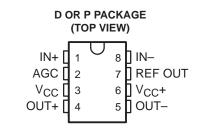
SLFS007A - JUNE 1985 - REVISED JULY 1990

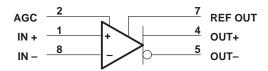
- Low Output Common-Mode Sensitivity to AGC Voltages
- Input and Output Impedances Independent of AGC Voltage
- Peak Gain . . . 38 dB Typ
- Wide AGC Range . . . 50 dB Typ
- 3-dB Bandwidth . . . 50 MHz
- Other Characteristics Similar to NE592 and uA733

#### description

This device is a monolithic two-stage highfrequency amplifier with differential inputs and outputs.



#### symbol



Internal feedback provides wide bandwidth, low phase distortion, and excellent gain stability. Variable gain based on signal summation provides large AGC control over a wide bandwidth with low harmonic distortion. Emitter-follower outputs enable the device to drive capacitive loads. All stages are current-source biased to obtain high common-mode and supply-voltage rejection ratios. The gain may be electronically attenuated by applying a control voltage to the AGC pin. No external compensation components are required.

This device is particularly useful in TV and radio IF and RF AGC circuits, as well as magnetic-tape and disk-file systems where AGC is needed. Other applications include video and pulse amplifiers where a large AGC range, wide bandwidth, low phase shift, and excellent gain stability are required.

The TL026C is characterized for operation from 0°C to 70°C.

#### absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Supply voltage, V <sub>CC+</sub> (see Note 1)	8 V
Supply voltage, V <sub>CC</sub> (see Note 1)	
Differential input voltage	
Common-mode input voltage	±6 V
Output current	±10 mA
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	– 65°C to 150°C
Lead temperature range 1.6 mm (1/16 inch) from case for 10 seconds	260°C

<sup>†</sup> Stresses beyond those listed under absolute maximum ratings may cause permanent damage to the device. This is a stress rating only, and functional operation of the device at these or any other conditions beyond those indicated in the recommended operating conditions section of this specification is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTE 1: All voltages are with respect to the midpoint of V<sub>CC+</sub> and V<sub>CC-</sub> except differential input and output voltages.

#### **DISSIPATION RATING TABLE**

PACKAGE	$T_{\mbox{A}} \le 25^{\circ}\mbox{C}$ POWER RATING	OPERATING FACTOR ABOVE T <sub>A</sub> = 25°C	T <sub>A</sub> = 70°C POWER RATING			
D	725 mW	5.8 mW/°C	464 mW			
Р	1000 mW	8.0 mW/°C	640 mW			



### TL026C DIFFERENTIAL HIGH-FREQUENCY AMPLIFIER WITH AGC

SLFS007A - JUNE 1985 - REVISED JULY 1990

#### recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, V <sub>CC +</sub>	3	6	8	V
Supply voltage, V <sub>CC</sub> _	-3	-6	-8	V
Operating free-air temperature range, TA	0		70	°C

# electrical characteristics at 25°C operating free-air temperature, $V_{CC+}$ = $\pm 6$ V, $V_{AGC}$ = 0, REF OUT pin open (unless otherwise specified)

	PARAMETER	FIGURE	TEST CONDITIONS	MIN	TYP	MAX	UNIT
A <sub>VD</sub>	Large-signal differential voltage amplification	1	$V_{O(PP)} = 3 \text{ V}, \qquad R_{L} = 2 \text{ k}\Omega$	65	85	105	V/V
ΔA <sub>VD</sub>	Change in voltage amplification	1	$V_{IPP}$ = 28.5 mV, $R_L$ = 2 k $\Omega$ , $V_{AGC}$ - $V_{ref}$ = ±180 mV		- 50		dB
V <sub>ref</sub>	Voltage at REF OUT		$I_{ref} = -1$ mA to 100 $\mu$ A	1.3		1.5	V
BW	Bandwidth (-3 dB)	2	$V_{O(PP)} = 1 \text{ V},$ $V_{AGC} - V_{ref} = \pm 180 \text{ mV}$		50		MHz
I <sub>IO</sub>	Input offset current				0.4	5	μΑ
I <sub>IB</sub>	Input bias current				10	30	μΑ
VICR	Common-mode input voltage range	3		±1			V
Voc	Common-mode output voltage	1	R <sub>L</sub> = ∞	3.25	3.75	4.25	V
$\Delta V_{OC}$	Change in common-mode output voltage	1	$V_{AGC} = 0 \text{ to } 2 \text{ V},  R_{L} = \infty$			300	mV
V <sub>00</sub>	Output offset voltage	1	$V_{ID} = 0,$ $R_L = \infty$			0.75	V
VO(PP)	Maximum peak-to-peak output voltage swing	1	$R_L = 2 k\Omega$	3	4		٧
rį	Input resistance at AGC, IN+, or IN -			10	30		kΩ
r <sub>O</sub>	Output resistance				20		Ω
CMRR	Common mode rejection retio	3	$V_{IC} = \pm 1 \text{ V}, \qquad f = 100 \text{ kHz}$	60	86		dB
CIVIKK	Common-mode rejection ratio	3	$V_{IC} = \pm 1 \text{ V}, \qquad f = 5 \text{ mHz}$		60		uБ
ksvr	Supply voltage rejection ratio (ΔV <sub>CC</sub> / ΔV <sub>IO</sub> )	4	$\Delta V_{CC+} = \pm 0.5 \text{ V},$ $\Delta V_{CC-} = \pm 0.5 \text{ V}$	50	70		dB
Vn	Broadband equivalent noise voltage	4	BW = 1 kHz to 10 MHz		12		μV
t <sub>pd</sub>	Propagation delay time	2	$\Delta V_{O} = 1 V$		6	10	ns
t <sub>r</sub>	Rise time	2	$\Delta V_{O} = 1 V$		4.5	12	ns
I <sub>sink(max)</sub>	Maximum output sink current		$V_{ID} = 1 \text{ V}, \qquad V_{O} = 3 \text{ V}$	3	4		mA
Icc	Supply current		No load, No signal		22	27	mA



## electrical characteristics over recommended operating free-air temperature range, $V_{CC\pm}$ = $\pm 6$ V, $V_{AGC}$ = 0, REF OUT pin open (unless otherwise specified)

	PARAMETER	FIGURE	TEST CONDITIONS	MIN	TYP	MAX	UNIT
AVD	Large-signal differential voltage amplification	1	$V_{O(PP)} = 3 \text{ V},  R_L = 2 \text{ k}\Omega$	55		115	V/V
IIO	Input offset current					6	μΑ
I <sub>IB</sub>	Input bias current					40	μΑ
VICR	Common-mode input voltage range	3		±1			V
Voo	Output offset voltage	1	$V_{ID} = 0,$ $R_L = \infty$			1.5	V
V <sub>O(PP)</sub>	Maximum peak-to-peak output voltage swing	1	$R_L = 2 k\Omega$	2.8			V
rį	Input resistance at AGC, IN+, or IN -			8			kΩ
CMRR	Common-mode rejection ratio	3	$V_{IC} = \pm 1 \text{ V}, \qquad f = 100 \text{ kHz}$	50			dB
ksvr	Supply voltage rejection ratio $(\Delta V_{CC} / \Delta V_{IO})$	4	$\Delta V_{CC} + = \pm 0.5 \text{ V},$ $\Delta V_{CC} - = \pm 0.5 \text{ V}$	50			dB
I <sub>sink(max)</sub>	Maximum output sink current		$V_{ID} = 1 \text{ V}, \qquad V_{O} = 3 \text{ V}$	2.8	4		mA
ICC	Supply current	1	No load, No signal			30	mA

#### PARAMETER MEASUREMENT INFORMATION

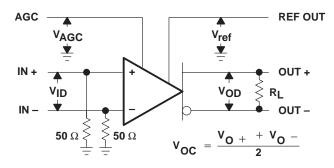


Figure 1. Test Circuit

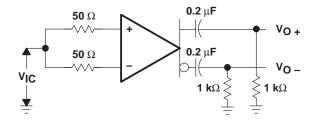


Figure 3. Test Circuit

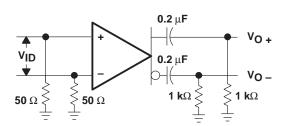


Figure 2. Test Circuit

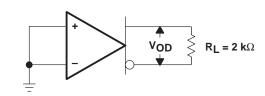


Figure 4. Test Circuit

#### TYPICAL CHARACTERISTICS

# DIFFERENTIAL VOLTAGE AMPLIFICATION vs DIFFERENTIAL GAIN-CONTROL VOLTAGE

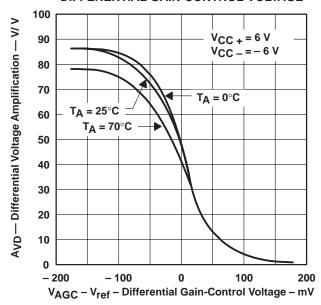


Figure 5

#### gain characteristics

Figure 5 shows the differential voltage amplification versus the differential gain-control voltage ( $V_{AGC} - V_{ref}$ ).  $V_{AGC}$  is the absolute voltage applied to the  $A_{GC}$  input and  $V_{ref}$  is the dc voltage at the REF OUT output. As  $V_{AGC}$  increases with respect to  $V_{ref}$ , the TL026C gain changes from maximum to minimum. As shown in Figure 5 for example,  $V_{AGC}$  would have to vary from approximately 180 mV less than  $V_{ref}$  to approximately 180 mV greater than  $V_{ref}$  to change the gain from maximum to minimum. The total signal change in  $V_{AGC}$  is defined by the following equation.

$$\Delta V_{AGC} = V_{ref} + 180 \text{ mV} - (V_{ref} - 180 \text{ mV})$$

$$\Delta V_{AGC} = 360 \text{ mV}$$
(1)

However, because  $V_{AGC}$  varies as the ac AGC signal varies and also differentially around  $V_{ref}$ , then  $V_{AGC}$  should have an ac signal component and a dc component. To preserve the dc and thermal tracking of the device, this dc voltage must be generated from  $V_{ref}$ . To apply proper bias to the AGC input, the external circuit used to generate  $V_{AGC}$  must combine these two voltages. Figures 6 and 7 show two circuits that will perform this operation and are easy to implement. The circuits use a standard dual operational amplifier for AGC feedback. By providing rectification and the required feedback gain, these circuits are also complete AGC systems.

#### circuit operation

Amplifier A1 amplifies and inverts the rectified and filtered AGC signal voltage  $V_C$  producing output voltage V1. Amplifier A2 is a differential amplifier that inverts V1 again and adds the scaled  $V_{ref}$  voltage. This conditioning makes  $V_{AGC}$  the sum of the signal plus the scaled  $V_{ref}$ . As the signal voltage increases,  $V_{AGC}$  increases and the gain of the TL026C is reduced. This maintains a constant output level.

#### feedback circuit equations

Following the AGC input signal (Figures 6 and 7) from the OUT output through the feedback amplifiers to the AGC input produces the following equations:

1. AC ouput to diode D1, assuming sinusoidal signals

$$V_{O} = V_{OP} (\sin (wt))$$
(2)

where:

 $V_{OP}$  = peak voltage of  $V_{O}$ 

2. Diode D1 and capacitor C1 output

$$V_{C} = V_{OP} - V_{F}$$
 (3)

where

VF = forward voltage drop of D1

V<sub>C</sub> = voltage across capacitor C1

3. A1 output

$$V1 = -\frac{R2}{R1} V_C \tag{4}$$

4. A2 output (R3 = R4)

$$V_{AGC} = \frac{R2}{R1} V_C + 2 \frac{R6}{R5 + R6} V_{ref}$$
 (5)



Amplifier A2 inverts V1 producing a positive AGC signal voltage. Therefore, the input voltage to the TL026C AGC pin consists of an AGC signal equal to:

$$\frac{R2}{R1}$$
 V<sub>C</sub> (6)

and a dc voltage derived from V<sub>ref</sub>, defined as the quiescent value of V<sub>AGC</sub>.

$$V_{AGC}(q) = 2\frac{R6}{R5 + R6}V_{ref}$$
 (7)

For the initial resistor calculations,  $V_{ref}$  is assumed to be typically 1.4 V making quiescent  $V_{AGC}$  approximately 1.22 V ( $V_{AGC}(q) = V_{ref} - 180$  mV). This voltage allows the TL026C to operate at maximum gain under no-signal and low-signal conditions. In addition, with  $V_{ref}$  used as both internal and external reference, its variation from device to device automatically adjusts the overall bias and makes AGC operation essentially independent of the absolute value of  $V_{ref}$ . The resistor divider needs to be calculated only once and is valid for the full tolerance of  $V_{ref}$ .

#### output voltage limits (see Figures 6 and 7)

The output voltage level desired must fall within the following limits:

- 1. Because the data sheet minimum output swing is 3 V peak-to-peak using a  $2-k\Omega$  load resistor, the user-selected design limit for the peak output swing should not exceed 1.5 V.
- The voltage drop of the rectifying diode determines the lower voltage limit. When a silicon diode is
  used, this voltage is approximately 0.7 V. The output voltage V<sub>O</sub> must have sufficient amplitude to
  exceed the rectifying diode drop. Aschottky diode can be used to reduce the V<sub>O</sub> level required.

#### gain calculations for a peak output voltage of 1 V

A peak output voltage of 1 V was chosen for gain calculations because it is approximately midway between the limits of conditions 1 and 2 in the preceding paragraph.

Using equation 3 ( $V_C = V_{OP} - V_d$ ),  $V_C$  is calculated as follows:

$$V_C = 1 V - 0.7 V$$

$$V_{C} = 0.3 V$$

Therefore, the gain of A1 must produce a voltage V1 that is equal to or greater than the total change in  $V_{AGC}$  for maximum TL026C gain change.

With a total change in V<sub>AGC</sub> of 360 mV and using equation 4, the calculation is as follows:

$$-\frac{V1}{V_C} = \frac{\Delta V_{AGC}}{V_C} = \frac{R2}{R1} = \frac{0.36}{0.3} = 1.2$$

If R1 is 10 k $\Omega$ , R2 is 1.2 time R1 or 12 k $\Omega$ .

Since the output voltage for this circuit must be between 0.85 V and 1.3 V, the component values in Figures 6 and 7 provide a nominal 1-V peak output limit. This limit is the best choice to allow for temperature variations of the diode and minimum output voltage specification.



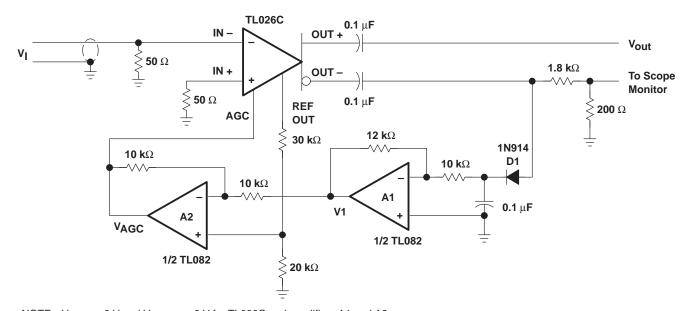
The circuit values in Figures 6 and 7 will produce the best results in this general application. Because of rectification and device input constraints, the circuit in Figure 6 will not provide attenuation and has about 32 dB of control range. The circuit shown in Figure 7 will have approximately 25% variation in the peak output voltage limit due to the variation in gain of the TL592 device to device. In addition, if a lower output voltage is desired, the output of the TL026C can be used for approximately 40 mV of controlled signal.

#### considerations for the use of the TL026C

To obtain the most reliable results, RF breadboarding techniques must be used. A groundplane board should be used and power supplies should be bypassed with 0.1-µF capacitors. Input leads and output leads should be as short as possible and separated from each other.

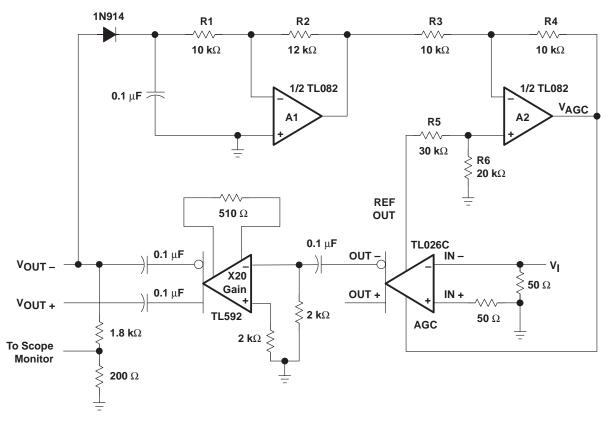
A peak input voltage greater than 200 mV will begin to saturate the input stages of the TL026C and, while the circuit is in the AGC mode, the output signal may become distorted.

To observe the output signal of TL026C or TL592, low-capacitance FET probes or the output voltage divider technique shown in Figure 6 should be used.



NOTE:  $V_{CC+} = 6 \text{ V}$  and  $V_{CC-} = -6 \text{ V}$  for TL026C and amplifiers A1 and A2.

Figure 6. Typical Application Circuit With No Attenuation



NOTE:  $V_{CC} + = 6 \text{ V}$  and  $V_{CC} - = -6 \text{ V}$  for TL026C and amplifiers A1 and A2.

Figure 7. Typical Application Circuit With Attenuation





11-Apr-2013

#### PACKAGING INFORMATION

Orderable Device	Status	Package Type	Package Drawing	Pins	Package Qty		Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Top-Side Markings	Samples
	(1)				-	(2)		(3)		(4)	
TL026CD	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	0 to 70	TL026C	Samples
TL026CDE4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	0 to 70	TL026C	Samples
TL026CDG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	0 to 70	TL026C	Samples
TL026CDR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	0 to 70	TL026C	Samples
TL026CDRE4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	0 to 70	TL026C	Samples
TL026CDRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	0 to 70	TL026C	Samples
TL026CP	ACTIVE	PDIP	Р	8	50	Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type	0 to 70	TL026CP	Samples
TL026CPE4	ACTIVE	PDIP	Р	8	50	Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type	0 to 70	TL026CP	Samples
TL026CPSR	ACTIVE	SO	PS	8	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	0 to 70	T026	Samples
TL026CPSRE4	ACTIVE	SO	PS	8	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	0 to 70	T026	Samples
TL026CPSRG4	ACTIVE	SO	PS	8	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	0 to 70	T026	Samples
TL026ID	OBSOLETE	E SOIC	D	8		TBD	Call TI	Call TI			

<sup>(1)</sup> The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free** (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

<sup>(2)</sup> Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.



#### PACKAGE OPTION ADDENDUM

11-Apr-2013

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above. **Green (RoHS & no Sb/Br):** TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) Multiple Top-Side Markings will be inside parentheses. Only one Top-Side Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Top-Side Marking for that device.

**Important Information and Disclaimer:** The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

#### PACKAGE MATERIALS INFORMATION

www.ti.com 26-Jan-2013

#### TAPE AND REEL INFORMATION





A0	Dimension designed to accommodate the component width
	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

#### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



#### \*All dimensions are nominal

ar amonorous are norminar												
Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TL026CDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TL026CPSR	SO	PS	8	2000	330.0	16.4	8.2	6.6	2.5	12.0	16.0	Q1

www.ti.com 26-Jan-2013



#### \*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TL026CDR	SOIC	D	8	2500	340.5	338.1	20.6
TL026CPSR	SO	PS	8	2000	367.0	367.0	38.0

## P (R-PDIP-T8)

#### PLASTIC DUAL-IN-LINE PACKAGE



- A. All linear dimensions are in inches (millimeters).
- B. This drawing is subject to change without notice.
- C. Falls within JEDEC MS-001 variation BA.



## D (R-PDSO-G8)

#### PLASTIC SMALL OUTLINE



- A. All linear dimensions are in inches (millimeters).
- B. This drawing is subject to change without notice.
- Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.006 (0,15) each side.
- Body width does not include interlead flash. Interlead flash shall not exceed 0.017 (0,43) each side.
- E. Reference JEDEC MS-012 variation AA.



## D (R-PDSO-G8)

## PLASTIC SMALL OUTLINE



- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
- E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.





NOTES: A. All linear dimensions are in millimeters.

B. This drawing is subject to change without notice.

C. Body dimensions do not include mold flash or protrusion, not to exceed 0,15.



## PS (R-PDSO-G8)

#### PLASTIC SMALL OUTLINE



- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
- E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.



#### IMPORTANT NOTICE

Texas Instruments Incorporated and its subsidiaries (TI) reserve the right to make corrections, enhancements, improvements and other changes to its semiconductor products and services per JESD46, latest issue, and to discontinue any product or service per JESD48, latest issue. Buyers should obtain the latest relevant information before placing orders and should verify that such information is current and complete. All semiconductor products (also referred to herein as "components") are sold subject to TI's terms and conditions of sale supplied at the time of order acknowledgment.

TI warrants performance of its components to the specifications applicable at the time of sale, in accordance with the warranty in TI's terms and conditions of sale of semiconductor products. Testing and other quality control techniques are used to the extent TI deems necessary to support this warranty. Except where mandated by applicable law, testing of all parameters of each component is not necessarily performed.

TI assumes no liability for applications assistance or the design of Buyers' products. Buyers are responsible for their products and applications using TI components. To minimize the risks associated with Buyers' products and applications, Buyers should provide adequate design and operating safeguards.

TI does not warrant or represent that any license, either express or implied, is granted under any patent right, copyright, mask work right, or other intellectual property right relating to any combination, machine, or process in which TI components or services are used. Information published by TI regarding third-party products or services does not constitute a license to use such products or services or a warranty or endorsement thereof. Use of such information may require a license from a third party under the patents or other intellectual property of the third party, or a license from TI under the patents or other intellectual property of TI.

Reproduction of significant portions of TI information in TI data books or data sheets is permissible only if reproduction is without alteration and is accompanied by all associated warranties, conditions, limitations, and notices. TI is not responsible or liable for such altered documentation. Information of third parties may be subject to additional restrictions.

Resale of TI components or services with statements different from or beyond the parameters stated by TI for that component or service voids all express and any implied warranties for the associated TI component or service and is an unfair and deceptive business practice. TI is not responsible or liable for any such statements.

Buyer acknowledges and agrees that it is solely responsible for compliance with all legal, regulatory and safety-related requirements concerning its products, and any use of TI components in its applications, notwithstanding any applications-related information or support that may be provided by TI. Buyer represents and agrees that it has all the necessary expertise to create and implement safeguards which anticipate dangerous consequences of failures, monitor failures and their consequences, lessen the likelihood of failures that might cause harm and take appropriate remedial actions. Buyer will fully indemnify TI and its representatives against any damages arising out of the use of any TI components in safety-critical applications.

In some cases, TI components may be promoted specifically to facilitate safety-related applications. With such components, TI's goal is to help enable customers to design and create their own end-product solutions that meet applicable functional safety standards and requirements. Nonetheless, such components are subject to these terms.

No TI components are authorized for use in FDA Class III (or similar life-critical medical equipment) unless authorized officers of the parties have executed a special agreement specifically governing such use.

Only those TI components which TI has specifically designated as military grade or "enhanced plastic" are designed and intended for use in military/aerospace applications or environments. Buyer acknowledges and agrees that any military or aerospace use of TI components which have *not* been so designated is solely at the Buyer's risk, and that Buyer is solely responsible for compliance with all legal and regulatory requirements in connection with such use.

TI has specifically designated certain components as meeting ISO/TS16949 requirements, mainly for automotive use. In any case of use of non-designated products, TI will not be responsible for any failure to meet ISO/TS16949.

#### Products Applications

Audio www.ti.com/audio Automotive and Transportation www.ti.com/automotive Communications and Telecom **Amplifiers** amplifier.ti.com www.ti.com/communications **Data Converters** dataconverter.ti.com Computers and Peripherals www.ti.com/computers **DLP® Products** www.dlp.com Consumer Electronics www.ti.com/consumer-apps

DSP **Energy and Lighting** dsp.ti.com www.ti.com/energy Clocks and Timers www.ti.com/clocks Industrial www.ti.com/industrial Interface interface.ti.com Medical www.ti.com/medical logic.ti.com Logic Security www.ti.com/security

Power Mgmt power.ti.com Space, Avionics and Defense www.ti.com/space-avionics-defense

Microcontrollers microcontroller.ti.com Video and Imaging www.ti.com/video

RFID www.ti-rfid.com

OMAP Applications Processors <a href="www.ti.com/omap">www.ti.com/omap</a> TI E2E Community <a href="e2e.ti.com">e2e.ti.com</a>

Wireless Connectivity <u>www.ti.com/wirelessconnectivity</u>