

300mA, Low Dropout, Low Noise Ultra-Fast Without Bypass Capacitor CMOS LDO Regulator

General Description

The RT9038 is a high-performance, 300mA LDO regulator, offering extremely high PSRR and ultra-low dropout. The RT9038 is Ideal for portable RF and wireless applications with demanding performance and space requirements.

The RT9038 provides quiescent current as low to as 25 μ A, further prolonging the battery life. The RT9038 also works with low-ESR ceramic capacitors, reducing the amount of board space necessary for power applications, which is critical in hand-held wireless devices.

The RT9038 consumes typical 0.7 μ A in shutdown mode and has fast turn-on time to be less than 40 μ s. The other features include ultra-low dropout voltage, high output accuracy, current limiting protection, and high ripple rejection ratio. The RT9038 is available in the WDFN-8L 2x2 package.

Ordering Information

RT9038-□□□□	
Package Type	QW : WDFN-8L 2x2 (W-Type)
Lead Plating System	G : Green (Halogen Free and Pb Free)
Fixed Output Voltage	12 : 1.2V
	13 : 1.3V
	:
	34 : 3.4V
	35 : 3.5V
	1B : 1.25V
	1H : 1.85V
	2H : 2.85V

Note :

Richtek products are :

- ▶ RoHS compliant and compatible with the current requirements of IPC/JEDEC J-STD-020.
- ▶ Suitable for use in SnPb or Pb-free soldering processes.

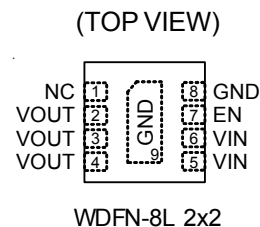
Features

- Wide Operating Voltage Range : 2.2V to 5.5V
- Low Dropout : 125mV at 300mA
- Ultra-Low-Noise for RF Application
- Ultra-Fast in Line/Load Transient Response
- Current Limiting Protection
- Thermal Shutdown Protection
- High Power Supply Rejection Ratio
- Only 1 μ F Output Capacitor Required for Stability
- TTL-Logic-Controlled Shutdown Input
- RoHS Compliant and Halogen Free

Applications

- CDMA/GSM Cellular Handsets
- Portable Information Appliances
- Laptop, Palmtops, Notebook Computers
- Hand-Held Instruments
- Mini PCI & PCI-Express Cards
- PCMCIA & New Cards

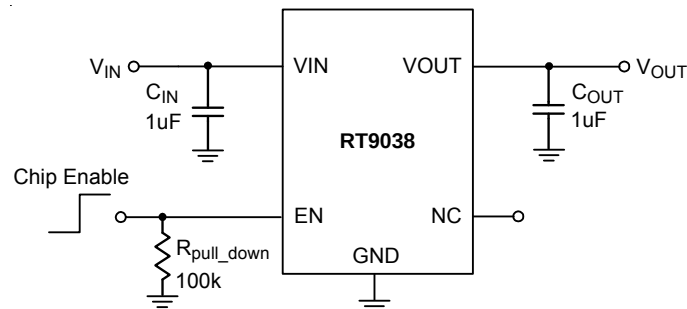
Pin Configurations



Marking Information

For marking information, contact our sales representative directly or through a Richtek distributor located in your area.

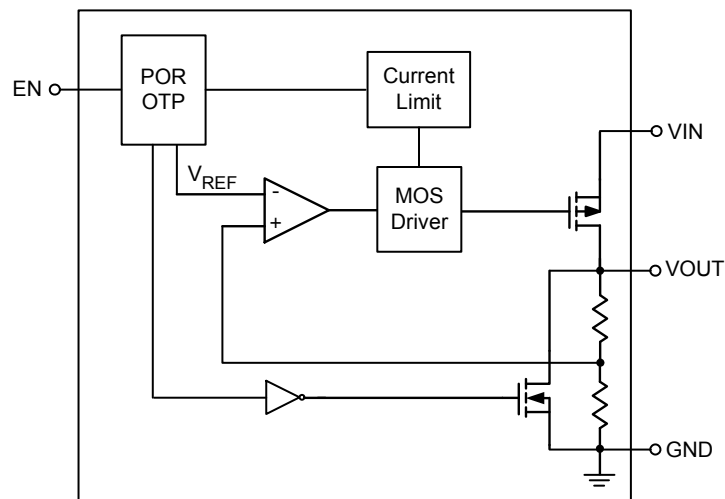
Typical Application Circuit



Functional Pin Description

Pin No.	Pin Name	Pin Function
1	NC	No Internal Connection.
2, 3, 4	VOUT	Regulator Output.
5, 6	VIN	Supply Input.
7	EN	Chip Enable (Active High). It is recommended to add a 100kΩ resistor between the EN and GND.
8, 9 (Exposed Pad)	GND	Common Ground. The exposed pad must be soldered to a large PCB and connected to GND for maximum power dissipation.

Function Block Diagram



Absolute Maximum Ratings (Note 1)

• Supply Input Voltage	6V
• EN Input Voltage	6V
• Power Dissipation, P_D @ $T_A = 25^\circ\text{C}$	
WDFN-8L 2x2	0.606W
• Package Thermal Resistance (Note 2)	
WDFN-8L 2x2	165°C/W
• Lead Temperature (Soldering, 10 sec.)	260°C
• Junction Temperature	150°C
• Storage Temperature Range	-65°C to 150°C
• ESD Susceptibility (Note 3)	
HBM	2kV
MM	200V

Recommended Operating Conditions (Note 4)

• Supply Input Voltage	2.2V to 5.5V
• Junction Temperature Range	-40°C to 125°C
• Ambient Temperature Range	-40°C to 85°C

Electrical Characteristics

($V_{IN} = V_{OUT} + 0.5\text{V}$, $V_{EN} = V_{IN}$, $C_{IN} = C_{OUT} = 1\mu\text{F}$ (Ceramic), $T_A = 25^\circ\text{C}$, unless otherwise specified)

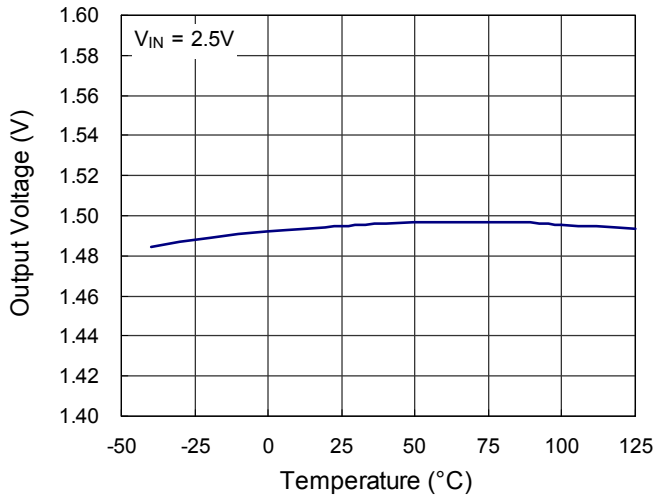
Parameter	Symbol	Test Conditions	Min	Typ	Max	Unit
Input Voltage Range	V_{IN}		2.2	--	5.5	V
Output Noise Voltage	V_{ON}	$V_{OUT} = 1.5\text{V}$, $C_{OUT} = 1\mu\text{F}$, $I_{OUT} = 0\text{mA}$	--	30	--	μV_{RMS}
Output Voltage Accuracy (Fixed Output Voltage)	ΔV_{OUT}	$I_{OUT} = 10\text{mA}$	-2	0	+2	%
Quiescent Current (Note 5)	I_Q	$V_{EN} = 5\text{V}$, $I_{OUT} = 0\text{mA}$	--	25	50	μA
Standby Current	I_{STBY}	$V_{EN} = 0\text{V}$	--	0.7	1.5	μA
Current Limit	I_{LIM}	$R_{\text{LOAD}} = 0\Omega$, $2.2\text{V} \leq V_{IN} < 5.5\text{V}$	300	400	600	mA
Dropout Voltage (Note 6)	V_{DROP}	$I_{OUT} = 300\text{mA}$	--	125	240	mV
Load Regulation (Note 7) (Fixed Output Voltage)	ΔV_{LOAD}	$1\text{mA} < I_{OUT} < 300\text{mA}$ $2.2\text{V} \leq V_{IN} \leq 5.5\text{V}$	--	--	1	%
EN Threshold	Logic-Low Voltage	V_{IL}	0	--	0.6	V
	Logic-High Voltage	V_{IH}	1.6	--	5.5	
Enable Pin Current	I_{EN}		--	0.1	1	μA
Power Supply Rejection Rate	PSRR	$I_{OUT} = 100\text{mA}$, $f = 100\text{kHz}$	--	40	--	dB
Line Regulation	ΔV_{LINE}	$V_{IN} = (V_{OUT} + 0.5)$ to 5.5V , $I_{OUT} = 1\text{mA}$	--	0.01	0.2	%/V
Thermal Shutdown Temperature	T_{SD}		--	170	--	$^\circ\text{C}$
Thermal Shutdown Hysteresis	ΔT_{SD}		--	30	--	

- Note 1.** Stresses beyond those listed “Absolute Maximum Ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions may affect device reliability.
- Note 2.** θ_{JA} is measured at $T_A = 25^{\circ}\text{C}$ on a low effective thermal conductivity single-layer test board per JEDEC 51-3.
- Note 3.** Devices are ESD sensitive. Handling precaution is recommended.
- Note 4.** The device is not guaranteed to function outside its operating conditions.
- Note 5.** Quiescent, or ground current, is the difference between input and output currents. It is defined by $I_Q = I_{IN} - I_{OUT}$ under no load condition ($I_{OUT} = 0\text{mA}$). The total current drawn from the supply is the sum of the load current plus the ground pin current.
- Note 6.** The dropout voltage is defined as $V_{IN} - V_{OUT}$, which is measured when V_{OUT} is $V_{OUT(NORMAL)} - 100\text{mV}$.
- Note 7.** Regulation is measured at constant junction temperature by using a 2ms current pulse. Devices are tested for load regulation in the load range from 10mA to 300mA.

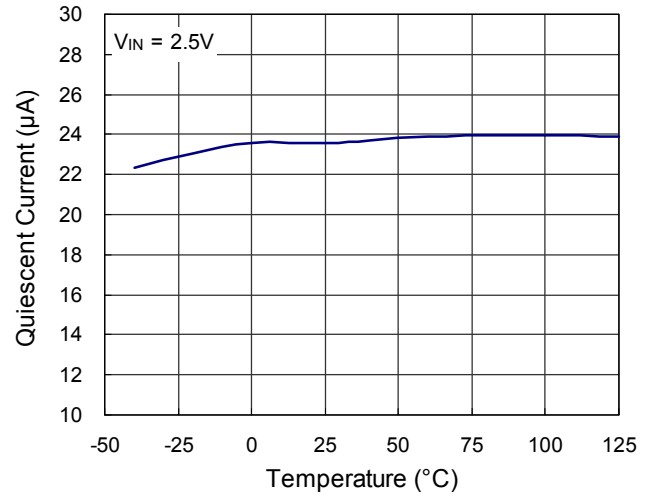
Typical Operating Characteristics

($C_{IN} = C_{OUT} = 1\mu/X7R$, unless otherwise specified)

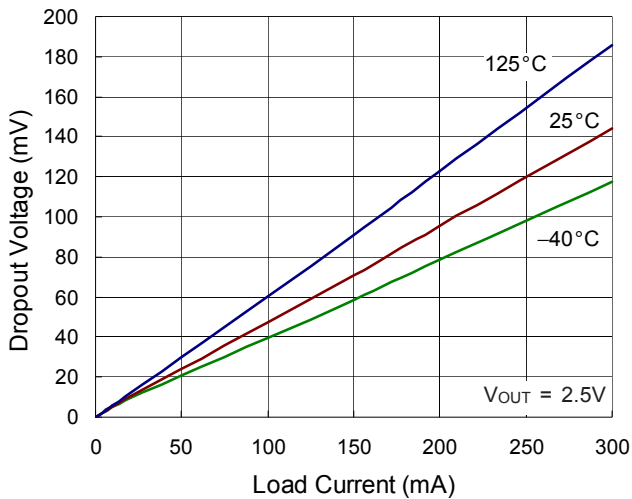
Output Voltage vs. Temperature



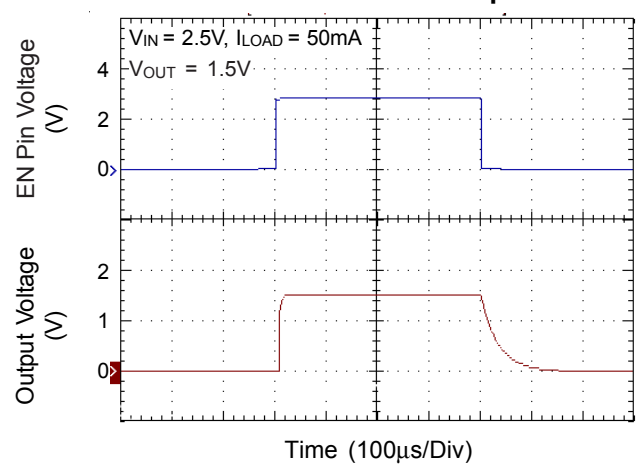
Quiescent Current vs. Temperature



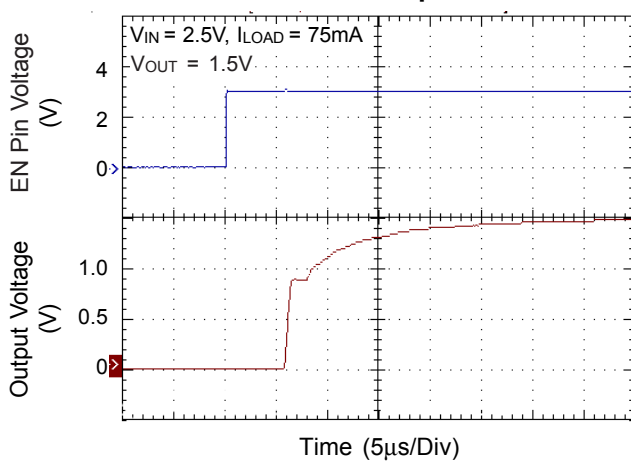
Dropout Voltage vs. Load Current



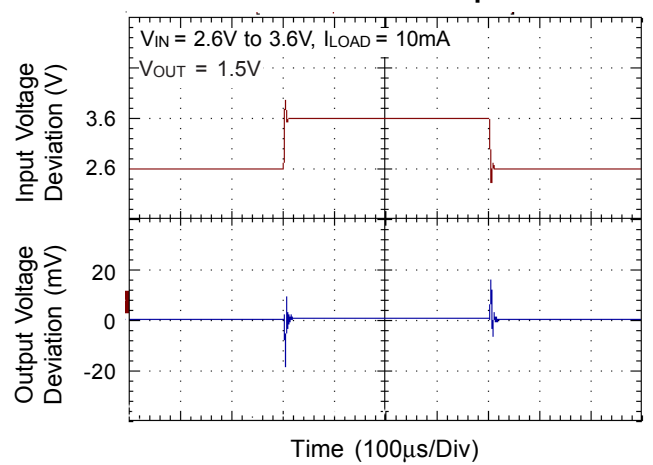
EN Pin Shutdown Response



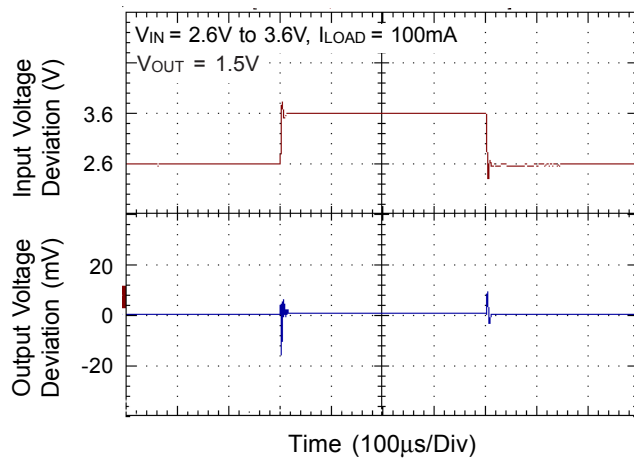
Start Up



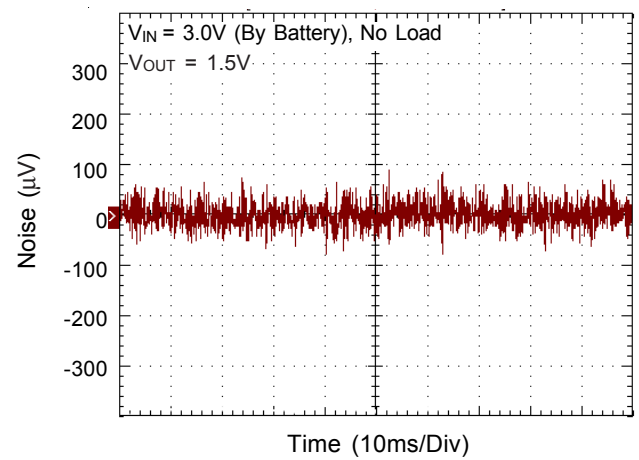
Line Transient Response



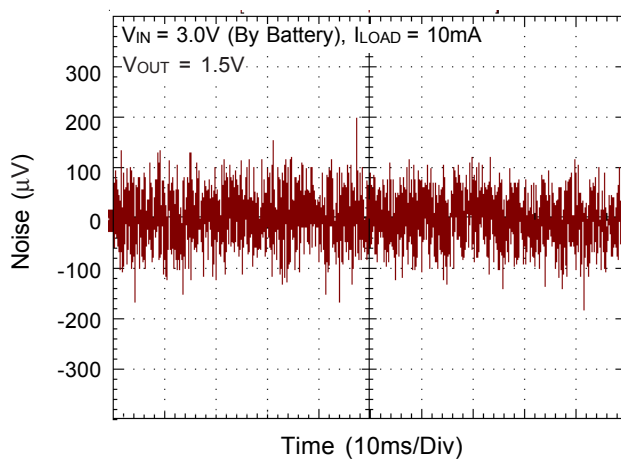
Line Transient Response



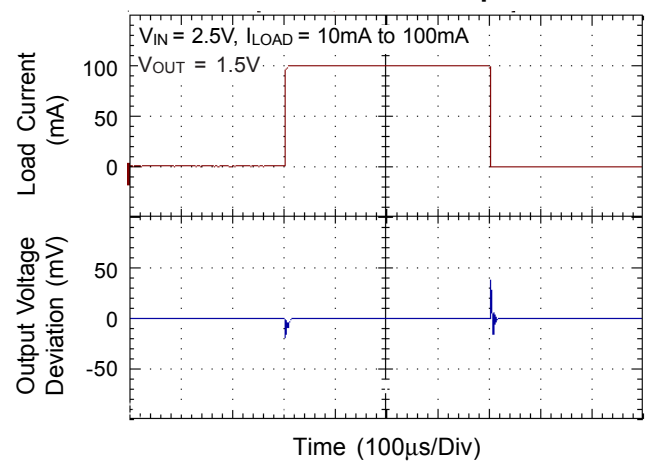
Noise



Noise



Load Transient Response



Applications Information

Like any low-dropout regulator, the external capacitors used with the RT9038 must be carefully selected for regulator stability and performance. Using a capacitor whose value is $> 1\mu\text{F}$ on the RT9038 input and the amount of capacitance can be increased without limit. The input capacitor must be located at a distance of not more than 0.5 inch from the input pin of the IC and returned to a clean analog ground. Any good quality ceramic can be used for this capacitor. The capacitor with larger value and lower ESR (equivalent series resistance) provides better PSRR and line-transient response.

The output capacitor must meet both requirements for minimum amount of capacitance and ESR in all LDOs application. The RT9038 is designed specifically to work with low ESR ceramic output capacitor in space-saving and performance consideration. Using a ceramic capacitor whose value is at least $1\mu\text{F}$ with ESR is $> 20\text{m}\Omega$ on the RT9038 output ensures stability. The RT9038 still works well with output capacitor of other types due to the wide stable ESR range. Figure 1. shows the curves of allowable ESR range as a function of load current for various output capacitor values. Output capacitor of larger capacitance can reduce noise and improve load transient response, stability, and PSRR. The output capacitor should be located not more than 0.5 inch from the VOUT pin of the RT9038 and returned to a clean analog ground.

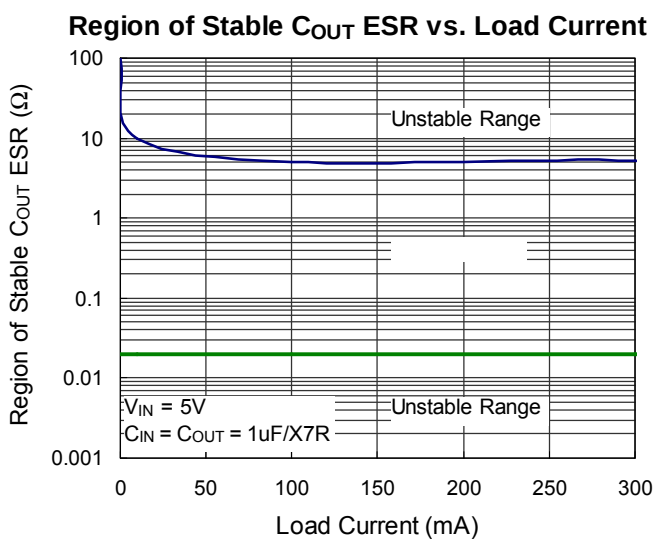


Figure 1

Enable

The RT9038 goes into sleep mode when the EN pin is in a logic low condition. During this condition, the RT9038 has an EN pin to turn on or turn off the regulator. When the EN pin is logic high the regulator will be turned on. The supply current is 0.7μA typical. The EN pin may be directly tied to V_{IN} to keep the part on. The Enable input is CMOS logic and cannot be left floating.

PSRR

The power supply rejection ratio (PSRR) is defined as the gain from the input to output divided by the gain from the supply to the output. The PSRR is found to be

$$\text{PSRR} = 20 \times \log \left(\frac{\Delta \text{Gain Error}}{\Delta \text{Supply}} \right)$$

Note that in heavy load measuring, Δsupply will cause $\Delta \text{temperature}$. And $\Delta \text{temperature}$ will cause Δoutput voltage. So the temperature effect is included in heavy load PSRR measuring.

Current limit

The RT9038 contains an independent current limiter, which monitors and controls the pass transistor's gate voltage, limiting the output current to 0.4A (typ.). The output can be shorted to ground indefinitely without damaging the part.

Thermal Considerations

Thermal protection limits power dissipation in the RT9038. When the operation junction temperature exceeds 170°C, the OTP circuit starts the thermal shutdown function and turns the pass element off. The pass element will be turned on again after the junction temperature cools by 30°C.

For continuous operation, do not exceed absolute maximum operation junction temperature 125°C. The power dissipation definition in the device is calculated as follows :

$$P_D = (V_{IN} - V_{OUT}) \times I_{OUT} + V_{IN} \times I_Q$$

The maximum power dissipation depends on the thermal resistance of IC package, PCB layout, the rate of surrounding airflow and temperature difference between

junction to ambient. The maximum power dissipation can be calculated by following formula :

$$P_{D(MAX)} = (T_{J(MAX)} - T_A) / \theta_{JA}$$

Where $T_{J(MAX)}$ is the maximum operation junction temperature, T_A is the ambient temperature and the θ_{JA} is the junction to ambient thermal resistance.

For recommended operating conditions specification, where $T_{J(MAX)}$ is the maximum junction temperature of the die (125°C) and T_A is the operated ambient temperature. The junction to ambient thermal resistance for WDFN-8L 2x2 package is 165°C/W on the standard JEDEC 51-3 single-layer thermal test board. The maximum power dissipation at $T_A = 25^\circ\text{C}$ can be calculated by following formula :

$$P_{D(MAX)} = (125^\circ\text{C} - 25^\circ\text{C}) / (165^\circ\text{C/W}) = 0.606\text{W for WDFN-8L 2x2 packages}$$

The maximum power dissipation depends on operating ambient temperature for fixed $T_{J(MAX)}$ and thermal resistance θ_{JA} . The Figure 2 of derating curve allow the designer to see the effect of rising ambient temperature on the maximum power allowed.

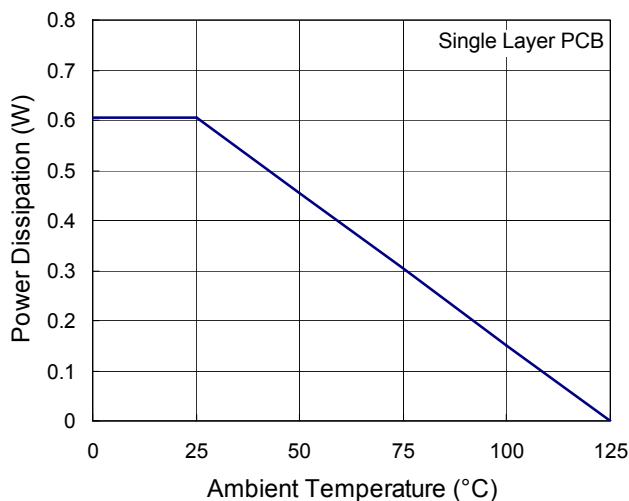
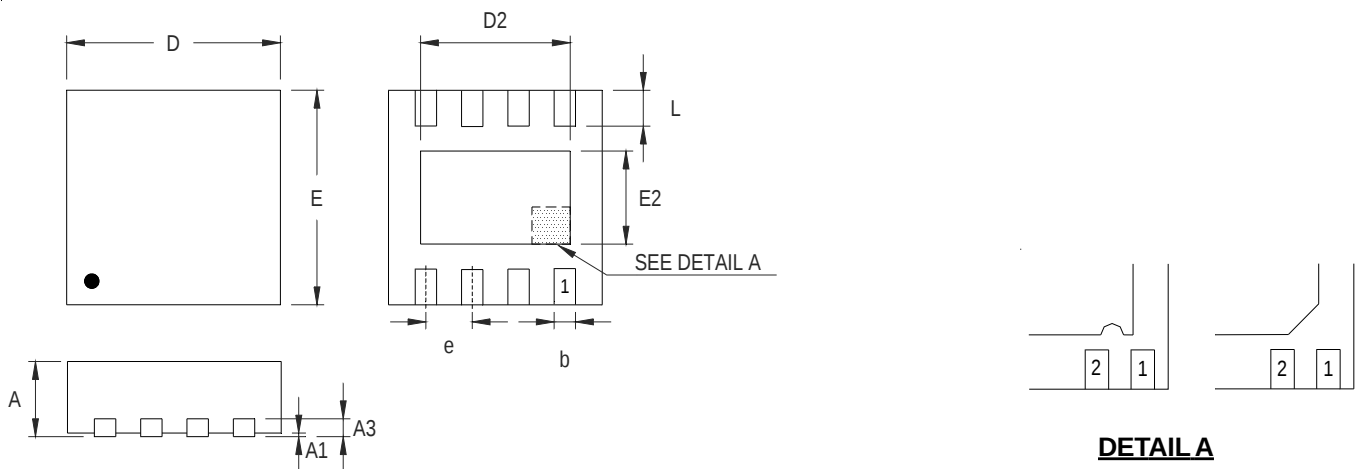


Figure 2. Derating Curve of Maximum Power Dissipation

Outline Dimension



DETAIL A

Pin #1 ID and Tie Bar Mark Options

Note : The configuration of the Pin #1 identifier is optional, but must be located within the zone indicated.

Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
A	0.700	0.800	0.028	0.031
A1	0.000	0.050	0.000	0.002
A3	0.175	0.250	0.007	0.010
b	0.200	0.300	0.008	0.012
D	1.950	2.050	0.077	0.081
D2	1.000	1.250	0.039	0.049
E	1.950	2.050	0.077	0.081
E2	0.400	0.650	0.016	0.026
e	0.500		0.020	
L	0.300	0.400	0.012	0.016

W-Type 8L DFN 2x2 Package

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