- High Resolution, 1/3-in Solid-State Image Sensor for NTSC Black and White **Applications**
- 340,000 Pixels per Field
- Frame Memory
- 658 (H) \times 496 (V) Active Elements in Image **Sensing Area Compatible With Electronic** Centering
- **Multimode Readout Capability**
 - Progressive Scan
 - Interlaced Scan
 - Dual-Line Readout
 - Image Area Line Summing
 - Smear Subtraction
- **Fast Single-Pulse Clear Capability**
- **Continuous Electronic Exposure Control** From 1/60-1/50,000 s
- 7.4-µm Square Pixels
- **Advanced Lateral Overflow Drain** Antiblooming
- **Low Dark Current**

(TOP VIEW) ODB 12 IAG1 IAG2 2 11 SAG SUB 3 10 SAG ADB 4 SUB 9 OUT1 5 SRG OUT2 6 RST

DUAL-IN-LINE PACKAGE

- **High Photoresponse Uniformity**
- **High Dynamic Range**
- **High Sensitivity**
- **High Blue Response**
- Solid-State Reliability With No Image Burn-In, Residual Imaging, Image Distortion, Image Lag, or **Microphonics**

description

The TC237B is a frame-transfer, charge-coupled device (CCD) image sensor designed for use in single-chip black and white National Television Standards Committee (NTSC) TV, computer, and special-purpose applications that require low cost and small size.

The image-sensing area of the TC237B device is configured into 500 lines with 680 elements in each line. Twenty-two elements are provided in each line for dark reference. The antiblooming feature of the sensor is based on an advanced lateral overflow drain concept. The sensor can be operated in a true interlace mode as a $658(H) \times 496(V)$ sensor with a low dark current. An important feature of the TC237B high-resolution sensor is the ability to capture a full 340,000 pixels per field. The image sensor also provides high-speed image transfer capability and a continuous electronic exposure control without the loss of sensitivity and resolution inherent in other technologies. Charge voltage is converted to signal voltage at 13 μV per electron by a high-performance structure with a reset and a voltage-reference generator. The signal is further buffered by a low-noise, two-stage, source-follower amplifier to provide high-output drive capability.

The TC237B sensor is built using TI-proprietary advanced virtual-phase (AVP) technology, which provides devices with high blue response, low dark current, high photoresponse uniformity, and single-phase clocking. The TC237B sensor is characterized for operation from -10° C to 45° C.



This MOS device contains limited built-in gate protection. During storage or handling, the device leads should be shorted together or the device should be placed in conductive foam. In a circuit, unused inputs should always be connected to Vss. Under no circumstances should pin voltages exceed absolute maximum ratings. Avoid shorting OUT to VSS during operation to prevent damage to the amplifier. The device can also be damaged if the output terminals are reverse-biased and an excessive current is allowed to flow. Specific guidelines for handling devices of this type are contained in the publication Guidelines for Handling Electrostatic-Discharge-Sensitive (ESDS) Devices and Assemblies available from Texas Instruments.



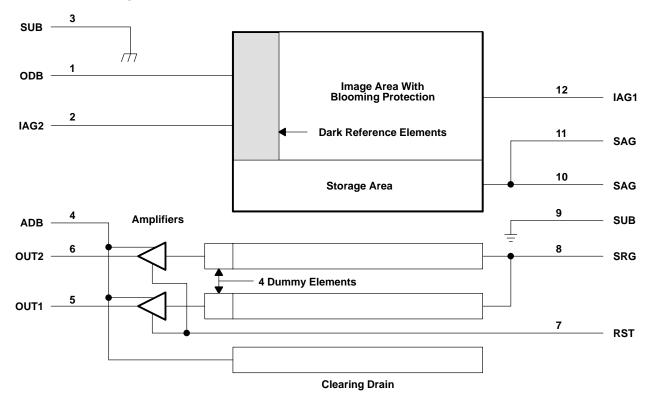
Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

TI is a trademark of Texas Instruments.

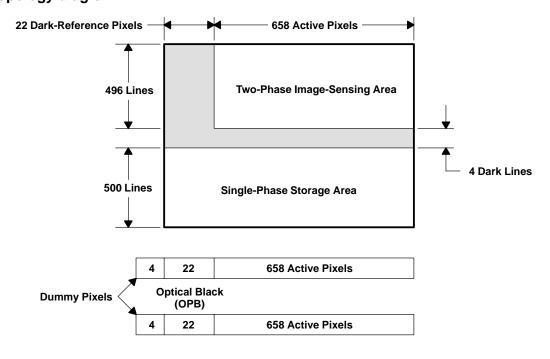
All other trademarks are the property of their respective owners.



functional block diagram



sensor topology diagram





Terminal Functions

TERM	TERMINAL		DESCRIPTION		
NAME	NO.	1/0	DESCRIPTION		
ADB	4	I	Supply voltage for amplifier drain bias		
IAG1	12	I	Image area gate 1		
IAG2	2	I	Image area gate 2		
ODB	1	I	Supply voltage for drain antiblooming bias		
OUT1	5	0	Output signal 1		
OUT2	6	0	Output signal 2		
RST	7	I	Reset gate		
SAG	10, 11	I	Storage area gate		
SRG	8	Ī	Serial register gate		
SUB	3, 9		Substrate		

detailed description

The TC237B CCD image sensor consists of four basic functional blocks: the image-sensing area, the image storage area, the serial register gates, and the low-noise signal processing amplifier block with charge detection nodes and independent resets. The location of each of these blocks is identified in the functional block diagram.

image-sensing and image storage areas

Figure 1 and Figure 2 show cross sections with potential well diagrams and top views of the image-sensing and storage area elements. As light enters the silicon in the image-sensing area, electrons are generated and collected in the wells of the sensing elements. Blooming protection is provided by applying a dc bias to the overflow drain bias pin. To clear the image before beginning a new integration time (for implementation of electronic fixed shutter or electronic auto-iris), apply a pulse of at least 1 µs to the overflow drain bias. After integration is complete, charge voltage is transferred into the storage area. The transfer timing depends on whether the readout mode is interlace or progressive scan. If the progressive-scan readout mode is selected, the readout may be performed by using one serial register or at high speed by using both serial registers (see Figure 3 through Figure 5). A line-summing operation, which is useful in off-chip smear subtraction, can be implemented before the parallel transfer (see Figure 6).

Twenty-two columns at the left edge of the image-sensing area are shielded from incident light; these elements provide the dark reference used in subsequent video-processing circuits to restore the video black level. In addition, four dark lines between the image-sensing and the image storage area prevent charge leakage from the image-sensing area into the image storage area.

advanced lateral overflow drain

The advanced lateral overflow drain structure is shared by two neighboring pixels and provides several unique features in the sensor. By varying the dc bias of the drain pin, you can control the blooming protection level and trade it for the well capacity.

To clear charge voltages in the image area, apply a 10-V pulse for a minimum duration of 1 μ s above the nominal dc bias level. This feature permits a precise control of the integration time on a frame-by-frame basis. The single-pulse clear capability also reduces smear by eliminating accumulated charge from the pixels before the start of the integration (single-sided smear).

Application of a negative 1-V pulse to the ODB signal during the parallel transfer is recommended to prevent slight column-to-column pixel well capacity variations in some artifacts.



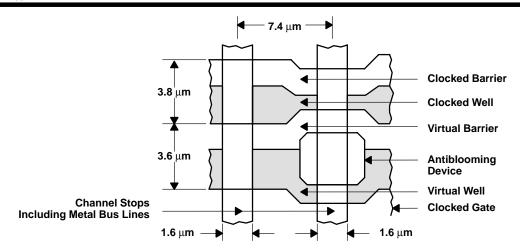


Figure 1. Image-Area Pixel Structure

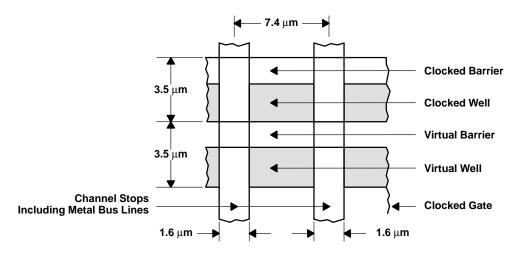


Figure 2. Storage-Area Pixel Structure



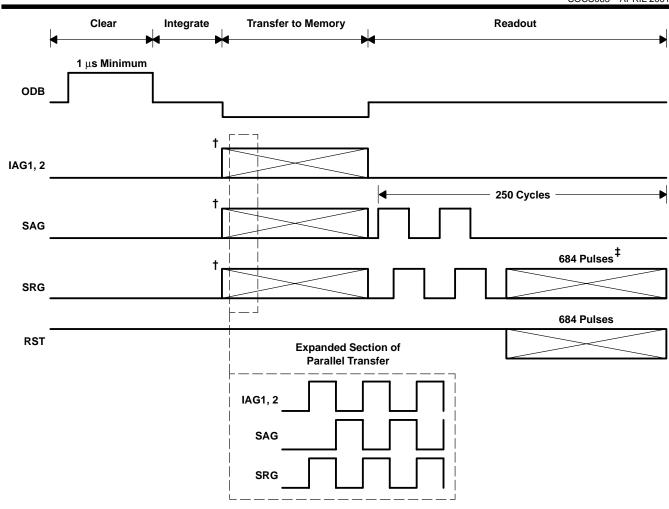
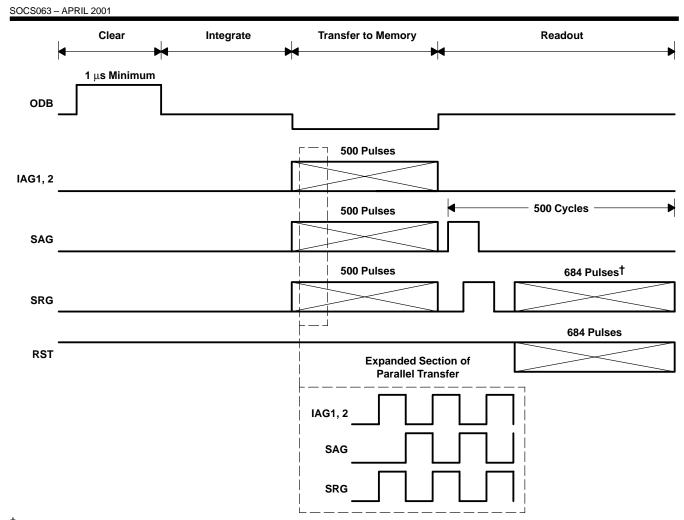


Figure 3. Interlace Timing

[†] The number of parallel-transfer pulses is field dependent. Field 1 has 500 pulses of IAG1, IAG2, SAG, and SRG with appropriate phasing. Field 2 has 501 pulses.

[‡] The readout is from register 2.



 \dagger The readout is from register 2.

Figure 4. Progressive-Scan Timing With Single Register Readout



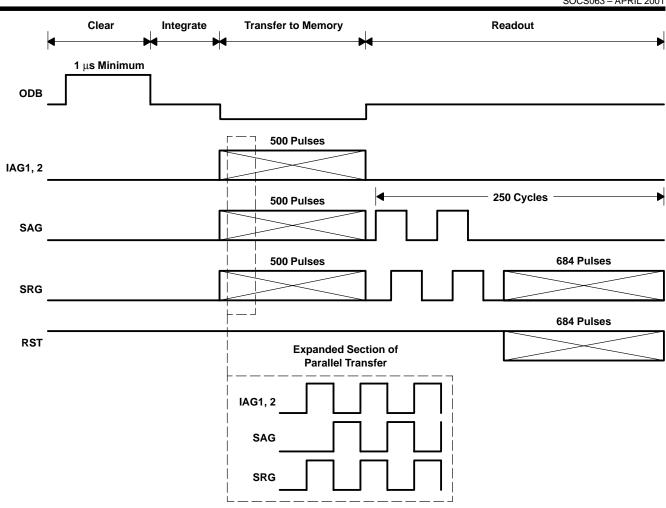


Figure 5. Progressive-Scan Timing With Dual-Register Readout

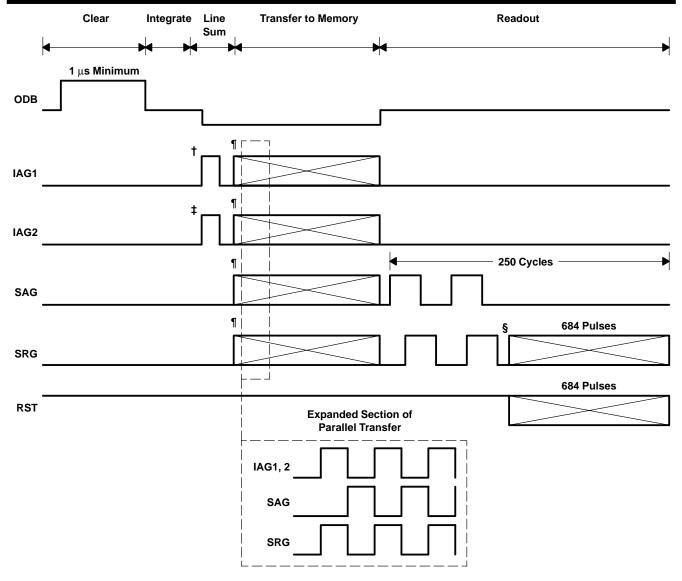


Figure 6. Line-Summing Timing

[†] This pulse occurs only during field 1.

[‡] This pulse occurs only during field 2.

[§] While readout is from register 2, register 1 can be read out for off-chip smear subtraction.

[¶] The number of parallel transfer pulses if field dependent. Field 1 has 500 pulses, and field 2 has 501 pulses.

serial registers

The storage area gate and serial gate(s) are used to transfer charge line-by-line from the storage area into the serial register(s). Depending on the readout mode, one or both serial registers are used. If both are used, the registers are read out in parallel.

readout and video processing

After transfer into the serial register(s), the pixels are clocked out and sensed by a charge detection node. The node must be reset to a reference level before the next pixel is placed onto it. The timing for the serial-register readout, which includes the external pixel clamp and sample-and-hold signals needed to implement correlated double sampling, is shown in Figure 7. As charge is transferred onto the detection node, the potential of the node changes in proportion to the amount of the charge received. The change is sensed by an MOS transistor; after proper buffering, the signal is supplied to the output terminal of the image sensor. Figure 8 shows the circuit diagram of the charge detection node and output amplifier. The detection nodes and amplifiers are placed a short distance from the edge of the storage area; therefore, each serial register contains 4 dummy elements that are used to span the distance between the serial registers and the amplifiers.

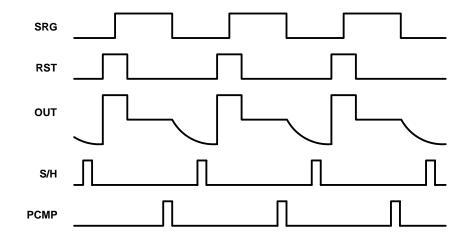


Figure 7. Serial Readout and Video-Processing Timing

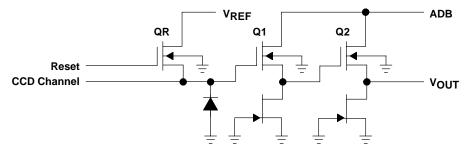


Figure 8. Output Amplifier and Charge Detection Node



absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Supply voltage range, ADB (see Note 1)	SUB to SUB + 15 V
Supply voltage range, ODB	SUB to SUB + 21 V
Input voltage range, V _I : IAG1, IAG2, SAG, SRG	0 V to 15 V
Operating free-air temperature range, T _A	10°C to 45°C
Storage temperature range, T _{stq}	30°C to 85°C
Operating case temperature range, T _C	–10°C to 55°C

[†] Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTE 1: All voltage values are with respect to SUB.

recommended operating conditions

				MIN	NOM	MAX	UNIT
	ADB			21	22	23	
Owner have a life and M	ODB	For antiblooming control		15.5	16	16.5	V
Supply voltage, V _{CC}		For clearing	For clearing		26	26.5	
		For transfer	For transfer		15	15.5	
Substrate bias voltage					10		V
Input voltage, V _I		14.04 14.00	High level	11.5	12	12.5	V
		IAG1, IAG2	Low level		0		
		SAG	High level	11.5	12	12.5	
			Low level		0		
		222 227	High level	11.5	12	12.5	
		SRG, RST	Low level		0		
Clock frequency, f _{clock}		IAG1, IAG2	IAG1, IAG2		12.5		
		SAG	SAG		12.5		MHz
		SRG, RST	SRG, RST		12.5		
Load capacitance		OUT1, OUT2	OUT1, OUT2			6	pF
Operating free-air temperature, T _A			-10		45	°C	



electrical characteristics over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

P.	MIN	TYP†	MAX	UNIT		
5	With CDS‡§		64		59 dB	
Dynamic range (see Note 2)	Without CDS [‡] §		58	59		
Charge conversion factor			13		μV/e	
Charge transfer efficiency (see Note 3)		0.9999	0.99995	1		
Signal response delay time, τ (see Note 4)			12.5		ns	
Gamma (see Note 5)				1		
Output resistance		300	400	500	Ω	
	With CDS‡	15	18	21	electrons	
Amplifier noise equivalent signal	Without CDS [‡]	30	36	42		
	ADB (see Note 6)		20			
Rejection ratio	SRG (see Note 7)		45		dB	
	ODB (see Note 8)		25		1	
Supply current			5	10	mA	
	IAG1, IAG2		2000			
	SRG		70		pF	
Input capacitance, Ci	RST		10			
	SAG		4000			

[†] All typical values are at $T_A = 25$ °C.

- NOTES: 2. Dynamic range is –20 times the logarithm of the mean-noise signal divided by the saturation output signal.
 - 3. Charge transfer efficiency is one minus the charge loss per transfer in the output register. The test is performed in the dark using an electrical input signal.
 - 4. Signal response delay time is the time between the falling edge of the SRG pulse and the output signal valid state.
 - 5. Gamma (γ) is the value of the exponent in the equation below for two points on the linear portion of the transfer function curve (this value represents points near saturation).

$$\left(\frac{\text{Exposure (2)}}{\text{Exposure (1)}}\right)^{\gamma} = \left(\frac{\text{Output signal (2)}}{\text{Output signal (1)}}\right)$$

- 6. ADB rejection ratio is -20 times the logarithm of the ac amplitude at the output divided by the ac amplitude at ADB.
- 7. SRG rejection ratio is -20 times the logarithm of the ac amplitude at the output divided by the ac amplitude at SRG.
- 8. ODB rejection ratio is -20 times the logarithm of the ac amplitude at the output divided by the ac amplitude at ODB.



[‡]CDS = Correlated double sampling, a signal-processing technique that improves noise performance by subtraction of reset noise.

[§] Performance depends on the particular implementation of the CDS technique and on the selected filter bandwidth that precedes sampling.

optical characteristics, $T_A = 40^{\circ}C$ (unless otherwise noted)

	MIN	TYP	MAX	UNIT		
Operation (page Marie O)	No IR filter With IR filter			256		
Sensitivity (see Note 9)				32		mV/lux
Saturation signal, V _{sat} (see Note 10)	Antiblooming disabled	T _A =45°C	320			mV
Maximum usable signal, V _{use}	Antiblooming enabled	T _A =45°C	120			mV
Blooming overload ratio (see Note 11)			500			
Image area well capacity				30K	38K	electrons
Smear (see Note 12)		See Note 13			-78	dB
Dark current		T _A = 21°C			0.05	nA/cm ²
Dark signal		T _A = 45°C			1	mV
Dark-signal uniformity	T _A = 45°C			0.5	mV	
Dark-signal shading	T _A = 45°C			0.5	mV	
0	Dark	T _A = 45°C			10	mV
Spurious nonuniformity	Illuminated, F#8	T _A = 45°C			15	%
Column uniformity					0.5	mV
Electronic shutter capability				1/60		S

NOTES: 9. Theoretical value

- 10. Saturation is the condition in which further increase in exposure does not lead to further increase in output signal.
- 11. Blooming is the condition in which charge is induced in an element by light incident on another element. Blooming-overload ratio is the ratio of blooming exposure to saturation exposure.
- 12. Smear is a measure of the error introduced by transferring charge through an illuminated pixel in shutterless operation. It is equivalent to the ratio of the single-pixel transfer time to the exposure time using an illuminated section that is 1/10 of the image-area vertical height with recommended clock frequencies.
- 13. The exposure time is 16.67 ms, the fast-dump clocking rate during vertical transfer is 12.5 MHz, and the illuminated section is 1/10 the height of the image section.

TYPICAL CHARACTERISTICS

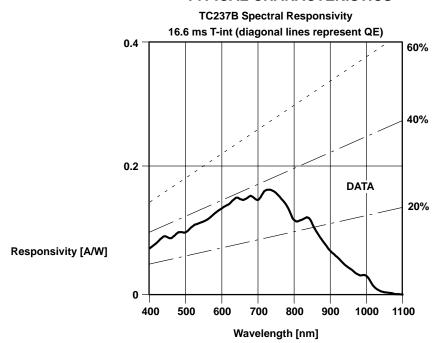
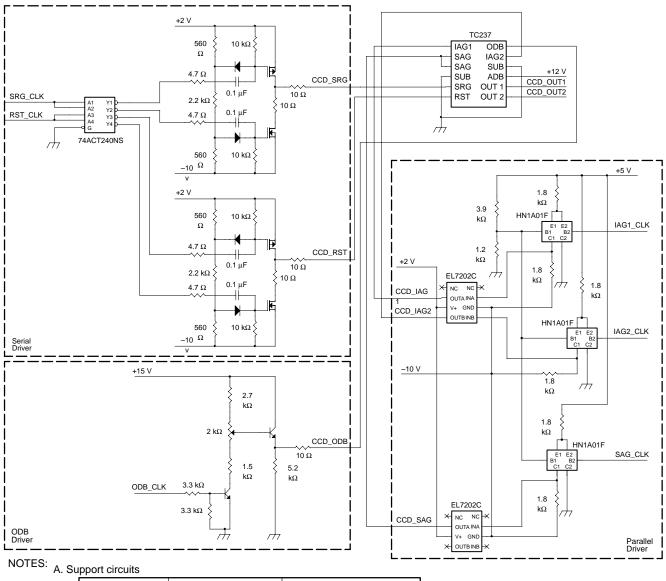


Figure 9. Spectral Characteristics of the TC237B CCD Sensor



APPLICATION INFORMATION



DEVICE	APPLICATION	FUNCTION
EL7202C	Parallel driver	Driver for IAG1, IAG2,SAG
74ACT240NS	Serial pre-driver	Driver for SRG, RST

B. Clock, DC voltages

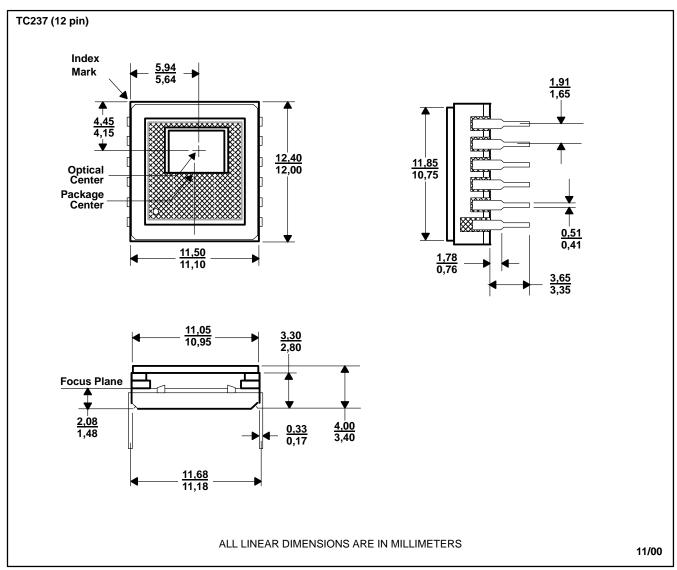
Clock	SRG_CLK, RST_CLK, ODB_CLK, IAG1_CLK, IAG2_CLK, SAG_CLK	TTL level
DC	+15 V, +12 V, +5 V, +2 V, 0 V, (Ground), -10 V	

Figure 10. Typical Application Circuit Diagram



MECHANICAL DATA

The package for the TC237B consists of a ceramic base, a glass window, and a 12-lead frame. The glass window is sealed to the package by an epoxy adhesive. The package leads are configured in a dual-in-line organization and fit into mounting holes with 1,78 mm center-to-center spacings.





IMPORTANT NOTICE

Texas Instruments and its subsidiaries (TI) reserve the right to make changes to their products or to discontinue any product or service without notice, and advise customers to obtain the latest version of relevant information to verify, before placing orders, that information being relied on is current and complete. All products are sold subject to the terms and conditions of sale supplied at the time of order acknowledgment, including those pertaining to warranty, patent infringement, and limitation of liability.

TI warrants performance of its products to the specifications applicable at the time of sale in accordance with TI's standard warranty. Testing and other quality control techniques are utilized to the extent TI deems necessary to support this warranty. Specific testing of all parameters of each device is not necessarily performed, except those mandated by government requirements.

Customers are responsible for their applications using TI components.

In order to minimize risks associated with the customer's applications, adequate design and operating safeguards must be provided by the customer to minimize inherent or procedural hazards.

TI assumes no liability for applications assistance or customer product design. TI does not warrant or represent that any license, either express or implied, is granted under any patent right, copyright, mask work right, or other intellectual property right of TI covering or relating to any combination, machine, or process in which such products or services might be or are used. TI's publication of information regarding any third party's products or services does not constitute TI's approval, license, warranty or endorsement thereof.

Reproduction of information in TI data books or data sheets is permissible only if reproduction is without alteration and is accompanied by all associated warranties, conditions, limitations and notices. Representation or reproduction of this information with alteration voids all warranties provided for an associated TI product or service, is an unfair and deceptive business practice, and TI is not responsible nor liable for any such use.

Resale of TI's products or services with <u>statements different from or beyond the parameters</u> stated by TI for that product or service voids all express and any implied warranties for the associated TI product or service, is an unfair and deceptive business practice, and TI is not responsible nor liable for any such use.

Also see: Standard Terms and Conditions of Sale for Semiconductor Products, www.ti.com/sc/docs/stdterms.htm

Mailing Address:

Texas Instruments Post Office Box 655303 Dallas, Texas 75265