PTH05000W —5-V Input

6-A, 5-V Input Non-Isolated Wide-Output Adjust Power Module



NOMINAL SIZE = 0.75 in x 0.5 in (19,05 mm x 12,7 mm)

Features

- Up to 6-A Output Current
- 5-V Input Voltage
- Wide-Output Voltage Adjust (0.9 V to 3.6 V)
- Efficiencies up to 94 %
- 160 W/in³ Power Density
- On/Off Inhibit
- Under-Voltage Lockout
- Output Current Limit
- Pre-Bias Startup Capability
- Over-Temperature Protection
- Surface Mountable
- Operating Temp: –40 to +85 °C
- Safety Agency Approvals (Pending): UL 1950, CSA 22.2 950, EN60950 & VDE

Description

The PTH05000 series of non-isolated power modules are small in size and high on performance. Using double-sided surface mount construction and synchronous rectification technology, these regulators deliver up to 6 A of output current while occupying a PCB area of about half the size of a standard postage stamp. They are an ideal choice for applications where space, performance and cost are important design constraints.

The series operates from an input voltage of 5 V to provide step-down power conversion to any output voltage over the range, 0.9 V to 3.6 V. The output voltage of the PTH05000W is set within this range using a single resistor.

Operating features include an on/off inhibit, output voltage adjust (trim), an output current limit, and over-temperature protection.

For high efficiency these parts employ a synchronous rectifier output stage. An output pre-bias holdoff capability ensures that the output will not sink current during startup.

Target applications include telecom, industrial, and general purpose circuits, including low-power dual-voltage systems that use a DSP, microprocessor, or ASIC.

Package options include both through-hole and surface mount configurations.

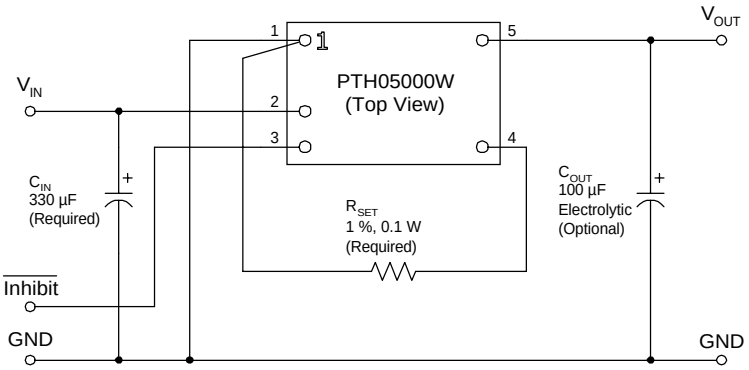
Pin Configuration

Pin	Function
1	GND
2	V_{in}
3	Inhibit *
4	V_o Adjust
5	V_{out}

* *Denotes negative logic:*
Open = Output On
Ground = Output Off

Standard Application

R_{set} = Required to set the output voltage to a value higher than 0.9 V. See spec. table for values.
C_{in} = Required 330 µF capacitor
C_{out} = Optional 100 µF capacitor



Ordering Information

Output Voltage (PTH05000❒xx)

Code	Voltage
W	0.9 V – 3.6 V (Adjust)

Package Options (PTH05000x❒❒) (1)

Code	Description	Pkg Ref. (2)
AH	Horiz. T/H	(EUS)
AS	SMD, Standard (3)	(EUT)

Notes: *(1) Add "T" to end of part number for tape and reel on SMD packages only.*
(2) Reference the applicable package reference drawing for the dimensions and PC board layout
(3) "Standard" option specifies 63/37, Sn/Pb pin solder material.

Pin Descriptions

Vin: The positive input voltage power node to the module, which is referenced to common GND.

Vout: The regulated positive power output with respect to the GND node.

GND: This is the common ground connection for the 'V_{in}' and 'V_{out}' power connections. It is also the 0 VDC reference for the 'Inhibit' and 'V_o Adjust' control input.

Inhibit: The Inhibit pin is an open-collector/drain negative logic input that is referenced to *GND*. Applying a low-level ground signal to this input disables the module's output and turns off the output voltage. When the Inhibit control is active, the input current drawn by the regulator is significantly reduced. If the Inhibit pin is left open-circuit, the module will produce an output whenever a valid input source is applied.

Vo Adjust: A 0.1 W 1 % resistor must be directly connected between this pin and the *GND* pin to set the output voltage to a value higher than 0.9 V. The temperature stability of the resistor should be 100 ppm/°C (or better). The set point range for the output voltage is from 0.9 V to 3.6 V. The resistor required for a given output voltage may be calculated from the following formula. If left open circuit, the output voltage will default to its lowest value. For further information on output voltage adjustment, consult the related application note.

$$R_{set} = 10\ k\Omega \cdot \frac{0.891\ V}{V_{out} - 0.9\ V} - 3.24\ k\Omega$$

The specification table gives the preferred resistor values for a number of standard output voltages.

Environmental & Absolute Maximum Ratings

Characteristics	Symbols	Conditions	Min	Typ	Max	Units
Operating Temperature Range	T_a	Over V_{in} Range	–40 (i)	—	+85	°C
Solder Reflow Temperature	T_{reflow}	Surface temperature of module body or pins			235 (ii)	°C
Storage Temperature	T_s	—	–40	—	+125	°C
Over Temperature Protection	OTP	IC junction temperature	—	150	—	°C
Mechanical Shock		Per Mil-STD-883D, Method 2002.3 1 msec, ½ sine, mounted	—	500	—	G's
Mechanical Vibration		Mil-STD-883D, Method 2007.2 20-2000 Hz	—	20	—	G's
Weight	—		—	2	—	grams
Flammability	—	Meets UL 94V-O				

Notes: *(i) For operation below 0 °C the external capacitors must have stable characteristics. Use either a low ESR tantalum, Os-con, or ceramic capacitor.*
(ii) During reflow of SMD package version do not elevate peak temperature of the module, pins or internal components above the stated maximum.

Electrical Specifications

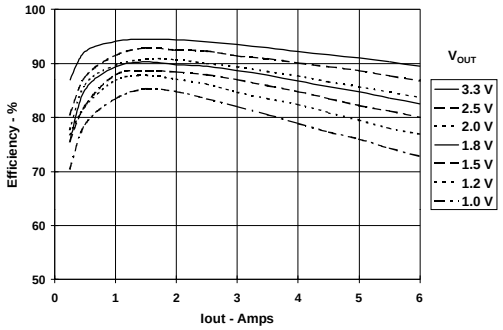
Unless otherwise stated, T_a =25 °C, V_{in} =5 V, V_o =3.3 V, C_{in} =330 µF, C_{out} =0 µF, and I_o =I_o(max)

Characteristics	Symbols	Conditions	PTH05000W Min	PTH05000W Typ	PTH05000W Max	Units
Output Current	I_o	0.9 V ≤ V_o ≤3.6 V, T_a =25 °C, natural convection	0	—	6 (1)	A
		T_a =60 °C, 200LFM	0	—	5.25 (1)	
Input Voltage Range	V_{in}	Over I_o range	4.5	—	5.5	V
Set-Point Voltage Tolerance	V_o tol		—	—	±2 (2)	%V_o
Temperature Variation	ΔReg_{temp}	–40 °C <T_a < +85 °C	—	±0.5	—	%V_o
Line Regulation	ΔReg_{line}	Over V_{in} range	—	±5	—	mV
Load Regulation	ΔReg_{load}	Over I_o range	—	±5	—	mV
Total Output Variation	ΔReg_{tot}	Includes set-point, line, load, –40 °C ≤ T_a ≤ +85 °C	—	—	±3 (2)	%V_o
Efficiency	η	V_{in} =5 V, I_o =4 A, R_{SET} = 475 Ω, V_o = 3.3 V	—	92	—	%
		R_{SET} = 2.32 kΩ, V_o = 2.5 V	—	90	—	
		R_{SET} = 4.87 kΩ, V_o = 2.0 V	—	88	—	
		R_{SET} = 6.65 kΩ, V_o = 1.8 V	—	87	—	
		R_{SET} = 11.5 kΩ, V_o = 1.5 V	—	84	—	
		R_{SET} = 26.1 kΩ, V_o = 1.2 V	—	82	—	
		R_{SET} = 84.5 kΩ, V_o = 1.0 V	—	79	—	
V_o Ripple (pk-pk)	V_r	20 MHz bandwidth, V_o ≥3.3 V	—	30	—	mVpp
		V_o ≤2.5 V	—	25	—	
Transient Response		1 A/µs load step, 50 to 100 % $I_o$max, V_o =1.8 V, C_{out} =100 µF				
	t_{tr}	Recovery time	—	70	—	µSec
	ΔV_{tr}	V_o over/undershoot	—	100	—	mV
Current Limit	I_{lim}	ΔV_o = –50 mV	—	13	—	A
Under-Voltage Lockout	UVLO	V_{in} increasing	—	3.8	4.3	V
		V_{in} decreasing	3.4	3.5	—	
Inhibit Control (pin 3)		Referenced to GND				
Input High Voltage	V_{IH}		V_{in} –0.5	—	Open (3)	V
Input Low Voltage	V_{IL}		–0.2	—	0.8	
Input Low Current	I_{IL}	Pin 3 to GND	—	–10	—	µA
Standby Input Current	I_{in} standby	pins 1 & 3 connected	—	1	—	mA
Switching Frequency	f_s	Over V_{in} and I_o ranges	—	700	—	kHz
External Input Capacitance	C_{in}		330 (4)	—	—	µF
External Output Capacitance	C_{out}	Capacitance value, non-ceramic	0	100 (5)	1,000 (6)	µF
		ceramic	0	—	300	
		Equiv. series resistance (non-ceramic)	4 (7)	—	—	mΩ
Reliability	MTBF	Per Bellcore TR-332 50 % stress, T_a =40 °C, ground benign	28	—	—	10^6 Hrs

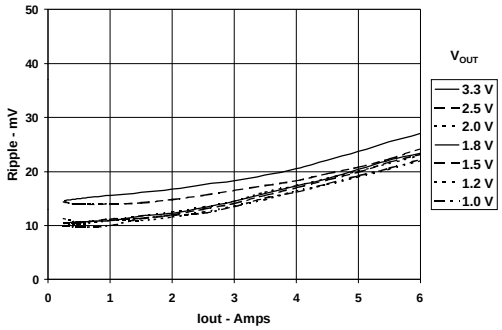
Notes: *(1) See SOA curves or consult factory for appropriate derating.*
(2) The set-point voltage tolerance is affected by the tolerance and stability of R_{SET}. The stated limit is unconditionally met if R_{SET} has a tolerance of 1 % with 100 ppm/°C or better temperature stability.
(3) The Inhibit control (pin 3) has an internal pull-up to Vin, and if left open-circuit the module will operate when input power is applied. A small low-leakage (<100 nA) MOSFET is recommended to control this input. See application notes for more information.
(4) The regulator requires a minimum of 330 µF input capacitor with a minimum 300 mArms ripple current rating. For further information, consult the related application note on Capacitor Recommendations.
(5) An external output capacitor is not required for basic operation. Adding 100 µF of distributed capacitance at the load will improve the transient response.
(6) This is the calculated maximum. The minimum ESR limitation will often result in a lower value. Consult the application notes for further guidance.
(7) This is the typical ESR for all the electrolytic (non-ceramic) output capacitance. Use 7 mΩ as the minimum when using max-ESR values to calculate.

Characteristic Data; V_{in} =5 V *(See Note A)*

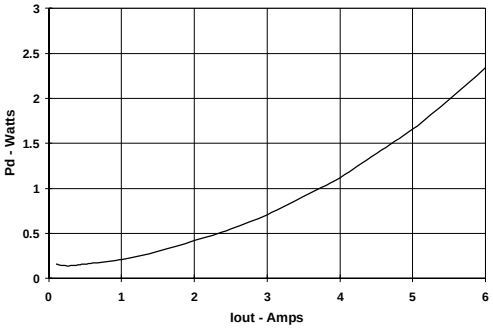
Efficiency vs Output Current



Ripple vs Output Current

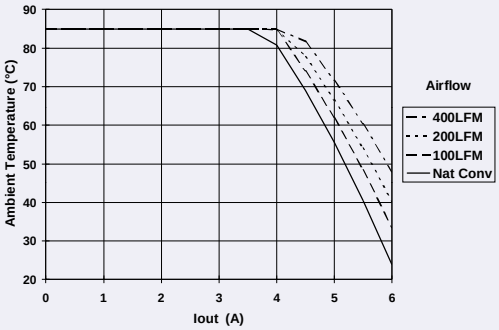


Power Dissipation vs Output Current



Safe Operating Area; V_{in} =5 V *(See Note B)*

All Output Voltages



Note A: *Characteristic data has been developed from actual products tested at 25 °C. This data is considered typical data for the Converter.*
Note B: *SOA curves represent the conditions at which internal components are at or below the manufacturer's maximum operating temperatures. Derating limits apply to modules soldered directly to a 4 in. × 4 in. double-sided PCB with 1 oz. copper.*

Capacitor Recommendations for the PTH05000W Wide-Output Adjust Power Modules

Input Capacitor

The recommended input capacitor(s) is determined by the 330 µF minimum capacitance and 300 mArms minimum ripple current rating.

Ripple current, less than 300 mΩ equivalent series resistance (ESR), and temperature are the major considerations when selecting input capacitors. Unlike polymer tantalum, regular tantalum capacitors have a recommended minimum voltage rating of 2 × (maximum DC voltage + AC ripple). This is standard practice to ensure reliability.

For improved ripple reduction on the input bus, ceramic capacitors [2] may used to complement electrolytic types, and achieve the minimum required capacitance.

Output Capacitors (Optional)

For applications with load transients (sudden changes in load current), regulator response will benefit from an external output capacitance. The recommended output capacitance of 100 µF will allow the module to meet its transient response specification (see product data sheet). For most applications, a high quality computer-grade aluminum electrolytic capacitor is adequate. These capacitors provide decoupling over the frequency range, 2 kHz to 150 kHz, and are suitable for ambient temperatures above 0 °C. For operation below 0 °C tantalum, ceramic or Os-Con type capacitors are recommended. When using one or more non-ceramic capacitors, the calculated equivalent ESR should be no lower than 4 mΩ (7 mΩ using the manufacturer's maximum ESR for a single capacitor). A list of preferred low-ESR type capacitors are identified in Table 1-1.

Ceramic Capacitors

Above 150 kHz the performance of aluminum electrolytic capacitors becomes less effective. To further improve the reflected input ripple current [2] or the output transient response, multilayer ceramic capacitors can also be added. Ceramic capacitors have very low ESR and their resonant frequency is higher than the bandwidth of the regulator. When used on the output their combined ESR is not critical as long as the total value of ceramic capacitance does not exceed 300 µF. Also, to prevent the formation of local resonances, do not place more than five identical ceramic capacitors in parallel with values of 10 µF or greater.

Tantalum Capacitors

Tantalum type capacitors can be used at both the input and output, and are recommended for applications where the ambient operating temperature can be less than 0 °C. The AVX TPS, Sprague 593D/594/595 and Kemet T495/T510 capacitor series are suggested over many other tantalum types due to their higher rated surge, power dissipation, and ripple current capability. As a caution many general purpose tantalum capacitors have considerably higher ESR, reduced power dissipation and lower ripple current capability. These capacitors are also less reliable as they have lower power dissipation and surge current ratings. Tantalum capacitors that do not have a stated ESR or surge current rating are not recommended for power applications.

When specifying Os-Con and polymer tantalum capacitors for the output, the minimum ESR limit will be encountered well before the maximum capacitance value is reached.

Capacitor Table

Table 1-1 identifies the characteristics of capacitors from a number of vendors with acceptable ESR and ripple current (rms) ratings. The recommended number of capacitors required at both the input and output buses is identified for each capacitor type.

This is not an extensive capacitor list. Capacitors from other vendors are available with comparable specifications. Those listed are for guidance. The RMS ripple current rating and ESR (at 100 kHz) are critical parameters necessary to insure both optimum regulator performance and long capacitor life.

Designing for Very Fast Load Transients

The transient response of the DC/DC converter has been characterized using a load transient with a di/dt of 1 A/µs. The typical voltage deviation for this load transient is given in the data sheet specification table using the optional value of output capacitance. As the di/dt of a transient is increased, the response of a converter's regulation circuit ultimately depends on its output capacitor decoupling network. This is an inherent limitation with any DC/DC converter once the speed of the transient exceeds its bandwidth capability. If the target application specifies a higher di/dt or lower voltage deviation, the requirement can only be met with additional output capacitor decoupling. In these cases special attention must be paid to the type, value and ESR of the capacitors selected.

If the transient performance requirements exceed that specified in the data sheet, the selection of output capacitors becomes more important. For further guidance consult the separate application note, "*Selecting Output Capacitors for PTH Products in High-Performance Applications.*"

Table 1-1: Input/Output Capacitors

Capacitor Vendor, Type/ Series (Style)	Capacitor Characteristics					Quantity		
	Working Voltage	Value (µF)	Max. ESR at 100 kHz	Max. Ripple Current at 85 °C (Irms)	Physical Size (mm)	Input Bus	Output Bus	Vendor Number
Panasonic WA, Poly-Aluminum (SMD) FC, Aluminum (SMD) FK, Aluminum (SMD) FC, Aluminum (Radial)	10 V 16 V 16 V 10 V	330 µF 330 µF 330 µF 330 µF	0.035 Ω 0.150 Ω 0.160 Ω 0.117 Ω	2800mA 670mA 600 mA 550 mA	8×6.9 10×10.2 8×10.2 8×11.5	1 1 1 1	≤3 ≤3 ≤3 ≤3	EEFWA1A121P EEVFC1C331P EEVFK1C331P EEUFC1A331
United Chemi–Con PXA, Poly-Alum (SMD) FS, Os-con (Radial) LXZ, Aluminum (Radial) MVZ, Alumimun (SMD) PSA, Poly-Aluminum (Radial)	 10 V 10 V 16 V 25 V 6.3 V	 330 µF 330 µF 330 µF 330 µF 390 µF	 0.024 Ω 0.025 Ω 0.120 Ω 0.170 Ω 0.008 Ω	 3770 mA 3500 mA 555 mA 450 mA 5080 mA	 10×7.7 10×10.5 8×12 8×10 8×11.5	 1 1 1 1 1	 ≤3 ≤3 ≤3 ≤3 ≤1	 PXA10VC331MJ80TP 10FS330M LXZ16VB331M8x12LL MVZ25VC331MH10TP PSA6.3VB390MH11
NichiconAluminum WG (SMD) PM (Radial)	 16 V 10 V	 330 µF 330 µF	 0.150 Ω 0.160 Ω	 670 mA 460 mA	 10×10 8×11.5	 1 1	 ≤3 ≤3	 UWG1C331MNR1GS UPM1A331MHH
Sanyo SVP, Os-con (SMD) SP, Os-con (Radial) TPE, Poscap Polymer (SMD)	 10 V 10 V 6.3 V	 330 µF 470 µF 330 µF	 0.025 Ω 0.015 Ω 0.025 Ω	 3700 mA 4500 mA 2400 mA	 10×8 10×10.5 7.3L×4.3W	 1 1 1	 ≤3 ≤2 ≤3	 10SVP330MX 10SP470M 6TPE330ML
AVX Tantalum TPS (SMD)	10 V 10 V	330 µF 330 µF	0.100 Ω 0.060 Ω	1100 mA 2000 mA	7.3L ×4.3W ×4.1H	1 1	≤3 ≤3	TPSV337M010R0100 TPSV337M010R0060
Kemet (SMD) T520, Poly Aluminum T530, Organic Poly- Alum.	10 V 10 V	330 µF 330 µF	0.040 Ω 0.015 Ω	1200 mA 1100 mA	7.3L ×5.7W ×4.0H	1 1	≤3 ≤2	T520X337M010AS T530X337M010AS
Vishay-Sprague 594D, Tantalum (SMD) 595D, Tantalum (SMD) 94SVP, Os-con (SMD) 94SA, Os-con (Radial)	10 V 10 V 6.3 V 6,3 V	330 µF 330 µF 330 µF 330 µF	0.045 Ω 0.140 Ω 0.025 Ω 0.025 Ω	1400 mA 1000 mA 3300 mA 3500 mA	7.3L ×6.0W ×4.1H 10×8 10×10.5	1 1 1 1	≤3 ≤3 ≤3 ≤3	594D337X0010R2T 595D337X0010D2T 94SVP337X06R3F8 94SA337X06R3FBP
Kemet, Ceramic X5R (SMD)	16 V 6.3 V	10 47	0.002 Ω 0.002 Ω	—	1210 case 3225 mm	1 [2] 1 [2]	≤3 ≤2	C1210C106M4PAC C1210C476K9PAC
Murata, Ceramic X5R (SMD)	6.3 V 6.3 V 16 V 16 V	100 47 22 10	0.002 Ω	—	1210 case 3225 mm	3 [1] 1 [2] 1 [2] 1 [2]	≤1 ≤2 ≤3 ≤3	GRM32ER60J107M GRM32ER60J476M GRM32ER61C226K GRM32DR61C106K
TDK, Ceramic X5R (SMD)	6.3 V 6.3 V 16 V 16 V	100 47 22 10	0.002 Ω	—	1210 case 3225 mm	3 [1] 1 [2] 1 [2] 1 [2]	≤1 ≤2 ≤3 ≤3	C3225X5R0J107MT C3225X5R0J476MT C3225X5R1C226MT C3225X5R1C106MT

[1] A total capacitance of 300 µF is acceptable based on the surge current capability of ceramic capacitors.
[2] A ceramic capacitor may be used to complement electrolytic types at the input to further reduce high-frequency ripple current.

Adjusting the Output Voltage of the PTH05000W Wide-Output Adjust Power Modules

The *V_o Adjust* control (pin 4) sets the output voltage of the PTH05000Wproduct. The adjustment range is from 0.9 V to 3.6 V. The adjustment method requires the addition of a single external resistor, R_{set}, that must be connected directly between the *V_o Adjust* and *GND* pins [1]. Table 2-1 gives the preferred value of the external resistor for a number of standard voltages, along with the actual output voltage that this resistance value provides.

For other output voltages the value of the required resistor can either be calculated using the following formula, or simply selected from the range of values given in Table 2-2. Figure 2-1 shows the placement of the required resistor.

$$R_{set} = 10\,k\Omega \cdot \frac{0.891\ V}{V_{out} - 0.9\ V} - 3.24\,k\Omega$$

Table 2-1; Preferred Values of R_{set} for Standard Output Voltages

V_{out} (Standard)	R_{set} (Pref'd Value)	V_{out} (Actual)
3.3 V	475 Ω	3.298V
2.5 V	2.32 kΩ	2.502 V
2 V	4.87 kΩ	1.999 V
1.8 V	6.65 kΩ	1.801 V
1.5 V	11.5 kΩ	1.504 V
1.2 V	26.1 kΩ	1.204 V
1 V	84.5 kΩ	1.001 V
0.9 V	Open	0.9 V

Figure 2-1; V_o Adjust Resistor Placement

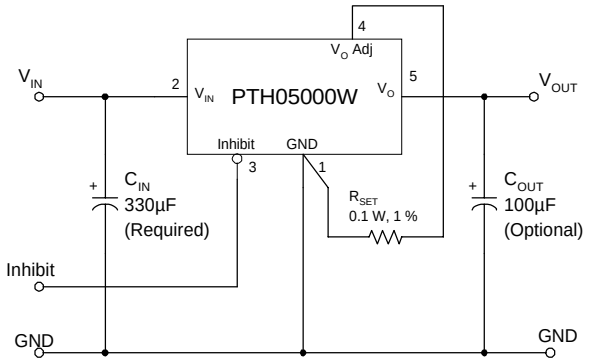


Table 2-2; Output Voltage Set-Point Resistor Values

V_a Req'd	R_{set}	V_a Req'd	R_{set}
0.900	Open	2.00	4.86 kΩ
0.925	353 kΩ	2.05	4.51 kΩ
0.950	175 kΩ	2.10	4.19 kΩ
0.975	116 kΩ	2.15	3.89 kΩ
1.000	85.9 kΩ	2.20	3.61 kΩ
1.025	68.0 kΩ	2.25	3.36 kΩ
1.050	56.2 kΩ	2.30	3.12 kΩ
1.075	47.7 kΩ	2.35	2.90 kΩ
1.100	41.3 kΩ	2.40	2.70 kΩ
1.125	36.4 kΩ	2.45	2.51 kΩ
1.150	32.4 kΩ	2.50	2.33 kΩ
1.175	29.2 kΩ	2.55	2.16 kΩ
1.200	26.5 kΩ	2.60	2.00 kΩ
1.225	24.2 kΩ	2.65	1.85 kΩ
1.250	22.2 kΩ	2.70	1.71 kΩ
1.275	20.5 kΩ	2.75	1.58 kΩ
1.300	19.0 kΩ	2.80	1.45 kΩ
1.325	17.7 kΩ	2.85	1.33 kΩ
1.350	16.6 kΩ	2.90	1.22 kΩ
1.375	15.5 kΩ	2.95	1.11 kΩ
1.400	14.6 kΩ	3.00	1.00 kΩ
1.425	13.7 kΩ	3.05	904 Ω
1.450	13.0 kΩ	3.10	810 Ω
1.475	12.3 kΩ	3.15	720 Ω
1.50	11.6 kΩ	3.20	634 Ω
1.55	10.5 kΩ	3.25	551 Ω
1.60	9.49 kΩ	3.30	473 Ω
1.65	8.64 kΩ	3.35	397 Ω
1.70	7.90 kΩ	3.40	324 Ω
1.75	7.24 kΩ	3.45	254 Ω
1.80	6.66 kΩ	3.50	187 Ω
1.85	6.14 kΩ	3.55	122 Ω
1.90	5.67 kΩ	3.60	60 Ω
1.95	5.25 kΩ		

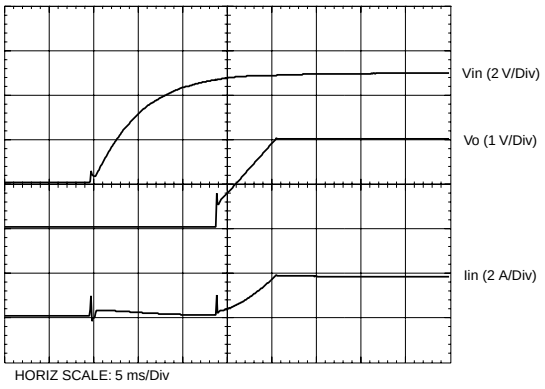
Notes:

1. Use a 0.1 W resistor. The tolerance should be 1 %, with a temperature stability of 100 ppm/°C (or better). Place the resistor as close to the regulator as possible. Connect the resistor directly between pins 4 and 1 using dedicated PCB traces.
2. Never connect capacitors from *V_o Adjust* to either *GND* or *V_{out}*. Any capacitance added to the *V_o Adjust* pin will affect the stability of the regulator.

Power-Up Characteristics

When configured per their standard application, the PTH03000 and PTH05000 series of power modules will produce a regulated output voltage following the application of a valid input source voltage. During power up, internal soft-start circuitry slows the rate that the output voltage rises, thereby limiting the amount of in-rush current that can be drawn from the input source. The soft-start circuitry introduces a short time delay (typically 10 ms) into the power-up characteristic. This is from the point that a valid input source is recognized. Figure 3-1 shows the power-up waveforms for a PTH05000W (5-V input), with the output voltage set point adjusted for a 2-V output. The waveforms were measured with a 5-A resistive load. The initial rise in input current when the input voltage first starts to rise is the charge current drawn by the input capacitors.

Figure 3-1



Current Limit Protection

The PTHxx000W modules protect against load faults with a continuous current limit characteristic. Under a load fault condition the output current cannot exceed the current limit value. Attempting to draw current that exceeds the current limit value causes the output voltage to be progressively reduced. Current is continuously supplied to the fault until it is removed. Upon removal of the fault, the output voltage will promptly recover.

Thermal Shutdown

Thermal shutdown protects the module's internal circuitry against excessively high temperatures. A rise in temperature may be the result of a drop in airflow, a high ambient temperature, or a sustained current limit condition. If the junction temperature of the internal components exceed 150 °C, the module will shutdown. This reduces the output voltage to zero. The module will start up automatically, by initiating a soft-start power up when the sensed temperature decreases 10 °C below the thermal shutdown trip point.

Output On/Off Inhibit

For applications requiring output voltage on/off control, the PTH03000W & PTH05000W power modules incorporate an output on/off *Inhibit* control (pin 3). The inhibit feature can be used wherever there is a requirement for the output voltage from the regulator to be turned off.

The power module functions normally when the *Inhibit* pin is left open-circuit, providing a regulated output whenever a valid source voltage is connected to V_{in} with respect to *GND*.

Figure 3-2 shows the typical application of the inhibit function. Note the discrete transistor (Q_1). The *Inhibit* control has its own internal pull-up to V_{in} potential. An open-collector or open-drain device is recommended to control this input.

Turning Q_1 on applies a low voltage to the *Inhibit* control pin and disables the output of the module. If Q_1 is then turned off, the module will execute a soft-start power-up sequence. A regulated output voltage is produced within 20 msec. Figure 3-3 shows the typical rise in the output voltage, following the turn-off of Q_1. The turn off of Q_1 corresponds to the fall in the waveform, Q_1 V_{gs}. The waveforms were measured with a 5-A resistive load.

Figure 3-2

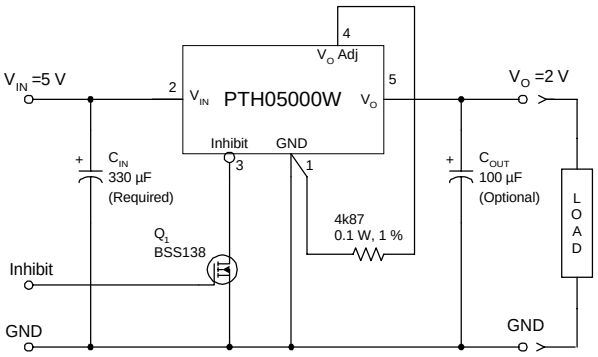
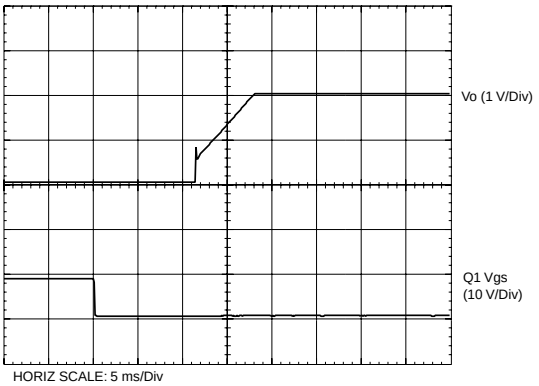


Figure 3-3



PTH05000W Startup with Output Pre-Bias

A pre-bias startup condition occurs as a result of an external voltage being present at the output of the power module prior to its output voltage rising. This often occurs in complex digital systems when current from another power source is backfed through a dual-supply logic component such as an FPGA or ASIC. Another path might be via clamp diodes (to a higher supply voltage) as part of a sequential power-up arrangement.

An output prebias can cause problems with power modules that incorporate synchronous rectifiers. This is because under most operating conditions, they can sink as well as source ouput current. Although the PTH05000W (5-V input) power module can sink current under normal operation, it will not do so during startup. [1] This is true as long as certain conditions are maintained. [2] Figure 3-1 shows an application schematic that demonstrates this capability. Figure 3-2 shows the waveforms of the circuit after input power is applied. Note that the module's output current (I_o) is never negative. Only positive current is sourced. This occurs when the output voltage is raised above that which is backfed from the 5-V input supply, via the diodes D_1 through D_4. [3]

Notes

1. Start up includes both the application of a valid input source voltage, or the removal of a ground signal from the *Inhibit** control (pin 3) with a valid input source applied. The output of the regulator is effectively off (tri-state), during the period that the *Inhibit** control is held low.
2. To ensure that the regulator does not sink current, the input voltage must always be greater or equal to the output voltage throughout the power-up and power-down sequence.
3. If during power up, the backfeeding source is greater than the module's set-point voltage, the module's output voltage will remain higher than its set point. The output will remain out of regulation until the back-feeding source is either reduced in voltage or removed.

Figure 3-2; Start-up with Output Pre-Bias

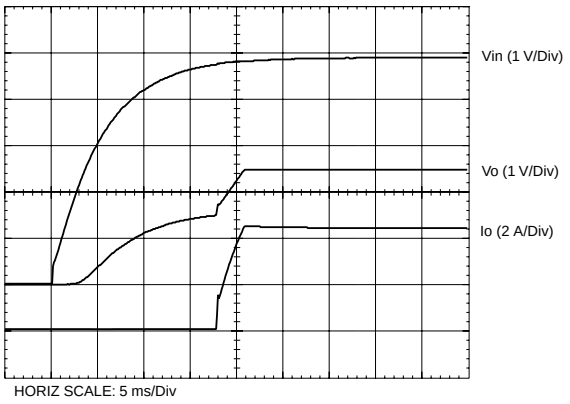
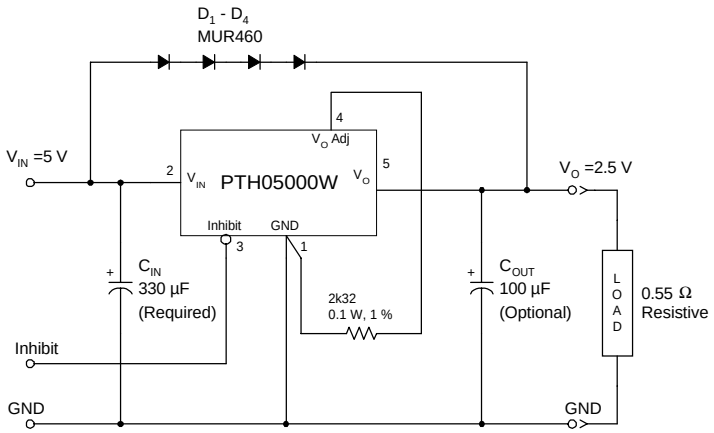


Figure 3-1; Schematic Demonstrating Startup with Output Pre-Bias



PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/ Ball Finish	MSL Peak Temp (3)	Samples (Requires Login)
PTH05000WAD	ACTIVE	Through-Hole Module	EUS	5	56	Pb-Free (RoHS)	SN	N / A for Pkg Type	
PTH05000WAH	ACTIVE	Through-Hole Module	EUS	5	56	Pb-Free (RoHS)	SN	N / A for Pkg Type	
PTH05000WAS	ACTIVE	Surface Mount Module	EUT	5	49	TBD	SNPB	Level-1-235C-UNLIM/ Level-3-260C-168HRS	
PTH05000WAST	ACTIVE	Surface Mount Module	EUT	5	250	TBD	SNPB	Level-1-235C-UNLIM/ Level-3-260C-168HRS	
PTH05000WAZ	ACTIVE	Surface Mount Module	EUT	5	49	Pb-Free (RoHS)	SNAGCU	Level-3-260C-168 HR	
PTH05000WAZT	ACTIVE	Surface Mount Module	EUT	5	250	Pb-Free (RoHS)	SNAGCU	Level-3-260C-168 HR	

(1) The marketing status values are defined as follows:
ACTIVE: Product device recommended for new designs.
LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.
NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.
PREVIEW: Device has been announced but is not in production. Samples may or may not be available.
OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.
TBD: The Pb-Free/Green conversion plan has not been defined.
Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.
Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.
Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

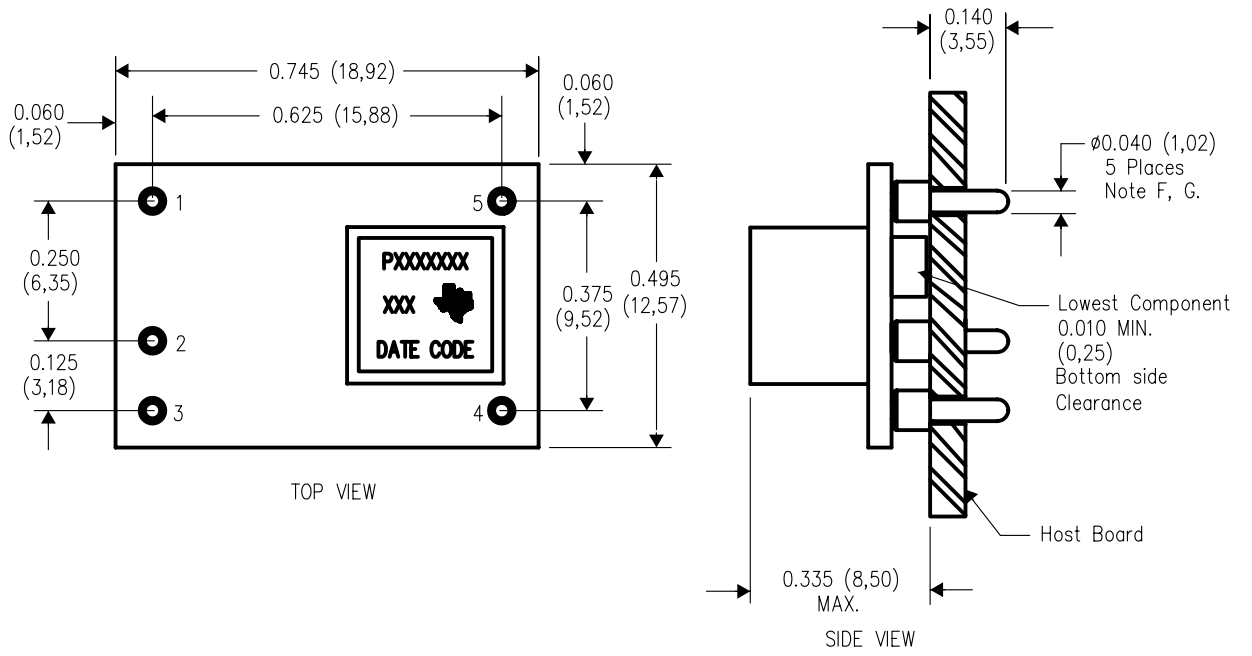
(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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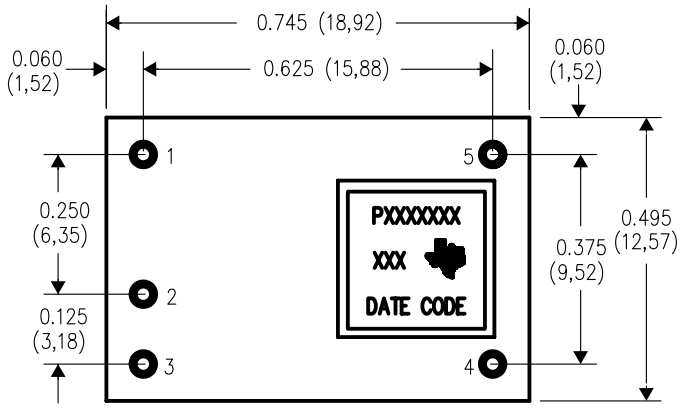
EUS (R–PDSS–T5)

DOUBLE SIDED MODULE

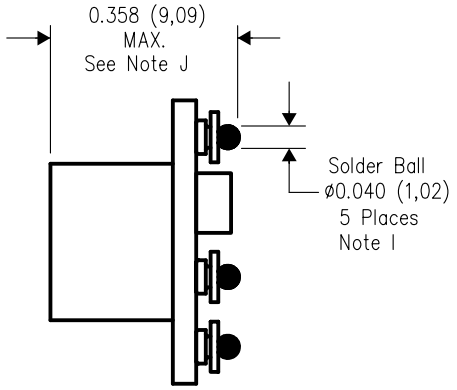


NOTES:
A. All linear dimensions are in inches (mm).
B. This drawing is subject to change without notice.
C. 2 place decimals are ±0.030 (±0,76mm).
D. 3 place decimals are ±0.010 (±0,25mm).
E. Recommended keep out area for user components.
F. Pins are 0.040" (1,02) diameter with 0.070" (1,78) diameter standoff shoulder.
G. All pins: Material – Copper Alloy
Finish – Tin (100%) over Nickel plate

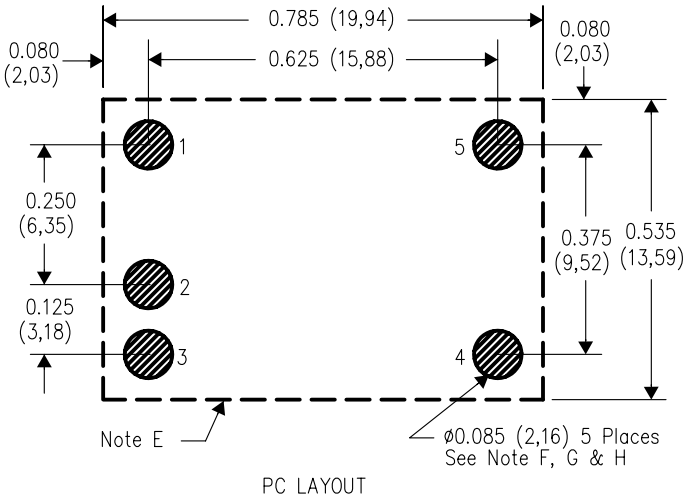
EUT (R–PDSS–B5) DOUBLE SIDED MODULE



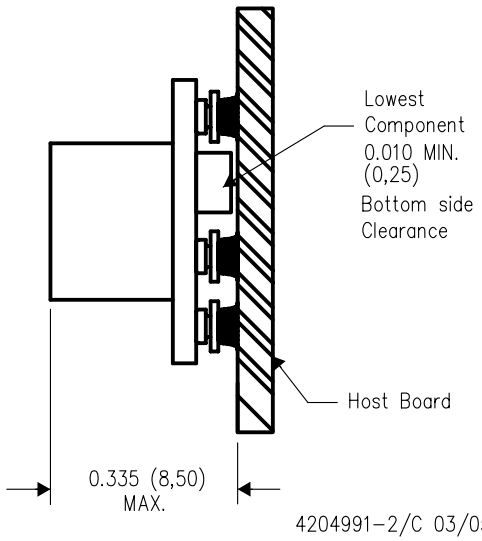
TOP VIEW



SIDE VIEW



PC LAYOUT



4204991–2/C 03/05

NOTES:
A. All linear dimensions are in inches (mm).
B. This drawing is subject to change without notice.
C. 2 place decimals are ±0.030 (±0,76mm).
D. 3 place decimals are ±0.010 (±0,25mm).
E. Recommended keep out area for user components.
F. Power pin connection should utilize two or more vias to the interior power plane of 0.025 (0,63) I.D. per input, ground and output pin (or the electrical equivalent).
G. Paste screen opening: 0.080 (2,03) to 0.085 (2,16). Paste screen thickness: 0.006 (0,15).
H. Pad type: Solder mask defined.
I. All pins: Material – Copper Alloy
Finish – Tin (100%) over Nickel plate
Solder Ball – See product data sheet.
J. Dimension prior to reflow solder.

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