# MAX30004

# Ultra-Low Power, Single-Channel Integrated Biopotential (R-to-R Detection) AFE

## General Description

The MAX30004 is a complete, biopotential, analog front-end solution for wearable applications. It offers high performance for clinical and fitness applications, with ultra-low power for long battery life. The MAX30004 is a single biopotential channel providing heart rate detection.

The biopotential channel has ESD protection, EMI filtering, internal lead biasing, DC leads-off detection, ultra-low power leads-on detection during standby mode. Soft power-up sequencing ensures no large transients are injected into the electrodes. The biopotential channel also has high input impedance, low noise, high CMRR, programmable gain, various low-pass and high-pass filter options, and a high resolution analog-to-digital converter. The biopotential channel is DC coupled, can handle large electrode voltage offsets, and has a fast recovery mode to quickly recover from overdrive conditions, such as defibrillation and electrosurgery.

The MAX30004 is available in a 28-pin TQFN and 30-bump wafer-level package (WLP), operating over the 0°C to +70°C commercial temperature range.

## Applications

- Single Lead Wireless Patches for At-Home/ In-Hospital Monitoring
- Chest Band Heart Rate Monitors for Fitness Applications

***Ordering Information** appears at end of data sheet.*

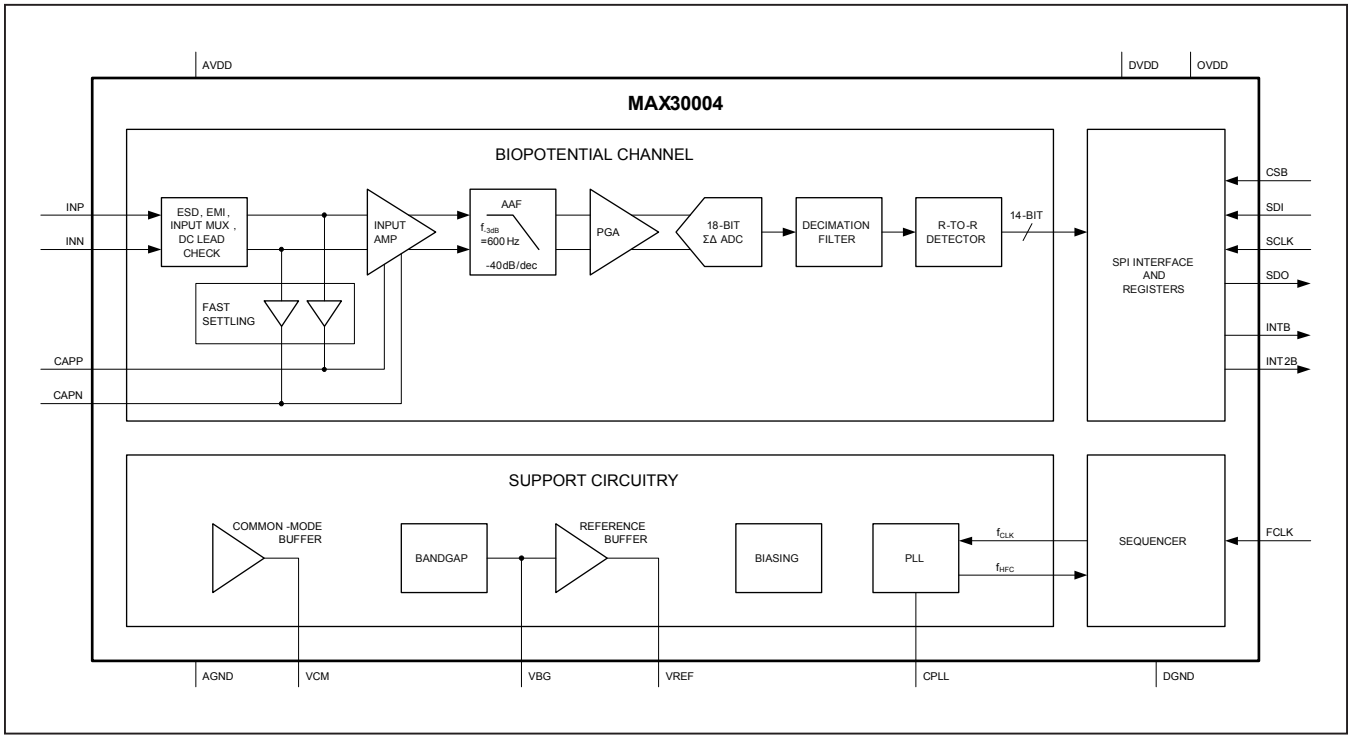
## Benefits and Features

- Heart Rate Detection with Interrupt Feature Eliminates the Need to Extract and Process the ECG Data on the Microcontroller
  - Robust R-R Detection in High Motion Environment at Extremely Low Power
- Clinical-Grade Biopotential AFE with High-Resolution Data Converter
  - 15.5 Bits Effective Resolution with 5µ$V_{P-P}$ Noise
- Better Dry Starts Due to Much Improved Real World CMRR and High Input Impedance
  - Fully Differential Input Structure with CMRR > 100dB
- Offers Better Common-Mode to Differential Mode Conversion Due to High Input Impedance
  - High Input Impedance > 500MΩ for Extremely Low Common-to-Differential Mode Conversion
- Minimum Signal Attenuation at the Input During Dry Start Due to High Electrode Impedance
- High DC Offset Range of ±650mV (1.8V, typ) Allows to Be Used with Wide Variety of Electrodes
- High AC Dynamic Range of 65m$V_{P-P}$ Will Help the AFE Not Saturate in the Presence of Motion/Direct Electrode Hits
- Longer Battery Life Compared to Competing Solutions
  - 85µW at 1.1V Supply Voltage
- Leads-On Interrupt Feature Keeps µC in Deep Sleep Mode with RTC Off Until Valid Lead Condition is Detected
  - Lead-On Detect Current: 0.7µA (typ)
- Configurable Interrupts Allows the µC Wake-Up Only on Every Heart Beat Reducing the Overall System Power
- High-Speed SPI Interface
- Shutdown Current of 0.5µA (typ)

*19-8732; Rev 0; 12/16*

maxim integrated™

## Functional Diagram



## Absolute Maximum Ratings

$AV_{DD}$ to AGND .....-0.3V to +2.0V
$DV_{DD}$ to DGND .....-0.3V to +2.0V
$AV_{DD}$ to $DV_{DD}$.....-0.3V to +0.3V
$OV_{DD}$ to DGND .....-0.3V to +3.6V
AGND to DGND .....-0.3V to +0.3V
CSB, SCLK, SDI, FCLK to DGND .....-0.3V to +3.6V
SDO, INTB, INT2B to
DGND .....-0.3V to the lower of (3.6V and $OV_{DD}$ + 0.3V)
All other pins to
AGND .....-0.3V to the lower of (2.0V and $AV_{DD}$ + 0.3V)
Maximum Current into Any Pin.....±50mA

Continuous Power Dissipation ($T_A$ = +70°C)
28-Pin TQFN (derate 34.5mW/°C above +70°C)...2758.6mW
30-Bump WLP (derate 24.3mW/°C
above +70°C).....1945.5mW
Operating Temperature Range.....0°C to +70°C
Junction Temperature.....+150°C
Storage Temperature Range..... -65°C to +150°C
Lead Temperature (Soldering, 10sec).....+300°C
Soldering Temperature (reflow).....+260°C

## Package Thermal Characteristics (Note 1)

TQFN
Junction-to-Ambient Thermal Resistance ($\theta_{JA}$) .....29°C/W
Junction-to-Case Thermal Resistance ($\theta_{JC}$).....2°C/W

WLP
Junction-to-Ambient Thermal Resistance ($\theta_{JA}$) .....44°C/W

**Note 1:** Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a four-layer board. For detailed information on package thermal considerations, refer to **www.maximintegrated.com/thermal-tutorial**.

*Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.*

## Electrical Characteristics

($V_{DVDD}$ = $V_{AVDD}$ = +1.1V to +2.0V, $V_{OVDD}$ = +1.65V to +3.6V, $f_{FCLK}$ = 32.768kHz, $T_A$ = $T_{MIN}$ to $T_{MAX}$, unless otherwise noted. Typical values are at $V_{DVDD}$ = $V_{AVDD}$ = +1.8V, $V_{OVDD}$ = +2.5V, $T_A$ = +25°C.) (Note 2)

| PARAMETER | SYMBOL | CONDITIONS | MIN | TYP | MAX | UNITS |
|---|---|---|---|---|---|---|
| **BIOPOTENTIAL CHANNEL** | | | | | | |
| AC Differential Input Range | | $V_{AVDD}$ = +1.1V, THD < 0.3% | -15 | | +15 | $mV_{p-p}$ |
| | | $V_{AVDD}$ = +1.8V, THD < 0.3% | | ±32.5 | | |
| DC Differential Input Range | | $V_{AVDD}$ = +1.1V, shift from nominal gain < 2% | -300 | | +300 | mV |
| | | $V_{AVDD}$ = +1.8V | | ±650 | | |
| Common Mode Input Range | | $V_{AVDD}$ = +1.1V, from $V_{MID}$, shift from nominal gain < 2% | -150 | | +150 | mV |
| | | $V_{AVDD}$ = +1.8V, from $V_{MID}$, shift from nominal gain < 2% | | ±550 | | |
| Common Mode Rejection Ratio | CMRR | 0Ω source impedance, f = 64Hz (Note 3) | 105 | 115 | | dB |
| | | (Note 4) | | 77 | | |
| Input Referred Noise | | BW = 0.05 - 150Hz, $G_{CH}$ = 20x | | 0.82 | | $\mu V_{RMS}$ |
| | | | | 5.4 | | $\mu V_{p-p}$ |
| | | BW = 0.05 - 40Hz, $G_{CH}$ = 20x (Note 3) | | 0.53 | 1.0 | $\mu V_{RMS}$ |
| | | | | 3.5 | 6.6 | $\mu V_{p-p}$ |
| Input Leakage Current | | $T_A$ = +25°C | -1 | 0.1 | +1 | nA |
| Input Impedance (INA) | | Common-mode, DC | | 45 | | GΩ |
| | | Differential, DC | | 1500 | | MΩ |

## Electrical Characteristics (continued)

($V_{DVDD}$ = $V_{AVDD}$ = +1.1V to +2.0V, $V_{OVDD}$ = +1.65V to +3.6V, $f_{FCLK}$ = 32.768kHz, $T_A$ = $T_{MIN}$ to $T_{MAX}$, unless otherwise noted. Typical values are at $V_{DVDD}$ = $V_{AVDD}$ = +1.8V, $V_{OVDD}$ = +2.5V, $T_A$ = +25°C.) (Note 2)

| PARAMETER | SYMBOL | CONDITIONS | | MIN | TYP | MAX | UNITS |
|---|---|---|---|---|---|---|---|
| Total Harmonic Distortion | THD | $V_{AVDD}$ = +1.80V, $V_{IN}$ = 65m$V_{p-p}$, $F_{IN}$ = 64Hz, $G_{CH}$ = 20x, electrode offset = ±300mV | | | 0.025 | | % |
| | | $V_{AVDD}$ = +1.1V, $V_{IN}$ = 30m$V_{p-p}$, $F_{IN}$ = 64Hz, $G_{CH}$ = 20x, electrode offset = ±300mV | | | | 0.3 | |
| Gain Setting | $G_{CH}$ | Programmable, see Register Map | | | 20 to 160 | | V/V |
| Gain Error | | $V_{AVDD}$ = +1.8V, $G_{CH}$ = 20x, INP = INN = VMID | | -2.5 | | +2.5 | % |
| | | $V_{AVDD}$ = +1.1V, $G_{CH}$ = 20x, INP = INN = VMID | | -4.5 | | +4.5 | % |
| Offset Error | | (Note 5) | | | 0.1 | | % of FSR |
| ADC Resolution | | | | | 18 | | Bits |
| ADC Sample Rate | | Programmable, see Register Map | | | 125 to 512 | | SPS |
| CAPP to CAPN Impedance | $R_{HPF}$ | FHP = 1/(2R x $R_{HPF}$ x $C_{HPF}$), $C_{HPF}$ = capacitance between CAPP and CAPN | | 320 | 450 | 600 | kΩ |
| Analog High-Pass Filter Slew Current | | Fast recovery enabled (1.8V) | | | 160 | | µA |
| | | Fast recovery enabled (1.1V) | | | 55 | | |
| | | Fast recovery disabled | | | 0.09 | | |
| Fast Settling Recovery Time | | $C_{HPF}$ = 10µF | | | 500 | | ms |
| Digital Low Pass Filter | | Linear phase FIR filter. | DLPF[0:1] = 01 | | 40 | | Hz |
| | | | DLPF[0:1] = 10 | | 100 | | |
| | | | DLPF[0:1] = 11 | | 150 | | |
| Digital High Pass Filter | | Phase-corrected 1st-order IIR filter. DHPF = 1 | | | 0.5 | | Hz |
| Power Supply Rejection | PSRR | Lead bias disabled, DC | | | 107 | | dB |
| | | Lead bias disabled, $f_{SW}$ = 64Hz | | | 110 | | |
| **INPUT MUX** | | | | | | | |
| DC Lead Off Check | | Pullup/ pulldown | DCLOFF_IMAG[2:0] = 001 | | 5 | | nA |
| | | | DCLOFF_IMAG[2:0] = 010 | | 10 | | |
| | | | DCLOFF_IMAG[2:0] = 011 | | 20 | | |
| | | | DCLOFF_IMAG[2:0] = 100 | | 50 | | |
| | | | DCLOFF_IMAG[2:0] = 101 | | 100 | | |

## Electrical Characteristics (continued)

($V_{DVDD}$ = $V_{AVDD}$ = +1.1V to +2.0V, $V_{OVDD}$ = +1.65V to +3.6V, $f_{FCLK}$ = 32.768kHz, $T_A$ = $T_{MIN}$ to $T_{MAX}$, unless otherwise noted. Typical values are at $V_{DVDD}$ = $V_{AVDD}$ = +1.8V, $V_{OVDD}$ = +2.5V, $T_A$ = +25°C.) (Note 2)

| PARAMETER | SYMBOL | CONDITIONS | | MIN | TYP | MAX | UNITS |
|---|---|---|---|---|---|---|---|
| DC Lead Off Comparator Low Threshold | | DCLOFF_VTH[1:0] = 11 (Note 6) | | | $V_{MID}$ - 0.50 | | V |
| | | DCLOFF_VTH[1:0] = 10 (Note 7) | | | $V_{MID}$ - 0.45 | | |
| | | DCLOFF_VTH[1:0] = 01 (Note 8) | | | $V_{MID}$ - 0.40 | | |
| | | DCLOFF_VTH[1:0] = 00 | | | $V_{MID}$ - 0.30 | | |
| DC Lead Off Comparator High Threshold | | DCLOFF_VTH[1:0] = 11 (Note 6) | | | $V_{MID}$ + 0.50 | | V |
| | | DCLOFF_VTH[1:0] = 10 (Note 7) | | | $V_{MID}$ + 0.45 | | |
| | | DCLOFF_VTH[1:0] = 01 (Note 8) | | | $V_{MID}$ + 0.40 | | |
| | | DCLOFF_VTH[1:0] = 00 | | | $V_{MID}$ + 0.30 | | |
| Lead Bias Impedance | | Lead bias enabled | RBIASV[1:0] = 00 | | 50 | | MΩ |
| | | | RBIASV[1:0] = 01 | | 100 | | |
| | | | RBIASV[1:0] = 10 | | 200 | | |
| Lead Bias Voltage | $V_{MID}$ | Lead bias enabled | | | $V_{AVDD}$/ 2.15 | | V |
| **INTERNAL REFERENCE/COMMON-MODE** | | | | | | | |
| $V_{BG}$ Output Voltage | $V_{BG}$ | | | | 0.650 | | V |
| $V_{BG}$ Output Impedance | | | | | 100 | | kΩ |
| External $V_{BG}$ Compensation Capacitor | $C_{VBG}$ | | | 1 | | | µF |
| $V_{REF}$ Output Voltage | $V_{REF}$ | $T_A$ = +25°C | | 0.995 | 1.000 | 1.005 | V |
| $V_{REF}$ Temperature Coefficient | $TC_{REF}$ | $T_A$ = 0°C to +70°C | | | 10 | | ppm/°C |
| $V_{REF}$ Buffer Line Regulation | | | | | 330 | | µV/V |
| $V_{REF}$ Buffer Load Regulation | | $I_{LOAD}$ = 0 to 100µA | | | 25 | | µV/µA |
| External $V_{REF}$ Compensation Capacitor | $C_{REF}$ | | | 1 | 10 | | µF |
| $V_{CM}$ Output Voltage | $V_{CM}$ | | | | 0.650 | | V |
| External $V_{CM}$ Compensation Capacitor | $C_{CM}$ | | | 1 | 10 | | µF |
| **DIGITAL INPUTS (SDI, SCLK, CSB, FCLK)** | | | | | | | |
| Input-Voltage High | $V_{IH}$ | | | 0.7 × $V_{OVDD}$ | | | V |
| Input-Voltage Low | $V_{IL}$ | | | | | 0.3 × $V_{OVDD}$ | V |
| Input Hysteresis | $V_{HYS}$ | | | | 0.05 × $V_{OVDD}$ | | V |

## Electrical Characteristics (continued)

($V_{DVDD}$ = $V_{AVDD}$ = +1.1V to +2.0V, $V_{OVDD}$ = +1.65V to +3.6V, $f_{FCLK}$ = 32.768kHz, $T_A$ = $T_{MIN}$ to $T_{MAX}$, unless otherwise noted. Typical values are at $V_{DVDD}$ = $V_{AVDD}$ = +1.8V, $V_{OVDD}$ = +2.5V, $T_A$ = +25°C.) (Note 2)

| PARAMETER | SYMBOL | CONDITIONS | | MIN | TYP | MAX | UNITS |
|---|---|---|---|---|---|---|---|
| Input Capacitance | $C_{IN}$ | | | | 10 | | pF |
| Input Current | $I_{IN}$ | | | -1 | | +1 | µA |
| **DIGITAL OUTPUTS (SDO, INTB, INT2B)** | | | | | | | |
| Output Voltage High | $V_{OH}$ | $I_{SOURCE}$ = 1mA | | $V_{OVDD}$ – 0.4 | | | V |
| Output Voltage Low | $V_{OL}$ | $I_{SINK}$ = 1mA | | | | 0.4 | V |
| Three-State Leakage Current | | | | -1 | | +1 | µA |
| Three-State Output Capacitance | | | | | 15 | | pF |
| **POWER SUPPLY** | | | | | | | |
| Analog Supply Voltage | $V_{AVDD}$ | Connect $V_{AVDD}$ to $V_{DVDD}$ | | 1.1 | | 2.0 | V |
| Digital Supply Voltage | $V_{DVDD}$ | Connect $V_{DVDD}$ to $V_{AVDD}$ | | 1.1 | | 2.0 | V |
| Interface Supply Voltage | $V_{OVDD}$ | Power for I/O drivers only | | 1.65 | | 3.6 | V |
| Supply Current | $I_{AVDD}$ + $I_{DVDD}$ | R-R Operation | $V_{AVDD}$ = $V_{DVDD}$ = +1.1V | | 76 | | µA |
| | | | $V_{AVDD}$ = $V_{DVDD}$ = +1.8V | | 100 | | |
| | | | $V_{AVDD}$ = $V_{DVDD}$ = +2.0V | | 110 | 122 | |
| | | ULP Lead On Detect | $T_A$ = +70°C | | 0.98 | | |
| | | | $T_A$ = +25°C | | 0.73 | 2.5 | |
| Interface Supply Current | $I_{OVDD}$ | $V_{OVDD}$ = +1.65V, ADC at 512sps (Note 9) | | | 0.2 | | µA |
| | | $V_{OVDD}$ = +3.6V, ADC at 512sps (Note 9) | | | 0.6 | 1.6 | |
| Shutdown Current | $I_{SAVDD}$ + $I_{SDVDD}$ | $V_{AVDD}$ = $V_{DVDD}$ = 2.0V | $T_A$ = +70°C | | 0.79 | | µA |
| | | | $T_A$ = +25°C | | 0.51 | 2.5 | |
| | $I_{SOVDD}$ | $V_{OVDD}$ = +3.6V, $V_{AVDD}$ = $V_{DVDD}$ = +2.0V | | | | 1.1 | |
| **ESD PROTECTION** | | | | | | | |
| INP, INN | | IEC61000-4-2 Contact Discharge (Note 10) | | | ±8 | | kV |
| | | IEC61000-4-2 Air-Gap Discharge (Note 10) | | | ±15 | | |
| | | HMM | | | ±8 | | |

## Timing Characteristics

($V_{DVDD}$ = $V_{AVDD}$ = +1.1V to +2.0V, $V_{OVDD}$ = +1.65V to +3.6V, $T_A$ = $T_{MIN}$ to $T_{MAX}$, unless otherwise noted. Typical values are at $V_{DVDD}$ = +1.8V, $V_{OVDD}$ = +2.5V, $T_A$ = +25°C.) (Notes 2, 3)

| PARAMETER | SYMBOL | CONDITIONS | MIN | TYP | MAX | UNITS |
|---|---|---|---|---|---|---|
| **TIMING CHARACTERISTICS** | | | | | | |
| SCLK Frequency | $f_{SCLK}$ | | 0 | | 12 | MHz |
| SCLK Period | $t_{CP}$ | | 83 | | | ns |
| SCLK Pulse Width High | $t_{CH}$ | | 15 | | | ns |
| SCLK Pulse Width Low | $t_{CL}$ | | 15 | | | ns |
| CSB Fall to SCLK Rise Setup Time | $t_{CSS0}$ | To 1st SCLK rising edge (RE) | 15 | | | ns |
| CSB Fall to SCLK Rise Hold Time | $t_{CSH0}$ | Applies to inactive RE preceding 1st RE | 0 | | | ns |
| CSB Rise to SCLK Rise Hold Time | $t_{CSH1}$ | Applies to 32nd RE, executed write | 10 | | | ns |
| CSB Rise to SCLK Rise | $t_{CSA}$ | Applies to 32nd RE, aborted write sequence | 15 | | | ns |
| SCLK Rise to CSB Fall | $t_{CSF}$ | Applies to 32nd RE | 100 | | | ns |
| CSB Pulse-Width High | $t_{CSPW}$ | | 20 | | | ns |
| SDI-to-SCLK Rise Setup Time | $t_{DS}$ | | 8 | | | ns |
| SDI to SCLK Rise Hold Time | $t_{DH}$ | | 8 | | | ns |
| SCLK Fall to SDO Transition | $t_{DOT}$ | $C_{LOAD}$ = 20pf | | | 40 | ns |
| | | $C_{LOAD}$ = 20pf, $V_{AVDD}$ = $V_{DVDD}$ ≥ 1.8V, $V_{DVDD}$ ≥ 2.5V | | | 20 | ns |
| SCLK Fall to SDO Hold | $t_{DOH}$ | $C_{LOAD}$ = 20pf | 2 | | | ns |
| CSB Fall to SDO Fall | $t_{DOE}$ | Enable time, $C_{LOAD}$ = 20pf | | | 30 | ns |

## Timing Characteristics (continued)

($V_{DVDD}$ = $V_{AVDD}$ = +1.1V to +2.0V, $V_{OVDD}$ = +1.65V to +3.6V, $T_A$ = $T_{MIN}$ to $T_{MAX}$, unless otherwise noted. Typical values are at $V_{DVDD}$ = +1.8V, $V_{OVDD}$ = +2.5V, $T_A$ = +25°C.) (Notes 2, 3)

| PARAMETER | SYMBOL | CONDITIONS | MIN | TYP | MAX | UNITS |
|---|---|---|---|---|---|---|
| CSB Rise to SDO Hi-Z | $t_{DOZ}$ | Disable time | | | 35 | ns |
| FCLK Frequency | $f_{FCLK}$ | External reference clock | 32.768 | | | kHz |
| FCLK Period | $t_{FP}$ | | | 30.52 | | µs |
| FCLK Pulse-Width High | $t_{FH}$ | 50% duty cycle assumed | | 15.26 | | µs |
| FCLK Pulse-Width Low | $t_{FL}$ | 50% duty cycle assumed | | 15.26 | | µs |

**Note 2:** Limits are 100% tested at $T_A$ = +25°C. Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization.
**Note 3:** Guaranteed by design and characterization. Not tested in production.
**Note 4:** One electrode drive with <10Ω source impedance, the other driven with 51kΩ in parallel with a 47nF per IEC60601-2-47.
**Note 5:** Inputs connected to 51kΩ in parallel with a 47nF to $V_{CM}$.
**Note 6:** Use this setting only for $V_{AVDD}$ = $V_{DVDD}$ ≥ 1.65V.
**Note 7:** Use this setting only for $V_{AVDD}$ = $V_{DVDD}$ ≥ 1.55V.
**Note 8:** Use this setting only for $V_{AVDD}$ = $V_{DVDD}$ ≥ 1.45V.
**Note 9:** $f_{SCLK}$ = 4MHz, burst mode, EFIT = 8, $C_{SDO}$ = $C_{INTB}$ = 50pF.
**Note 10:** ESD test performed with 1kΩ series resistor designed to withstand 8kV surge voltage.

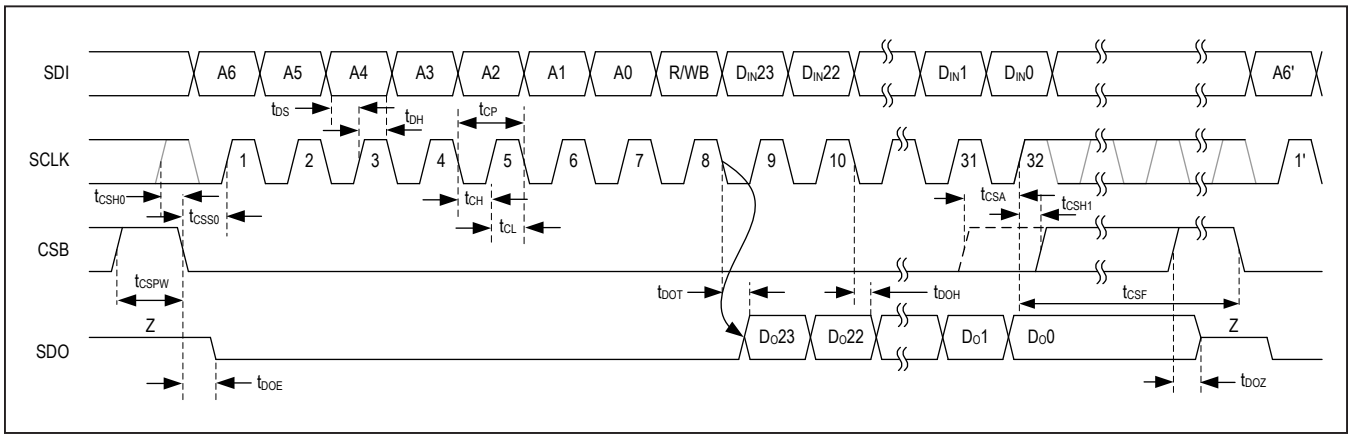



*Figure 1a. SPI Timing Diagram*

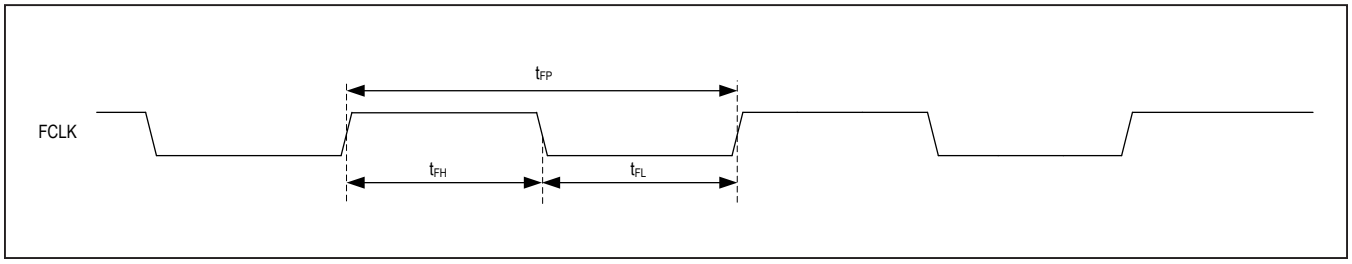
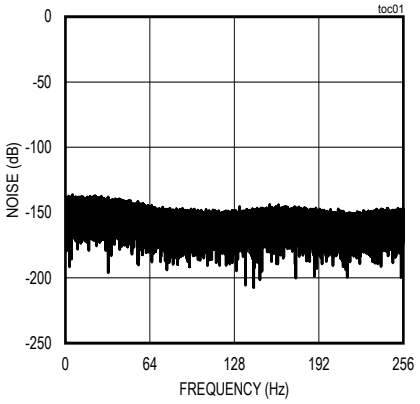



*Figure 1b. FCLK Timing Diagram*

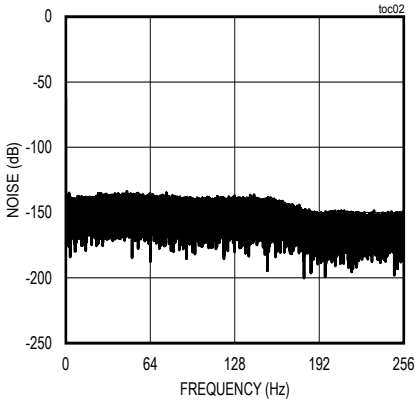
## Typical Operating Characteristics

($V_{DVDD}$ = $V_{AVDD}$ = +1.8V, $V_{OVDD}$ = 2.5V, $T_A$ = +25°C, unless otherwise noted.)

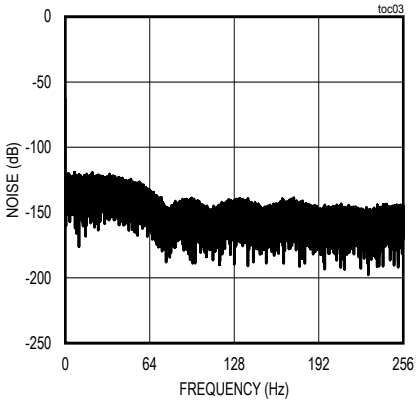
CHANNEL NOISE SPECTRUM vs. FREQUENCY
INPUTS SHORTED, GAIN = 20, LPF = 40Hz



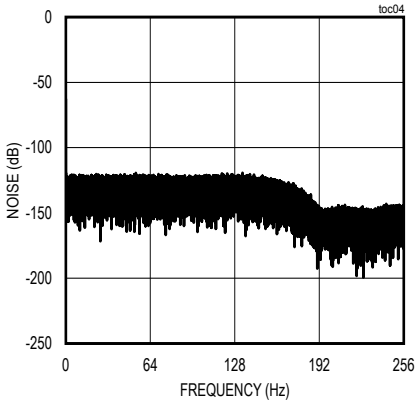

CHANNEL NOISE SPECTRUM vs. FREQUENCY
INPUTS SHORTED, GAIN = 20, LPF = 150Hz



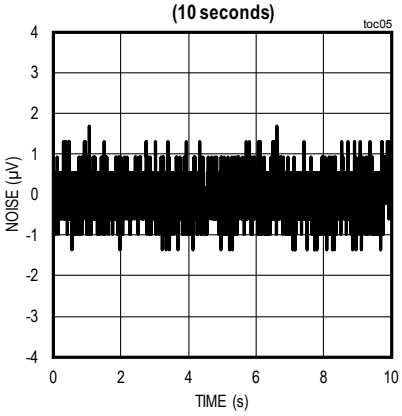

CHANNEL NOISE SPECTRUM vs. FREQUENCY
INPUTS SHORTED, GAIN = 160, LPF = 40Hz



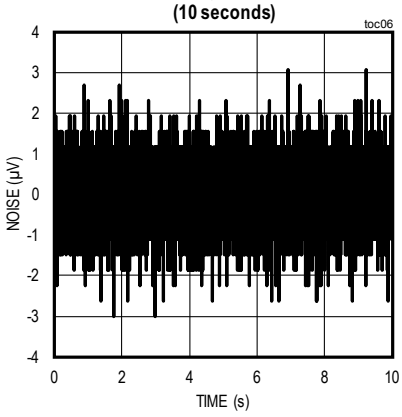

CHANNEL NOISE SPECTRUM vs. FREQUENCY
INPUTS SHORTED, GAIN = 160, LPF = 150Hz



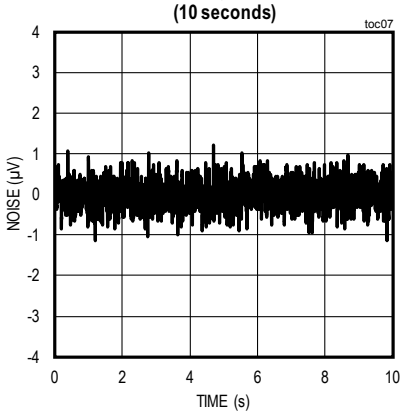

CHANNEL INPUT-REFERRED NOISE vs.TIME
GAIN = 20, LPF = 40Hz
(10 seconds)




CHANNEL INPUT-REFERRED NOISE vs. TIME
GAIN = 20, LPF = 150Hz
(10 seconds)



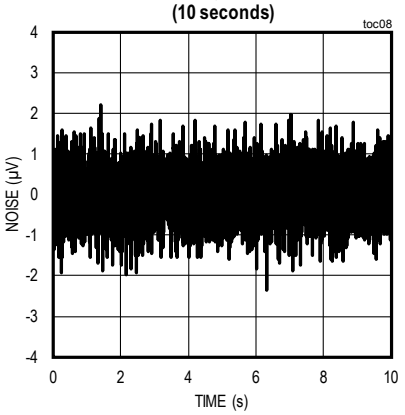

CHANNEL INPUT-REFERRED NOISE vs. TIME
GAIN = 160, LPF = 40Hz
(10 seconds)



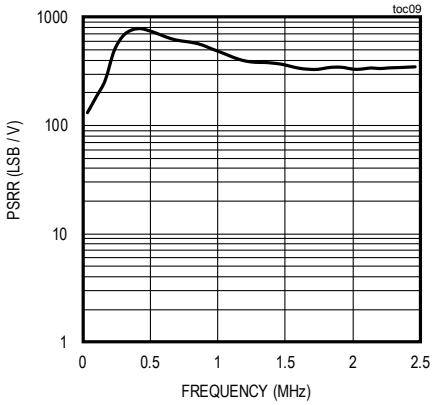
## Typical Operating Characteristics

($V_{DVDD}$ = $V_{AVDD}$ = +1.8V, $V_{OVDD}$ = 2.5V, $T_A$ = +25°C, unless otherwise noted.)

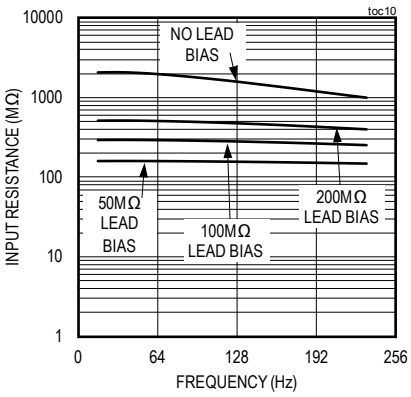
**CHANNEL INPUT-REFERRED NOISE vs. TIME GAIN = 160, LPF = 150Hz (10 seconds)**



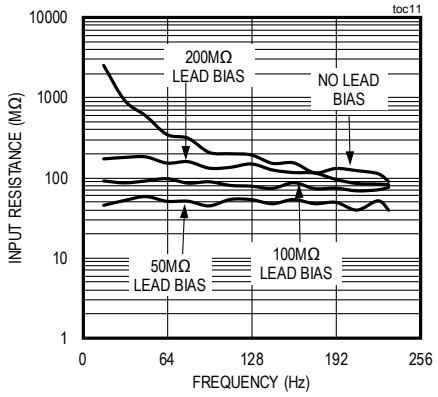

**CHANNEL PSRR vs. FREQUENCY**



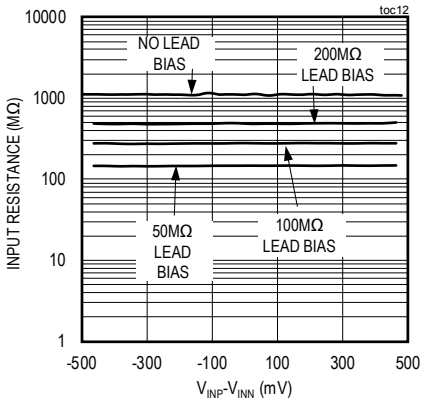

**DIFFERENTIAL INPUT RESISTANCE vs. FREQUENCY**



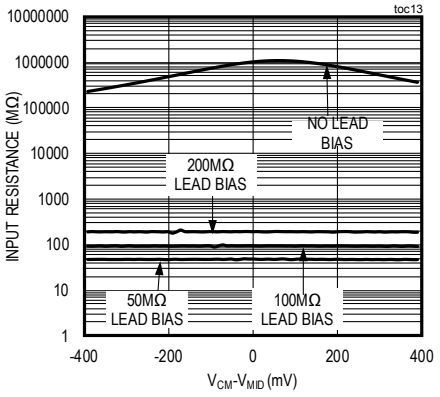

**COMMON-MODE INPUT RESISTANCE vs. FREQUENCY**



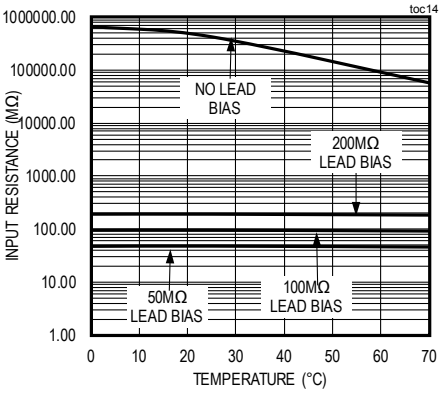

**DIFFERENTIAL INPUT RESISTANCE vs. VOLTAGE**




**COMMON-MODE INPUT RESISTANCE vs. VOLTAGE**



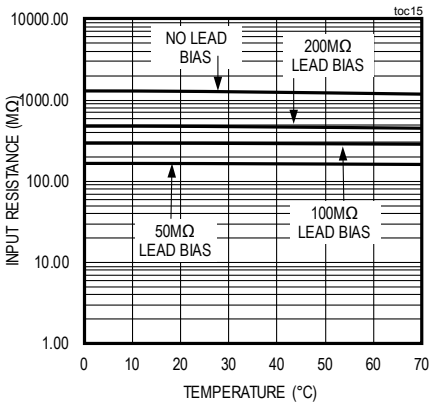

**COMMON-MODE INPUT RESISTANCE vs. TEMPERATURE**



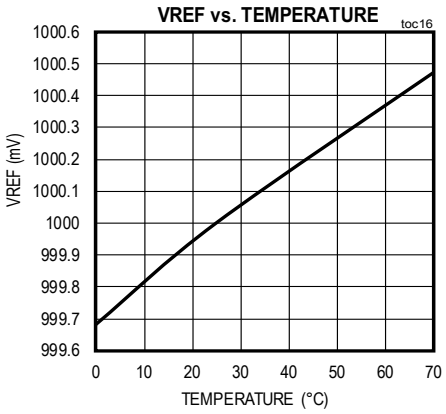
## Typical Operating Characteristics

($V_{DVDD}$ = $V_{AVDD}$ = +1.8V, $V_{OVDD}$ = 2.5V, $T_A$ = +25°C, unless otherwise noted.)

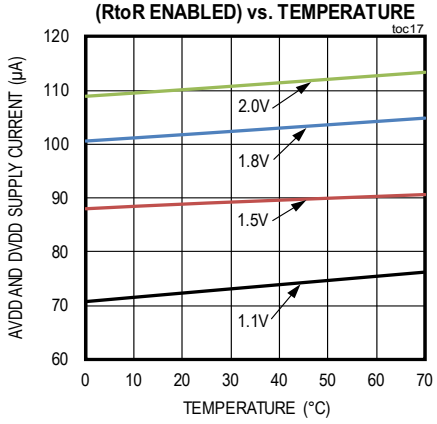
**DIFFERENTIAL INPUT RESISTANCE vs. TEMPERATURE**



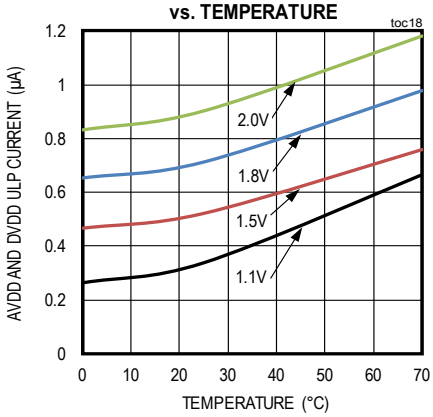

**VREF vs. TEMPERATURE**



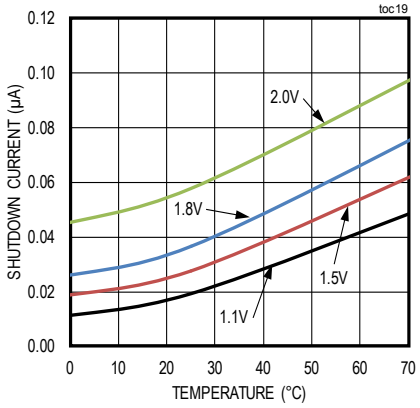

**AVDD AND DVDD SUPPLY CURRENT (RtoR ENABLED) vs. TEMPERATURE**



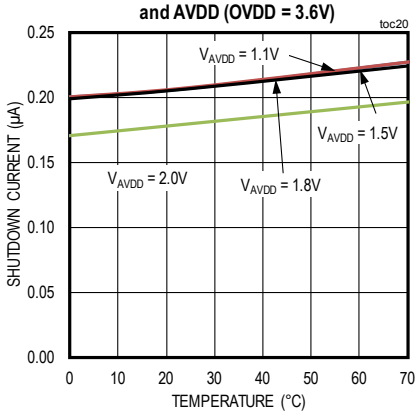

**AVDD AND DVDD ULP CURRENT vs. TEMPERATURE**



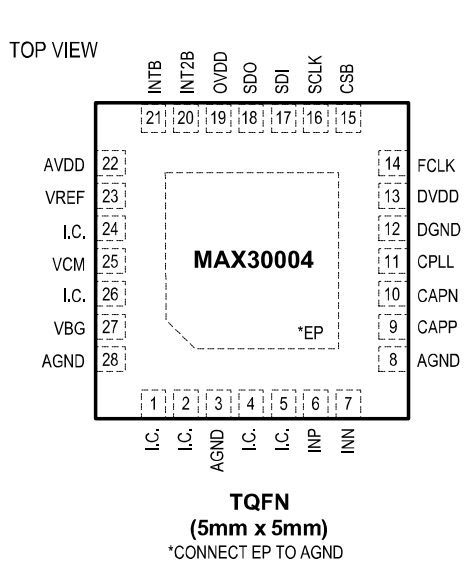

**AVDD SHUTDOWN CURRENT**



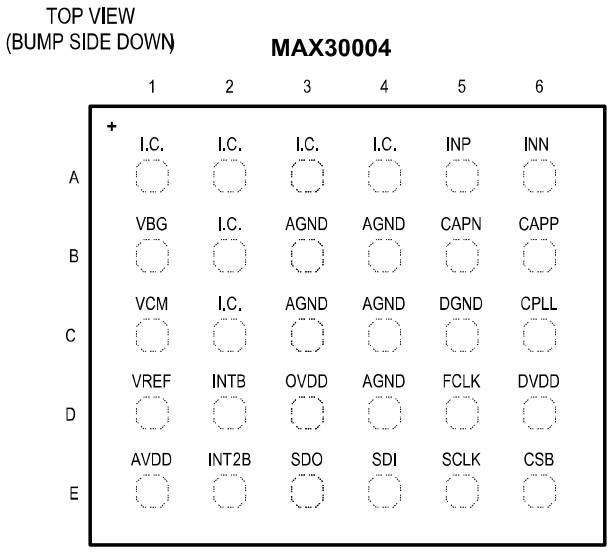

**OVDD SHUTDOWN CURRENT vs. TEMPERATURE and AVDD (OVDD = 3.6V)**



## Pin Configurations




**TQFN**
**(5mm x 5mm)**
*CONNECT EP TO AGND




**WLP**
**(2.7mm x 2.9mm)**

## Pin Description

| PIN | BUMP | NAME | FUNCTION |
|---|---|---|---|
| TQFN | WLP | | |
| 1, 2, 4, 5, 24, 26 | A1, A2, A3, A4, B2, C2 | I.C. | Internally Connected. Connect to AGND. |
| 3,8,28 | B3, B4, C3, C4, D4 | AGND | Analog Power and Reference Ground. Connect into the printed circuit board ground plane. |
| 6 | A5 | INP | Positive Input |
| 7 | A6 | INN | Negative Input |
| 9 | B6 | CAPP | Analog High-Pass Filter Input. Connect a 1µF X7R capacitor (CHPF) between CAPP and CAPN to form a 0.5Hz high-pass response in the channel. |
| 10 | B5 | CAPN | Analog High-Pass Filter Input. Connect a 1µF X7R capacitor (CHPF) between CAPP and CAPN to form a 0.5Hz high-pass response in the channel. |
| 11 | C6 | CPLL | PLL Loop Filter Input. Connect 1nf COG cap between CPLL and AGND. |

## Pin Description (continued)

| PIN | BUMP | NAME | FUNCTION |
|---|---|---|---|
| TQFN | WLP | | |
| 12 | C5 | DGND | Digital Ground for Both Digital Core and I/O Pad Drivers. Recommended to connect to AGND plane. |
| 13 | D6 | DVDD | Digital Core Supply Voltage. Connect to AVDD |
| 14 | D5 | FCLK | External 32.768kHz Clock that Controls the Sampling of the Internal Sigma-Delta Converters and Decimator. |
| 15 | E6 | CSB | Active-Low Chip-Select Input. Enables the serial interface. |
| 16 | E5 | SCLK | Serial Clock Input. Clocks data in and out of the serial interface when CSB is low. |
| 17 | E4 | SDI | Serial Data Input. SDI is sampled into the device on the rising edge of SCLK when CSB is low. |
| 18 | E3 | SDO | Serial Data Output. SDO will change state on the falling edge of SCLK when CSB is low. SDO is three-stated when CSB is high. |
| 19 | D3 | OVDD | Logic Interface Supply Voltage |
| 20 | E2 | INT2B | Interrupt 2 Output. INT2B is an active-low status output. It can be used to interrupt an external device. |
| 21 | D2 | INTB | Interrupt Output. INTB is an active low status output. It can be used to interrupt an external device. |
| 22 | E1 | AVDD | Analog Core Supply Voltage. Connect to DVDD. |
| 23 | D1 | VREF | ADC Reference Buffer Output. Connect a 10µF X5R ceramic capacitor between VREF and AGND. |
| 25 | C1 | VCM | Common Mode Buffer Output. Connect a 10µF X5R ceramic capacitor between VCM and AGND. |
| 27 | B1 | VBG | Bandgap Noise Filter Output. Connect a 1.0µF X7R ceramic capacitor between VBG and AGND. |
| EP | — | — | Exposed Paddle. Connect to AGND. |

## Detailed Description

### Biopotential Channel

Figure 2 illustrates the biopotential channel block diagram, excluding the ADC. The channel comprises an input MUX, a fast-recovering instrumentation amplifier, an anti-alias filter, and a programmable gain amplifier. The MUX includes several features such as ESD protection, EMI filtering, lead biasing, leads off checking, and ultra-low power leads-on checking. The output of this analog channel drives a high-resolution ADC.

### Input MUX

The input MUX shown in Figure 3 contains integrated ESD and EMI protection, DC leads off detect current sources, lead-on detect, series isolation switches, and lead biasing.

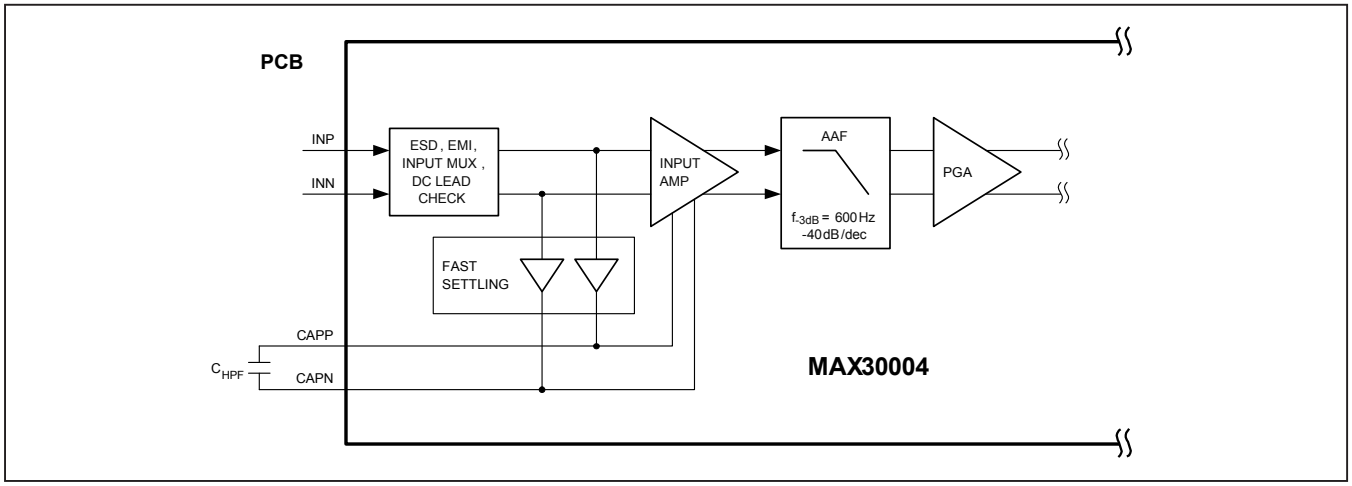



*Figure 2. Channel Input Amplifier and PGA Excluding the ADC*

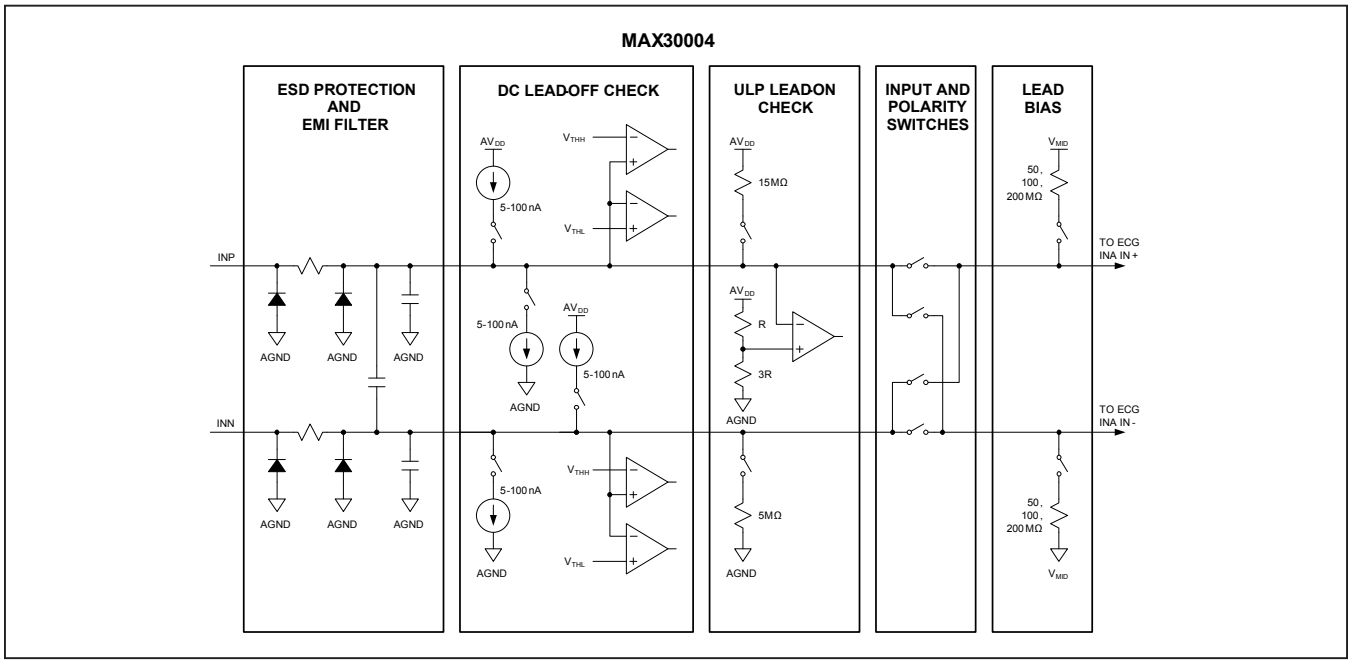



*Figure 3. Input MUX*

## EMI Filtering and ESD Protection

EMI filtering of the INP and INN inputs consists of a single pole, low pass, differential, and common mode filter with the pole located at approximately 2MHz. The INP and INN inputs also have input clamps that protect the inputs from ESD events.

- ±8kV using the Contact Discharge method specified in IEC61000-4-2 ESD
- ±15kV using the Air Gap Discharge method specified in IEC61000-4-2 ESD
- ±8kV HBM
- For IEC61000-4-2 ESD protection, use 1kΩ series resistors on INP and INN that is rated to withstand 8kV surge voltages.

### DC Leads-Off Detection and ULP Leads-On Detection

The input MUX leads-off detect circuitry consists of programmable sink/source DC current sources that allow for DC leads-off detection, while the channel is powered up in normal operation and an ultra-low-power (ULP) leads-on detect while the channel is powered-down.

The MAX30004 accomplishes DC leads-off detection by applying a DC current to pull the input voltage up to above $V_{MID}$ + $V_{TH}$ or down to below $V_{MID}$ - $V_{TH}$. The current sources have user selectable values of 0nA, 5nA, 10nA, 20nA, 50nA, and 100nA that allow coverage of dry and wet electrode impedance ranges. Supported thresholds are $V_{MID}$ ± 0.30V (recommended), $V_{MID}$ ± 0.40V, $V_{MID}$ ± 0.45V, and $V_{MID}$ ± 0.50V. A threshold of 400mV, 450mV, and 500mV should only be used when $V_{AVDD}$ ≥ 1.45V, 1.55V, and 1.65V, respectively. A dynamic comparator protects against false flags generated by the input amplifier and input chopping. The comparator checks for a minimum continuous violation (or threshold exceeded) of 115ms to 140ms depending on the setting of FMSTR[1:0] before asserting any one of the LDOFF_* interrupt flags (Figure 4). See registers CNFG_GEN (0x10) and CNFG_MUX 0x14) for configuration settings and see Table 1 for recommended values given electrode type and supply voltage.

The ULP lead on detect operates by pulling INN low with a pulldown resistance larger than 5mΩ and pulling INP high with a pullup resistance larger than 15mΩ. A low-power comparator determines if INP is pulled below a predefined threshold that occurs when both electrodes make contact with the body. When the impedance between INP and INN is less than 20Ω, an interrupt LONINT is asserted, alerting the µC to a leads-on condition.

A 0nA/$V_{MID}$ ± 300mV selection is available allowing monitoring of the input compliance of the INA during non-DC lead-off checks.

### Lead Bias

The MAX30004 limits the INP and INN DC input common mode range to $V_{MID}$ ±150mV. This range can be maintained either through external/internal lead-biasing.

Internal DC lead-biasing consists of 50MΩ, 100MΩ, or 200MΩ selectable resistors to $V_{MID}$ that drive the electrodes within the input common mode requirements of the channel and can drive the connected body to the proper common mode voltage level. See register CNFG_GEN (0x10) to select a configuration.

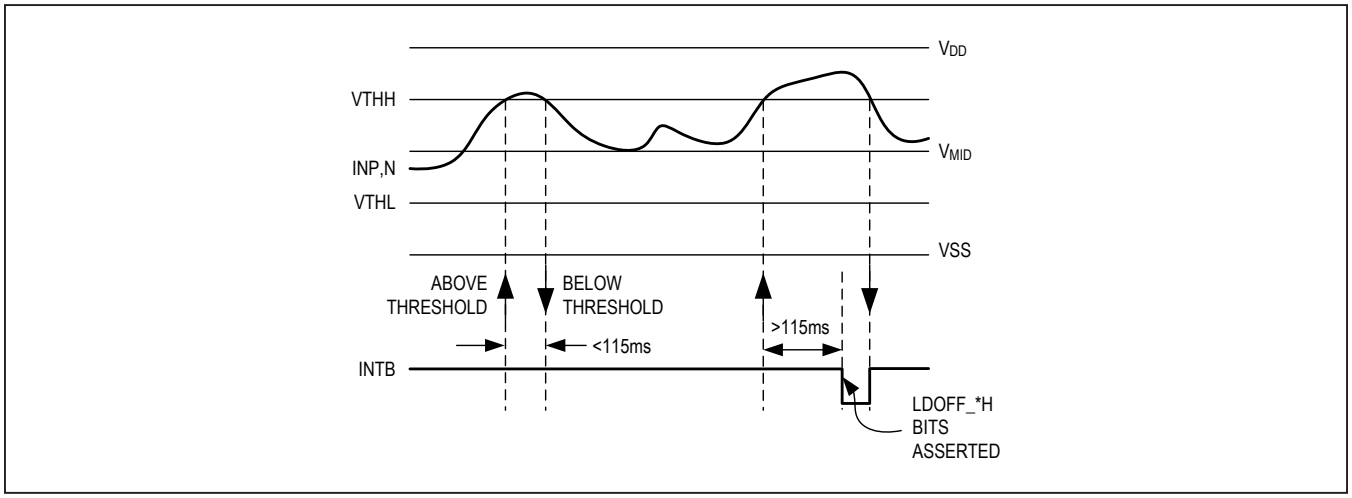



*Figure 4. Lead-Off Detect Behavior*

## Table 1. Recommended Lead Bias (Rb), Current Source Values, and Thresholds for Electrode Impedances

<table>
<tr><th rowspan="2">$I_{DC}$<br>$V_{TH}$</th><th colspan="8">ELECTRODE IMPEDANCE</th></tr>
<tr><th><100kΩ</th><th>100kΩ - 200kΩ</th><th>200kΩ - 400kΩ</th><th>400kΩ - 1MΩ</th><th>1MΩ - 2MΩ</th><th>2MΩ - 4 MΩ</th><th>4MΩ - 10MΩ</th><th>10MΩ - 20MΩ</th></tr>
<tr><td>$I_{DC}$ = 10nA</td><td colspan="8">All settings of $R_b$<br>$V_{TH}$ = $V_{MID}$<br>± 300mV, ±400mV</td></tr>
<tr><td>$I_{DC}$ = 20nA</td><td colspan="7">All settings of $R_b$<br>All settings of $V_{TH}$</td><td>All settings of $R_b$<br>$V_{TH}$ = $V_{MID}$<br>±400mV,<br>±450mV,<br>±500mV</td></tr>
<tr><td>$I_{DC}$ = 50nA</td><td colspan="6">All settings of $R_b$<br>All settings of $V_{TH}$</td><td>All settings of $R_b$<br>$V_{TH}$ = $V_{MID}$<br>±450mV,<br>±500mV</td><td></td></tr>
<tr><td>$I_{DC}$ = 100nA</td><td colspan="5">All settings of $R_b$<br>All settings of $V_{TH}$</td><td>All settings of Rb<br>$V_{TH}$ = $V_{MID}$<br>±400mV,<br>±450mV,<br>±500mV</td><td></td><td></td></tr>
</table>

### Isolation Switches

The series switches in the MAX30004 isolate INP and INN pins (subject) from the internal signal path. the series switches are disabled by default. They must be enabled to record R-to-R data.

### Gain Settings and Input Range

The device's biopotential channel contains an input instrumentation amplifier that provides low-noise, fixed 20V/V gain amplification of the differential signal, rejects differential DC voltage due to electrode polarization, rejects common-mode interference primarily due to AC mains interference, and provides high input impedance to guarantee high CMRR even in the presence of severe electrode impedance mismatch (see Figure 2). The differential DC rejection corner frequency is set by an external capacitor ($C_{HPF}$) placed between pins CAPP and CAPN, refer to Table 2 for appropriate value selection. There are three recommended options for the cutoff frequency: 5Hz, 0.5Hz, and 0.05Hz. Setting the cutoff frequency to 5Hz provides the most motion artifact rejection, making it best suited for heart rate monitoring. 0.5Hz and 0.05Hz can be used for applications requiring moderate and no motion artifact rejection respectively. The high-pass corner frequency is calculated by the following equation:

$$1/(2\pi \times R_{HPF} \times C_{HPF})$$

RHPF is specified in the *Electrical Characteristics* table.

Following the instrumentation amplifier is a 2-pole active anti-aliasing filter with a 600Hz -3dB frequency that provides 57dB of attenuation at half the modulator sampling rate (approximately 16kHz) and a PGA with programmable gains of 1, 2, 4, and 8V/V for an overall gain of 20, 40, 80, and 160V/V. The instrumentation amplifier and PGA are chopped to minimize offset and 1/f noise. Gain settings are configured through the CNFG_CH (0x15) register. The useable common-mode range is $V_{MID}$ ±150mV, internal lead biasing can be used to meet this requirement. The useable DC differential range is ±300mV to allow for electrode polarization voltages on each electrode. The input AC differential range is ±32.5mV or ±65$mV_{P-P}$.

## Table 2. Analog HPF Corner Frequency Selection

| $C_{HPF}$ | HPF CORNER FREQUENCY |
|---|---|
| 0.1µ | ≤ 5Hz |
| 1.0µ | ≤ 0.5Hz |
| 10µ | ≤ 0.05Hz |

### Fast Recovery Mode

The input instrumentation amplifier has the ability to rapidly recover from an excessive overdrive event such as a defibrillation pulse, high-voltage external pacing, and electro-surgery interference. There are two modes of recovery that can be used: automatic or manual recovery. The mode is programmed by the FAST[1:0] bits in the MNGR_DYN (0x05) register.

Automatic mode engages once the saturation counter exceeds approximately 125ms ($t_{SAT}$). The counter is activated the first time the ADC output exceeds the symmetrical threshold defined by the FAST_TH[5:0] bits in the MNGR_DYN (0x05) register and accumulates the time that the ADC output exceeds either the positive or negative threshold. If the saturation counter exceeds 125ms, it triggers the fast settling mode (if enabled) and resets. The saturation counter can also be reset prior to triggering the fast settling mode if the ADC output falls below the threshold continuously for 125ms ($t_{BLW}$). This feature is designed to avoid false triggers due to the QRS complex. Once triggered, fast settling mode will be engaged for 500ms, see Figure 5.

In manual mode, a user algorithm running on the host microcontroller or an external stimulus input will generate the trigger to enter fast recovery mode. The host microcontroller then enables the manual fast recovery mode in the MNGR_DYN (0x05) register. The manual fast recovery mode can be of a much shorter duration than the automatic mode and allows for more rapid recovery. One such example is recovery from external high-voltage pacing signals in a few milliseconds to allow the observation of a subsequent p-wave.

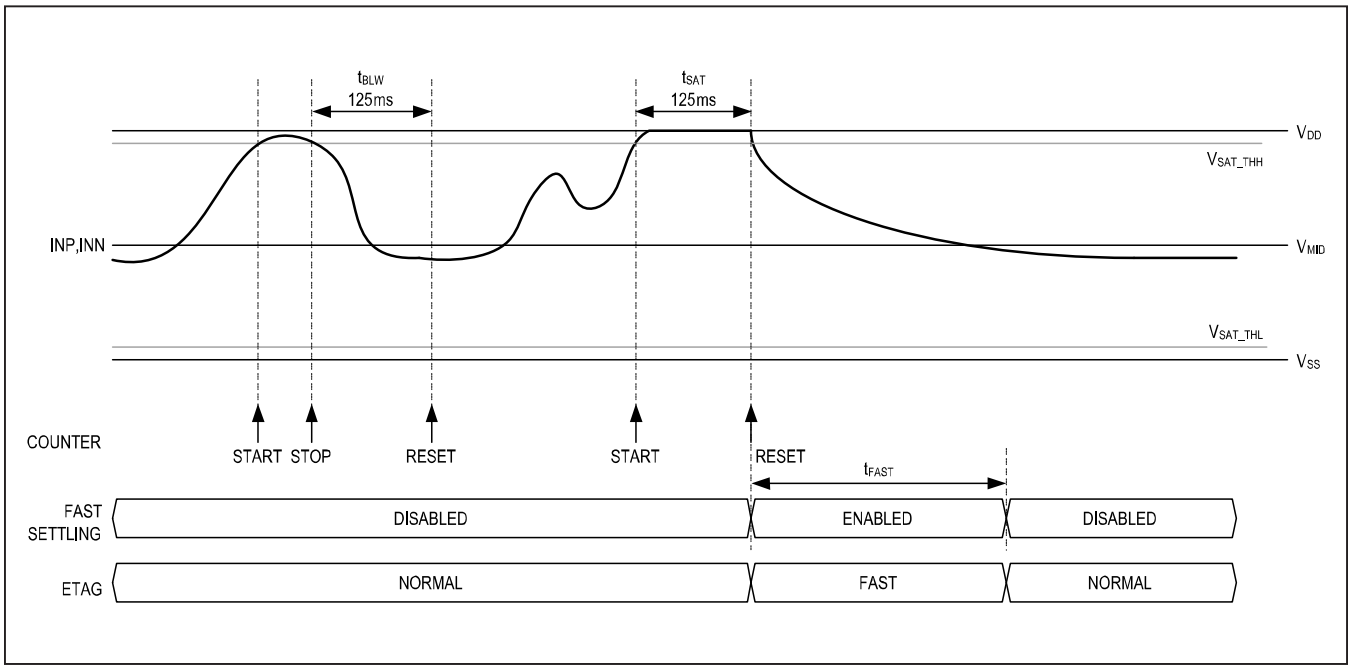



*Figure 5. Automatic Fast Settling Behavior*

## R-to-R Detection

The MAX30004 contains built-in hardware to detect R-to-R intervals using an adaptation of the Pan-Tompkins QRS detection algorithm[1]. The timing resolution of the R-R interval is approximately 8ms and depends on the setting of FMSTR [1:0] in CNFG_GEN (0x10) register. See Table 19 for the timing resolution of each setting.

When an R event is identified, the RRINT status bit is asserted and the RTOR_REG (0x25) register is updated with the count seen since the last R event. Figure 6 illustrates the R-R interval on a QRS complex. Refer to registers CNFG_RTOR1 (0x1D) and CNFG_RTOR2 (0x1E) for configuration details.

The latency of the R-to-R value written to the RTOR Interval Memory Register is the sum of the R-to-R decimation delay and the R-to-R detection delay blocks. The R-to-R decimation factor is fixed at 256 and the decimation delay ($t_{R2R_DEC}$) is always 3,370 FMSTR clocks, as shown in Table 3.

The detection circuit consists of several digital filters and signal processing delays. These depend on the WNDW[3:0] bits in the CNFG_RTOR (0x1D) register. The detection delay ($t_{R2R_DET}$) is described by the following equation:

$t_{R2R_DET}$ = 5,376 + 256 x WNDW in FMSTR clocks where WNDW is an integer from 0 to 15 and the total latency ($t_{R2R_DEL}$) is the sum of the two delays and summarized in the equation below:

$t_{R2R_DEL} = t_{R2R_DEC} + t_{R2R_DET}$ = 3,370 + 5,376 + 256 x WNDW in FMSTR clocks where WNDW is an integer from 0 to 15.

## Reference and Common Mode Buffer

The MAX30004 features internally generated reference voltages. The bandgap output ($V_{BG}$) pin requires an external 1.0µF capacitor to AGND and the reference output ($V_{REF}$) pin requires a 10µF external capacitor to AGND for compensation and noise filtering.

A common-mode buffer is provided to buffer 650mV which is used to drive common mode voltages for internal blocks. Use a 10µF external capacitor between $V_{CM}$ to AGND to provide compensation and noise filtering.

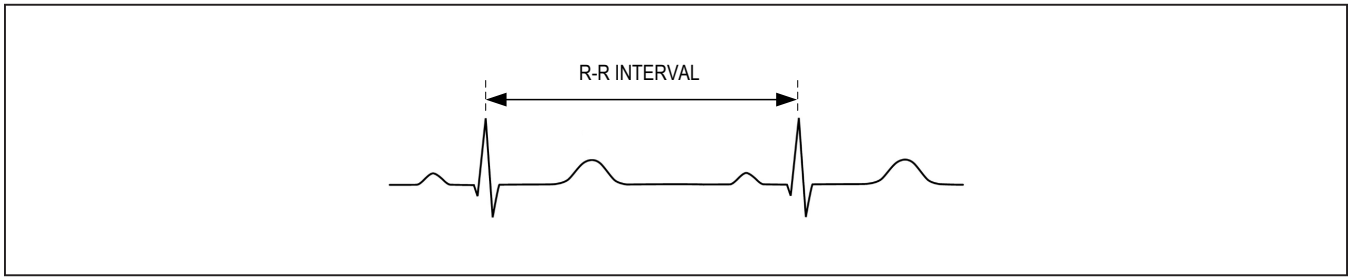



*Figure 6. R-to-R Interval Illustration*

## Table 3. R-to-R Decimation Delay in ms and FMSTR CLK vs. Register Settings, FCLK = 32.768Hz

| FMSTR [1:0] | FMSTR FREQ IN FCLKs | FMSTR FREQ (Hz) | DECIMATION | RTOR TIME RESOLUTION (ms) | DELAY IN R-TO-R DECIMATION | |
|---|---|---|---|---|---|---|
| | | | | | IN FMSTR CLKs | IN ms |
| 00 | FCLK | 32,768 | 256 | 7.8125 | 3370 | 102.844 |
| 01 | FCLK x 625/640 | 32,000 | 256 | 8.0 | 3370 | 105.313 |
| 10 | FCLK x 625/640 | 32,000 | 256 | 8.0 | 3370 | 105.313 |
| 11 | FCLK x 640/656 | 31,968.78 | 256 | 8.0078 | 3370 | 105.415 |

[1] *J. Pan and W.J. Tompkins, "A Real-Time QRS Detection Algorithm," IEEE Trans. Biomed. Eng., vol. 32, pp. 230-236*

## Sample Synchronization Pulse

The MAX30004 offers a sample synchronization pulse that allows the direct observation of the channel sample instant for either synchronization between multiple devices or as a monitoring feature during debug. When enabled (EN_SAMP in either register EN_INT or EN_INT2), the MAX30004 generates an interrupt either every sample instant, or every second, fourth, or 16th sample instant, based on the setting of SAMP_IT[1:0] in register MNGR_INT. The sample instants are defined by the channel ADC, and are not R-to-R samples. Therefore, the interval between individual sample instants is dependent on the channel data rate as defined by FMSTR[1:0] and RATE[1:0]. The clear behavior of the sample synchronization pulse is affected by the CLR_SAMP bit in register MNGR_INT. When this feature is used, it is recommended to use a dedicated interrupt output for just the sample synchronization pulse.

## SPI Interface Description

### 32-Bit Read/Write Sequences

The MAX30004 interface is SPI/QSPI/Micro-wire/DSP compatible. The operation of the SPI interface is shown in Figure 1. Data is strobed into the MAX30004 on SCLK rising edges. The device is programmed and accessed by a 32 cycle SPI instruction framed by a CSB low interval. The content of the SPI operation consists of a one byte command word (comprised of a seven bit address and a Read/$\overline{\text{Write}}$ mode indicator (i.e., A[6:0] + R/$\overline{\text{W}}$) followed by a three-byte data word. The MAX30004 is compatible with the CPOL = 0/CPHA = 0 and CPOL = 1/CPHA = 1 modes of operation.

Write mode operations will be executed on the 32nd SCLK rising edge using the first four bytes of data available. In write mode, any data supplied after the 32nd SCLK rising edge will be ignored. Subsequent writes require CSB to de-assert high and then assert low for the next write command. In order to abort a command sequence, the rise of CSB must precede the updating (32nd) rising-edge of SCLK, meeting the $t_{CSA}$ requirement.

Read mode operations will access the requested data on the 8th SCLK rising edge, and present the MSB of the requested data on the following SCLK falling edge, allowing the µC to sample the data MSB on the 9th SCLK rising edge. Configuration and status data are all available via normal mode read back sequences. If more than 32 SCLK rising edges are provided in a normal read sequence then the excess edges will be ignored and the device will read back zeros.

If accessing the STATUS register, all interrupt updates will be made in response to the 30th SCLK rising edge, allowing for internal synchronization operations to occur.

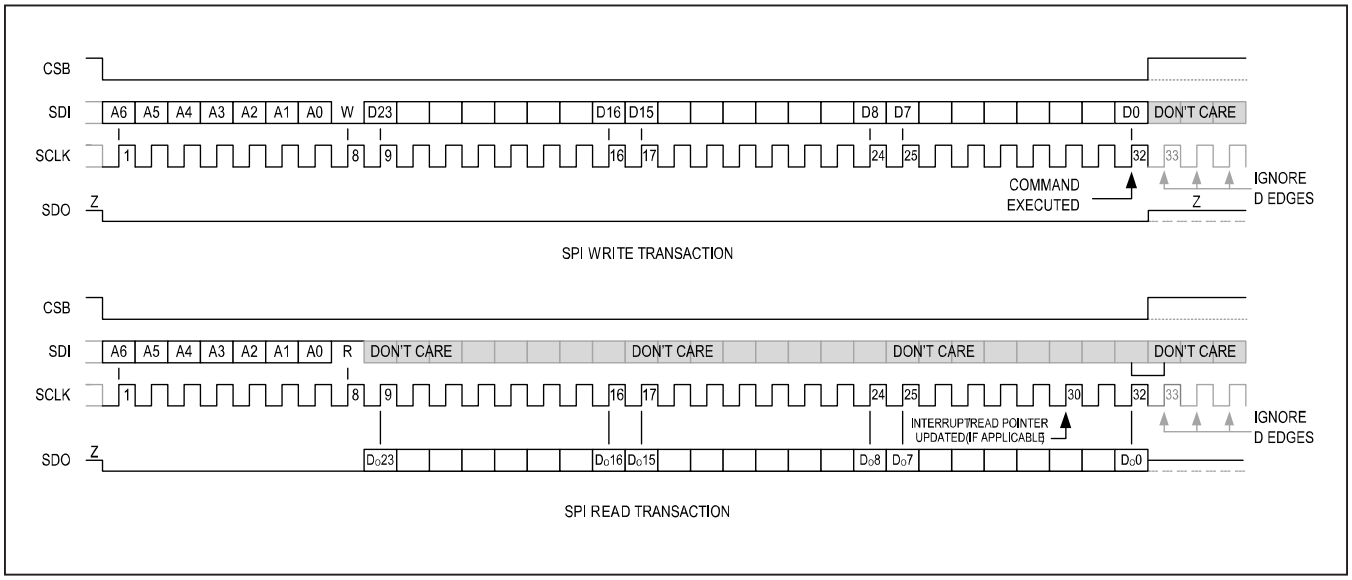



*Figure 7. SPI Normal Mode Transaction Diagrams*

## User Command and Register Map

| REG [6:0] | NAME | R/W MODE | DATA INDEX ($D_{INDEX}$) | | | | | | | |
|---|---|---|---|---|---|---|---|---|---|---|
| | | | 23 / 15 / 7 | 22 / 14 / 6 | 21 / 13 / 5 | 20 / 12 / 4 | 19 / 11 / 3 | 18 / 10 / 2 | 17 / 9 / 1 | 16 / 8 / 0 |
| 0x00 | NO-OP | R/W | x / x / x | x / x / x | x / x / x | x / x / x | x / x / x | x / x / x | x / x / x | x / x / x |
| 0x01 | STATUS | R | x | x | FSTINT | DCLOFFINT | x | x | x | x |
| | | | x | x | x | x | LONINT | RRINT | SAMP | PLLINT |
| | | | 0 | 0 | x | x | LOFF_PH | LOFF_PL | LOFF_NH | LOFF_NL |
| 0x02<br>0x03 | EN_INT<br>EN_INT2 | R/W | x | x | EN_FSTINT | EN_DCLOFFINT | x | x | x | x |
| | | | x | x | x | x | EN_LONINT | EN_RRINT | EN_SAMP | EN_PLLINT |
| | | | x | x | x | x | x | x | INTB_TYPE[1:0] | |
| 0x04 | MNGR_INT | R/W | x | x | x | x | x | x | x | x |
| | | | x | x | x | x | x | x | x | x |
| | | | x | CLR_FAST | CLR_RRINT[1:0] | | x | CLR_SAMP | SAMP_IT[3:0] | |
| 0x05 | MNGR_DYN | R/W | FAST[1:0] | | FAST_TH[5:0] | | | | | |
| | | | x | x | x | x | x | x | x | x |
| | | | x | x | x | x | x | x | x | x |
| 0x08 | SW_RST | W | Data Required for Execution = 0x000000 | | | | | | | |
| 0x09 | RESTART | W | Data Required for Execution = 0x000000 | | | | | | | |
| 0x0A | RTOR_RST | W | Data Required for Execution = 0x000000 | | | | | | | |
| 0x0F | INFO | R | 0 | 1 | 0 | 1 | REV_ID[3:0] | | | |
| | | | x | x | 0 | 0 | x | x | x | x |
| | | | x | x | x | x | x | x | x | x |
| 0x10 | CNFG_GEN | R/W | EN_ULP_LON[1:0] | | FMSTR[1:0] | | EN_CH | x | x | x |
| | | | x | x | EN_DCLOFF[1:0] | | IPOL | IMAG[2:0] | | |
| | | | VTH[1:0] | | EN_RBIAS[1:0] | | RBIASV[1:0] | | RBIASP | RBIASN |

## User Command and Register Map (continued)

| REG [6:0] | NAME | R/W MODE | DATA INDEX ($D_{INDEX}$) | | | | | | | |
|---|---|---|---|---|---|---|---|---|---|---|
| | | | 23 / 15 / 7 | 22 / 14 / 6 | 21 / 13 / 5 | 20 / 12 / 4 | 19 / 11 / 3 | 18 / 10 / 2 | 17 / 9 / 1 | 16 / 8 / 0 |
| 0x14 | CNFG_ MUX | R/W | POL | x | OPENP | OPENN | x | x | x | x |
| | | | x | x | x | x | x | x | x | x |
| | | | x | x | x | x | x | x | x | x |
| 0x15 | CNFG_ | R/W | RATE[1:0] | | x | x | x | x | GAIN[1:0] | |
| | | | x | DHPF | DLPF[1:0] | | x | x | x | x |
| | | | x | x | x | x | x | x | x | x |
| 0x1D | CNFG_ RTOR1 | R/W | WNDW[3:0] | | | | RGAIN[3:0] | | | |
| | | | EN_RTOR | x | PAVG[1:0] | | PTSF[3:0] | | | |
| | | | x | x | x | x | x | x | x | x |
| 0x1E | CNFG_ RTOR2 | R/W | x | x | HOFF[5:0] | | | | | |
| | | | x | x | RAVG[1:0] | | x | RHSF[2:0] | | |
| | | | x | x | x | x | x | x | x | x |
| 0x25 | RTOR | R | R-to-R Interval Register Read Back | | | | See *RTOR DATA* description for details | | | |
| 0x7F | NO-OP | R/W | x/x/x | x/x/x | x/x/x | x/x/x | x/x/x | x/x/x | x/x/x | x/x/x |

## Register Descriptions

### NO_OP (0x00 and 0x7F) Registers

No Operation (NO_OP) registers are read-write registers that have no internal effect on the device. If these registers are read back, DOUT remains zero for the entire SPI transaction. Any attempt to write to these registers is ignored without impact to internal operation.

### STATUS (0x01) Register

STATUS is a read-only register that provides a comprehensive overview of the current status of the device. The first two bytes indicate the state of all interrupt terms (regardless of whether interrupts are enabled in registers EN_INT (0x02) or EN_INT2 (0x03)). All interrupt terms are active high. The last byte includes detailed status information for conditions associated with the interrupt terms.

## Table 4. STATUS (0x01) Register Map

| REG | NAME | R/W | 23 / 15 / 7 | 22 / 14 / 6 | 21 / 13 / 5 | 20 / 12 / 4 | 19 / 11 / 3 | 18 / 10 / 2 | 17 / 9 / 1 | 16 / 8 / 0 |
|---|---|---|---|---|---|---|---|---|---|---|
| 0x01 | STATUS | R | x | x | FSTINT | DCLOFF INT | x | x | x | x |
| | | | x | x | x | x | LONINT | RRINT | SAMP | PLLINT |
| | | | x | x | x | x | LDOFF_ PH | LDOFF_ PL | LDOFF_ NH | LDOFF_ NL |

## Table 5. Status (0x01) Register Meaning

| INDEX | NAME | MEANING |
|---|---|---|
| D[21] | FSTINT | Fast Recovery Mode. Issued when the Fast Recovery Mode is engaged (either manually or automatically).<br>Status and Interrupt Clear behavior is defined by CLR_FAST, see MNGR_INT for details. |
| D[20] | DCLOFFINT | DC Lead-Off Detection Interrupt. Indicates that the MAX30004 has determined it is in leads off condition (as selected in CNFG_GEN) for more than 115ms.<br>Remains active as long as the leads-off condition persists, then held until cleared by STATUS read back (32nd SCLK). |

## Table 5. Status (0x01) Register Meaning (continued)

| INDEX | NAME | MEANING |
|---|---|---|
| D[11] | LONINT | Ultra-Low Power (ULP) Leads-On Detection Interrupt. Indicates that the MAX30004 has determined it is in a leads-on condition (as selected in CNFG_GEN).<br>LONINT is asserted whenever EN_ULP_LON[1:0] in register CNFG_GEN is set to either 0b01 or 0b10 to indicate that the ULP leads on detection mode has been enabled. The STATUS register has to be read back once after ULP leads-on detection mode has been activated to clear LONINT and enable leads-on detection.<br>LONINT remains active while the leads-on condition persists, then held until cleared by STATUS read back (32nd SCLK). |
| D[10] | RRINT | R-to-R Detector R Event Interrupt. Issued when the R-to-R detector has identified a new R event.<br>Clear behavior is defined by CLR_RRINT[1:0]; see MNGR_INT for details. |
| D[9] | SAMP | Sample Synchronization Pulse. Issued on the ADC base-rate sampling instant, for use in assisting µC monitoring and synchronizing other peripheral operations and data, generally recommended for use as a dedicated interrupt.<br>Frequency is selected by SAMP_IT[1:0], see MNGR_INT for details.<br>Clear behavior is defined by CLR_SAMP, see MNGR_INT for details. |
| D[8] | PLLINT | PLL Unlocked Interrupt. Indicates that the PLL has not yet achieved or has lost its phase lock.<br>PLLINT will only be asserted when the PLL is powered up and active (Channel enabled).<br>Remains asserted while the PLL unlocked condition persists, then held until cleared by STATUS read back (32nd SCLK). |
| D[3] | LDOFF_PH | DC Lead Off Detection Detailed Status. Indicates that the MAX30004 has determined (as selected by CNFG_GEN):<br>INP is above the high threshold ($V_{THH}$), INP is below the low threshold ($V_{THL}$),<br>INN is above the high threshold ($V_{THH}$), INN is below the low threshold ($V_{THL}$), respectively.<br>Remains active as long as the leads-off detection is active and the leads-off condition persists, then held until cleared by STATUS read back (32nd SCLK). LDOFF_PH to LDOFF_NL are detailed status bits that are asserted at the same time as DCLOFFINT. |
| D[2] | LDOFF_PL | |
| D[1] | LDOFF_NH | |
| D[0] | LDOFF_NL | |

### EN_INT (0x02) and EN_INT2 (0x03) Registers

EN_INT and EN_INT2 are read/write registers that govern the operation of the INTB output and INT2B output, respectively. The first two bytes indicate which interrupt input terms are included in the interrupt output OR term (ex. a one in an EN_INT register indicates that the corresponding input term is included in the INTB interrupt output OR term). See the STATUS register for detailed descriptions of the interrupt terms. The power-on reset state of all EN_INT terms is 0 (ignored by INT).

EN_INT and EN_INT2 can also be used to mask persistent interrupt conditions in order to perform other interrupt-driven operations until the persistent conditions are resolved.

INTB_TYPE[1:0] allows the user to select between a CMOS or an open-drain NMOS mode INTB output. If using open-drain mode, an option for an internal 125kΩ pullup resistor is also offered.

All INTB and INT2B types are active-low (INTB low indicates the device requires servicing by the µC); however, the open-drain mode allows the INTB line to be shared with other devices in a wired-or configuration.

In general, it is suggested that INT2B be used to support specialized/dedicated interrupts of use in specific applications, such as the self-clearing versions of SAMP or RRINT.

## Table 6. EN_INT (0x02) and EN_INT2 (0x03) Register Maps

| REG | NAME | R/W | 23 / 15 / 7 | 22 / 14 / 6 | 21 / 13 / 5 | 20 / 12 / 4 | 19 / 11 / 3 | 18 / 10 / 2 | 17 / 9 / 1 | 16 / 8 / 0 |
|---|---|---|---|---|---|---|---|---|---|---|
| 0x02<br>0x03 | EN_INT<br>EN_INT2 | R/W | x | x | EN_<br>FSTINT | EN_DCL<br>OFFINT | x | x | x | x |
| | | | x | x | x | x | EN_<br>LONINT | EN_<br>RRINT | EN_<br>SAMP | EN_<br>PLLINT |
| | | | x | x | x | x | x | x | INTB_TYPE[1:0] | |

## Table 7. EN_INT (0x02 and 0x03) Register Meaning

| INDEX | NAME | DEFAULT | FUNCTION |
|---|---|---|---|
| D[23:20]<br>D[11:8] | EN_FSTINT<br>EN_DCLOFFINT<br>EN_LONINT<br>EN_RRINT<br>EN_SAMP<br>EN_PLLINT | 0x0000 | Interrupt Enables for interrupt terms in STATUS[23:8]<br>0 = Individual interrupt term is not included in the interrupt OR term<br>1 = Individual interrupt term is included in the interrupt OR term |
| D[1:0] | INTB_TYPE[1:0] | 11 | INTB Port Type (EN_INT Selections)<br>00 = Disabled (Three-state)<br>01 = CMOS Driver<br>10 = Open-Drain NMOS Driver<br>11 = Open-Drain NMOS Driver with Internal 125kΩ Pullup Resistance |
| | | 11 | INT2B Port Type (EN_INT2 Selections)<br>00 = Disabled (three-state)<br>01 = CMOS Driver<br>10 = Open-Drain nMOS Driver<br>11 = Open-Drain nMOS Driver with Internal 125kΩ Pullup Resistance |

### MNGR_INT (0x04)

MNGR_INT is a read/write register that manages the operation of the configurable interrupt bits. Finally, this register contains the configuration bits supporting the sample synchronization pulse (SAMP) and RTOR heart rate detection interrupt (RRINT).

## Table 8. MNGR_INT (0x04) Register Map

| REG | NAME | R/W | 23 / 15 / 7 | 22 / 14 / 6 | 21 / 13 / 5 | 20 / 12 / 4 | 19 / 11 / 3 | 18 / 10 / 2 | 17 / 9 / 1 | 16 / 8 / 0 |
|---|---|---|---|---|---|---|---|---|---|---|
| 0x04 | MNGR_INT | R/W | x | x | x | x | x | x | x | x |
| | | | x | x | x | x | x | x | x | x |
| | | | x | CLR_FAST | CLR_RRINT[1:0] | | x | CLR_SAMP | SAMP_IT[1:0] | |

## Table 9. MNGR_INT (0x04) Register Functionality

| INDEX | NAME | DEFAULT | FUNCTION |
|---|---|---|---|
| D[6] | CLR_FAST | 0 | FAST MODE Interrupt Clear Behavior:<br>0 = FSTINT remains active until the FAST mode is disengaged (manually or automatically), then held until cleared by STATUS read back (32nd SCLK).<br>1 = FSTINT remains active until cleared by STATUS read back (32nd SCLK), even if the MAX30004 remains in FAST recovery mode. Once cleared, FSTINT will not be re-asserted until FAST mode is exited and re-entered, either manually or automatically. |
| D[5:4] | CLR_RRINT[1:0] | 00 | RTOR R Detect Interrupt (RRINT) Clear Behavior:<br>00 = Clear RRINT on STATUS Register Read Back<br>01 = Clear RRINT on RTOR Register Read Back<br>10 = Self-Clear RRINT after one data rate cycle, approximately 2ms to 8ms<br>11 = Reserved. Do not use. |
| D[2] | CLR_SAMP | 1 | Sample Synchronization Pulse (SAMP) Clear Behavior:<br>0 = Clear SAMP on STATUS Register Read Back (recommended for debug/evaluation only).<br>1 = Self-clear SAMP after approximately one-fourth of one data rate cycle. |
| D[1:0] | SAMP_IT[1:0] | 00 | Sample Synchronization Pulse (SAMP) Frequency<br>00 = issued every sample instant<br>01 = issued every 2nd sample instant<br>10 = issued every 4th sample instant<br>11 = issued every 16th sample instant |

### MNGR_DYN (0x05)

MNGR_DYN is a read/write register that manages the settings of any general/dynamic modes within the device. The Fast Recovery modes and thresholds are managed here. Unlike many CNFG registers, changes to dynamic modes do not require a RESTART operation.

### SW_RST (0x08)

SW_RST (Software Reset) is a write-only register/command that resets the MAX30004 to its original default conditions at the end of the SPI SW_RST transaction (i.e. the 32nd SCLK rising edge). Execution occurs only if DIN[23:0] = 0x000000. The effect of a SW_RST is identical to power-cycling the device.

## Table 10. MNGR_DYN (0x05) Register Map

| REG | NAME | R/W | 23 / 15 / 7 | 22 / 14 / 6 | 21 / 13 / 5 | 20 / 12 / 4 | 19 / 11 / 3 | 18 / 10 / 2 | 17 / 9 / 1 | 16 / 8 / 0 |
|---|---|---|---|---|---|---|---|---|---|---|
| 0x05 | MNGR_DYN | R/W | FAST[1:0] | | FAST_TH[5:0] | | | | | |
| | | | x | x | x | x | x | x | x | x |
| | | | x | x | x | x | x | x | x | x |

## Table 11. MNGR_DYN (0x05) Register Functionality

| INDEX | NAME | DEFAULT | FUNCTION |
|---|---|---|---|
| D[23:22] | FAST[1:0] | 00 | Fast Recovery Mode Selection (High-Pass Filter Bypass):<br>00 = Normal Mode (Fast Recovery Mode Disabled)<br>01 = Manual Fast Recovery Mode Enable (remains active until disabled)<br>10 = Automatic Fast Recovery Mode Enable (Fast Recovery automatically activated when/while outputs are saturated, using FAST_TH).<br>11 = Reserved. Do not use. |
| D[21:16] | FAST_TH[5:0] | 0x3F | Automatic Fast Recovery Threshold:<br>If FAST[1:0] = 10 and the output of a measurement exceeds the symmetric thresholds defined by 2048*FAST_TH for more than 125ms, the Fast Recovery mode will be automatically engaged and remain active for 500ms.<br>For example, the default value (FAST_TH = 0x3F) corresponds to an output upper threshold of 0x1F800, and an output lower threshold of 0x20800. |

## Table 12. SW_RST (080x) Register Map

| REG | NAME | R/W | 23 / 15 / 7 | 22 / 14 / 6 | 21 / 13 / 5 | 20 / 12 / 4 | 19 / 11 / 3 | 18 / 10 / 2 | 17 / 9 / 1 | 16 / 8 / 0 |
|---|---|---|---|---|---|---|---|---|---|---|
| 0x08 | SW_RST | R/W | D[23:16] = 0x00 | | | | | | | |
| | | | D[15:8] = 0x00 | | | | | | | |
| | | | D[7:0] = 0x00 | | | | | | | |

### RESTART (0x09)

RESTART (Restart) is a write-only register/command that begins new RTOR operations and recording, beginning on the internal MSTR clock edge following the end of the SPI RESTART transaction (i.e., the 32nd SCLK rising edge). Execution occurs only if DIN[23:0] = 0x000000. In addition to restarting the operations of any active R-to-R circuitry, RESTART also resets and clears the DSP filters (to midscale), allowing the user to effectively set the “Time Zero” for the RTOR data. No configuration settings are impacted. For best results, users should wait until the PLL has achieved lock before restarting if the CNFG_GEN settings have been altered.

Once the device is initially powered up, it needs to be fully configured prior to launching recording operations. Likewise, anytime a change to CNFG_GEN or CNFG_CH registers are made discontinuities in the RTOR record may occur. The RESTART command provides a means to restart operations cleanly following any such disturbances.

### RTOR_RST (0x0A)

RTOR_RST is a write-only register/command that begins a new data recording by resetting the memories and resuming the record with the next available data. Execution occurs only if DIN[23:0] = 0x000000. Unlike the RESTART command, the operations of any active data and R-to-R circuitry are not impacted by RTOR_RST, so, therefore, no settling/recovery transients apply.

## Table 13. RESTART (0x09) Register Map

| REG | NAME | R/W | 23 / 15 / 7 | 22 / 14 / 6 | 21 / 13 / 5 | 20 / 12 / 4 | 19 / 11 / 3 | 18 / 10 / 2 | 17 / 9 / 1 | 16 / 8 / 0 |
|---|---|---|---|---|---|---|---|---|---|---|
| 0x09 | RESTART | R/W | D[23:16] = 0x00 | | | | | | | |
| | | | D[15:8] = 0x00 | | | | | | | |
| | | | D[7:0] = 0x00 | | | | | | | |

## Table 14. RTOR_RST (0x0A) Register Map

| REG | NAME | R/W | 23 / 15 / 7 | 22 / 14 / 6 | 21 / 13 / 5 | 20 / 12 / 4 | 19 / 11 / 3 | 18 / 10 / 2 | 17 / 9 / 1 | 16 / 8 / 0 |
|---|---|---|---|---|---|---|---|---|---|---|
| 0x0A | RTOR_RST | R/W | D[23:16] = 0x00 | | | | | | | |
| | | | D[15:8] = 0x00 | | | | | | | |
| | | | D[7:0] = 0x00 | | | | | | | |

### INFO (0x0F)

INFO is a read-only register that provides information about the MAX30004. The first nibble contains an alternating bit pattern to aide in interface verification. The second nibble contains the revision ID. The third nibble includes part ID information. The final 3 nibbles contain a serial number for Maxim internal use—note that individual units are not given unique serial numbers, and these bits should not be used as serial numbers for end products, though they may be useful during initial development efforts.

**Note:** Due to internal initialization procedures, this command will not read-back valid data if it is the first command executed following either a power-cycle event, or a SW_RST event.

### CNFG_GEN (0x10)

CNFG_GEN is a read/write register which governs general settings, most significantly the master clock rate for all internal timing operations. Anytime a change to CNFG_GEN is made, there may be discontinuities in the data record may occur. The RESTART command can be used to restore internal synchronization resulting from configuration changes. Note when EN_CH is logic-low, the device is in one of two ultra-low power modes (determined by EN_ULP_LON).

Table 19 shows the data rates that can be realized with various setting of FMSTR, along with RATE configuration bits available in the CNFG_CH register.

## Table 15. INFO (0x0F) Register Map

| REG | NAME | R/W | 23 / 15 / 7 | 22 / 14 / 6 | 21 / 13 / 5 | 20 / 12 / 4 | 19 / 11 / 3 | 18 / 10 / 2 | 17 / 9 / 1 | 16 / 8 / 0 |
|---|---|---|---|---|---|---|---|---|---|---|
| 0x0F | INFO | R/W | 0 | 1 | 0 | 1 | REV_ID[3:0] | | | |
| | | | x | X | 0 | 0 | x | x | x | x |
| | | | x | x | x | x | x | x | x | x |

## Table 16. INFO (0x0F) Register Meaning

| INDEX | NAME | MEANING |
|---|---|---|
| D[19:16] | REV_ID[3:0] | Revision ID |

## Table 17. CNFG_GEN (0x10) Register Map

| REG | NAME | R/W | 23 / 15 / 7 | 22 / 14 / 6 | 21 / 13 / 5 | 20 / 12 / 4 | 19 / 11 / 3 | 18 / 10 / 2 | 17 / 9 / 1 | 16 / 8 / 0 |
|---|---|---|---|---|---|---|---|---|---|---|
| 0x10 | CNFG_GEN | R/W | EN_ULP_LON[1:0] | | FMSTR[1:0] | | EN_RTOR | x | x | x |
| | | | x | x | EN_DCLOFF[1:0] | | IPOL | IMAG[2:0] | | |
| | | | VTH[1:0] | | EN_RBIAS[1:0] | | RBIASV[1:0] | | RBIASP | RBIASN |

## Table 18. CNFG_GEN (0x10) Register Functionality

| INDEX | NAME | DEFAULT | FUNCTION |
|---|---|---|---|
| D[23:22] | EN_ULP_LON [1:0] | 00 | Ultra-Low Power Lead-On Detection Enable<br>00 = ULP Lead-On Detection disabled<br>01 = ULP Lead-On Detection enabled<br>10 = Reserved. Do not use.<br>11 = Reserved. Do not use.<br>ULP mode is only active when the RTOR channel is powered down/disabled. |
| D[21:20] | FMSTR[1:0] | 00 | Master Clock Frequency. Selects the Master Clock Frequency (FMSTR), and Timing Resolution ($T_{RES}$). These are generated from FCLK, which is always 32.768Khz.<br>00 = $F_{MSTR}$ = 32768Hz, $T_{RES}$ = 15.26µs (512Hz data progressions)<br>01 = $F_{MSTR}$ = 32000Hz, $T_{RES}$ = 15.63µs (500Hz data progressions)<br>10 = $F_{MSTR}$ = 32000Hz, $T_{RES}$ = 15.63µs (200Hz data progressions)<br>11 = $F_{MSTR}$ = 31968.78Hz, $T_{RES}$ = 15.64µs (199.8049Hz data progressions) |
| D[19] | EN_CH | 0 | Channel Enable<br>0 = Channel disabled<br>1 = Channel enabled<br>Note: The channel must be enabled to allow R-to-R operation. |
| D[13:12] | EN_DCLOFF | 00 | DC Lead-Off Detection Enable<br>00 = DC Lead-Off Detection disabled<br>01 = DCLOFF Detection applied to the INP/INN pins<br>10 = Reserved. Do not use.<br>11 = Reserved. Do not use.<br>DC Method, requires active selected channel, enables DCLOFF interrupt and status bit behavior.<br>Uses current sources and comparator thresholds set below. |
| D[11] | DCLOFF_ IPOL | 0 | DC Lead-Off Current Polarity (if current sources are enabled/connected)<br>0 = INP - Pullup INN – Pulldown<br>1 = INP - Pulldown INN – Pullup |
| D[10:8] | DCLOFF_ IMAG[2:0] | 000 | DC Lead-Off Current Magnitude Selection<br>000 = 0nA (Disable and Disconnect Current Sources)<br>001 = 5nA<br>010 = 10nA<br>011 = 20nA<br>100 = 50nA<br>101 = 100nA<br>110 = Reserved. Do not use.<br>111 = Reserved. Do not use. |
| D[7:6] | DCLOFF_ VTH[1:0] | 00 | DC Lead-Off Voltage Threshold Selection<br>00 = $V_{MID}$ ± 300mV<br>01 = $V_{MID}$ ± 400mV<br>10 = $V_{MID}$ ± 450mV<br>11 = $V_{MID}$ ± 500mV |
| D[5:4] | EN_RBIAS[1:0] | 00 | Enable and Select Resistive Lead Bias Mode<br>00 = Resistive Bias disabled<br>01 = Resistive Bias enabled if EN_CH is also enabled<br>10 = Reserved. Do not use.<br>11 = Reserved. Do not use.<br>If EN_CH is not asserted at the same time as prior to EN_RBIAS[1:0] being set to 01, then EN_RBIAS[1:0] will remain set to 00. |

## Table 18. CNFG_GEN (0x10) Register Functionality (continued)

| INDEX | NAME | DEFAULT | FUNCTION |
|---|---|---|---|
| D[3:2] | RBIASV[1:0] | 01 | Resistive Bias Mode Value Selection<br>00 = $R_{BIAS}$ = 50MΩ<br>01 = $R_{BIAS}$ = 100MΩ<br>10 = $R_{BIAS}$ = 200MΩ<br>11 = Reserved. Do not use. |
| D[1] | RBIASP | 0 | Enables Resistive Bias on Positive Input<br>0 = INP is not resistively connected to $V_{MID}$<br>1 = INP is connected to $V_{MID}$ through a resistor (selected by RBIASV). |
| D[0] | RBIASN | 0 | Enables Resistive Bias on Negative Input<br>0 = INN is not resistively connected to $V_{MID}$<br>1 = INN is connected to $V_{MID}$ through a resistor (selected by RBIASV). |

## Table 19. Master Frequency Summary Table

| FMSTR [1:0] | MASTER FREQUENCY ($f_{MSTR}$) (Hz) | DATA RATES & RELATED TIMING (RATE SELECTIONS) | | CALIBRATION TIMING RESOLUTION (CAL_RES) (µs) |
|---|---|---|---|---|
| | | CHANNEL | RTOR | |
| | | DATA RATE (sps) | TIMING RESOLUTION (RTOR_RES) (ms) | |
| 00 | 32768 | 00 = 512<br>01 = 256<br>10 = 128 | 7.8125 | 30.52 |
| 01 | 32000 | 00 = 500<br>01 = 250<br>10 = 125 | 8.000 | 31.25 |
| 10 | 32000 | 10 = 200 | 8.000 | 31.25 |
| 11 | 31968.78 | 10 =199.8049 | 8.008 | 31.28 |

### CNFG_MUX (0x14)

CNFG_MUX is a read/write register which configures the operation, settings, and functionality of the Input Multiplexer.

### CNFG_CH (0x15)

CNFG_CH is a read/write register that configures the operation, settings, and functionality of the Biopotential channel. Anytime a change to CNFG_CH is made, there may be discontinuities in the data record.

## Table 20. CNFG_EMUX (0x14) Register Map

| REG | NAME | R/W | 23 / 15 / 7 | 22 / 14 / 6 | 21 / 13 / 5 | 20 / 12 / 4 | 19 / 11 / 3 | 18 / 10 / 2 | 17 / 9 / 1 | 16 / 8 / 0 |
|---|---|---|---|---|---|---|---|---|---|---|
| 0x14 | CNFG_ MUX | R/W | POL | x | OPENP | OPENN | x | x | x | x |
| | | | x | x | x | x | x | x | x | x |
| | | | x | x | x | x | x | x | x | x |

## Table 21. CNFG_MUX (0x14) Register Functionality

| INDEX | NAME | DEFAULT | FUNCTION |
|---|---|---|---|
| D[23] | POL | 0 | Input Polarity Selection<br>0 = Non-Inverted<br>1 = Inverted |
| D[21] | OPENP | 1 | Open the INP Input Switch (most often used for testing and calibration studies)<br>0 = INP is internally connected to the AFE Channel<br>1 = INP is internally isolated from the AFE Channel |
| D[20] | OPENN | 1 | Open the INN Input Switch (most often used for testing and calibration studies)<br>0 = INN is internally connected to the AFE Channel<br>1 = INN is internally isolated from the AFE Channel |

## Table 22. CNFG_CH (0x15) Register Map

| REG | NAME | R/W | 23 / 15 / 7 | 22 / 14 / 6 | 21 / 13 / 5 | 20 / 12 / 4 | 19 / 11 / 3 | 18 / 10 / 2 | 17 / 9 / 1 | 16 / 8 / 0 |
|---|---|---|---|---|---|---|---|---|---|---|
| 0x15 | CNFG_ CH | R/W | RATE[1:0] | | x | x | x | x | GAIN[1:0] | |
| | | | x | DHPF | DLPF[1:0] | | x | x | x | x |
| | | | x | x | x | x | x | x | x | x |

## Table 23. CNFG_CH (0x15) Register Functionality

| INDEX | NAME | DEFAULT | FUNCTION |
|---|---|---|---|
| D[23:22] | RATE[1:0] | 10 | Sample Data Rate (also dependent on $F_{MSTR}$ selection, see CNFG_GEN Table 24): |
| | | | FMSTR = 00: $f_{MSTR}$ = 32768Hz, $t_{RES}$ = 15.26µs (512Hz data progressions)<br>00 = 512sps<br>01 = 256sps<br>10 = 128sps<br>11 = Reserved. Do not use. |
| | | | FMSTR = 01: $f_{MSTR}$ = 32000Hz, $t_{RES}$ = 15.63µs (500Hz data progressions)<br>00 = 500sps<br>01 = 250sps<br>10 = 125sps<br>11 = Reserved. Do not use. |
| | | | FMSTR = 10: $f_{MSTR}$ = 32000Hz, $t_{RES}$ = 15.63µs (200Hz data progressions)<br>00 = Reserved. Do not use.<br>01 = Reserved. Do not use.<br>10 = 200sps<br>11 = Reserved. Do not use. |
| | | | FMSTR = 11: $f_{MSTR}$ = 31968Hz, $t_{RES}$ = 15.64µs (199.8Hz data progressions)<br>00 = Reserved. Do not use.<br>01 = Reserved. Do not use.<br>10 = 199.8sps<br>11 = Reserved. Do not use. |
| D[17:16] | GAIN[1:0] | 00 | Gain Setting<br>00 = 20V/V<br>01 = 40V/V<br>10 = 80V/V<br>11 = 160V/V |

## Table 23. CNFG_CH (0x15) Register Functionality (continued)

| INDEX | NAME | DEFAULT | FUNCTION |
|---|---|---|---|
| D[14] | DHPF | 1 | Digital High-Pass Filter Cutoff Frequency<br>0 = Bypass (DC)<br>1 = 0.50Hz |
| D[13:12] | DLPF[1:0] | 01 | Digital Low-Pass Filter Cutoff Frequency<br>00 = Bypass (Decimation only, no FIR filter applied)<br>01 = approximately 40Hz<br>10 = approximately 100Hz (Available for 512, 256, 500, and 250sps data Rate selections only)<br>11 = approximately 150Hz (Available for 512 and 500sps data rate selections only)<br>**Note:** See Table 24. If an unsupported DLPF setting is specified, the 40Hz setting (DLPF[1:0] = 01) will be used internally; the CNFG_CH register will continue to hold the value as written, but return the effective internal value when read back. |

## Table 24. Supported RATE and DLPF Options

| CNFG_GEN FMSTR[1:0] | RATE[1:0] SAMPLE RATE (sps) | DLPF[1:0]/DIGITAL LPF CUTOFF | | | |
|---|---|---|---|---|---|
| | | 00 | 01 (Hz) | 10 (Hz) | 11 (Hz) |
| 00 = 32768Hz | 00 = 512 | Bypass | 40.96 | 102.4 | 153.6 |
| | 01 = 256 | Bypass | 40.96 | 102.4 | **40.96** |
| | 10 = 128 | Bypass | 40.96 | **40.96** | **40.96** |
| 01 = 32000Hz | 00 = 500 | Bypass | 40.00 | 100.0 | 150.0 |
| | 01 = 250 | Bypass | 40.00 | 100.0 | **40.00** |
| | 10 = 125 | Bypass | 40.00 | **40.00** | **40.00** |
| 10 = 32000Hz | 10 = 200 | Bypass | 40.00 | **40.00** | **40.00** |
| 11 = 31968Hz | 10 = 199.8 | Bypass | 39.96 | **39.96** | **39.96** |

***Note:*** *Combinations shown in grey are unsupported and will be internally mapped to the default settings shown.*

### CNFG_RTOR1 and CNFG_RTOR2 (0x1D & 0x1E)

CNFG_RTOR is a two-part read/write register that configures the operation, settings, and function of the RTOR heart rate detection block. The first register contains algorithmic voltage gain and threshold parameters, the second contains algorithmic timing parameters.

### RTOR Interval Memory Register (1 Word x 24 Bits)

The RTOR Interval (RTOR) memory register is a single read-only register consisting of 14 bits of timing interval information, left justified (and 8 unused bits, set to zero).

The RTOR register stores the time interval between the last two R events, as identified by the RTOR detection circuitry, which operates on the channel output data. Each LSB in the RTOR register is approximately equal to 8ms (CNFG_GEN for exact figures). The resulting 14-bit storage interval can thus be approximately 130 seconds in length, again depending on device settings.

Each time the RTOR detector identifies a new R event, the RTOR register is updated, and the RRINT interrupt term is asserted (see *STATUS (0x01) Register* for details).

Users wishing to log heart rate based on RTOR register data should set CLR_RRINT equals 01 in the MNGR_INT register. This will clear the RRINT interrupt term after the RTOR register has been read back, preparing the device for identification of the next RTOR interval.

Users wishing to log heart rate based on the time elapsed between RRINT assertions using the µC to keep track of the time base (and ignoring the RTOR register data) have two choices for interrupt management. If CLR_RRINT equals 00 in the MNGR_INT register, the RRINT interrupt term will clear after each STATUS register read back, preparing the device for identification of the next RTOR interval. If CLR_RRINT equals 10 in the MNGR_INT register, the RRINT interrupt term will self-clear after each one full data cycle has passed, preparing the device for identification of the next RTOR interval (this mode is recommended only if using the INT2B as a dedicated heart rate indicator).

If the RTOR detector reaches an overflow state after several minutes without detection of an R event, the counter will simply roll over, and the lack of the RRINT activity on the dedicated INT2B line will inform the µC that no RTOR activity was detected.

## Table 25. CNFG_RTOR and CNFG_RTOR2 (0x1D and 0x1E) Register Maps

| REG | NAME | R/W | 23 / 15 / 7 | 22 / 14 / 6 | 21 / 13 / 5 | 20 / 12 / 4 | 19 / 11 / 3 | 18 / 10 / 2 | 17 / 9 / 1 | 16 / 8 / 0 |
|---|---|---|---|---|---|---|---|---|---|---|
| 0x1D | CNFG_ RTOR1 | R/W | WNDW[3:0] | | | | GAIN[3:0] | | | |
| | | | EN_ RTOR | x | PAVG[1:0] | | PTSF[3:0] | | | |
| | | | x | x | x | x | x | x | x | x |
| 0x1E | CNFG_ RTOR2 | R/W | x | x | HOFF[5:0] | | | | | |
| | | | x | x | RAVG[1:0] | | x | RHSF[2:0] | | |
| | | | x | x | x | x | x | x | x | x |

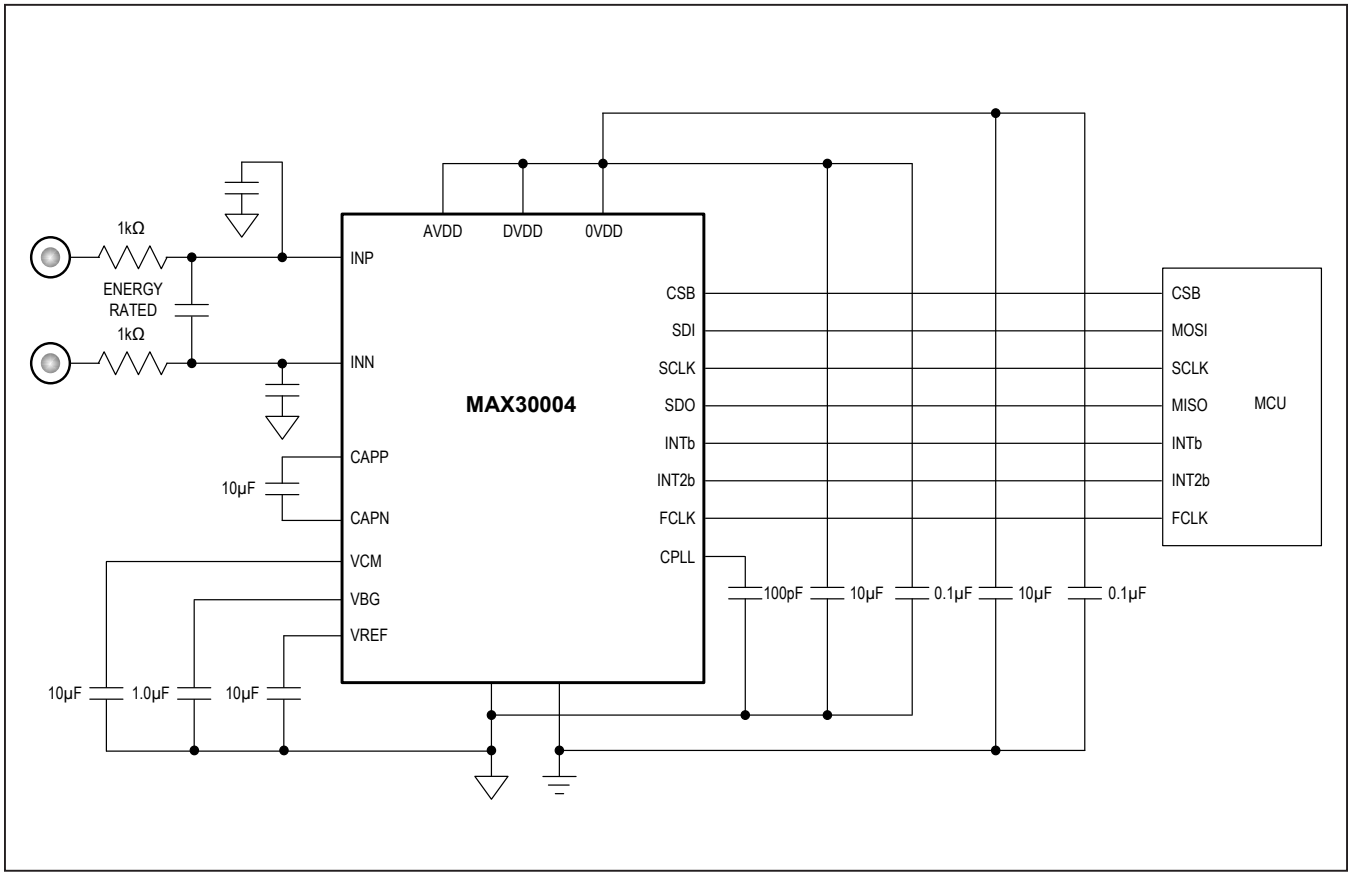
## Table 26. CNFG_RTOR and CNFG_RTOR2 (0x1D and 0x1E) Register Functionality

| INDEX | NAME | DEFAULT | FUNCTION |
|---|---|---|---|
| | | | **CNFG_RTOR1 (0x1D)** |
| D[23:20] | WNDW[3:0] | 0011 | This is the width of the averaging window, which adjusts the algorithm sensitivity to the width of the QRS complex.<br>R-to-R Window Averaging (Window Width = RTOR_WNDW[3:0]*8ms)<br>0000 = 6<br>0001 = 8<br>0010 = 10<br>0011 = 12 (default = 96ms)<br>0100 = 14<br>0101 = 16<br>0110 = 18<br>0111 = 20<br>1000 = 22<br>1001 = 24<br>1010 = 26<br>1011 = 28<br>1100 = Reserved. Do not use.<br>1101 = Reserved. Do not use.<br>1110 = Reserved. Do not use.<br>1111 = Reserved. Do not use. |
| D[19:16] | GAIN[3:0] | 1111 | R-to-R Gain (where Gain = 2^GAIN[3:0], plus an auto-scale option). This is used to maximize the dynamic range of the algorithm.<br>0000 = 1 1000 = 256<br>0001 = 2 1001 = 512<br>0010 = 4 1010 = 1024<br>0011 = 8 1011 = 2048<br>0100 = 16 1100 = 4096<br>0101 = 32 1101 = 8192<br>0110 = 64 1110 = 16384<br>0111 = 128 1111 = Auto-Scale (default)<br>In Auto-Scale mode, the initial gain is set to 64. |
| D[15] | EN_RTOR | 0 | RTOR Detection Enable<br>0 = RTOR Detection disabled<br>1 = RTOR Detection enabled if EN_CH is also enabled. |

## Table 26. CNFG_RTOR and CNFG_RTOR2 (0x1D and 0x1E) Register Functionality (continued)

| INDEX | NAME | DEFAULT | FUNCTION |
|---|---|---|---|
| D[13:12] | PAVG[1:0] | 10 | R-to-R Peak Averaging Weight Factor<br>This is the weighting factor for the current RTOR peak observation vs. past peak observations when determining peak thresholds. Lower numbers weight current peaks more heavily.<br>00 = 2<br>01 = 4<br>10 = 8 (default)<br>11 = 16<br>Peak_Average(n) =<br>[Peak(n) + (RTOR_PAVG-1) x Peak_Average(n-1)]/RTOR_PAVG. |
| D[11:8] | PTSF[3:0] | 0011 | R-to-R Peak Threshold Scaling Factor<br>This is the fraction of the Peak Average value used in the Threshold computation. Values of 1/16 to 16/16 are selected directly by (RTOR_PTSF[3:0]+1)/16, default is 4/16. |
| CNFG_ RTOR2 (0x1E)D [21:16] | HOFF[5:0] | 10_0000 | R-to-R Minimum Hold Off<br>This sets the absolute minimum interval used for the static portion of the Hold Off criteria.<br>Values of 0 to 63 are supported, default is 32<br>$t_{HOLD_OFF_MIN}$ = HOFF[5:0] * $t_{RTOR}$, where $t_{RTOR}$ is ~8ms, as determined by FMSTR[1:0] in the CNFG_GEN register.<br>(representing approximately ¼ second).<br>The R-to-R Hold Off qualification interval is<br>$t_{Hold_Off}$ = MAX($t_{Hold_Off_Min}$, $t_{Hold_Off_Dyn}$) (see below). |
| D[13:12] | RAVG[1:0] | 10 | R-to-R Interval Averaging Weight Factor<br>This is the weighting factor for the current RtoR interval observation vs. the past interval observations when determining dynamic holdoff criteria. Lower numbers weight current intervals more heavily.<br>00 = 2<br>01 = 4<br>10 = 8 (default)<br>11 = 16<br>Interval_Average(n) = [Interval(n) + (RAVG-1) x Interval_Average(n-1)]/RAVG. |
| D[10:8] | RHSF[2:0] | 100 | R-to-R Interval Hold Off Scaling Factor<br>This is the fraction of the RtoR average interval used for the dynamic portion of the holdoff criteria ($t_{HOLD_OFFDYN}$).<br>Values of 0/8 to 7/8 are selected directly by RTOR_RHSF[3:0]/8, default is 4/8.<br>If 000 (0/8) is selected, then no dynamic factor is used and the holdoff criteria is determined by HOFF[5:0] only (see above). |

## Typical Application Circuit



## Application Diagrams

### Two Electrode Heart Rate Fitness

See Figure 8 for an example of a fitness monitoring configuration.

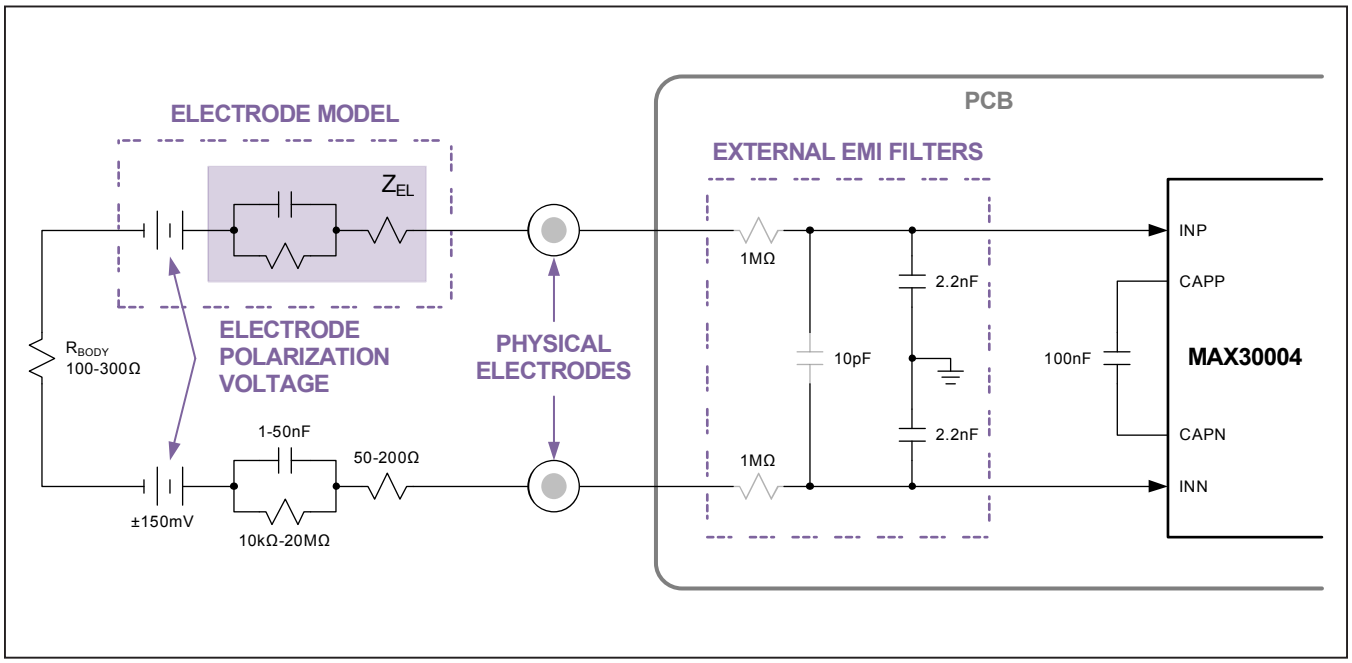



*Figure 8. Two Electrode Heart Rate Monitoring for Fitness*

## Ordering Information

| PART | TEMP RANGE | PIN-PACKAGE |
|---|---|---|
| MAX30004CTI+T* | 0°C to +70°C | 28 TQFN-EP** |
| MAX30004CWV+ | 0°C to +70°C | 30 WLP |

*+Denotes a lead(Pb)-free/RoHS-compliant package.*
*T = Tape and reel.*
**Future product-contact factory for availability.*
***EP = Exposed pad.*

## Chip Information

PROCESS: CMOS

## Package Information

For the latest package outline information and land patterns (footprints), go to **www.maximintegrated.com/packages**. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

| PACKAGE TYPE | PACKAGE CODE | OUTLINE NO. | LAND PATTERN NO. |
|---|---|---|---|
| 28 TQFN | T2855+8 | 21-0140 | 90-0028 |
| 30 WLP | W302L2+1 | 21-100074 | Refer to Application Note 1891 |

## Revision History

| REVISION NUMBER | REVISION DATE | DESCRIPTION | PAGES CHANGED |
|---|---|---|---|
| 0 | 12/16 | Initial release | — |

For pricing, delivery, and ordering information, please contact Maxim Direct at 1-888-629-4642, or visit Maxim Integrated's website at www.maximintegrated.com.

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