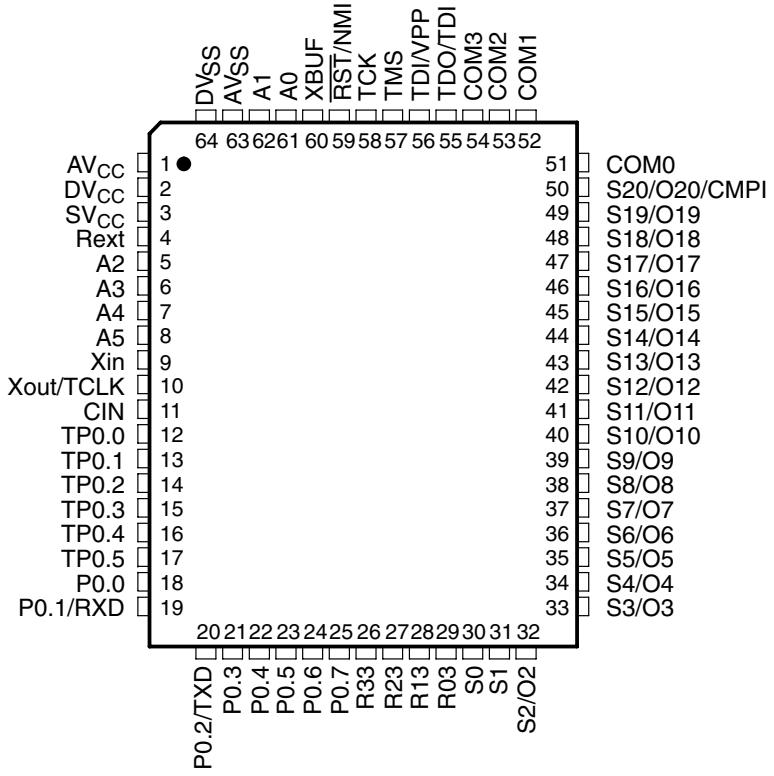
- **Low Supply Voltage Range, 2.5 V – 5.5 V**
- **Low Operation Current, 400 μA at 1 MHz, 3 V**
- **Ultra-Low Power Consumption (Standby Mode Down to 0.1 μA)**
- **Five Power-Saving Modes**
- **Wakeup From Standby Mode in 6 μs**
- **16-Bit RISC Architecture, 300 ns Instruction Cycle Time**
- **Single Common 32 kHz Crystal, Internal System Clock up to 3.3 MHz**
- **Integrated LCD Driver for up to 84 Segments**
- **Integrated 12+2 Bit A/D Converter**
- **Family Members Include:**
  - **MSP430C323, 8KB ROM, 256 Byte RAM**
  - **MSP430C325, 16KB ROM, 512 Byte RAM**
  - **MSP430P325A, 16KB OTP, 512 Byte RAM**
- **EPROM Version Available for Prototyping: PMS430E325A**
- **Serial Onboard Programming**
- **Programmable Code Protection by Security Fuse**
- **Avaliable in 64 Pin Quad Flatpack (QFP), 68 Pin Plastic J-Leaded Chip Carrier (PLCC), 68 Pin J-Leaded Ceramic Chip Carrier (JLCC) Package (EPROM Version)**

description

The Texas Instruments MSP430 is an ultra-low power mixed-signal microcontroller family consisting of several devices which feature different sets of modules targeted to various applications. The microcontroller is designed to be battery operated for an extended application lifetime. With 16-bit RISC architecture, 16-bit integrated registers on the CPU, and a constant generator, the MSP430 achieves maximum code efficiency. The digitally-controlled oscillator, together with the frequency-locked-loop (FLL), provides a wakeup from a low-power mode to active mode in less than 6 μs.

PG Package
(TOP VIEW)



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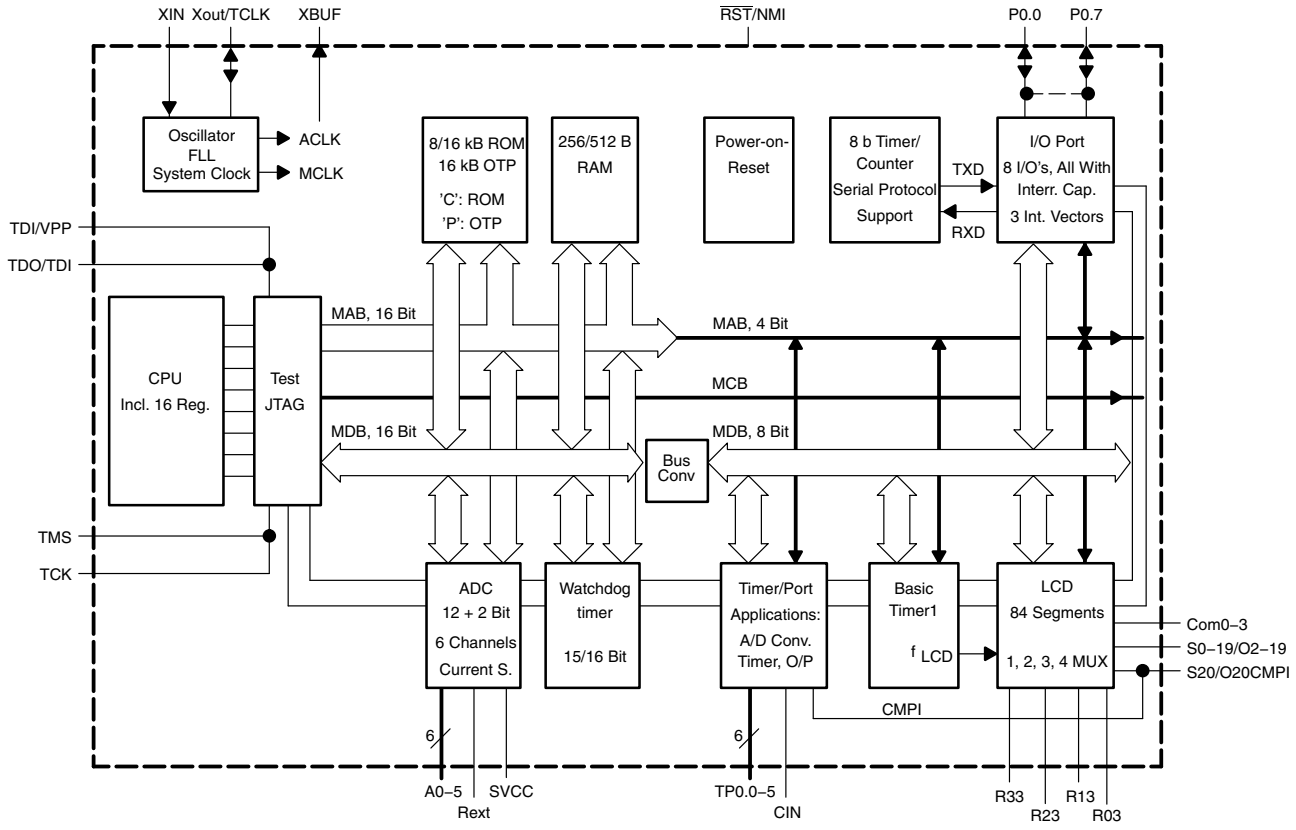
description (continued)

Typical applications include sensor systems that capture analog signals, convert them to digital values, and then process the data and display them or transmit them to a host system. The MSP430x32x offers an integrated 12+2 bit A/D converter with six multiplexed inputs.

AVAILABLE OPTIONS

T_A	PACKAGED DEVICES			
	PLASTIC 64-PIN QFP (PG)	PLASTIC 64-PIN QFP (PM)	PLASTIC 68-PIN PLCC (FN)	CERAMIC 68-PIN JLCC (FZ)
–40°C to 85°C	MSP430C323IPG MSP430C325IPG MSP430P325AIPG	MSP430C323IPM MSP430C325IPM MSP430P325AIPM	MSP430C323IFN MSP430C325IFN MSP430P325AIFN	—
25°C	—	—	—	PMS430E325AFZ

functional block diagram



Terminal Functions

TERMINAL NAME	TERMINAL NO.	I/O	DESCRIPTION
AV_{CC}	1		Positive analog supply voltage
AV_{SS}	63		Analog ground reference
A0	61	I	Analog-to-digital converter input port 0 or digital input port 0
A1	62	I	Analog-to-digital converter input port 1 or digital input port 1
A2–A5	5–8	I	Analog-to-digital converter inputs ports 2–5 or digital inputs ports 2–5
CIN	11	I	Input used as enable of counter TPCNT1 – Timer/Port
COM0–3	51–54	O	Common outputs, used for LCD backplanes – LCD
DV_{CC}	2		Positive digital supply voltage
DV_{SS}	64		Digital ground reference
P0.0	18	I/O	General-purpose digital I/O
P0.1/RXD	19	I/O	General-purpose digital I/O, receive digital input port, 8-bit Timer/Counter
P0.2/TXD	20	I/O	General-purpose digital I/O, transmit data output port, 8-bit Timer/Counter
P0.3–P0.7	21–25	I/O	Five general-purpose digital I/Os, bit 3 to bit 7
Rext	4	I	Programming resistor input of internal current source
$\overline{RST}$/NMI	59	I	Reset input or non-maskable interrupt input
R03	29	I	Input of fourth positive analog LCD level (V4) – LCD
R13	28	I	Input of third positive analog LCD level (V3) – LCD
R23	27	I	Input of second positive analog LCD level (V2) – LCD
R33	26	O	Output of first positive analog LCD level (V1) – LCD
SV_{CC}	3		Switched AV_{CC} to analog-to-digital converter
S0	30	O	Segment line S0 – LCD
S1	31	O	Segment line S1 – LCD
S2–S5/O2–O5	32–35	O	Segment lines S2 to S5 or digital output ports O2–O5, group 1 – LCD
S20/O20/CMPI	50	I/O	Segment line S20 can be used as comparator input port CMPI – Timer/Port
S6–S9/O6–O9	36–39	O	Segment lines S6 to S9 or digital output ports O6–O9, group 2 – LCD
S10–S13/O10–O13	40–43	O	Segment lines S10 to S13 or digital output ports O10–O13, group 3 – LCD
S14–S17/O14–O17	44–47	O	Segment lines S14 to S17 or digital output ports O14 to O17, group 4 – LCD
S18-S19/O18-O19	48, 49	O	Segment lines S18 and S19 or digital output port O18 and O19, group 5 – LCD
TCK	58	I	Test clock, clock input terminal for device programming and test
TDO/TDI	55	I/O	Test data output, data output terminal or data input during programming
TDI/VPP	56	I	Test data input, data input terminal or input of programming voltage
TMS	57	I	Test mode select, input terminal for device programming and test
TP0.0	12	O	General-purpose 3-state digital output port, bit 0 – Timer/Port
TP0.1	13	O	General-purpose 3-state digital output port, bit 1 – Timer/Port
TP0.2	14	O	General-purpose 3-state digital output port, bit 2 – Timer/Port
TP0.3	15	O	General-purpose 3-state digital output port, bit 3 – Timer/Port
TP0.4	16	O	General-purpose 3-state digital output port, bit 4 – Timer/Port
TP0.5	17	I/O	General-purpose digital input/output port, bit 5 – Timer/Port
XBUF	60	O	Clock signal output of system clock MCLK or crystal clock ACLK
Xin	9	I	Input terminal of crystal oscillator
Xout/TCLK	10	I/O	Output terminal of crystal oscillator or test clock input

short-form description

processing unit

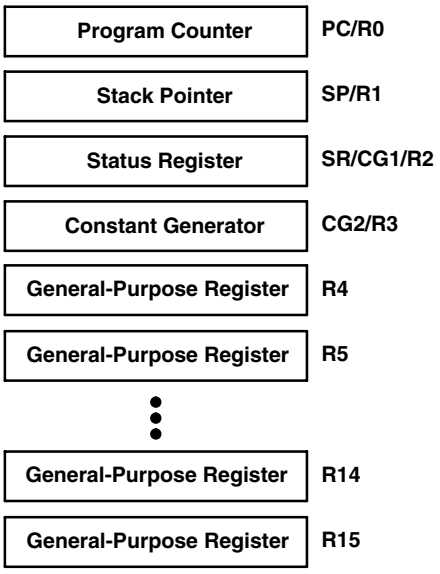
The processing unit is based on a consistent and orthogonally-designed CPU and instruction set. This design structure results in a RISC-like architecture, highly transparent to the application development and is distinguished due to ease of programming. All operations other than program-flow instructions are consequently performed as register operations in conjunction with seven addressing modes for source and four modes for destination operand.

CPU

Sixteen registers are located inside the CPU, providing reduced instruction execution time. This reduces a register-register operation execution time to one cycle of the processor frequency.

Four of the registers are reserved for special use as a program counter, a stack pointer, a status register and a constant generator. The remaining registers are available as general-purpose registers.

Peripherals are connected to the CPU using a data address and control bus and can be handled easily with all instructions for memory manipulation.



instruction set

The instruction set for this register-register architecture provides a powerful and easy-to-use assembler language. The instruction set consists of 51 instructions with three formats and seven addressing modes. Table 1 provides a summation and example of the three types of instruction formats; the addressing modes are listed in Table 2.

Table 1. Instruction Word Formats

Dual operands, source-destination	e.g. ADD R4, R5	R4 + R5 → R5
Single operands, destination only	e.g. CALL R8	PC → (TOS), R8 → PC
Relative jump, un-/conditional	e.g. JNE	Jump-on equal bit = 0

Each instruction that operates on word and byte data is identified by the suffix B.

Examples:	Instructions for word operation		Instructions for byte operation	
	MOV	EDE, TONI	MOV.B	EDE, TONI
	ADD	#235h, &MEM	ADD.B	#35h, &MEM
	PUSH	R5	PUSH.B	R5
	SWPB	R5	—	

Table 2. Address Mode Descriptions

ADDRESS MODE	s	d	SYNTAX	EXAMPLE	OPERATION
Register	√	√	MOV Rs, Rd	MOV R10, R11	R10 → R11
Indexed	√	√	MOV X(Rn), Y(Rm)	MOV 2(R5), 5(R6)	M(2 + R5) → M(6 + R6)
Symbolic (PC relative)	√	√	MOV EDE, TONI		M(EDE) → M(TONI)
Absolute	√	√	MOV &MEM, &TCDAT		M(MEM) → M(TCDAT)
Indirect	√		MOV @Rn, Y(Rm)	MOV @R10, Tab(R6)	M(R10) → M(Tab + R6)
Indirect autoincrement	√		MOV @Rn+, RM	MOV @R10+, R11	M(R10) → R11, R10 + 2 → R10
Immediate	√		MOV #X, TONI	MOV #45, TONI	#45 → M(TONI)

NOTE: s = source d = destination

Computed branches (BR) and subroutine calls (CALL) instructions use the same addressing modes as the other instructions. These addressing modes provide *indirect* addressing, ideally suited for computed branches and calls. The full use of this programming capability permits a program structure different from conventional 8- and 16-bit controllers. For example, numerous routines can easily be designed to deal with pointers and stacks instead of using flag type programs for flow control.

operation modes and interrupts

The MSP430 operating modes support various advanced requirements for ultra low power and ultra low energy consumption. This is achieved by the intelligent management of the operations during the different module operation modes and CPU states. The requirements are fully supported during interrupt event handling. An interrupt event awakens the system from each of the various operating modes and returns with the RETI instruction to the mode that was selected before the interrupt event. The clocks used are ACLK and MCLK. ACLK is the crystal frequency and MCLK is a multiple of ACLK and is used as the system clock.

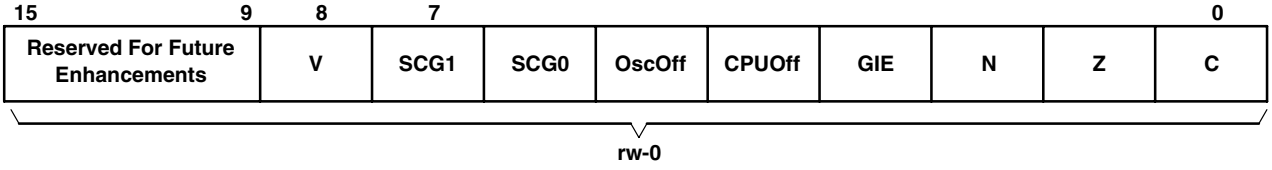
The software can configure five operating modes:

- Active mode (AM). The CPU is enabled with different combinations of active peripheral modules.
- Low power mode 0 (LPM0). The CPU is disabled, peripheral operation continues, ACLK and MCLK signals are active, and loop control for MCLK is active.
- Low power mode 1 (LPM1). The CPU is disabled, peripheral operation continues, ACLK and MCLK signals are active, and loop control for MCLK is inactive.
- Low power mode 2 (LPM2). The CPU is disabled, peripheral operation continues, ACLK signal is active, and MCLK and loop control for MCLK are inactive.
- Low power mode 3 (LPM3). The CPU is disabled, peripheral operation continues, ACLK signal is active, MCLK and loop control for MCLK are inactive, and the dc generator for the digital controlled oscillator (DCO) (→MCLK generator) is switched off.
- Low power mode 4 (LPM4). The CPU is disabled, peripheral operation continues, ACLK signal is inactive (crystal oscillator stopped), MCLK and loop control for MCLK are inactive, and the dc generator for the DCO is switched off.

The special function registers (SFR) include module-enable bits that stop or enable the operation of the specific peripheral module. All registers of the peripherals may be accessed if the operational function is stopped or enabled. However, some peripheral current-saving functions are accessed through the state of local register bits. An example is the enable/disable of the analog voltage generator in the LCD peripheral, which is turned on or off using one register bit.

operation modes and interrupts (continued)

The most general bits that influence current consumption and support fast turnon from low-power operating modes are located in the status register (SR). Four of these bits control the CPU and the system clock generator: SCG1, SCG0, OscOff, and CPUOff.



interrupt vector addresses

The interrupt vectors and the power-up starting address are located in the ROM with an address range of 0FFFFh-0FFE0h. The vector contains the 16-bit address of the appropriate interrupt handler instruction sequence.

INTERRUPT SOURCE	INTERRUPT FLAG	SYSTEM INTERRUPT	WORD ADDRESS	PRIORITY
Power-up, external reset, watchdog	WDTIFG (see Note1)	Reset	0FFFEh	15, highest
NMI, oscillator fault	NMIIFG (see Notes 1 and 3) OFIFG (see Notes 1 and 4)	Non-maskable, (Non)-maskable	0FFFCh	14
Dedicated I/O P0.0	P0.0IFG	maskable	0FFFAh	13
Dedicated I/O P0.1 or 8-bit Timer/Counter RXD	P0.1IFG	maskable	0FFF8h	12
			0FFF6h	11
Watchdog Timer	WDTIFG	maskable	0FFF4h	10
			0FFF2h	9
			0FFF0h	8
			0FFEEh	7
			0FFECh	6
ADC	ADCIFG	maskable	0FFEAh	5
Timer/Port	RC1FG, RC2FG, EN1FG (see Note 2)	maskable	0FFE8h	4
			0FFE6h	3
			0FFE4h	2
Basic Timer1	BTIFG	maskable	0FFE2h	1
I/O port 0, P0.2–7	P0.27IFG (see Note 1)	maskable	0FFE0h	0, lowest

NOTE 1: Multiple source flags
NOTE 2: Timer/Port interrupt flags are located in the T/P registers
NOTE 3: Non-maskable: neither the individual nor the general interrupt enable bit will disable an interrupt event.
NOTE 4: (Non)-maskable: the individual interrupt enable bit can disable on interrupt event, but the general interrupt enable bit cannot.

operation modes and interrupts (continued)

special function registers

Most interrupt and module enable bits are collected into the lowest address space. Special function register bits that are not allocated to a functional purpose are not physically present in the device. Simple SW access is provided with this arrangement.

interrupt enable 1 and 2

Address	7	6	5	4	3	2	1	0
0h					P0IE.1	P0IE.0	OFIE	WDTIE
					rw-0	rw-0	rw-0	rw-0

WDTIE: Watchdog Timer enable signal
OFIE: Oscillator fault enable signal
P0IE.0: Dedicated I/O P0.0
P0IE.1: P0.1 or 8-bit Timer/Counter, RXD

Address	7	6	5	4	3	2	1	0
01h	BTIE				TPIE	ADIE		
	rw-0				rw-0	rw-0		

ADIE: A/D converter enable signal
TPIE: Timer/Port enable signal
BTIE: Basic Timer1 enable signal

interrupt flag register 1 and 2

Address	7	6	5	4	3	2	1	0
02h				NMIIFG	P0IFG.1	P0IFG.0	OFIFG	WDTIFG
				rw-0	rw-0	rw-0	rw-1	rw-0

WDTIFG: Set on overflow or security key violation
or
Reset on V_{CC} power on or reset condition at $\overline{RST}$/NMI-pin
OFIFG: Flag set on oscillator fault
P0.0IFG: Dedicated I/O P0.0
P0.1IFG: P0.1 or 8-bit Timer/Counter, RXD
NMIIFG: Signal at $\overline{RST}$/NMI-pin

Address	7	6	5	4	3	2	1	0
03h	BTIFG					ADIFG		
	rw					rw-0		

BTIFG Basic Timer1 flag
ADFIG Analog-to-digital converter flag

operation modes and interrupts (continued)

module enable register 1 and 2

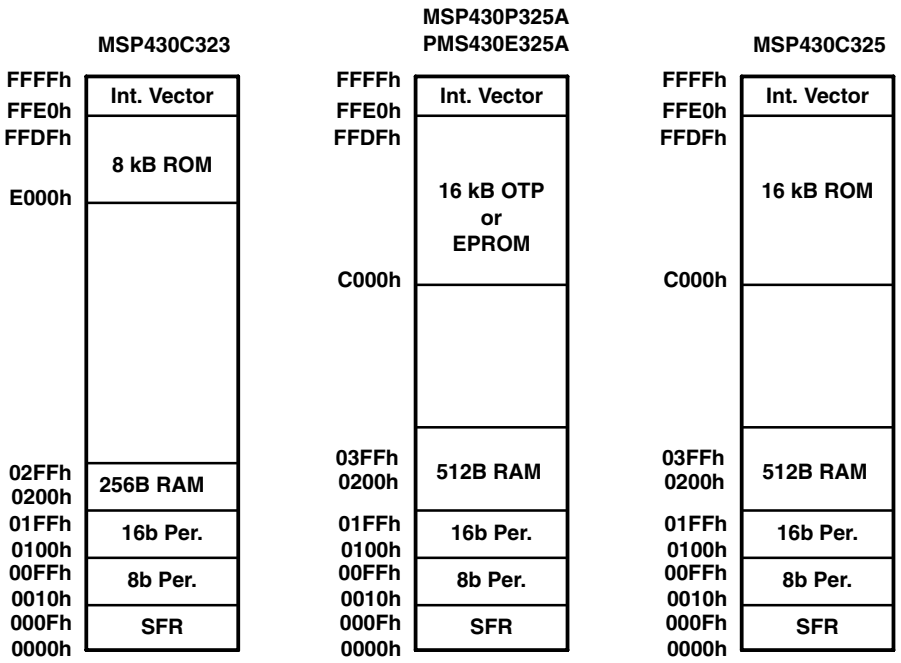
Address	7	6	5	4	3	2	1	0
04h								

Address	7	6	5	4	3	2	1	0
05h								

Legend

rw:	Bit can be read and written.
rw-0:	Bit can be read and written. It is reset by PUC.
	SFR bit not present in device.

memory organization



peripherals

Peripherals connect to the CPU through data, address, and control busses and can be handled easily with all instructions for memory manipulation.

peripheral file map

PERIPHERALS WITH WORD ACCESS			
Watchdog	Watchdog Timer control	WDTCTL	0120h
ADC	Data register Reserved Control register Input enable register Input register	ADAT ACTL AEN AIN	0118h 0116h 0114h o112h 0110h
PERIPHERALS WITH BYTE ACCESS			
EPROM	EPROM control	EPCTL	054h
Crystal buffer	Crystal buffer control	CBCTL	053h
System clock	SCG frequency control SCG frequency integrator SCG frequency integrator	SCFQCTL SCFI1 SCFI0	052h 051h 050h
Timer/Port	Timer/Port enable Timer/Port data Timer/Port counter2 Timer/Port counter1 Timer/Port control	TPE TPD TPCNT2 TPCNT1 TPCTL	04Fh 04Eh 04Dh 04Ch 04Bh
8-Bit Timer/Counter	8-Bit Timer/Counter data 8-Bit Timer/Counter preload 8-Bit Timer/Counter control	TCDAT TCPLD TCCTL	044h 043h 042h
Basic Timer1	Basic Timer counter2 Basic Timer counter1 Basic Timer control	BTCNT2 BTCNT1 BTCTL	047h 046h 040h
LCD	LCD memory 15 : LCD memory 1 LCD control & mode	LCDM15 : LCDM1 LCDCTL	03Fh : 031h 030h
Port P0	Port P0 interrupt enable Port P0 interrupt edge select Port P0 interrupt flag Port P0 direction Port P0 output Port P0 input	P0IE P0IES P0IFG P0DIR P0OUT P0IN	015h 014h 013h 012h 011h 010h
Special function	SFR interrupt flag2 SFR interrupt flag1 SFR interrupt enable2 SFR interrupt enable1	IFG2 IFG1 IE2 IE1	003h 002h 001h 000h

oscillator and system clock

Two clocks are used in the system, the system (master) clock (MCLK) and the auxiliary clock (ACLK). The MCLK is a multiple of the ACLK. The ACLK runs with the crystal oscillator frequency. The special design of the oscillator supports the feature of low current consumption and the use of a 32 768 Hz crystal. The crystal is connected across two terminals without any other external components being required.

The oscillator starts after applying VCC, due to a reset of the control bit (OscOff) in the status register (SR). It can be stopped by setting the OscOff bit to a 1. The enabled clock signals ACLK, ACLK/2, ACLK/4, or MCLK are accessible for use by external devices at output terminal XBUF.

oscillator and system clock (continued)

The controller system clock has to operate with different requirements according to the application and system conditions. Requirements include:

- High frequency in order to react quickly to system hardware requests or events
- Low frequency in order to minimize current consumption, EMI, etc.
- Stable frequency for timer applications e.g. real time clock (RTC)
- Enable start-stop operation with a minimum of delay

These requirements cannot all be met with fast frequency high-Q crystals or with RC-type low-Q oscillators. The compromise selected for the MSP430 uses a low-crystal frequency which is multiplied to achieve the desired nominal operating range:

$$f_{(system)} = (N+1) \times f_{(crystal)}$$

The crystal frequency multiplication is acheived with a frequency locked loop (FLL) technique. The factor N is set to 31 after a power-up clear condition. The FLL technique, in combination with a digital controlled oscillator (DCO) provides immediate start-up capability together with long term crystal stability. The frequency variation of the DCO with the FLL inactive is typically 330 ppm which means that with a cycle time of 1 μs the maximum possible variation is 0.33 ns. For more precise timing, the FLL can be used which forces longer cycle times if the previous cycle time was shorter than the selected one. This switching of cycle times makes it possible to meet the chosen system frequency over a long period of time.

The start-up operation of the system clock depends on the previous machine state. During a power up clear (PUC), the DCO is reset to its lowest possible frequency. The control logic starts operation immediately after recognition of PUC. Connect operation of the FLL control logic requires the presence of a stable crystal oscillator.

digital I/O

One 8-bit I/O port (Port0) is implemented. Six control registers give maximum flexibility of digital input/output to the application:

- All individual I/O bits are programmable independently.
- Any combination of input, output, and interrupt conditions is possible.
- Interrupt processing of external events is fully implemented for all eight bits of port P0.
- Provides read/write access to all registers with all instructions.

The six registers are:

- Input register — Contains information at the pins
- Output register — Contains output information
- Direction register — Controls direction
- Interrupt flags — Indicates if interrupt(s) are pending
- Interrupt edge select — Contains input signal change necessary for interrupt
- Interrupt enable — Contains interrupt enable pins

All six registers contain eight bits except for the interrupt flag register and the interrupt enable register. The two LSBs of the interrupt flag and interrupt enable registers are located in the special functions register (SFR). Three interrupt vectors are implemented, one for Port0.0, one for Port0.1, and one commonly used for any interrupt event on Port0.2 to Port0.7. The Port0.1 and Port0.2 pin function is shared with the 8-bit Timer/Counter.

LCD drive

Liquid crystal displays (LCDs) for static, 2-, 3- and 4-MUX operations can be driven directly. The controller LCD logic operation is defined by software using memory-bit manipulation. LCD memory is part of the LCD module, not part of data memory. Eight mode and control bits define the operation and current consumption of the LCD drive. The information for the individual digits can be easily obtained using table programming techniques combined with the correct addressing mode. The segment information is stored in LCD memory using instructions for memory manipulation.

The drive capability is mainly defined by the external resistor divider that supports the analog levels for 2-, 3- and 4-MUX operation. Groups of the LCD segment lines can be selected for digital output signals. The MSP430x32x configuration has four common signal lines and 21 segment lines.

A/D converter

The analog-to-digital converter (ADC) is a cascaded converter type that converts analog signals from V_{CC} to GND. It is a 12+2 bit converter with a software or automatically-controlled range select. Five inputs can be selected for analog or digital function. A ratiometric current source can be used on four of the analog pins. The current is adjusted by an external resistor and is enabled/disabled by bits located in the control registers. The conversion is started by setting the start-of-conversion bit (SOC) in the control register and the end-of-conversions sets the interrupt flag. The analog input signal is sampled starting with SOC during the next twelve MCLK clock pulses. The power-down bit in the control register controls the operating mode of the ADC peripheral. The current consumption and operation is stopped when it is set. The system reset PUC sets the power-down bit.

Basic Timer1

The Basic Timer1 (BT1) divides the frequency of MCLK or ACLK, as selected with the SSEL bit, to provide low frequency control signals. This is done within the system by one central divider, the Basic Timer1, to support low current applications. The BTCTL control register contains the flags which controls or selects the different operational functions. When the supply voltage is applied or when a reset of the device ($\overline{\text{RST}}$/NMI pin), a watchdog overflow, or a watchdog security key violation occurs, all bits in the register hold undefined or unchanged status. The user software usually configures the operational conditions on the BT1 during initialization.

The Basic Timer1 has two 8-bit timers which can be cascaded to a 16-bit timer. Both timers can be read and written by software. Two bits in the SFR address range handle the system control interaction according to the function implemented in the Basic Timer1. These two bits are the Basic Timer1 interrupt flag (BTIFG) and the Basic Timer1 interrupt enable (BTIE) bit.

Watchdog Timer

The primary function of the Watchdog Timer (WDT) module is to perform a controlled system restart after a software upset has occurred. If the selected time interval expires, a system reset is generated. If this watchdog function is not needed in an application, the module can work as an interval timer, which generates an interrupt after the selected time interval.

The Watchdog Timer counter (WDTCNT) is a 15/16-bit up-counter which is not directly accessible by software. The WDTCNT is controlled using the Watchdog Timer control register (WDTCTL), which is an 8-bit read/write register. Writing to WDTCTL, in both operating modes (watchdog or timer) is only possible by using the correct password in the high-byte. The low-byte stores data written to the WDTCTL. The high-byte password is 05Ah. If any value other than 05Ah is written to the high-byte of the WDTCTL, a system reset PUC is generated. *When the password is read it's value is 069h*. This minimizes accidental write operations to the WDTCTL register. In addition to the Watchdog Timer control bits, two bits included in the WDTCTL configure the NMI pin.

8-bit Timer/Counter

The 8-bit interval timer supports three major functions for the application:

- Serial communication or data exchange
- Pulse counting or pulse accumulation
- Timer

The 8-bit Timer/Counter peripheral includes the following major blocks: an 8-bit Up-Counter with preload register, an 8-bit control register, an Input clock selector, an edge detection (e.g. Start bit detection for asynchronous protocols), and an input and output data latch, triggered by the carry-out-signal from the 8-bit counter.

The 8-bit counter counts up with an input clock which is selected by two control bits from the control register. The four possible clock sources are MCLK, ACLK, the external signal from terminal P0.1, and the signal from the logical AND of MCLK and terminal P0.1.

Two counter inputs (load, enable) control the counter operation. The load input controls load operations. A write-access to the counter results in loading the content of the preload register into the counter. The software writes or reads the preload register with all instructions. The preload register acts as a buffer and can be written immediately after the load of the counter is completed. The enable input enables the count operation. When the enable signal is set HIGH, the counter will count-up each time a positive clock edge is applied to the clock input of the counter.

Serial protocols, like UART protocol, need start-bit edge-detection to determine, at the receiver, the start of a data transmission. When this function is activated, the counter starts counting after the start-bit condition is detected. The first signal level is sampled into the RXD input data-latch after completing the first timing interval, which is programmed into the counter. Two latches are used for input and output data (RXD_FF and TXD_FF) are clocked by the counter after the programmed timing interval has elapsed.

UART

The serial communication is realized by using software and the 8-bit Timer/Counter hardware. The hardware supports the output of the serial data stream, bit-by-bit, with the timing determined by the counter. The software/hardware interface connects the mixed signal controller to external devices, systems, or networks.

Timer/Port

The Timer/Port module has two 8-bit counters, an input that triggers one counter, and six 3-state digital outputs. Both counters have an independent clock-selector for selecting an external signal or one of the internal clocks (ACLK or MCLK). One of the counters has an extended control capability to halt, count continuously, or gate the counter by selecting one of two external signals. This gate signal sets the interrupt flag, if an external signal is selected, and the gate stops the counter.

Both timers can be read from and written to by software. The two 8-bit counters can be cascaded to a 16-bit counter. A common interrupt vector is implemented. The interrupt flag can be set from three events in the 8-bit counter mode (gate signal, overflow from the counters) or from two events in the 16-bit counter mode (gate signal, overflow from the MSB of the cascaded counter).

absolute maximum ratings†

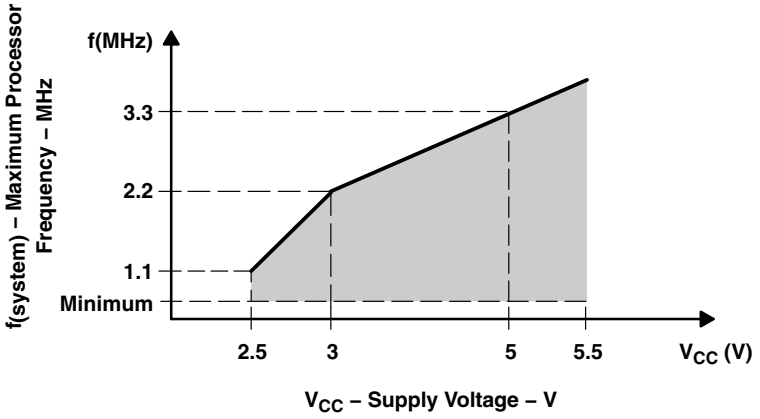
Voltage applied at V_{CC} to V_{SS} (see Note 5) . . . . . –0.3 V to 6 V
Voltage applied to any pin (referenced to V_{SS}) . . . . . –0.3 V to V_{CC} + 0.3 V
Diode current at any device terminal . . . . . ± 2 mA
Storage temperature, T_{stg} (unprogrammed device) . . . . . –55°C to 150°C
T_{stg} (programmed device) . . . . . –40°C to 85°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTE 5: All voltage values relative to V_{SS}.

recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, V_{CC} (MSP430C32x)		2.5		5.5	V
Supply voltage, V_{CC} (MSP430P/E325A)		2.5		5.5	V
Supply voltage, during programming OTP/EPROM ($AV_{CC} = DV_{CC} = V_{CC}$)	MSP430P325A, PMS430E325A	4.5	5	5.5	V
Supply voltage, V_{SS}			0		V
Operating free-air temperature range, T_A	MSP430C32x, MSP430P325A	–40		85	°C
	PMS430E325A		25		
XTAL frequency, $f_{(XTAL)}$			32 768		Hz
Processor frequency (signal MCLK), $f_{(system)}$	V_{CC} = 3 V	DC		2.2	MHz
	V_{CC} = 5 V	DC		3.3	
Low-level input voltage, V_{IL} (excluding Xin, Xout)	V_{CC} = 3 V/5 V	V_{SS}		V_{SS}+0.8	V
High-level input voltage, V_{IH} (excluding Xin, Xout)		0.7 V_{CC}		V_{CC}	
Low-level input voltage, $V_{IL(Xin, Xout)}$		V_{SS}		0.2×V_{CC}	V
High-level input voltage, $V_{IH(Xin, Xout)}$		0.8×V_{CC}		V_{CC}	



NOTE: Minimum processor frequency is defined by system clock.

Figure 1. Processor Frequency vs Supply Voltage

electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

supply current into $AV_{CC}+DV_{CC}$ excluding external current, f_{system} = 1 MHz

PARAMETER			TEST CONDITIONS		MIN	TYP	MAX	UNIT
$I_{(AM)}$	Active mode, A/D conversion in power-down	C32x	T_A = –40°C to 85°C,	V_{CC} = 3 V		400	500	µA
			T_A = –40°C to 85°C,	V_{CC} = 5 V		800	900	
		P325A	T_A = –40°C to 85°C,	V_{CC} = 3 V		500	550	
			T_A = –40°C to 85°C,	V_{CC} = 5 V		950	1050	
$I_{(CPUOff)}$	Low power mode, (LPM0, LPM1)	C32x	T_A = –40°C to 85°C,	V_{CC} = 3 V		50	70	µA
			T_A = –40°C to 85°C,	V_{CC} = 5 V		100	130	
		P325A	T_A = –40°C to 85°C,	V_{CC} = 3 V		50	70	
			T_A = –40°C to 85°C,	V_{CC} = 5 V		100	130	
$I_{(LPM2)}$	Low power mode, (LPM2)		T_A = –40°C to 85°C,	V_{CC} = 3 V		6	12	µA
			T_A = –40°C to 85°C,	V_{CC} = 5 V		15	25	
$I_{(LPM3)}$	Low power mode, (LPM3)		T_A = –40°C	V_{CC} = 3 V		1.5	2.4	µA
			T_A = 25°C			1.3	2	
			T_A = 85°C			1.6	2.8	
			T_A = –40°C	V_{CC} = 5 V		5.2	7	
			T_A = 25°C			4.2	6.5	
			T_A = 85°C			4	7	
$I_{(LPM4)}$	Low power mode, (LPM4)		T_A = –40°C	V_{CC} = 3 V/5 V		0.1	0.8	µA
			T_A = 25°C			0.1	0.8	
			T_A = 85°C			0.4	1.3	

NOTE: All inputs are tied to 0 V or V_{CC}. Outputs do not source or sink any current. The current consumption in LPM2, LPM3 and LPM4 are measured with active Basic Timer1 (ACLK selected) and LCD Module ($f_{(LCD)}$=1024 Hz, 4 MUX).

current consumption of active mode versus system frequency

$$I_{AM} = I_{AM[1\ MHz]} \times f_{system}\ [MHz]$$

current consumption of active mode versus supply voltage

$$I_{AM} = I_{AM[3\ V]} + 200\ \mu A/V \times (V_{CC}–3\ V)$$

Schmitt-trigger inputs Port 0, P0.x Timer/Port, CIN, TP 0.5

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
V_{IT+}	Positive-going input threshold voltage	V_{CC} = 3 V	1.2		2.1	V
		V_{CC} = 5 V	2.3		3.4	
$V_{IT–}$	Negative-going input threshold voltage	V_{CC} = 3 V	0.5		1.35	
		V_{CC} = 5 V	1.4		2.3	
V_{hys}	Hysteresis (V_{IT+}–$V_{IT–}$)	V_{CC} = 3 V	0.3		1	
		V_{CC} = 5 V	0.6		1.4	

electrical characteristics over recommended operating free-air temperature range (unless otherwise noted) (continued)

outputs – Port 0: P0.x; Timer/Port: TP0.0...5; LCD: Sxx/Oxx; XBUF, (see Note 6)

PARAMETER		TEST CONDITIONS			MIN	TYP	MAX	UNIT
V_{OH}	High-level output current	I_{OH} = –1.2 mA,	V_{CC} = 3 V,	See Note 6	V_{CC}–0.4		V_{CC}	V
		I_{OH} = –3.5 mA,	V_{CC} = 3 V,	See Note 7	V_{CC}–1		V_{CC}	
		I_{OH} = –1.5 mA,	V_{CC} = 5 V,	See Note 6	V_{CC}–0.4		V_{CC}	
		I_{OH} = –4.5 mA,	V_{CC} = 5 V,	See Note 7	V_{CC}–1		V_{CC}	
V_{OL}	Low-level output voltage	I_{OL} = 1.2 mA,	V_{CC} = 3 V,	See Note 6	V_{SS}		V_{SS}+0.4	V
		I_{OL} = 3.5 mA,	V_{CC} = 3 V,	See Note 7	V_{SS}		V_{SS}+1	
		I_{OL} = 1.5 mA,	V_{CC} = 5 V,	See Note 6	V_{SS}		V_{SS}+0.4	
		I_{OL} = 4.5 mA,	V_{CC} = 5 V,	See Note 7	V_{SS}		V_{SS}+1	

NOTES: 6. The maximum total current, I_{OH}max and I_{OL}max, for all outputs combined, should not exceed ±9.6 mA to satisfy the maximum voltage drop specified.
7. The maximum total current, I_{OH}max and I_{OL}max, for all outputs combined, should not exceed ±20 mA to satisfy the maximum voltage drop specified.

leakage current (see Note 8)

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
$I_{lkg(TP)}$	Leakage current, Timer/Port	Timer/Port: $V_{(TP0.x,CIN)}$ (see Note 9)	V_{CC} = 3 V/5 V			±50	nA
$I_{lkg(P0x)}$	Leakage current, port 0	Port 0: $V_{(P0.x)}$ (see Note 10)				±50	nA
$I_{lkg(S20)}$	Leakage current, S20	$V_{(S20)}$ = V_{SS} to V_{CC}				±50	nA
$I_{lkg(Ax)}$	Leakage current, ADC	ADC: Ax, x= 0 to 5 (see Note 11)				±30	nA
$I_{lkg(\overline{RST}/NMI)}$	Leakage current, $\overline{RST}$/NMI					±50	nA

NOTES: 8. The leakage current is measured with V_{SS} or V_{CC} applied to the corresponding pin(s), unless otherwise noted.
9. All Timer/Port pins TP0.0 to TP0.5 are Hi-Z. Pins CIN and TP.0 to TP0.5 are connected together during leakage current measurement. In the leakage measurement the input CIN is included. The input voltage is V_{SS} or V_{CC}.
10. The port pin must be selected for input and there must be no optional pullup or pulldown resistor.
11. The input voltage is $V_{(IN)}$ = V_{SS} to V_{CC} , the current source is off, AEN.x bit is normally reset to stop throughput current flowing from V_{CC} to V_{SS} terminal.

optional resistors (see Note 12)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$R_{(opt1)}$	Resistors, individually programmable with ROM code, all port pins, values applicable for pulldown and pullup	V_{CC} = 3 V/5 V	1.2	2.4	4.8	kΩ
$R_{(opt2)}$		V_{CC} = 3 V/5 V	1.8	3.6	7.2	kΩ
$R_{(opt3)}$		V_{CC} = 3 V/5 V	3.6	7.3	14.6	kΩ
$R_{(opt4)}$		V_{CC} = 3 V/5 V	5.5	11	22	kΩ
$R_{(opt5)}$		V_{CC} = 3 V/5 V	11	22	44	kΩ
$R_{(opt6)}$		V_{CC} = 3 V/5 V	22	44	88	kΩ
$R_{(opt7)}$		V_{CC} = 3 V/5 V	33	66	132	kΩ
$R_{(opt8)}$		V_{CC} = 3 V/5 V	55	110	220	kΩ
$R_{(opt9)}$		V_{CC} = 3 V/5 V	77	154	310	kΩ
$R_{(opt10)}$		V_{CC} = 3 V/5 V	100	200	400	kΩ

NOTE 12: Optional resistors $R_{(optx)}$ for pulldown or pullup are not programmed in standard OTP/EPROM devices P/E 325A.

electrical characteristics over recommended operating free-air temperature range (unless otherwise noted) (continued)

input frequency – Port 0: P0.1; Timer/Port: CIN, TP0.5

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
$f_{(IN)}$	Input frequency	P0.x, CIN, TP.5		DC		$f_{(system)}$	MHz
$t_{(H)}$ or $t_{(L)}$	High level or low level time		3 V	300			ns
			5 V	125			ns

output frequency

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
f_{XBUF}		XBUF, C_L = 20 pF				$f_{(system)}$	MHz
t_{Xdc}	Duty cycle of O/P frequency	XBUF, C_L = 20 pF, V_{CC} = 3 V/5 V	f_{MCLK} = 1.1 MHz	40%		60%	
			$f_{XBUF} = f_{ACLK}$	35%		65%	
			$f_{XBUF} = f_{ACLK/n}$		50%		

external interrupt timing

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{(int)}$	Port P0: External trigger signal for the interrupt flag (see Notes 13 and 14)	1.5			cycle

NOTES: 13. The external signal sets the interrupt flag every time $t_{(int)}$ is met. It may be set even with trigger signals shorter than $t_{(int)}$. The conditions to set the flag must be met independently of this timing constraint. Input frequency ($t_{(int)}$) is defined in MCLK cycles.

14. The external signal needs additionally a timing resulting from the maximum input frequency constraint.

RAM

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
V_{RAMh}	CPU halted (see Note 15)		1.8			V

NOTE 15: This parameter defines the minimum supply voltage when the data in the program memory RAM remains unchanged. No program execution should take place during this supply voltage condition.

electrical characteristics over recommended operating free-air temperature range (unless otherwise noted) (continued)

DCO

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
$f_{(NOM)}$	DCO	N_{DCO} = 1A0h, FN_4=FN_3=FN_2=0	V_{CC} = 3 V/5 V		1		MHz
$f_{(NOM)}$	f_{DCO3}	N_{DCO} = 00 0110 0000, FN_4=FN_3=FN_2=0	V_{CC} = 3 V	0.15		0.6	MHz
			V_{CC} = 5 V	0.18		0.62	
	f_{DCO26}	N_{DCO} = 11 0100 0000 FN_4=FN_3=FN_2=0	V_{CC} = 3 V	1.25		4.7	
			V_{CC} = 5 V	1.45		5.5	
$2xf_{(NOM)}$	f_{DCO3}	N_{DCO} = 00 0110 0000, FN_4=FN_3=0, FN_2=1	V_{CC} = 3 V	0.36		1.05	MHz
			V_{CC} = 5 V	0.39		1.2	
	f_{DC26}	N_{DCO} = 11 0100 0000, FN_4=FN_3=0, FN_2=1	V_{CC} = 3 V	2.5		8.1	
			V_{CC} = 5 V	3		9.9	
$3xf_{(NOM)}$	f_{DCO3}	N_{DCO} = 00 0110 0000, FN_4=0, FN_3= 1, FN_2=X	V_{CC} = 3 V	0.5		1.5	MHz
			V_{CC} = 5 V	0.6		1.8	
	f_{DCO26}	N_{DCO} = 11 0100 0000, FN_4= 0, FN_3=1, FN_2=X	V_{CC} = 3 V	3.7		11	
			V_{CC} = 5 V	4.5		13.8	
$4xf_{(NOM)}$	f_{DCO3}	N_{DCO} = 00 0110 0000 FN_4 =1, FN_3=FN_2=X	V_{CC} = 3 V	0.7		1.85	MHz
			V_{CC} = 5 V	0.8		2.4	
	f_{DCO26}	N_{DCO} = 11 0100 0000, FN_4=1, FN_3=FN_2=X	V_{CC} = 3 V	4.8		13.3	
			V_{CC} = 5 V	6		17.7	
N_{DCO}		$f_{MCLK} = f_{NOM}$, FN_4=FN_3=FN_2=0	V_{CC} = 3 V/5 V	A0h	1A0h	340h	
S		$f_{NDCO+1} = S \times f_{NDCO}$	V_{CC} = 3 V/5 V	1.07		1.13	

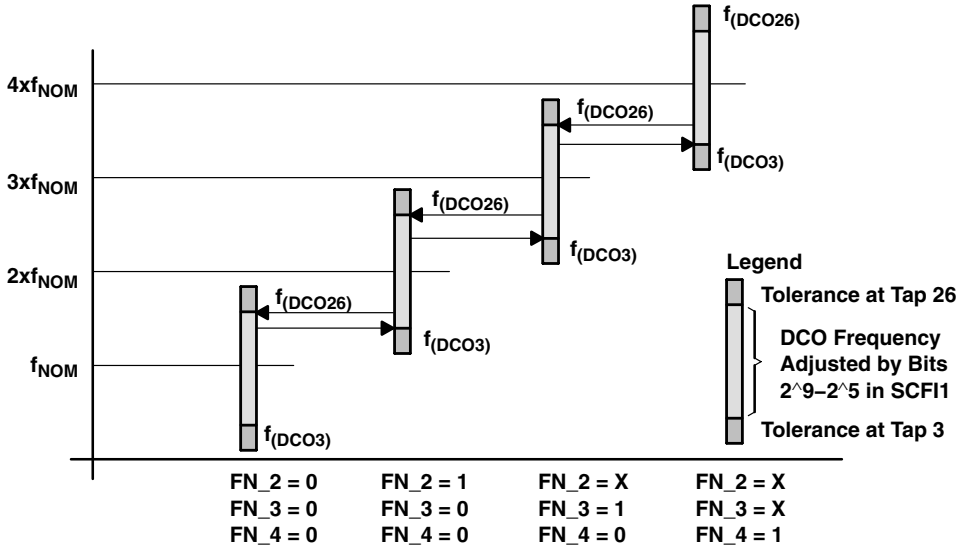


Figure 2

electrical characteristics over recommended operating free-air temperature range (unless otherwise noted) (continued)

crystal oscillator

PARAMETER		TEST CONDITIONS	MIN	NOM	MAX	UNIT
$C_{(Xin)}$	Integrated capacitance at input	V_{CC} = 3 V/5 V		12		pF
$C_{(Xout)}$	Integrated capacitance at output	V_{CC} = 3 V/5 V		12		pF

PUC/POR

PARAMETER		TEST CONDITIONS		MIN	NOM	MAX	UNIT
$t_{(POR_delay)}$	POR		V_{CC} = 3 V/5 V		150	250	µs
$V_{(POR)}$	POR	T_A = –40°C		1.5		2.4	V
		T_A = 25°C		1.2		2.1	V
		T_A = 85°C		0.9		1.8	V
$V_{(min)}$	POR			0		0.4	V
$t_{(reset)}$	PUC/POR	Reset is accepted internally		2			µs

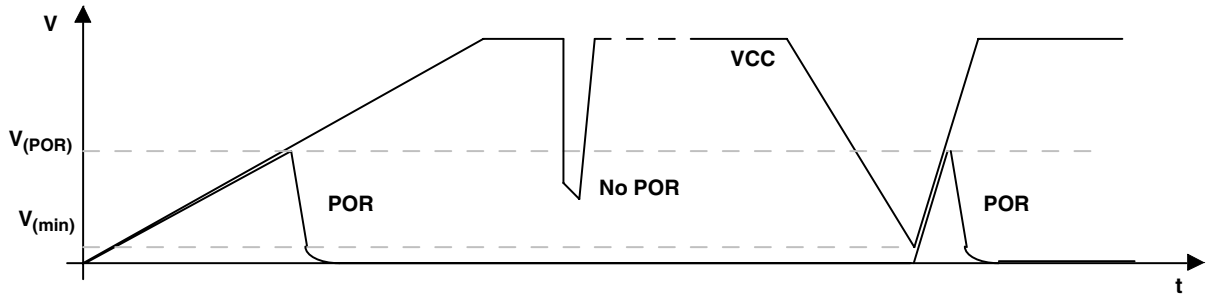


Figure 3. Power-On Reset (POR) vs Supply Voltage

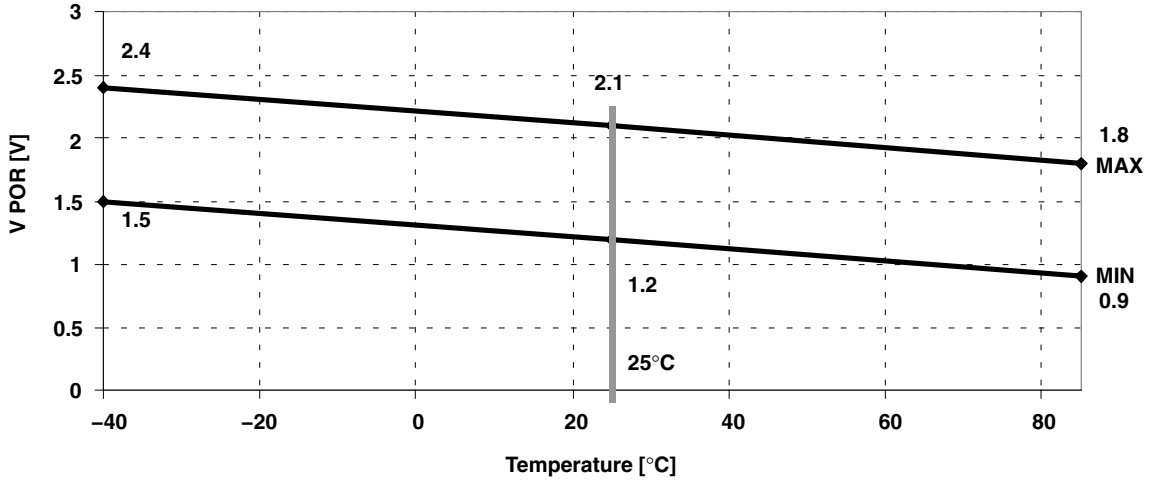


Figure 4. $V_{(POR)}$ vs Temperature

TEXAS INSTRUMENTS
POST OFFICE BOX 655303 • DALLAS, TEXAS 75265
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electrical characteristics over recommended operating free-air temperature range (unless otherwise noted) (continued)

LCD

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
$V_{O(HLCD)}$	Output 1 (HLCD)	$I_{(HLCD)}$ <= 10 nA	V_{CC} = 3 V/5 V	V_{CC}–0.125		V_{CC}	V
$V_{O(LLCD)}$	Output 0 (LLCD)	$I_{(LLCD)}$ <= 10 nA		V_{SS}		V_{SS}+0.125	
$I_{I(R03)}$	Input leakage	R03 = V_{SS}, No load at all seg and com pins	V_{CC} = 3 V/5 V			±20	nA
$I_{I(R13)}$		R13 = V_{CC}/ 3, No load at all seg and com pins					
$I_{I(R23)}$		R23 = 2 V_{CC}/ 3, No load at all seg and com pins					
$r_{o(Rx3\ to\ Sxx)}$	Resistance	$I_{(SXX)}$ = –3 µA,	V_{CC} = 3 V/5 V			50	kΩ

comparator (Timer/Port)

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
$I_{(com)}$	Comparator (Timer/Port)	CPON = 1	V_{CC} = 3 V		250	350	µA
			V_{CC} = 5 V		450	600	
$V_{ref(com)}$	Internal reference voltage at (–) terminal	CPON = 1	V_{CC} = 3 V/5 V	0.23×V_{CC}	0.25×V_{CC}	0.26×V_{CC}	V
$V_{hys(com)}$	Input hysteresis (comparator)	CPON = 1	V_{CC} = 3 V		5	37	mV
			V_{CC} =5 V		10	42	

wake-up LPM3

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
$t_{(LPM3)}$	Delay time	f = 1 MHz	V_{CC} = 3 V			6	µs
			V_{CC} = 5 V				
		f = 2 MHz	V_{CC} = 3 V			6	
			V_{CC} = 5 V				
		f = 3 MHz	V_{CC} = 5 V			6	

ADC supply current ($f_{(ADCLK)}$ = 1 MHz)

PARAMETER		TEST CONDITIONS		MIN	NOM	MAX	UNIT
$I_{(ADC)}$	ADC current	SV_{CC} on, current source off,	V_{CC} = 3 V		200	400	µA
$I_{(ADC)}$		SV_{CC} on, current source off,	V_{CC} = 5 V		300	740	µA

SV_{CC} (switched AV_{CC})

PARAMETER		TEST CONDITIONS		MIN	NOM	MAX	UNIT
$V_{(SVCC)}$		SV_{CC} on, $I_{(SVCC)}$ = –8 mA,	V_{CC} = 2.5 V	V_{CC}–0.2 V		V_{CC}	V
$I_{(SVCC)}$		SV_{CC} off, SV_{CC} = 0 V,	V_{CC} = 5 V			±0.1	µA
$Z_{(SVCC)}$	Input impedance	SV_{CC} off,	V_{CC} = 3 V/5 V	40		100	kΩ

electrical characteristics over recommended operating free-air temperature range (unless otherwise noted) (continued)

current source (ADC)

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
$V_{(Rext)}$	Voltage, (Rext)	$V_{(Rext)} = V_{(SVCC)} - V_{(RI)}$, $I_{(RI)}$ = 6 mA,	V_{CC} = 3 V/5 V,	0.246 × $V_{(SVCC)}$	0.249 × $V_{(SVCC)}$	0.252 × $V_{(SVCC)}$	V
$R_{(ext)}$	External resistor		V_{CC} = 3 V/5 V	95		1600	Ω
ΔI_S	Load compliance	VA0..A3 = 0 .. 0.4 × $V_{(SVCC)}$, I_S = $V_{(Rext)}/R_{(ext)}$ = 1 mA	V_{CC} = 3 V,	–1		1	µA
		VA0..A3 = 0 .. 0.4 × $V_{(SVCC)}$, $I_S = V_{(Rext)}/R_{(ext)}$ = 6 mA	V_{CC} = 3 V,	–3.2		3.2	µA
		VA0..A3 = 0 .. 0.5 × $V_{(SVCC)}$ $I_S = V_{(Rext)}/R_{(ext)}$= 1 mA	V_{CC} = 5 V,	–1.5		1.5	µA
		VA0..A3 = 0 .. 0.5 × $V_{(SVCC)}$ $I_S = V_{(Rext)}/R_{(ext)}$= 6 mA	V_{CC} = 5 V,	–3.2		3.2	µA

A/D converter ($f_{(ADCLK)}$ = 1 MHz)

PARAMETER		TEST CONDITIONS			MIN	TYP	MAX	UNIT
Resolution						12 + 2		bits
$f_{(con)}$	Conversion frequency	$f_{(con)} = f_{(ADCLK)}$	12-bit conversion	V_{CC} = 3 V/5 V	0.1		1.5	MHz
			12+2-bit conversion		0.14		1.5	
$f_{(concyc)}$	Conversion cycles	$f_{(ADCLK)} = f_{(MCLK)}/N$	12-bit conversion	V_{CC} = 3 V/5 V		96		cycles of ADCLK
			12+2-bit conversion			132		
LSB Voltage				V_{CC} = 3 V/5 V		0.000061×V_{SVCC}		V
INL_1	Integral nonlinearity, (see Note 18)	0 ≤ DDV ≤ 127		V_{CC} = 3 V/5 V	–2		2	LSB
INL_2		128 ≤ DDV ≤ 255		V_{CC} = 3 V/5 V	–3		3	LSB
INL_3		256 ≤ DDV ≤ 2047		V_{CC} = 3 V/5 V	–7		7	LSB
INL_4		2048 ≤ DDV ≤ 4095		V_{CC} = 3 V/5 V	–10		10	LSB
DNL	Differential nonlinearity, (see Note 19)			V_{CC} = 3 V/5 V	–1		1	LSB
dN/dT	Temperature stability	$V_{(Rext)}/R_{(ext)}$ = 6mA, Range A		V_{CC} = 3 V/5 V		0.008		LSB/°C
		Range B				0.015		
$dN/dV_{(SVCC)}$	$V_{(SVCC)}$rejection ratio	Range A, B, $V_{(Rext)}/R_{(ext)}$ = 1 mA, SV_{CC} ±10%		V_{CC} = 3 V/5 V			1.25	LSB/V
Conversion offset 12 bit analog input to digital value (see Note 16)		Range A		V_{CC} = 3 V/5 V	–1.2	–0.49	0.24	% FSR_A (see Note 17)
		Range B		V_{CC} = 3 V/5 V	–1.7	–0.6	0.49	% FSR_B (see Note 17)
		Range C		V_{CC} = 3 V/5 V	–1.8	–0.6	0.6	% FSR_C (see Note 17)
		Range D		V_{CC} = 3 V/5 V	–1.7	0.6	0.49	% FSR_D (see Note 17)
Conversion offset 14 bit analog input to digital value (see Note 16)		Range ABCD		V_{CC} = 3 V/5 V	–0.27	–0.06	0.13	%FSR_{ABCD} (see Note 17)
Slope 12 bit				V_{CC} = 3 V/5 V	0.9925	1	1.0075	
Slope 14 bit				V_{CC} = 3 V/5 V	0.9982	1	1.0018	
$C_{(IN)}$	Input capacitance			V_{CC} = 3 V/5 V		40	45	pF
$R_{(SIN)}$	Serial input resistance			V_{CC} = 3 V/5 V		2		kΩ

NOTES: 16. Offset referred to full scale 12/14 bit
17. FSRx: full scale range, separate for the four 12-bit ranges and the 14-bit (12+2) range.
18. DDV is short form of delta digital value. The DDV is a span of conversion results. It is assumed that the conversion is of 12 bit not 12+2 bit.
19. DNL is valid for all 12-bit ranges and the 14-bit (12+2) range.

electrical characteristics over recommended operating free-air temperature range (unless otherwise noted) (continued)

JTAG

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
$f_{(TCK)}$	JTAG/Test	TCK frequency	V_{CC} = 3 V	DC		5	MHz
			V_{CC} = 5 V	DC		10	
$R_{(TEST)}$		Pullup resistors on TMS, TCK, TDI (see Note 20)	V_{CC} = 3 V/ 5 V	25	60	90	kΩ
$V_{(FB)}$	JTAG/Fuse (see Note 21)	Fuse blow voltage, C versions (see Note 22)	V_{CC} = 3 V/ 5 V	5.5		6	V
		Fuse blow voltage, E/P versions (see Note 22)	V_{CC} = 3 V/ 5 V	11		12	
$I_{(FB)}$		Supply current on TDI to blow fuse				100	mA
$t_{(FB)}$		Time to blow the fuse				1	ms
$V_{(PP)}$	EPROM (E) and OTP(P) – versions only	Programming voltage, applied to TDI/VPP		12	12.5	13	V
$I_{(PP)}$		Current from programming voltage source				70	mA
$t_{(pps)}$		Programming time, single pulse		5			ms
$t_{(ppf)}$		Programming time, fast algorithm			100		µs
P_n		Number of pulses for successful programming		4		100	
		Data retention T_J < 55°C		10			year
$t_{(erase)}$	EPROM (E) versions only	Erase time wave length 2537 Å at 15 Ws/cm^2 (UV lamp of 12 mW/ cm^2)		30			min
		Write/Erase cycles		1000			cycles

NOTES: 20. The TMS and TCK pullup resistors are implemented in all C-, P-, and E-versions. The pullup resistor on TDI is implemented in C-versions only.

21. Once the JTAG fuse is blown, no further access to the MSP430 JTAG/test feature is possible. The JTAG block switches to by-pass mode.

22. The voltage supply to blow the JTAG fuse is applied to TDI/VPP pin when fuse blowing is desired.

TYPICAL CHARACTERISTICS

DIGITAL CONTROLLED OSCILLATOR FREQUENCY vs OPERATING FREE-AIR TEMPERATURE

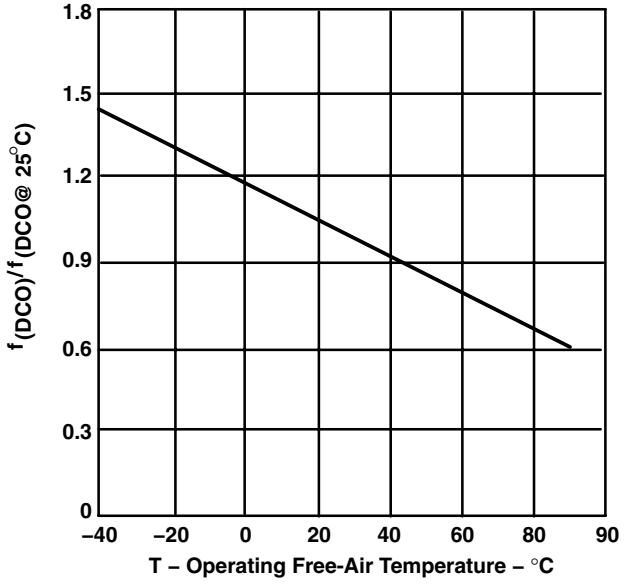


Figure 5

DIGITAL CONTROLLED OSCILLATOR FREQUENCY vs SUPPLY VOLTAGE

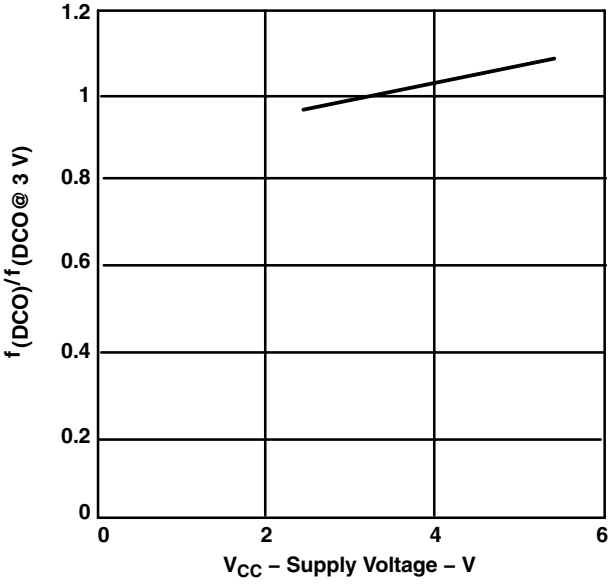
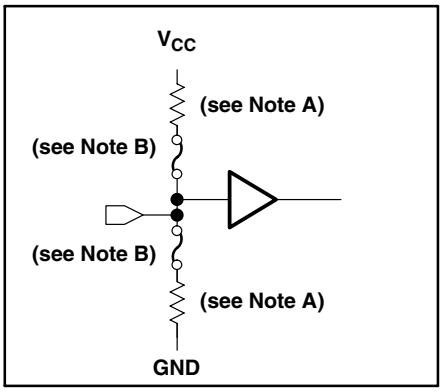


Figure 6

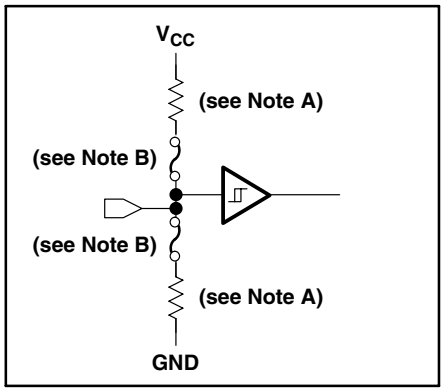
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TYPICAL CHARACTERISTICS

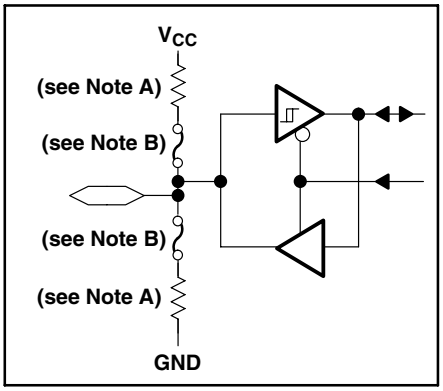
typical input/output schematics



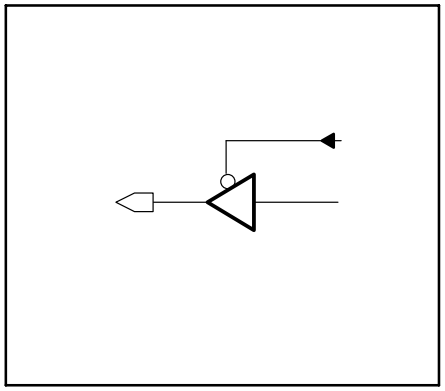
CMOS INPUT ($\overline{\text{RST}}$/NMI)



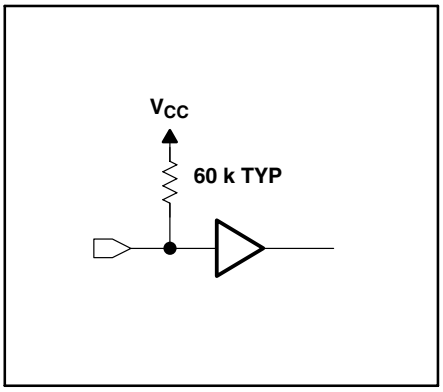
CMOS SCHMITT-TRIGGER INPUT (CIN)



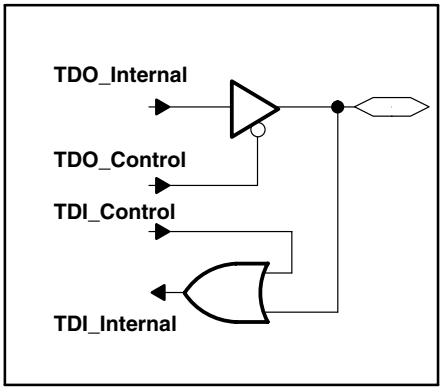
I/O WITH SCHMITT-TRIGGER INPUT (P0.x, TP5)



CMOS 3-STATE OUTPUT
(TP0–4, XBUF)



MSP430C32x: TMS, TCK, TDI
MSP430P/E325A: TMS, TCK

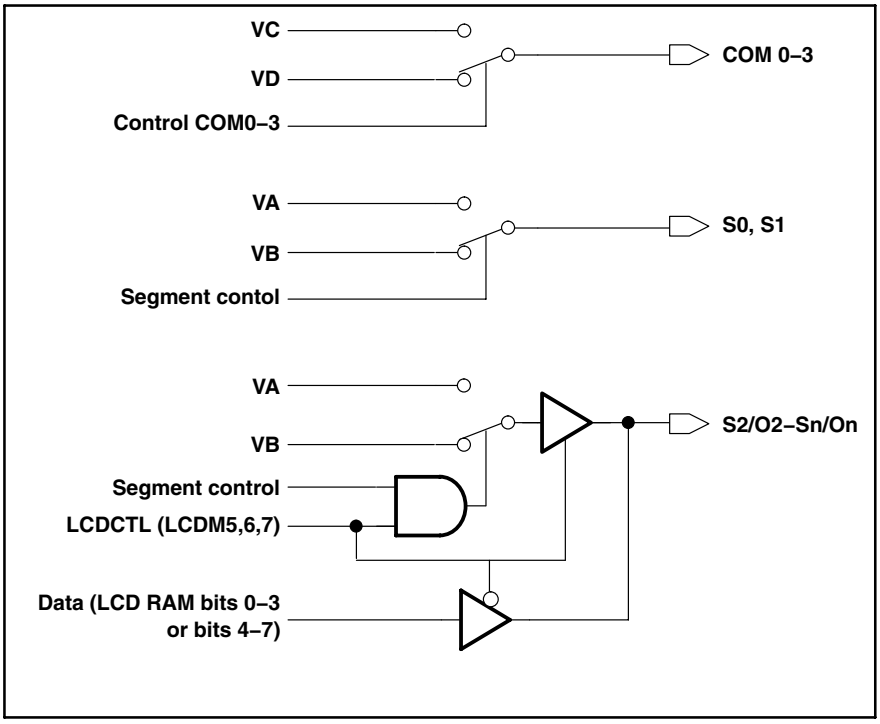


MSP430C32x: TDO/TDI
MSP430P/E325A: TDO/TDI

NOTES: A. Optional selection of pullup or pulldown resistors with ROM (masked) versions. Anti-parallel diodes are connected between AV_{SS} and DV_{SS}.
B. Fuses for the optional pullup and pulldown resistors can only be programmed at the factory.

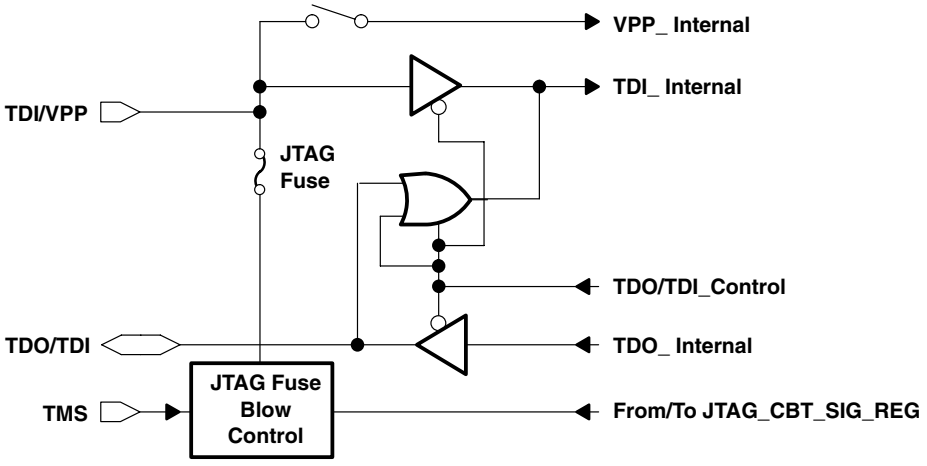
TYPICAL CHARACTERISTICS

typical input/output schematics



LCD OUTPUT (COM0–4, Sn, Sn/On)

NOTE: The signals VA, VB, VC, and VD come from the LCD module analog voltage generator.



NOTES: A. During programming activity and when blowing the JTAG enable fuse, the TDI/VPP terminal is used to apply the correct voltage source. The TDO/TDI terminal is used to apply the test input data for JTAG circuitry.

B. The TDI/VPP terminal of the ’P325A and ’E325A does not have an internal pullup resistor. An external pulldown resistor is recommended to avoid a floating node which could increase the current consumption of the device.

C. The TDO/TDI terminal is in a high-impedance state after POR. The ’P325A and ’E325A needs a pullup or a pulldown resistor to avoid floating a node which could increase the current consumption of the device.

Figure 7. MSP430P325A/E325A: TDI/VPP, TDO/TDI

TYPICAL CHARACTERISTICS

JTAG fuse check mode

MSP430 devices that have the fuse on the TDI/VPP terminal have a fuse check mode that tests the continuity of the fuse the first time the JTAG port is accessed after a power-on reset (POR). When activated, a fuse check current, I_{TF}, of 1 mA at 3 V, 2.5 mA at 5 V can flow from the TDI/VPP pin to ground if the fuse is not burned. Care must be taken to avoid accidentally activating the fuse check mode and increasing overall system power consumption.

Activation of the fuse check mode occurs with the first negative edge on the TMS pin after power up or if TMS is being held low during power up. The second positive edge on the TMS pin deactivates the fuse check mode. After deactivation, the fuse check mode remains inactive until another POR occurs. After each POR the fuse check mode has the potential to be activated.

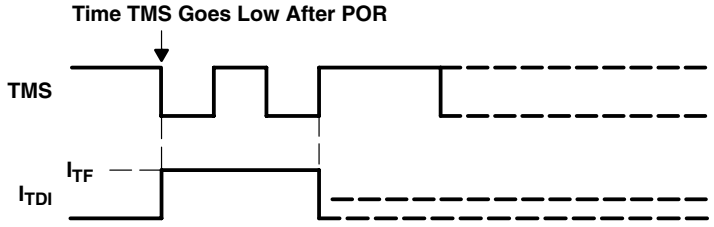


Figure 8. Fuse Check Mode Current, MSP430P/E325A, C32x

Care must be taken to avoid accidentally activating the fuse check mode, including guarding against EMI/ESD spikes that could cause signal edges on the TMS pin.

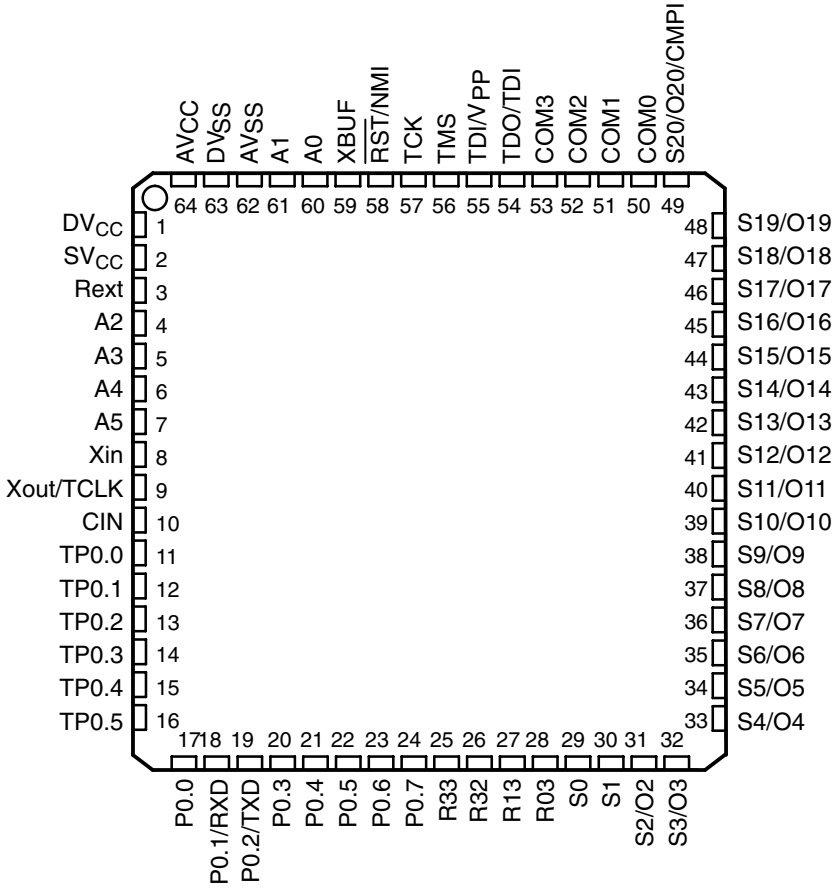
Configuration of TMS, TCK, TDI/VPP and TDO/TDI pins in applications.

	C3xx	P/E3xx
TDI	Open	68k, pulldown
TDO	Open	68k, pulldown
TMS	Open	Open
TCK	Open	Open

MECHANICAL DATA

pinning MSP43C323, MSP430C325, MSP430P325A (PM package)

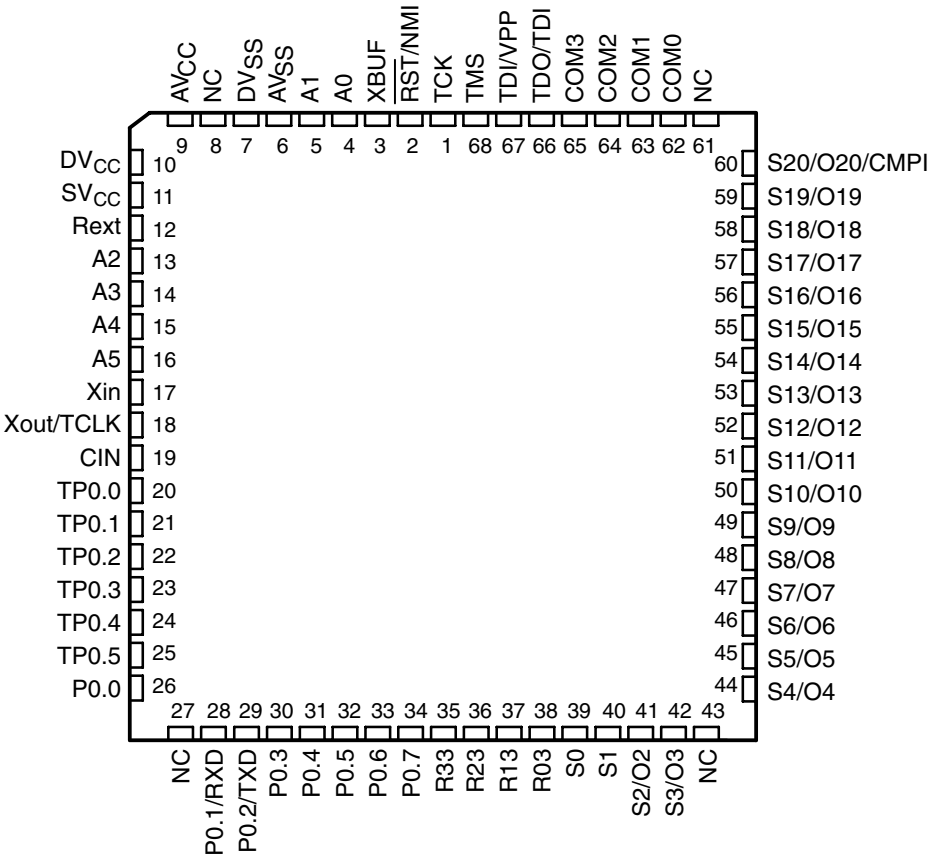
PM PACKAGE
(TOP VIEW)



pinning MSP43C323, MSP430C325, MSP430P325A (FN package)

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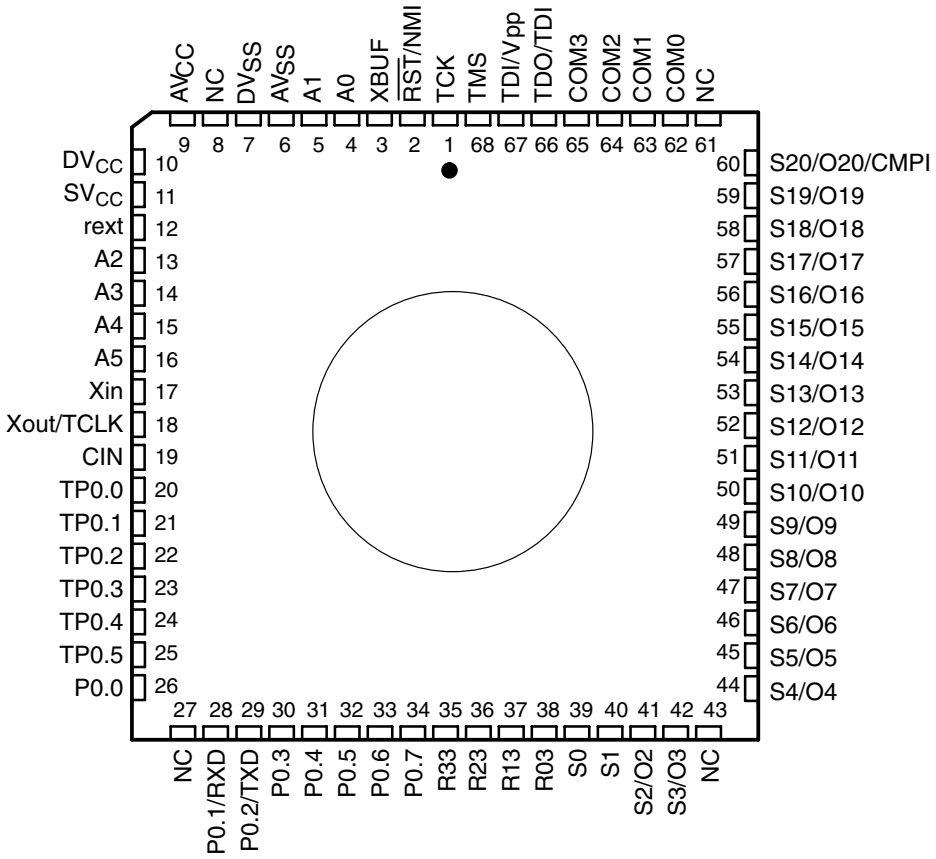
FN PACKAGE
(TOP VIEW)



NC – No internal connection

pinning PMS430E325A (FZ package)

FZ PACKAGE
(TOP VIEW)

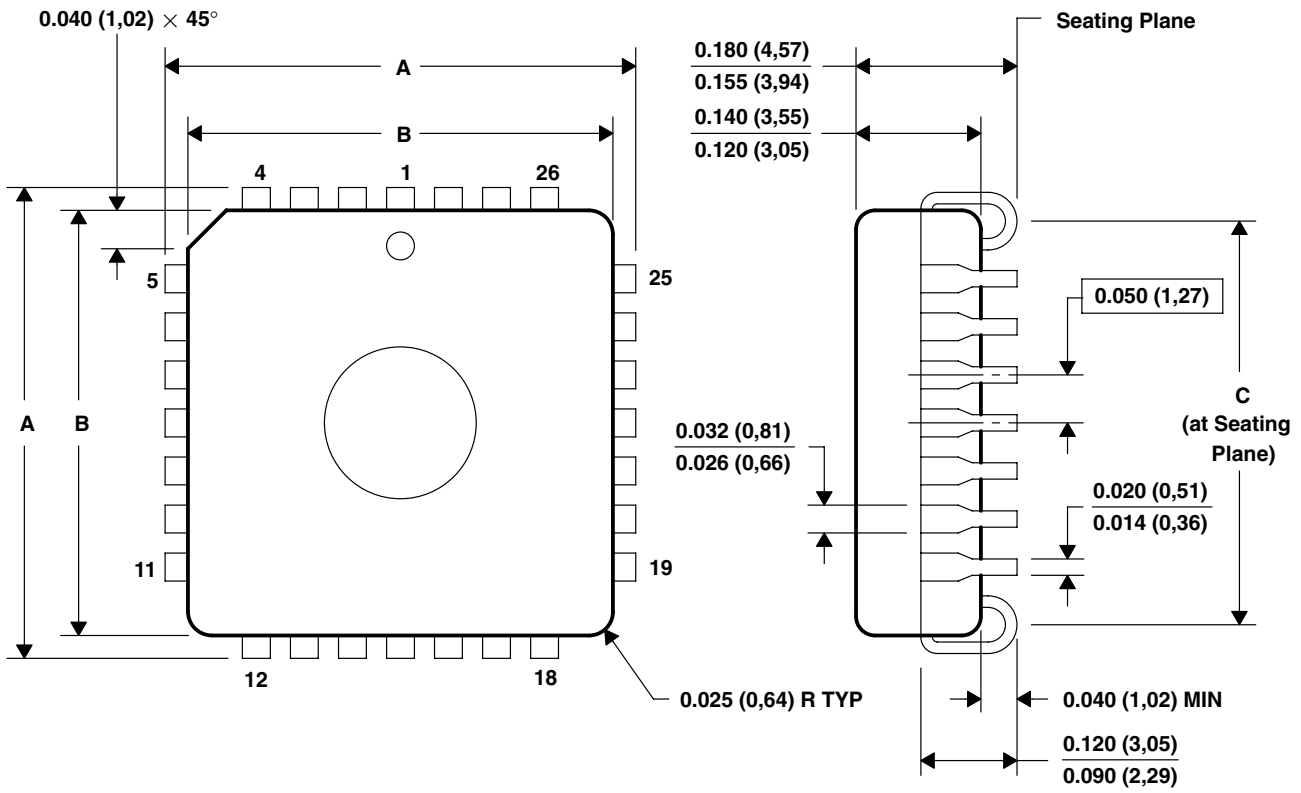


NC – No internal connection

MECHANICAL DATA

FZ (S-CQCC-J)** **J-LEADED CERAMIC CHIP CARRIER**

28 LEAD SHOWN



JEDEC OUTLINE	NO. OF PINS**	A		B		C	
		MIN	MAX	MIN	MAX	MIN	MAX
MO-087AA	28	0.485 (12,32)	0.495 (12,57)	0.430 (10,92)	0.455 (11,56)	0.410 (10,41)	0.430 (10,92)
MO-087AB	44	0.685 (17,40)	0.695 (17,65)	0.630 (16,00)	0.655 (16,64)	0.610 (15,49)	0.630 (16,00)
MO-087AC	52	0.785 (19,94)	0.795 (20,19)	0.730 (18,54)	0.765 (19,43)	0.680 (17,28)	0.740 (18,79)
MO-087AD	68	0.985 (25,02)	0.995 (25,27)	0.930 (23,62)	0.955 (24,26)	0.910 (23,11)	0.930 (23,62)

4040219/B 03/95

NOTES: A. All linear dimensions are in inches (millimeters).
B. This drawing is subject to change without notice.
C. This package can be hermetically sealed with a ceramic lid using glass frit.

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
MSP430P325AIFN	OBSOLETE	PLCC	FN	68		TBD	Call TI	Call TI	-40 to 85	M430P325A	
MSP430P325AIPG	OBSOLETE	QFP	PG	64		TBD	Call TI	Call TI	-40 to 85	M430P325A	
MSP430P325AIPM	OBSOLETE	LQFP	PM	64		TBD	Call TI	Call TI	-40 to 85	M430P325A	
PMS430E325FZ	OBSOLETE	JLCC	FZ	68		TBD	Call TI	Call TI			

(1) The marketing status values are defined as follows:
ACTIVE: Product device recommended for new designs.
LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.
NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.
PREVIEW: Device has been announced but is not in production. Samples may or may not be available.
OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.
TBD: The Pb-Free/Green conversion plan has not been defined.
Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.
Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.
Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

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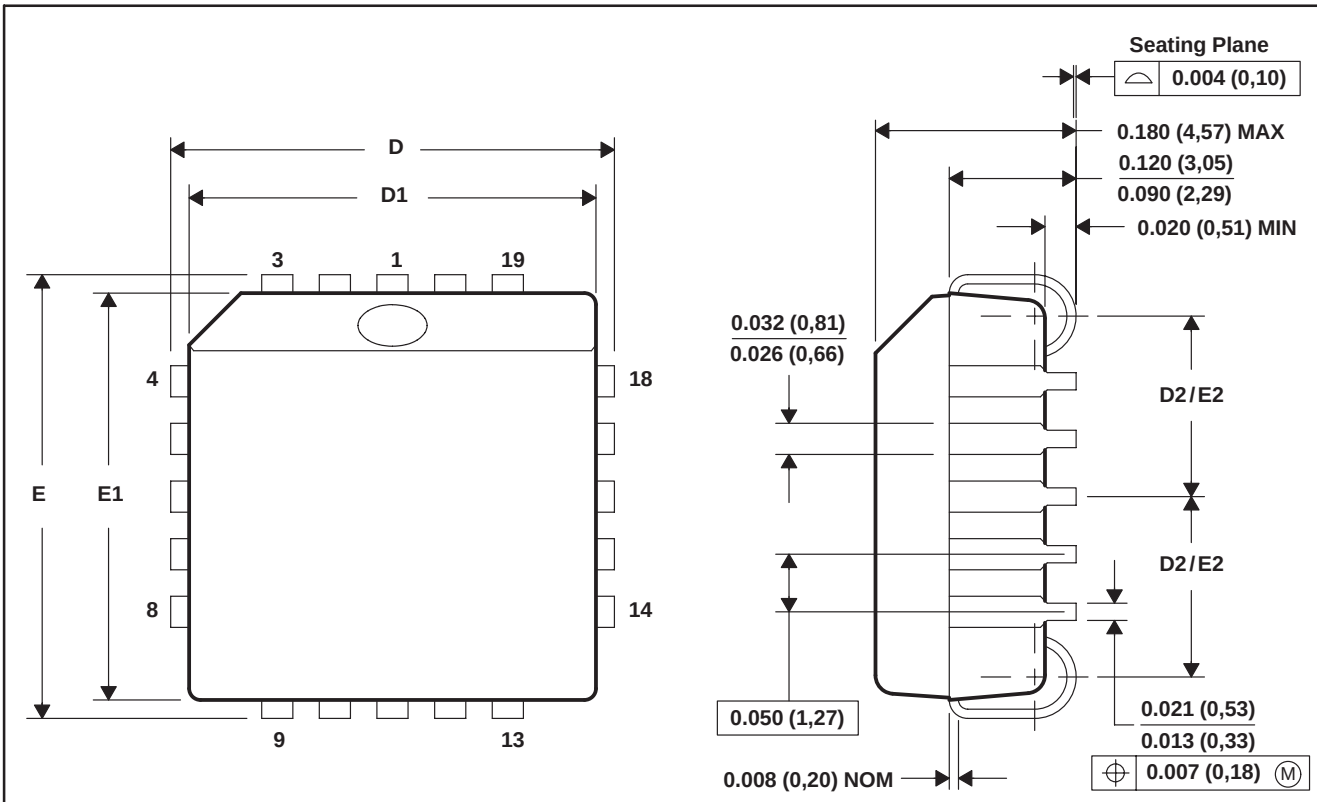
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MPLC004A – OCTOBER 1994

FN (S-PQCC-J**)

PLASTIC J-LEADED CHIP CARRIER

20 PIN SHOWN

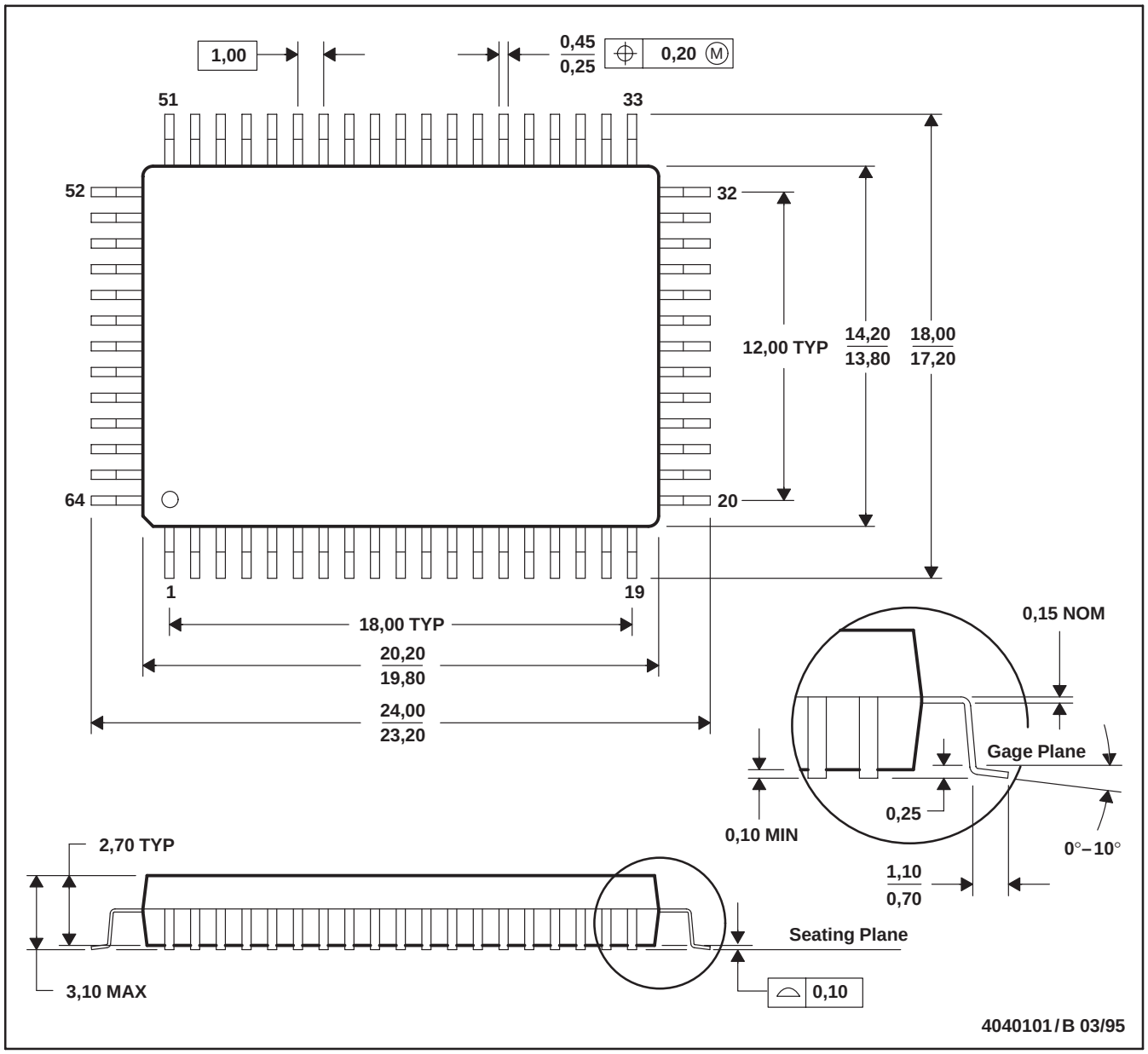


NO. OF PINS **	D/E		D1/E1		D2/E2	
	MIN	MAX	MIN	MAX	MIN	MAX
20	0.385 (9,78)	0.395 (10,03)	0.350 (8,89)	0.356 (9,04)	0.141 (3,58)	0.169 (4,29)
28	0.485 (12,32)	0.495 (12,57)	0.450 (11,43)	0.456 (11,58)	0.191 (4,85)	0.219 (5,56)
44	0.685 (17,40)	0.695 (17,65)	0.650 (16,51)	0.656 (16,66)	0.291 (7,39)	0.319 (8,10)
52	0.785 (19,94)	0.795 (20,19)	0.750 (19,05)	0.756 (19,20)	0.341 (8,66)	0.369 (9,37)
68	0.985 (25,02)	0.995 (25,27)	0.950 (24,13)	0.958 (24,33)	0.441 (11,20)	0.469 (11,91)
84	1.185 (30,10)	1.195 (30,35)	1.150 (29,21)	1.158 (29,41)	0.541 (13,74)	0.569 (14,45)

4040005/B 03/95

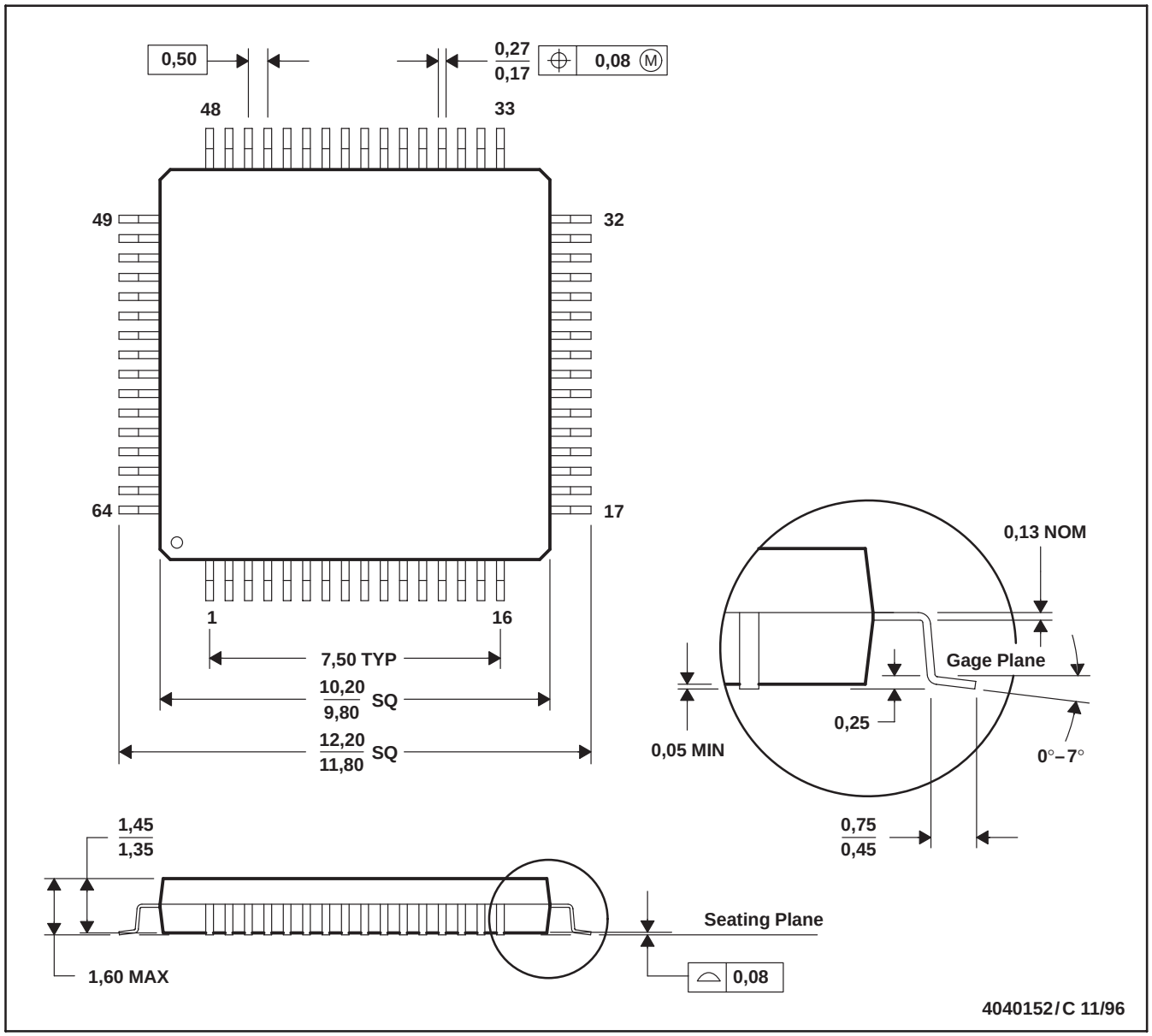
NOTES: A. All linear dimensions are in inches (millimeters).
B. This drawing is subject to change without notice.
C. Falls within JEDEC MS-018

PG (R-PQFP-G64) **PLASTIC QUAD FLATPACK**



NOTES: A. All linear dimensions are in millimeters.
B. This drawing is subject to change without notice.
C. Contact field sales office to determine if a tighter coplanarity requirement is available for this package.

PM (S-PQFP-G64) **PLASTIC QUAD FLATPACK**



NOTES: A. All linear dimensions are in millimeters.
B. This drawing is subject to change without notice.
C. Falls within JEDEC MS-026
D. May also be thermally enhanced plastic with leads connected to the die pads.

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