# 24-Bit, 192-kHz Sampling, Enhanced Multilevel, Delta-Sigma, Audio Digital-to-Analog Converter

#### **FEATURES**

- 24-Bit Resolution
- Analog Performance (V<sub>CC</sub> = 5 V):
  - Dynamic Range: 106 dBSNR: 106 dB, Typical
  - THD+N: 0.002%, Typical
  - Full-Scale Output: 3.9 Vp-p, Typical
- 4x/8x Oversampling Digital Filter:
  - Stop-Band Attenuation: -50 dB
  - Pass-Band Ripple: ±0.04 dB
- Sampling Frequency: 5 kHz to 200 kHz
- System Clock: 128 f<sub>S</sub>, 192 f<sub>S</sub>, 256 f<sub>S</sub>, 384 f<sub>S</sub>, 512 f<sub>S</sub>, 768 f<sub>S</sub>, 1152 f<sub>S</sub> With Autodetect
- Software Control (PCM1780, PCM1782):
  - Accepts 16-, 18-, 20-, and 24-Bit Audio Data
  - Formats: Right-Justified, I<sup>2</sup>S, and Left-Justified
  - Digital Attenuation: Mode Selectable
    - 0 dB to -63 dB, 0.5 dB/step
    - 0 dB to -100 dB, 1 dB/step
  - Digital De-Emphasis
  - Digital Filter Rolloff: Sharp or Slow
  - Soft Mute
  - Zero Flags for Each Output
  - Open-Drain Output Zero Flag (PCM1782)
- Hardware Control (PCM1781):
  - I<sup>2</sup>S and 16-Bit Word, Right-Justified
  - Digital De-Emphasis
  - Soft Mute
  - Zero Flag for L-, R-Channel Common Output
- Power Supply: 5-V Single Supply
- Small, 16-Lead SSOP Package (150 mil)
- Pin-Compatible with PCM1680

#### **APPLICATIONS**

- A/V Receivers
- DVD Movie Players
- DVD Add-On Cards For High-End PCs
- DVD Audio Players
- HDTV Receivers
- Car Audio Systems
- Other Applications Requiring 24-Bit Audio

#### DESCRIPTION

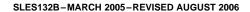
The PCM1780/81/82 is a CMOS. monolithic. integrated circuit, which includes stereo digital-to-analog converters and support circuitry in a small 16-lead SSOP package. The data converters use TI's enhanced multilevel delta-sigma architecture to achieve excellent dynamic performance tolerance to clock improved jitter. PCM1780/81/82 accepts industry standard audio data formats with 16- to 24-bit data, providing easy interfacing to audio DSP and decoder chips. Sampling rates up to 200 kHz are supported. The PCM1780/82 provides а full user-programmable functions through a three-wire serial control port, which supports register write functions. The PCM1781 provides a subset of user-programmable functions through four control pins.

The PCM1780 is pin-compatible with the PCM1680 (8-channel DAC).



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This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage. ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

#### **ABSOLUTE MAXIMUM RATINGS**

over operating free-air temperature range (unless otherwise noted)(1)

		PCM1780, PCM1781, PCM1782
Supply voltage	V <sub>CC</sub>	–0.3 V to 6.5 V
Input voltage		-0.3 V to V <sub>CC</sub> + 0.3 V, < 6.5 V
Input current (any pins except supplies)		±10 mA
Ambient temperature under bias		- 40°C to 125°C
Storage temperature		– 55°C to 150°C
Junction temperature		150°C
Lead temperature (soldering)		260°C, 5 s
Package temperature (IR reflow, peak)		260°C

<sup>(1)</sup> Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

#### RECOMMENDED OPERATING CONDITIONS

over operating free-air temperature range

		MIN	NOM	MAX	UNIT
Digital and analog supply voltage, \	/ <sub>cc</sub>	4.5	5	5.5	V
Digital input logic family			TTL		
Disital input along from the	System clock	8.192		36.864	MHz
Digital input clock frequency	Sampling clock	32		192	kHz
Analog output load resistance		5			kΩ
Analog output load capacitance				50	pF
Digital output load capacitance				20	pF
Operating free-air temperature, T <sub>A</sub>		-25		85	°C

#### **ELECTRICAL CHARACTERISTICS**

All specifications at  $T_A = 25$ °C,  $V_{CC} = 5$  V,  $f_S = 48$  kHz, system clock = 512  $f_S$ , and 24-bit data (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
Resolution				24		Bits
DATA FORMAT						
Audio data interface format PCM1780, PCM1782			Right-justified, I <sup>2</sup> S, left-justified			
	PCM1781		I <sup>2</sup> S, right-justified			
Adia data bit lawath	PCM1780, PCM1782		,	3-, 20-, 24- electable	bit	
Audio data bit length	PCM1781			bit I <sup>2</sup> S, 16 nt-justified	-bit	
Audio data format			MSB-first,	2s comple	ement	
f <sub>S</sub> Sampling frequency			5		200	kHz
System clock frequency			128 f <sub>S</sub> , 192 512 f <sub>S</sub> , 7	f <sub>S</sub> , 256 f <sub>S</sub> , 768 f <sub>S</sub> , 115	384 f <sub>S</sub> , 52 f <sub>S</sub>	



# **ELECTRICAL CHARACTERISTICS (continued)**

All specifications at  $T_A = 25$ °C,  $V_{CC} = 5$  V,  $f_S = 48$  kHz, system clock = 512  $f_S$ , and 24-bit data (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
DIGITAL	_ INPUT/OUTPUT					
	Logic family		TTI	L compatib	le	
$V_{IH}$	Input logic lovel	nout logic lovel		2		
V <sub>IL</sub>	Input logic level				0.8	Vdc
I <sub>IH</sub> <sup>(1)</sup>		$V_{IN} = V_{CC}$			10	
I <sub>IL</sub> <sup>(1)</sup>	Input logic current	$V_{IN} = 0 V$			-10	^
I <sub>IH</sub> <sup>(2)</sup>	Input logic current	$V_{IN} = V_{CC}$		65	100	μΑ
I <sub>IL</sub> <sup>(2)</sup>		$V_{IN} = 0 V$			-10	
V <sub>OH</sub> <sup>(3)</sup>	Output logic level	$I_{OH} = -1 \text{ mA}$	2.4			Vdc
V <sub>OL</sub> <sup>(4)</sup>	Output logic level	I <sub>OL</sub> = 1 mA			0.4	vuc
DYNAM	IC PERFORMANCE <sup>(5)</sup>					
		$V_{OUT} = 0 \text{ dB}, f_S = 48 \text{ kHz}$		0.002%	0.006%	
THD+N	Total harmonic distortion + noise	$V_{OUT} = 0$ dB, $f_S = 96$ kHz, system clock = 256 $f_S$		0.003%		
		$V_{OUT} = 0$ dB, $f_S = 192$ kHz, system clock = 128 $f_S$		0.004%		
		EIAJ, A-weighted, $f_S = 48 \text{ kHz}$	100	106		
	Dynamic range	A-weighted, $f_S = 96 \text{ kHz}$ , system clock = 256 $f_S$		104		dB
		A-weighted, $f_S = 192 \text{ kHz}$ , system clock = 128 $f_S$		102		
		EIAJ, A-weighted, f <sub>S</sub> = 48 kHz	100	106		
SNR	Signal-to-noise ratio	A-weighted, $f_S = 96$ kHz, system clock = 256 $f_S$		104		dB
		A-weighted, $f_S = 192 \text{ kHz}$ , system clock = 128 $f_S$		102		
		f <sub>S</sub> = 48 kHz	97	103		
	Channel separation	f <sub>S</sub> = 96 kHz, system clock = 256 f <sub>S</sub>		101		dB
		$f_S$ = 192 kHz, system clock = 128 $f_S$		100		
DC ACC	CURACY					
	Gain error			±1	±6	% of FSR
	Gain mismatch, channel-to-channel			±1	±6	% of FSR
	Bipolar zero error	$V_{OUT} = 49\%$ of $V_{CC}$ at BPZ input		±30	±80	mV
ANALO	G OUTPUT		ı			
	Output voltage	Full scale (-0 dB)		0.78 V <sub>CC</sub>		Vp-p
	Bipolar zero voltage			0.49 V <sub>CC</sub>		Vdc
	Load impedance	AC-coupled load	5			kΩ
DIGITAL	FILTER PERFORMANCE					
Filter Cl	haracteristics (Sharp Rolloff)					
	Pass band	±0.04 dB			0.454 f <sub>S</sub>	
	Stop band		0.546 f <sub>S</sub>			-
	Pass-band ripple				0.04	dB
	Stop-band attenuation	Stop band = 0.546 f <sub>S</sub>	-50			dB

<sup>(1)</sup> Pins 5, 6, 7, 8: SCK, DATA, BCK, LRCK

<sup>(2)</sup> Pins 2, 3, 4: MS, MC, MD (PCM1780/PCM1782). Pins 1, 2, 3, 4: FMT, DEMP0, DEMP1, MUTE (PCM1781) (3) Pins 1, 16: ZEROL, ZEROR (PCM1780). Pin 16: ZEROA (PCM1781) (4) Pins 1, 16: ZEROL, ZEROR (PCM1780/PCM1782). Pin 16: ZEROA (PCM1781)

<sup>(5)</sup> Analog performance specifications are measured using the System Two™ Cascade audio measurement system by Audio Precision™.



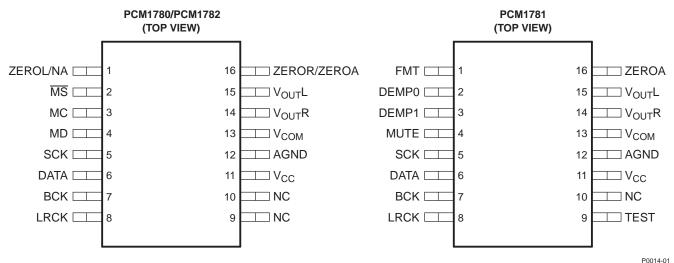
# **ELECTRICAL CHARACTERISTICS (continued)**

All specifications at  $T_A = 25$ °C,  $V_{CC} = 5$  V,  $f_S = 48$  kHz, system clock = 512  $f_S$ , and 24-bit data (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT	
Filter	Characteristics (Slow Rolloff, PCM1780/PC	M1782)	•		'		
	Pass band	±0.5 dB			0.198 f <sub>S</sub>		
	Stop band		0.884 f <sub>S</sub>				
	Pass-band ripple				±0.5	dB	
	Stop-band attenuation	Stop band = 0.884 f <sub>S</sub>	-35			dB	
	Delay time			20/f <sub>S</sub>		s	
	De-emphasis error			±0.1		dB	
ANAL	OG FILTER PERFORMANCE	·					
	Fraguenay reconomic	at 20 kHz		-0.02		dB	
	Frequency response	at 44 kHz		-0.07		uБ	
POWE	R SUPPLY REQUIREMENTS						
$V_{CC}$	Voltage range		4.5	5	5.5	Vdc	
		f <sub>S</sub> = 48 kHz		25	40		
$I_{CC}$	Supply current	f <sub>S</sub> = 96 kHz, system clock = 256 f <sub>S</sub>		30		mA	
		f <sub>S</sub> = 192 kHz, system clock = 128 f <sub>S</sub>		30			
		f <sub>S</sub> = 48 kHz		125	200		
	Power dissipation	f <sub>S</sub> = 96 kHz, system clock = 256 f <sub>S</sub>		150		mW	
		f <sub>S</sub> = 192 kHz, system clock = 128 f <sub>S</sub>		150		1	
TEMP	ERATURE RANGE		•		1		
$T_A$	Operation temperature		-25		85	°C	
$\theta_{JA}$	Thermal resistance			115		°C/W	

## **DEVICE INFORMATION**

## **PIN ASSIGNMENTS**



F0014-01



# DEVICE INFORMATION (continued) TERMINAL FUNCTIONS—PCM1780/PCM1782

TERMINAL		1/0	DESCRIPTION		
NAME	NO.	I/O	JESCRIPTION		
AGND	12	-	Ground		
ВСК	7	I	Audio data bit clock input <sup>(1)</sup>		
DATA	6	ı	Audio data digital input <sup>(1)</sup>		
LRCK	8	ı	Audio data left and right clock input <sup>(1)</sup>		
MC	3	ı	Mode control clock input <sup>(1)(2)</sup>		
MD	4	I	Mode control data input (1)(2)		
MS	2	I	Mode control select input <sup>(1)(2)</sup>		
NC	9, 10	_	No connection		
SCK	5	ı	System clock input <sup>(1)</sup>		
$V_{CC}$	11	_	Power supply, 5-V		
V <sub>COM</sub>	13	_	Common voltage decoupling		
V <sub>OUT</sub> L	15	0	Analog output for L-channel		
$V_{OUT}R$	14	0	Analog output for R-channel		
ZEROL/NA	1	0	Zero flag output for L-channel / No assign (3)		
ZEROR/ZEROA	16	0	Zero flag output for R-channel / Zero flag output for L- and R-channels <sup>(3)</sup>		

- (1) Schmitt-trigger input
- (2) Pulldown
- (3) Open-drain output (PCM1782)

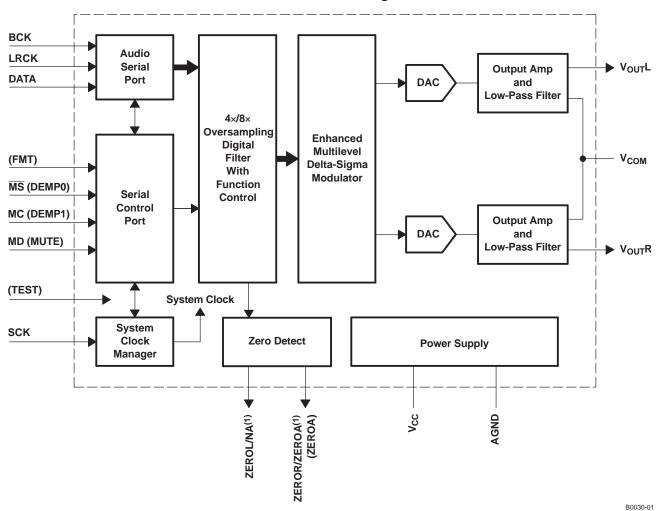
## **TERMINAL FUNCTIONS—PCM1781**

TERMINAL		1/0	DECORPTION		
NAME	NO.	1/0	DESCRIPTION		
AGND	12	_	Ground		
BCK	7	I	Audio data bit clock input <sup>(1)</sup>		
DATA	6	I	Audio data digital input <sup>(1)</sup>		
DEMP0	2	I	De-emphasis control <sup>(1)(2)</sup>		
DEMP1	3	I	De-emphasis control <sup>(1)(2)</sup>		
FMT	1	I	Data format select <sup>(1)(2)</sup>		
LRCK	8	I	Audio data left and right clock input <sup>(1)</sup>		
MUTE	4	I	Soft mute control <sup>(1)(2)</sup>		
NC	10	_	No connection		
SCK	5	I	System clock input <sup>(1)</sup>		
TEST	9	_	Test pin for factory use. Must be LOW or open <sup>(1)(2)</sup>		
$V_{CC}$	11	_	Power supply, 5-V		
V <sub>COM</sub>	13	_	Common voltage decoupling		
$V_{OUT}L$	15	0	Analog output for L-channel		
V <sub>OUT</sub> R	14	0	Analog output for R-channel		
ZEROA	16	0	Zero flag output for L- and R-channels		

- (1) Schmitt-trigger input
- (2) Pulldown



## **Functional Block Diagram**



(1) Open-drain output for the PCM1782

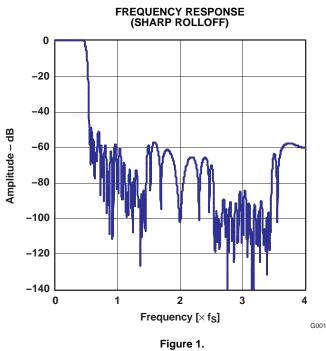
NOTE: Signal names in parentheses ( ) are for the PCM1781.



## TYPICAL PERFORMANCE CURVES

All specifications at  $T_A = 25$ °C,  $V_{CC} = 5$  V,  $f_S = 48$  kHz, system clock = 512  $f_S$ , and 24-bit data, unless otherwise noted

## **Digital Filter (De-Emphasis Off)**



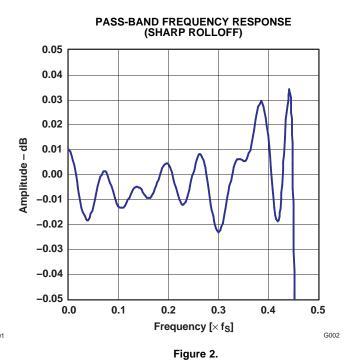
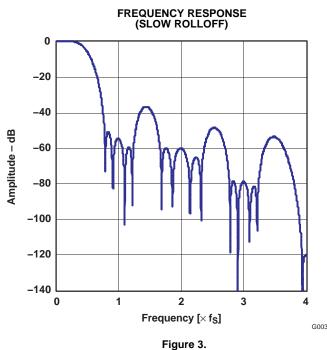


Figure 1.



TRANSITION CHARACTERISTICS (SLOW ROLLOFF)

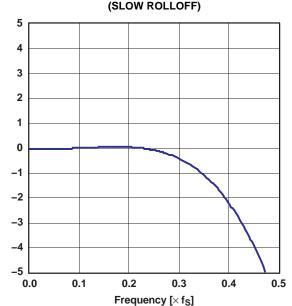


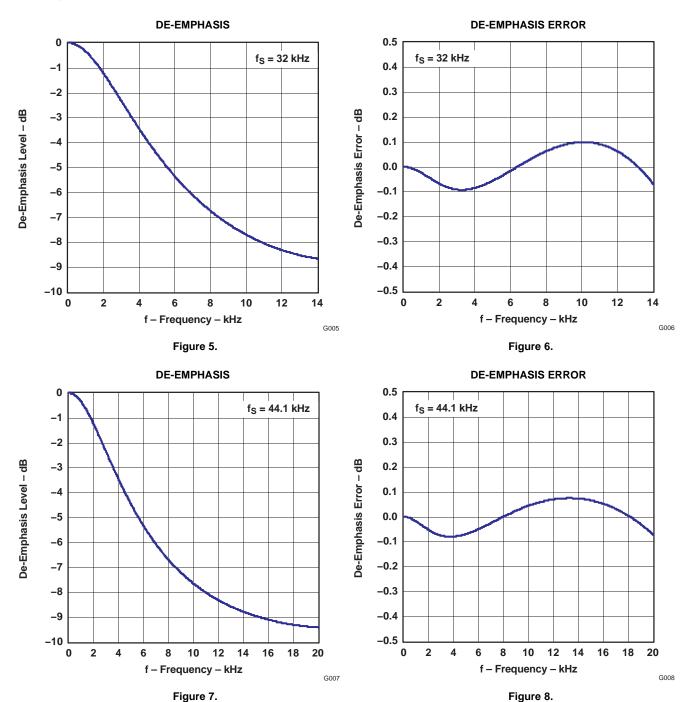
Figure 4.

Amplitude – dB



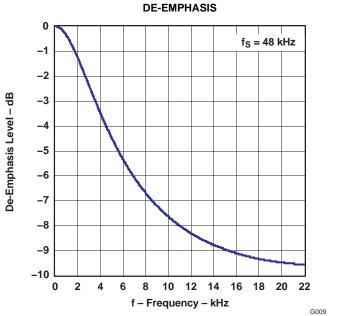
All specifications at  $T_A = 25$ °C,  $V_{CC} = 5$  V,  $f_S = 48$  kHz, system clock = 512  $f_S$ , and 24-bit data, unless otherwise noted

# **De-Emphasis Filter**





All specifications at  $T_A = 25$ °C,  $V_{CC} = 5$  V,  $f_S = 48$  kHz, system clock = 512  $f_S$ , and 24-bit data, unless otherwise noted



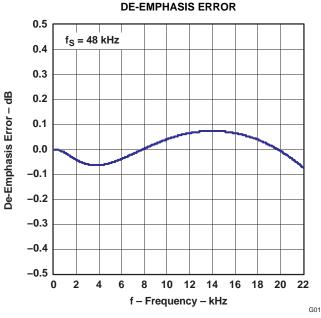
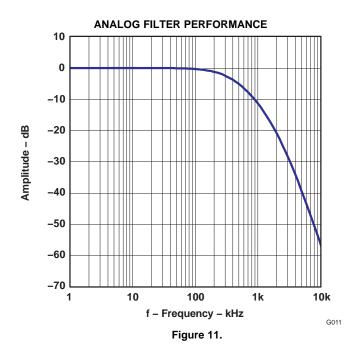


Figure 9. Figure 10.

# **Analog Filter**

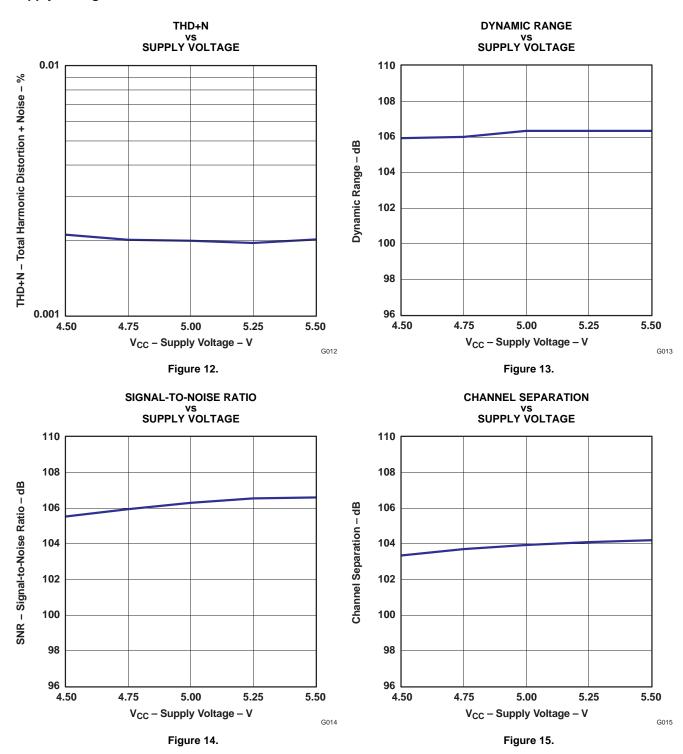




All specifications at  $T_A = 25$ °C,  $V_{CC} = 5$  V,  $f_S = 48$  kHz, system clock = 512  $f_S$ , and 24-bit data, unless otherwise noted

# **Analog Dynamic Performance**

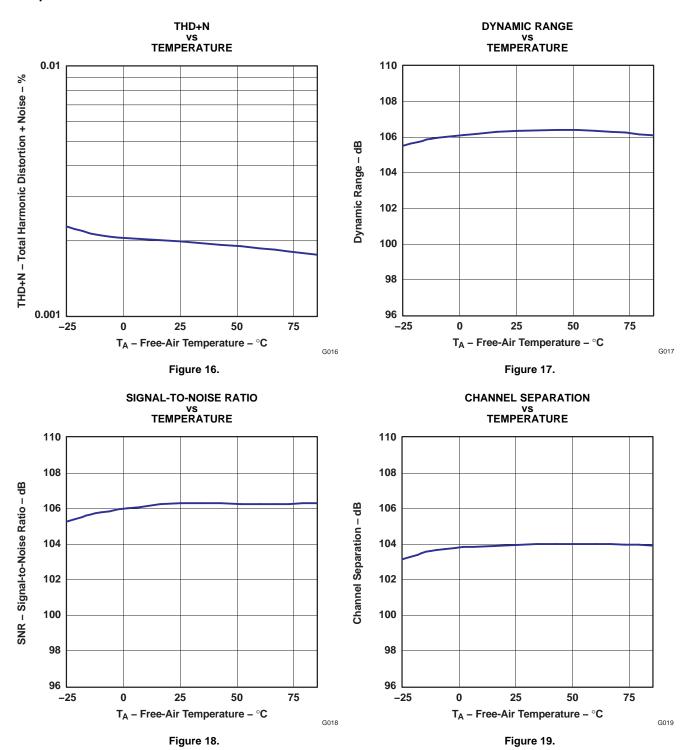
## **Supply Voltage Characteristics**





All specifications at  $T_A = 25$ °C,  $V_{CC} = 5$  V,  $f_S = 48$  kHz, system clock = 512  $f_S$ , and 24-bit data, unless otherwise noted

## **Temperature Characteristics**





## SYSTEM CLOCK AND RESET FUNCTIONS

#### **System Clock Input**

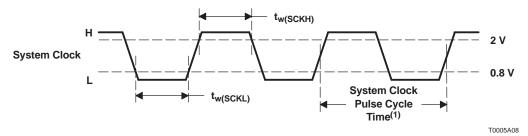
The PCM1780/81/82 requires a system clock for operating the digital interpolation filters and multilevel delta-sigma modulators. The system clock is applied at the SCK input (pin 5). Table 1 shows examples of system clock frequencies for common audio sampling rates.

Figure 20 shows the timing requirements for the system clock input. For optimal performance, it is important to use a clock source with low phase jitter and noise. Ti's PLL170x family of multiclock generators is an excellent choice for providing the PCM1780/81/82 system clock.

Table 1. System Clock Frequencies for Common Audio Sampling Frequencies

SAMPLING		SYSTEM CLOCK FREQUENCY (f <sub>SCK</sub> ), MHz							
FREQUENCY	128 f <sub>S</sub>	192 f <sub>S</sub>	256 f <sub>S</sub>	384 f <sub>S</sub>	512 f <sub>S</sub>	768 f <sub>S</sub>	1152 f <sub>S</sub>		
8 kHz	1.024	1.536	2.048	3.072	4.096	6.144	9.216		
16 kHz	2.048	3.072	4.096	6.144	8.192	12.288	18.432		
32 kHz	4.096	6.144	8.192	12.288	16.384	24.576	36.864		
44.1 kHz	5.6448	8.4672	11.2896	16.9344	22.5792	33.8688	_(1)		
48 kHz	6.144	9.216	12.288	18.432	24.576	36.864	_(1)		
88.2 kHz	11.2896	16.9344	22.5792	33.8688	_(1)	_(1)	_(1)		
96 kHz	12.288	18.432	24.576	36.864	_(1)	_(1)	_(1)		
192 kHz	24.576	36.864	_(1)	_(1)	_(1)	_(1)	_(1)		

(1) This system clock frequency is not supported for the given sampling frequency.



(1)  $1/128 f_S$ ,  $1/192 f_S$ ,  $1/256 f_S$ ,  $1/384 f_S$ ,  $1/512 f_S$ ,  $1/768 f_S$ , or  $1/1152 f_S$ 

	PARAMETER				UNIT
t <sub>w(SCKH)</sub>	System clock pulse duration, HIGH	7			ns
t <sub>w(SCKL)</sub>	System clock pulse duration, LOW	7			ns

Figure 20. System Clock Input Timing



#### **Power-On-Reset Functions**

The PCM1780/81/82 includes a power-on-reset function. Figure 21 shows the operation of this function. With the system clock active and  $V_{CC} > 3$  V (typical, 2.2 V to 3.7 V), the power-on-reset function is enabled. The initialization sequence requires 3072 system clocks from the time  $V_{CC} > 3$  V (typical, 2.2 V to 3.7 V). After the initialization period, the PCM1780/82 is set to its reset default state, as described in the *Mode Control Register* section of this data sheet.

During the reset period (3072 system clocks), the analog output is forced to the common voltage ( $V_{COM}$ ), or  $V_{CC}/2$ . After the reset period, the internal register is initialized in the next  $1/f_{S}$  period and if SCK, BCK, and LRCK are provided continuously, the PCM1780/81/82 provides the proper analog output with a group delay corresponding to the input data.

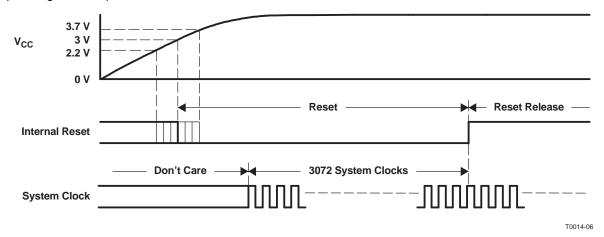


Figure 21. Power-On-Reset Timing

#### **Audio Serial Interface**

The audio serial interface for the PCM1780/81/82 consists of a three-wire synchronous serial port. It includes LRCK (pin 8), BCK (pin 7), and DATA (pin 6). BCK is the serial audio bit clock, and it is used to clock the serial data present on DATA into the serial shift register of the audio interface. Serial data is clocked into the PCM1780/81/82 on the rising edge of BCK. LRCK is the serial audio left/right word clock. It is used to latch serial data into the internal registers of the serial audio interface.

Both LRCK and BCK should be synchronous with the system clock. Ideally, it is recommended that LRCK and BCK be derived from the system clock input, SCK. LRCK is operated at the sampling frequency, f<sub>S</sub>. BCK can be operated at 32, 48, or 64 times the sampling frequency.

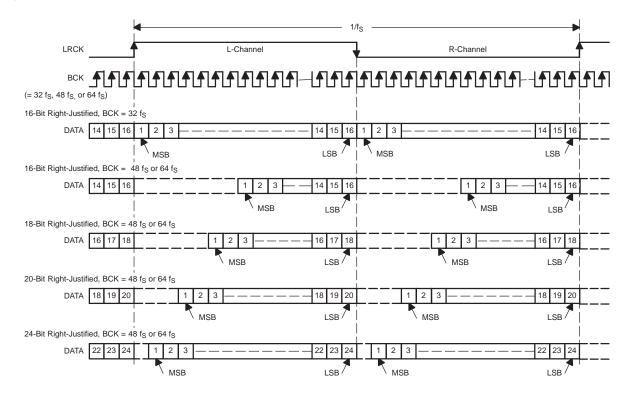
Internal operation of the PCM1780/81/82 is synchronized with LRCK. Accordingly, internal operation of the device is suspended when the sampling rate clock, LRCK, is changed or SCK and/or BCK is interrupted at least for three bit-clock cycles. If SCK, BCK, and LRCK are provided continuously after this suspended condition, the internal operation is resynchronized automatically within a period of less than 3/f<sub>S</sub>. External resetting is not required.

#### **Audio Data Formats and Timing**

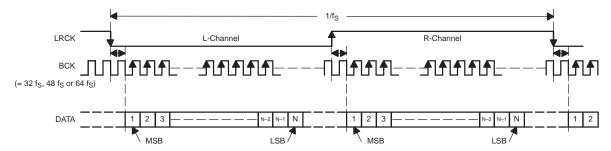
The PCM1780/82 supports industry-standard audio data formats, including right-justified, I²S, and left-justified. The PCM1781 supports I²S and 16-bit-word, right-justified. The data formats are shown in Figure 22. Data formats are selected for the PCM1780/82 using the format bits, FMT[2:0], located in control register 20, and are selected for the PCM1781 using the FMT pin. The default data format is 24-bit, left-justified. All formats require binary 2s complement, MSB-first audio data. Figure 23 shows a detailed timing diagram for the serial audio interface.



## (1) Right-Justified Data Format; L-Channel = HIGH, R-Channel = LOW



## (2) I2S Data Format; L-Channel = LOW, R-Channel = HIGH



#### (3) Left-Justified Data Format; L-Channel = HIGH, R-Channel = LOW

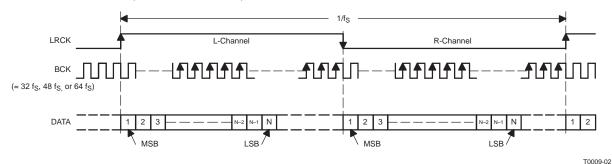
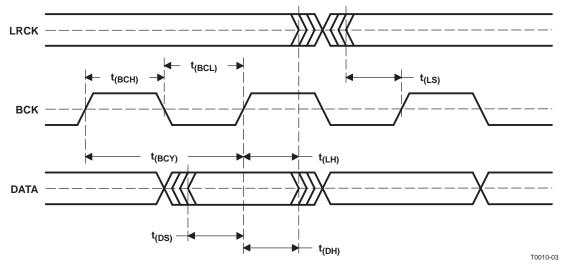


Figure 22. Audio Data Input Formats





	PARAMETER	MIN	UNIT
t <sub>(BCY)</sub>	BCK pulse cycle time	1/(32 f <sub>S</sub> ), 1/(48 f <sub>S</sub> ), 1/(64 f <sub>S</sub> ) <sup>(1)</sup>	
t <sub>(BCH)</sub>	BCK pulse duration, HIGH	35	ns
t <sub>(BCL)</sub>	BCK pulse duration, LOW	35	ns
t <sub>(LS)</sub>	LRCK setup time to BCK rising edge	10	ns
t <sub>(LH)</sub>	LRCK hold time to BCK rising edge	10	ns
t <sub>(DS)</sub>	DATA setup time	10	ns
t <sub>(DH)</sub>	DATA hold time	10	ns

<sup>(1)</sup>  $f_S$  is the sampling frequency (e.g., 44.1 kHz, 48 kHz, 96 kHz, etc.).

Figure 23. Audio Interface Timing



#### OVERSAMPLING RATE CONTROL

The PCM1780/81/82 automatically controls the oversampling rate of the delta-sigma D/A converters with the system clock frequency. The oversampling rate is set to  $64\times$  oversampling with an  $1152\text{-}f_S$ ,  $768\text{-}f_S$ , or  $512\text{-}f_S$  system clock, to  $32\times$  oversampling with a  $384\text{-}f_S$  or  $256\text{-}f_S$  system clock, or to  $16\times$  oversampling with a  $192\text{-}f_S$  or  $128\text{-}f_S$  system clock.

## ZERO FLAGS (PCM1780/82)

## **Zero-Detect Condition**

Zero detection for each output channel is independent from the other. If the data for a given channel remains at a 0 level for 1024 sample periods (or LRCK clock periods), a zero-detect condition exists for that channel.

## **Zero-Flag Outputs**

Each channel has a corresponding zero-flag pin, ZEROL (pin 1) or ZEROR (pin 16). Given that a zero-detect condition exists for one or more channels, the zero-flag pins for those channels are set to a logic-1 state. The zero-flag pins can be used to operate external mute circuits, or used as status indicators for a microcontroller, audio signal processor, or other digitally controlled function.

The active polarity of the zero-flag outputs can be inverted by setting the ZREV bit of control register 22 to 1. The reset default is active-high output, or ZREV = 0.

The L-channel and R-channel common zero flag can be selected by setting the AZRO bit of control register 22 to 1. The reset default is for independent L-channel and R-channel zero flags, or AZRO = 0.

On the PCM1782, ZEROL and ZEROR are open-drain outputs.

## **ZERO FLAG (PCM1781)**

The PCM1781 has a zero-flag pin, ZEROA (pin 16). ZEROA is the L-channel and R-channel common zero-flag pin. If the data for L-channel and R-channel remains at a 0 level for 1024 sampling periods (or LRCK clock periods), ZEROA is set to a logic-1 state.

#### **HARDWARE CONTROL (PCM1781)**

The digital functions of the PCM1781 are capable of hardware control. Table 2 shows selectable formats, Table 3 shows de-emphasis control, and Table 4 shows muting control.

**Table 2. Data Format Selection** 

FMT (PIN 1)	DATA FORMAT
LOW	16- to 24-bit, I <sup>2</sup> S format
HIGH	16-bit right-justified

Table 3. De-Emphasis Control

DEMP1 (PIN 3)	DEMP0 (PIN 2)	DE-EMPHASIS FUNCTION
LOW	LOW	OFF
LOW	HIGH	48-kHz de-emphasis ON
HIGH	LOW	44.1-kHz de-emphasis ON
HIGH	HIGH	32-kHz de-emphasis ON

**Table 4. Mute Control** 

MUTE (PIN 4)	MUTE STATUS
LOW	Mute OFF
HIGH	Mute ON



## **SOFTWARE CONTROL (PCM1780/82)**

The PCM1780/82 has many programmable functions that can be controlled in the software control mode; the functions are controlled by programming the internal registers using MS, MC, and MD.

The serial control interface is a 3-wire serial port which operates asynchronously to the audio serial interface. The serial control interface is used to program the on-chip mode registers. The control interface includes MD (pin 4), MC (pin 3), and  $\overline{\text{MS}}$  (pin 2). MD is the serial data input, used to program the mode registers. MC is the serial bit clock, used to shift data into the control port.  $\overline{\text{MS}}$  is the select input, used to enable the mode control port.

#### **Register Write Operation**

All write operations for the serial control port use 16-bit data words. Figure 24 shows the control data word format. The most significant bit must be a 0. Seven bits, labeled IDX[6:0], set the register index (or address) for the write operation. The least significant eight bits, D[7:0], contain the data to be written to the register specified by IDX[6:0].

Figure 25 shows the functional timing diagram for writing to the serial control port.  $\overline{\text{MS}}$  is held at a logic-1 state until a register needs to be written. To start the register write cycle,  $\overline{\text{MS}}$  is set to logic 0. Sixteen clocks are then provided on MC, corresponding to the 16 bits of the control data word on MD. After the sixteenth clock cycle has completed,  $\overline{\text{MS}}$  is set to logic 1 to latch the data into the indexed mode control register.

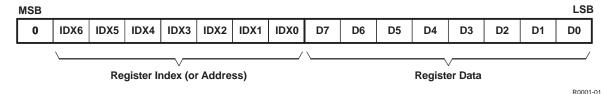


Figure 24. Control Data Word Format for MD

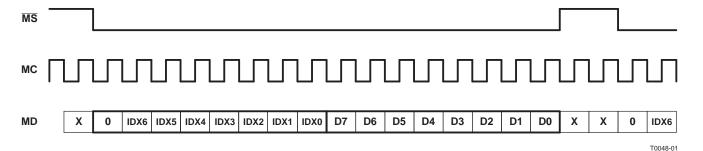
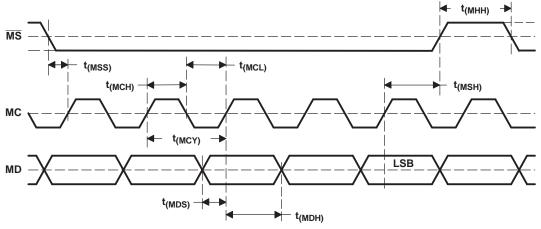


Figure 25. Register Write Operation

#### **Control Interface Timing Requirements**

Figure 26 shows a detailed timing diagram for the serial control interface. These timing parameters are critical for proper control port operation.





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	PARAMETER	MIN	UNIT
t <sub>(MCY)</sub>	MC pulse cycle time	100	ns
t <sub>(MCL)</sub>	MC low-level time	50	ns
t <sub>(MCH)</sub>	MC high-level time	50	ns
t <sub>(MHH)</sub>	MS high-level time	$3/(256 \times f_S)^{(2)}$	ns
t <sub>(MSS)</sub>	MS falling edge to MC rising edge	20	ns
t <sub>(MSH)</sub>	MS hold time <sup>(1)</sup>	20	ns
t <sub>(MDH)</sub>	MD hold time	15	ns
t <sub>(MDS)</sub>	MD setup time	20	ns

- (1) MC rising edge for LSB to  $\overline{\text{MS}}$  rising edge
- (2) f<sub>S</sub>: sampling rate

Figure 26. Control Interface Timing



## **MODE CONTROL REGISTERS (PCM1780/82)**

#### **User-Programmable Mode Controls**

The PCM1780/82 includes a number of user-programmable functions, which are accessed via control registers. The registers are programmed using the serial control interface, which was previously discussed in the *Software Control* section of this data sheet. Table 5 lists the available mode control functions, along with their reset default conditions and associated register index.

## **Register Map**

The mode control register map is shown in Table 6. Each register includes an index (or address) indicated by the IDX[6:0] bits.

**Table 5. User-Programmable Mode Controls** 

FUNCTION	RESET DEFAULT	REGISTER	BIT(S)
Digital attenuation control, 0 dB to -63 dB in 0.5-dB steps	0 dB, no attenuation	16 and 17	AT1[7:0], AT2[7:0]
Soft mute control	Mute disabled	18	MUT[2:0]
Oversampling rate control	x64, x32, x16	18	OVER
Soft reset control	Reset disabled	18	SRST
DAC operation control	DAC1 and DAC2 enabled	19	DAC[2:1]
De-emphasis function control	De-emphasis disabled	19	DM12
De-emphasis sample rate selection	44.1 kHz	19	DMF[1:0]
Audio data format control	24-bit, left-justified	20	FMT[2:0]
Digital filter rolloff control	Sharp rolloff	20	FLT
Digital attenuation mode select	0 to -63 dB, 0.5 dB/step	21	DAMS
Output phase select	Normal Phase	22	DREV
Zero-flag polarity select	High	22	ZREV
Zero-flag function select	L-, R-channels independent	22	AZRO

## **Table 6. Mode Control Register Map**

IDX (B8-B14)	REGIST ER	B15	B14	B13	B12	B11	B10	В9	В8	В7	В6	В5	В4	В3	B2	B1	В0
10h	16	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	AT17	AT16	AT15	AT14	AT13	AT12	AT11	AT10
11h	17	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	AT27	AT26	AT25	AT24	AT23	AT22	AT21	AT20
12h	18	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	SRST	OVER	RSV	RSV	RSV	RSV	MUT2	MUT1
13h	19	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	RSV	DMF1	DMF0	DM12	RSV	RSV	DAC2	DAC1
14h	20	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	RSV	RSV	FLT	RSV	RSV	FMT2	FMT1	FMT0
15h	21	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	RSV	DAMS						
16h	22	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	RSV	RSV	RSV	RSV	0	AZRO	ZREV	DREV



#### **Register Definitions**

	B15	B14	B13	B12	B11	B10	B9	В8	B7	B6	B5	B4	В3	B2	B1	B0
REGISTER 16	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	AT17	AT16	AT15	AT14	AT13	AT12	AT11	AT10
	B15	B14	B13	B12	B11	B10	В9	В8	B7	В6	B5	B4	В3	B2	B1	В0
REGISTER 17	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	AT27	AT26	AT25	AT24	AT23	AT22	AT21	AT20

## ATx[7:0]: Digital Attenuation Level Setting

Where x = 1 or 2, corresponding to the DAC output  $V_{OUT}L$  (x = 1) and  $V_{OUT}R$  (x = 2)

Default value: 1111 1111b

ATx[7:0]	DECIMAL VALUE	ATTENUATION	LEVEL SETTING
		DAMS = 0	DAMS = 1
1111 1111b	255	0 dB, no attenuation (default)	0 dB, no attenuation (default)
1111 1110b	254	−0.5 dB	-1 dB
1111 1101b	253	-1 dB	−2 dB
:	÷	· ·	:
1001 1100b	156	-49.5 dB	-99 dB
1001 1011b	155	-50 dB	-100 dB
1001 1010b	154	−50.5 dB	Mute
:	:		:
1000 0010b	130	−62.5 dB	Mute
1000 0001b	129	–63 dB	Mute
1000 0000b	128	Mute	Mute
:	:	:	:
0000 0000b	0	Mute	Mute

Each DAC channel ( $V_{OUT}L$  or  $V_{OUT}R$ ) includes a digital attenuation function. The attenuation level can be set from 0 dB to R dB, in S-dB steps. Changes in attenuator levels are made by incrementing or decrementing by one step (S dB) every  $8/f_S$  time internal until the programmed attenuator setting is reached. Alternatively, the attenuation level can be set to infinite attenuation (or mute).

R (range) and S (step) are -63 and 0.5 for DAMS = 0 and -100 and 1 for DAMS = 1, respectively. The DAMS bit is defined in register 21.

The attenuation data for each channel can be set individually. The attenuation level can be calculated using the following formula:

Attenuation level (dB) =  $S \bullet (ATx[7:0]_{DEC} - 255)$ 

where  $ATx[7:0]_{DEC} = 0$  through 255. For  $ATx[7:0]_{DEC} = 0$  through 128 with DAMS = 0, or for  $ATx[7:0]_{DEC} = 0$  through 154 with DAMS = 1, the attenuation level is set to infinite attenuation (mute).



	B15	B14	B13	B12	B11	B10	В9	B8	В7	В6	B5	B4	В3	B2	B1	В0
REGISTER 18	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	SRST	OVER	RSV	RSV	RSV	RSV	MUT2	MUT1
				NOT	E: RS\	/ indica	tes a re	eservec	bit that	should b	e set to	0.				<u> </u>

#### **MUTx: Soft Mute Control**

Where x = 1 or 2, corresponding to the DAC output  $V_{OUT}L$  (x = 1) and  $V_{OUT}R$  (x = 2).

#### Default value: 0

MUTx = 0	Mute disabled (default)
MUTx = 1	Mute enabled

The mute bits, MUT1 and MUT2, are used to enable or disable the soft mute function for the corresponding DAC outputs,  $V_{OUT}L$  and  $V_{OUT}R$ . The soft mute function is incorporated into the digital attenuators. When mute is disabled (MUTx = 0), the attenuator and DAC operate normally. When mute is enabled by setting MUTx = 1, the digital attenuator for the corresponding output is decreased from the current setting to infinite attenuation, one attenuator step (S dB) for every 8/fS seconds. This provides pop-free muting of the DAC output. The step size, S, is 0.5 dB for DAMS = 0 and 1 dB for DAMS = 1.

By setting MUTx = 0, the attenuator is increased one step for every 8/f<sub>S</sub> seconds to the previously programmed attenuation level.

#### **OVER: Oversampling Rate Control**

Default value: 0

System clock frequency = 512 f<sub>S</sub>, 768 f<sub>S</sub>, or 1152 f<sub>S</sub>

```
OVER = 0 ×64 oversampling (default)

OVER = 1 ×128 oversampling (applicable only if sampling clock frequency ≤ 24 kHz)
```

#### System clock frequency = 256 f<sub>S</sub> or 384 f<sub>S</sub>

```
OVER = 0 ×32 oversampling (default)

OVER = 1 ×64 oversampling (applicable only if sampling clock frequency ≤ 48 kHz)
```

#### System clock frequency = 128 f<sub>S</sub>, 192 f<sub>S</sub>

```
OVER = 0 ×16 oversampling (default)

OVER = 1 ×32 oversampling (applicable only if sampling clock frequency ≤ 96 kHz)
```

The OVER bit is used to control the oversampling rate of the delta-sigma D/A converters.

Setting OVER = 1 is recommended under the following conditions:

- System clock frequency = 512 f<sub>S</sub>, 768 f<sub>S</sub>, or 1152 f<sub>S</sub>, and sampling clock frequency ≤ 24 kHz
- System clock frequency = 256 f<sub>S</sub> or 384 f<sub>S</sub> and sampling clock frequency ≤ 48 kHz
- System clock frequency = 128 f<sub>S</sub> or 192 f<sub>S</sub> and sampling clock frequency ≤ 96 kHz

#### **SRST: Reset**

#### Default value: 0

SRST = 0	Reset disabled (default)
SRST = 1	Reset enabled

The SRST bit is used to enable or disable the soft reset function. The operation is the same as for the power-on-reset function with the exception of the reset period, which is 1024 system clocks for the SRST function. All registers are initialized.



	B15	B14	B13	B12	B11	B10	В9	B8	B7	В6	B5	B4	В3	B2	B1	В0
REGISTER 19	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	RSV	DMF1	DMF0	DM12	RSV	RSV	DAC2	DAC1
				NOT	E: RS\	/ indica	tes a re	eserved	bit that	should b	e set to	0.				

#### **DACx: DAC Operation Control**

Where x = 1 or 2, corresponding to the DAC output  $V_{OUT}L$  (x = 1) or  $V_{OUT}R$  (x = 2).

Default value: 0

DACx = 0	DAC operation enabled (default)
DACx = 1	DAC operation disabled

The DAC operation controls are used to enable and disable the DAC outputs,  $V_{OUT}L$  and  $V_{OUT}R$ . When DACx = 0, the corresponding output generates the audio waveform dictated by the data present on the DATA pin. When DACx = 1, the corresponding output is set to the dc common voltage ( $V_{COM}$ ), equal to  $V_{CC}/2$ .

## **DM12: Digital De-Emphasis Function Control**

Default value: 0

DM12 = 0	De-emphasis disabled (default)
DM12 = 1	De-emphasis enabled

The DM12 bit is used to enable or disable the digital de-emphasis function. Refer to the plots shown in the *Typical Performance Curves* section of this data sheet.

## DMF[1:0]: Sampling Frequency Selection for the De-Emphasis Function

Default value: 00

DMF[1:0]	De-Emphasis Sample Rate Selection
00	44.1 kHz (default)
01	48 kHz
10	32 kHz
11	Reserved

The DMF[1:0] bits are used to select the sampling frequency used for the digital de-emphasis function when it is enabled.



	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	В3	B2	B1	В0
REGISTER 20	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	RSV	RSV	FLT	RSV	RSV	FMT2	FMT1	FMT0

NOTE: RSV indicates a reserved bit that should be set to 0.

#### FMT[2:0]: Audio Interface Data Format

Default value: 101

FMT[2:0]	Audio Data Format Selection
000	24-bit right-justified format, standard data
001	20-bit right-justified format, standard data
010	18-bit right-justified format, standard data
011	16-bit right-justified format, standard data
100	I <sup>2</sup> S format, 16 to 24 bits
101	Left-justified format, 16 to 24 bits (default)
110	Reserved
111	Reserved

The FMT[2:0] bits are used to select the data format for the serial audio interface. The preceding table shows the available format options.

## **FLT: Digital Filter Rolloff Control**

Default value: 0

FLT = 0	Sharp rolloff (default)
FLT = 1	Slow rolloff

The FLT bit allows users to select the digital filter rolloff that is best suited to their application. Two filter rolloff selections are available: sharp and slow. The filter responses for these selections are shown in the *Typical Performance Curves* section of this data sheet

	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	В0
REGISTER 21	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	RSV	RSV	RSV	RSV	RSV	RSV	RSV	DAMS
				NOT	E: RS\	/ indica	ites a re	eserved	bit that	should b	e set to	0.	•			

## **DAMS: Digital Attenuation Mode Select**

Default value: 0

DAMS = 0	Fine step, 0.5 dB/step for 0 to -63 dB range (default)
DAMS = 1	Wide range, 1 dB/step for 0 to -100 dB range

The DAMS bit is used to select the digital attenuation mode.

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	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	В3	B2	B1	В0
REGISTER 22	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	RSV	RSV	RSV	RSV	0	AZRO	ZREV	DREV

NOTE: RSV indicates a reserved bit that should be set to 0.

# **DREV: Output Phase Select**

Default value: 0

DREV = 0	Normal output (default)
DREV = 1	Inverted output

The DREV bit allows the user to control the phase of the analog output signal.

# **ZREV: Zero-Flag Polarity Select**

Default value: 0

ZREV = 0	Zero-flag pins HIGH at a zero detect (default)
ZREV = 1	Zero-flag pins LOW at a zero detect

The ZREV bit allows the user to select the polarity of the zero-flag pins.

## **AZRO: Zero Flag Function Select**

Default value: 0

AZRO = 0	Pin 1: ZEROL; zero-flag output for L-channel
	Pin 16: ZEROR; zero-flag output for R-channel
AZRO = 1	Pin 1: NA; not assigned
	Pin 16: ZEROA; zero-flag output for L-/R-channel

The AZRO bit allows the user to select the function of zero-flag pins.



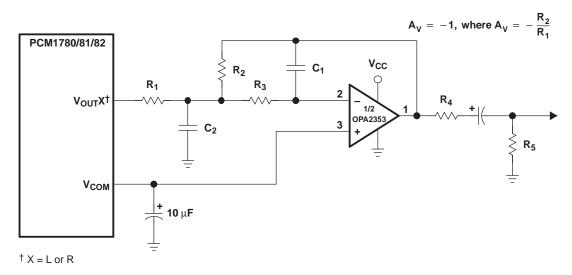
#### **ANALOG OUTPUTS**

The PCM1780/81/82 includes two independent output channels,  $V_{OUT}L$  and  $V_{OUT}R$ . These are unbalanced outputs, each capable of driving 3.9 Vp-p typical into a 5-k $\Omega$  ac-coupled load. The internal output amplifiers for  $V_{OUT}L$  and  $V_{OUT}R$  are biased to the dc common voltage, equal to 0.5  $V_{CC}$ .

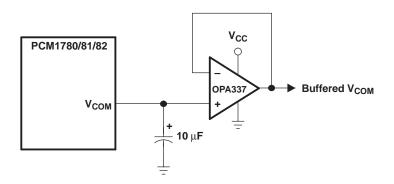
The output amplifiers include an RC continuous-time filter, which helps to reduce the out-of-band noise energy present at the DAC outputs due to the noise-shaping characteristics of the PCM1780/81/82 delta-sigma D/A converters. The frequency response of this filter is shown as ANALOG FILTER in the *Typical Performance Curves* section. By itself, this filter is not enough to attenuate the out-of-band noise to an acceptable level for many applications. An external low-pass filter is required to provide sufficient out-of-band noise rejection. Further discussion of DAC post-filter circuits is provided in the *Application Information* section of this data sheet.

## **V<sub>COM</sub> Output**

One unbuffered common voltage output pin,  $V_{COM}$  (pin 13), is brought out for decoupling purposes. This pin is nominally biased to the dc common voltage, equal to  $V_{CC}/2$ . This pin can be used to bias external circuits. Figure 27 shows an example of using the  $V_{COM}$  pin for external biasing applications.



(a) Using V<sub>COM</sub> to Bias a Single-Supply Filter Stage



(b) Using a Voltage Follower to Buffer V<sub>COM</sub> When Biasing Multiple Nodes

S0054-01

Figure 27. Biasing External Circuits Using the V<sub>COM</sub> Pin

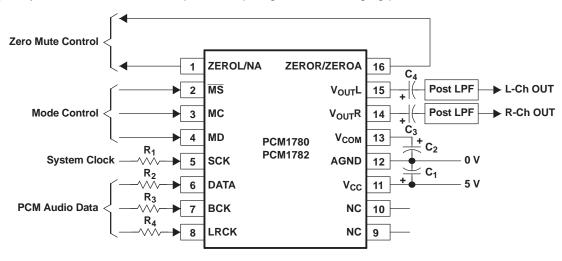


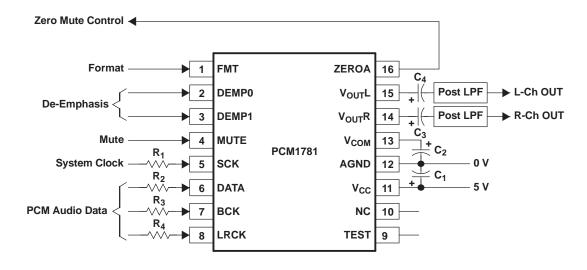
#### **APPLICATION INFORMATION**

## **Connection Diagrams**

A basic connection diagram is shown in Figure 28, with the necessary power supply bypassing and decoupling components. TI recommends using the component values shown in Figure 28 for all designs.

The use of series resistors (22  $\Omega$  to 100  $\Omega$ ) is recommended for the SCK, LRCK, BCK, and DATA inputs. The series resistor combines with the stray PCB and device input capacitance to form a low-pass filter which reduces high-frequency noise emissions and helps to dampen glitches and ringing present on clock and data lines.





 $C_1$ : 0.1- $\mu F$  Ceramic and 10- $\mu F$  Electrolytic

C2: 10-µF Electrolytic

 $C_3$ ,  $C_4$ : 4.7- $\mu$ F to 10- $\mu$ F Electrolytic

 $R_1$ - $R_4$ : 22  $\Omega$  to 100  $\Omega$ 

S0055-01

Figure 28. Basic Connection Diagram



## **APPLICATION INFORMATION (continued)**

#### **Power Supplies and Grounding**

The PCM1780/81/82 requires a 5-V supply for V<sub>CC</sub>.

Proper power supply bypassing is shown in Figure 28. The  $0.1-\mu F$  ceramic capacitor and the  $10-\mu F$  electrolytic capacitor are recommended.

## **D/A Output Filter Circuits**

Delta-sigma D/A converters use noise-shaping techniques to improve in-band signal-to-noise ratio (SNR) performance at the expense of generating increased out-of-band noise above the Nyquist frequency, or f<sub>S</sub>/2. The out-of-band noise must be low-pass filtered in order to provide optimal converter performance. This is accomplished by a combination of on-chip and external low-pass filtering.

Figure 27(a) and Figure 29 show the recommended external low-pass active filter circuits for single- and dual-supply applications. These circuits are second-order Butterworth filters using the multiple feedback (MFB) circuit arrangement, which reduces sensitivity to passive component variations over frequency and temperature. For more information regarding MFB active filter design, see *Dynamic Performance Testing of Digital Audio D/A Converters* (SBAA055), available from the TI Web site at http://www.ti.com.

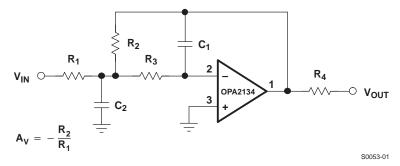


Figure 29. Dual-Supply Filter Circuit

## **PCB LAYOUT GUIDELINES**

A typical PCB floor plan for the PCM1780/81/82 is shown in Figure 30. A ground plane is recommended, with the analog and digital sections being isolated from one another using a split or cut in the circuit board. The PCM1780/81/82 should be oriented with the digital I/O pins facing the ground plane split/cut to allow for short, direct connections to the digital audio interface and control signals originating from the digital section of the board.

Separate power supplies are recommended for the digital and analog sections of the board. This prevents the switching noise present on the digital supply from contaminating the analog power supply and degrading the dynamic performance of the PCM1780/81/82. In cases where a common 5-V supply must be used for the analog and digital sections, an inductance (RF choke, ferrite bead) should be placed between the analog and digital 5-V supply connections to avoid coupling of the digital switching noise into the analog circuitry. Figure 31 shows the recommended approach for single-supply applications.



# **APPLICATION INFORMATION (continued)**

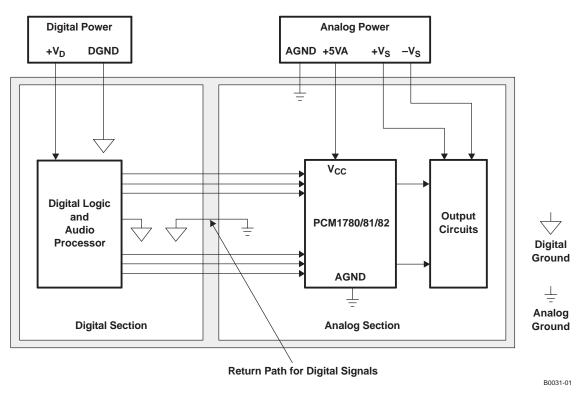


Figure 30. Recommended PCB Layout

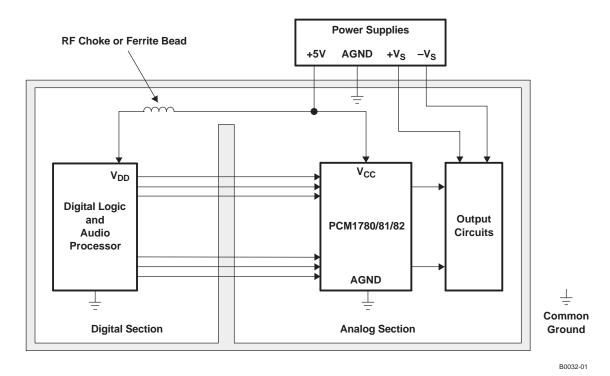


Figure 31. Single-Supply PCB Layout





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#### PACKAGING INFORMATION

Orderable Device	Status	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
PCM1780DBQ	ACTIVE	SSOP	DBQ	16	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-20 to 85	PCM1780	Samples
PCM1780DBQG4	ACTIVE	SSOP	DBQ	16	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-20 to 85	PCM1780	Samples
PCM1780DBQR	ACTIVE	SSOP	DBQ	16	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM		PCM1780	Samples
PCM1781DBQ	ACTIVE	SSOP	DBQ	16	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-25 to 85	PCM1781	Samples
PCM1781DBQG4	ACTIVE	SSOP	DBQ	16	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-25 to 85	PCM1781	Samples
PCM1781DBQR	ACTIVE	SSOP	DBQ	16	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-25 to 85	PCM1781	Samples
PCM1782DBQ	ACTIVE	SSOP	DBQ	16	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-25 to 85	PCM1782	Samples
PCM1782DBQR	ACTIVE	SSOP	DBQ	16	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-25 to 85	PCM1782	Samples

<sup>&</sup>lt;sup>(1)</sup> The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

TBD: The Pb-Free/Green conversion plan has not been defined.

**Pb-Free** (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes. **Pb-Free** (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

<sup>(2)</sup> Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

<sup>(3)</sup> MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.



## PACKAGE OPTION ADDENDUM

24-Apr-2015

- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead/Ball Finish Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

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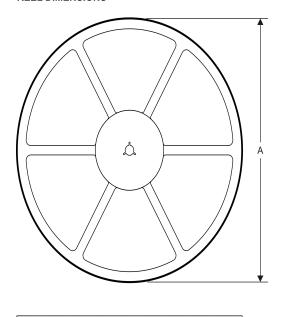
In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

# PACKAGE MATERIALS INFORMATION

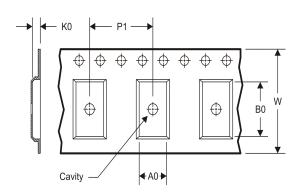
www.ti.com 16-Aug-2012

# TAPE AND REEL INFORMATION

## **REEL DIMENSIONS**



## **TAPE DIMENSIONS**



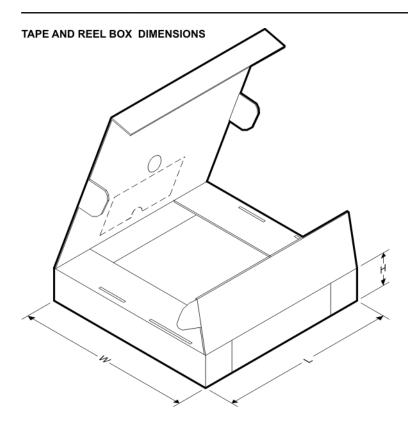
A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

## TAPE AND REEL INFORMATION

\*All dimensions are nominal

All difficultions are norminal												
Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
PCM1780DBQR	SSOP	DBQ	16	2000	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
PCM1781DBQR	SSOP	DBQ	16	2000	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
PCM1782DBQR	SSOP	DBQ	16	2000	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1

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\*All dimensions are nominal

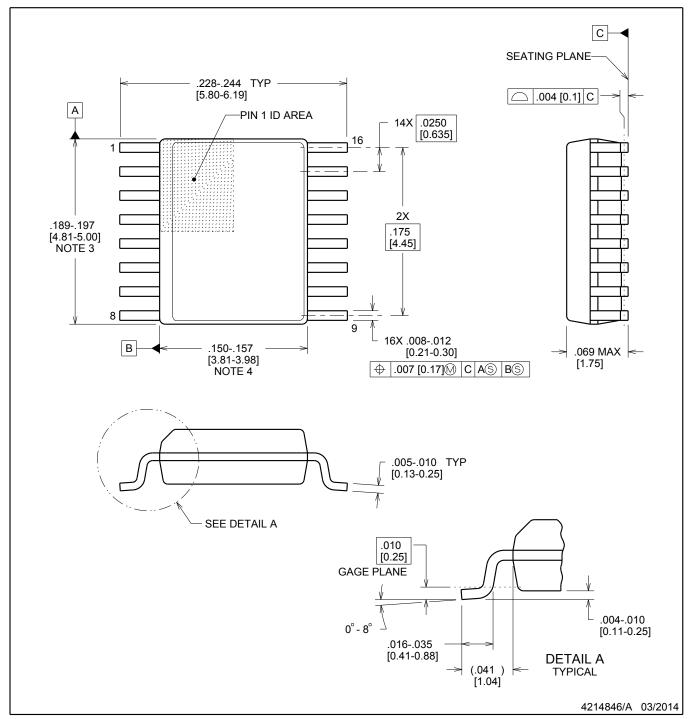
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	Device	Package Type	Package Drawing	Pins SPQ		Length (mm)	Width (mm)	Height (mm)	
	PCM1780DBQR	SSOP	DBQ	16	2000	367.0	367.0	35.0	
	PCM1781DBQR	SSOP	DBQ	16	2000	367.0	367.0	35.0	
	PCM1782DBQR	SSOP	DBQ	16	2000	367.0	367.0	35.0	



Images above are just a representation of the package family, actual package may vary. Refer to the product data sheet for package details.



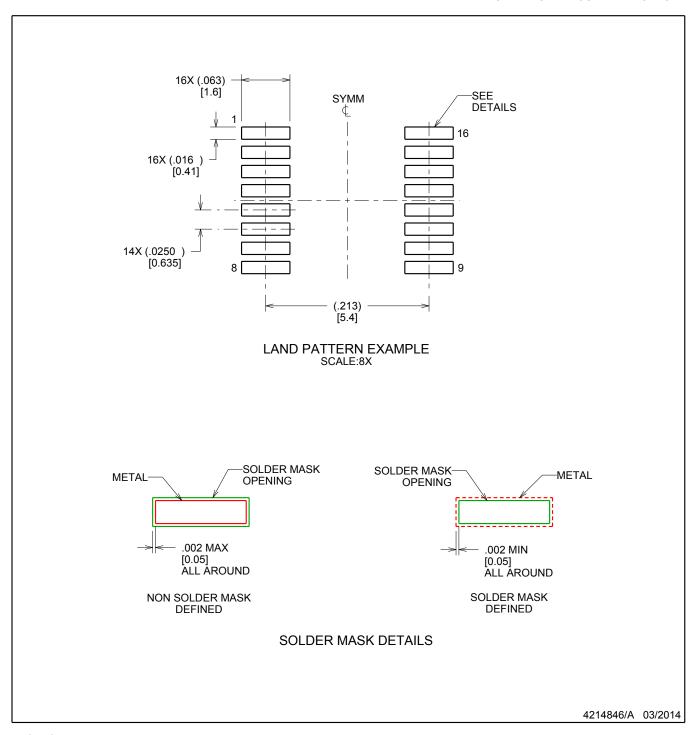




## NOTES:

- 1. Linear dimensions are in inches [millimeters]. Dimensions in parenthesis are for reference only. Controlling dimensions are in inches. Dimensioning and tolerancing per ASME Y14.5M.
- 2. This drawing is subject to change without notice.
- 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed .006 inch, per side.
- 4. This dimension does not include interlead flash.5. Reference JEDEC registration MO-137, variation AB.



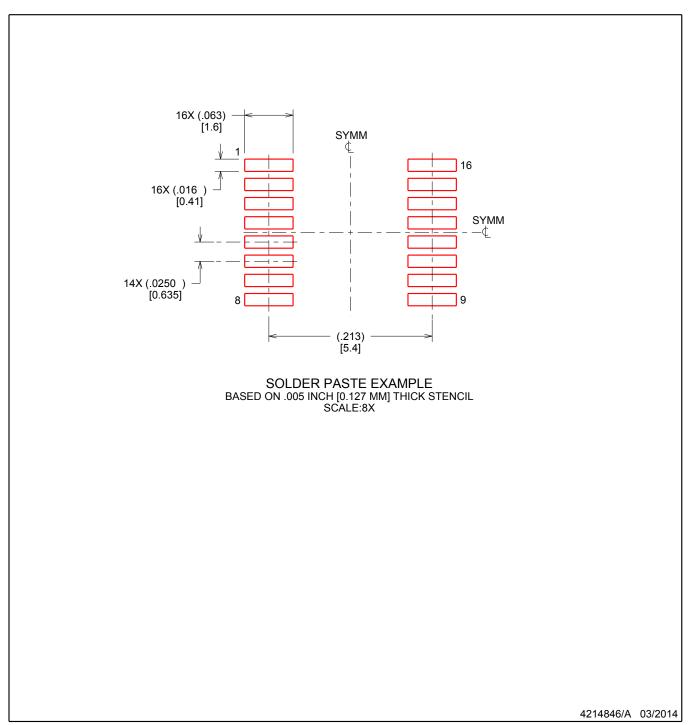


NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.





NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.



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