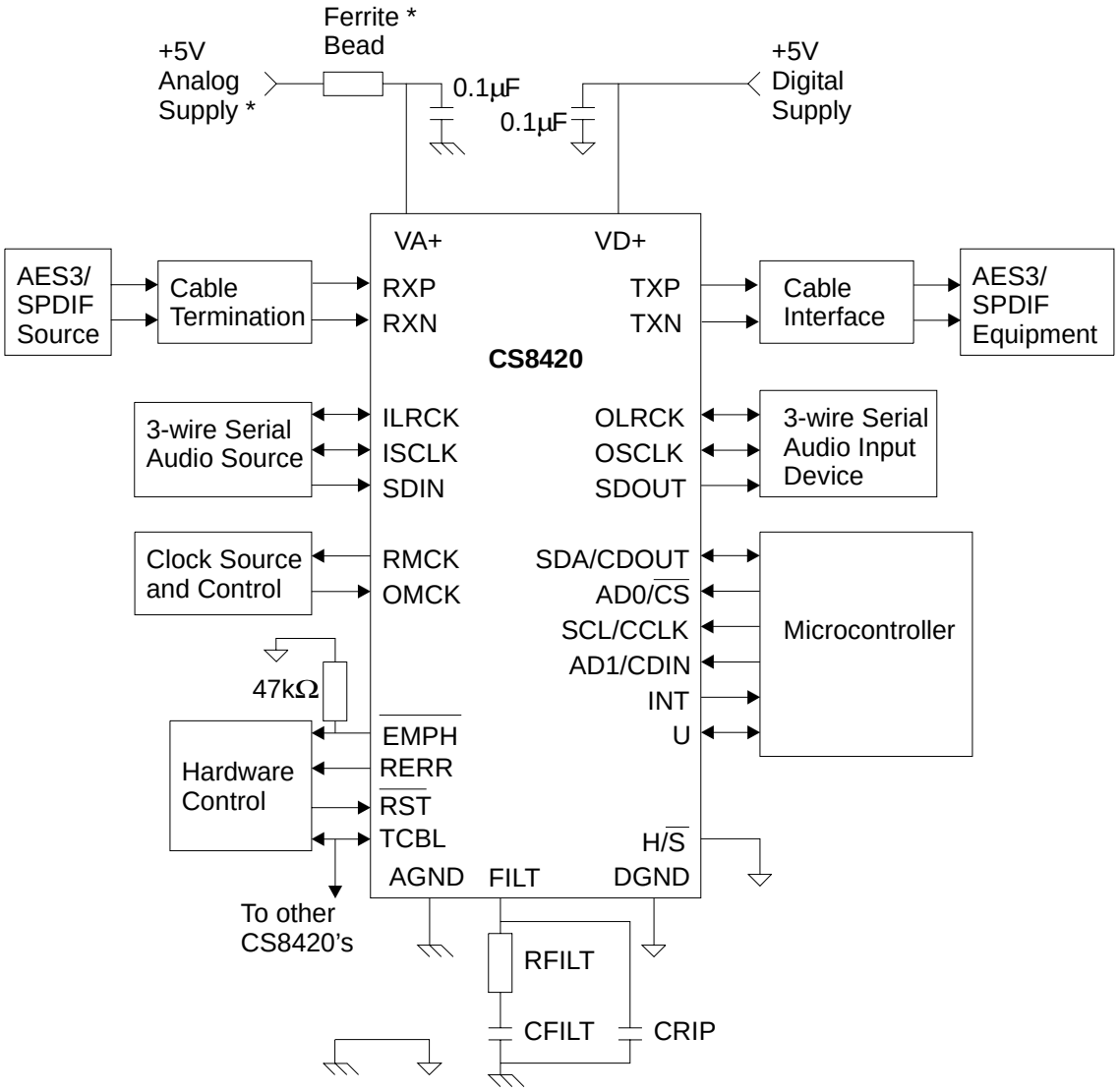


Figure 4. I²C Mode Timing

Specifications are subject to change without notice

2. TYPICAL CONNECTION DIAGRAM



* A separate analog supply is only necessary in applications where RMCK is used for a jitter sensitive task. For applications where RMCK is not used for a jitter sensitive task, connect VA+ to VD+ via a ferrite bead. Keep the decoupling capacitor between VA+ and AGND.

Figure 5. Recommended Connection Diagram for Software Mode

3. GENERAL DESCRIPTION

The CS8420 is a fully asynchronous sample rate converter plus AES3 transceiver intended to be used in digital audio systems. Such systems include digital mixing consoles, effects processors, tape recorders and computer multimedia systems. The CS8420 is intended for 16, 20, and 24-bit applications where the input sample rate is unknown, or is known to be asynchronous to the system sample rate.

On the input side of the CS8420, AES3 or a 3-wire serial format can be chosen. The output side produces both AES3 and a 3-wire serial format. An I^2C/SPI compatible microcontroller interface allows full block processing of channel status and user data via block reads from the incoming AES3 data stream and block writes to the outgoing AES3 data stream. The user can also access information decoded from the input AES3 data stream, such as the presence of non-audio data and pre-emphasis, as well as control the various modes of the device. For users who prefer not to use a micro-controller, six hardware modes have been provided, documented towards the end of this data sheet. In these modes, flexibility is limited, with pins providing some programmability.

When used for AES3 in, AES3 out applications, the CS8420 can automatically transceive user data that conforms to the IEC60958 recommended format. The CS8420 also allows access to the relevant bits in the AES3 data stream to comply with the serial copy management system (SCMS).

The diagram on the cover of this data sheet shows the main functional blocks of the CS8420. Figure 5 shows the supply and external connections to the device.

Familiarity with the AES3 and IEC60958 specifications are assumed throughout this document. The Application Note: “Overview of Digital Audio Interface Data Structures”, contains a tutorial on digital audio specifications. The paper “An Understanding and Implementation of the SCMS Serial Copy Management System for Digital Audio Transmission”, by Clif Sanchez, is an excellent tutorial on SCMS. It may be obtained from Crystal Semiconductor, or from the AES.

To guarantee system compliance, the proper standards documents should be obtained. The latest AES3 standard should be obtained from the Audio Engineering Society or ANSI, the latest IEC60958 standard from the International Electrotechnical Commission and the latest EIAJ CP-1201 standard from the Japanese Electronics Bureau.

4. DATA I/O FLOW AND CLOCKING OPTIONS

The CS8420 can be configured for nine connectivity alternatives, called data flows. Each data flow has an associated clocking set-up. Figure 6 shows the data flow switching, along with the control register bits which control the switches; this drawing only shows the audio data paths for simplicity.

The AESBP switch allows a TTL level, already biphase mark encoded, data stream connected to RXP to be routed to the TXP and TXN pin drivers. The TXOFF switch causes the TXP and TXN outputs to be driven to ground.

In modes including the SRC function, there are two audio data related clock domains. One domain includes the input side of SRC, plus the attached data source. The second domain includes the output side of the SRC, plus any attached output ports.

There are two possible clock sources. The first is known as the recovered clock, is the output of a PLL, and is connected to the RCMK pin. The input to the PLL can be either the incoming AES3 data stream, or the ILRCK word rate clock from the serial audio input port. The second clock is input via the OMCK pin, and would normally be a crystal derived stable clock. The Clock Source Control Register bits determine which clock is connected to which domain.

By studying the following drawings, and appropriately setting the Data Flow Control and Clock Source Control register bits, the CS8420 can be configured to fit a variety of customer requirements.

The following drawings illustrate the possible valid data flows. The audio data flow is indicated by the thin lines; the clock routing is indicated by the bold lines. The register settings for the Data Flow Control register and the Clock Source Register are also shown for each data flow. Some of the register settings may appear to be not relevant to the particular data flow in question, but have been assigned a particular state. This is done to minimize power consumption. The AESBP data path from the RXP pin to the AES3 output drivers, and the TXOFF control, have been omitted for clarity, but are present and functional in all modes where the AES3 transmitter is in use.

Figures 7 and 8 show audio data entering via the serial audio input port, then passing through the sample rate converter, and then output both to the serial audio output port and to the AES3 transmitter. Figure 7 shows the PLL recovering the input clock from ILRCK word clock. Figure 8 shows using a

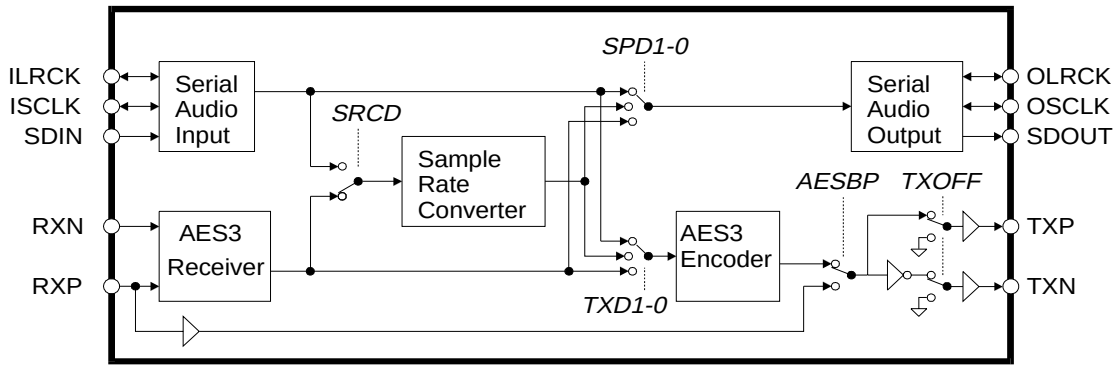


Figure 6. Software Mode Audio Data Flow Switching Options

direct 256*Fsi clock input via the RMCK pin, instead of the PLL.

Figure 9 shows audio data entering via the AES3 Receiver. The PLL locks onto the pre-ambles in the incoming audio stream, and generates a 256*Fsi clock. The rate converted data is then output via the serial audio output port and via the AES3 transmitter.

Figure 10 shows the same data flow as Figure 7. The input clock is derived from an incoming AES3 data stream. The incoming data must be synchronous to the AES3 data stream.

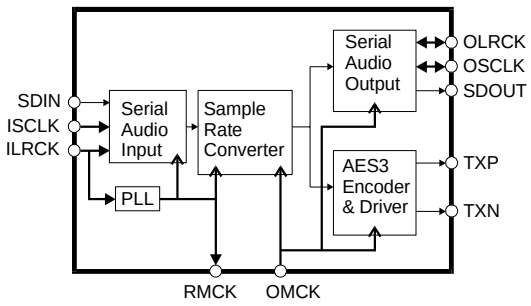
Figure 11 shows the same data flow as Figure 7. The input data must be synchronous to OMCK. The output data is clocked by the recovered PLL clock from an AES3 input stream. This may be used to implement a “house sync” architecture.

Figure 8 shows audio data entering via the AES3 receiver, passing through the sample rate converter, and then exiting via the serial audio output port. Synchronous audio data may then be input via the serial audio input port and output via the AES3 transmitter.

Figure 13 is the same as Figure 12, but without the sample rate converter. The whole data path is clocked via the PLL generated recovered clock.

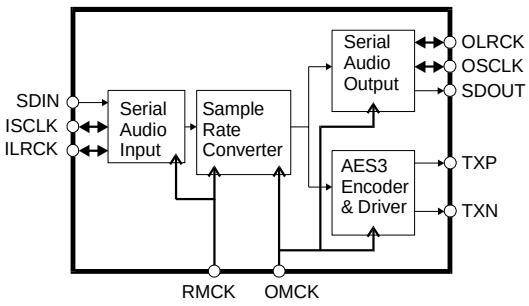
Figure 14 illustrates a standard AES3 receiver function, with no rate conversion.

Figure 15 shows a standard AES3 transmitter function, with no rate conversion.



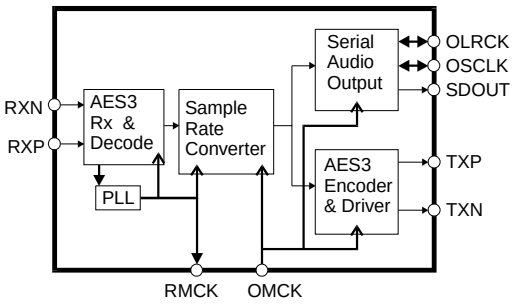
Data Flow Control Bits		Clock Source Control Bits	
TXD1-0:	00	*OUTC:*	0
SPD1-0:	00	*INC:*	0
SRCD:	0	*RXD1-0:*	00

Figure 7. Serial Audio Input, using PLL, SRC enabled



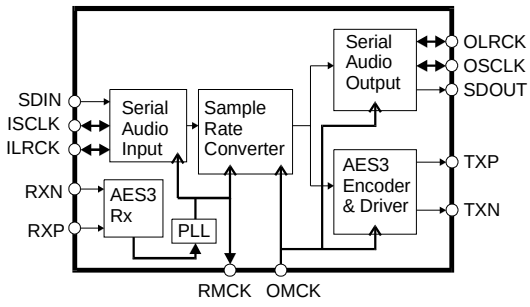
Data Flow Control Bits		Clock Source Control Bits	
TXD1-0:	00	*OUTC:*	0
SPD1-0:	00	*INC:*	0
SRCD:	0	*RXD1-0:*	10

Figure 8. Serial Audio Input, No PLL, SRC enabled



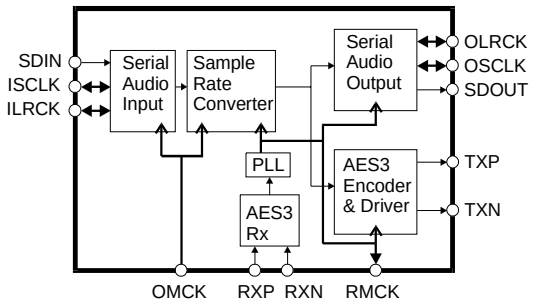
Data Flow Control Bits		Clock Source Control Bits	
TXD1-0:	00	*OUTC:*	0
SPD1-0:	00	*INC:*	0
SRCD:	1	*RXD1-0:*	01

Figure 9. AES3 Input, SRC enabled



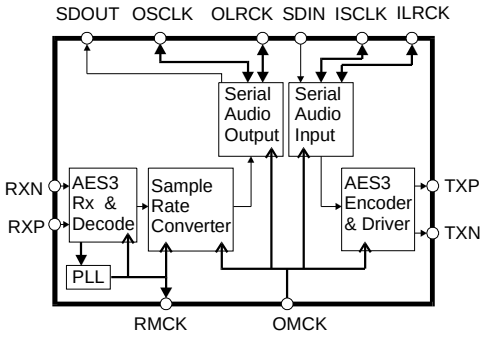
Data Flow Control Bits		Clock Source Control Bits	
TXD1-0:	00	*OUTC:*	0
SPD1-0:	00	*INC:*	0
SRCD:	0	*RXD1-0:*	01

Figure 10. Serial Audio Input, AES3 Input Clock



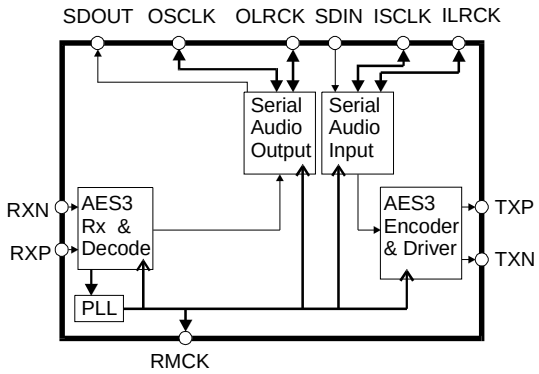
Data Flow Control Bits		Clock Source Control Bits	
TXD1-0:	00	*OUTC:*	1
SPD1-0:	00	*INC:*	1
SRCD:	0	*RXD1-0:*	01

Figure 11. Serial Audio Input, SRC Output clocked by AES3 Recovered Clock



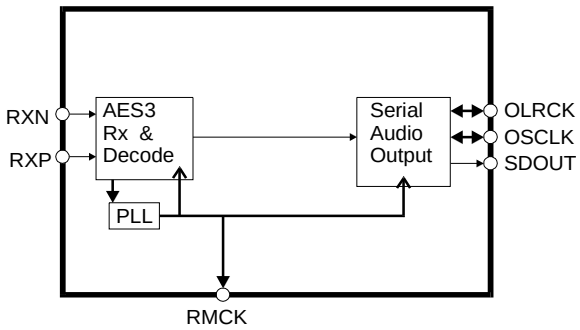
Data Flow Control Bits		Clock Source Control Bits	
TXD1-0:	01	*OUTC:*	0
SPD1-0:	00	*INC:*	0
SRCD:	1	*RXD1-0:*	01

Figure 12. AES3 Input, SRC to Serial Audio Output, Serial Audio Input to AES3 Out



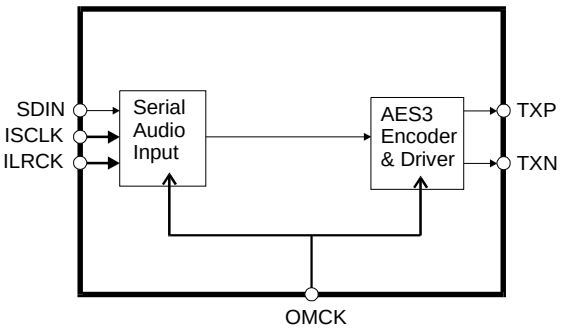
Data Flow Control Bits		Clock Source Control Bits	
TXD1-0:	01	*OUTC:*	1
SPD1-0:	10	*INC:*	0
SRCD:	0	*RXD1-0:*	01

Figure 13. AES3 Input to Serial Audio Output, Serial Audio Input to AES3 Out, no SRC



Data Flow Control Bits		Clock Source Control Bits	
TXD1-0:	10	*OUTC:*	1
SPD1-0:	10	*INC:*	0
SRCD:	0	*RXD1-0:*	01
TXOFF:	1		

Figure 14. AES3 Input to Serial Audio Output Only



Data Flow Control Bits		Clock Source Control Bits	
TXD1-0:	01	*OUTC:*	0
SPD1-0:	01	*INC:*	1
SRCD:	0	*RXD1-0:*	00

Figure 15. Input Serial Port to AES3 Transmitter

5. SAMPLE RATE CONVERTER (SRC)

Multirate digital signal processing techniques are used to conceptually upsample the incoming data to very high rate and then downsample to the outgoing rate, resulting in a 24 bit output, regardless of the width of the input. The filtering is designed so that a full input audio bandwidth of 20 kHz is preserved if the input sample and output sample rates are greater than 44.1 kHz. When the output sample rate becomes less than the input sample rate, the input is automatically bandlimited to avoid aliasing products in the output. Careful design ensures minimum ripple and distortion products are added to the incoming signal. The SRC also determines the ratio between the incoming and outgoing sample rates, and sets the filter corner frequencies appropriately. Any jitter in the incoming signal has little impact on the dynamic performance of the rate converter, and has no influence on the output clock.

5.1 Dither

When using the AES3 input, and when using the serial audio input port in left justified and I^2S modes, all input data is treated as 24-bits wide. Any truncation that has been done prior to the CS8420 to less than 24-bits should have been done using an appropriate dither process. If the serial audio input port is used to feed the SRC, and the port is in right justified mode, then the input data will be truncated to the SIRES bit setting value. If SIRES bits are set to 16 or 20-bits, and the input data is 24-bits wide, then truncation distortion will occur. Similarly, in any serial audio input port mode, if an inadequate number of bit clocks are entered (say 16 instead of 20), then the input words will be truncated, causing truncation distortion at low levels. In summary, there is no dithering mechanism on the input side of the CS8420, and care must be taken to ensure that no truncation occurs.

Dithering is used internally where appropriate inside the SRC block.

The output side of the SRC can be set to 16, 20 or 24 bits. Optional dithering can be applied, and is automatically scaled to the selected output word length. This dither is not correlated between left and right channels. It is recommended that the dither control bit be left in its default on state.

5.2 SRC Locking, Varispeed and the Sample Rate Ratio Register

The SRC calculates the ratio between the input sample rate and the output sample rate, and uses this information to set up various parameters inside the SRC block. The SRC takes some time to make this calculation. For a worst case 3:1 to 1:3 input sample rate transition, the SRC will take 9400/Fso to settle (195 ms at Fso of 48 kHz). For a power-up situation, the SRC will start from 1:1, the worst case time becomes 8300/Fso (172 ms at Fso of 48 kHz).

If the PLL is in use (either AES3 or serial input port), then the worst case locking time for the PLL and the SRC is the sum of each locking time.

If Fsi is changing, for example in a varispeed application, the REUNLOCK interrupt will occur, and the SRC will track the incoming sample rate. During this tracking mode, the SRC will still rate convert the audio data, but at increased distortion levels. Once the incoming sample rate is stable, then the REUNLOCK interrupt will become false, and the SRC will return to normal levels of audio quality.

The VFIFO interrupt occurs if the data buffer in the SRC overflows, which can occur if the input sample rate changes at >10%/second.

Varispeed at Fsi slew rates approaching 10%/sec is only supported when the input is via the serial audio input port. When using the AES3 input, high frame rate slew rates will cause the PLL to lose lock.

The sample rate ratio is also made available as a register, accessible via the control port. The upper

2 bits of this register form the integer part of the ratio, while the lower 6 bits form the fractional part. Since, in many instances, Fso is known, this allows the calculation of the incoming sample rate by the host microcontroller.

6. THREE-WIRE SERIAL AUDIO PORTS

A 3-wire serial audio input port and a 3-wire serial audio output port is provided. Each port can be adjusted to suit the attached device via control registers. The following parameters are adjustable: master or slave, serial clock frequency, audio data resolution, left or right justification of the data relative to left/right clock, optional 1 bit cell delay of the 1st data bit, the polarity of the bit clock and the polarity of the left/right clock. By setting the appropriate control bits, many formats are possible.

Figure 16 shows a selection of common input formats, along with the control bit settings. The clocking of the input section of the CS8420 may be derived from the incoming ILRCK word rate clock, using the on-chip PLL. The PLL operation is described in the AES receiver description on page 19. In the case of use with the serial audio input port, the PLL locks onto the leading edges of the ILRCK clock.

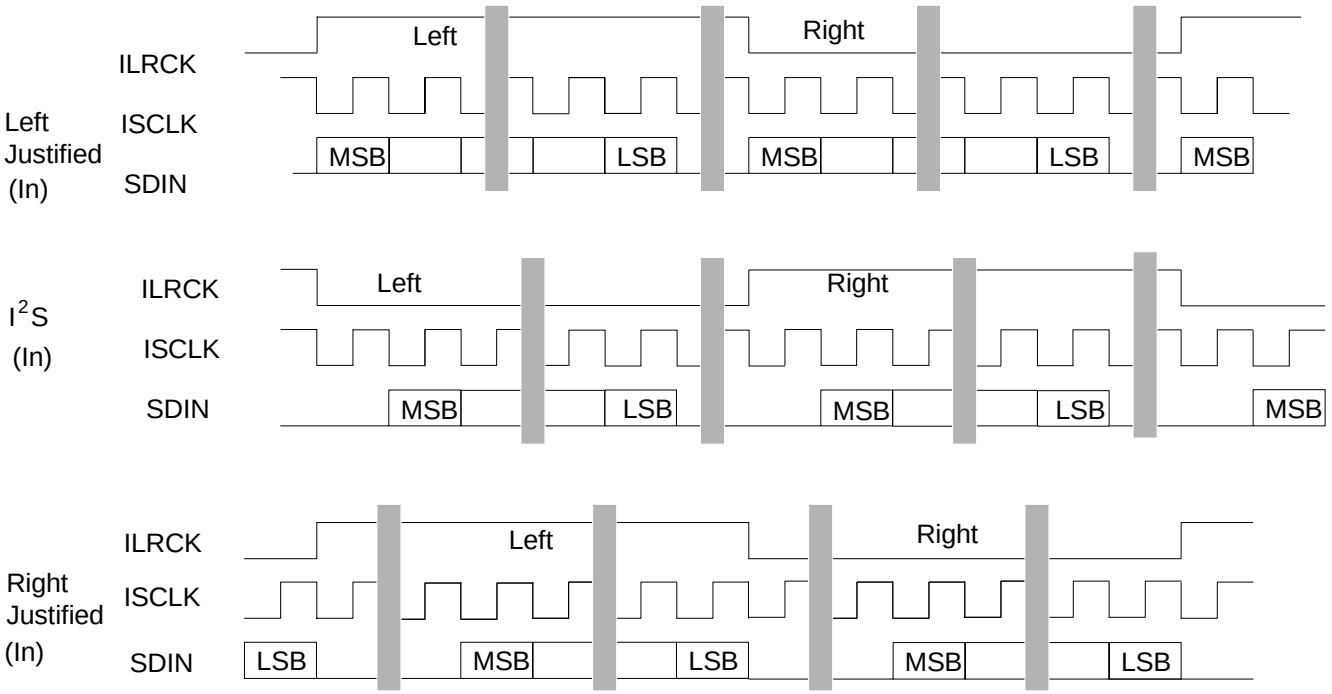
Figure 17 shows a selection of common output formats, along with the control bit settings. A special AES3 direct output format is included, which allows serial output port access to the V, U, and C bits embedded in the serial audio data stream. The P bit is replaced by a bit indicating the location of the start of a block. This format is only available when the serial audio output port is being clocked by the AES3 receiver recovered clock. Also, the received channel status block start signal is only available in hardware mode 5, as the RCBL pin.

In master mode, the left/right clock and the serial bit clock are outputs, derived from the appropriate clock domain master clock.

In slave mode, the left/right clock and the serial bit clock are inputs. The left/right clock must be synchronous to the appropriate master clock, but the serial bit clock can be asynchronous and discontinuous if required. By appropriate phasing of the left/right clock and control of the serial clocks, multiple CS8420's can share one serial port. The left/right clock should be continuous, but the duty cycle does not have to be 50%, provided that enough serial clocks are present in each phase to clock all the data bits. When in slave mode, the serial audio output port must be set to left justified or I^2S data.

When using the serial audio output port in slave mode with an OLRCK input which is asynchronous to the port's data source, then an interrupt bit is provided to indicate when repeated or dropped samples occur.

The CS8420 allows immediate mute of the serial audio output port audio data via a control register bit.



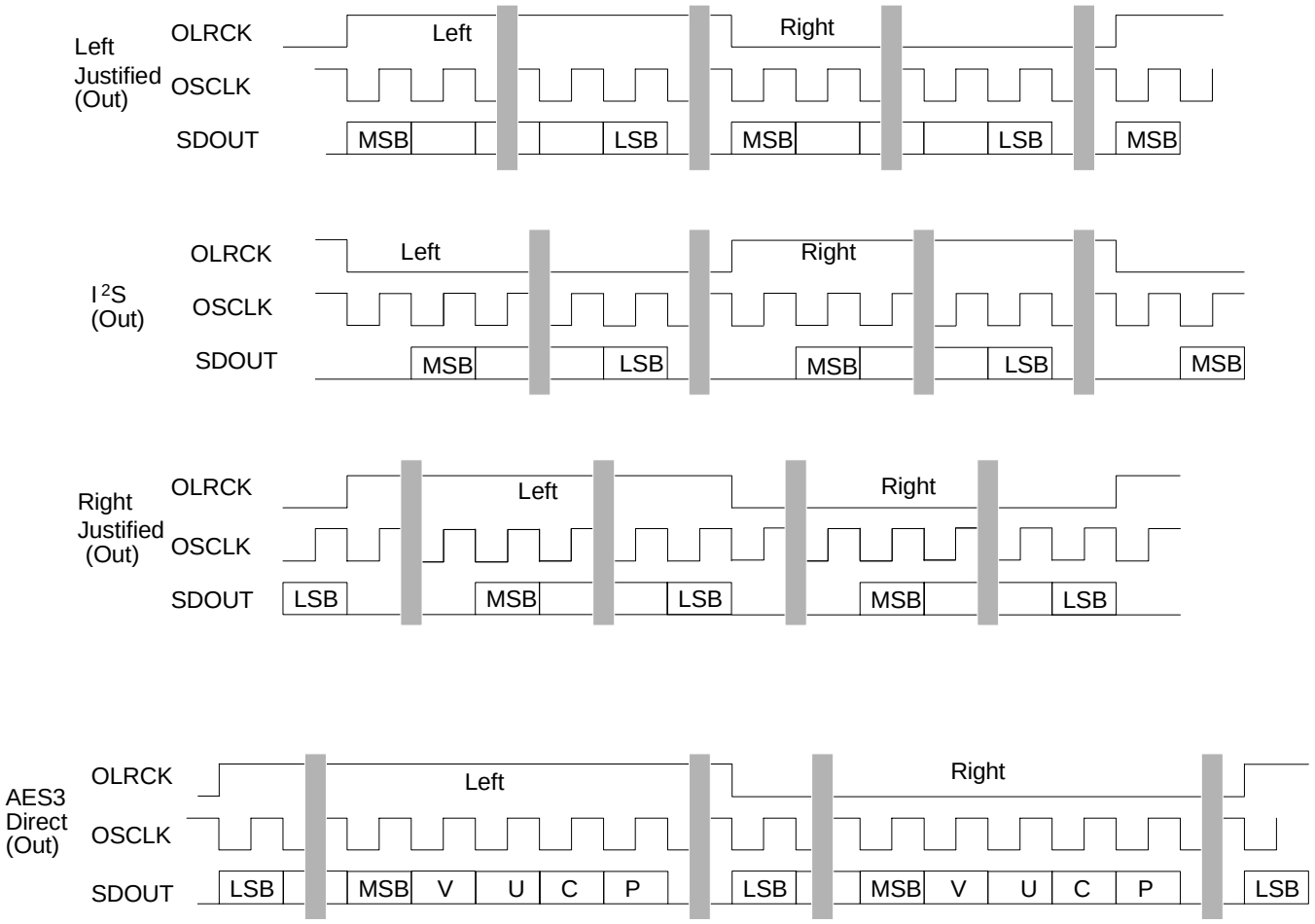
	SIMS	SISF	SIRES1/0	SIJUST	SIDEL	SISPOL	SILRPOL
Left Justified	X	X	00	0	0	0	0
I^2S	X	X	00+	0	1	0	1
Right Justified	X	X	XX*	1	0	0	0

X = don't care to match format, but does need to be set to the desired setting

+ I^2S can accept an arbitrary number of bits, determined by the number of ISCLK cycles

* not 11 - See Serial Input Port Data Format Register Bit Descriptions for an explanation of the meaning of each bit

Figure 16. Serial Audio Input Example Formats



	SOMS	SOSF	SORES1/0	SOJUST	SODEL	SOSPOL	SOLRPOL
Left Justified	X	X	XX*	0	0	0	0
I^2S	X	X	XX*	0	1	0	1
Right Justified	1	X	XX*	1	0	0	0
AES3 Direct	X	X	11	0	0	0	0

X = don't care to match format, but does need to be set to the desired setting

* not 11 - See Serial Output Data Format Register Bit Descriptions for an explanation of the meaning of each bit

Figure 17. Serial Audio Output Example Formats

7. AES3 TRANSMITTER AND RECEIVER

The CS8420 includes an AES3 type digital audio receiver and an AES3 type digital audio transmitter. A comprehensive buffering scheme provides read/write access to the channel status and user data. This buffering scheme is described in the Appendix: Channel Status and User Data Buffer Management on page 72.

7.1 AES3 Receiver

The AES3 receiver accepts and decodes audio and digital data according to the AES3, IEC60958 (S/PDIF), and EIAJ CP-1201 interface standards. The receiver consists of a differential input stage, accessed via pins RXP and RXN, a PLL based clock recovery circuit, and a decoder which separates the audio data from the channel status and user data.

External components are used to terminate and isolate the incoming data cables from the CS8420. These components are detailed in the Appendix "External AES/SPDIF/IEC60958 Transmitter and Receiver Components" on page 70.

7.1.1 PLL, Jitter Attenuation, and Varispeed

An on-chip Phase Locked Loop (PLL) is used to recover the clock from the incoming data stream. Although the on-chip sample rate converter is immune to large amounts of jitter, there are some applications where low jitter in the recovered clock, presented on the RMCK pin, is important. For this reason, the PLL has been designed to have good jitter attenuation characteristics, shown in Figures 18, 19 & 20. In addition, the PLL has been designed to only use the preambles of the AES3 stream to provide lock update information to the PLL. This results in the PLL being immune to data dependent jitter affects, since the AES3 preambles do not vary with the data. The PLL has the ability to lock onto a wide range of input sample rates, with no external component changes. If the sample rate of the input subsequently changes, for example in a varispeed application, then the PLL will only track up to ±12.5% from the nominal center sample rate. The nominal center sample rate is the sample rate that the PLL first locks onto upon application of an AES3 data stream, or after enabling the CS8420 clocks by setting the RUN control bit. If the 12.5% sample rate limit is exceeded, the PLL will return to its wide lock range mode, and re-acquire a new nominal center sample rate.

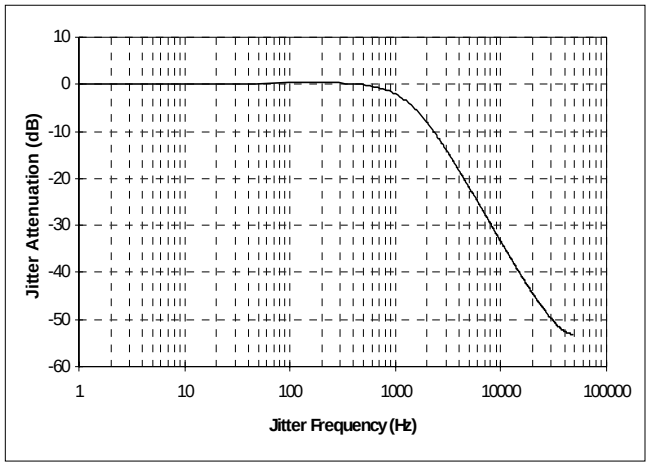


Figure 18. Jitter Attenuation Characteristics of PLL with "slow" Filter Components

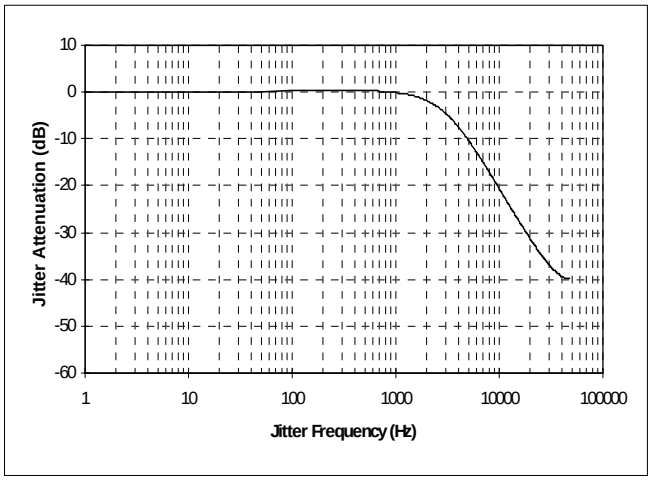


Figure 19. Jitter Attenuation Characteristics of PLL with "medium" Filter Components

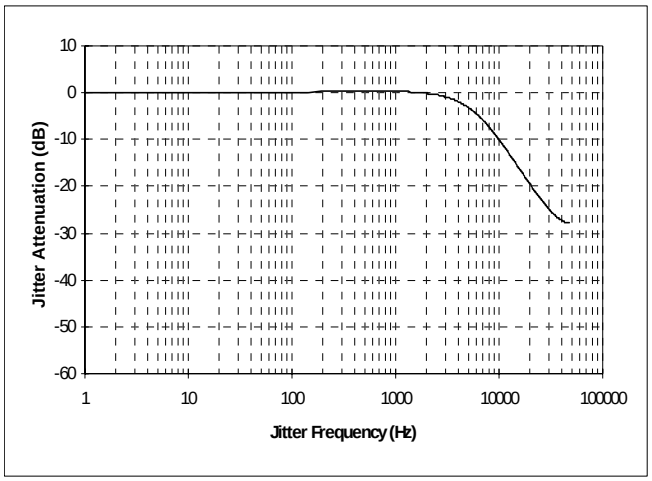


Figure 20. Jitter Attenuation Characteristics of PLL with "fast" Filter Components

8. OMCK OUT ON RMCK

A special mode is available that allows the clock that is being input through the OMCK pin to be output through the RMCK pin. This feature is controlled by the SWCLK bit in register 4 of the control registers. When the PLL loses lock the frequency of the VCO drops to 300 kHz. The SWCLK function allows the clock from RMCK to be used as a clock in the system without any disruption when input is removed from the Receiver.

9. PLL EXTERNAL COMPONENTS

The PLL behavior is affected by the external filter component values. Figure 5 shows the configuration of the required 2 capacitors and 1 resistor. Two alternate sets of component values are recommended, depending on the requirements of the application (see Table 1). The default set, called "fast", accommodates input sample rates of 16 kHz to 108 Hz with no component changes. It has the highest corner frequency jitter attenuation curve, and takes the shortest time to lock. The alternate component set, called "medium" allows the lowest input sample rate to be 8 kHz, and increases the lock time of the PLL. Lock times are worst case for an Fsi transition of 96 kHz.

9.1 Error Reporting and Hold Function

While decoding the incoming AES3 data stream, the CS8420 can identify several kinds of error, indicated in the Receiver Error register. The UNLOCK bit indicates whether the PLL is locked to the incoming AES3 data. The V bit reflects the current validity bit status. The CONF (confidence) bit indicates the amplitude of the eye pattern opening, indicating a link that is close to generating errors. The BIP (bi-phase) error bit indicates an error in incoming bi-phase coding. The PAR (parity) bit indicates a received parity error.

The error bits are "sticky": they are set on the first occurrence of the associated error, and will remain set until the user reads the register via the control port. This enables the register to log all unmasked errors that occurred since the last time the register was read.

The Receiver Error Mask register allows masking of individual errors. The bits in this register serve as masks for the corresponding bits of the Receiver Error Register. If a mask bit is set to 1, the error is considered unmasked, meaning that its occurrence will be reported in the receiver error register, will affect the RERR pin, will invoke the occurrence of a RERR interrupt, and will affect the current audio sample according to the status of the HOLD bits. The HOLD bits allow a choice of holding the previous sample, replacing the current sample with zero (mute), or do not change the current audio sample. If a mask bit is set to 0, the error is considered masked, meaning that its occurrence will not be reported in the receiver error register, will not induce a pulse on RERR or generate a RERR interrupt, and will not affect the current audio sample. The QCRC and CCRC errors do not affect the current audio sample, even if unmasked.

9.2 Channel Status Data Handling

The first 2 bytes of the Channel Status block are decoded into the Receiver Channel Status register. The setting of the CHS bit in the Channel Status Data Buffer Control register determines whether

Type	RFILT (kΩ)	CFILT (µF)	CRIP (nF)	Fsi Range (kHz)	PLL Lock Time (ms)
Medium	0.909	1.8	33	8 to 96	56
Fast	1.78	0.47	8.2	16 to 108	15

Table 1. PLL External Component Values

the channel status decodes are from the A channel (CHS = 0) or B channel (CHS = 1).

The PRO (professional) bit is extracted directly. Also, for consumer data, the COPY (copyright) bit is extracted, and the category code and L bits are decoded to determine SCMS status, indicated by the ORIG (original) bit. Finally, the $\overline{\text{AUDIO}}$ bit is extracted, and used to set an $\overline{\text{AUDIO}}$ indicator, as described in the Non-Audio Auto Detection section below.

If 50/15 µs pre-emphasis is detected, then this is reflected in the state of the $\overline{\text{EMPH}}$ pin.

The encoded sample word length channel status bits are decoded according to AES3-1992 or IEC 60958. If the AES3 receiver is the data source for the SRC, then the SRC audio input data is truncated according to the channel status word length settings. Audio data routed to the serial audio output port is unaffected by the word length settings; all 24 bits are passed on as received.

The Appendix: Channel Status and User Data Buffer Management (page 72) describes the overall handling of CS and U data.

9.3 User Data Handling

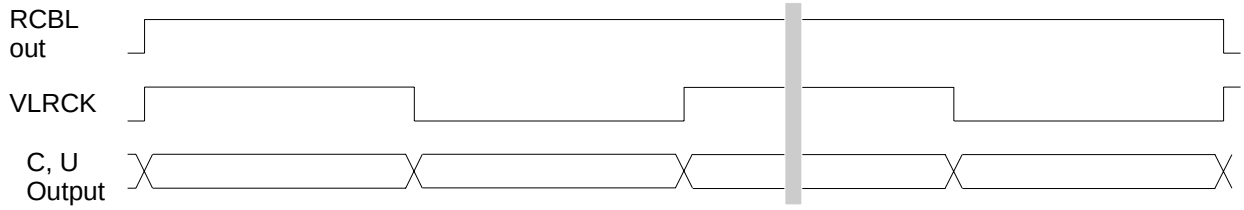
The incoming user data is buffered in a user accessible buffer. Various automatic modes of re-transmitting received U data are provided. The Appendix: Channel Status and User Data Buffer Management (page 72) describes the overall handling of CS and U data.

Received U data may also be output to the U pin, under the control of a control register bit. Depending on the data flow and clocking options selected, there may not be a clock available to qualify the U data output. Figure 21 illustrates the timing.

If the incoming user data bits have been encoded as Q-channel subcode, then the data is decoded and presented in 10 consecutive register locations. An interrupt may be enabled to indicate the decoding of a new Q-channel block, which may be read via the control port.

9.4 Non-Audio Auto Detection

Since it is possible to convey non-audio data in an AES3 data stream, it is important to know whether the incoming AES3 data stream is digital audio or not. This information is typically conveyed in channel status bit 1 ($\overline{\text{AUDIO}}$), which is extracted



RCBL and C output are only available in hardware mode 5.
RCBL goes high 2 frames after receipt of a Z pre-amble, and is high for 16 frames.
VLRCK is a virtual word clock, which may not exist, but is used to illustrate the CU timing.
VLRCK duty cycle is 50%. VLRCK frequency is always equal to the incoming frame rate.
If no SRC is used, and the serial audio output port is in master mode, VLRCK = OLRCK.
If the serial audio output port is in slave mode, then VLRCK needs to be externally created, if required.
C, U transitions are aligned within ±1% of VLRCK period to VLRCK edges

Figure 21. AES3 Receiver Timing for C & U pin output data

automatically by the CS8420. However, certain non-audio sources, such as AC3 or MPEG encoders, may not adhere to this convention, and the bit may not be properly set. The CS8420 AES3 receiver can detect such non-audio data. This is accomplished by looking for a 96-bit sync code, consisting of 0x0000, 0x0000, 0x0000, 0x0000, 0xF872, and 0x4E1F. When the sync code is detected, an internal AUTODETECT signal will be asserted. If no additional sync codes are detected within the next 4096 frames, AUTODETECT will be de-asserted until another sync code is detected. The $\overline{\text{AUDIO}}$ bit in the Receiver Channel Status register is the logical OR of AUTODETECT and the received channel status bit 1. If non-audio data is detected, the data is still processed exactly as if it were normal audio. It is up to the user to mute the outputs as required.

9.5 AES3 Transmitter

The AES3 transmitter encodes and transmits audio and digital data according to the AES3, IEC60958 (S/PDIF), and EIAJ CP-1201 interface standards. Audio and control data are multiplexed together and bi-phase mark encoded. The resulting bit stream is then driven directly, or through a transformer, to an output connector.

The transmitter is usually clocked from the output side clock domain of the sample rate converter. This clock may be derived from the clock input pin OMCK, or from the incoming data. In data flows with no SRC, and where OMCK is asynchronous to the data source, an interrupt bit is provided that will go high every time a data sample is dropped or repeated.

The channel status (C) and user channel (U) bits in the transmitted data stream are taken from storage areas within the CS8420. The user can manipulate the contents of the internal storage with a microcontroller. The CS8420 will also run in one of several automatic modes. The Appendix: Channel Status and User Data Buffer Management (page 72) provides detailed descriptions of each automatic mode, and describes methods for accessing the storage areas. The transmitted user data can optionally be input via the U pin, under the control of a control port register bit. Figure 22 shows the timing requirements for inputting U data via the U pin.

9.5.1 Transmitted Frame and Channel Status Boundary Timing

The TCBL pin may be an input or an output, and is used to control or indicate the start of transmitted channel status block boundaries.

In some applications, it may be necessary to control the precise timing of the transmitted AES3 frame boundaries. This may be achieved in 3 ways:

a) With TCBL configured as an input, when TCBL transitions high for >3 OMCK clocks, it will cause a frame start, and a new channel status block start.

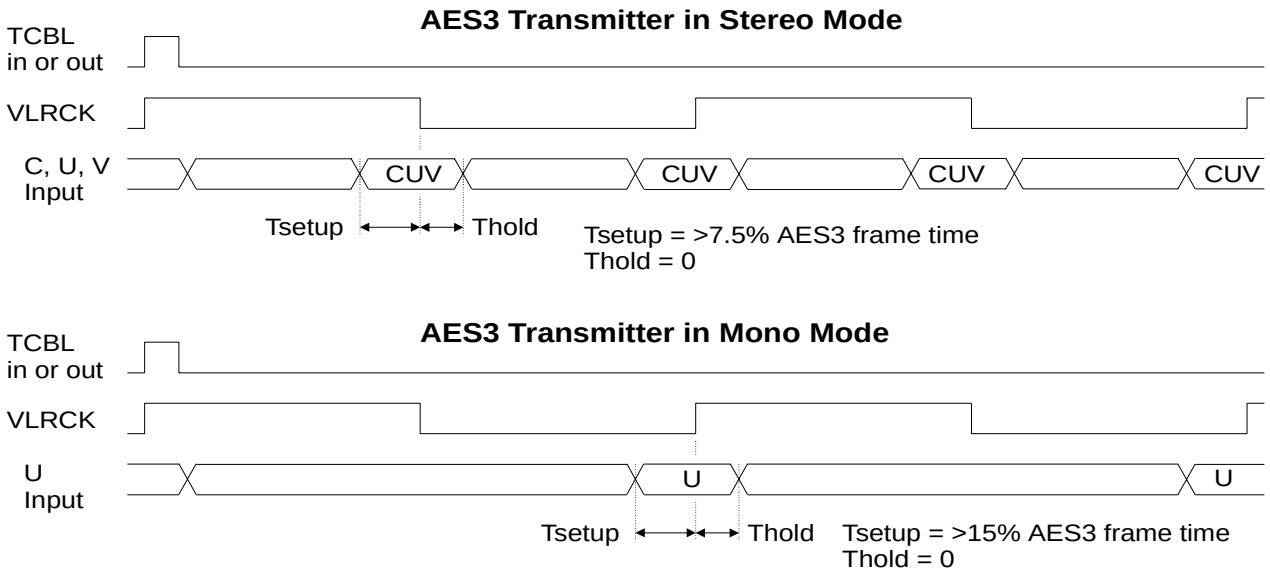
b) If the AES3 output comes from the AES3 input, while there is no SRC, setting TCBL as output will cause AES3 output frame boundaries to align with AES3 input frame boundaries.

c) If the AES3 output comes from the serial audio input port while the port is in slave mode, and TCBL is set to output, then the start of the A channel sub-frame will be aligned with the leading edge of ILRCK.

9.5.2 TXN and TXP Drivers

The line drivers are low skew, low impedance, differential outputs capable of driving cables directly. Both drivers are set to ground during reset ($\overline{\text{RST}}$ = low), when no AES3 transmit clock is provided, and optionally under the control of a register bit. The CS8420 also allows immediate mute of the AES3 transmitter audio data via a control register bit.

External components are used to terminate and isolate the external cable from the CS8420. These components are detailed in the Appendix “External



VLRCK is a virtual word clock, which may not exist, but is used to illustrate the CUV timing.
VLRCK duty cycle is 50%.
In stereo mode, VLRCK = AES3 frame rate. In mono mode, VLRCK = 2*AES3 frame rate
If the serial audio output port is in master mode, and TCBL is an output, and the SRC is not in use, then VLRCK = OLRCK.
If the serial audio input port is in master mode, and TCBL is an input, and the SRC is not between the serial audio input port and the AES3 transmitter, then VLRCK = ILRCK.
Otherwise, VLRCK needs to be externally created, if required

Figure 22. AES3 Transmitter Timing for C, U and V pin input data

AES/SPDIF/IEC60958 Transmitter and Receiver Components" on page 70.

9.6 Mono Mode Operation

Currently, the AES3 standard is being updated to include options for 96 kHz sample rate operation. One method is to double the frame rate of the current format. This results in a 96 kHz sample rate, stereo signal carried over a single twisted pair cable. An alternate method is where the 2 sub-frames in a 48 kHz frame rate AES3 signal are used to carry consecutive samples of a mono signal, resulting in a 96 kHz sample rate stream. This allows older equipment, whose AES3 transmitters and receivers are not rated for 96 kHz frame rate operation, to handle 96 kHz sample rate information. In this "mono mode", 2 AES3 cables are needed for stereo data transfer. The CS8420 offers mono mode operation, both for the AES3 receiver and for the AES3 transmitter. Figure 23 shows the operation of mono mode in comparison with normal stereo mode. The receiver and transmitter sections may be independently set to mono mode via the MMR and MMT control bits.

The receiver mono mode effectively doubles Fsi compared to the input frame rate. The clock output on the RMCK pin tracks Fsi, and so is doubled in frequency compared to stereo mode. In mono mode, A and B sub-frames are routed to the SRC inputs as consecutive samples.

When the transmitter is in mono mode, either A or B SRC consecutive outputs are routed alternately to A and B sub-frames in the AES3 output stream. Which channel status block is transmitted is also selectable.

For the AES3 input to serial audio port output data flow, in receiver mono mode, then the receiver will

run at a frame rate of Fsi/2, and the serial audio output port will run at Fsi. Identical data will appear in both left and right data fields on the SDOUT pin.

For the serial audio input port to AES3 transmitter data flow, in transmitter mono mode, then the input port will run at Fso audio sample rate, while the AES3 transmitter frame rate will be at Fso/2. The data from either consecutive left, or right, positions will be selected for transmitting in A and B sub-frames.

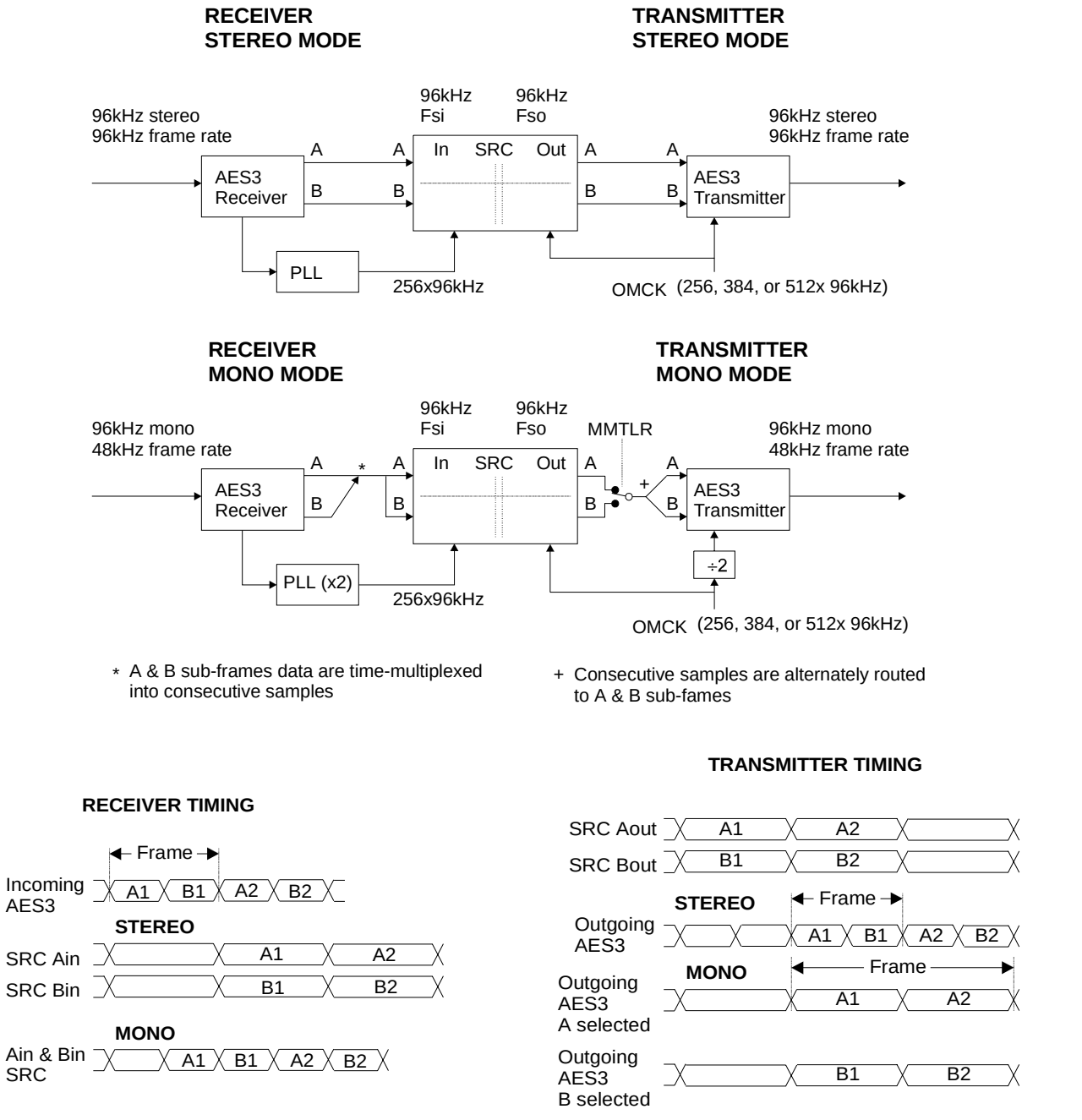


Figure 23. Mono Mode Operation Compared to Normal Stereo Operation

10. CONTROL PORT DESCRIPTION AND TIMING

The control port is used to access the registers, allowing the CS8420 to be configured for the desired operational modes and formats. In addition, Channel Status and User data may be read and written via the control port. The operation of the control port may be completely asynchronous with respect to the audio sample rates. However, to avoid potential interference problems, the control port pins should remain static if no operation is required.

The control port has 2 modes: SPI and I^2C, with the CS8420 acting as a slave device. SPI mode is selected if there is a high to low transition on the AD0/$\overline{CS}$ pin, after the $\overline{RST}$ pin has been brought high. I^2C mode is selected by connecting the AD0/$\overline{CS}$ pin to VD+ or DGND, thereby permanently selecting the desired AD0 bit address state.

10.1 SPI Mode

In SPI mode, $\overline{CS}$ is the CS8420 chip select signal, CCLK is the control port bit clock (input into the CS8420 from the microcontroller), CDIN is the input data line from the microcontroller, CDOUT is the output data line to the microcontroller. Data is clocked in on the rising edge of CCLK and out on the falling edge.

Figure 24 shows the operation of the control port in SPI mode. To write to a register, bring $\overline{CS}$ low. The first 7 bits on CDIN form the chip address and must be 0010000. The eighth bit is a read/write indicator (R/$\overline{W}$), which should be low to write. The next 8 bits form the Memory Address Pointer (MAP), which is set to the address of the register that is to be updated. The next 8 bits are the data which will be placed into the register designated by the MAP. During writes, the CDOUT output stays in the Hi-Z state. It may be externally pulled high or low with a 47 kΩ resistor, if desired.

There is a MAP auto increment capability, enabled by the INCR bit in the MAP register. If INCR is a zero, then the MAP will stay constant for successive read or writes. If INCR is set to a 1, then the MAP will auto-increment after each byte is read or written, allowing block reads or writes of successive registers.

To read a register, the MAP has to be set to the correct address by executing a partial write cycle which finishes ($\overline{CS}$ high) immediately after the MAP byte. The MAP auto increment bit (INCR)

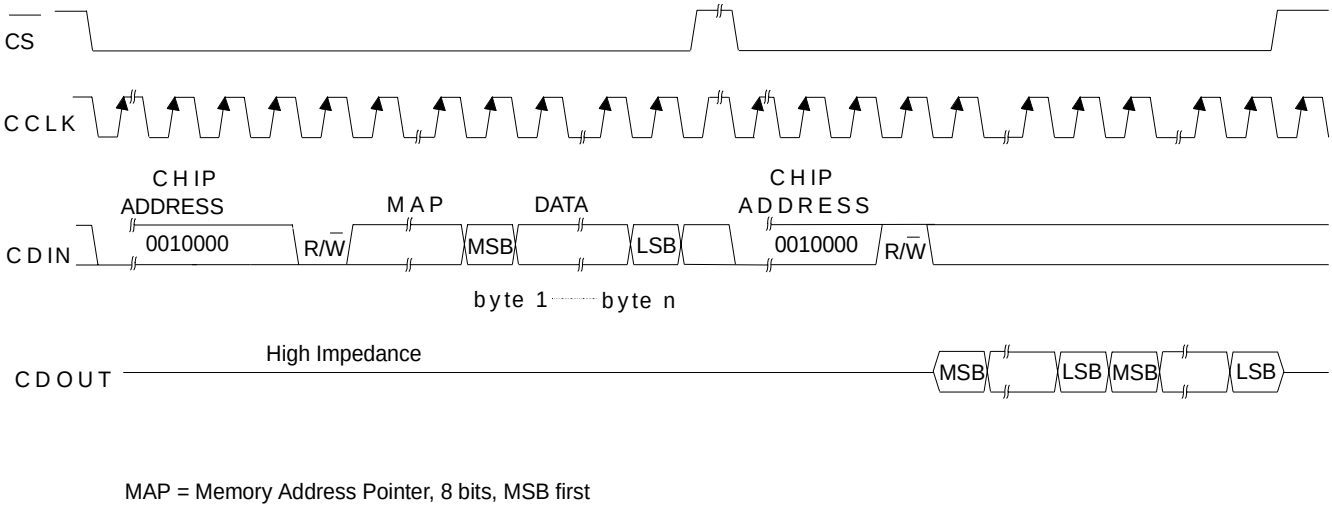


Figure 24. Control Port Timing in SPI Mode

may be set or not, as desired. To begin a read, bring $\overline{CS}$ low, send out the chip address and set the read/write bit (R/$\overline{W}$) high. The next falling edge of CCLK will clock out the MSB of the addressed register (CDOUT will leave the high impedance state). If the MAP auto increment bit is set to 1, the data for successive registers will appear consecutively.

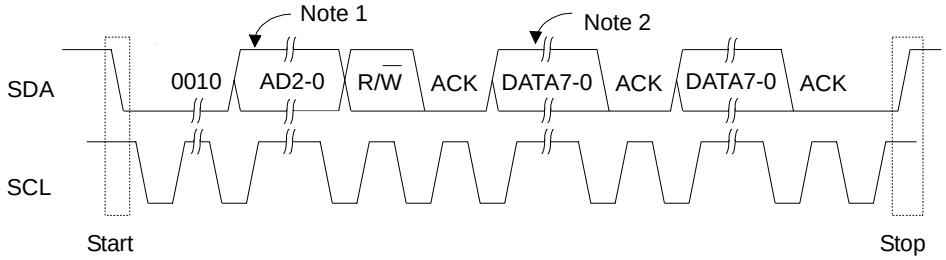
10.2 I²C Mode

In I²C mode, SDA is a bidirectional data line. Data is clocked into and out of the part by the clock, SCL, with the clock to data relationship as shown in Figure 25. There is no $\overline{CS}$ pin. Each individual CS8420 is given a unique address. Pins AD0, AD1 form the 2 least significant bits of the chip address, and should be connected to VD+ or DGND as desired. The $\overline{EMPH}$ pin is used to set the AD2 bit, by connecting a resistor from the $\overline{EMPH}$ pin to VD+ or to DGND. The state of the pin is sensed while the CS8420 is being reset. The upper 4 bits of the 7-bit address field are fixed at 0010. To communicate with a CS8420, the chip address field, which is the first byte sent to the CS8420, should match 0010 followed by the settings of the $\overline{EMPH}$, AD1, and AD0. The eighth bit of the address is the R/$\overline{W}$ bit. If the operation is a write, the next byte is the Memory Address Pointer (MAP) which selects the register to be read or written. If the operation is a read, the contents of the register pointed to by the MAP will be output. Setting the auto increment bit in MAP allows successive reads or writes of consecutive registers. Each byte is separated by an acknowledge bit. The ACK bit is output from the CS8420 after each input byte is read, and is input to the CS8420 from the microcontroller after each transmitted byte. I²C is a registered trademark of Philips Semiconductors.

10.3 Interrupts

The CS8420 has a comprehensive interrupt capability. The INT output pin is intended to drive the interrupt input pin on the host microcontroller. The INT pin may be set to be active low, active high or active low with no active pull-up transistor. This last mode is used for active low, wired-OR hookups, with multiple peripherals connected to the microcontroller interrupt input pin.

Many conditions can cause an interrupt, as listed in the interrupt status register descriptions. Each source may be masked off via mask registers. In addition, each source may be set to rising edge, falling edge or level sensitive. Combined with the option of level sensitive or edge sensitive modes within the microcontroller, many different set-ups are possible, depending on the needs of the equipment designer.



Note 1: AD2 is derived from a resistor attached to the $\overline{EMPH}$ pin, AD1 and AD0 are determined by the state of the corresponding pins

Note 2: If operation is a write, this byte contains the Memory Address Pointer, MAP

Figure 25. Control Port Timing in I²C Mode

11. CONTROL PORT REGISTER BIT DEFINITIONS

11.1 Memory Address Pointer (MAP)

7	6	5	4	3	2	1	0
INCR	MAP6	MAP5	MAP4	MAP3	MAP2	MAP1	MAP0

This register defaults to 01

INCR — Auto Increment Address Control Bit
- 0 - Auto increment address off
- 1 - Auto increment address on

MAP6-MAP0 — Register address and function list
- 0 - Reserved
- 1 - Misc. Control 1
- 2 - Misc. Control 2
- 3 - Data Flow Control
- 4 - Clock Source Control
- 5 - Serial Audio Input Port Data Format
- 6 - Serial Audio Output Port Data Format
- 7 - Interrupt Register 1 Status
- 8 - Interrupt Register 2 Status
- 9 - Interrupt Register 1 Mask
- 10 - Interrupt Register1 Mode (MSB)
- 11 - Interrupt Register 1 Mode (LSB)
- 12 - Interrupt Register 2 Mask
- 13 - Interrupt Register 2 Mode (MSB)
- 14 - Interrupt Register 2 Mode (LSB)
- 15 - Receiver Channel Status Bits
- 16 - Receiver Error Status
- 17 - Receiver Error Mask
- 18 - Channel Status Data Buffer Control
- 19 - User Data Buffer Control
- 20 to 29 - Q-channel Subcode Bytes 0 to 9
- 30 - Sample Rate Ratio
- 31 - Reserved
- 32 to 55 - C-bit or U-bit Data Buffer
- 56 to 126 - Reserved
- 127 - Chip ID and version register

Reserved registers must not be written to during normal operation. Some reserved registers are used for test modes, which can completely alter the normal operation of the CS8420.

Addr	Function	7	6	5	4	3	2	1	0
1	Control 1	SWCLK	VSET	MUTESAO	MUTEAES	DITH	INT1	INT0	TCBLD
2	Control 2	TRUNC	HOLD1	HOLD0	RMCKF	MMR	MMT	MMTCS	MMTLR
3	Data Flow Control	AMLL	TXOFF	AESBP	TXD1	TXD0	SPD1	SPD0	SRCD
4	Clock Source Control	0	RUN	CLK1	CLK0	OUTC	INC	RXD1	RXD0
5	Serial Input Format	SIMS	SISF	SIRES1	SIRES0	SIJUST	SIDEL	SISPOL	SILRPOL
6	Serial Output Format	SOMS	SOSF	SORES1	SORES0	SOJUST	SODEL	SOSPOL	SOLRPOL
7	Interrupt 1 Status	TSLIP	OSLIP	SRE	OVRGL	OVRGR	DETC	EFTC	RERR
8	Interrupt 2 Status	0	0	VFIFO	REUNLOCK	DETU	EFTU	QCH	UOVW
9	Interrupt 1 Mask	TSLIPM	OSLIPM	SREM	OVRGLM	OVRGRM	DETCM	EFTCM	RERRM
10	Interrupt 1 Mode (MSB)	TSLIP1	OSLIP1	SRE1	OVRGL1	OVRGR1	DETC1	EFTC1	RERR1
11	Interrupt 1 Mode (LSB)	TSLIP0	OSLIP0	SRE0	OVRGL0	OVRGR0	DETC0	EFTC0	RERR0
12	Interrupt 2 Mask	0	0	VFIFOM	REUNLOCKM	DETUM	EFTUM	QCHM	UOVWM
13	Interrupt 2 Mode (MSB)	0	0	VFIFO1	REUNLOCK1	DETU1	EFTU1	QCH1	UOVW1
14	Interrupt 2 Mode (LSB)	0	0	VFIFO0	REUNLOCK0	DETU0	EFTU0	QCH0	UOVW0
15	Receiver CS Data	AUX3	AUX2	AUX1	AUX0	PRO	AUDIO	COPY	ORIG
16	Receiver Errors	0	QCRC	CCRC	UNLOCK	V	CONF	BIP	PAR
17	Receiver Error Mask	0	QCRCM	CCRCM	UNLOCKM	VM	CONFM	BIPM	PARM
18	CS Data Buffer Control	0	0	BSEL	CBMR	DETCI	EFTCI	CAM	CHS
19	U Data Buffer Control	0	0	0	UD	UBM1	UBM0	DETUI	EFTUI
20-29	Q sub-code Data								
30	Sample Rate Ratio	SRR7	SRR6	SRR5	SRR4	SRR3	SRR2	SRR1	SRR0
32-55	C or U Data Buffer								
127	ID and Version	ID3	ID2	ID1	ID0	VER3	VER2	VER1	VER0

Table 2. Summary of all Bits in the Control Register Map

11.2 Miscellaneous Control 1 (1)

7	6	5	4	3	2	1	0
SWCLK	VSET	MUTESAO	MUTEAES	DITH	INT1	INT0	TCBLD

SWCLK — Controls the output of OMCK on the RMCK pin in the absence of input to the Receiver
0 - RMCK default function
1 - OMCK is switched to output through RMCK in the absence of input to the Receiver

VSET — Transmitted V bit level
0 - Transmit a 0 for the V bit, indicating that the data is valid, and is normally linear PCM audio (default)
1 - Transmit a 1 for the V bit, indicating that the data is invalid or is not linear PCM audio data

MUTESAO — Mute control for the serial audio output port
0 - Normal output (default)
1 - Mute the serial audio output port

MUTEAES — Mute control for the AES3 transmitter output
0 - Normal output (default)
1 - Mute the AES3 transmitter output

DITH — Dither Control
0 - Triangular PDF dither applied to output data. The level of the dither is automatically adjusted to be appropriate for the output word length selected by the SORES bits (default)
1 - No dither applied to output data.

INT1-INT0 — Interrupt (INT) output pin control
00 - Active high, high output indicates an interrupt condition has occurred (default)
01 - Active low, low output indicates an interrupt condition has occurred
10 - Open drain, active low. This setting requires an external pull up resistor on the INT pin.
11 - Reserved

TCBLD — Transmit Channel Status Block pin (TCBL) direction specifier
0 - TCBL is an input (default)
1 - TCBL is an output

11.3 Miscellaneous Control 2 (2)

7	6	5	4	3	2	1	0
TRUNC	HOLD1	HOLD0	RMCKF	MMR	MMT	MMTCS	MMTLR

TRUNC Determines whether the word length is set according to the incoming Channel Status data
0 - Data to the SRC is not truncated (default)
1 - Data to the SRC is set according to the AUX field in the incoming data stream

HOLD1-0The HOLD bits determine how the received audio sample is affected when a receiver error occurs.
00 - Hold the last valid audio sample (default)
01 - Replace the current audio sample with 00 (mute)
10 - Do not change the received audio sample
11 - Reserved

RMCKFSelect recovered master clock output pin frequency.
0 - RMCK is equal to 256 * Fsi (default)
1 - RMCK is equal to 128 * Fsi

MMR Select AES3 receiver mono or stereo operation
0 - Interpret A and B subframes as two independent channels (normal stereo operation, default)
1 - Interpret A and B subframes as consecutive samples of one channel of data. This data is duplicated to both left and right parallel outputs of the AES receiver block. The input sample rate (Fsi) is doubled compared to MMR=0

MMT Select AES3 transmitter mono or stereo operation
0 - Outputs left channel input into A subframe and right channel input into B subframe (normal stereo operation, default).
1 - Output either left or right channel inputs into consecutive subframe outputs (mono mode, left or right is determined by MMTLR bit)

MMTCS Select A or B channel status data to transmit in mono mode
0 - Use channel A CS data for the A sub-frame slot and use channel B CS data for the B sub-frame slot (default)
1 - Use the same CS data for both the A and B sub-frame output slots. If MMTLR = 0, use the left channel CS data. If MMTLR = 1, use the right channel CS data.

MMTLR Channel Selection for AES Transmitter mono mode
0 - Use left channel input data for consecutive sub-frame outputs (default)
1- Use right channel input data for consecutive sub-frame outputs

11.4 Data Flow Control (3)

7	6	5	4	3	2	1	0
AMLL	TXOFF	AESBP	TXD1	TXD0	SPD1	SPD0	SRCD

The Data Flow Control register configures the flow of audio data to/from the following blocks: Serial Audio Input Port, Serial Audio Output Port, AES3 receiver, AES3 transmitter, and Sample Rate Converter. In conjunction with the Clock Source Control register, multiple Receiver/Transmitter/Transceiver modes may be selected. The output data should be muted prior to changing bits in this register to avoid transients.

AMLL — Auto Mutes the SRC data sink when Receiver lock is lost, zero data is transmitted. The SRC data sink may be either, or both, the Transmitter and the Serial Audio Output Port.
0 - Disables Auto Mute on loss of lock (default)
1 - Enables Auto Mute on loss of lock

TXOFF — AES3 Transmitter Output Driver Control
0 - AES3 transmitter output pin drivers normal operation (default)
1 - AES3 transmitter output pin drivers drive to 0V.

AESBP — AES3 bypass mode selection
0 - normal operation
1 - Connect the AES3 transmitter driver input directly to the RXP pin, which become a normal TTL threshold digital input.

TXD1 - TXD0 — AES3 Transmitter Data Source
00 - SRC output (default)
01 - Serial audio input port
10 - AES3 receiver
11 - Reserved

SPD1 - SPD0 — Serial Audio Output Port Data Source
00 - SRC output (default)
01 - Serial Audio Input Port
10 - AES3 receiver
11 - Reserved

SRCD — Input Data Source for SRC
0 - Serial Audio Input Port (default)
1 - AES3 Receiver

11.5 Clock Source Control (4)

7	6	5	4	3	2	1	0
0	RUN	CLK1	CLK0	OUTC	INC	RXD1	RXD0

This register configures the clock sources of various blocks. In conjunction with the Data Flow Control register, various Receiver/Transmitter/Transceiver modes may be selected.

RUN

The RUN bit controls the internal clocks, allowing the CS8420 to be placed in a "powered down", low current consumption, state.

0 - Internal clocks are stopped. Internal state machines are reset. The fully static control port is operational, allowing registers to be read or changed. Reading and writing the U and C data buffers is not possible. Power consumption is low (default).

1 - Normal part operation. This bit must be written to the 1 state to allow the CS8420 to begin operation. All input clocks should be stable in frequency and phase when RUN is set to 1.

CLK1-0

Output side master clock input (OMCK) frequency to output sample rate (Fso) ratio selector. If these bits are changed during normal operation, then always stop the CS8420 first (RUN = 0), then write the new value, then start the CS8420 (RUN = 1).

00 - OMCK frequency is 256*Fso(default)

01 - OMCK frequency is 384*Fso

10 - OMCK frequency is 512*Fso

11 - reserved

OUTC

Output Time Base

0 - OMCK input pin (modified by the selected divide ratio bits CLK1 & CLK0, default)

1 - Recovered Input Clock

INC

Input Time Base Clock Source

0 - Recovered Input Clock (default)

1 - OMCK input pin (modified by the selected divide ratio bits CLK1 & CLK0)

RXD1-0

Recovered Input Clock Source

00 - 256*Fsi, where Fsi is derived from the ILRCK pin (only possible when the serial audio input port is in slave mode, default)

01 - 256*Fsi, where Fsi is derived from the AES3 input frame rate

10 - Bypass the PLL and apply an external 256*Fsi clock via the RMCK pin. The AES3 receiver is held in synchronous reset. This setting is useful to prevent UNLOCK interrupts when using an external RMCK and inputting data via the serial audio input port.

11 - Reserved

11.6 Serial Audio Input Port Data Format (5)

7	6	5	4	3	2	1	0
SIMS	SISF	SIRES1	SIRES0	SIJUST	SIDEL	SISPOL	SILRPOL

SIMS — Master/Slave Mode Selector
0 - Serial audio input port is in slave mode (default)
1 - Serial audio input port is in master mode

SISF — ISCLK frequency (for master mode)
0 - 64*Fsi (default)
1 - 128*Fsi

SIRES1-0 — Resolution of the input data, for right-justified formats
00 - 24 bit resolution (default)
01 - 20 bit resolution
10 - 16 bit resolution
11 - Reserved

SIJUST — Justification of SDIN data relative to ILRCK
0 - Left-justified (default)
1 - Right-justified

SIDEL — Delay of SDIN data relative to ILRCK, for left-justified data formats
0 - MSB of SDIN data occurs in the first ISCLK period after the ILRCK edge (default)
1 - MSB of SDIN data occurs in the second ISCLK period after the ILRCK edge

SISPOL — ISCLK clock polarity
0 - SDIN sampled on rising edges of ISCLK (default)
1 - SDIN sampled on falling edges of ISCLK

SILRPOL — ILRCK clock polarity
0 - SDIN data is for the left channel when ILRCK is high (default)
1 - SDIN data is for the right channel when ILRCK is high

11.7 Serial Audio Output Port Data Format (6)

7	6	5	4	3	2	1	0
SOMS	SOSF	SORES1	SORES0	SOJUST	SODEL	SOSPOL	SOLRPOL

SOMS — Master/Slave Mode Selector
0 - Serial audio output port is in slave mode (default)
1 - Serial audio output port is in master mode

SOSF — OSCLK frequency (for master mode)
0 - 64*Fso (default)
1 - 128*Fso

SORES1-0 — Resolution of the output data on SDOUT and on the AES3 output
00 - 24 bit resolution (default)
01 - 20 bit resolution
10 - 16 bit resolution
11 - Direct copy of the received NRZ data from the AES3 receiver (including C, U, and V bits, the time slot normally occupied by the P bit is used to indicate the location of the block start, SDOUT pin only, serial audio output port clock must be derived from the AES3 receiver recovered clock)

SOJUST — Justification of SDOUT data relative to OLRCK
0 - Left-justified (default)
1 - Right-justified (master mode only)

SODEL — Delay of SDOUT data relative to OLRCK, for left-justified data formats
0 - MSB of SDOUT data occurs in the first OSCLK period after the OLRCK edge (default)
1 - MSB of SDOUT data occurs in the second OSCLK period after the OLRCK edge

SOSPOL — OSCLK clock polarity
0 - SDOUT transitions occur on falling edges of OSCLK (default)
1 - SDOUT transitions occur on rising edges of OSCLK

SOLRPOL — OLRCK clock polarity
0 - SDOUT data is for the left channel when OLRCK is high (default)
1 - SDOUT data is for the right channel when OLRCK is high

11.8 Interrupt 1 Register Status (7) (Read Only)

7	6	5	4	3	2	1	0
TSLIP	OSLIP	SRE	OVRGL	OVRGR	DETC	EFTC	RERR

For all bits in this register, a “1” means the associated interrupt condition has occurred at least once since the register was last read. A ”0” means the associated interrupt condition has NOT occurred since the last reading of the register. Reading the register resets all bits to 0, unless the interrupt mode is set to level and the interrupt source is still true. Status bits that are masked off in the associated mask register will always be “0” in this register. This register defaults to 00.

TSLIP — AES3 transmitter source data slip interrupt. In data flows with no SRC, and where OMCK, which clocks the AES3 transmitter, is asynchronous to the data source, this bit will go high every time a data sample is dropped or repeated. Also, when TCBL is an input, and when the SRC is not in use, this bit will go high on receipt of a new TCBL signal.

OSLIP — Serial audio output port data slip interrupt. When the serial audio output port is in slave mode, and OLRCK is asynchronous to the port data source, this bit will go high every time a data sample is dropped or repeated. Also, when the SRC is used, and the SRC output goes to the output serial port configured in slave mode, this bit will indicate if the ratio of OMCK frequency to OLRCK frequency does not match what is set in the CLK1 and CLK0 bits.

SRE — Sample rate range exceeded indicator. Occurs if Fsi/Fso or Fso/Fsi exceeds 3.

OVRGL — Over-range indicator for left (A) channel SRC output. Occurs on internal over-range for left channel data. Note that the CS8420 automatically clips over-ranges to plus or minus full-scale.

OVRGR — Over-range indicator for right (B) channel SRC output. Occurs on internal over-range for right channel data. Note that the CS8420 automatically clips over-ranges to plus or minus full-scale

DETC — D to E C-buffer transfer interrupt. The source for this bit is true during the D to E buffer transfer in the C bit buffer management process.

EFTC — E to F C-buffer transfer interrupt. The source for this bit is true during the E to F buffer transfer in the C bit buffer management process.

RERR — A receiver error has occurred. The Receiver Error register may be read to determine the nature of the error which caused the interrupt.

11.9 Interrupt Register 2 Status (8) (Read Only)

7	6	5	4	3	2	1	0
0	0	VFIFO	REUNLOCK	DETU	EFTU	QCH	UOVW

For all bits in this register, a “1” means the associated interrupt condition has occurred at least once since the register was last read. A ”0” means the associated interrupt condition has NOT occurred since the last reading of the register. Reading the register resets all bits to 0, unless the interrupt mode is set to level and the interrupt source is still true. Status bits that are masked off in the associated mask register will always be “0” in this register. This register defaults to 00.

VFIFO — Varispeed FIFO overflow indicator. Occurs if the data buffer in the SRC overflows. This will occur if the input sample rate slows too fast.

REUNLOCK — Sample rate converter unlock indicator. This interrupt occurs if the SRC is still tracking a changing input or output sample rate.

DETU — D to E U-buffer transfer interrupt. The source of this bit is true during the D to E buffer transfer in the U bit buffer management process (block mode only).

EFTU — E to F U-buffer transfer interrupt. The source of this bit is true during the E to F buffer transfer in the U bit buffer management process (block mode only).

QCH — A new block of Q-subcode data is available for reading. The data must be completely read within 588 AES3 frames after the interrupt occurs to avoid corruption of the data by the next block.

UOVW — U-bit FIFO Overwrite. This interrupt occurs on an overwrite in the U-bit FIFO.

11.10 Interrupt 1 Register Mask (9)

7	6	5	4	3	2	1	0
TSLIPM	OSLIPM	SREM	OVRGLM	OVRGRM	DETCM	EFTCM	RERRM

The bits of this register serve as a mask for the Interrupt 1 Register. If a mask bit is set to 1, the error is considered unmasked, meaning that its occurrence will affect the INT pin and the status register. If a mask bit is set to 0, the error is considered masked, meaning that its occurrence will not affect the INT pin or the status register. The bit positions align with the corresponding bits in Interrupt Register 1. This register defaults to 00.

11.11 Interrupt Register 1 Mode Registers MSB & LSB(10,11)

7	6	5	4	3	2	1	0
TSLIP1	OSLIP1	SRE1	OVRGL1	OVRGR1	DETC1	EFTC1	RERR1
TSLIP0	OSLIP0	SRE0	OVRGL0	OVRGR0	DETC0	EFTC0	RERR0

The two Interrupt Mode registers form a 2-bit code for each Interrupt Register 1 function. This code determines whether the INT pin is set active on the arrival of the interrupt condition, on the removal of the interrupt condition, or on the continuing occurrence of the interrupt condition. These registers default to 00.

00 - Rising edge active
01 - Falling edge active
10 - Level active
11 - Reserved

11.12 Interrupt 2 Register Mask (12)

7	6	5	4	3	2	1	0
0	0	VFIFOM	REUNLOCKM	DETUM	EFTUM	QCHM	UOVWM

The bits of this register serve as a mask for the Interrupt 2 Register. If a mask bit is set to 1, the error is considered unmasked, meaning that its occurrence will affect the INT pin and the status register. If a mask bit is set to 0, the error is considered masked, meaning that its occurrence will not affect the INT pin or the status register. The bit positions align with the corresponding bits in Interrupt Register 2. This register defaults to 00.

11.13 Interrupt Register 2 Mode Registers MSB & LSB(13,14)

7	6	5	4	3	2	1	0
0	0	VFIFO1	REUNLOCK1	DETU1	EFTU1	QCH1	UOVW1
0	0	VFIFO0	REUNLOCK0	DETU0	EFTU0	QCH0	UOVW0

The two Interrupt Mode registers form a 2-bit code for each Interrupt 2 register function. This code determines whether the INT pin is set active on the arrival of the interrupt condition, on the removal of the interrupt condition, or on the continuing occurrence of the interrupt condition. These registers default to 00.

00 - Rising edge active
01 - Falling edge active
10 - Level active
11 - Reserved

11.14 Receiver Channel Status (15) (Read Only)

7	6	5	4	3	2	1	0
AUX3	AUX2	AUX1	AUX0	PRO	$\overline{\text{AUDIO}}$	COPY	ORIG

The bits in this register can be associated with either channel A or B of the received data. The desired channel is selected with the CHS bit of the Channel Status Data Buffer Control Register.

AUX3-0 The AUX3-0 bits indicate the width of the incoming auxiliary data field, as indicated by the incoming channel status bits, decoded according to IEC60958 and AES3.
0000 - Auxiliary data is not present
0001 - Auxiliary data is 1 bit long
0010 - Auxiliary data is 2 bits long
0011 - Auxiliary data is 3 bits long
0100 - Auxiliary data is 4 bits long
0101 - Auxiliary data is 5 bits long
0110 - Auxiliary data is 6 bits long
0111 - Auxiliary data is 7 bits long
1000 - Auxiliary data is 8 bits long
1001 - 1111 Reserved

PRO Channel status block format indicator
0 - Received channel status block is in consumer format
1 - Received channel status block is in professional format

$\overline{\text{AUDIO}}$ Audio indicator
0 - Received data is linearly coded PCM audio
1 - Received data is not linearly coded PCM audio

COPY SCMS copyright indicator
0 - Copyright asserted
1 - Copyright not asserted

ORIG SCMS generation indicator. This is decoded from the category code and the L bit.
0 - Received data is 1st generation or higher
1 - Received data is original

Note: COPY and ORIG will both be set to 1 if the incoming data is flagged as professional, or if the receiver is not in use.

11.15 Receiver Error (16) (Read Only)

7	6	5	4	3	2	1	0
0	QCRC	CCRC	UNLOCK	V	CONF	BIP	PAR

This register contains the AES3 receiver and PLL status bits. Unmasked bits will go high on occurrence of the error, and will stay high until the register is read. Reading the register resets all bits to 0, unless the error source is still true. Bits that are masked off in the receiver error mask register will always be 0 in this register. This register defaults to 00.

QCRCQ-subcode data CRC error has occurred. Updated on Q-subcode block boundaries.
0 - No error
1 - Error

CCRCChannel Status Block Cyclic Redundancy Check bit. Updated on CS block boundaries. This bit is valid in professional mode only.
0 - No error
1 - Error

UNLOCK — PLL lock status bit. Updated on CS block boundaries.
0 - PLL locked
1 - PLL out of lock

V — Received AES3 Validity bit status. Updated on sub-frame boundaries.
0 - Data is valid and is normally linear coded PCM audio
1 - Data is invalid, or may be valid compressed audio

CONF — Confidence bit. Updated on sub-frame boundaries.
0 - No error
1 - Confidence error. This indicates that the received data eye opening is less than half a bit period, indicating a poor link that is not meeting specifications.

BIP — Bi-phase error bit. Updated on sub-frame boundaries.
0 - No error
1 - Bi-phase error. This indicates an error in the received bi-phase coding.

PAR — Parity bit. Updated on sub-frame boundaries.
0 - No error
1 - Parity error

11.16 Receiver Error Mask (17)

7	6	5	4	3	2	1	0
0	QCRCM	CCRCM	UNLOCKM	VM	CONFM	BIPM	PARM

The bits in this register serve as masks for the corresponding bits of the Receiver Error Register. If a mask bit is set to 1, the error is considered unmasked, meaning that its occurrence will appear in the receiver error register, will affect the RERR pin, will affect the RERR interrupt, and will affect the current audio sample according to the status of the HOLD bit. If a mask bit is set to 0, the error is considered masked, meaning that its occurrence will not appear in the receiver error register, will not affect the RERR pin, will not affect the RERR interrupt, and will not affect the current audio sample. The CCRC and QCRC bits behave differently from the other bits: they do not affect the current audio sample even when unmasked. This register defaults to 00.

11.17 Channel Status Data Buffer Control (18)

7	6	5	4	3	2	1	0
0	0	BSEL	CBMR	DETCI	EFTCI	CAM	CHS

BSEL — Selects the data buffer register addresses to contain User data or Channel Status data
0 - Data buffer address space contains Channel Status data (default)
1 - Data buffer address space contains User data

CBMR — Control for the first 5 bytes of channel status "E" buffer
0 - Allow D to E buffer transfers to overwrite the first 5 bytes of channel status data (default)
1 - Prevent D to E buffer transfers from overwriting first 5 bytes of channel status data

DETCI — D to E C-data buffer transfer inhibit bit.
0 - Allow C-data D to E buffer transfers (default)
1 - Inhibit C-data D to E buffer transfers

EFTCI — E to F C-data buffer transfer inhibit bit.
0 - Allow C-data E to F buffer transfers (default)
1 - Inhibit C-data E to F buffer transfers

CAM — C-data buffer control port access mode bit
0 - One byte mode
1 - Two byte mode

CHS — Channel select bit
0 - Channel A information is displayed at the $\overline{\text{EMPH}}$ pin and in the receiver channel status register. Channel A information is output during control port reads when CAM is set to 0 (One Byte Mode)
1 - Channel B information is displayed at the $\overline{\text{EMPH}}$ pin and in the receiver channel status register. Channel B information is output during control port reads when CAM is set to 0 (One Byte Mode)

11.18 User Data Buffer Control (19)

7	6	5	4	3	2	1	0
0	0	0	UD	UBM1	UBM0	DETUI	EFTUI

UD — User data pin (U) direction specifier
0 - The U pin is an input. The U data is latched in on both rising and falling edges of OLRCK. This setting also chooses the U pin as the source for transmitted U data (default).
1 - The U pin is an output. The received U data is clocked out on both rising and falling edges of ILRCK. This setting also chooses the U data buffer as the source of transmitted U data.

UBM1-0 — Sets the operating mode of the AES3 U bit manager
00 - Transmit all zeros mode (default)
01 - Block mode
10 - Reserved
11 - IEC consumer mode 4

DETUI — D to E U-data buffer transfer inhibit bit (valid in block mode only).
0 - Allow U-data D to E buffer transfers (default)
1 - Inhibit U-data D to E buffer transfers

EFTUI — E to F U-data buffer transfer inhibit bit (valid in block mode only).
0 - Allow U-data E to F buffer transfers (default)
1 - Inhibit U-data E to F buffer transfer

11.19 Q-Channel Subcode Bytes 0 to 9 (20 - 29) (Read Only)

The following 10 registers contain the decoded Q-channel subcode data

7	6	5	4	3	2	1	0
ADDRESS	ADDRESS	ADDRESS	ADDRESS	CONTROL	CONTROL	CONTROL	CONTROL
TRACK	TRACK	TRACK	TRACK	TRACK	TRACK	TRACK	TRACK
INDEX	INDEX	INDEX	INDEX	INDEX	INDEX	INDEX	INDEX
MINUTE	MINUTE	MINUTE	MINUTE	MINUTE	MINUTE	MINUTE	MINUTE
SECOND	SECOND	SECOND	SECOND	SECOND	SECOND	SECOND	SECOND
FRAME	FRAME	FRAME	FRAME	FRAME	FRAME	FRAME	FRAME
ZERO	ZERO	ZERO	ZERO	ZERO	ZERO	ZERO	ZERO
ABS MINUTE	ABS MINUTE	ABS MINUTE	ABS MINUTE	ABS MINUTE	ABS MINUTE	ABS MINUTE	ABS MINUTE
ABS SECOND	ABS SECOND	ABS SECOND	ABS SECOND	ABS SECOND	ABS SECOND	ABS SECOND	ABS SECOND
ABS FRAME	ABS FRAME	ABS FRAME	ABS FRAME	ABS FRAME	ABS FRAME	ABS FRAME	ABS FRAME

11.20 Sample Rate Ratio (30) (Read Only)

7	6	5	4	3	2	1	0
SRR7	SRR6	SRR5	SRR4	SRR3	SRR2	SRR1	SRR0

The Sample Rate Ratio is Fso divided by Fsi. This value is represented as an integer and a fractional part. The value is meaningful only after the both the PLL and SRC have reached lock, and the SRC output is being used

SRR7-6The integer part of the sample rate ratio

SRR5-0 The fractional part of the sample rate ratio

11.21 C-bit or U-bit Data Buffer (32 - 55)

Either channel status data buffer E or user data buffer E (provided UBM bits are set to block mode) is accessible via these register addresses.

11.22 CS8420 I.D. and Version Register (127) (Read Only)

7	6	5	4	3	2	1	ID3
ID3	ID2	ID1	ID0	VER3	VER2	VER1	VER0

ID3-0 ID code for the CS8420. Permanently set to 0001

VER3-0 CS8420 revision level. Revision B is coded as 0001, Revision C is coded as 0011, Revision D is coded as 0100

12. SYSTEM AND APPLICATIONS ISSUES

12.1 Reset, Power Down and Start-up Options

When $\overline{\text{RST}}$ is low, the CS8420 enters a low power mode and all internal states are reset, including the control port and registers, and the outputs are muted. When $\overline{\text{RST}}$ is high, the control port becomes operational and the desired settings should be loaded into the control registers. Writing a 1 to the RUN bit will then cause the part to leave the low power state and begin operation. After the PLL and the SRC have settled, the AES3 and serial audio outputs will be enabled.

Some options within the CS8420 are controlled by a start-up mechanism. During the reset state, some of the output pins are reconfigured internally to be inputs. Immediately upon exiting the reset state, the level of these pins is sensed. The pins are then switched to be outputs. This mechanism allows output pins to be used to set alternative modes in the CS8420 by connecting a 47kΩ resistor to between the pin and either VD+ (HI) or DGND (LO). For each mode, every start-up option select pin MUST have an external pull-up or pull-down resistor. In software mode, the only start-up option pin is $\overline{\text{EMPH}}$, which is used to set a chip address bit for the control port in I^2C mode. Hardware modes use many start-up options, which are detailed in the hardware definition section at the end of this data sheet.

12.2 ID Code and Revision Code

The CS8420 has a register that contains a 4-bit code to indicate that the addressed device is a CS8420. This is useful when other CS84XX family members are resident in the same system, allowing common software modules.

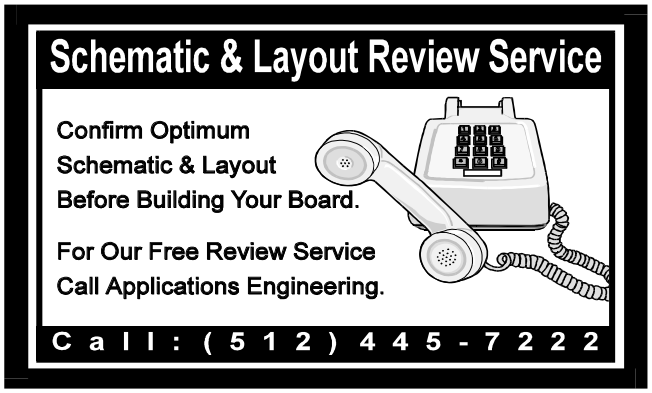
The CS8420 4-bit revision code is also available. This allows the software driver for the CS8420 to identify which revision of the device is in a particular system, and modify its behavior accordingly. To allow for future revisions, it is strongly recommend that the revision code is read into a variable area within the microcontroller, and used wherever appropriate as revision details become known.

12.3 Power Supply, Grounding, and PCB layout

For most applications, the CS8420 can be operated from a single +5V supply, following normal supply decoupling practice (see Figure 5. “Recommended Connection Diagram for Software Mode” on page 10). For applications where the recovered input clock, output on the RMCK pin, is required to be low jitter, then use a separate, quiet, analog +5V supply for VA+, decoupled to AGND. In addition, a separate region of analog ground plane around the FILT, AGND, VA+, RXP and RXN pins is recommended.

The VD+ supply should be well decoupled with a 0.1µF capacitor to DGND to minimize AES3 transmitter induced transients.

Extensive use of power and ground planes, ground plane fill in unused areas and surface mount decoupling capacitors are recommended. Make sure decoupling capacitors are mounted on the same side of the board as the CS8420, to minimize via inductance effects. All decoupling capacitors should be as close to the CS8420 as possible.



12.4 Synchronization of Multiple CS8420s

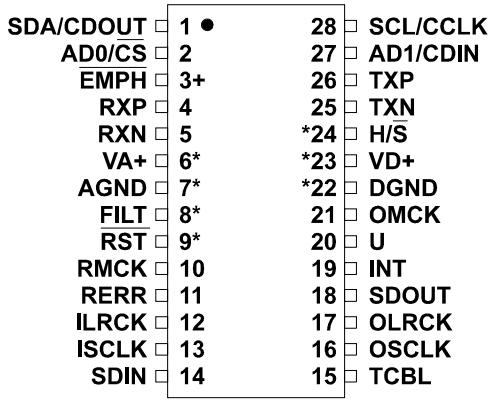
The serial audio output ports of multiple CS8420s can be synchronized by sharing the same master clock, OSCLK, OLRCK, and $\overline{\text{RST}}$ line and ensuring that all devices leave the reset state on the same master clock falling edge. Either all the ports need to be in slave mode, or one can be set as a master.

The AES3 transmitters may be synchronized by sharing the same master clock, TCBL, and $\overline{\text{RST}}$ signals, and all devices leave the reset state on the same master clock falling edge. The TCBL pin is used to synchronize multiple CS8420 AES3 transmitters at the channel status block boundaries. One CS8420 must have its TCBL set to master; the others must be set to slave TCBL. Alternatively, TCBL can be derived from some external logic, in which case all the CS8420 devices should be set to slave TCBL.

12.5 Extended Range Sample Rate Conversion

For handling sampling rate conversion ratios greater than 3:1 or less than 1:3, the user can use a cascade of two devices. The product of the conversion ratio of the two devices should equal the target conversion ratio.

13. SOFTWARE MODE - PIN DESCRIPTION



* Pins which remain the same function in all modes.
\+ Pins which require a pull up or pull down resistor to select the desired startup option.

The above diagram and the following pin descriptions apply to software mode. In hardware mode, some pins change their function as described in subsequent sections of this data sheet. Fixed function pins are marked with a *, and will be described once in this section. Pins marked with a + are used upon reset to select various start-up options, and require a pull-up or pull-down resistor.

Power Supply Connections:

VD+ - Positive Digital Power *

Positive supply for the digital section. Nominally +5V.

VA+ - Positive Analog Power *

Positive supply for the analog section. Nominally +5V. This supply should be as quiet as possible since noise on this pin will directly affect the jitter performance of the recovered clock.

DGND - Digital Ground *

Ground for the digital section. DGND should be connected to the same ground as AGND.

AGND - Analog Ground *

Ground for the analog section. AGND should be connected to the same ground as DGND.

Clock Related Pins:

OMCK - Output Section Master Clock Input

Output section master clock input. The frequency must be 256x, 384x, or 512x the output sample rate (Fso).

RMCK - Input Section Recovered Master Clock Output

Input section recovered master clock output. Will be at a frequency of 128x or 256x the input sample rate (Fsi).

FILT - PLL Loop Filter *

An RC network should be connected between this pin and ground. Recommended schematic and component values are given in the Receiver section of this data sheet.

Overall Device Control:

H/$\overline{S}$ - Hardware or Software Control Mode Select *

The H/$\overline{S}$ pin determines the method of controlling the operation of the CS8420, and the method of accessing CS and U data. In software mode, device control and CS and U data access is primarily via the control port, using a microcontroller. In hardware mode, alternate modes and access to CS and U data is provided by pins. This pin should be permanently tied to VD+ or DGND.

$\overline{RST}$ - Reset Input *

When $\overline{RST}$ is low, the CS8420 enters a low power mode and all internal states are reset. On initial power up, $\overline{RST}$ must be held low until the power supply is stable, and all input clocks are stable in frequency and phase. This is particularly true in hardware mode with multiple CS8420 devices, where synchronization between devices is important.

INT - Interrupt Output

The INT output pin indicates errors and key events during the operation of the CS8420. All bits affecting INT are maskable via control registers. The condition(s) that initiated interrupt are readable via a control register. The polarity of the INT output, as well as selection of a standard or open drain output, is set via a control register. Once set true, the INT pin goes false only after the interrupt status registers have been read, and the interrupt status bits have returned to zero.

Audio Input Interface:

SDIN - Serial Audio Input Port Data Input

Audio data serial input pin.

ISCLK - Serial Audio Input Port Bit Clock input or output

Serial bit clock for audio data on the SDIN pin.

ILRCK - Serial Audio Input Port Left/Right Clock input or output

Word rate clock for the audio data on the SDIN pin. The frequency will be at the input sample rate (Fsi)

AES3/SPDIF Receiver Interface:

RXP, RXN - Differential Line Receiver Inputs

Differential line receiver inputs, carrying AES3 type data.

RERR - Receiver Error Indicator

When high, indicates a problem with the operation of the AES3 receiver. The status of this pin is updated once per sub-frame of incoming AES3 data. Conditions that can cause RERR to go high are: validity, parity error, bi-phase coding error, confidence, QCRC and CCRC errors, as well as loss of lock in the PLL. Each condition may be optionally masked from affecting the RERR pin using the Receiver Error Mask Register. The RERR pin tracks the status of the unmasked errors: the pin goes high as soon as an unmasked error occurs and goes low immediately when all unmasked errors go away.

$\overline{EMPH}$ - Pre-emphasis Indicator Output

$\overline{EMPH}$ is low when the incoming AES3 data indicates the presence of 50/15 µs pre-emphasis. When the AES3 data indicates the absence of pre-emphasis or the presence of other than 50/15 µs pre-emphasis $\overline{EMPH}$ is high. This is also a start-up option pin, and requires a 47 kΩ resistor to either VD+ or DGND, which determines the AD2 address bit for the control port in I^2C mode.

Audio Output Interface:

SDOUT - Serial Audio Output Port Data Output

Audio data serial output pin.

OSCLK - Serial Audio Output Port Bit Clock input or output

Serial bit clock for audio data on the SDOUT pin.

OLRCK - Serial Audio Output Port Left/Right Clock input or output

Word rate clock for the audio data on the SDOUT pin. The frequency will be at the output sample rate (Fso)

AES3/SPDIF Transmitter Interface:

TCBL - Transmit Channel Status Block Start

This pin can be configured as an input or output. When operated as output, TCBL is high during the first sub-frame of a transmitted channel status block, and low at all other times. When operated as input, driving TCBL high for at least three OMCK (or RMCK, depending on which clock is operating the AES3 encoder block) clocks will cause the next transmitted sub-frame to be the start of a channel status block.

TXN, TXP - Differential Line driver outputs

Differential line driver outputs, transmitting AES3 type data. Drivers are pulled to low while the CS8420 is in the reset state.

Control Port Signals:

SCL/CCLK - Control Port clock

SCL/CCLK is the serial control interface clock, and is used to clock control data bits into and out of the CS8420.

AD0/$\overline{CS}$ - Address Bit 0 (I^2C) / Control Port Chip Select (SPI)

A falling edge on this pin puts the CS8420 into SPI control port mode. With no falling edge, the CS8420 defaults to I^2C mode. In I^2C mode, AD0 is a chip address pin. In SPI mode, $\overline{CS}$ is used to enable the control port interface on the CS8420.

AD1/CDIN - Address Bit 1 (I^2C) / Serial Control data in (SPI)

In I^2C mode, AD1 is a chip address pin. In SPI mode, CDIN is the input data line for the control port interface

SDA/CDOUT - Serial Control Data I/O (I^2C) / data out (SPI)

In I^2C mode, SDA is the control I/O data line. SDA is open drain, and requires an external pull-up resistor to VD+. In SPI mode, CDOUT is the output data from the control port interface on the CS8420.

Miscellaneous pins:

U - User Data

The U pin may optionally be used to input User data for transmission by the AES3 transmitter, see Figure 22 for timing information. Alternatively, the U pin may be set to output User data from the AES3 receiver, see Figure 21 for timing information. If not driven, a 47kΩ pull-down resistor is recommended for the U pin, since the default state of the UD direction bit sets the U pin as an input. The pull-down resistor ensures that the transmitted user data will be zero. If the U pin is always set to be an output, thereby causing the U bit manager to be the source of the U data, then the resistor is not necessary. The U pin should not be tied directly to ground, in case it is programmed to be an output, and subsequently tries to output a logic high. This situation may affect the long term reliability of the device. If the U pin is driven by a logic level output, then a 100 Ω series resistor is recommended.

14. HARDWARE MODES

14.1 Overall Description

The CS8420 has six hardware modes, which allow use of the device without using a micro-controller to access the device control registers and CS & U data. The flexibility of the CS8420 is necessarily limited in hardware mode. Various pins change function in hardware mode, and various data paths are also possible. These alternatives are identified by hardware mode numbers 1 through 6. The following sections describe the data flows and pin definitions for each hardware mode.

14.1.1 Hardware Mode Definitions

Hardware mode is selected by connecting the H/$\overline{S}$ pin to '1'. In hardware mode, 3 pins (DFC0, DFC1 & S/$\overline{AES}$) determine the hardware mode number, according to Table 3Start-up options are used extensively in hardware mode. Options include whether the serial audio output ports are master or slave, the serial audio ports' format and whether TCBL is an input or an output. Which output pins are used to set which modes depends on which hardware mode is being used.

DFC1	DFC0	S/$\overline{AES}$	Hardware Mode Number
0	0	0	1 - Default Data Flow, AES3 input
0	0	1	2 - Default Data Flow, serial input
0	1	-	3 - Transceive Flow, with SRC
1	0	-	4 - Transceive Flow, no SRC
1	1	0	5 - AES3 Rx only, AES3 input
1	1	1	6 - AES3 Tx only, serial input

Table 3. Hardware Mode Definitions

14.1.2 Serial Audio Port Formats

In hardware mode, only a limited number of alternative serial audio port formats are available. These formats are described by Tables 4 & 5, which define the equivalent software mode bit settings for each format. Timing diagrams are shown in Figures 16 and 17.

For each hardware mode, the following pages contain a data flow diagram, a pin-out drawing, a pin descriptions list and a definition of the available start-up options.

	SOSF	SORES1/0	SOJUST	SODEL	SOSPOL	SOLRPOL
OF1 - Left Justified	0	00	0	0	1	0
OF2 - I^2S 24-bit data	0	00	0	1	0	1
OF3 - Right Justified, master mode only	0	00	1	0	0	0
OF4 - I^2S 16 bit data	0	10	0	1	0	1
OF5 - Direct AES3 data	0	11	0	0	1	0

Table 4. Serial Audio Output Formats Available in Hardware Mode

	SISF	SIRES1/0	SIJUST	SIDEL	SISPOL	SILRPOL
IF1 - Left Justified	0	00	0	0	1	0
IF2 - I^2S	0	00	0	1	0	1
IF3 - Right Justified 24-bit data	0	00	1	0	0	0
IF4 - Right Justified 16-bit data	0	10	1	0	0	0

Table 5. Serial Audio Input Formats Available in Hardware Mode

14.2 Hardware Mode 1 Description (DEFAULT Data Flow, AES3 Input)

Hardware Mode 1 data flow is shown in Figure 26. Audio data is input via the AES3 receiver, and rate converted. The audio data at the new rate is then output both via the serial audio output port and via the AES3 transmitter.

The channel status data, user data and validity bit information are handled in 2 alternative modes: 1A and 1B, determined by a start-up resistor on the COPY pin. In mode 1A, the received PRO, COPY, ORIG, $\overline{\text{EMPH}}$, and $\overline{\text{AUDIO}}$ channel status bits are output on pins. The transmitted channel status bits are copied from the received channel status data, and the transmitted U and V bits are 0.

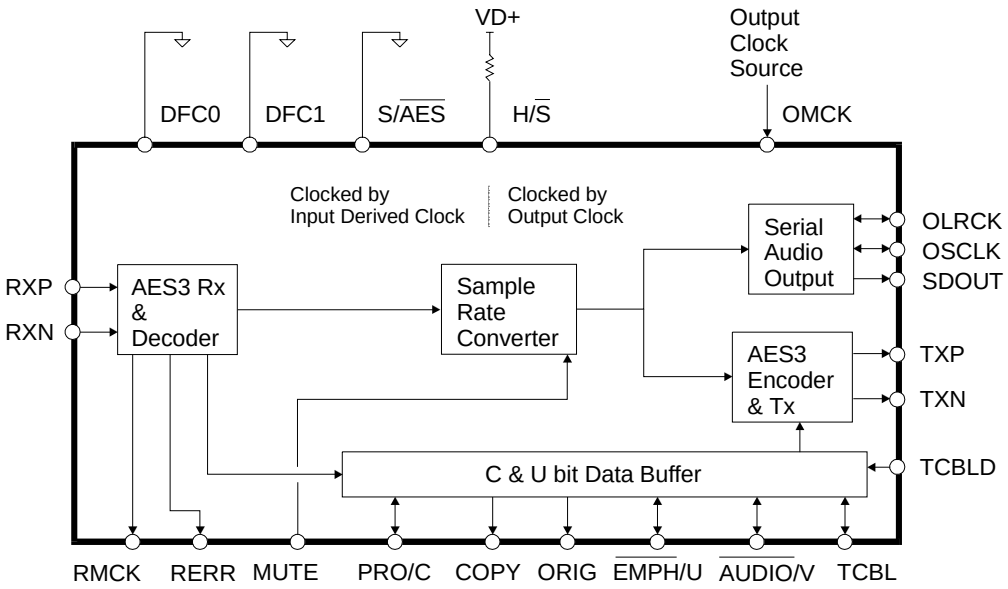
In mode 1B, only the COPY and ORIG pins are output, and reflect the received channel status data. The transmitted channel status bits, user data and validity bits are input serially via the PRO/C, $\overline{\text{EMPH}}$/U and $\overline{\text{AUDIO}}$/V pins. Figure 22 shows the timing requirements.

Start-up options are shown in Table 6, and allow choice of the serial audio output port as a master or slave, choice of 4 serial audio output port formats, and the source for transmitted C, U and V data. The following pages contain the detailed pin descriptions for hardware mode 1.

If a validity, parity, bi-phase or lock receiver error occurs, the current audio sample will be held.

SDOUT	RMCK	RERR	COPY	Function
LO	-	-	-	Serial Output Port is Slave
HI	-	-	-	Serial Output Port is Master
-	-	-	LO	Mode1A: C transmitted data is copied from received data, U & V = 0, received PRO, $\overline{\text{EMPH}}$, $\overline{\text{AUDIO}}$ are visible.
-	-	-	HI	Mode 1B: CUV transmitted data is input serially on pins, received PRO, $\overline{\text{EMPH}}$, $\overline{\text{AUDIO}}$ are not visible
-	LO	LO		Serial Output Format OF1
-	LO	HI		Serial Output Format OF2
-	HI	LO		Serial Output Format OF3
-	HI	HI		Serial Output Format OF4

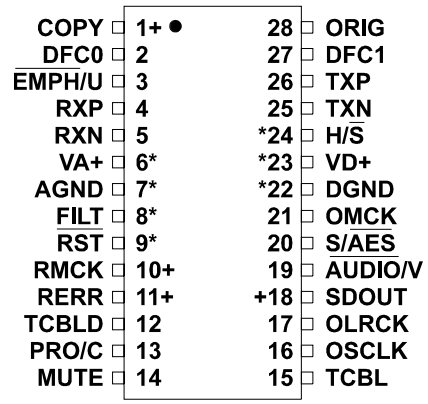
Table 6. Hardware Mode 1 Start-up Options



Power supply pins (VD+, VA+, DGND, AGND), the reset pin ($\overline{\text{RST}}$) and the PLL filter pin (FILT) are omitted from this diagram. Please refer to the Typical Connection Diagram for hook-up details.

Figure 26. Hardware Mode 1 - Default Data Flow, AES3 Input

14.2.1 Pin Description - Hardware Mode 1



* Pins which remain the same function in all modes.
\+ Pins which require a pull up or pull down resistor to select the desired startup option.

Overall Device Control:

DFC0, DFC1 - Data Flow Control Inputs

DFC0 and DFC1 inputs determine the major data flow options available in hardware mode, according to Table 3.

S/$\overline{\text{AES}}$ - Serial Audio or AES3 Input Select

S/$\overline{\text{AES}}$ is connected to ground in hardware mode 1, in order to select the AES3 input.

MUTE - Mute Output Data Input

If MUTE is low, audio data is passed normally. If MUTE is high, then both the AES3 transmitted audio data and the serial audio output port data is set to digital zero.

OMCK - Output Section Master Clock Input

Output section master clock input. The frequency must be 256x the output sample rate (Fso).

AES3/SPDIF Receiver Interface:

RXP, RXN - Differential Line Receiver Inputs

Differential line receiver inputs, carrying AES3 type data.

RMCK - Input Section Recovered Master Clock Output

Input section recovered master clock output. Will be at a frequency of 256x the input sample rate (Fsi). This is also a start-up option pin, and requires a pull-up or pull-down resistor.

RERR - Receiver Error Indicator

When high, indicates a problem with the operation of the AES3 receiver. The status of this pin is updated once per sub-frame of incoming AES3 data. Conditions that cause RERR to go high are: parity error, and bi-phase coding error, as well as loss of lock in the PLL. This is also a start-up option pin, and requires a pull-up or pull-down resistor.

$\overline{\text{EMPH}}$/U - Pre-emphasis Indicator Output or U-bit Data Input

The $\overline{\text{EMPH}}$/U pin reflects either the state of the $\overline{\text{EMPH}}$ channel status bits in the incoming AES3 type data stream, or is the serial U-bit input for the AES3 type transmitted data, clocked by OLRCK. When indicating emphasis $\overline{\text{EMPH}}$/U is low if the incoming data indicates 50/15 µs pre-emphasis and high otherwise.

COPY - Copy Channel Status Bit Output

The COPY pin reflects the state of the COPY Channel Status bit in the incoming AES3 type data stream. This is also a start-up option pin, and requires a pull-up or pull-down resistor.

ORIG - Original Channel Status Output

SCMS generation indicator. This is decoded from the incoming category code and the L bit. A low output indicates that the audio data stream is 1st generation or higher. A high indicates that the audio data stream is original.

PRO/C - Professional Channel Status Bit Output or C-bit Data Input

The PRO/C pin either reflects the state of the Professional/Consumer Channel Status bit in the incoming AES3 type data stream, or is the serial C-bit input for the AES3 type transmitted data, clocked by OLRCK.

$\overline{\text{AUDIO}}$/V - Audio Channel Status Bit Output or V-bit Data Input

The $\overline{\text{AUDIO}}$/V pin either reflects the state of the audio/non audio Channel Status bit in the incoming AES3 type data stream, or is the V-bit data input for the AES3 type transmitted data stream, clocked by OLRCK.

Audio Output Interface:

SDOUT - Serial Audio Output Port Data Output

Audio data serial output pin. This is also a start-up option pin, and requires a pull-up or pull-down resistor.

OSCLK - Serial Audio Output Port Bit Clock Input or Output

Serial bit clock for audio data on the SDOUT pin.

OLRCK - Serial Audio Output Port Left/Right Clock Input or Output

Word rate clock for the audio data on the SDOUT pin. The frequency will be at the output sample rate (Fso)

AES3/SPDIF Transmitter Interface:

TCBL - Transmit Channel Status Block Start

When operated as output, TCBL is high during the first sub-frame of a transmitted channel status block, and low at all other times. When operated as input, driving TCBL high for at least three OMCK clocks will cause the current transmitted sub-frame to be the start of a channel status block.

TCBLD - Transmit Channel Status Block Direction Input

Connect TCBLD to VD+ to set TCBL as an output. Connect TCBLD to DGND to set TCBL as an input.

TXN, TXP - Differential Line Driver Outputs

Differential line driver outputs, transmitting AES3 type data. Drivers are pulled to low while the CS8420 is in the reset state.

14.3 Hardware Mode 2 Description (DEFAULT Data Flow, Serial Input)

Hardware Mode 2 data flow is shown in Figure 27. Audio data is input via the serial audio input port, and rate converted. The audio data at the new rate is then output both via the serial audio output port and via the AES3 transmitter.

The C, U and V bits in the AES3 output stream may be set in 2 methods, selected by the CUVEN pin. When CUVEN is low, mode 2A is selected, where COPY/C, ORIG/U and $\overline{\text{EMPH}}$/V pins allow selected channel status data bits to be set. The COPY and ORIG pins are used to set the pro bit, the copy bit and the L bit, as shown in Table 7. In consumer mode, the transmitted category code shall be '0101100', which indicates sample rate converter. The transmitted U and V bits are 0.

COPY/C	ORIG/U	Function
0	0	PRO=0, COPY=0, L=0
0	1	PRO=0, COPY=0, L=1
1	0	PRO=0, COPY=1, L=0
1	1	PRO=1

Table 7. HW Mode 2A COPY/C and ORIG/U Pin Function

When the CUVEN pin is high, mode 2B is selected, where COPY/C, ORIG/U and $\overline{\text{EMPH}}$/V become serial bit inputs for C, U and V data. This data is clocked by both edges of OLRCK, and the channel status block start is indicated or determined by TCBL. Figure 22 shows the timing requirements.

Audio serial port data formats are selected as shown in Tables 8, 4, and 5.

SFMT1	SFMT0	Function
0	0	Serial Input & Output Format IF1&OF1
0	1	Serial Input & Output Format IF2&OF2
1	0	Serial Input & Output Format IF3&OF3
1	1	Serial Input & Output Format IF4&OF3

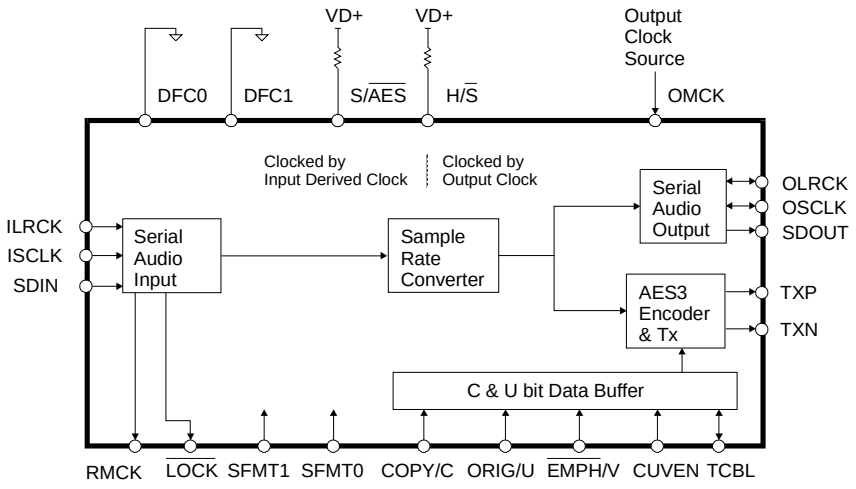
Table 8. HW Mode 2 Serial Audio Port Format Selection

Start-up options are shown in Table 9, and allow choice of the serial audio output port as a master or slave and whether TCBL is an input or an output. The serial audio input port is always a slave.

SDOUT	$\overline{\text{LOCK}}$	Function
LO	-	Serial Output Port is Slave
HI	-	Serial Output Port is Master
-	LO	TCBL is an input
-	HI	TCBL is an output

Table 9. Hardware Mode 2 Start-up Options

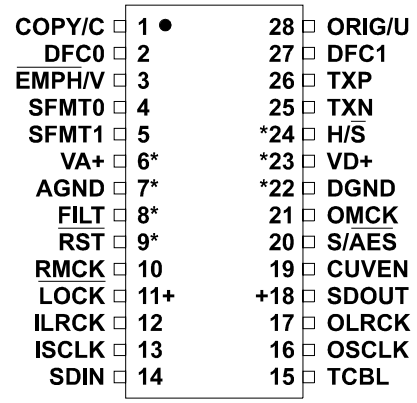
The following pages contain the detailed pin descriptions for hardware mode 2.



Power supply pins (VD+, VA+, DGND, AGND) & the reset pin ($\overline{\text{RST}}$) and the PLL filter pin (FILT) are omitted from this diagram. Please refer to the Typical Connection Diagram for hook-up details.

Figure 27. Hardware Mode 2 - Default Data Flow, Serial Audio Input

14.3.1 Pin Description - Hardware Mode 2



* Pins which remain the same function in all modes.
\+ Pins which require a pull up or pull down resistor to select the desired startup option.

Overall Device Control:

DFC0, DFC1 - Data Flow Control Inputs

DFC0 and DFC1 inputs determine the major data flow options available in hardware mode, according to Table 3.

S/$\overline{\text{AES}}$ - Serial Audio or AES3 Input Select

S/$\overline{\text{AES}}$ is connected to VD+ in hardware mode 2, in order to select the serial audio input.

SFMT0, SFMT1 - Serial Audio Port Data Format Select Inputs

SFMT0 and SFMT1 select the serial audio input and output ports' format. See Table 8.

OMCK - Output Section Master Clock Input

Output section master clock input. The frequency must be 256x the output sample rate (Fso).

Audio Input Interface:

SDIN - Serial Audio Input Port Data Input

Audio data serial input pin.

ISCLK - Serial Audio Input Port Bit Clock Input or Output

Serial bit clock for audio data on the SDIN pin.

ILRCK - Serial Audio Input Port Left/Right Clock Input or Output

Word rate clock for the audio data on the SDIN pin. The frequency will be at the input sample rate (Fsi)

RMCK - Input Section Recovered Master Clock Output

Input section recovered master clock output. Will be at a frequency of 256x the input sample rate (Fsi).

$\overline{\text{LOCK}}$ - PLL Lock Indicator Output

$\overline{\text{LOCK}}$ low indicates that the PLL is locked. This is also a start-up option pin, and requires a pull-up or pull-down resistor.

Audio Output Interface:

SDOUT - Serial Audio Output Port Data Output

Audio data serial output pin. This is also a start-up option pin, and requires a pull-up or pull-down resistor.

OSCLK - Serial Audio Output Port Bit Clock Input or Output

Serial bit clock for audio data on the SDOUT pin.

OLRCK - Serial Audio Output Port Left/Right Clock Input or Output

Word rate clock for the audio data on the SDOUT pin. The frequency will be at the output sample rate (Fso).

AES3/SPDIF Transmitter Interface:

TXN, TXP - Differential Line Driver Outputs

Differential line driver outputs, transmitting AES3 type data. Drivers are pulled to low while the CS8420 is in the reset state.

TCBL - Transmit Channel Status Block Start

When operated as output, TCBL is high during the first sub-frame of a transmitted channel status block, and low at all other times. When operated as input, driving TCBL high for at least three OMCK clocks will cause the current transmitted sub-frame to be the start of a channel status block.

CUVEN - C, U and V bit Input Enable Mode Input

The CUVEN pin determines how the channel status data, user data and validity bit is input. When CUVEN is low, hardware mode 2A is selected, where the EMPH/V, COPY/C and ORIG/U pins are used to enter selected channel status data. When CUVEN is high, hardware 2B is selected, where the EMPH/V, COPY/C and ORIG/U pins are used to enter serial C, U and V data.

$\overline{\text{EMPH}}$/V - Pre-emphasis Indicator Input or V bit Input

In mode 2A $\overline{\text{EMPH}}$/V low sets the 3 $\overline{\text{EMPH}}$ channel status bits to indicate 50/15 µs pre-emphasis. $\overline{\text{EMPH}}$/V high sets the 3 $\overline{\text{EMPH}}$ bits to 000 indicating no pre-emphasis. In mode 2B $\overline{\text{EMPH}}$/V low sets the V bit to indicate valid audio. $\overline{\text{EMPH}}$/V high sets the V-bit to indicate non-valid audio.

COPY/C - COPY Channel Status bit Input or C bit Input

In mode 2A, the COPY/C pin determines the state of the COPY, PRO and L Channel Status bits in the outgoing AES3 type data stream (See Table 7). In mode 2B, COPY/C becomes the direct C bit input data pin.

ORIG/U - ORIG Channel Status bit Input or U bit Input

In mode 2A, the ORIG/U pin determines the state of the COPY, PRO and L Channel Status bits in the outgoing AES3 type data stream. (See Table 7). In mode 2B, ORIG/U becomes the direct U bit input data pin.

14.4 Hardware Mode 3 Description (Transceive Data Flow, with SRC)

Hardware Mode 3 data flow is shown in Figure 28. Audio data is input via the AES3 receiver, and rate converted. The audio data at the new rate is then output via the serial audio output port. Different audio data, synchronous to OMCK, may be input into the serial audio input port, and output via the AES3 transmitter.

The channel status data, user data and validity bit information are handled in 2 alternative modes: 3A and 3B, determined by a start-up resistor on the COPY pin. In mode 3A, the received PRO, COPY, ORIG, and $\overline{\text{AUDIO}}$ channel status bits are output on pins. The transmitted channel status bits are copied from the received channel status data, and the transmitted U and V bits are 0.

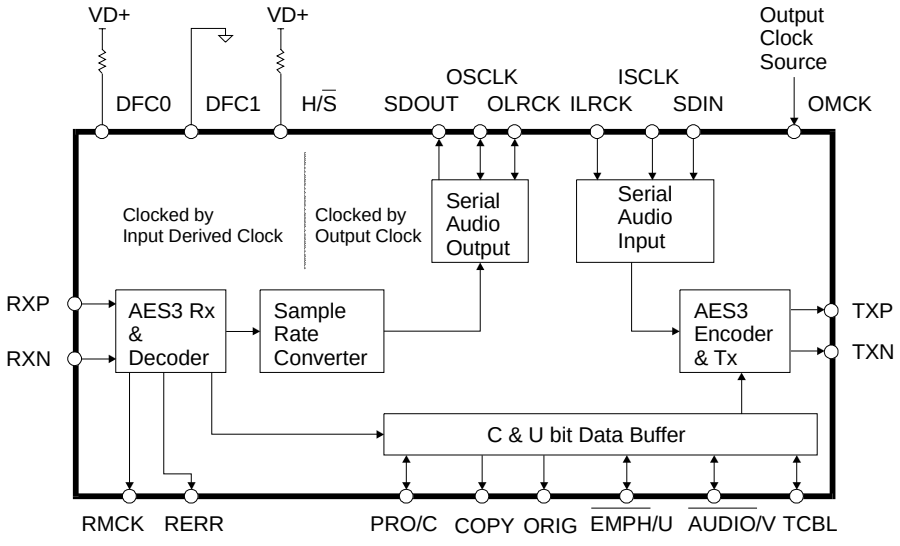
In mode 3B, only the COPY and ORIG pins are output, and reflect the received channel status data. The transmitted channel status bits, user data and validity bits are input serially via the PRO/C, $\overline{\text{EMPH}}$/U and $\overline{\text{AUDIO}}$/V pins. Figure 22 shows the timing requirements.

The serial audio input port is always a slave.

If a validity, parity, bi-phase or lock receiver error occurs, the current audio sample will be held.

Start-up options are shown in Table 10, and allow choice of the serial audio output port as a master or slave, whether TCBL is an input or an output, the serial audio ports formats and the source of the transmitted C, U and V data.

The following pages contain the detailed pin descriptions for hardware mode 3.



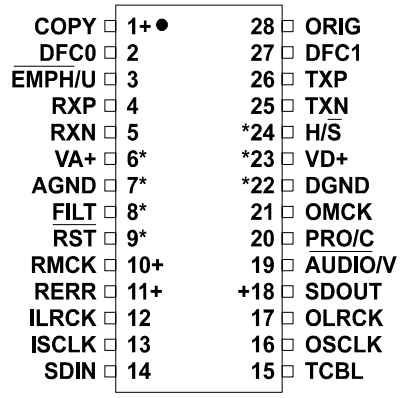
Power supply pins (VD+, VA+, DGND, AGND) & the reset pin ($\overline{\text{RST}}$) and the PLL filter pin (FILT) are omitted from this diagram. Please refer to the Typical Connection Diagram for hook-up details.

Figure 28. Hardware Mode 3 - Transceive Data Flow, with SRC

SDOUT	RMCK	RERR	ORIG	COPY	Function
LO	-	-	-	-	Serial Output Port is Slave
HI	-	-	-	-	Serial Output Port is Master
-	-	-	-	LO	Mode 3A: C transmitted data is copied from received data, U & V =0, received PRO, EMPH, AUDIO is visible
-	-	-	-	HI	Mode 3B: CUV transmitted data is input serially on pins, received PRO, EMPH and AUDIO is not visible
-	LO	LO	-	-	Serial Input & Output Format IF1&OF1
-	LO	HI	-	-	Serial Input & Output Format IF2&OF2
-	HI	LO	-	-	Serial Input & Output Format IF3&OF3
-	HI	HI	-	-	Serial Input & Output Format IF2&OF4
-	-	-	LO	-	TCBL is an input
-	-	-	HI	-	TCBL is an output

Table 10. Hardware Mode 3 Start-up Options

14.4.1 Pin Description - Hardware Mode 3



* Pins which remain the same function in all modes.
\+ Pins which require a pull up or pull down resistor to select the desired startup option.

Overall Device Control:

DFC0, DFC1 - Data Flow Control Inputs

DFC0 and DFC1 inputs determine the major data flow options available in hardware mode, according to Table 3.

OMCK - Output Section Master Clock Input

Output section master clock input. The frequency must be 256x the output sample rate (Fso).

Audio Input Interface:

SDIN - Serial Audio Input Port Data Input

Audio data serial input pin. This data will be transmitted out the AES3 port.

ISCLK - Serial Audio Input Port Bit Clock Input

Serial bit clock for audio data on the SDIN pin.

ILRCK - Serial Audio Input Port Left/Right Clock Input

Word rate clock for the audio data on the SDIN pin. The frequency will be at the output sample rate (Fso)

Audio Output Interface:

SDOUT - Serial Audio Output Port Data Output

Audio data serial output pin. This is also a start-up option pin, and requires a pull-up or pull-down resistor.

OSCLK - Serial Audio Output Port Bit Clock Input or Output

Serial bit clock for audio data on the SDOUT pin.

OLRCK - Serial Audio Output Port Left/Right Clock Input or Output

Word rate clock for the audio data on the SDOUT pin. The frequency will be at the output sample rate (Fso).

AES3/SPDIF Transmitter Interface:

TXN, TXP - Differential Line Driver Outputs

Differential line driver outputs, transmitting AES3 type data. Drivers are pulled to low while the CS8420 is in the reset state.

TCBL - Transmit Channel Status Block Start

When operated as output, TCBL is high during the first sub-frame of a transmitted channel status block, and low at all other times. When operated as input, driving TCBL high for at least three OMCK clocks will cause the current transmitted sub-frame to be the start of a channel status block.

AES3/SPDIF Receiver Interface:

RXP, RXN - Differential Line Receiver Inputs

Differential line receiver inputs, carrying AES3 type data.

RMCK - Input Section Recovered Master Clock Output

Input section recovered master clock output. Will be at a frequency of 256x the input sample rate (Fsi). This is also a start-up option pin, and requires a pull-up or pull-down resistor.

RERR - Receiver Error Indicator Output

When high, indicates a problem with the operation of the AES3 receiver. The status of this pin is updated once per sub-frame of incoming AES3 data. Conditions that cause RERR to go high are: parity error, and bi-phase coding error, as well as loss of lock in the PLL. This is also a start-up option pin, and requires a pull-up or pull-down resistor.

$\overline{\text{EMPH}}$/U - Pre-emphasis Indicator Output or U-bit Data Input

The $\overline{\text{EMPH}}$/U pin either reflects the state of the $\overline{\text{EMPH}}$ channel status bits in the incoming AES3 type data stream, or is the serial U-bit input for the AES3 type transmitted data, clocked by OLRCK. If indicating emphasis $\overline{\text{EMPH}}$/U is low when the incoming data indicates 50/15 µs pre-emphasis and high otherwise.

COPY - Copy Channel Status bit Output

The COPY pin reflects the state of the COPY Channel Status bit in the incoming AES3 type data stream. This is also a start-up option pin, and requires a pull-up or pull-down resistor.

ORIG - Original Channel Status Output

SCMS generation indicator. This is decoded from the incoming category code and the L bit. A low output indicates that the audio data stream is 1st generation or higher. A high indicates that the audio data stream is original. This is also a start-up option pin, and requires a pull-up or pull-down resistor.

PRO/C - Professional Channel Status bit Output or C-bit Data Input

The PRO/C pin either reflects the state of the Professional/Consumer Channel Status bit in the incoming AES3 type data stream, or is the serial C-bit input for the AES3 type transmitted data, clocked by OLRCK.

$\overline{\text{AUDIO}}$/V - Audio Channel Status bit Output or V-bit Data Input

The $\overline{\text{AUDIO}}$/V pin either reflects the state of the audio/non audio Channel Status bit in the incoming AES3 type data stream, or is the V-bit data input for the AES3 type transmitted data stream, clocked by OLRCK.

14.5 Hardware Mode 4 Description (Transceive Data Flow, No SRC)

Hardware Mode 4 data flow is shown in Figure 29. Audio data is input via the AES3 receiver, and routed to the serial audio output port. Different audio data synchronous to RMCK may be input into the serial audio input port, and output via the AES3 transmitter.

The channel status data, user data and validity bit information are handled in 2 alternative modes: 4A and 4B, determined by a start-up resistor on the COPY pin. In mode 4A, the received PRO, COPY, ORIG, $\overline{\text{EMPH}}$, and $\overline{\text{AUDIO}}$ channel status bits are output on pins. The transmitted channel status bits are copied from the received channel status data, and the transmitted U and V bits are 0.

In mode 4B, only the COPY and ORIG pins are output, and reflect the received channel status data. The transmitted channel status bits, user data and validity bits are input serially via the PRO/C, $\overline{\text{EMPH}}$/U and $\overline{\text{AUDIO}}$/V pins. Figure 22 shows the timing requirements.

The APMS pin allows the serial audio input port to be set to master or slave.

If a validity, parity, bi-phase or lock receiver error occurs, the current audio sample is passed unmodified to the serial audio output port.

Start-up options are shown in Table 11, and allow choice of the serial audio output port as a master or slave, whether TCBL is an input or an output, and the audio serial ports formats and the source of the transmitted C, U and V data.

The following pages contain the detailed pin descriptions for hardware mode 4.

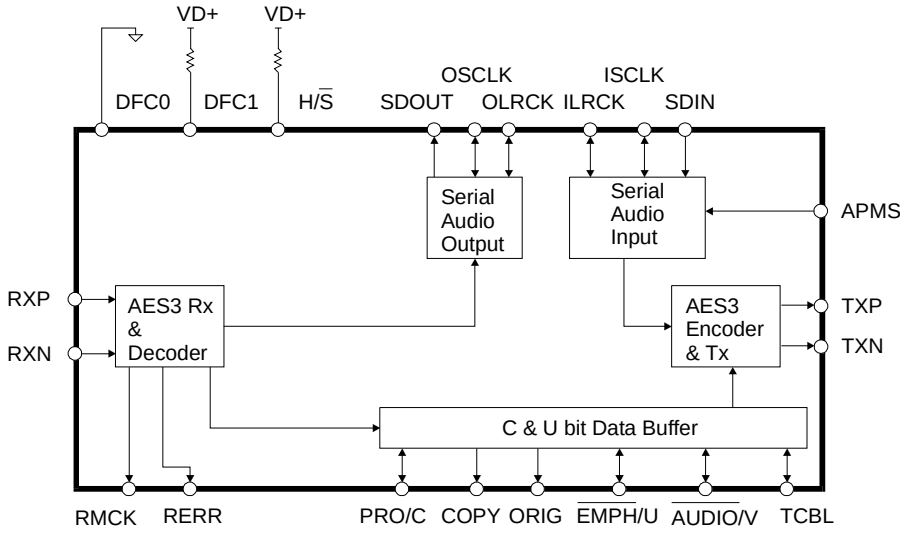
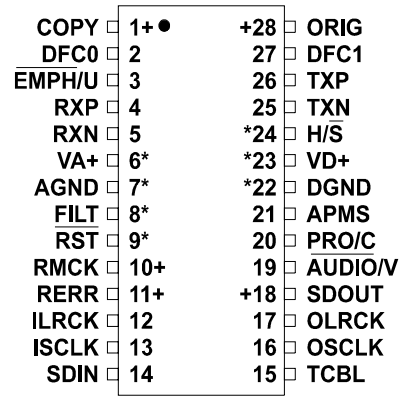


Figure 29. Hardware Mode 4 - Transceive Data Flow, without SRC

SDOUT	RMCK	RERR	ORIG	COPY	Function
LO	-	-	-	-	Serial Output Port is Slave
HI	-	-	-	-	Serial Output Port is Master
-	-	-	-	LO	Mode 4A: C transmitted data is copied from received data, U & V =0, received PRO, EMPH, AUDIO is visible
-	-	-	-	HI	Mode 4B: CUV transmitted data is input serially on pins, received PRO, EMPH and AUDIO is not visible
-	LO	LO	-	-	Serial Input & Output Format IF1&OF1
-	LO	HI	-	-	Serial Input & Output Format IF2&OF2
-	HI	LO	-	-	Serial Input & Output Format IF3&OF3
-	HI	HI	-	-	Serial Input & Output Format IF1&OF5
-	-	-	LO	-	TCBL is an input
-	-	-	HI	-	TCBL is an output

Table 11. Hardware Mode 4 Start-up Options

14.5.1 Pin Description - Hardware Mode 4



* Pins which remain the same function in all modes.
\+ Pins which require a pull up or pull down resistor to select the desired startup option.

Overall Device Control:

DFC0, DFC1 - Data Flow Control Inputs

DFC0 and DFC1 inputs determine the major data flow options available in hardware mode, according to Table 3.

Audio Input Interface:

SDIN - Serial Audio Input Port Data Input

Audio data serial input pin. This data will be transmitted out the AES3 port.

ISCLK - Serial Audio Input Port Bit Clock Input or Output

Serial bit clock for audio data on the SDIN pin.

ILRCK - Serial Audio Input Port Left/Right Clock Input or Output

Word rate clock for the audio data on the SDIN pin. The frequency will be at the input sample rate (Fsi)

APMS - Serial Audio Input Port Master or Slave

APMS should be connected to VD+ to set serial audio input port as a master, or connected to DGND to set the port as a slave.

Audio Output Interface:

SDOUT - Serial Audio Output Port Data Output

Audio data serial output pin. This is also a start-up option pin, and requires a pull-up or pull-down resistor.

OSCLK - Serial Audio Output Port Bit Clock Input or Output

Serial bit clock for audio data on the SDOUT pin.

OLRCK - Serial Audio Output Port Left/Right Clock Input or Output

Word rate clock for the audio data on the SDOUT pin. The frequency will be at the input sample rate (Fsi).

AES3/SPDIF Transmitter Interface:

TXN, TXP - Differential Line Driver Outputs

Differential line driver outputs, transmitting AES3 type data. Drivers are pulled to low while the CS8420 is in the reset state.

TCBL - Transmit Channel Status Block Start

When operated as output, TCBL is high during the first sub-frame of a transmitted channel status block, and low at all other times. When operated as input, driving TCBL high for at least three RMCK clocks will cause the current transmitted sub-frame to be the start of a channel status block.

AES3/SPDIF Receiver Interface:

RXP, RXN - Differential Line Receiver Inputs

Differential line receiver inputs, carrying AES3 type data.

RMCK - Input Section Recovered Master Clock Output

Input section recovered master clock output. Will be at a frequency of 256x the input sample rate (Fsi). This is also a start-up option pin, and requires a pull-up or pull-down resistor.

RERR - Receiver Error Indicator Output

When high, indicates a problem with the operation of the AES3 receiver. The status of this pin is updated once per sub-frame of incoming AES3 data. Conditions that cause RERR to go high are: parity error, and bi-phase coding error, as well as loss of lock in the PLL. This is also a start-up option pin, and requires a pull-up or pull-down resistor.

$\overline{\text{EMPH}}$/U - Pre-emphasis Indicator Output or U-bit Data Input

The $\overline{\text{EMPH}}$/U pin either reflects the state of the $\overline{\text{EMPH}}$ channel status bit in the incoming AES3 type data stream, or is the serial U-bit input for the AES3 type transmitted data, clocked by OLRCK. If indicating emphasis $\overline{\text{EMPH}}$/U is high when the incoming data indicates 50/15 µs pre-emphasis and low otherwise.

COPY - Copy Channel Status bit Output

The COPY pin reflects the state of the COPY Channel Status bit in the incoming AES3 type data stream. This is also a start-up option pin, and requires a pull-up or pull-down resistor.

ORIG - Original Channel Status Output

SCMS generation indicator. This is decoded from the incoming category code and the L bit. A low output indicates that the audio data stream is 1st generation or higher. A high indicates that the audio data stream is original. This is also a start-up option pin, and requires a pull-up or pull-down resistor.

PRO/C - Professional Channel Status bit Output or C-bit Data Input

The PRO/C pin either reflects the state of the Professional/Consumer Channel Status bit in the incoming AES3 type data stream, or is the serial C-bit input for the AES3 type transmitted data, clocked by OLRCK.

$\overline{\text{AUDIO}}$/V - Audio Channel Status bit Output or V-bit Data Input

The $\overline{\text{AUDIO}}$/V pin either reflects the state of the audio/non audio Channel Status bit in the incoming AES3 type data stream, or is the V-bit data input for the AES3 type transmitted data stream, clocked by OLRCK.

14.6 Hardware Mode 5 Description (AES3 Receiver Only)

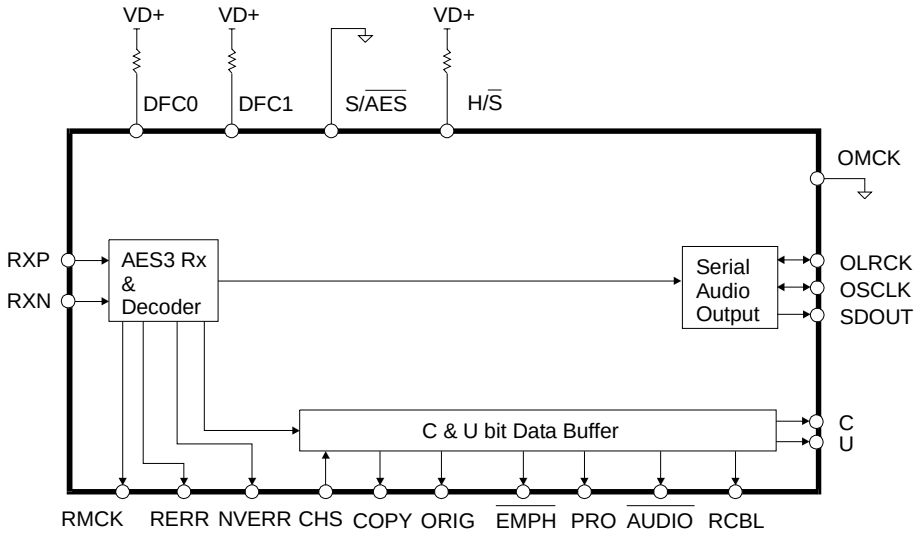
Hardware Mode 5 data flow is shown in Figure 30. Audio data is input via the AES3 receiver, and routed to the serial audio output port. The PRO, COPY, ORIG, $\overline{\text{EMPH}}$, and $\overline{\text{AUDIO}}$ channel status bits are output on pins. The decoded C and U bits are also output, clocked by both edges of OLRCK (master mode only, see Figure 21).

If a validity, parity, bi-phase or lock receiver error occurs, the current audio sample is passed unmodified to the serial audio output port.

Start-up options are shown in Table 12, and allow choice of the serial audio output port as a master or slave, and the serial audio port format. The following pages contain the detailed pin descriptions for hardware mode 5.

SDOUT	ORIG	$\overline{\text{EMPH}}$	Function
LO	-	-	Serial Output Port is Slave
HI	-	-	Serial Output Port is Master
-	LO	LO	Serial Output Format OF1
-	LO	HI	Serial Output Format OF2
-	HI	LO	Serial Output Format OF3
-	HI	HI	Serial Output Format OF5

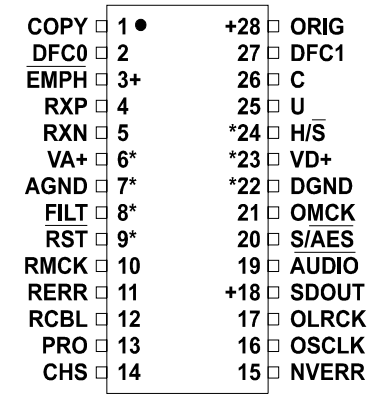
Table 12. Hardware Mode 5 Start-up Options



Power supply pins (VD+, VA+, DGND, AGND) & the reset pin ($\overline{\text{RST}}$) and the PLL filter pin (FILT) are omitted from this diagram. Please refer to the Typical Connection Diagram for hook-up details.

Figure 30. Hardware Mode 5 - AES3 Receiver Only

14.6.1 Pin Description - Hardware Mode 5



* Pins which remain the same function in all modes.
\+ Pins which require a pull up or pull down resistor to select the desired startup option.

Overall Device Control:

DFC0, DFC1 - Data Flow Control Inputs

DFC0 and DFC1 inputs determine the major data flow options available in hardware mode, according to Table 3.

S/$\overline{\text{AES}}$ - Serial Audio or AES3 Input Select

S/$\overline{\text{AES}}$ is connected to DGND in hardware mode 5, in order to select the AES3 input.

OMCK - Output Section Master Clock Input

Output section master clock input. This pin is not used in this mode and should be connected to DGND.

Audio Output Interface:

SDOUT - Serial Audio Output Port Data Output

Audio data serial output pin. This is also a start-up option pin, and requires a pull-up or pull-down resistor.

OSCLK - Serial Audio Output Port Bit Clock Input or Output

Serial bit clock for audio data on the SDOUT pin.

OLRCK - Serial Audio Output Port Left/Right Clock Input or Output

Word rate clock for the audio data on the SDOUT pin. The frequency will be at the input sample rate (Fsi).

AES3/SPDIF Receiver Interface:

RXP, RXN - Differential Line Receiver Inputs

Differential line receiver inputs, carrying AES3 type data.

RMCK - Input Section Recovered Master Clock Output

Input section recovered master clock output. Will be at a frequency of 256x the input sample rate (Fsi).

RERR - Receiver Error Indicator

When high, indicates a problem with the operation of the AES3 receiver. The status of this pin is updated once per sub-frame of incoming AES3 data. Conditions that cause RERR to go high are: validity, parity error, and bi-phase coding error, as well as loss of lock in the PLL.

NVERR - No Validity Receiver Error Indicator

When high, indicates a problem with the operation of the AES3 receiver. The status of this pin is updated once per frame of incoming AES3 data. Conditions that cause NVERR to go high are: parity error, and bi-phase coding error, as well as loss of lock in the PLL.

$\overline{\text{EMPH}}$ - Pre-emphasis Indicator Output

$\overline{\text{EMPH}}$ is low when the incoming AES3 data indicates the presence of 50/15 µs pre-emphasis. When the AES3 data indicates the absence of pre-emphasis or the presence of non 50/15 µs pre-emphasis $\overline{\text{EMPH}}$ is high. This is also a start-up option pin, and requires a pull-up or pull-down resistor.

COPY - Copy Channel Status bit Output

The COPY pin reflects the state of the COPY Channel Status bit in the incoming AES3 type data stream.

ORIG - Original Channel Status Output

SCMS generation indicator. This is decoded from the incoming category code and the L bit. A low output indicates that the audio data stream is 1st generation or higher. A high indicates that the audio data stream is original. This is also a start-up option pin, and requires a pull-up or pull-down resistor.

PRO - Professional Channel Status bit Output

The PRO pin reflects the state of the Professional/Consumer Channel Status bit in the incoming AES3 type data stream.

$\overline{\text{AUDIO}}$ - Audio Channel Status bit Output

The $\overline{\text{AUDIO}}$ pin reflects the state of the audio/non audio Channel Status bit in the incoming AES3 type data stream.

RCBL - Receiver Channel Status Block Output

RCBL indicates the beginning of a received channel status block. RCBL goes high 2 frames after the reception of a Z preamble, remains high for 16 frames while COPY, ORIG, AUDIO, $\overline{\text{EMPH}}$ and PRO are updated, and returns low for the remainder of the block. RCBL changes on rising edges of RMCK.

CHS - Channel Select Input

Selects which sub-frame's channel status data is output on the $\overline{\text{EMPH}}$, COPY, ORIG, PRO and $\overline{\text{AUDIO}}$ pins. Channel A is selected when CHS is low, channel B is selected when CHS is high.

U - User Data Output

The U pin outputs user data from the AES3 receiver, clocked by rising and falling edges of OLRCK.

C - Channel Status Data Output

The C pin outputs channel status data from the AES3 receiver, clocked by rising and falling edges of OLRCK.

14.7 Hardware Mode 6 Description (AES3 Transmitter Only)

Hardware Mode 6 data flow is shown in Figure 31. Audio data is input via the serial audio input port and routed to the AES3 transmitter.

The transmitted channel status, user and validity data may be input in 2 alternative methods, determined by the state of the CEN pin. Mode 6A is selected when the CEN pin is low. In mode 6A, the user data and validity bit are input via the U and V pins, clocked by both edges of ILRCK. The channel status data is derived from the state of the COPY/C, ORIG, $\overline{\text{EMPH}}$, and $\overline{\text{AUDIO}}$ pins. Table 13 shows how the COPY/C and ORIG pins map to channel status bits. In consumer mode, the transmitted category code shall be set to Sample Rate Converter (0101100).

Mode 6B is selected when the CEN pin is high. In mode 6B, the channel status, user data and validity bit are input serially via the COPY/C, U and V pins. These pins are clocked by both edges of ILRCK (if the port is in master mode). Figure 22 shows the timing requirements.

COPY/C	ORIG	Function
0	0	PRO=0, COPY=0, L=0
0	1	PRO=0, COPY=0, L=1
1	0	PRO=0, COPY=1, L=0
1	1	PRO=1

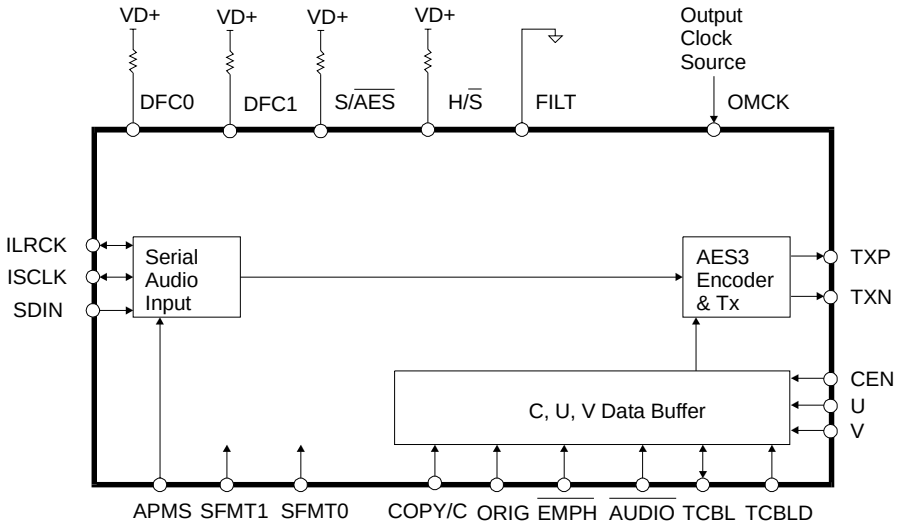
Table 13. HW 6C COPY/C and ORIG pin function

The channel status block pin (TCBL) may be an input or an output, determined by the state of the TCBLD pin. The serial audio input port data format is selected as shown in Table 14, and may be set to master or slave by the state of the APMS input pin.

SFMT1	SFMT0	Function
0	0	Serial Input Format IF1
0	1	Serial Input Format IF2
1	0	Serial Input Format IF3
1	1	Serial Input Format IF4

Table 14. HW 6 Serial Audio Port Format Selection

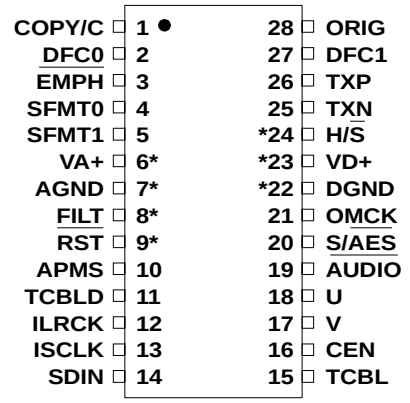
The following pages contain the detailed pin descriptions for hardware mode 6.



Power supply pins (VD+, VA+, DGND, AGND) & the reset pin ($\overline{\text{RST}}$)
are omitted from this diagram. Please refer to the Typical Connection Diagram for hook-up details.

Figure 31. Hardware Mode 6 - AES3 Transmitter Only

14.7.1 Pin Description - Hardware Mode 6



* Pins which remain the same function in all modes.

Overall Device Control:

DFC0, DFC1 - Data Flow Control Inputs

DFC0 and DFC1 inputs determine the major data flow options available in hardware mode, according to Table 3.

S/AES - Serial Audio or AES3 Input Select

S/AES is connected to VD+ in hardware mode 6, in order to select the serial audio input.

SFMT0, SFMT1 - Serial Audio Input Port Data Format Select Inputs

SFMT0 and SFMT1 select the serial audio input port format. See Table 14.

OMCK - Output Section Master Clock Input

Output section master clock input. The frequency must be 256x the output sample rate (Fso).

Audio Input Interface:

SDIN - Serial Audio Input Port Data Input

Audio data serial input pin.

ISCLK - Serial Audio Input Port Bit Clock Input or Output

Serial bit clock for audio data on the SDIN pin.

ILRCK - Serial Audio Input Port Left/Right Clock Input or Output

Word rate clock for the audio data on the SDIN pin.

APMS - Serial Audio Input Port Master or Slave

APMS should be connected to VD+ to set serial audio input port as a master, or connected to DGND to set the port as a slave.

AES3/SPDIF Transmitter Interface:

TXN, TXP - Differential Line Driver Outputs

Differential line driver outputs, transmitting AES3 type data. Drivers are pulled to low while the CS8420 is in the reset state.

TCBL - Transmit Channel Status Block Start

When operated as output, TCBL is high during the first sub-frame of a transmitted channel status block, and low at all other times. When operated as input, driving TCBL high for at least three OMCK clocks will cause the current transmitted sub-frame to be the start of a channel status block.

TCBLD - Transmit Channel Status Block Direction Input

Connect TCBLD to VD+ to set TCBL as an output. Connect TCBLD to DGND to set TCBL as an input.

$\overline{\text{EMPH}}$ - Pre-emphasis Indicator Input

In mode 6B, $\overline{\text{EMPH}}$ pin low sets the 3 $\overline{\text{EMPH}}$ channel status bits to indicate 50/15 µs pre-emphasis. If $\overline{\text{EMPH}}$ is high the 3 $\overline{\text{EMPH}}$ channel status bits are set to 000 indicating no pre-emphasis.

COPY/C - COPY Channel Status bit Input or C bit Input

In mode 6B, the COPY/C pin determines the state of the COPY, PRO and L Channel Status bits in the outgoing AES3 type data stream (See Table 13). In mode 6A, the COPY/C pin becomes the direct C bit input data pin.

ORIG - ORIG Channel Status bit Input

In mode 6B, the ORIG pin determines the state of the COPY, PRO and L Channel Status bits in the outgoing AES3 type data stream. See Table 13.

$\overline{\text{AUDIO}}$ - Audio Channel Status bit Input

In mode 6B, the $\overline{\text{AUDIO}}$ pin determines the state of the audio/non audio Channel Status bit in the outgoing AES3 type data stream.

V - Validity bit Input

In modes 6A and 6B, the V pin input determines the state of the validity bit in the outgoing AES3 transmitted data. This pin is sampled on both edges of the ILRCK.

U - User Data bit Input

In modes 6A and 6B, the U pin input determines the state of the user data bit in the outgoing AES3 transmitted data. This pin is sampled on both edges of the ILRCK.

CEN - C bit Input Enable Mode Input

The CEN pin determines how the channel status data bits are input. When CEN is low, hardware mode 6A is selected, where the COPY/C, ORIG, $\overline{\text{EMPH}}$ and $\overline{\text{AUDIO}}$ pins are used to enter selected channel status data. When CEN is high, hardware mode 6B is selected, where the COPY/C pin is used to enter serial channel status data.

15. APPENDIX A: EXTERNAL AES3/SPDIF/IEC60958 TRANSMITTER AND RECEIVER COMPONENTS

This section details the external components required to interface the AES3 transmitter and receiver to cables and fiber-optic components.

15.1 AES3 Transmitter External Components

The output drivers on the CS8420 are designed to drive both the professional and consumer interfaces. The AES3 specification for professional/broadcast use calls for a 110 Ω source impedance and a balanced drive capability. Since the transmitter output impedance is very low, a 110 Ω resistor should be placed in series with one of the transmit pins. The specifications call for a balanced output drive of 2-7 volts peak-to-peak into a 110 Ω load with no cable attached. Using the circuit in Figure 32, the output of the transformer is short-circuit protected, has the proper source impedance, and provides a 5 volt peak-to-peak signal into a 110 Ω load. Lastly, the two output pins should be attached to an XLR connector with male pins and a female shell, and with pin 1 of the connector grounded.

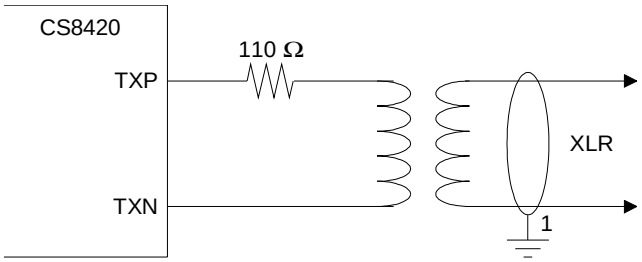


Figure 32. Professional Output Circuit

In the case of consumer use, the IEC60958 specifications call for an unbalanced drive circuit with an output impedance of 75 Ω and a output drive level of 0.5 V peak-to-peak ±20% when measured across a 75 Ω load using no cable. The circuit shown in Figure 33 only uses the TXP pin and provides the proper output impedance and drive level using standard 1% resistors. The connector for a consumer application would be an RCA phono socket. This circuit is also short circuit protected.

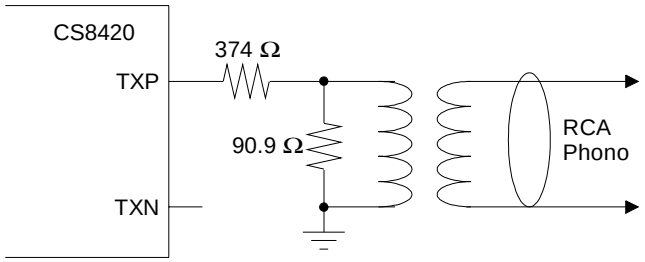


Figure 33. Consumer Output Circuit

The TXP pin may be used to drive TTL or CMOS gates as shown in Figure 34. This circuit may be used for optical connectors for digital audio since they usually have TTL or CMOS compatible inputs. This circuit is also useful when driving multiple digital audio outputs since RS422 line drivers have TTL compatible inputs.

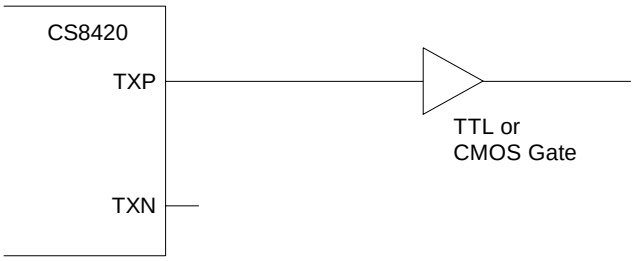


Figure 34. TTL/CMOS Output Circuit

15.2 AES3 Receiver External Components

The CS8420 AES3 receiver is designed to accept both the professional and consumer interfaces. The digital audio specifications for professional use call for a balanced receiver, using XLR connectors, with 110 Ω ±20% impedance. The XLR connector on the receiver should have female pins with a male shell. Since the receiver has a very high input impedance, a 110 Ω resistor should be placed across the receiver terminals to match the line impedance, as shown in Figure 35. Although transformers are not required by the AES, they are, however, strongly recommended.

If some isolation is desired without the use of transformers, a 0.01 µF capacitor should be placed in se-

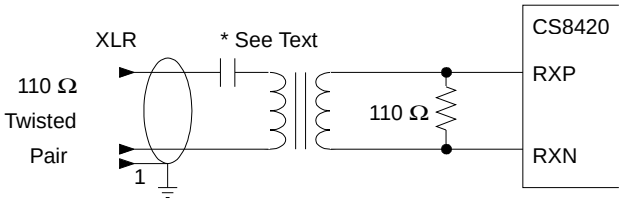


Figure 35. Professional Input Circuit

ries with each input pin (RXP and RXN) as shown in Figure 36. However, if a transformer is not used, high frequency energy could be coupled into the receiver, causing degradation in analog performance.

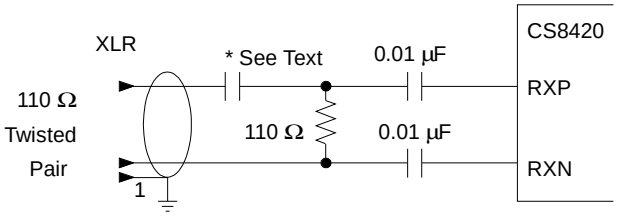


Figure 36. Transformerless Professional Input Circuit

Figures 35 and 36 show an optional DC blocking capacitor (0.1 µF to 0.47 µF) in series with the cable input. This improves the robustness of the receiver, preventing the saturation of the transformer, or any DC current flow, if a DC voltage is present on the cable.

In the configuration of systems, it is important to avoid ground loops and DC current flowing down the shield of the cable that could result when boxes with different ground potentials are connected. Generally, it is good practice to ground the shield to the chassis of the transmitting unit, and connect the shield through a capacitor to chassis ground at the receiver. However, in some cases it is advantageous to have the ground of two boxes held to the same potential, and the cable shield might be depended upon to make that electrical connection. Generally, it may be a good idea to provide the option of grounding or capacitively coupling the shield to the chassis.

In the case of the consumer interface, the standards call for an unbalanced circuit having a receiver impedance of 75 Ω ±5%. The connector for the consumer interface is an RCA phono socket. The receiver circuit for the consumer interface is shown in Figure 37.

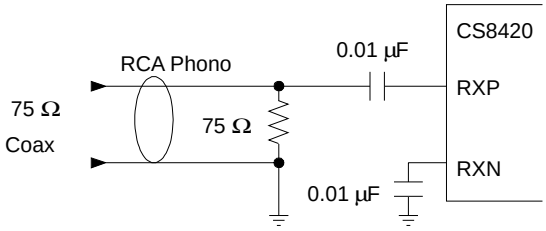


Figure 37. Consumer Input Circuit

The circuit shown in Figure 38 may be used when external RS422 receivers, optical receivers or other TTL/CMOS logic outputs drive the CS8420 receiver section.

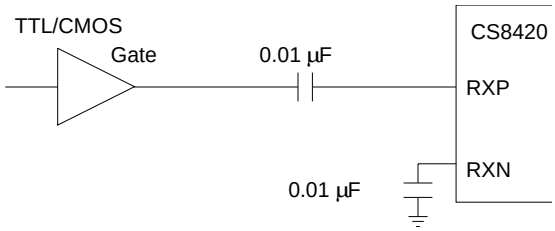


Figure 38. TTL/CMOS Input Circuit

15.3 Isolating Transformer Requirements

The transformer should be capable of operating from 1.5 to 14 MHz, which is equivalent to an audio data rate of 25 kHz to 108 kHz after bi-phase mark encoding. Transformers provide isolation from ground loops, 60Hz noise, and common mode noise and interference. One of the important considerations when choosing transformers is minimizing shunt capacitance between primary and secondary windings. The higher the shunt capacitance, the lower the isolation between primary and secondary, and the more coupling of high frequency energy. This energy appears in the form of common mode noise on the receive side ground and has the potential to degrade analog performance. Therefore, for best performance, shielded transformers optimized for minimum shunt capacitance should be used. See Application Note 134 for a selection of manufacturers and their part numbers.

16. APPENDIX B: CHANNEL STATUS AND USER DATA BUFFER MANAGEMENT

The CS8420 has a comprehensive channel status (C) and user (U) data buffering scheme, which allows automatic management of channel status blocks and user data. Alternatively, sufficient control and access is provided to allow the user to completely manage the C and U data via the control port.

16.1 AES3 Channel Status(C) Bit Management

The CS8420 contains sufficient RAM to store a full block of C data for both A and B channels (192x2 = 384 bits), and also 384 bits of U information. The user may read from or write to these RAMs via the control port.

Unlike the audio data, it is not possible to 'sample-rate' convert the C bits. This is because specific meanings are associated with fixed-length data patterns, which should not be altered. Since the output data rate of the CS8420 will differ from the input rate when sample-rate conversion is done, it is not feasible to directly transfer incoming C data to the output. The CS8420 manages the flow of channel status data at the block level, meaning that entire blocks of channel status information are buffered at the input, synchronized to the output timebase, and then transmitted. The buffering scheme involves a cascade of 3 block-sized buffers, named D,E and F, as shown in Figure 39. The MSB of each byte represents the first bit in the serial C data stream. For example, the MSB of byte 0 (which is at control port address 32) is the consumer/professional bit for channel status block A.

The first buffer, D, accepts incoming C data from the AES receiver. The 2nd buffer, E, accepts entire blocks of data from the D buffer. The E buffer is also accessible from the control port, allowing read and writing of the C data. The 3rd buffer (F) is used as the source of C data for the AES3 transmitter. The F buffer accepts block transfers from the E buffer.

If the input rate is slower than the output rate (so that in a given time interval, more channel status blocks are transmitted than received), some buffered C blocks will be transmitted multiple times. If the input rate is faster than the output rate, some will not be transmitted at all. This is illustrated in Figure 40). In this manner, channel status block integrity is maintained. If the transmitted sample count bits are important in the application, then they will need to be updated via the control port by the microcontroller for every outgoing block.

16.1.1 Manually accessing the E buffer

The user can monitor the data being transferred by reading the E buffer, which is mapped into the register space of the CS8420, via the control port. The user can modify the data to be transmitted by writing to the E buffer.

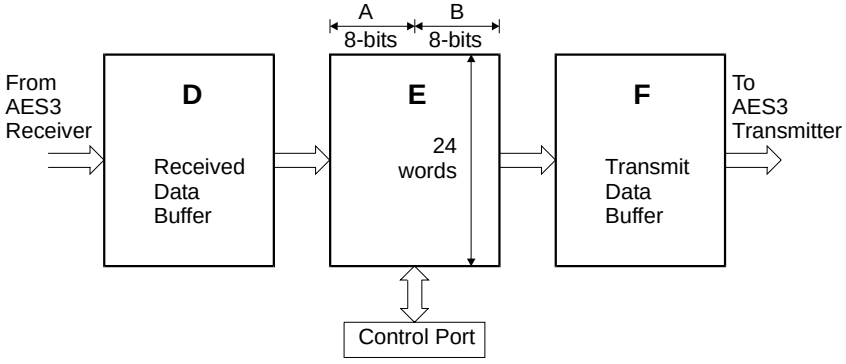


Figure 39. Channel Status Data Buffer Structure

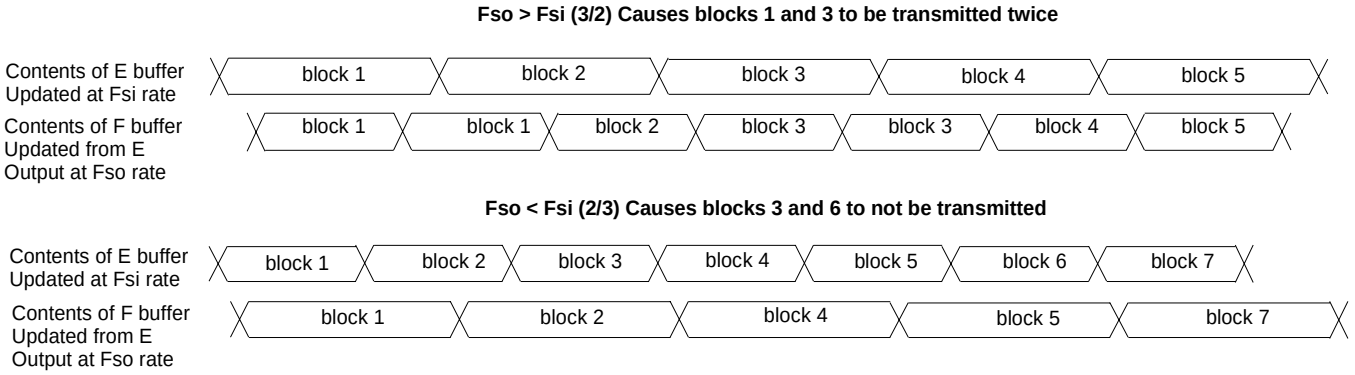


Figure 40. Channel Status Block Handling When Fso is Not Equal to Fsi

The user can configure the interrupt enable register to cause interrupts to occur whenever "D to E" or "E to F" buffer transfers occur. This allows determination of the allowable time periods to interact with the E buffer.

Also provided are "D to E" and "E to F" inhibit bits. The associated buffer transfer is disabled whenever the user sets these bits. These may be used whenever "long" control port interactions are occurring. They can also be used to align the behavior of the buffers with the selected audio data flow. For example, if the audio data flow is serial port in to AES3 out, then it is necessary to inhibit "D toE" transfers, since these would overwrite the desired transmit C data with invalid data.

Flowcharts for reading and writing to the E buffer are shown in Figures 41 and 42. For reading, since a D to E interrupt just occurred, then there a substantial time interval until the next D to E transfer (approximately 192 frames worth of time). This is usually plenty of time to access the E data without having to inhibit the next transfer.

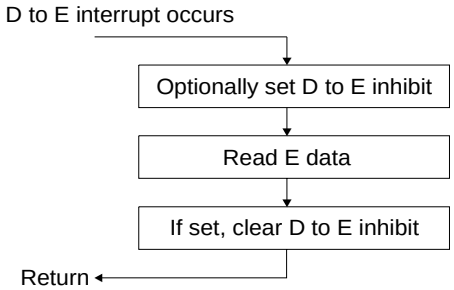


Figure 41. Flowchart for Reading the E Buffer

For writing, the sequence starts after a E to F transfer, which is based on the output timebase. Since a D to E transfer could occur at any time (this is based on the input timebase), then it is important to inhibit D to E transfers while writing to the E buffer until all writes are complete. Then wait until the next E to F transfer occurs before enabling D to E transfers. This ensures that the data written to the E buffer actually gets transmitted and not overwritten by a D to E transfer.

If the channel status block to transmit indicates PRO mode, then the CRCC byte is automatically calculated by the CS8420, and does not have to be written into the last byte of the block by the host microcontroller.

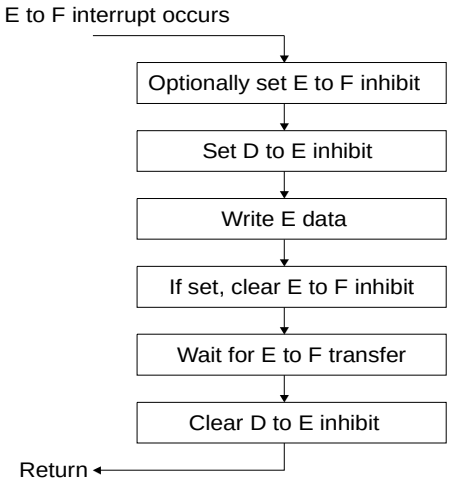


Figure 42. Flowchart for Writing the E Buffer

16.1.2 Reserving the first 5 bytes in the E buffer

D to E buffer transfers periodically overwrite the data stored in the E buffer. This can be a problem for users who want to transmit certain channel status settings which are different from the incoming settings. In this case, the user would have to superimpose his settings on the E buffer after every D to E overwrite.

To avoid this problem, the CS8420 has the capability of reserving the first 5 bytes of the E buffer for user writes only. When this capability is in use, internal D to E buffer transfers will NOT affect the first 5 bytes of the E buffer. Therefore, the user can set values in these first 5 E bytes once, and the settings will persist until the next user change. This mode is enabled via the Channel Status Data Buffer Control register.

16.1.3 Serial Copy Management System (SCMS)

In software mode, the CS8420 allows read/modify/write access to all the channel status bits. For consumer mode SCMS compliance, the host microcontroller needs to read and manipulate the Category Code, Copy bit and L bit appropriately.

In hardware mode, the SCMS protocol can be followed by either using the COPY and ORIG input pins, or by using the C bit serial input pin. These options are documented in the hardware mode section of this data sheet (starting on14 page 49)

16.1.4 Channel Status Data E Buffer Access

The E buffer is organized as 24 x 16-bit words. For each word the MS Byte is the A channel data, and the LS Byte is the B channel data (see Figure 39).

There are two methods of accessing this memory, known as one byte mode and two byte mode. The desired mode is selected via a control register bit.

16.1.5 One Byte mode

In many applications, the channel status blocks for the A and B channels will be identical. In this situation, if the user reads a byte from one of the channel's blocks, the corresponding byte for the other channel will be the same. Similarly, if the user wrote a byte to one channel's block, it would be necessary to write the same byte to the other block. One byte mode takes advantage of the often identical nature of A and B channel status data.

When reading data in one byte mode, a single byte is returned, which can be from channel A or B data, depending on a register control bit. If a write is being done, the CS8420 expects a single byte to be input to its control port. This byte will be written to both the A and B locations in the addressed word.

One byte mode saves the user substantial control port access time, as it effectively accesses 2 bytes worth of information in 1 byte's worth of access time. If the control port's auto-increment addressing is used in combination with this mode, multi-byte accesses such as full-block reads or writes can be done especially efficiently.

16.1.6 Two Byte mode

There are those applications in which the A and B channel status blocks will not be the same, and the user is interested in accessing both blocks. In these situations, two byte mode should be used to access the E buffer.

In this mode, a read will cause the CS8420 to output two bytes from its control port. The first byte out will represent the A channel status data, and the 2nd byte will represent the B channel status data. Writing is similar, in that two bytes must now be input to the CS8420's control port. The A channel status data is first, B channel status data second.

16.2 AES3 User (U) Bit Management

The CS8420 U bit manager has four operating modes:

Mode 1. Transmit all zeros.

Mode 2. Block mode.

Mode 3. Reserved

Mode 4. IEC Consumer B.

16.2.1 Mode 1: Transmit All Zeros

Mode 1 causes only zeros to be transmitted in the output U data, regardless of E buffer contents or U data embedded in an input AES3 data stream. This mode is intended for the user who does not want to transceive U data, and simply wants the output U channel to contain no data.

16.2.2 Mode 2: Block Mode

Mode 2 is very similar to the scheme used to control the C bits. Entire blocks of U data are buffered from input to output, using a cascade of 3 block-sized RAMs to perform the buffering. The user has access to the second of these 3 buffers, denoted the E buffer, via the control port. Block mode is designed for use in AES3 in, AES3 out situations in which input U data is decoded using a microcontroller via the control port. It is also the only mode in which the user can merge his own U data into the transmitted AES3 data stream.

The U buffer access only operates in two byte mode, since there is no concept of A and B blocks for user data. The arrangement of the data in the each byte is that the MSB is the first received bit and is the first transmitted bit. The first byte read is the first byte received, and the first byte sent is the first byte transmitted.

16.2.3 IEC60958 Recommended U Data Format For Consumer Applications

Modes (3) and (4) are intended for use in AES3 in, AES3 out situations, in which the input U data is formatted as recommended in the “IEC60958 Digital Audio Interface, part 3: Consumer applications” document.

In this format, “messages” are formed in the U data from Information Units or IUs. An IU is 8 bits long, and the MSB is always 1, and is called the start bit, or 'P' bit. The remaining 7 bits are called Q R S T U V & W, and carry the desired data.

A “message” consists of 3 to 129 IUs. Multiple IUs are considered to be in the same message if they are separated by zero to eight 0s, denoted here as filler. A filler sequence of nine or more 0s indicates an inter-message gap. The desired information is normally carried in the sequence of corresponding bits in the IUs. For example, the sequential Q bits from each IU make up the Q sub-code data that is used to indicate Compact Disk track information. This data is automatically extracted from the received IEC60958 stream, and is presented in the control port register map space.

Where incoming U data is coded in the above format, and needs to be re-transmitted, the data transfer cannot be done using shift registers, because of the different Fsi and Fso sampling clocks. Instead, input data must be buffered in a FIFO structure, and then read out by the AES3 transmitter at appropriate times.

Each bit of each IU must be transceived; unlike the audio samples, there can be no sample rate conversion of the U data. Therefore, there are 2 potential problems:

(1) Message Partitioning

When Fso > Fsi, more data is transmitted than received per unit time. The FIFO will frequently be completely emptied. Sensible behavior must occur when the FIFO is empty, otherwise, a single incoming message may be erroneously be partitioned into multiple, smaller, messages.

(2) Overwriting

When Fso < Fsi, more data is received than transmitted per unit time. There is a danger of the FIFO

becoming completely full, allowing incoming data to overwrite data that has not yet been output through the AES3 transmitter.

16.2.4 Mode (3): Reserved

This mode has been removed. Use IEC Consumer mode B.

16.2.5 Mode (4): IEC Consumer B

In this mode, the partitioning problem is solved by buffering an entire message before starting to transmit it. In this scheme, zero-segments between messages will be expanded when Fso > Fsi, but the integrity of individual messages is preserved.

The overwriting problem (when Fso < Fsi) is solved by only storing a portion of the input U data in the FIFO. Specifically, only the IUs themselves are stored (and not the zeroes that provide inter-IU and inter-message “filler”). An inter-IU filler segment of fixed length (OF) will be added back to the messages at the FIFO output, where the length of OF is equal to the shortest observed input filler segment (IF).

Storing only IUs (and not filler) within the FIFO makes it possible for the slower AES3 transmitter to “catch up” to the faster AES3 receiver as data is read out of the FIFO. This is because nothing is written into the FIFO when long strings of zeroes are input to the AES-EBU receiver. During this time of no writing, the transmitter can read out data that had previously accumulated, allowing the FIFO to empty out. If the FIFO becomes completely empty, zeroes are transmitted until a complete message is written into the FIFO.

Mode 4 is not fail-safe; the FIFO can still get completely full if there isn't enough “zero-padding” between incoming messages. It is up to the user to provide proper padding, as defined below:

Minimum padding

$$= (Fsi/Fso - 1)*[8N + (N-1)*IF +9] + 9$$

where N is the number of IUs in the message, IF is the number of filler bits between each IU, and Fso ≤ Fsi.

Example 1: Fsi/Fso = 2, N=4, IF=1: minimum proper padding is 53 bits.

Example 2: Fsi/Fso = 1, N=4, IF=7: min proper padding is 9 bits.

The CS8420 detects when an overwrite has occurred in the FIFO, and synchronously resets the entire FIFO structure to prevent corrupted U data from being merged into the transmitted AES3 data stream. The CS8420 can be configured to generate an interrupt when this occurs.

Mode 4 is recommended for properly formatted U data where mode 3 cannot provide acceptable performance, either because of a too-extreme Fsi/Fso ratio, or because it's unacceptable to change the lengths of filler segments. Mode 4 provides error-free performance over the complete range of Fsi/Fso ratios (provided that the input messages are properly zero-padded for Fsi > Fso).

17. PARAMETER DEFINITIONS

Input Sample Rate (Fsi)

The sample rate of the incoming digital audio.

Input Frame Rate

The frame rate of the received AES3 format data.

Output Sample Rate (Fso)

The sample rate of the outgoing digital audio.

Output Frame Rate

The frame rate of the transmitted AES3 format data.

Dynamic Range

The ratio of the maximum signal level to the noise floor.

Total Harmonic Distortion and Noise

The ratio of the noise and distortion to the test signal level. Normally referenced to 0 dBFS.

Peak Idle Channel Noise Component

With an all-zero input, what is the amplitude of the largest frequency component visible with a 16K point FFT. The value is in dB ratio to full-scale.

Input Jitter Tolerance

The amplitude of jitter on the AES3 stream, or in the ILRCK clock, that will cause measurable artifacts in the SRC output. Test signal is full scale 9 kHz, Fsi is 48 kHz, Fso is different 48 kHz, jitter is 2 kHz sinusoidal, and audio band white noise.

AES3 Transmitter Output Jitter

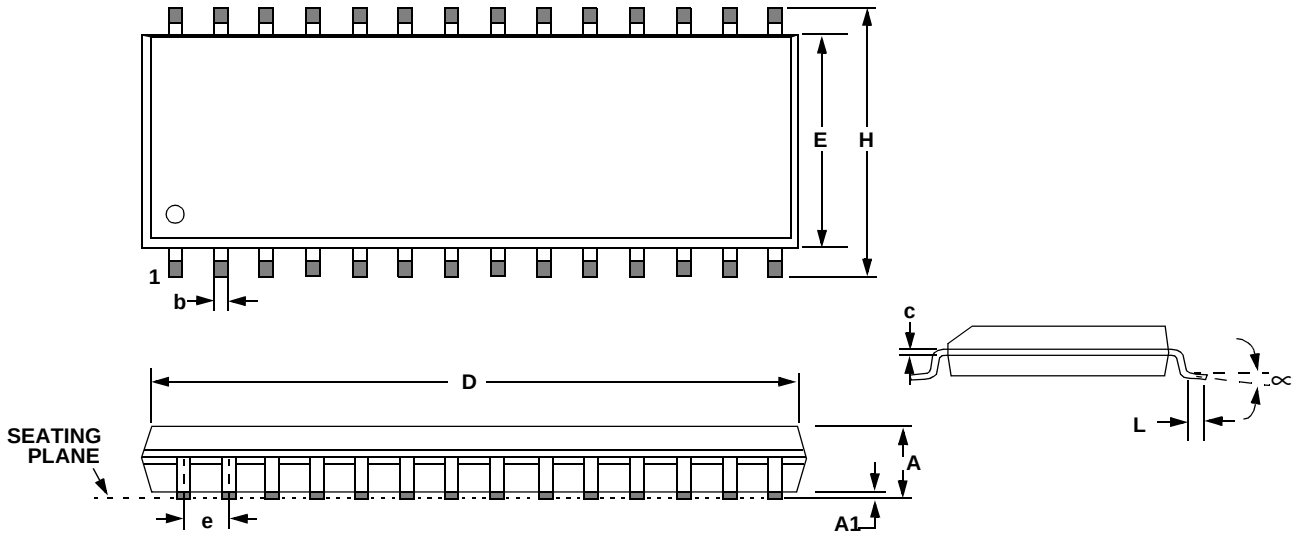
With a jitter free OMCK clock, what is the jitter added by the AES3 transmitter.

Gain Error

The difference in amplitude between the output and the input signal level, within the passband of the digital filter in the SRC.

18. PACKAGE DIMENSIONS

28L SOIC (300 MIL BODY) PACKAGE DRAWING



	INCHES		MILLIMETERS	
DIM	MIN	MAX	MIN	MAX
A	0.093	0.104	2.35	2.65
A1	0.004	0.012	0.10	0.30
B	0.013	0.020	0.33	0.51
C	0.009	0.013	0.23	0.32
D	0.697	0.713	17.70	18.10
E	0.291	0.299	7.40	7.60
e	0.040	0.060	1.02	1.52
H	0.394	0.419	10.00	10.65
L	0.016	0.050	0.40	1.27
∝	0°	8°	0°	8°

• Notes •


SMART
Analog
™