

Precision Voltage Reference

FEATURES

- ◆ Very High Accuracy: +2.5 V Output, $\pm 200 \mu\text{V}$
- ◆ Extremely Low Drift: $0.6 \text{ ppm}/^\circ\text{C}$ (25°C to $+85^\circ\text{C}$)
- ◆ Low Warm-up Drift: 1 ppm Typical
- ◆ Excellent Stability: 6 ppm/1000 Hrs. Typical
- ◆ Excellent Line Regulation: 6 ppm/V Typical
- ◆ Hermetic 20-Terminal Ceramic LCC
- ◆ Military Processing Option

APPLICATIONS

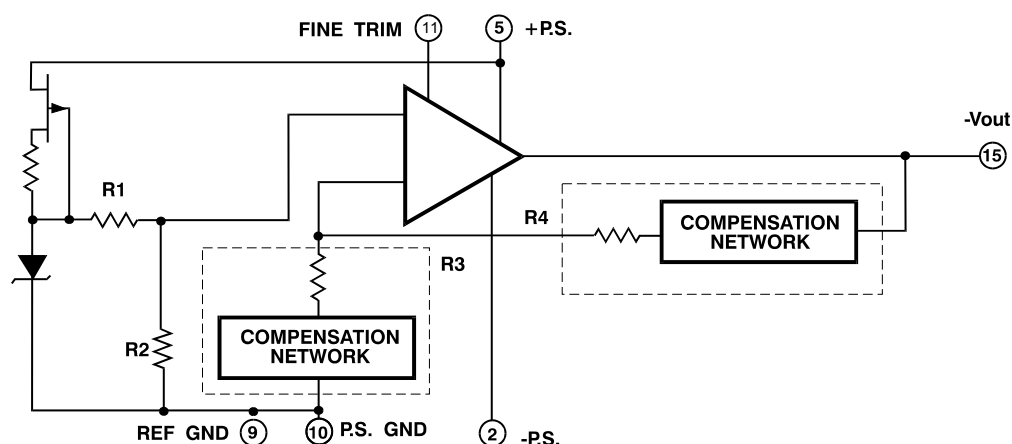
- ◆ Precision A/D and D/A Converters
- ◆ Transducer Excitation
- ◆ Accurate Comparator Threshold Reference
- ◆ High Resolution Servo Systems
- ◆ Digital Voltmeters
- ◆ High Precision Test and Measurement Instruments

DESCRIPTION

VRE202 Series Precision Voltage References provide ultrastable +2.5 V outputs with $\pm 200 \mu\text{V}$ initial accuracy and temperature coefficient as low as $0.6 \text{ ppm}/^\circ\text{C}$ over the full military temperature range. This improvement in accuracy is made possible by a unique, proprietary multipoint laser compensation technique. Significant improvements have been made in other performance parameters as well, including initial accuracy, warm-up drift, line regulation, and longterm stability, making the VRE202 series the most accurate and stable 2.5 V surface mount references available.

VRE202 devices are available in two operating temperature ranges, -25°C to $+85^\circ\text{C}$ and -55°C to $+125^\circ\text{C}$, and two electrical performance grades. All devices are packaged in 20-terminal ceramic LCC packages for maximum long-term stability. "M" versions are screened for high reliability and quality.

Figure 1. BLOCK DIAGRAM



SELECTION GUIDE

Model	Output (V)	Temperature Operating Range	Volt Deviation (MAX)
VRE202C	+2.5V	-25°C to $+85^\circ\text{C}$	$\pm 200 \mu\text{V}$
VRE202CA	+2.5V	-25°C to $+85^\circ\text{C}$	$\pm 100 \mu\text{V}$
VRE202M	+2.5V	-55°C to $+125^\circ\text{C}$	$\pm 400 \mu\text{V}$
VRE202MA	+2.5V	-55°C to $+125^\circ\text{C}$	$\pm 200 \mu\text{V}$



**20-terminal Ceramic LCC
Package Style HD**

1. CHARACTERISTICS AND SPECIFICATIONS

ELECTRICAL SPECIFICATIONS

$V_{PS} = +15V$, $T = +25^{\circ}C$, $R_L = 10K\Omega$ UNLESS OTHERWISE NOTED.

Grade	C			CA			
Parameter	Min	Typ	Max	Min	Typ	Max	Units
ABSOLUTE MAXIMUM RATINGS							
Power Supply	+13.5		+22	*		*	V
Operating Temperature	-25		+85	*		*	°C
Storage Temperature	-65		+150	*		*	°C
Short Circuit Protection	Continuous			*			
OUTPUT VOLTAGE							
VRE202		+2.5			*		V
OUTPUT VOLTAGE ERRORS							
Initial Error			±300			±200	μV
Warmup Drift		2			1		ppm
T _{MIN} - T _{MAX} (Note1)			200			100	μV
Long-Term Stability		6			*		ppm/1000hrs.
Noise (0.1 - 10Hz)		1.5			*		μVpp
OUTPUT CURRENT							
Range	±10			*			mA
REGULATION							
Line		6	10		*	*	ppm/V
Load		3			*		ppm/mA
OUTPUT ADJUSTMENT							
Range		10			*		mV
Temperature Coefficient		4			*		μV/°C/mV
POWER SUPPLY CURRENT (Note 2)							
VRE202 +PS		5	7		*	*	mA
VRE202 -PS		5	7		*	*	mA

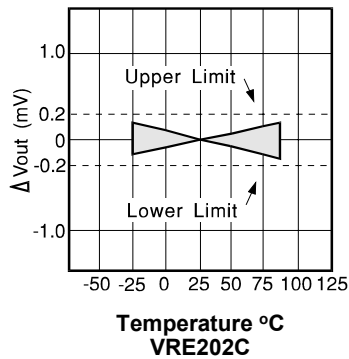
NOTES:

* Same as C Models.

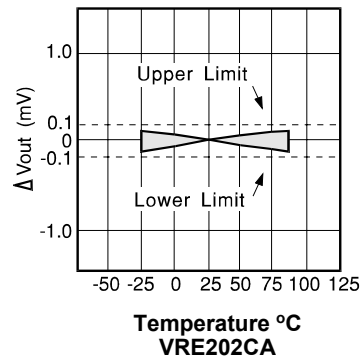
- Using the box method, the specified value is the maximum deviation from the output voltage at 25°C over the specified operating temperature range.
- The specified values are unloaded.

2. TYPICAL PERFORMANCE CURVES

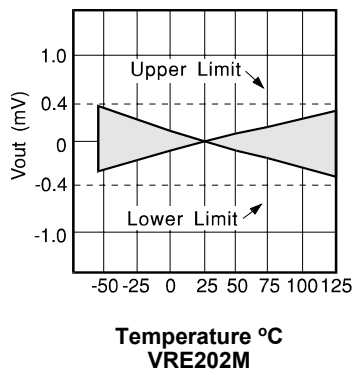
V_{OUT} vs. TEMPERATURE



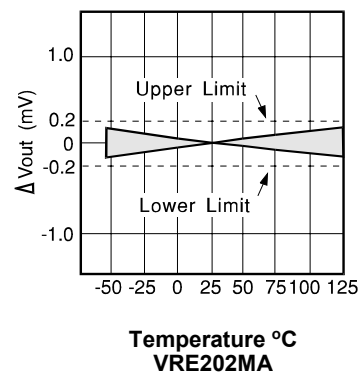
V_{OUT} vs. TEMPERATURE



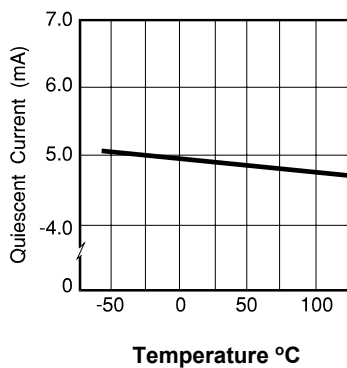
V_{OUT} vs. TEMPERATURE



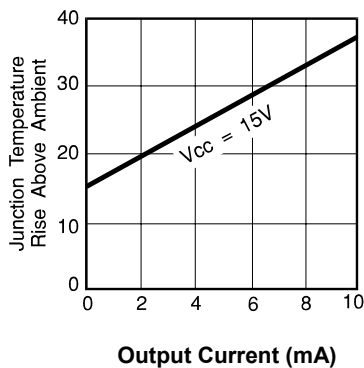
V_{OUT} vs. TEMPERATURE



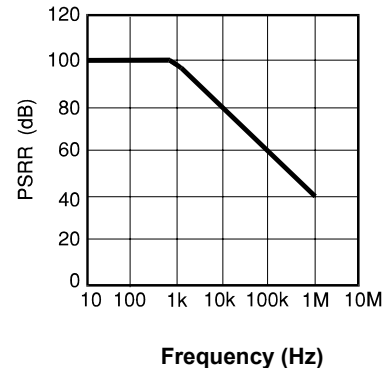
QUIESCENT CURRENT vs. TEMP



JUNCTION TEMP. RISE vs. OUTPUT CURRENT



PSRR vs. FREQUENCY



3. THEORY OF OPERATION

The following discussion refers to the block diagram in Figure 1. A FET current source is used to bias a 6.3 V zener diode. The zener voltage is divided by the resistor network R1 and R2. This voltage is then applied to the noninverting input of the operational amplifier which amplifies the voltage to produce a 2.5 V output. The gain is determined by the resistor networks R3 and R4: $G=1 + R4/R3$. The 6.3 V zener diode is used because it is the most stable diode over time and temperature.

The current source provides a closely regulated zener current, which determines the slope of the references' voltage vs. temperature function. By trimming the zener current a lower drift over temperature can be achieved. But since the voltage vs. temperature function is nonlinear this compensation technique is not well suited for wide temperature ranges.

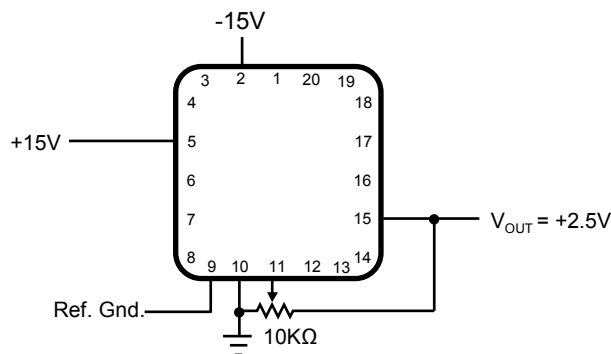
A nonlinear compensation network of thermistors and resistors is used in the VRE series voltage references. This proprietary network eliminates most of the nonlinearity in the voltage vs. temperature function. Then by adjusting the slope, a very stable voltage over wide temperature ranges is produced. This network is less than 2% of the overall network resistance so it has a negligible effect on long term stability. By using highly stable resistors in our network, a voltage reference is produced that also has very good long term stability.

4. APPLICATION INFORMATION

The proper connection of the VRE202 series voltage references with the optional trim resistors is shown below. Pay careful attention to the circuit layout to avoid noise pickup and voltage drops in the lines.

The VRE202 series voltage references have the ground terminal brought out on two pins (pin 9 and pin 10) which are connected together internally. This allows the user to achieve greater accuracy when using a socket. Voltage references have a voltage drop across their power supply ground pin due to quiescent current flowing through the contact resistance. If the contact resistance was constant with time and temperature, this voltage drop could be trimmed out. When the reference is plugged into a socket, this source of error can be as high as 20 ppm. By connecting pin 10 to the power supply ground and pin 9 to a high impedance ground point in the measurement circuit, the error due to the contact resistance can be eliminated. If the unit is soldered into place, the contact resistance is sufficiently small that it does not effect performance.

EXTERNAL CONNECTIONS



PIN CONFIGURATION

