Dual 7 ns Single Supply Comparator

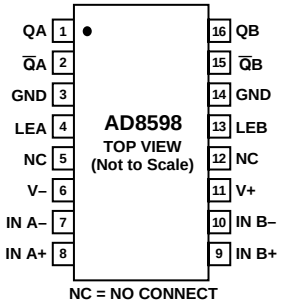
AD8598

FEATURES
7 ns Propagation Delay
Single Supply Operation: +3 V to +10 V
Low Power
Symmetrical Layout
Latch Function
TSSOP Packages

APPLICATIONS
Clock Recovery and Clock Distribution
High Speed Data
Line Receivers
Phase Detectors
Digital Communications
Land Q Detection
High Speed Sampling
Upgrade for MAX912
Satellite Receivers
PCMCIA Cards
Wireless Data Links
Battery Operated Instrumentation

PIN CONFIGURATIONS

N-16, RU-16 and R-16A



NC = NO CONNECT

OBSOLETE

GENERAL DESCRIPTION

The AD8598 is a dual 7 ns comparator with digital latches. Separate supplies enable the input stage to be operated from +5 V to as high as ±5 V.

Ultrafast 7 ns propagation delay makes the AD8598 a good choice for timing circuits and line receivers. Propagation delays for rising and falling signals are closely matched and track over temperature. This matched delay makes the AD8598 a good choice for clock recovery, since the duty cycle of the output will match the duty cycle of the input.

The AD8598 has the same pinout as the DIP version of the AD9698. For a single comparator like the AD8598, please refer to the AD8561 data sheet.

The AD8598 is specified over the industrial (–40°C to +85°C) temperature range. The AD8598 is available in both the 16-lead plastic DIP, 16-lead TSSOP or narrow R-16A surface mount packages.

REV. A

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AD8598–SPECIFICATIONS

ELECTRICAL SPECIFICATIONS (@ V+ = +5.0 V, V– = V_{GND} = 0 V, T_A = +25°C unless otherwise noted)

Parameter	Symbol	Conditions	Min	Typ	Max	Units
INPUT CHARACTERISTICS						
Offset Voltage	V_{OS}			2.3	7	mV
		–40°C ≤ T_A ≤ +85°C			8	mV
Offset Voltage Drift	$\Delta V_{OS}/\Delta T$			4		µV/°C
Input Bias Current	I_B	V_{CM} = 0 V	–6	–3		µA
	I_B	–40°C ≤ T_A ≤ +85°C	–7	–3.5		µA
Input Offset Current	I_{OS}	V_{CM} = 0 V			±4	µA
Input Common-Mode Voltage Range	V_{CM}		0.0		+3.0	V
Common-Mode Rejection Ratio	CMRR	0 V ≤ V_{CM} ≤ +3.0 V	65	85		dB
Large Signal Voltage Gain	A_{VO}	R_L = 10 kΩ		+3,000		V/V
Input Capacitance	C_{IN}			3.0		pF
LATCH ENABLE INPUT						
Logic "1" Voltage Threshold	V_{IH}		+2.0	+1.65		V
Logic "0" Voltage Threshold	V_{IL}			+1.60	+0.8	V
Logic "1" Current	I_{IH}	V_{LH} = +3.0 V	–1.0	–0.3		µA
Logic "0" Current	I_{IL}	V_{LL} = +0.3 V	–4	–2		µA
Latch Enable						
Pulsewidth	$t_{PW(E)}$			6		ns
Setup Time	t_S			1		ns
Hold Time	t_H			1.2		ns
DIGITAL OUTPUTS						
Logic "1" Voltage	V_{OH}	I_{OH} = –50 µA, ΔV_{IN} > 250 mV	+3.5			V
Logic "1" Voltage	V_{OH}	I_{OH} = –3.2 mA, ΔV_{IN} > 250 mV	+2.4	+3.5		V
Logic "0" Voltage	V_{OL}	I_{OL} = 3.2 mA, ΔV_{IN} > 250 mV		+0.25	+0.4	V
DYNAMIC PERFORMANCE						
Propagation Delay	t_P	200 mV Step with 100 mV Overdrive		6.75	9.8	ns
		–40°C ≤ T_A ≤ +85°C		8	12	ns
Propagation Delay	t_P	100 mV Step with 5 mV Overdrive		8		ns
Differential Propagation Delay (Rising Propagation Delay vs. Falling Propagation Delay)	Δt_P	100 mV Step with 100 mV Overdrive[1]		0.5	2.0	ns
Rise Time		20% to 80%		3.8		ns
Fall Time		80% to 20%		1.5		ns
POWER SUPPLY						
Power Supply Rejection Ratio	PSRR	+4.5 V ≤ V+ ≤ +5.5 V	50	65		dB
Positive Supply Current (V+ Pin)	I+			9.0	12.0	mA
		–40°C ≤ T_A ≤ +85°C			15.0	mA
Ground Supply Current (GND Pins)	I_{GND}	V_O = 0 V, R_L = ∞		4.4	6.6	mA
		–40°C ≤ T_A ≤ +85°C			7.6	mA
Analog Supply Current (V– Pin)	I–			4.6	9.0	mA
		–40°C ≤ T_A ≤ +85°C			11.0	mA

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NOTES

[1]Guaranteed by design.

Specifications subject to change without notice.

ELECTRICAL SPECIFICATIONS (@ V+ = +5.0 V, V_{GND} = 0 V, V– = –5 V, T_A = +25°C unless otherwise noted)

Parameter	Symbol	Conditions	Min	Typ	Max	Units
INPUT CHARACTERISTICS						
Offset Voltage	V_{OS}			1	7	mV
		–40°C ≤ T_A ≤ +85°C			8	mV
Offset Voltage Drift	$\Delta V_{OS}/\Delta T$			4		µV/°C
Input Bias Current	I_B	V_{CM} = 0 V	–6	–3		µA
	I_B	–40°C ≤ T_A ≤ +85°C	–7	–2.5		µA
Input Offset Current	I_{OS}	V_{CM} = 0 V			±4	µA
Input Common-Mode Voltage Range	V_{CM}		–5.0		+3.0	V
Common-Mode Rejection Ratio	CMRR	–5.0 V ≤ V_{CM} ≤ +3.0 V	65	85		dB
Large Signal Voltage Gain	A_{VO}	R_L = 10 kΩ		+3,000		V/V
Input Capacitance	C_{IN}			3.0		pF
LATCH ENABLE INPUT						
Logic "1" Voltage Threshold	V_{IH}		+2.0	+1.65		V
Logic "0" Voltage Threshold	V_{IL}			+1.60	+0.8	V
Logic "1" Current	I_{IH}	V_{LH} = +3.0 V	–1	–0.5		µA
Logic "0" Current	I_{IL}	V_{LL} = +0.3 V	–4	–2		µA
Latch Enable						
Pulsewidth	$t_{PW(E)}$			6		ns
Setup Time	t_S			1.0		ns
Hold Time	t_H			1.2		ns
DIGITAL OUTPUTS						
Logic "1" Voltage	V_{OH}	I_{OH} = –3.2 mA	+2.6	+3.5		V
Logic "0" Voltage	V_{OL}	I_{OL} = 3.2 mA		+0.2	+0.3	V
DYNAMIC PERFORMANCE						
Propagation Delay	t_P	200 mV Step with 100 mV Overdrive		6.5	9.8	ns
		–40°C ≤ T_A ≤ +85°C		8	13	ns
Propagation Delay	t_P	100 mV Step with 5 mV Overdrive		7		ns
Differential Propagation Delay (Rising Propagation Delay vs. Falling Propagation Delay)	Δt_P	100 mV Step with 100 mV Overdrive[1]		0.5	2	ns
Rise Time		20% to 80%		3.8		ns
Fall Time		80% to 20%		1.5		ns
Dispersion				1		ns
POWER SUPPLY						
Power Supply Rejection Ratio	PSRR	±4.5 V ≤ V_{CC} and V_{EE} ≤ ±5.5 V	55	70		dB
Supply Current		V_O = 0 V, R_L = ∞				
Positive Supply Current (V+ Pin)	I+			9.4	13.0	mA
		–40°C ≤ T_A ≤ +85°C			15.0	mA
Ground Supply Current (GND Pins)	I_{GND}	V_O = 0 V, R_L = ∞		4.4	6.6	mA
		–40°C ≤ T_A ≤ +85°C			7.6	mA
Analog Supply Current (V– Pin)	I–			4.8	9.0	mA
		–40°C ≤ T_A ≤ +85°C			11.0	mA

OBSOLETE

NOTES
[1]Guaranteed by design.

Specifications subject to change without notice.

AD8598–SPECIFICATIONS

ELECTRICAL SPECIFICATIONS (@ V+ = +3.0 V, V– = V_{GND} = 0 V, T_A = +25°C unless otherwise noted)

Parameter	Symbol	Conditions	Min	Typ	Max	Units
INPUT CHARACTERISTICS						
Offset Voltage	V_{OS}				7	mV
Input Bias Current	I_B	V_{CM} = 0 V	–6	–3.0		µA
	I_B	–40°C ≤ T_A ≤ +85°C	–7	–4		µA
Input Common-Mode Voltage Range	V_{CM}		0		+1.5	V
Common-Mode Rejection Ratio	CMRR	+0.1 V ≤ V_{CM} ≤ +1.5 V	50			dB
OUTPUT CHARACTERISTICS						
Output High Voltage[1]	V_{OH}	I_{OH} = –3.2 mA, V_{IN} > 250 mV	+1.2			V
Output Low Voltage	V_{OL}	I_{OL} = +3.2 mA, V_{IN} > 250 mV			+0.3	V
POWER SUPPLY						
Power Supply Rejection Ratio	PSRR	+2.7 V ≤ V_{CC}, V_{EE} ≤ +6 V		40		dB
Supply Currents		V_O = 0 V, R_L = ∞				
Positive Supply Current (V+ Pin)	I+			8.0	9.0	mA
		–40°C ≤ T_A ≤ +85°C			11.0	mA
Ground Supply Current (GND Pins)	I_{GND}			3.2	5.0	mA
		–40°C ≤ T_A ≤ +85°C			6.0	mA
Analog Supply Current (V– Pin)	I–			4.8	6.6	mA
		–40°C ≤ T_A ≤ +85°C			7.6	mA
DYNAMIC PERFORMANCE						
Propagation Delay	t_P	100 mV Step with 20 mV Overdrive[2]		8.5	9.8	ns

NOTES
[1]Output high voltage without pull-up resistor. It may be useful to have a pull-up resistor to V+ for +3 V operation.
[2]Guaranteed by design.

Specifications subject to change without notice.

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ABSOLUTE MAXIMUM RATINGS

Total Analog Supply Voltage . . . . . +14 V
Digital Supply Voltage . . . . . +7 V
Analog Positive Supply–Digital Positive Supply . . . . . –600 mV
Input Voltage[1] . . . . . ±7 V
Differential Input Voltage . . . . . ±8 V
Output Short-Circuit Duration to GND . . . . . Indefinite
Storage Temperature Range
N, R, RU Package . . . . . –65°C to +150°C
Operating Temperature Range . . . . . –40°C to +85°C
Junction Temperature Range
N, R, RU Package . . . . . –65°C to +150°C
Lead Temperature Range (Soldering, 10 sec) . . . . . +300°C

Package Type	θ_{JA}[2]	θ_{JC}	Units
16-Lead Plastic DIP	103	43	°C/W
16-Lead SOIC	158	43	°C/W
16-Lead TSSOP	240	43	°C/W

NOTES
[1]The analog input voltage is equal to ±7 V or the analog supply voltage, whichever is less.
[2]θ_{JA} is specified for the worst case conditions, i.e., θ_{JA} is specified for device in socket for P-DIP and θ_{JA} is specified for device soldered in circuit board for SOIC and TSSOP packages.

ORDERING GUIDE

Model	Temperature Range	Package Descriptions	Package Options
AD8598AN	–40°C to +85°C	16-Lead Plastic DIP	N-16
AD8598ARU	–40°C to +85°C	16-Lead Thin Shrink Small Outline (TSSOP)	RU-16
AD8598AR	–40°C to +85°C	16-Lead Narrow Body IC	R-16A

CAUTION

ESD (electrostatic discharge) sensitive device. Electrostatic charges as high as 4000 V readily accumulate on the human body and test equipment and can discharge without detection. Although the AD8598 features proprietary ESD protection circuitry, permanent damage may occur on devices subjected to high energy electrostatic discharges. Therefore, proper ESD precautions are recommended to avoid performance degradation or loss of functionality.



Typical Performance Characteristics (V+ = +5 V, V– = 0 V, T_A = +25°C unless otherwise noted)

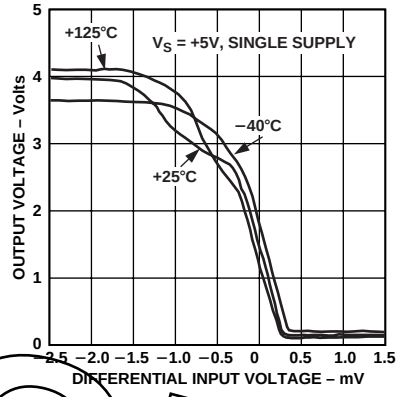


Figure 1. Output Voltage vs. Differential Input Voltage

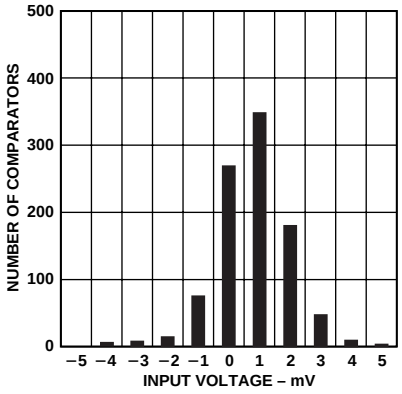


Figure 2. Typical Distribution of Input Offset Voltage

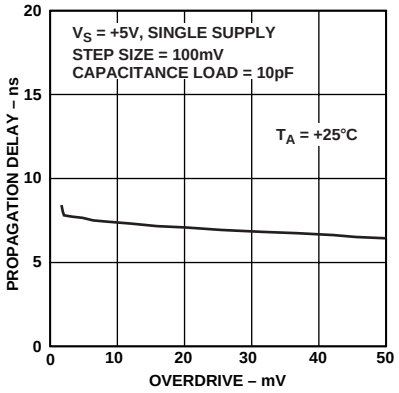


Figure 3. Propagation Delay vs. Overdrive

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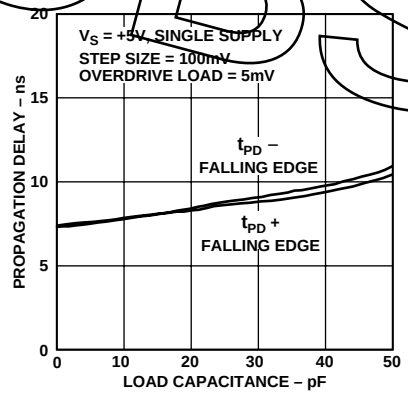


Figure 4. Propagation Delay vs. Load Capacitance

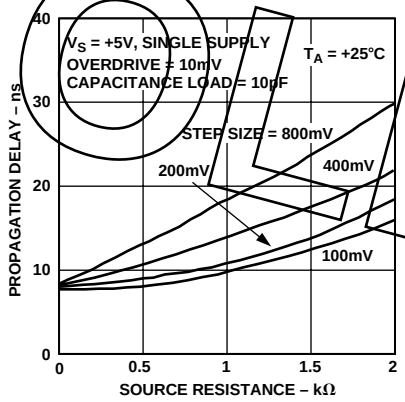


Figure 5. Propagation Delay vs. Source Resistance

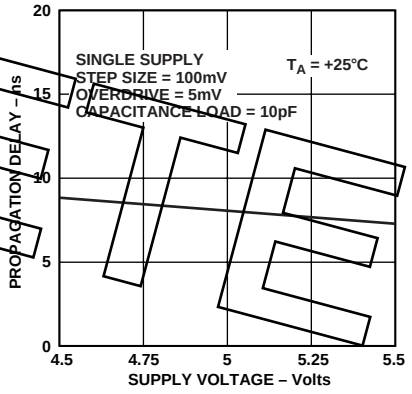


Figure 6. Propagation Delay vs. Positive Supply Voltage

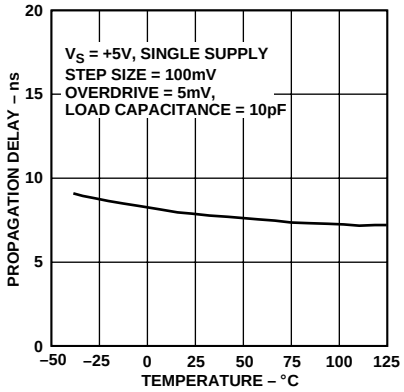


Figure 7. Propagation Delay vs. Temperature

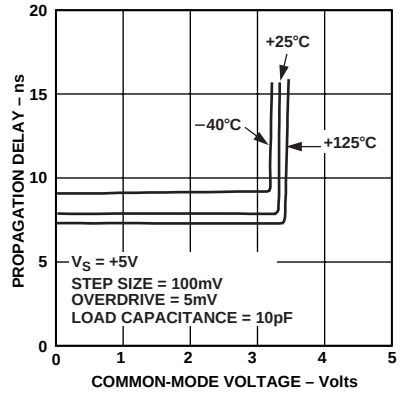


Figure 8. Propagation Delay vs. V_{CM}

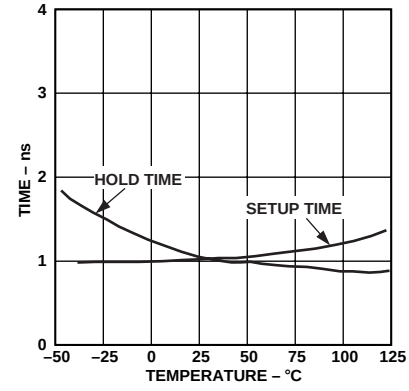


Figure 9. Latch Setup-and-Hold Time vs. Temperature

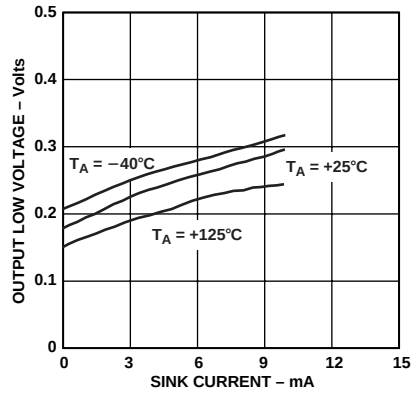


Figure 10. Output Low Voltage, V_{OL} vs. Sink Current

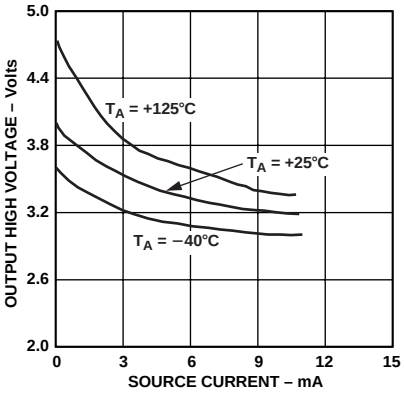


Figure 11. Output High Voltage, V_{OH} vs. Source Current

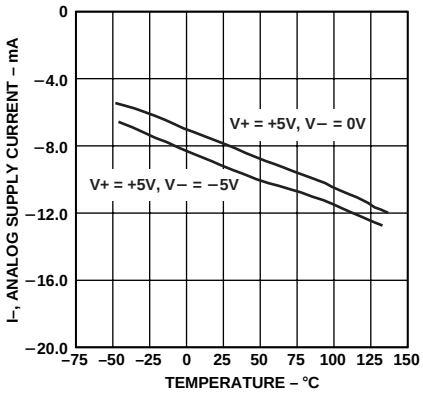


Figure 12. Analog Supply Current vs. Temperature for ±5 V Supplies

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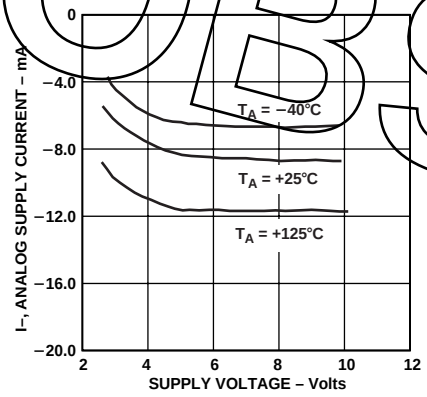


Figure 13. Analog Supply Current vs. Supply Voltage

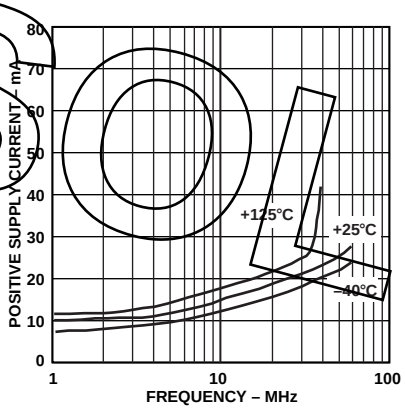


Figure 14. Positive Supply Current vs. Input Frequency

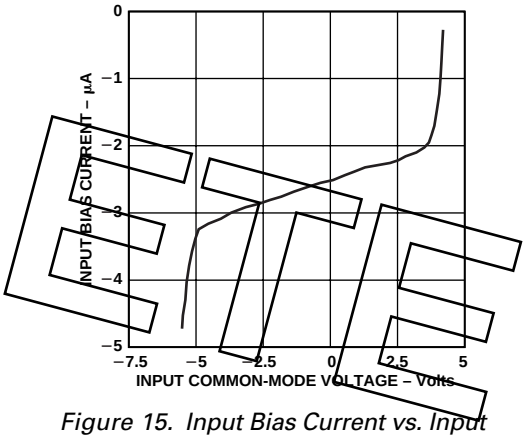


Figure 15. Input Bias Current vs. Input Common-Mode Voltage for ±5 V Supplies

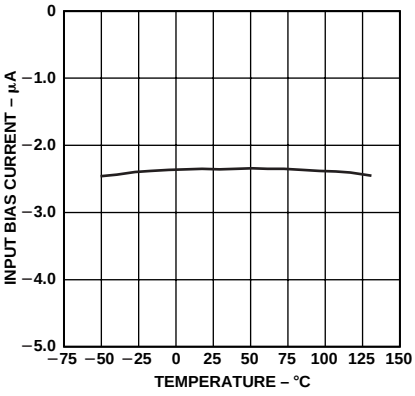


Figure 16. Input Bias Current vs. Temperature

APPLICATIONS

Optimizing High Speed Performance

As with any high speed comparator or amplifier, proper design and layout techniques should be used to ensure optimal performance from the AD8598. The performance limits of high speed circuitry can easily be a result of stray capacitance, improper ground impedance or other layout issues.

Minimizing resistance from source to the input is an important consideration in maximizing the high speed operation of the AD8598. Source resistance in combination with equivalent input capacitance could cause a lagged response at the input, thus delaying the output. The input capacitance of the AD8598, in combination with stray capacitance from an input pin to ground could result in several picofarads of equivalent capacitance. A combination of 3 kΩ source resistance and 5 pF of input capacitance yields a time constant of 15 ns, which is slower than the 5 ns capability of the AD8598. Source impedances should be less than 1 kΩ for the best performance.

It is also important to provide bypass capacitors for the power supply in a high speed application. A 1 μF electrolytic bypass capacitor should be placed within 0.5 inches of each power supply pin to ground. These capacitors will reduce any potential voltage ripples from the power supply. In addition, a 10 nF ceramic capacitor should be placed as close as possible from the power supply pins to ground. These capacitors act as a charge reservoir for the device during high frequency switching.

A ground plane is recommended for proper high speed performance. This can be created by using a continuous conductive plane over the surface of the circuit board, only allowing breaks in the plane for necessary current paths. The ground plane provides a low inductance ground, eliminating any potential differences at different ground points throughout the circuit board caused from "ground bounce." A proper ground plane also minimizes the effects of stray capacitance on the circuit board.

Replacing the MAX912

The AD8598 is pin compatible with the MAX912 comparator. While it is easy to replace the MAX912 with the higher performance AD8598, please note that there are differences, and it is useful to check these to ensure proper operation.

There are five major differences between the AD8598 and the MAX912; input voltage range, input bias currents, speed, output swing and power consumption.

When operated on a +5 V single supply, the MAX912 has an input voltage range from –0.2 V to +3.5 V. The AD8598 has an input range from 0 V to +3.0 V. Signals above +3.0 V may result in slower response times (see Figure 8). If both signals exceed +3.0 V, the signals may be shifted or attenuated to bring them into range, keeping in mind the note about source resistance in Optimizing High Speed Performance. If only one of the signals exceeds +3.0 V only slightly, and the other signal is always well within the 0 V to +3 V range, the comparator may operate without changes to the circuit.

Example: A comparator compares a fast moving signal to a fixed +2.5 V reference. Since the comparator only needs to operate when the signal is near +2.5 V, both signals will be within the input range (near +2.5 V and well under +3.0 V) when the comparator needs to change output.

Note that signals much greater than +3.0 V will result in increased input currents and may cause the device to operate more slowly.

The input bias current of the AD8598 is the same magnitude (–3 μA typical) as the MAX912 (+3 μA typical), and the current flows out of the AD8598 and into MAX912. If relatively low value resistors and/or low impedance sources are used on the inputs, the voltage shift due to bias current should be small.

The AD8598 (6.75 ns typical) is faster than the MAX912 (10 ns typical). While this is beneficial to many systems, timing may need to be adjusted to take advantage of the higher speed.

The AD8598 has slightly more output voltage swing when the output is lightly loaded.

The AD8598 uses less current (typically 10 mA) than the MAX912 (typically 12 mA).

Increasing Output Swing

Although not required for normal operation, the output voltage swing of the AD8598 can be increased by connecting a 5 kΩ resistor from the output of the device to the V+ power supply. This configuration can be useful in low voltage power supply applications where maximizing output voltage swing is important. Adding a 5 kΩ pull-up resistor to the device's output will not adversely affect the specifications of the AD8598.

Output Loading Considerations

The AD8598 output can deliver up to 40 mA of output current without any significant increase in propagation delay. The output of the device should not be connected to more than twenty (20) TTL input logic gates, nor drive a load resistance less than 100 Ω.

To ensure the best performance from the AD8598 it is important to minimize capacitive loading of the output of the device. Capacitive loads greater than 50 pF will cause ringing on the output waveform and will reduce the operating bandwidth of the comparator.

Setup and Hold Times for Latching the Output

The latch inputs can be used to retain data at the outputs of the AD8598. When the voltage at the latch input goes high, the output of the device will remain constant regardless of the input voltages. The setup time for the latch is 2 ns–3 ns and the hold time is 3 ns. This means that to ensure data retention at the output, the input signal must be valid at least 5 ns before the latch pin goes high and must remain valid at least 3 ns after the latch pin goes high. Once the latch input voltage goes low, new output data will appear in approximately 8 ns.

A logic high for the latch input is a minimum of +2.0 V and a logic low is a maximum of +0.8 V. This makes the latch input easily interface with TTL or CMOS logic gates. The latch circuitry in the AD8598 has no built-in hysteresis.

Input Stage and Bias Currents

The AD8598 uses a PNP differential input stage that enables the input common-mode range to extend all the way from the negative supply rail to within +2.2 V of the positive supply rail. The input common-mode voltage can be found as the average of the voltage at the two inputs of the device. To ensure the fastest response time, care should be taken not to allow the input common-mode voltage to exceed either of these voltages.

OBSOLETE
LATEST

The input bias current for the AD8598 is 3 μA. As with any PNP differential input stage, this bias current will go to zero on an input that is high and will double on an input that is low. Care should be taken in choosing resistor values to be connected to the inputs as large resistors could cause significant voltage drops due to the input bias current.

The input capacitance for the AD8598 is typically 3 pF. This is measured by inserting a 5 kΩ source resistance to the input and measuring the change in propagation delay.

Using Hysteresis

Hysteresis can easily be added to a comparator through the addition of positive feedback. Adding hysteresis to a comparator offers an advantage in noisy environments where it is not desirable for the output to toggle between states when the input signal is near the switching threshold. Figure 17 shows a method for configuring the AD8598 with hysteresis.

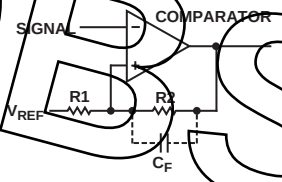


Figure 17. Configuring the AD8598 with Hysteresis

The input signal is directly connected to the noninverting input of the comparator. The output is fed back to the inverting input through R1 and R2. The ratio of R1 to R1 + R2 establishes the width of the hysteresis window with V_{REF} setting the center of the window, or the average switching voltage. The Q output will switch high when the input voltage is greater than V_{HI} and will not switch low again until the input voltage is less than V_{LO} as given in Equation 1:

$$V_{HI} = \left(V_+ - 1 - V_{REF}\right)\frac{R1}{R1+R2} + V_{REF}$$

$$V_{LO} = V_{REF}\left(1 - \frac{R1}{R1+R2}\right) \quad (1)$$

where V_+ is the positive supply voltage.

The capacitor, C_F, can also be added to introduce a pole into the feedback network. This has the effect of increasing the amount of hysteresis at high frequencies. This can be useful when comparing a relatively slow signal in a high frequency noise environment. At frequencies greater than $f_P = \frac{1}{2\pi C_F R2}$, the hysteresis window approaches $V_{HI} = V_+ - 1$ V and $V_{LO} = 0$ V. At frequencies less than f_P the threshold voltages remain as in Equation 1.

OBSOLETE

SPICE Model

```
* AD8598 SPICE Macro-Model Typical Values
* 4/98, Ver. 1.0
* TAM / ADSC
*
* Node assignments
*                non-inverting input
*                |       inverting input
*                |       |       positive supply
*                |       |       |       negative supply
*                |       |       |       |       Latch
*                |       |       |       |       |       DGND
*                |       |       |       |       |       |       Q
*                |       |       |       |       |       |       |       QNOT
*                |       |       |       |       |       |       |       |
.SUBCKT AD8598   1       2       99      50      80      51      45      65
*
* INPUT STAGE
*
*
Q1     4  3 5 PIX
Q2     6  2 5 PIX
IBIAS 99  5 800E-6
RC1    4 50 1E3
RC2    6 50 1E3
CL1    4  6 1E-12
CIN    1  2 3E-12
VCM1  99  7 1
D1     5  7 DX
EOS    3  1 POLY(1) (31,98) 1E-3 1
*
* Reference Voltage
*
EREF 98 0 POLY(2) (99,0) (50,0) 0 0.5 0.5
RREF 98 0 100E3
*
* CMRR=80dB, ZERO AT 1kHz
*
ECM1 30 98 POLY(2) (1,98) (2,98) 0 0.5 0.5
RCM1 30 31 10E3
RCM2 31 98 1
CCM1 30 31 15.9E-9
*
* Latch Section
*
RX 80 51 100E3
E1 10 98 (4,6) 1
S1 10 11 (80,51) SLATCH1
R2 11 12 1
C3 12 98 10E-12
E2 13 98 (12,98) 1
R3 12 13 500
*
* Power Supply Section
*
```

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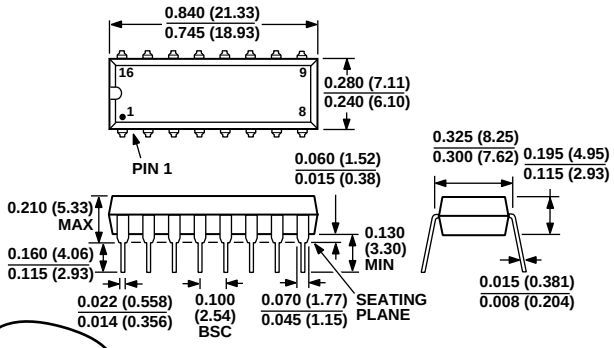
```
GSY1 99 52 POLY(1) (99,50) 4E-3 -2.6E-4
GSY2 52 50 POLY(1) (99,50) 3.7E-3 -.6E-3
RSY  52 51 10
*
* Gain Stage Av=250 fp=100MHz
*
G2 98 20 (12,98) 0.25
R1 20 98 1000
C1 20 98 10E-13
D2 20 21 DX
D3 22 20 DX
V1 99 21 DC 0.8
V2 22 50 DC 0.8
*
* Q Output
*
Q3  99 41 46 NOX
Q4  47 42 50 NOX
RB1 43 41 200
RB2 40 42 5E3
CB1 99 41 10E-12
CB2 42 50 5E-12
RO1 46 45 2E3
RO2 47 45 500
EO1 98 43 POLY(1) (20,98) 0 1
EO2 40 98 POLY(1) (20,98) 0 1
*
* Q NOT Output
*
Q5  99 61 66 NOX
Q6  67 62 50 NOX
RB3 63 61 200
RB4 60 62 5E3
CB3 99 61 10E-12
CB4 62 50 5E-12
RO3 66 65 2E3
RO4 67 65 500
EO3 63 98 POLY(1) (20,98) 0 1
EO4 98 60 POLY(1) (20,98) 0 1
*
* MODELS
*
.MODEL PIX PNP(BF=100,IS=1E-16)
.MODEL NOX NPN(BF=100,VAF=130,IS=1E-14)
.MODEL DX D(IS=1E-16)
.MODEL SLATCH1 VSWITCH(ROFF=1E6,RON=500,VOFF=2.1,VON=1.4)
.ENDS AD8598
```

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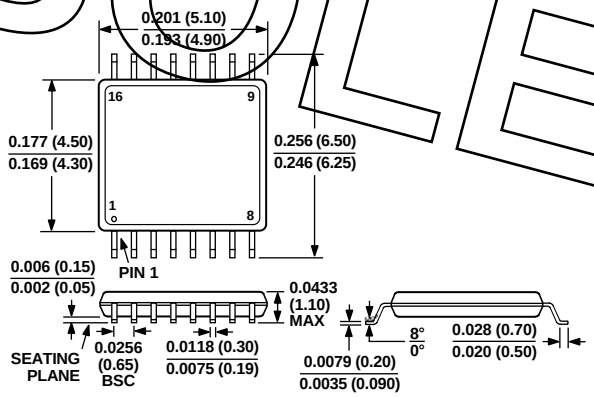
OUTLINE DIMENSIONS

Dimensions shown in inches and (mm).

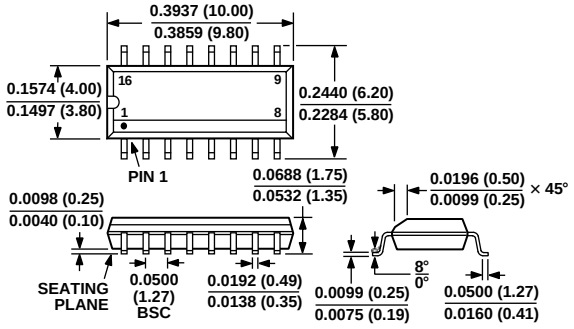
16-Lead Plastic DIP
(N-16)



16-Lead Thin Shrink Small Outline
(RU-16)



16-Lead Narrow Body IC
(R-16A)



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