

## 300 to 930MHz Receiver Evaluation Board Description

#### **Features**

Dual RF input for afficing space and frequency diversity, LIVA cascading of differential feeding
Fully integrated PLL-based synthesizer
2 <sup>nd</sup> mixer with image rejection
Reception of ASK or FSK modulated signals
Wide operating voltage and temperature ranges
Very low standby current consumption
Low operating current consumption
External IF filters 455kHz or 10.7MHz
Internal FSK demodulator
Average or peak detection data slicer mode
RSSI output with high dynamic range for RF level indication
Output noise cancellation filter
MCU clock output
High over-all frequency accuracy

Dual DE input for entenne anges and frequency diversity. LNA assessing or differential feeding

### **Ordering Information**

Part No. (see paragraph 4)

EVB71120-315-C EVB71120-868-C EVB71120-915-C

**Note 1:** Peak detection mode, IF2 selection = 10.7MHz is default population.

## **Application Examples**

General digital and analog RF receivers
at 300 to 930MHz
Tire pressure monitoring systems (TPMS)
Remote keyless entry (RKE)
Low power telemetry systems
Alarm and security systems
Active RFID tags
Remote controls
Garage door openers
Home and building automation

## **General Description**

The MLX71120 is a multi-band, single-channel RF receiver based on a double-conversion super-heterodyne architecture. It can receive FSK and ASK modulated signals. The IC is designed for general purpose applications for example in the European bands at 433MHz and 868MHz or for similar applications in North America or Asia, e.g. at 315MHz or 915MHz. It is also well-suited for narrow-band applications according to the ARIB STD-T67 standard in the frequency range 426MHz to 470MHz.

The receiver's extended temperature and supply voltage ranges make the device a perfect fit for automotive or similar applications where harsh environmental conditions are expected.



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## 1 Theory of Operation

#### 1.1 General

The MLX71120 receiver architecture is based on a double-conversion super-heterodyne approach. The two LO signals are derived from an on-chip integer-N PLL frequency synthesizer. The PLL reference frequency is derived from a crystal (XTAL). As the first intermediate frequency (IF1) is very high, a reasonably high degree of image rejection is provided even without using an RF front-end filter. At applications asking for very high image rejections, cost-efficient RF front-end filtering can be realized by using a SAW filter in front of the LNA. The second mixer MIX2 is an image-reject mixer.

The receiver signal chain is setup by one (or two) low noise amplifier(s) (LNA1, LNA2), two down-conversion mixers (MIX1, MIX2) and an external IF filter with an on-chip amplifier (IFA). By choosing the required modulation via an FSK/ASK switch (at pin MODSEL), either the on-chip FSK demodulator (FSK DEMOD) or the RSSI-based ASK detector is selected. A second order data filter (OA1) and a data slicer (OA2) follow the demodulator. The data slicer threshold can be generated from the mean-value of the data stream or by means of the positive and negative peak detectors (PKDET+/-). A digital post-processing of the sliced data signal can be performed by a noise filter (NF) building block.

The dual LNA configuration can be used for antenna space diversity or antenna frequency diversity or to setup an LNA cascade (to further improve the input sensitivity). The two LNAs can also be setup to feed the RF signal differentially.

A sequencer circuit (SEQ) controls the timing during start-up. This is to reduce start-up time and to minimize power dissipation.

A clock output, which is a divide-by-8 version of the crystal oscillator signal, can be used to drive a microcontroller. The clock output is open drain and gets activated through a load connected to positive supply.

#### 1.2 Technical Data Overview

Input frequency ranges: 300 to 470MHz 610 to 930MHz	Image rejection: 65dB 1 <sup>st</sup> IF (with external RF front-end filter)
Power supply range: 2.1 to 5.5V	25dB 2 <sup>nd</sup> IF (internal image rejection)
Temperature range: -40 to +125°C	Maximum data rate: 50kps RZ (bi-phase) code,
Shutdown current: 50 nA	100kps NRZ
Operating current: 6.5 to 8.1mA	Spurious emission: < -54dBm
Selectable IF2 frequency: 10.7MHz or 455kHz	Linear RSSI range: > 70dB
FSK deviation range: ±10kHz to ±100kHz (WB) ±2kHz to ±10kHz (NB)	Crystal reference frequency: 16 to 27MHz MCU clock frequency: 2.0 to 3.4

☐ Input Sensitivity: at 4kbps NRZ, BER = 3·10 <sup>-3</sup>					
	Frequency	315 MHz	433.92 MHz	868.3 MHz	915 MHz
FSK	wide band 180kHz BW, IF2=10.7MHz $\Delta f = \pm 20$ kHz	-109dBm	-108dBm	-106dBm	-104dBm
rsk	narrow band 20kHz BW, IF2=455kHz $\Delta f = \pm 5$ kHz	-114dBm	-112dBm	-111dBm	-109dBm
ASK	wide band 180kHz BW, IF2=10.7MHz	-113dBm	-113dBm	-111dBm	-109dBm

Note: - Sensitivities given for RF input 1 (without SAW filter)

- Sensitivity for RF input 2 is about 2 to 3dB worse (because of SAW filter loss)

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#### 1.3 Block Diagram

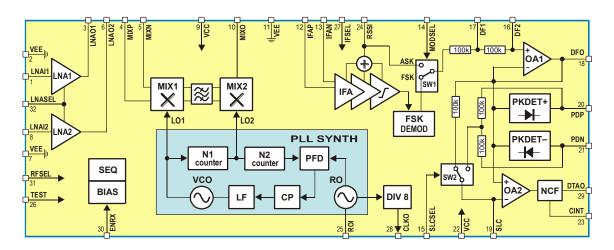


Fig. 1: MLX71120 block diagram

The MLX71120 receiver IC consists of the following building blocks:

- PLL synthesizer (PLL SYNTH) to generate the first and second local oscillator signals LO1 and LO2. The PLL SYNTH consists of a fully integrated voltage-controlled oscillator (VCO), a distributed feedback divider chain (N1, N2), a phase-frequency detector (PFD) a charge pump (CP), a loop filter (LF) and a crystal-based reference oscillator (RO).
- Two low-noise amplifiers (LNA) for high-sensitivity RF signal reception
- First mixer (MIX1) for down-conversion of the RF signal to the first IF (intermediate frequency)
- Second mixer (MIX2) with image rejection for down-conversion from the first to the second IF
- IF amplifier (IFA) to provide a high voltage gain and an RSSI signal output
- FSK demodulator (FSK DEMOD)
- Operational amplifiers OA1 and OA2 for low-pass filtering and data slicing, respectively
- Positive (PKDET+) and negative (PKDET-) peak detectors
- Switches SW1 to select between FSK and ASK as well as SW2 to chose between averaging or peak detection mode.
- Noise cancellation filter (NF)
- Sequencer circuit (SEQ) and biasing (BIAS) circuit
- Clock output (DIV8)

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#### 1.4 Operating Modes

ENRX	Description		
0	Shutdown mode		
1	Receive mode		

Note: ENRX is pulled down internally.

#### 1.5 Frequency Range

Two different receive frequency ranges can be selected by the control signal RFSEL.

RFSEL	Description		
0	Input frequency range 300 to 470MHz		
1	Input frequency range 610 to 930MHz		

#### 1.6 LNA Selection

LNASEL	Description
0	LNA1 active, LNA2 shutdown
Hi-Z	LNA1 and LNA2 active
1	LNA1 shutdown, LNA2 active

Note: Hi-Z state means pin LNASEL is left floating (pin is internally pulled to V<sub>CC</sub>/2 in this case).

#### 1.7 External IF2 Selection

IFSEL	Description
0	IF2 = 455 kHz
1	IF2 = 10.7 MHz

#### 1.8 Demodulation Selection

MODSEL	Description
0	ASK demodulation
1	FSK demodulation

#### 1.9 Data Slicer

SLCSEL	Description		
0	Averaging detection mode		
1	Peak detection mode		

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## 2 Frequency Planning

Because of the double conversion architecture that employs two mixers and two IF signals, there are four different combinations for injecting the LO1 and LO2 signals:

As a result, four different radio frequencies (RFs) could yield one and the same second IF (IF2). Fig. 2 shows this for the case of receiving at  $f_{RF}$ (high-high). In the example of Fig. 2, the image signals at  $f_{RF}$ (low-high) and  $f_{RF}$ (low-low) are suppressed by the bandpass characteristic provided by the RF front-end. The bandpass shape can be achieved either with a SAW filter (featuring just a couple of MHz bandwidth), or by the tank circuits at the LNA input and output (this typically yields 30 to 60MHz bandwidth). In any case, the high value of the first IF (IF1) helps to suppress the image signals at  $f_{RF}$ (low-high) and  $f_{RF}$ (low-low).

The two remaining signals at IF1 resulting from  $f_{RF}(high-high)$  and  $f_{RF}(high-low)$  are entering the second mixer MIX2. This mixer features image rejection with so-called single-sideband (SSB) selection. This means either the upper or lower sideband of IF1 can be selected. In the example of Fig. 2, LO2 high-side injection has been chosen to select the IF2 signal resulting from  $f_{RF}(high-high)$ .

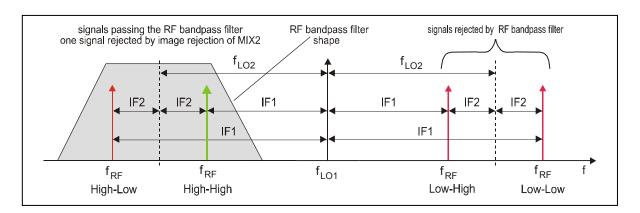


Fig. 2: The four receiving frequencies in a double conversion superhet receiver

It can be seen from the block diagram of Fig. 1 that there is a fixed relationship between the LO signal frequencies ( $f_{LO1}$ ,  $f_{LO2}$ ) and the reference oscillator frequency  $f_{RO}$ .

$$\mathbf{f}_{\text{LO1}} = \mathbf{N}_1 \cdot \mathbf{f}_{\text{LO2}} \qquad \qquad \mathbf{f}_{\text{LO2}} = \mathbf{N}_2 \cdot \mathbf{f}_{\text{RO}}$$

The IF2 frequency can be selected to 455kHz or 10.7MHz via the logic level at the IFSEL control pin. At the same time the output impedance of the 2<sup>nd</sup> mixer at pin MIXO is set according to the IF2 (please refer to pin description for details). Of course, also the operating frequency of the FSK demodulator (FSK DEMOD) is set accordingly.

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### 2.1 Calculation of Frequency Settings

The receiver has two predefined receive frequency plans which can be selected by the RFSEL control pin. Depending on the logic level of RFSEL pin the sideband selection of the second mixer and the counter settings for N1 and N2 are changed accordingly. (see in 1.5)

RFSEL	Injection	f <sub>RFmin</sub> [MHz]	f <sub>RFmax</sub> [MHz]	N <sub>1</sub>	N <sub>2</sub>
0	high-low	300	470	4	6
1	low-high	610	930	2	12

The following table shows the relationships of several internal receiver frequencies for the two input frequency ranges.

f <sub>RF</sub> [MHz]	f <sub>IF1</sub> f <sub>LO1</sub>		f <sub>LO2</sub>	f <sub>RO</sub>	
300 to 470	$\frac{\mathbf{f}_{RF} + \mathbf{N}_{1}\mathbf{f}_{IF2}}{\mathbf{N}_{1} - 1}$	$\frac{N_{1}(f_{RF} + f_{IF2})}{N_{1} - 1}$	$\frac{f_{RF}+f_{IF2}}{N_1-1}$	$\frac{f_{RF} + f_{IF2}}{N_2(N_1 - 1)}$	
610 to 930	$\frac{\mathbf{f}_{RF} - \mathbf{N}_{1} \mathbf{f}_{IF2}}{\mathbf{N}_{1} + 1}$	$\frac{N_1(f_{RF} + f_{IF2})}{N_1 + 1}$	$\frac{f_{RF}+f_{IF2}}{N_1+1}$	$\frac{f_{RF} + f_{IF2}}{N_2(N_1 + 1)}$	

Given IF2 is selectable at either 455kHz or 10.7MHz and the corresponding  $N_1$ ,  $N_2$  counter settings, above equations can be transferred into the following table.

#### IF2=455kHz

f <sub>RF</sub> [MHz]	f <sub>IF1</sub>	f <sub>LO1</sub>	f <sub>LO2</sub>	f <sub>RO</sub>
300 to 470	$\frac{f_{RF} + 1.82MHz}{3}$	$\frac{4(f_{RF} + 0.455M  Hz)}{3}$	$f_{RF} + 0.455MHz$	$\frac{f_{RF} + 0.455MHz}{18}$
610 to 930	$\frac{f_{RF} - 0.91MHz}{3}$	$\frac{2(f_{RF} + 0.455MHz)}{3}$	3	$\frac{f_{RF} + 0.455MHz}{36}$

#### IF2=10.7MHz

f <sub>RF</sub> [MHz]	f <sub>IF1</sub>	f <sub>LO1</sub>	f <sub>LO2</sub>	f <sub>RO</sub>	
300 to 470	$\frac{f_{RF} + 42.8MHz}{3}$	$\frac{4(f_{RF} + 10.7M \text{ Hz})}{3}$	$f_{RF} + 10.7MHz$	$\frac{f_{RF} + 10.7MHz}{18}$	
610 to 930	930 $\frac{f_{RF} - 21.4MHz}{3}$ $\frac{2(f_{RF} + 10.7MHz)}{3}$		3	$\frac{f_{RF} + 10.7MHz}{36}$	

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## 2.2 Standard Frequency Plans

IF2 = 455kHz

f <sub>RF</sub> [MHz]	f <sub>IF1</sub> [MHz]	f <sub>LO1</sub> [MHz]	f <sub>LO2</sub> [MHz]	f <sub>RO</sub> [MHz]
315	105.6067	420.6067	105.1517	17.525277
433.92	145.2467	579.1667	144.7917	24.131944
868.3	289.1300	579.1700	289.5850	24.132083
915	304.6967	610.3033	305.1517	25.429305

IF2 = 10.7MHz

f <sub>RF</sub> [MHz]	f <sub>IF1</sub> [MHz]	f <sub>LO1</sub> [MHz]	f <sub>LO2</sub> [MHz]	f <sub>RO</sub> [MHz]
315	119.2667	434.2667	108.5667	18.094444
433.92	158.0667	592.8267	148.2067	24.701111
868.3	282.3000	586.0000	293.0000	24.416666
915	297.8667	617.1333	308.5667	25.713888

#### 2.3 433/868MHz Frequency Diversity

The receiver's multi-band functionality can be used to operate at two different frequency bands just by changing the logic level at pin RFSEL and without changing the crystal. This feature is applicable for common use of the 433 and 868MHz bands. Below table shows the corresponding frequency plans.

IF2 = 455kHz

RFSEL	f <sub>RF</sub> [MHz]	f <sub>IF1</sub> [MHz]	f <sub>LO1</sub> [MHz]	f <sub>LO2</sub> [MHz]	f <sub>RO</sub> [MHz]	
0	433.9225	145.2483	579.17	144.7925	24 122002	
1	868.3	289.1300	579.17	289.5850	24.132083	



# 3 Dual-Channel Application Circuits for FSK & ASK Reception

#### 3.1 Peak Detector Data Slicer

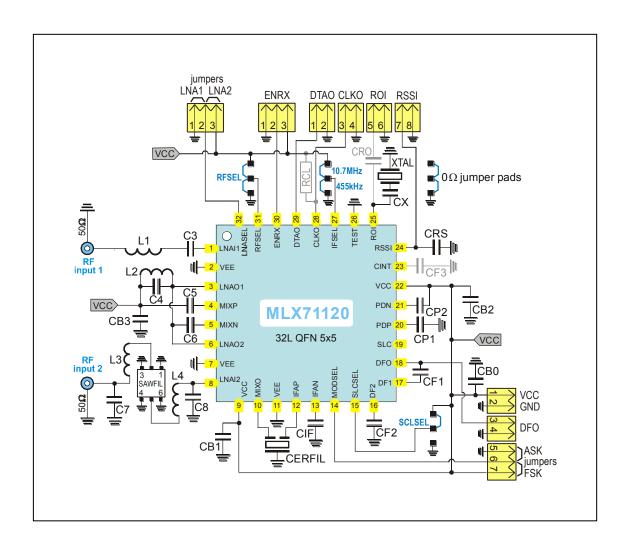


Fig. 3: Circuit schematic



#### 3.1.1 Component Arrangement Top Side (Peak Detection Data Slicer, IF2 = 10.7MHz)

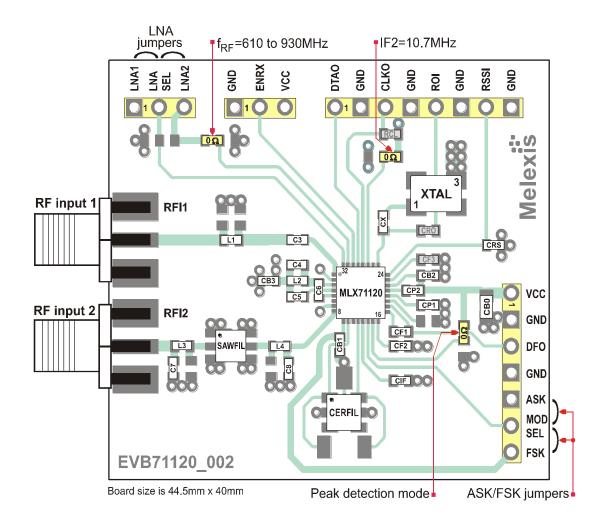


Fig. 4: PCB top-side view



## 3.1.2 Component Arrangement Top Side (Peak Detection Data Slicer, IF2 = 455kHz)

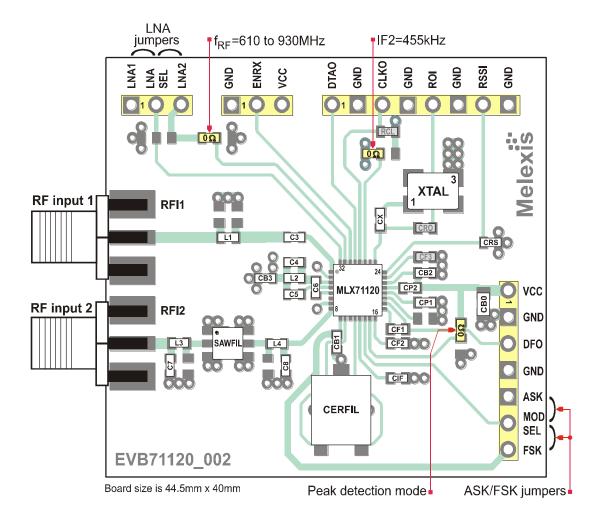


Fig. 5: PCB top-side view



## 3.2 Averaging Data Slicer Configured for Bi-Phase Codes

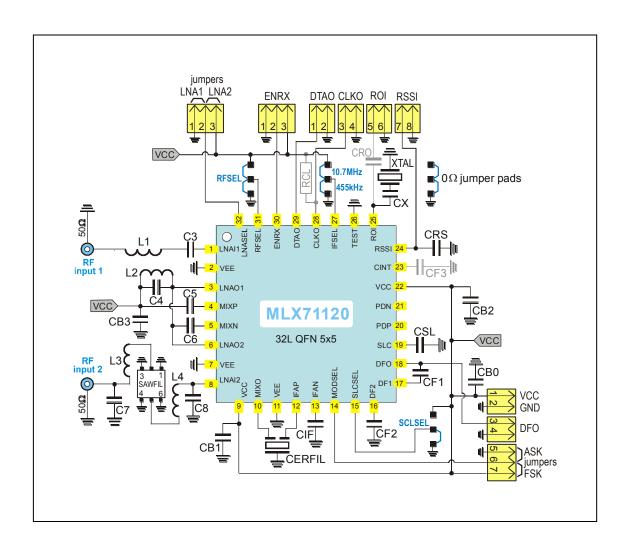


Fig. 6: Circuit schematic



# **Evaluation Board Description**

#### 3.2.1 Component Arrangement Top Side (Averaging Data Slicer, IF2 = 10.7MHz)

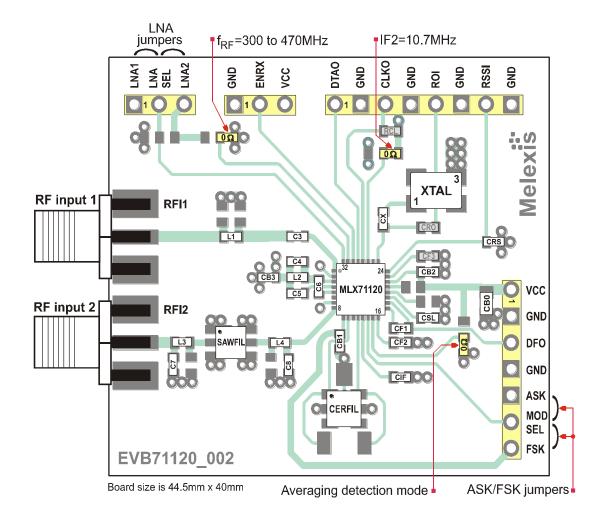


Fig. 7: PCB top-side view

## 300 to 930MHz Receiver Evaluation Board Description

## 3.3 Component List for Dual-Channel Application

Below table is valid for test circuits shown in Figures 3.1 to 3.2.

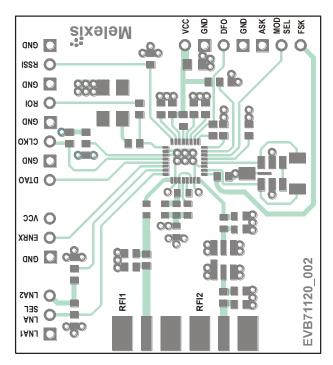
Part   Size				ircuits shown					
C4	Part	Size	Value @ 315 MHz	Value @ 433.92 MHz	Value @ 868.3 MHz	Value @ 915 MHz	Tol.		Description
C5	C3	0603	100 pF	100 pF	100 pF	100 pF	±5%	LNA input filter	ring capacitor
CR   0603   100 pF   100 pF   100 pF   100 pF   15%   MIX1 negative input matching capacitor	C4	0603	4.7 pF	3.9 pF	2.2 pF	1.5 pF	±5%	LNA output tar	nk capacitor
C7	C5	0603	100 pF	100 pF	100 pF	100 pF	±5%	MIX1 positive	input matching capacitor
C8	C6	0603	100 pF	100 pF	100 pF	100 pF	±5%	MIX1 negative	input matching capacitor
CB0	C7	0603	NIP	4.7 pF	3.9 pF	NIP	±5%	matching capa	acitor
CB1	C8	0603	NIP	NIP	1.0 pF	NIP	±5%	matching capa	acitor
CB2	CB0	0805	330 nF	330 nF	330 nF	330 nF	±10%	decoupling cap	pacitor
CB3	CB1	0603	330 pF	330 pF	330 pF	330 pF	±10%	decoupling cap	pacitor
CF1	CB2	0603	330 pF	330 pF	330 pF	330 pF	±10%	decoupling cap	pacitor
CF2   O603   330 pF   330 p	CB3	0603	330 pF	330 pF	330 pF	330 pF	±10%	decoupling cap	pacitor
CF2   0603   330 pF   330 pF   330 pF   330 pF   330 pF   350 pF   500 pF   500 pF   500 point of rate at rate of 4 kbps NRZ	CF1	0603	680 pF	680 pF	680 pF	680 pF	±10%		
CF3   O603   Connected to ground if noise filter not used	CF2	0603	330 pF	330 pF	330 pF	330 pF	±10%		
CIF   0603	CE3	0603		value according	to the data rate		+10%	•	citor for noise cancellation
CP1   0603   33 nF   33 nF   33 nF   33 nF   33 nF   ±10%   positive PKDET capacitor, for data rate of 4 kbps NRZ	OI 3	0003	conr	nected to ground it	f noise filter not i	used	±10 /6	filter	
CP2	CIF	0603	1 nF	1 nF	1 nF	1 nF	±10%	IFA feedback	capacitor
CRS   0603   1 nF   1 nF   1 nF   1 nF   1 nF   ±10%   RSSI output low pass capacitor   optional capacitor, to couple external RO signal   optional capacitor, to couple texternal RO signal   optional capacit	CP1	0603	33 nF	33 nF	33 nF	33 nF	±10%	for data rate of 4 kbps NRZ	
CRO         0603         1 nF         1 nF         1 nF         1 nF         ±5%         optional capacitor, to couple external RO signal data slicer capacitor, for overaging detection mode only           CX         0603         27 pF         27 pF         27 pF         ±10% for averaging detection mode only         ±10% for data rate of 4 kbps NRZ           CX         0603         27 pF         27 pF         27 pF         ±5% crystal series capacitor for data rate of 4 kbps NRZ           L1         0603         56 nH         27 nH         15 nH         3.9 nH         3.9 nH         ±5% matching inductor           L2         0603         27 nH         15 nH         3.9 nH         3.9 nH         ±5% matching inductor           L3         0603         0 Ω         47 nH         22 nH         0 Ω         ±5% matching inductor           L4         0603         56 nH         68 nH         22 nH         0 Ω         ±5% matching inductor           RCL         0603         3.3 kΩ         3.3 kΩ         3.3 kΩ         3.3 kΩ         ±5% optional CLK output resistor, to clock output signal generated           SAW FIL         SMD 3x3 signal         RF3446 (433.92 MHz)         SAFCC868M SL0X00 (863 MHz)         SAFCC915M AL0N00 (915 MHz)         low-loss SAW filter from Murata or equivalent part	CP2	0603	33 nF	33 nF	33 nF	33 nF	±10%		
CSL   0603   100 nF   100 nF   100 nF   100 nF   100 nF   110 n	CRS	0603	1 nF	1 nF	1 nF	1 nF	±10%	RSSI output low pass capacitor	
To raveraging detection mode only   \$\frac{\pmath{to r}}{\text{for a averaging detection mode only}}   \$\frac{\pmath{to r}}{\text{for data rate of 4 kbps NRZ}}   \$\frac{\pmath{to r}}{\text{cystal series capacitor}}   \$\frac{\pmath{to r}}{\pmath{to r}}   \$\pmath{	CRO	0603			1 nF	1 nF	±5%		
CX   O603   27 pF   27 pF   27 pF   27 pF   ±5%   crystal series capacitor	CSL	0603					+10%		
L1							= . 0 / 0	for data rate of	f 4 kbps NRZ
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	CX		27 pF	27 pF	27 pF	27 pF		crystal series	capacitor
L3					0 Ω	0 Ω			
L4         0603         56 nH         68 nH         22 nH         0 Ω         ±5%         matching inductor           RCL         0603         3.3 kΩ         3.3 kΩ         3.3 kΩ         ±5%         optional CLK output resistor, to clock output signal generated           SAW FIL SAW			27 nH	15 nH		3.9 nH	±5%	LNA output tar	nk inductor
RCL   0603   3.3 kΩ   3.3 kΩ   3.3 kΩ   3.3 kΩ   ±5%   optional CLK output resistor, to clock output signal generated	L3	0603	0 Ω	47 nH	22 nH	0 Ω	±5%	matching indu	ctor
SAW FIL   SMD SMOTOO (315 MHz)   SFECF10M7HA00 (868.3 MHz)   SMD (3.45x3.1   SMD (5.5x6.0)   SMD (5.5x6.0)   SMD (5.5x6.0   SMD (5.5x6.0   SMD (5.5x6.0)   SMD (5.5x6.0   SMD (5	L4	0603	56 nH	68 nH	22 nH	0 Ω	±5%	matching indu	ctor
SAW FILE   SAW SMD   SMOTOO (315 MHz)   SL0X00 (868.3 MHz)   SL0X00 (915 MHz)   SMD (915 MHz)   SFECF10M7HA00	RCL	0603	3.3 kΩ	3.3 kΩ	3.3 kΩ	3.3 kΩ	±5%		
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$			SM0T00		SL0X00	AL0N00			
FIL SMD 6.5x6.0	CER							IF2=10.7MHz	corariio iiitoi
XTAL SMD SMD 5x3.2	FIL							IF2=455kHz	1
XTAL SMD 5x3.2									
MHz MHz MHz IF2=455KHZ	XTAL		18.094444M HZ-12-50-F- H20-T2075- W2-T	24.701111MHZ -12-50-F-H20- T2075-W2-T	24.416667MH Z-12-50-F- H20-T2075- W2-T	25.713889M HZ-12-50-F- H20-T2075- W2-T		IF2=10.7MHz	
±20ppm cal., ±30ppm temp.								IF2=455kHz	
				±20ppm cal., ±	30ppm temp.				

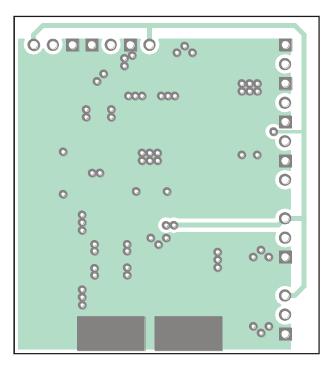
**Note:** NIP – not in place, may be used optionally

## 300 to 930MHz Receiver Evaluation Board Description

#### 3.4 PCB Layouts for Antenna Space Diversity

• Board layout data in Gerber format is available, board size is 40mm x 44.5mm.





PCB top view

PCB bottom view

#### 4 Board Variants

Туре	Frequency/MHz		Board Execution		
EVB71120	-315		-A	antenna version	
	-433		-C	connector version	
	-868				
	-915				

Note:

available EVB setups

## 300 to 930MHz Receiver Evaluation Board Description

## 5 Package Description



The device MLX71120 is RoHS compliant.

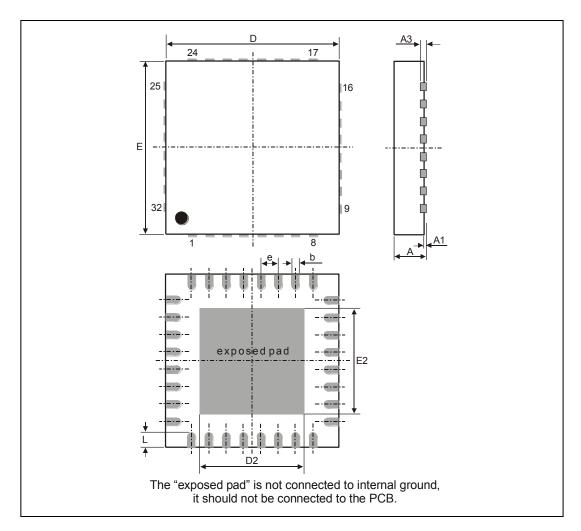


Fig. 8: 32L QFN 5x5 Quad

all Dime	all Dimension in mm									
	D	Е	D2	E2	Α	<b>A1</b>	A3	L	е	b
min	4.75	4.75	3.00	3.00	0.80	0	0.20	0.3	0.50	0.18
max	5.25	5.25	3.25	3.25	1.00	0.05	0.20	0.5	0.50	0.30
all Dime	all Dimension in inch									
min	0.187	0.187	0.118	0.118	0.0315	0	0.0079	0.0118	0.0197	0.0071
max	0.207	0.207	0.128	0.128	0.0393	0.002	0.0079	0.0197	0.0197	0.0118

#### 5.1 Soldering Information

 The device MLX71120 is qualified for MSL3 with soldering peak temperature 260 deg C according to JEDEC J-STD-20



## 300 to 930MHz Receiver Evaluation Board Description

### 6 Reliability Information

This Melexis device is classified and qualified regarding soldering technology, solderability and moisture sensitivity level, as defined in this specification, according to following test methods:

#### Reflow Soldering SMD's (Surface Mount Devices)

- IPC/JEDEC J-STD-020
  - "Moisture/Reflow Sensitivity Classification for Nonhermetic Solid State Surface Mount Devices (classification reflow profiles according to table 5-2)"
- EIA/JEDEC JESD22-A113
  - "Preconditioning of Nonhermetic Surface Mount Devices Prior to Reliability Testing (reflow profiles according to table 2)"

#### Wave Soldering SMD's (Surface Mount Devices) and THD's (Through Hole Devices)

- FN60749-20
  - "Resistance of plastic- encapsulated SMD's to combined effect of moisture and soldering heat"
- EIA/JEDEC JESD22-B106 and EN60749-15
  - "Resistance to soldering temperature for through-hole mounted devices"

#### Iron Soldering THD's (Through Hole Devices)

EN60749-15

"Resistance to soldering temperature for through-hole mounted devices"

#### Solderability SMD's (Surface Mount Devices) and THD's (Through Hole Devices)

 EIA/JEDEC JESD22-B102 and EN60749-21 "Solderability"

For all soldering technologies deviating from above mentioned standard conditions (regarding peak temperature, temperature gradient, temperature profile etc) additional classification and qualification tests have to be agreed upon with Melexis.

The application of Wave Soldering for SMD's is allowed only after consulting Melexis regarding assurance of adhesive strength between device and board.

Melexis is contributing to global environmental conservation by promoting **lead free** solutions. For more information on qualification of **RoHS** compliant products (RoHS = European directive on the Restriction Of the Use of Certain Hazardous Substances) please visit the quality page on our website:

http://www.melexis.com/quality\_leadfree.aspx

### 7 ESD Precautions

Electronic semiconductor products are sensitive to Electro Static Discharge (ESD). Always observe Electro Static Discharge control procedures whenever handling semiconductor products.



## 300 to 930MHz Receiver Evaluation Board Description

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Europe, Africa: Americas: Asia:

Phone: +32 1367 0495 Phone: +1 603 223 2362 Phone: +32 1367 0495 E-mail: <a href="mailto:sales\_europe@melexis.com">sales\_europe@melexis.com</a> E-mail: <a href="mailto:sales\_europe@melexis.com">sales\_europe@melexis.com</a> E-mail: <a href="mailto:sales\_europe@melexis.com">sales\_europe@melexis.com</a>

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