FA5310BP(S), FA5314P(S), FA5316P(S)
FA5311BP(S), FA5315P(S), FA5317P(S)

■ Description

The FA531X series are bipolar ICs for switching power supply control that can drive a power MOSFET.
These ICs contain many functions in a small 8-pin package. With these ICs, a high-performance and compact power supply can be created because not many external discrete components are needed.

■ Features

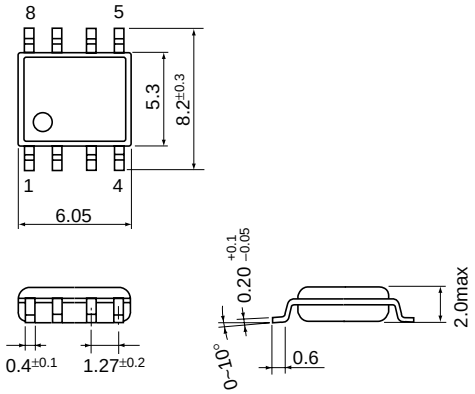
- Drive circuit for connecting a power MOSFET
- Wide operating frequency range (5 to 600kHz)
- Pulse-by-pulse overcurrent limiting function
- Overload cutoff function (Latch or non-protection mode selectable)
- Output ON/OFF control function by external signal
- Overvoltage cutoff function in latch mode
- Undervoltage malfunction prevention function
- Low standby current (90μA typical)
- Exclusive choices by circuits (See selection guide on page 25)
- 8-pin package (DIP/SOP)

■ Applications

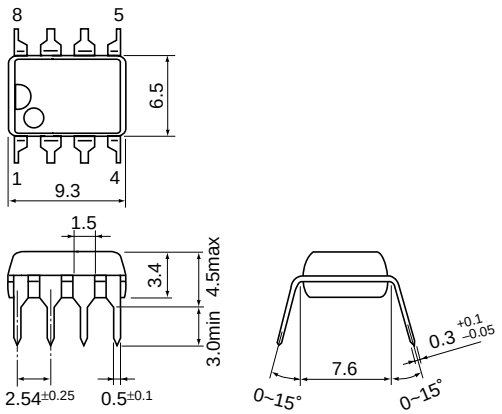
- Switching power supply for general equipment

■ Dimensions, mm

● **SOP-8**

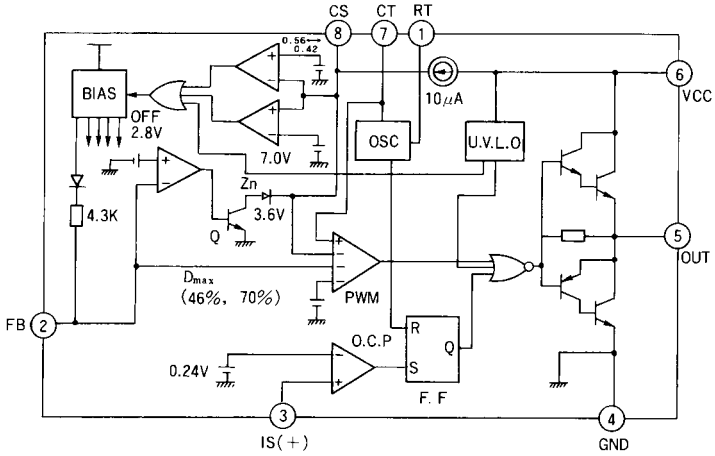


● **DIP-8**



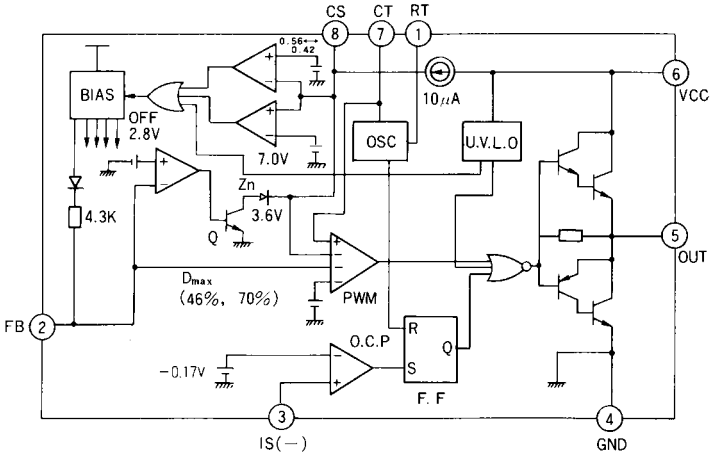
■ Block diagram

● **FA5310BP(S)/FA5311BP(S)/FA5316P(S)/FA5317P(S)**



Pin No.	Pin symbol	Description
1	RT	Oscillator timing resistor
2	FB	Feedback
3	IS (+)	Overcurrent (+) detection
4	GND	Ground
5	OUT	Output
6	VCC	Power supply
7	CT	Oscillator timing capacitor
8	CS	Soft-start and ON/OFF control

● **FA5314P(S)/FA5315P(S)**



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6	VCC	Power supply
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8	CS	Soft-start and ON/OFF control

■ Selection guide

Type	Max. duty cycle (typ.)	Polarity of overcurrent detection	UVLO (typ.)		Max. output current	Application
			ON threshold	OFF threshold		
FA5310BP(S)	46%	+	16.0V	8.70V	1.5A	Forward type
FA5311BP(S)	70%	+	16.0V	8.70V	1.5A	Flyback type
FA5314P(S)	46%	–	15.5V	8.40V	1.5A	Forward type
FA5315P(S)	70%	–	15.5V	8.40V	1.5A	Flyback type
FA5316P(S)	46%	+	15.5V	8.40V	1.0A	Forward type
FA5317P(S)	70%	+	15.5V	8.40V	1.0A	Flyback type

■ Absolute maximum ratings

Item	Symbol	Rating	Unit
Supply voltage	V_{CC}	31	V
Output current FA5310/11/14/15 FA5316/17	I_O	 ±1.5 ±1.0	A
Feedback terminal input voltage	V_{FB}	4	V
Overcurrent detection terminal input voltage	V_{IS}	–0.3 to +4	V
CS terminal input current	I_{CS}	2	mA
Total power dissipation	P_d	800 (DIP-8) *1	mW
(Ta=25°C)		550 (SOP-8) *2	
Operating temperature	T_{opr}	–30 to +85	°C
Junction temperature	T_j	125	°C
Storage temperature	T_{stg}	–40 to +150	°C

Notes:
*1 Derating factor Ta > 25°C : 8.0mW/°C (on PC board)
*2 Derating factor Ta > 25°C : 5.5mW/°C (on PC board)

■ Recommended operating conditions

Item	Symbol	Min.	Max.	Unit
Supply voltage	V_{CC}	10	30	V
Oscillator timing resistance FA5310/11 FA5314/15/16/17	R_T	 3.3 1	 10 10	kΩ
Soft-start capacitor	C_S	0.1	1	μF
Oscillation frequency	f_{OSC}	5	600	kHz

■ Electrical characteristics (Ta=25°C, Vcc=18V, f_{OSC}=135kHz)

Oscillator section

Item	Symbol	Test condition	Min.	Typ.	Max.	Unit
Oscillation frequency	f_{OSC}	R_T=5.1kΩ, C_T=360pF	125	135	145	kHz
Frequency variation 1 (due to supply voltage change)	f_{dV}	V_{CC}=10 to 30V		±1		%
Frequency variation 1 (due to temperature change)	f_{dr}	Ta=–30 to +85°C		±1.5		%

Pulse width modulation circuit section

Item	Symbol	Test condition	FA5310/14/16			FA5311/15/17			Unit
			Min.	Typ.	Max.	Min.	Typ.	Max.	
Feedback terminal source current	I_{FB}	V_{FB}=0	–660	–800	–960	–660	–800	–960	μA
Input threshold voltage (Pin 2)	$V_{TH\ FBO}$	Duty cycle =0%		0.75			0.75		V
	$V_{TH\ FBM}$	Duty cycle =D_{MAX}		1.80			2.30		V
Maximum duty cycle	D_{MAX}		43	46	49	66	70	74	%

Soft-start circuit section

Item	Symbol	Test condition	FA5310/14/16			FA5311/15/17			Unit
			Min.	Typ.	Max.	Min.	Typ.	Max.	
Charge current (Pin 8)	I_{CHG}	Pin 8=0V	–15	–10	–5	–15	–10	–5	μA
Input threshold voltage (Pin 8)	$V_{TH\ CSO}$	Duty cycle =0%		0.90			0.90		V
	$V_{TH\ CSM}$	Duty cycle =D_{MAX}		1.90			2.30		V

Overcurrent limiting circuit section

Item	Symbol	Test condition	FA5310/11/16/17			FA5314/15			Unit
			Min.	Typ.	Max.	Min.	Typ.	Max.	
Input threshold voltage (Pin 3)	$V_{TH\ IS}$		0.21	0.24	0.27	–0.21	–0.17	–0.14	V
Overcurrent detection terminal source current	I_{IS}	Pin 3=0V	–300	–200	–100	–240	–160	–80	µA
Delay time	$T_{PD\ IS}$			150			200		ns

Latch-mode cutoff circuit section

Item	Symbol	Test condition	Min.	Typ.	Max.	Unit
CS terminal sink current	$I_{SINK\ CS}$	Pin 8=6V, Pin 2=1V	25	45	65	µA
Cutoff threshold voltage (Pin 8)	$V_{TH\ CS}$		6.5	7.0	7.5	V

Overload cutoff circuit section

Item	Symbol	Test condition	Min.	Typ.	Max.	Unit
Cutoff-start voltage (Pin 2)	$V_{TH\ FB}$		2.6	2.8	3.1	V

Undervoltage lockout circuit section

Item	Symbol	Test condition	FA5310/11			FA5314/15/16/17			Unit
			Min.	Typ.	Max.	Min.	Typ.	Max.	
OFF-to-ON threshold voltage	$V_{CC\ ON}$		15.5	16.0	16.5	14.8	15.5	16.2	V
ON-to-OFF threshold voltage	$V_{CC\ OFF}$		8.20	8.70	9.20	7.70	8.40	9.10	V

Output section

Item	Symbol	Test condition		Min.	Typ.	Max.	Unit
		FA5310/11/14/15	FA5316/17				
L-level output Voltage	V_{OL}	I_O=100mA	I_O=50mA		1.30	1.80	V
H-level output Voltage	V_{OH}	I_O=–100mA V_{CC}=18V	I_O=–50mA V_{CC}=18V	16.0	16.5		V
Rise time	tr	No load	No load		50		ns
Fall time	tf	No load	No load		50		ns

Output ON/OFF circuit section

Item	Symbol	Test condition	Min.	Typ.	Max.	Unit
CS terminal source current	$I_{source\ cs}$	Pin 8=0V	–15	–10	–5	µA
OFF-to-ON threshold Voltage (Pin 8)	$V_{TH\ ON}$	CS terminal voltage OFF→ON		0.56		V
ON-to-OFF threshold Voltage (Pin 8)	$V_{TH\ OFF}$	CS terminal voltage ON→OFF		0.42		V

Overall device

Item	Symbol	Test condition	Min.	Typ.	Max.	Unit
Standby current	$I_{CC\ ST}$	V_{CC}=14V		90	150	µA
Operating-state supply current	$I_{CC\ OP}$			9	15	mA
OFF-state supply current	$I_{CC\ OFF}$			1.1	1.8	mA
Cutoff-state supply current	I_{CCL}			1.1	1.8	mA

■ Description of each circuit

1. Oscillator (See block diagram)
The oscillator generates a triangular waveform by charging and discharging a capacitor. CT pin voltage oscillates between an upper limit of approx. 3.0V and a lower limit of approx. 1.0V. The oscillation frequency is determined by a external resistance and capacitance shown in figure 1, and approximately given by the following equation:

$$f\ (kHz) = \frac{10^6}{4R_T\ (k\Omega) \bullet C_T\ (pF)} \quad .....(1)$$

The recommended oscillation range is between 5k and 600kHz.
The oscillator output is connected to a PWM comparator.

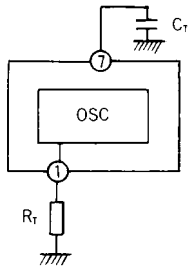


Fig. 1 Oscillator

2. Feedback pin circuit
Figure 2 gives an example of connection in which an optocoupler is used to couple the feedback signal to the FB pin. It is designed to be strong against noise and will not create parasitic oscillation so much, because the output impedance at the FB pin is as low as 4k to 5k. If this circuit causes power supply instability, the frequency gain can be decreased by connecting R_4 and C_4 as shown in figure 2. R_4 should be between several tens of ohms to several kiloohms and C_4 should be between several thousand picofarads to one microfarads.

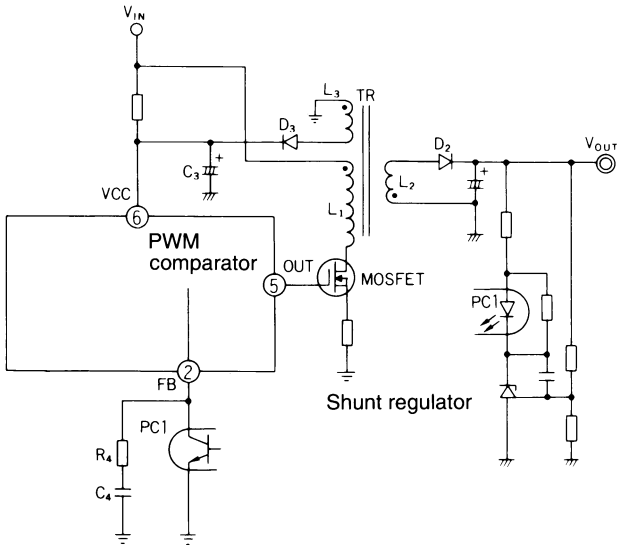


Fig. 2 Configuration with optocoupler (FB pin input)

3. PWM comparator
The PWM comparator has four inputs as shown in Figure 3. Oscillator output ① is compared with CS pin voltage ②, FB pin voltage ③, and DT voltage ④. The lowest of three inputs ②, ③, and ④ is compared with output ①. If it is lower than the oscillator output, the PWM comparator output is high, and if it is higher than the oscillator output, the PWM comparator output is low (see Fig. 4).
The IC output voltage is high during when the comparator output is low, and the IC output voltage is low during when the comparator output is high.
When the IC is powered up, CS pin voltage ② controls soft start operation. The output pulse then begins to widen gradually. During normal operation, the output pulse width is determined within the maximum duty cycle set by DT voltage ④ under the condition set by feedback signal ③, to stabilize the output voltage.

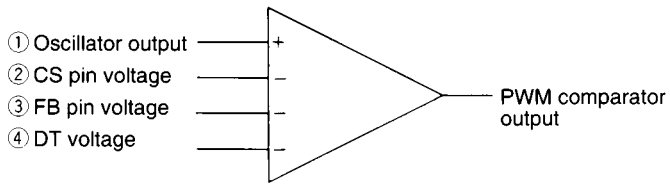


Fig. 3 PWM comparator

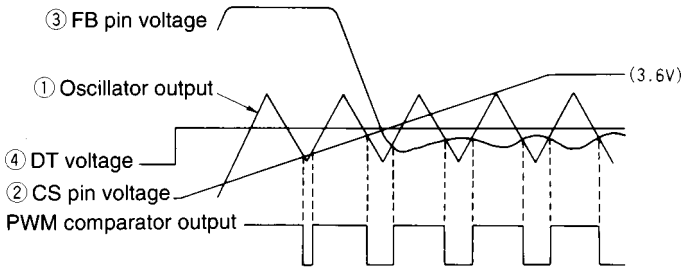


Fig. 4 PWM comparator timing chart

4. CS pin circuit

As shown in Figure 5 capacitor Cs is connected to the CS pin. When power is turned on, the constant current source (10μA) begins to charge capacitor Cs. Accordingly, the CS pin voltage rises as shown in Figure 6. The CS pin is connected to an input of the PWM comparator. The device is in soft-start mode while the CS pin voltage is between 0.9V and 1.9V (FA5310/14/16) and between 0.9V and 2.3V(FA5311/15/17). During normal operation, the CS pin is clamped at 3.6V by internal zener diode Zn. If the output voltage drops due to an overload, etc., the clamp voltage shifts from 3.6V to 8.0V. As a result, the CS pin voltage rises to 8.0V. The CS pin is also connected to latch comparator C2. If the pin voltage rises above 7.0V, the output of comparator C2 goes high to turn off the bias circuit, thereby shutting the output down. Comparator C2 can be used not only for shutdown in response to an overload, but also for shutdown in response to an overvoltage. Comparator C1 is also connected to the CS pin, and the bias circuit is turned off and the output is shut down if the CS pin voltage drops below 0.42V. In this way, comparator C1 can also be used for output on/off control.

As explained above, the CS pin can be used for soft-start operation, overload and overvoltage output shutdown and output on/off control.

Further details on the four functions of the CS pin are given below.

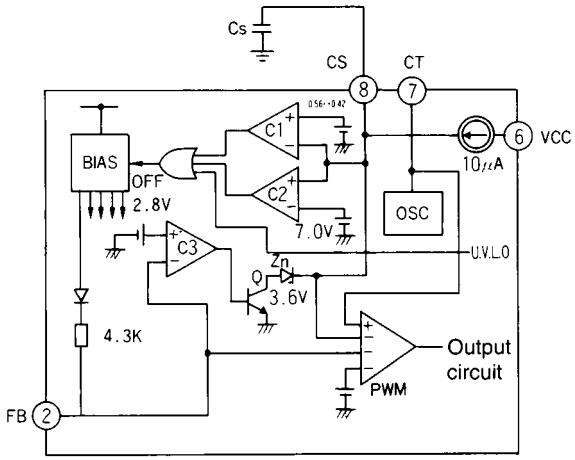


Fig. 5 CS pin circuit

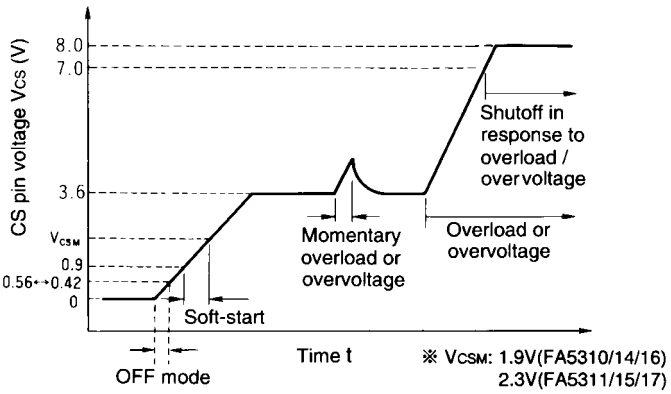


Fig. 6 CS pin waveform

Fig. 6 CS pin waveform

4.1 Soft start function

Figure 7 shows the soft start circuit. Figure 8 is the soft-start operation timing chart. The CS pin is connected to capacitor Cs. When power is turned on, a 10μA constant-current source begins to charge the capacitor. As shown in the timing chart, the CS pin voltage rises slowly in response to the charging current. The CS pin is connected internally to the PWM comparator. The comparator output pulse slowly widens as shown in the timing chart.

The soft start period can be approximately evaluated by the period ts from the time the IC is activated to the time the output pulse width widens to 30%. Period ts is given by the following equation:

ts(ms)=160Cs(μF).....(2)

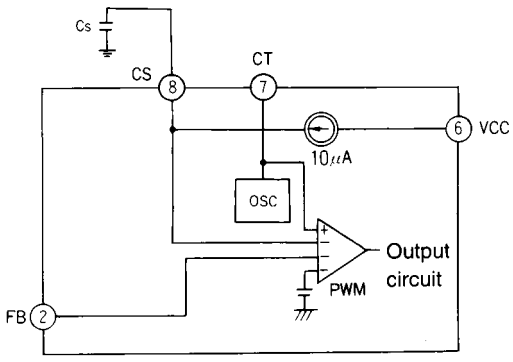


Fig. 7 Soft-start circuit

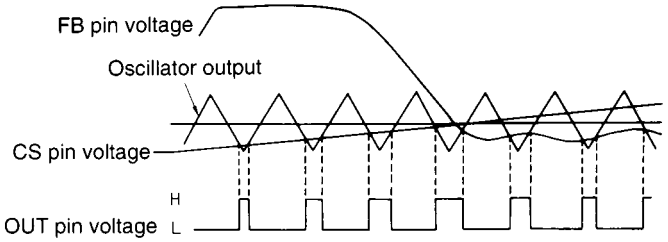


Fig. 8 Soft-start timing chart

4.2 Overload shutdown

Figure 9 shows the overload shutdown circuit, and Figure 10 is a timing chart which illustrates overload shutdown operation.
If the output voltage drops due to an overload or short-circuit, the output voltage of the FB pin rises. If FB pin voltage exceeds the reference voltage (2.8V) of comparator C3, the output of comparator C3 switches low to turn transistor Q off.
In normal operation, transistor Q is on and the CS pin is clamped at 3.6V by zener diode Zn. With Q off, the clamp is released and the 10μA constant-current source begins to charge capacitor C_S again and the CS pin voltage rises. When the CS pin voltage exceeds the reference voltage (7.0V) of comparator C2, the output of comparator C2 switches high to turn the bias circuit off. The IC then enters the latched mode and shuts the output down. Shutdown current consumption is 400μA(V_{CC}=9V). This current must be supplied through the startup resistor. The IC then discharges the MOSFET gates.
Shutdown operation initiated by an overload can be reset by lowering supply voltage V_{CC} below $V_{CC\ OFF}$ or forcing the CS pin voltage below 7.0V.
The period t_{OL} from the time that the output is short-circuited to the time that the bias circuit turns off is given by the following equation:

$$t_{OL}(ms)=340C_S(\mu F) \quad \text{..... ..... (3)}$$

4.3 Overvoltage shutdown

Figure 11 shows the overvoltage shutdown circuit, and Figure 12 is a timing chart which illustrates overvoltage shutdown operation.
The optocoupler PC1 is connected between the CS and V_{CC} pins. If the output voltage rises too high, the PC1 turns on to raise the voltage at the CS pin via resistor R_6. When the CS pin voltage exceeds the reference voltage (7.0V) of comparator C2, comparator C2 switches high to turn the bias circuit off. The IC then enters the latched mode and shuts the output down. The shutdown current consumption of the IC is 400μA(V_{CC}=9V). This current must be applied via startup resistor R_5.
The IC then discharges the MOSFET gates.
The shutdown operation initiated by an overvoltage condition can be reset by lowering supply voltage V_{CC} below $V_{CC\ OFF}$ or forcing the CS pin voltage below 7.0V.
During normal operation, the CS pin is clamped by a 3.6V zener diode with a sink current of 65μA max. Therefore, a current of 65μA or more must be supplied by the optocoupler in order to raise the CS pin voltage above 7.0V.

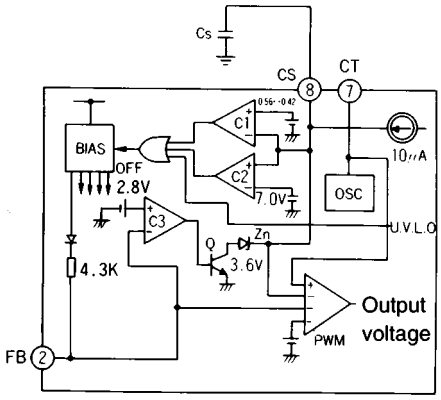


Fig. 9 Overload shutdown circuit

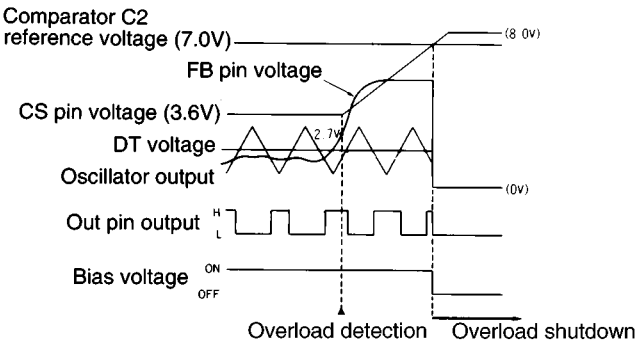


Fig. 10 Overload shutdown timing chart

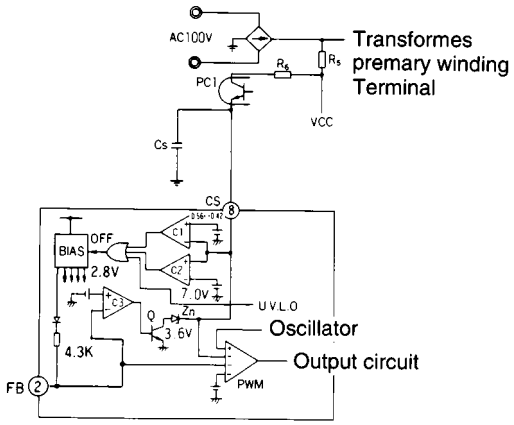


Fig. 11 Overvoltage shutdown circuit

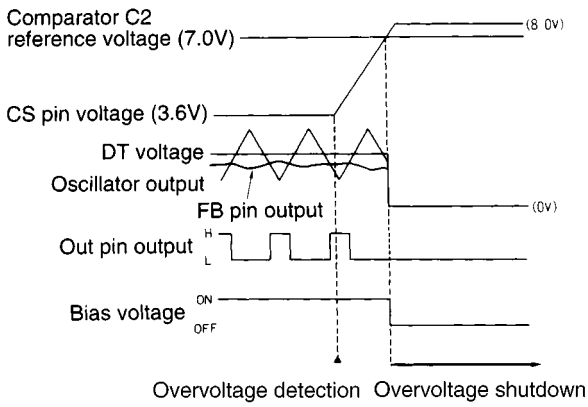


Fig. 12 Overvoltage shutdown timing chart

4.4 Output ON/OFF control

The IC can be turned on and off by an external signal applied to the CS pin.
Figure 13 shows the external output on/off control circuit, and Figure 14 is the timing chart.
The IC is turned off if the CS pin voltage falls below 0.42V. The output of comparator C1 switches high to turn the bias circuit off. This shuts the output down. The IC then discharges the MOSFET gates.
The IC turns on if the CS pin is opened for automatic soft start. The power supply then restarts operation.

5. Overcurrent limiting circuit

The overcurrent limiting circuit detects the peak value of every drain current pulse of the main switching MOSFET to limit the overcurrent.
The detection threshold is + 0.24V for FA5310B/11B/16/17 with respect to ground as shown in Figure 15.
The drain current of the MOSFET is converted to voltage by resistor R_7 and fed to the IS pin of the IC. If the voltage exceeds the reference voltage (0.24V) of comparator C4, the output of comparator C4 goes high to set flip-flop output Q high. The output is immediately turned off to shut off the current. Flip-flop output Q is reset on the next cycle by the output of the oscillator to turn the output on again. This operation is repeated to limit the overcurrent.
If the overcurrent limiting circuit malfunctions due to noise, place an RC filter between the IS pin and the MOSFET.
Figure 16 is a timing chart which illustrates current-limiting operations.

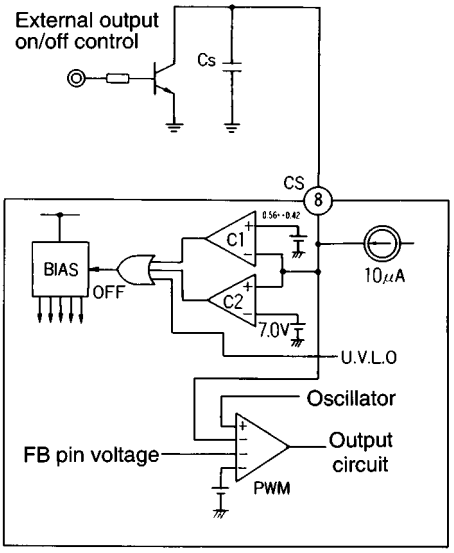


Fig. 13 External output on/off control circuit

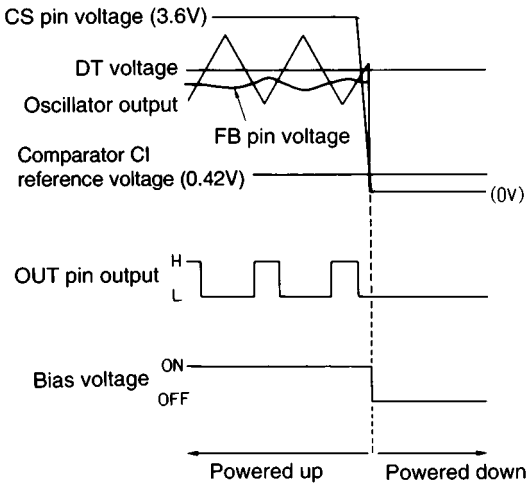


Fig. 14 Timing chart for external output on/off control

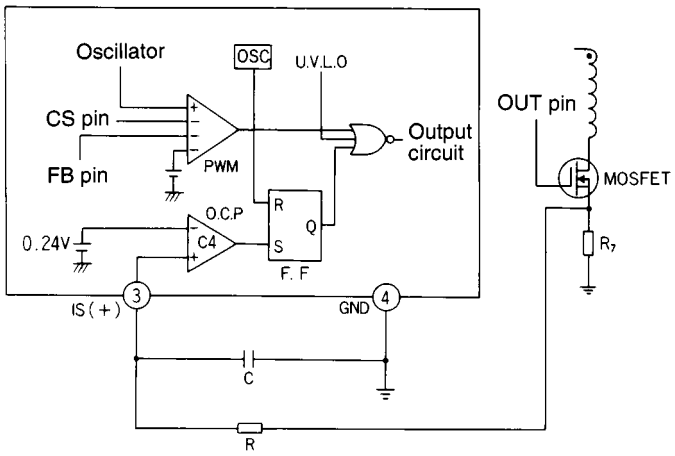


Fig. 15 Overcurrent limiting circuit for FA5310/11/16/17

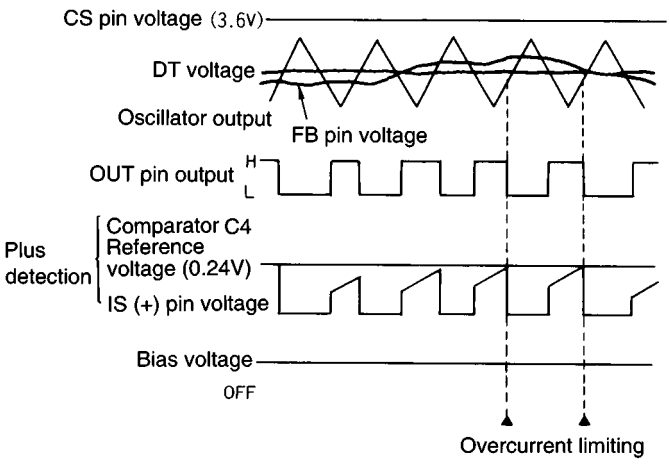


Fig. 16 Overcurrent timing chart for FA5310/11/16/17

The detection threshold is -0.17V for FA5314/15 with respect to ground as shown in Figure 17.
The operation is similar to that of FA5310B/11B/16/17 except the threshold is minus voltage compared to that which is plus voltage for FA5310B/11B/16/17.
Figure 18 is a timing chart which illustrates current limiting operations.

6. Undervoltage lockout circuit
The IC incorporates a circuit which prevents the IC from malfunctioning when the supply voltage drops. When the supply voltage is raised from 0V, the IC starts operation with $V_{CC}=V_{CC\ ON}$.
If the supply voltage drops, the IC shuts its output down when $V_{CC}=V_{CC\ OFF}$. When the undervoltage lockout circuit operates, the CS pin goes low to reset the IC.

7. Output circuit
As shown in Figure 19, the IC's totem-pole output can directly drive the MOSFET. The OUT pin can source and sink currents of up to 1.5A or 1.0A.
If IC operation stops when the undervoltage lockout circuit operates, the gate voltage of the MOSFET goes low and the MOSFET is shut down.

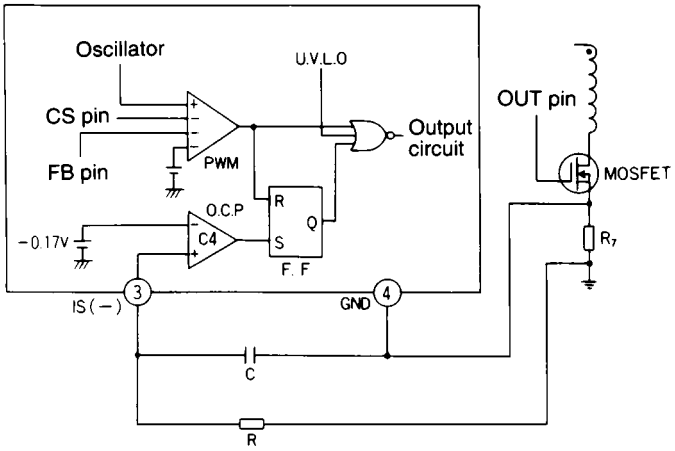


Fig. 17 Overcurrent limiting circuit for FA5314/15

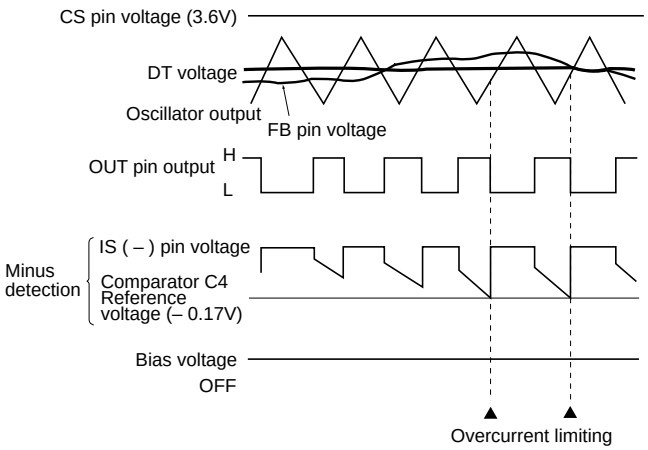


Fig. 18 Overcurrent timing chart for FA5314/15

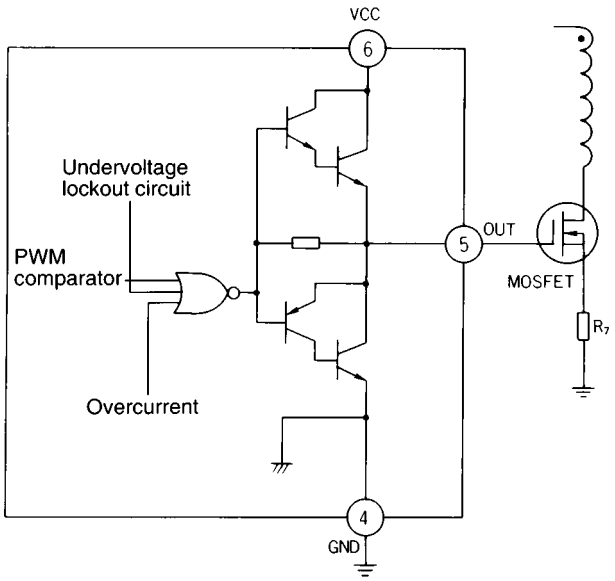


Fig. 19 Output circuit

■ Design advice

1. Startup circuit

It is necessary to start-up IC that the voltage inclination of VCC terminal “dVcc/dt” satisfies the following equation(4).

dVcc/dt(V/s)>1.8/Cs(μF).....(4)
Cs : Capacitor connected between CS terminal and GND

Note that equation (4) must be satisfied in any condition. Also, it is necessary to keep “latch mode” for overload protection or overvoltage protection that the current supplied to VCC terminal through startup resistor satisfies the following equation(5).

Icc(Lat)> 0.4mA for Vcc ≦ 9.2V .....(5)
Icc(Lat) : Cutoff-state(= Latch mode) supply current

The detail is explained as follows.

(1) Startup circuit connected to AC line directly

Fig. 20 shows a typical startup circuit that a startup resistor Rc is connected to AC line directly. The period from power-on to startup is determined by Rc, R_D and C_A. Rc, R_D and C_A must be designed to satisfy the following equations.

dVcc/dt(V/s)=
$(1/C_A) \cdot \{(V_{AVE} - Vccon)/R_C - Vccon/R_D - Iccst\} >$
1.8/(Cs(μF)).....(6)

$Rc(k\Omega) < (V_{AVE} - 9.2(V))/\{0.4\ (mA) + (9.2(V)/R_D(k\Omega)\}$ .....(7)
$V_{AVE} = Vac \cdot \sqrt{2}/\pi$: Average voltage applied to AC line side of Rc
Vac: AC input effective voltage
Vccon: ON threshold of UVLO, 16.5V(max.) or 16.2V(max.)
Iccst: Standby current, 0.15 mA(max.)

In this method, Vcc voltage includes ripple voltage influenced by AC voltage. Therefore, enough dVcc/dt required by equation (6) tend to be achieved easily when Vcc reaches to Vccon even if Vcc goes up very slowly. After power-off, Vcc does not rise up because a voltage applied from bias winding to VCC terminal decreases and the current flowing Rc becomes zero, therefore, re-startup does not occur after Vcc falls down below OFF threshold of UVLO until next power-on.

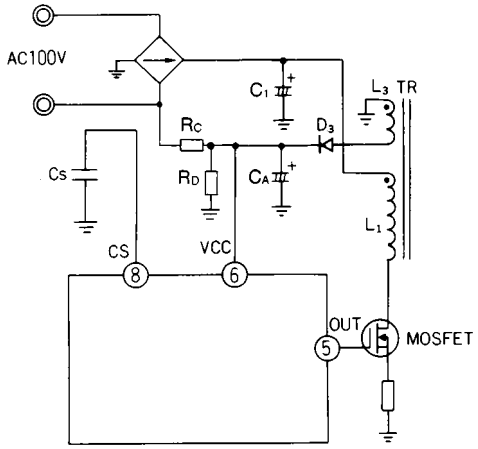


Fig. 20 Startup circuit example(1)

(2) Startup circuit connected to rectified line

This method is not suitable for FA531X, especially concerned with re-startup operation just after power-off or startup which AC input voltage goes up slowly. Fig. 21 shows a startup circuit that a startup resistor R_A is connected to rectified line directly.

The period from power-on to startup is determined by R_A, R_B and C_A. R_A, R_B and C_A must be designed to satisfy the following equations.

$$dVcc/dt(V/s) = (1/C_A)\cdot\{(V_{IN} - Vccon)/R_A - Vccon/R_B - Iccst \} > 1.8/(Cs(\mu F)) \quad \text{.....(8)}$$

$$R_A(k\Omega) < (V_{IN} - 9.2(V))/\{ 0.4(mA) + (9.2(V)/R_B(k\Omega)) \} \quad \text{....(9)}$$

V_{IN} : $\sqrt{2}$ • (AC input effective voltage)

After power-off, once V_{CC} falls down below OFF threshold voltage, V_{CC} rises up again and re-startup occurs while the capacitor C_1 is discharged until approximately zero because V_{CC} voltage rises up by the current flowing R_A.

This operation is repeated several times.

After the repeated operation, IC stops in the condition that V_{CC} voltage is equal to Vccon (=ON threshold) because capacitor C_1 is discharged gradually and the decreased V_{CC} inclination is out of the condition required by equation (4). After that, re-startup by power-on can not be guaranteed even when equation (8) is satisfied.

The image of that the startup is impossible is shown in Fig. 22.

It is necessary to startup IC that supply current Icc(startup) to VCC is over 4mA in the condition of Tj < 100°C during Vcc is kept at Vccon(≒16V, balance state at Vccon after the repeated operation.

Icc(start-up) > 4mA
at Vcc=Vccon, Tj<100°C, after power-off

This balance state that startup is impossible tends to occur at higher temperature. If power-on is done when Vcc is not kept at Vccon (for example:power-off is done and after enough time that C1 is discharged until Vcc can not be pulled up to Vccon), the IC can startup in the condition given by equation(8).

In some cases, such as when the load current of power supply is changed rapidly, you may want to prolong the hold time of the power supply output by means of maintaining Vcc over the off threshold.

For this purpose, connect diode D4 and electrolytic capacitor C4 as shown in Fig. 23. This prolongs the hold time of the power supply voltage Vcc regardless of the period from power-on to startup.

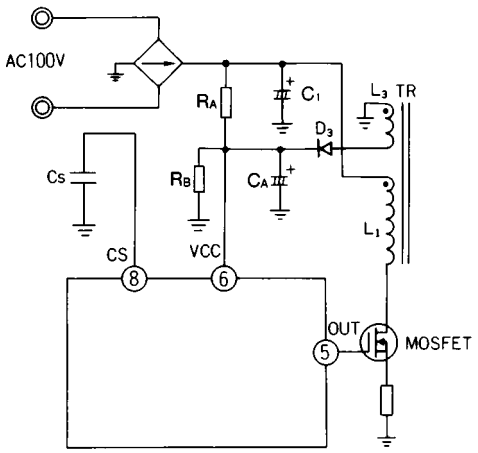


Fig. 21 Startup circuit example(2)

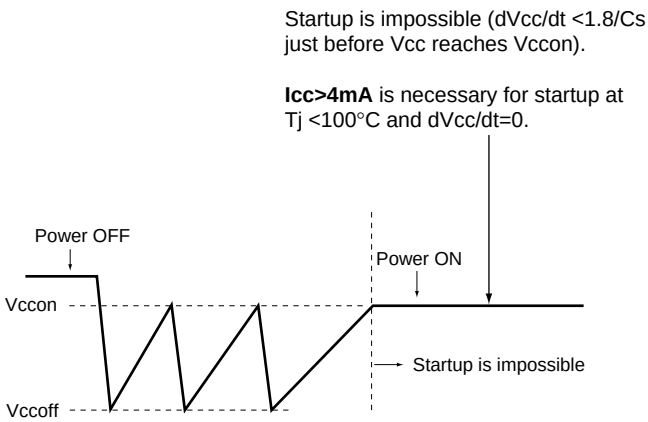


Fig. 22 A image of waveform when re-startup is impossible

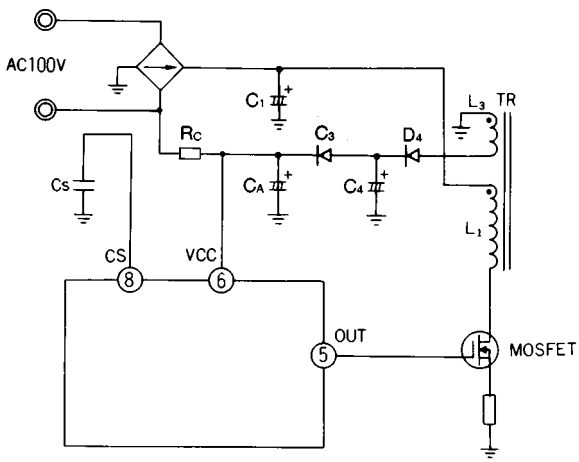


Fig. 23 Startup circuit example(3)

2. Disabling overload shutdown function

As shown in Figure 24, connect a 11kΩ resistor between the FB pin and ground. Then, the CS pin voltage does not rise high enough to reach the reference voltage (7.0V) of the latch comparator, and the IC does not enter the OFF latch mode. With this connection, the overvoltage shutdown function is available.

3. Setting soft start period and OFF latch delay independently

Figure 25 shows a circuit for setting the soft start period and OFF latch delay independently. In this circuit, capacitance C_S determines the soft start period, and capacitance C_L determines the OFF latch delay. If the overload shutdown and overvoltage shutdown functions raise the CS pin voltage to around 5V, zener diode Zn becomes conductive to charge C_L. The OFF latch delay can be thus prolonged by C_L.

4. Laying out Vcc and ground lines

Figure 26 and 27 show the recommended layouts of V_{CC} and ground lines. The bold lines represent paths carrying large currents. The lines must have an adequate thickness.

5.Sink current setting for CS terminal

A sink current to CS terminal must be satisfied the following condition to prevent from the malfunction which uncontrolled pulse output generates at OUT terminal when latch-mode protection should be operated for overvoltage.

65μA < Ics(sink) < 500μA at Vcs= 6.5(V)

Ics(sink) : Sink current to CS terminal

Example (for the circuit shown in Fig. 28)

Ics(sink) = (28(V)–18(V)– 6.5(V))/7.5(kΩ)

≒ 467 (μA) < 500 (μA)

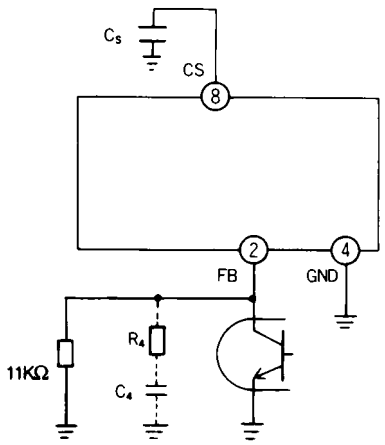


Fig. 24 Disabling overload shutdown function

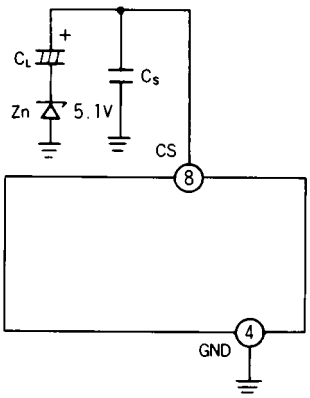


Fig. 25 Independent setting of soft-start period and OFF latch delay

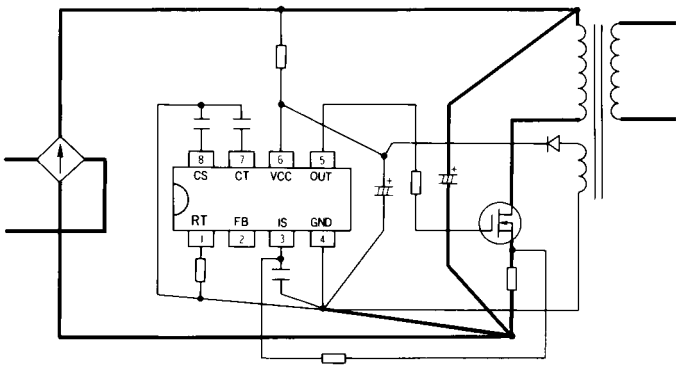


Fig. 26 Vcc line and ground line (1) for FA5310B/11B/16/17

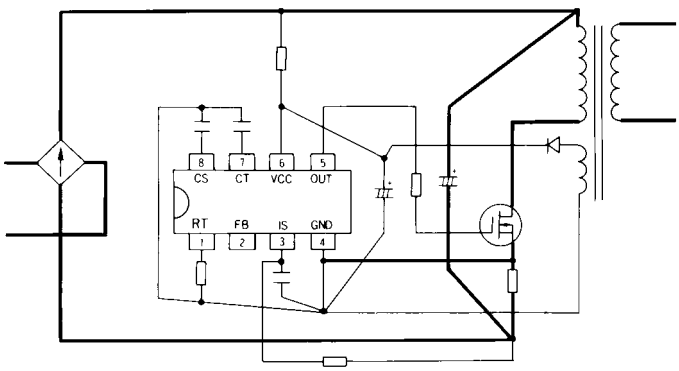


Fig. 27 Vcc line and ground line (2) for FA5314/15

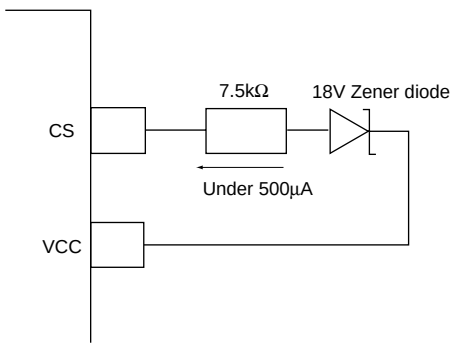
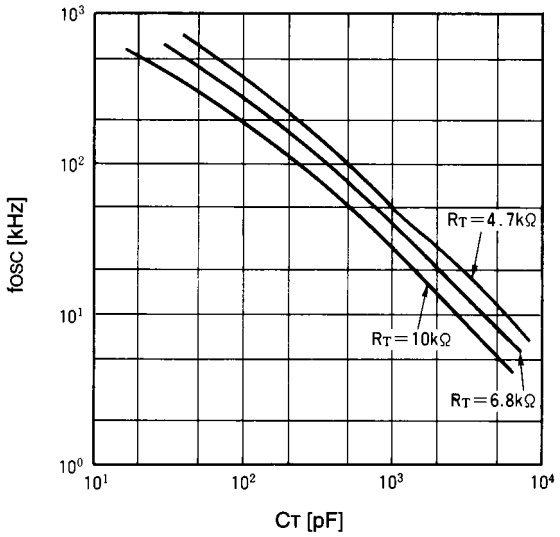


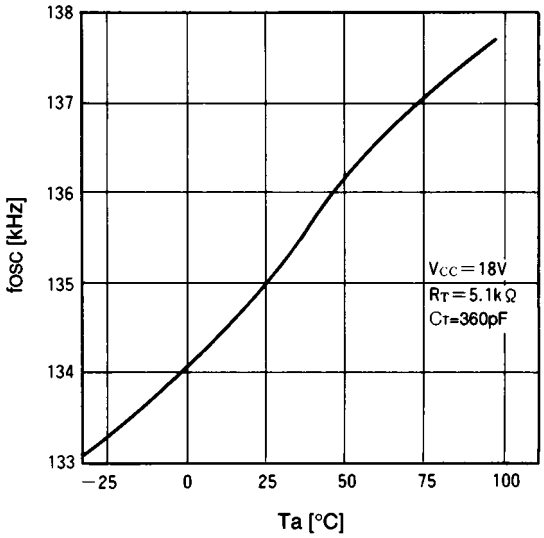
Fig. 28 Setting sink current for CS terminal

■ Characteristic curves (Ta=25°C)

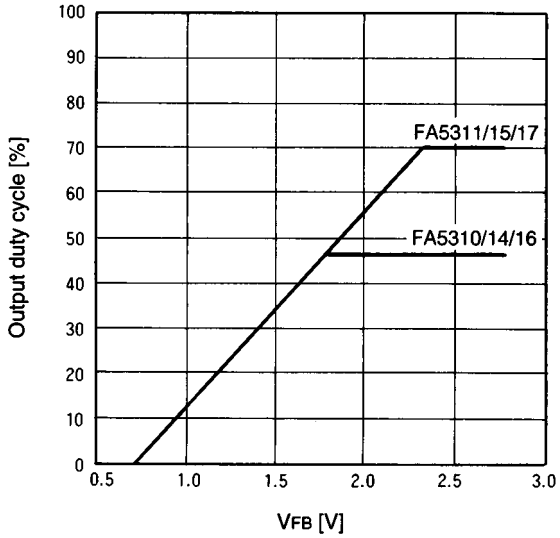
Oscillation frequency (fosc) vs. timing capacitor capacitance (C_T)



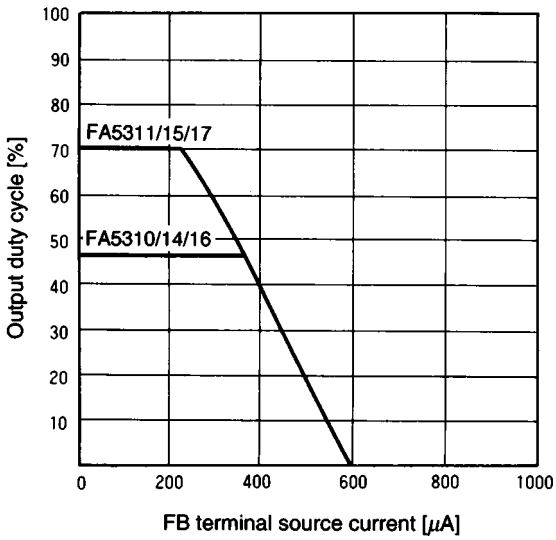
Oscillation frequency (fosc) vs. ambient temperature (Ta)



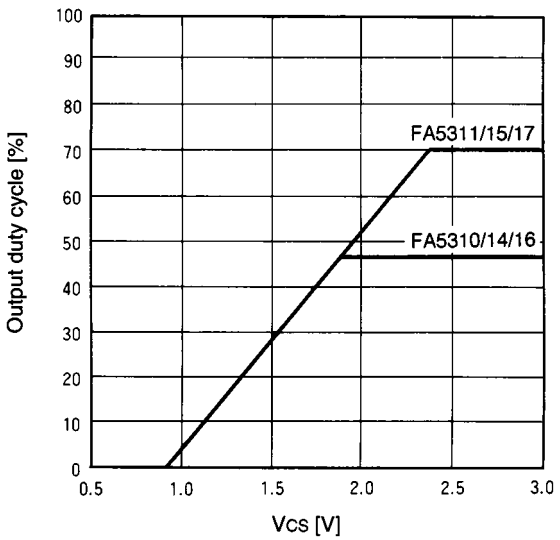
Output duty cycle vs. FB terminal voltage (V_{FB})



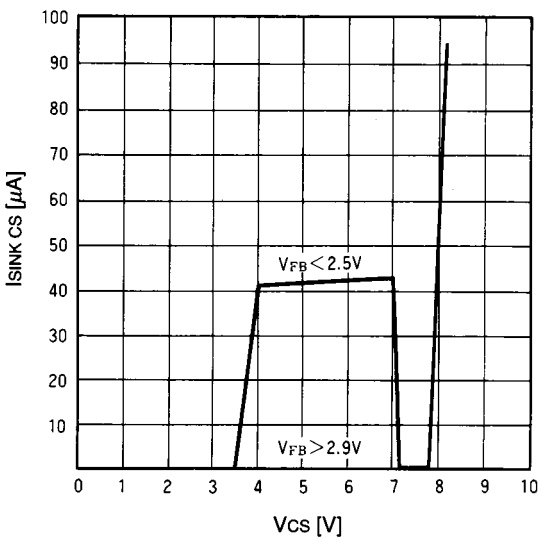
Output duty cycle vs. FB terminal source current (I_{SOURCE})



Output duty cycle vs. CS terminal voltage (V_{CS})

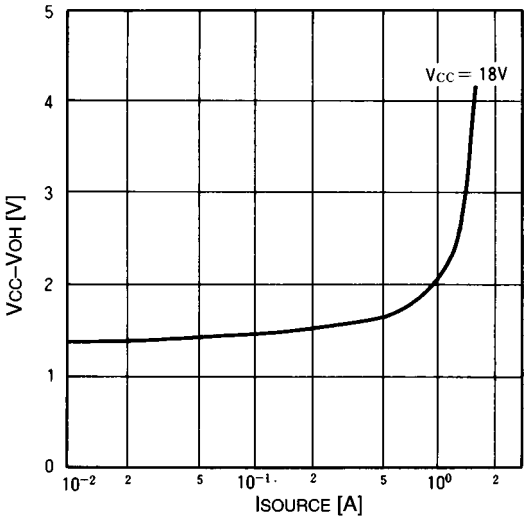


CS terminal sink current ($I_{SINK\ CS}$) vs. CS terminal voltage (V_{CS})

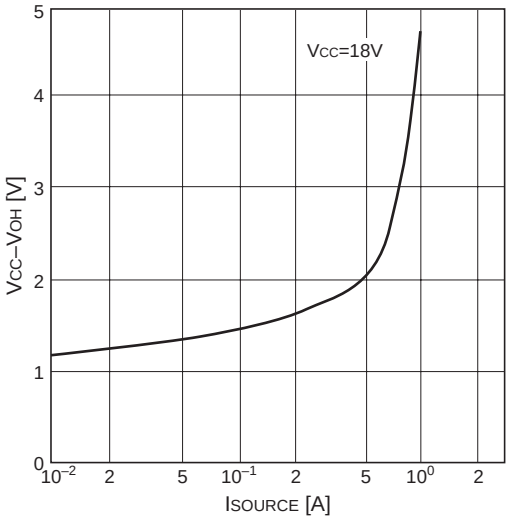


H-level output voltage (V_{OH}) vs. output source current (I_{SOURCE})

FA5310/11/14/15

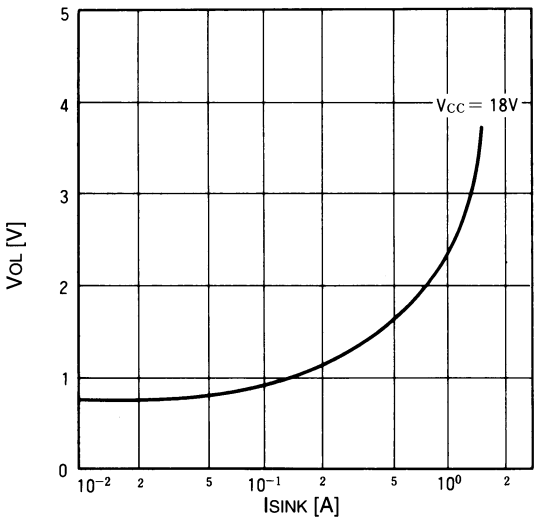


FA5316/17

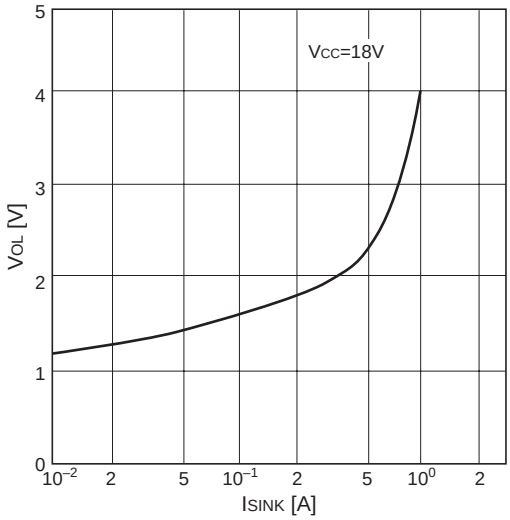


L-level output voltage(V_{OL}) vs. output sink current (I_{SINK})

FA5310/11/14/15

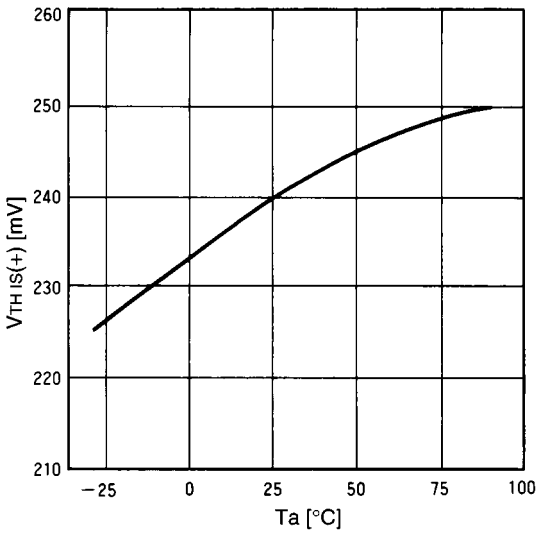


FA5316/17



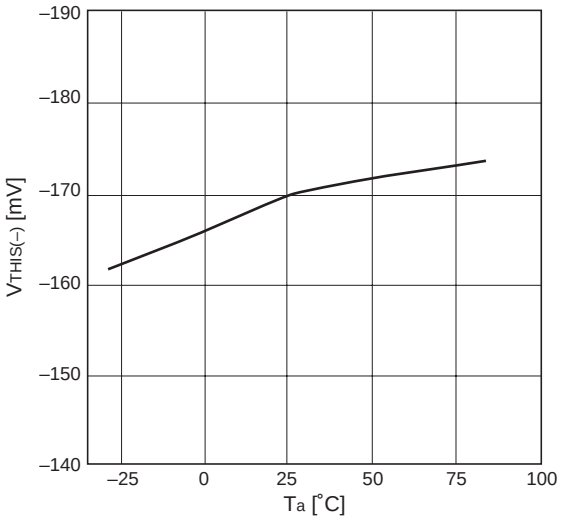
IS (+) terminal threshold voltage ($V_{TH\ IS(+)}$) vs. ambient temperature (Ta)

FA5310/11/16/17

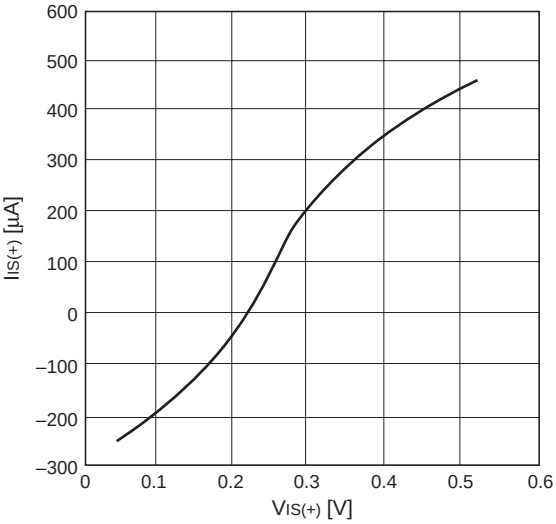


IS (–) terminal threshold voltage ($V_{TH\ IS(-)}$) vs. ambient temperature (Ta)

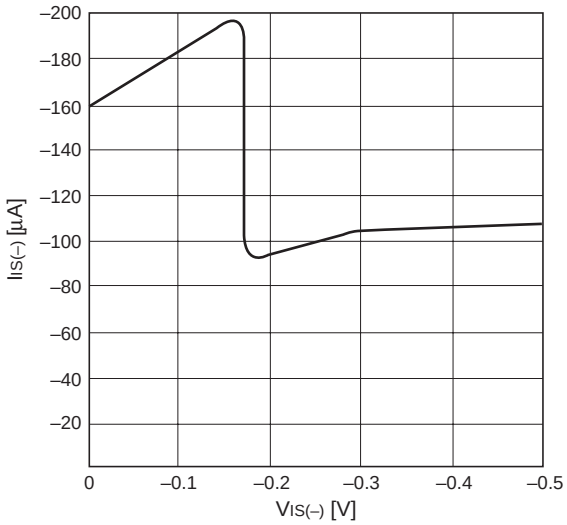
FA5314/15



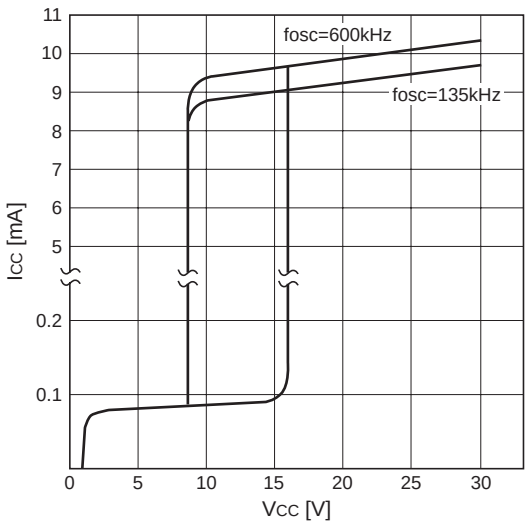
IS (+) terminal current ($I_{IS(+)}$) vs.
IS (+) terminal voltage ($V_{IS(+)}$)
FA5310/11/16/17



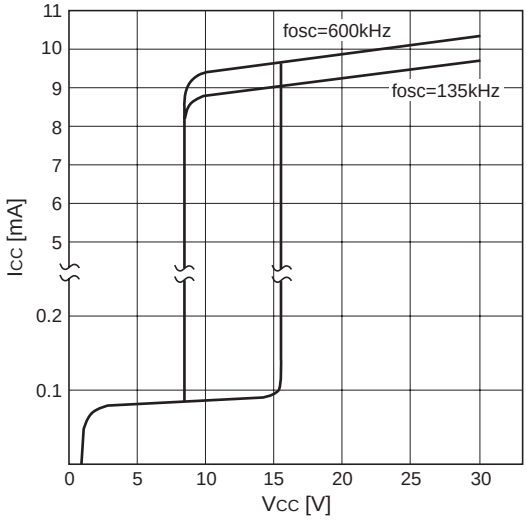
IS (-) terminal current ($I_{IS(-)}$) vs.
IS (-) terminal voltage ($I_{IS(-)}$)
FA5314/15



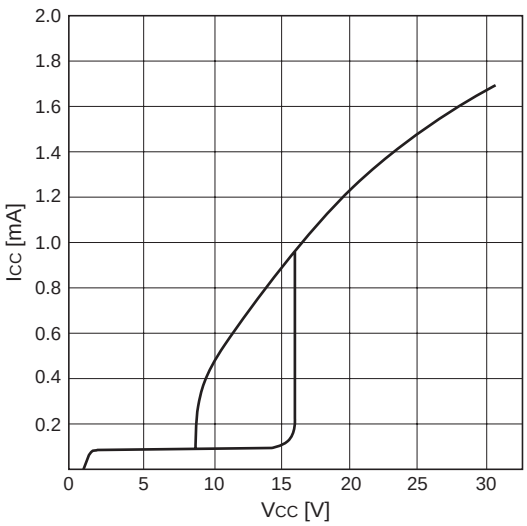
Supply current (I_{CC}) vs. supply voltage (V_{CC})
Ordinary operation
FA5310/11



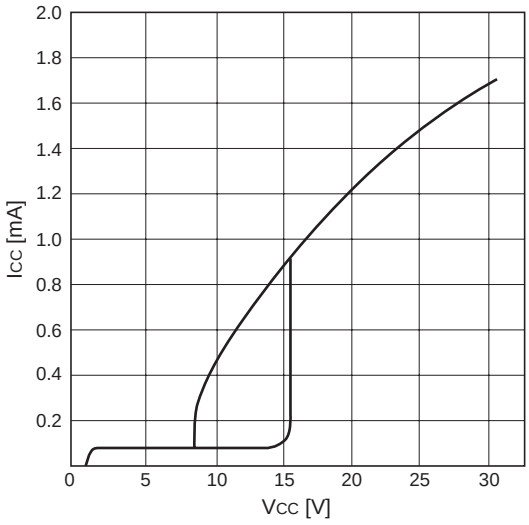
FA5314/15/16/17



Supply current (I_{CC}) vs. supply voltage (V_{CC})
OFF or OFF latch mode
FA5310/11

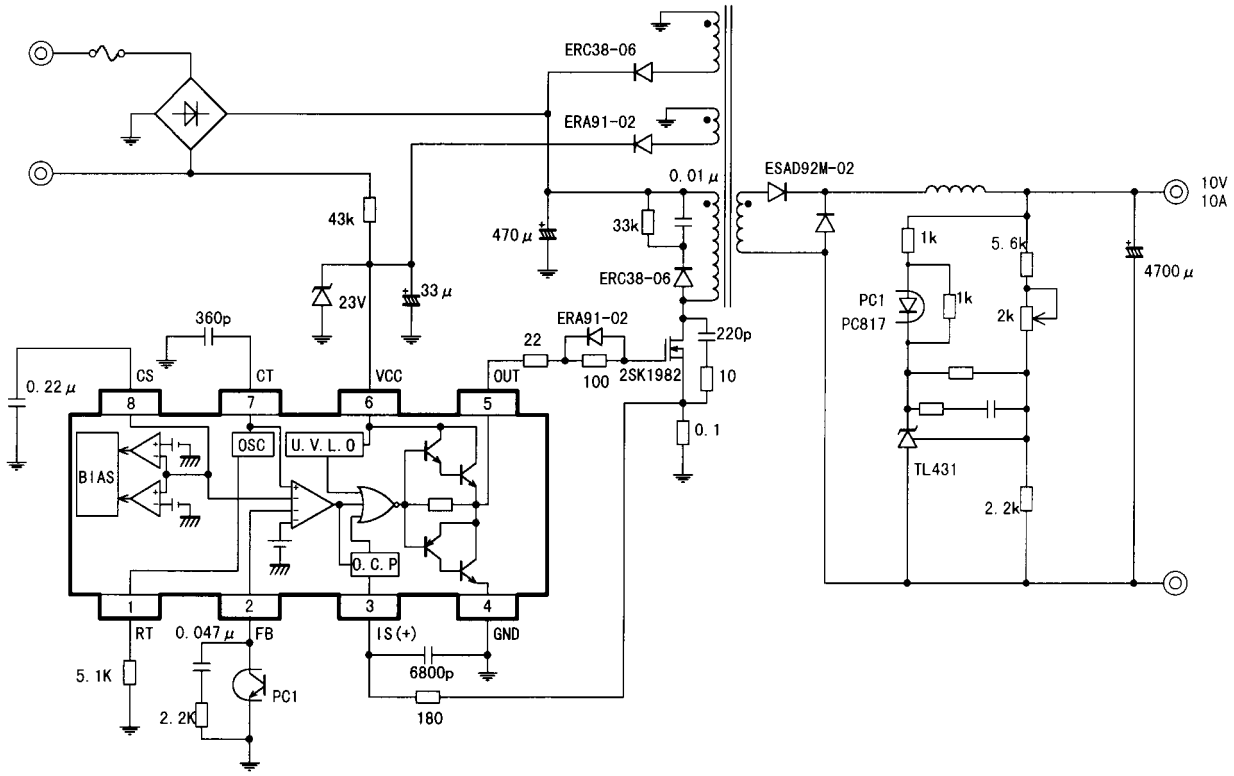


FA5314/15/16/17

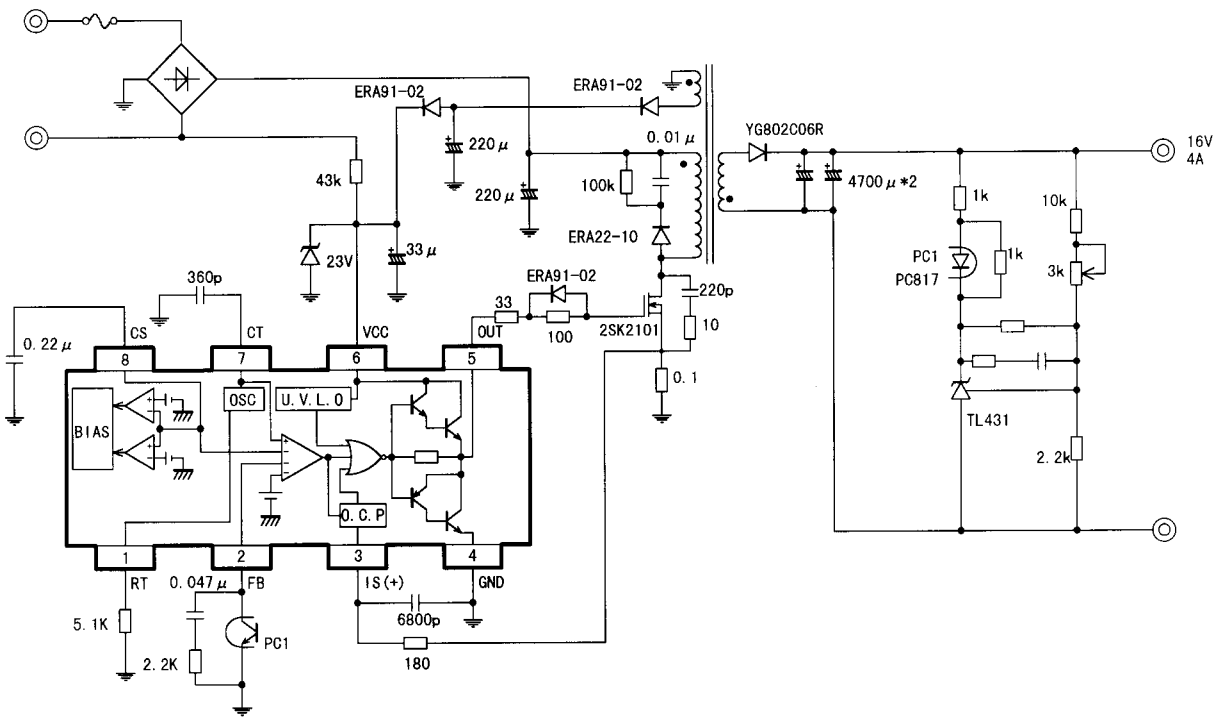


■ Application circuit

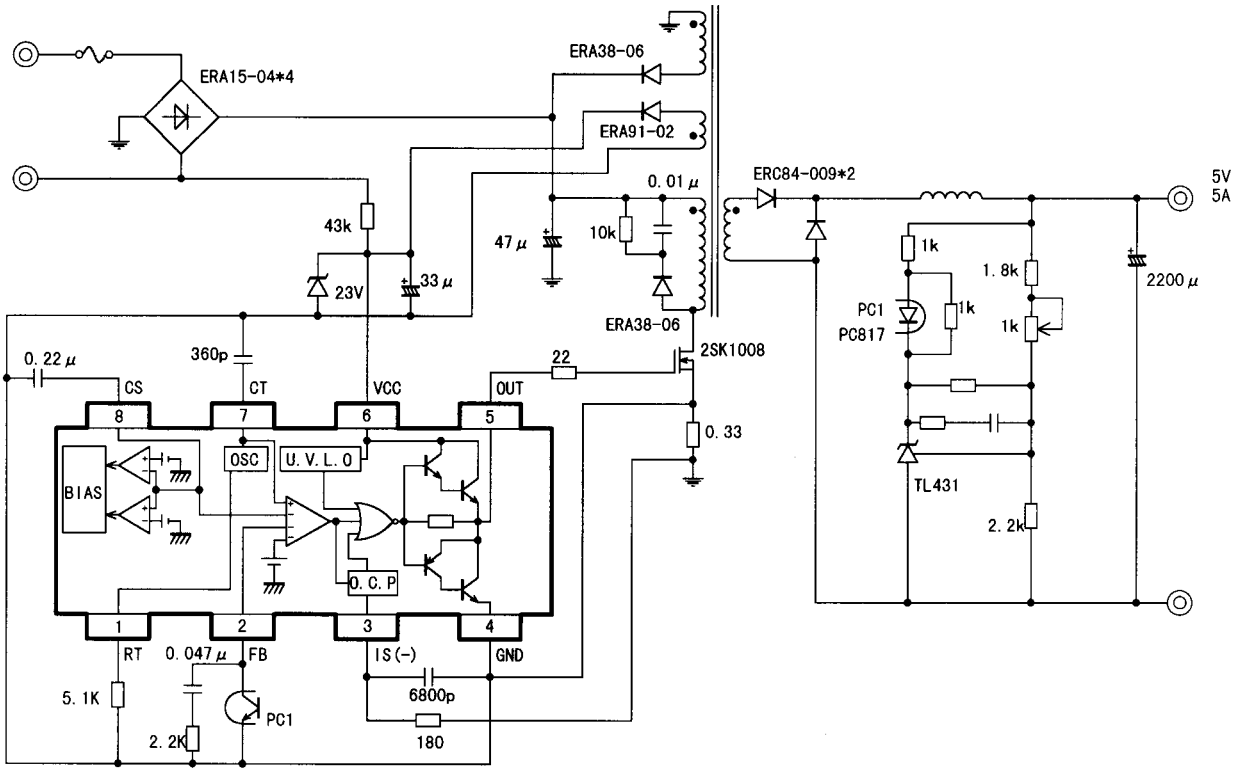
● Example of FA5310B application circuit



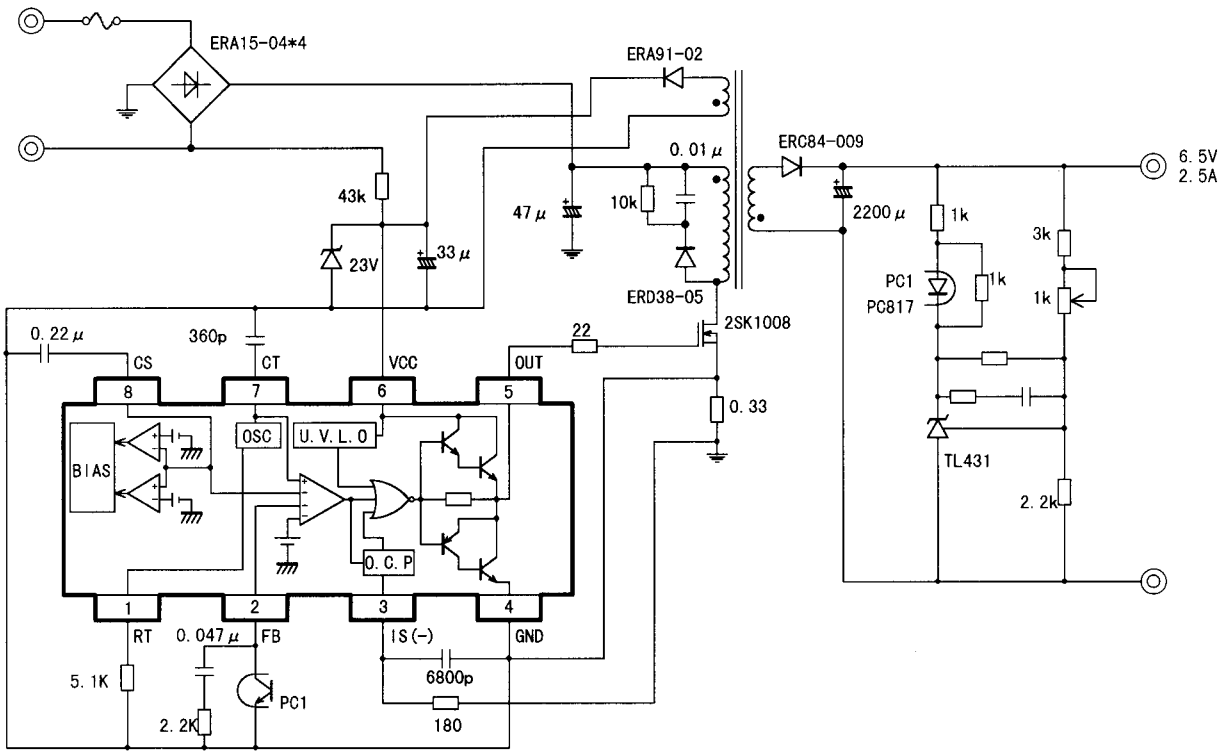
● Example of FA5311B application circuit



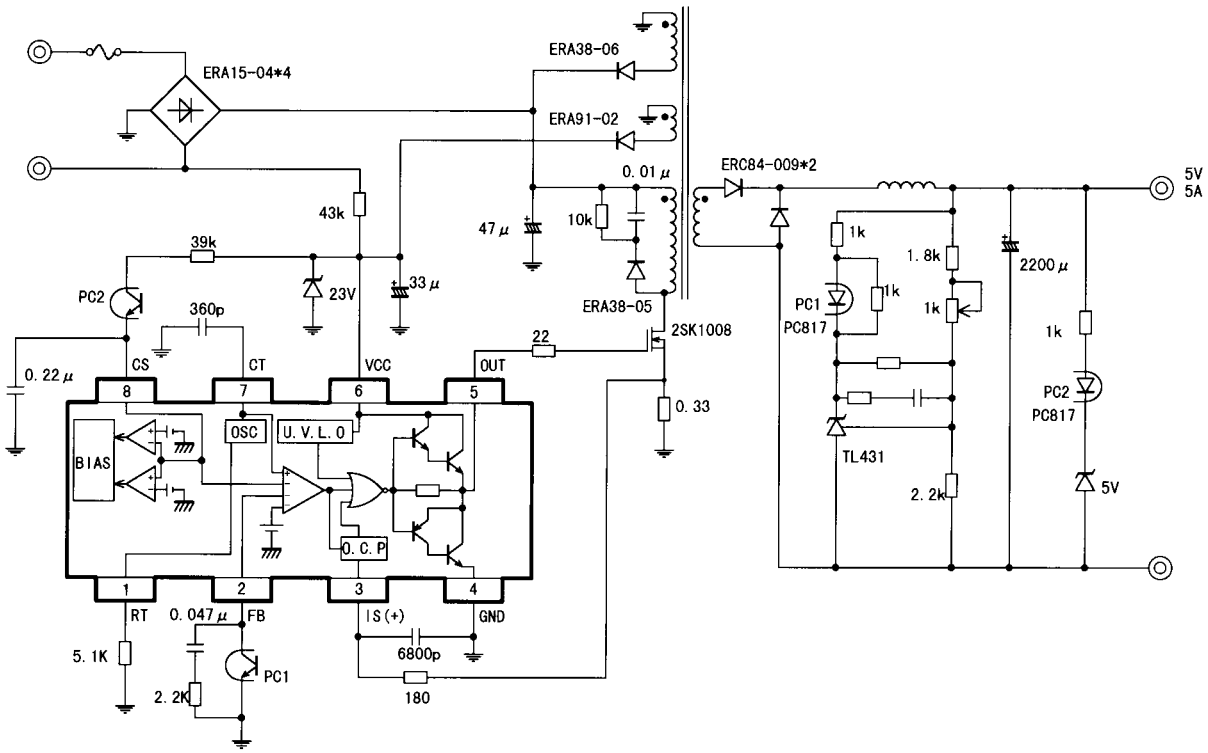
● **Example of FA5314 application circuit**



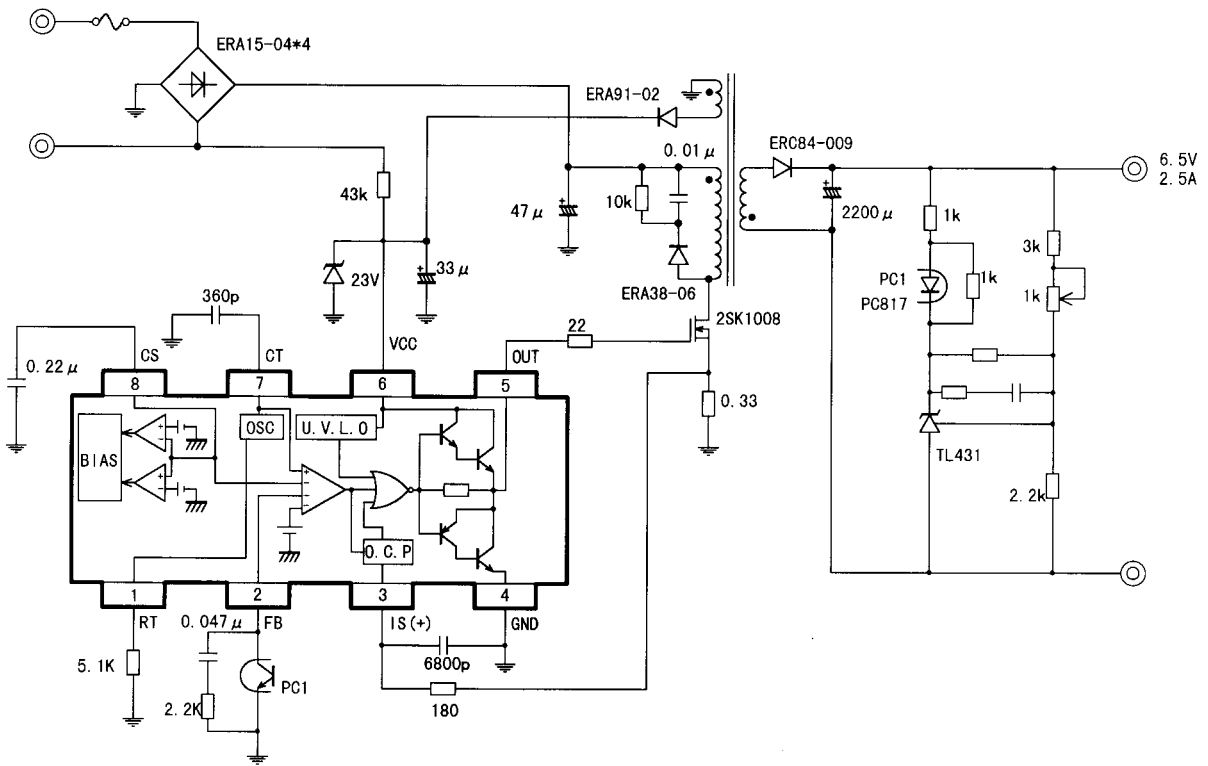
● **Example of FA5315 application circuit**



● Example of FA5316 application circuit



● Example of FA5317 application circuit



Parts tolerances characteristics are not defined in the circuit design sample shown above. When designing an actual circuit for a product, you must determine parts tolerances and characteristics for safe and economical operation.