NSIC2050BT3G

Constant Current Regulator & LED Driver for A/C off-line Applications

120 V, 50 mA ± 15%, 3 W Package

The linear constant current regulator (CCR) is a simple, economical and robust device designed to provide a cost−effective solution for regulating current in LEDs (similar to Constant Current Diode, CCD). The CCR is based on Self−Biased Transistor (SBT) technology and regulates current over a wide voltage range. It is designed with a negative temperature coefficient to protect LEDs from thermal runaway at extreme voltages and currents.

The CCR turns on immediately and is at 20% of regulation with only 0.5 V Vak. It requires no external components allowing it to be designed as a high or low−side regulator.

The 120 V anode−cathode voltage rating is designed to withstand the high peak voltage incurred in A/C offline applications. The high anode−cathode voltage also protects surges common in Industrial and Commercial Signage applications. The CCR comes in thermally robust packages and is UL94−V0 Certified.

Features

- Robust Power Package: 2.3 W
- Wide Operating Voltage Range
- Immediate Turn-On
- Voltage Surge Suppressing − Protecting LEDs
- UL94−V0 Certified
- SBT (Self−Biased Transistor) Technology
- Negative Temperature Coefficient
- Also available in 30 mA (NSIC2030BT1G) and 20 mA (NSIC2020BT1G)
- These Devices are Pb−Free, Halogen Free/BFR Free and are RoHS Compliant

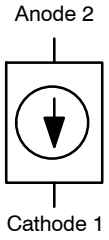
Typical Applications

- AC Lighting Panels, Display Signage, Decorative Lighting, Channel Lettering
- Application Note AND8433/D – A/C Application
- Application Note AND8493/D – A/C Capacitive Drop Design
- Design Note DN05013 – A/C Design
- Design Note DN06065 – A/C Design with PFC
- Application Notes AND8391/D, AND9008/D − Power Dissipation Considerations
- Automotive Applications − Consult Factory

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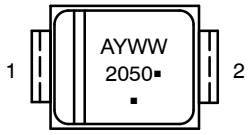
$I_{reg(SS)}$ = 50 mA
@ Vak = 7.5 V





SMB
CASE 403A

MARKING DIAGRAM



2050	= Specific Device Code
A	= Assembly Location
Y	= Year
WW	= Work Week
▪	= Pb−Free Package

(Note: Microdot may be in either location)

ORDERING INFORMATION

Device	Package	Shipping†
NSIC2050BT3G	SMB (Pb−Free)	2500 / Tape & Reel

†For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, BRD8011/D.

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MAXIMUM RATINGS (T_A = 25°C unless otherwise noted)

Rating	Symbol	Value	Unit
Anode−Cathode Voltage	Vak Max	120	V
Reverse Voltage	V_R	500	mV
Operating Junction and Storage Temperature Range	T_J, T_{stg}	−55 to +175	°C
ESD Rating: Human Body Model Machine Model	ESD	Class 3A (4000 V) Class C (400 V)	

Stresses exceeding Maximum Ratings may damage the device. Maximum Ratings are stress ratings only. Functional operation above the Recommended Operating Conditions is not implied. Extended exposure to stresses above the Recommended Operating Conditions may affect device reliability.

ELECTRICAL CHARACTERISTICS (T_A = 25°C unless otherwise noted)

Characteristic	Symbol	Min	Typ	Max	Unit
Steady State Current @ Vak = 7.5 V (Note 1)	$I_{reg(SS)}$	42.5	50	57.5	mA
Voltage Overhead (Note 2)	$V_{overhead}$		1.8		V
Pulse Current @ Vak = 7.5 V (Note 3)	$I_{reg(P)}$	48.1	57.4	66.7	mA

1. $I_{reg(SS)}$ steady state is the voltage (Vak) applied for a time duration ≥ 80 sec, using 100 mm^2 , 1 oz. Cu (or equivalent), in still air.
2. $V_{overhead} = V_{in} - V_{LEDs}$. $V_{overhead}$ is typical value for 80% $I_{reg(SS)}$.
3. $I_{reg(P)}$ non−repetitive pulse test. Pulse width t ≤ 360 μsec.

THERMAL CHARACTERISTICS

Characteristic	Symbol	Max	Unit
Total Device Dissipation (Note 1) T_A = 25°C Derate above 25°C	P_D	1210 8.0	mW mW/°C
Thermal Resistance, Junction−to−Ambient (Note 1)	$R_{\theta JA}$	124	°C/W
Thermal Reference, Junction−to−Tab (Note 1)	$R_{\psi JL}$	17.5	°C/W
Total Device Dissipation (Note 2) T_A = 25°C Derate above 25°C	P_D	1282 8.5	mW mW/°C
Thermal Resistance, Junction−to−Ambient (Note 2)	$R_{\theta JA}$	117	°C/W
Thermal Reference, Junction−to−Tab (Note 2)	$R_{\psi JL}$	18.2	°C/W
Total Device Dissipation (Note 3) T_A = 25°C Derate above 25°C	P_D	1667 11.1	mW mW/°C
Thermal Resistance, Junction−to−Ambient (Note 3)	$R_{\theta JA}$	90	°C/W
Thermal Reference, Junction−to−Tab (Note 3)	$R_{\psi JL}$	16.4	°C/W
Total Device Dissipation (Note 4) T_A = 25°C Derate above 25°C	P_D	1765 11.8	mW mW/°C
Thermal Resistance, Junction−to−Ambient (Note 4)	$R_{\theta JA}$	85	°C/W
Thermal Reference, Junction−to−Tab (Note 4)	$R_{\psi JL}$	16.7	°C/W
Total Device Dissipation (Note 5) T_A = 25°C Derate above 25°C	P_D	1948 13	mW mW/°C
Thermal Resistance, Junction−to−Ambient (Note 5)	$R_{\theta JA}$	77	°C/W
Thermal Reference, Junction−to−Tab (Note 5)	$R_{\psi JL}$	15.5	°C/W
Total Device Dissipation (Note 6) T_A = 25°C Derate above 25°C	P_D	2055 12.7	mW mW/°C
Thermal Resistance, Junction−to−Ambient (Note 6)	$R_{\theta JA}$	73	°C/W
Thermal Reference, Junction−to−Tab (Note 6)	$R_{\psi JL}$	15.6	°C/W
Total Device Dissipation (Note 7) T_A = 25°C Derate above 25°C	P_D	2149 14.3	mW mW/°C
Thermal Resistance, Junction−to−Ambient (Note 7)	$R_{\theta JA}$	69.8	°C/W
Thermal Reference, Junction−to−Tab (Note 7)	$R_{\psi JL}$	14.8	°C/W
Total Device Dissipation (Note 8) T_A = 25°C Derate above 25°C	P_D	2269 15.1	mW mW/°C
Thermal Resistance, Junction−to−Ambient (Note 8)	$R_{\theta JA}$	66.1	°C/W
Thermal Reference, Junction−to−Tab (Note 8)	$R_{\psi JL}$	14.8	°C/W
Total Device Dissipation (Note 9) T_A = 25°C Derate above 25°C	P_D	2609 17.4	mW mW/°C
Thermal Resistance, Junction−to−Ambient (Note 9)	$R_{\theta JA}$	57.5	°C/W
Thermal Reference, Junction−to−Tab (Note 9)	$R_{\psi JL}$	13.9	°C/W
Total Device Dissipation (Note 10) T_A = 25°C Derate above 25°C	P_D	2500 16.7	mW mW/°C
Thermal Resistance, Junction−to−Ambient (Note 10)	$R_{\theta JA}$	60	°C/W
Thermal Reference, Junction−to−Tab (Note 10)	$R_{\psi JL}$	16	°C/W
Total Device Dissipation (Note 11) T_A = 25°C Derate above 25°C	P_D	3000 20	mW mW/°C
Thermal Resistance, Junction−to−Ambient (Note 11)	$R_{\theta JA}$	50	°C/W
Thermal Reference, Junction−to−Tab (Note 11)	$R_{\psi JL}$	16	°C/W

NOTE: Lead measurements are made by non−contact methods such as IR with treated surface to increase emissivity to 0.9.
Lead temperature measurement by attaching a T/C may yield values as high as 30% higher °C/W values based upon empirical measurements and method of attachment.

1. 100 mm^2, 1 oz. Cu, still air.
2. 100 mm^2, 2 oz. Cu, still air.
3. 300 mm^2, 1 oz. Cu, still air.
4. 300 mm^2, 2 oz. Cu, still air.
5. 500 mm^2, 1 oz. Cu, still air.
6. 500 mm^2, 2 oz. Cu, still air.
7. 700 mm^2, 1 oz. Cu, still air.
8. 700 mm^2, 2 oz. Cu, still air.
9. 1000 mm^2, 3 oz. Cu, still air.
10. 400 mm^2, PCB is DENKA K1, 1.5 mm Al, 2kV Thermally conductive dielectric, 2 oz. Cu, or equivalent, still air.
11. 900 mm^2, PCB is DENKA K1, 1.5 mm Al, 2kV Thermally conductive dielectric, 2 oz. Cu, or equivalent, still air.

TYPICAL PERFORMANCE CURVES

(Minimum FR−4 @ 100 mm^2, 1 oz. Copper Trace, Still Air)

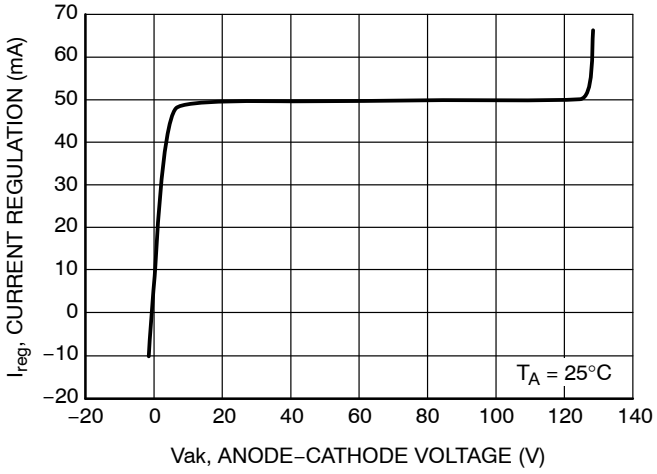


Figure 1. General Performance Curve for CCR

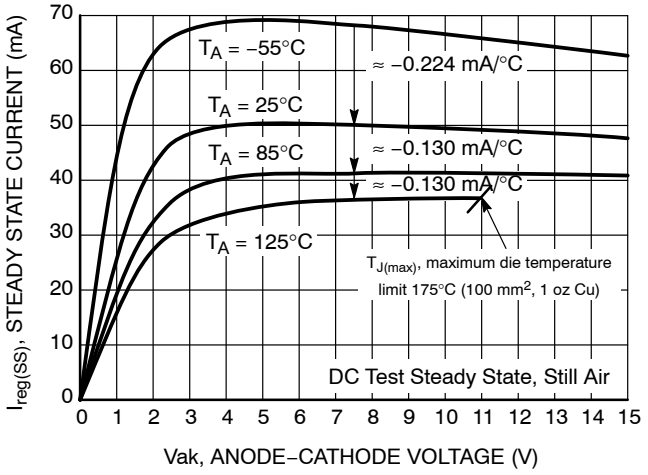


Figure 2. Steady State Current ($I_{reg(SS)}$) vs. Anode−Cathode Voltage (Vak)

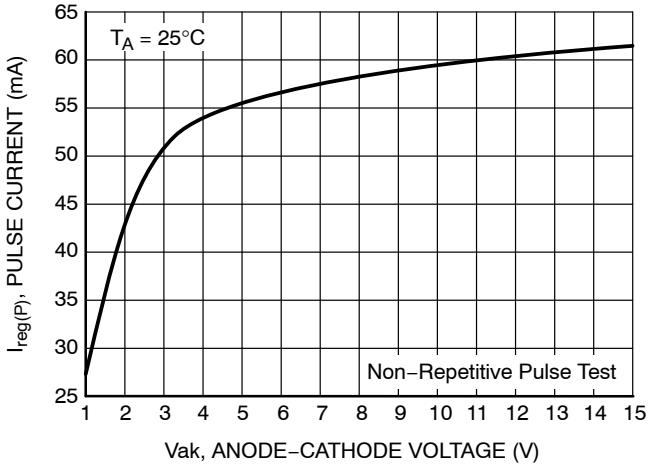


Figure 3. Pulse Current ($I_{reg(P)}$) vs. Anode−Cathode Voltage (Vak)

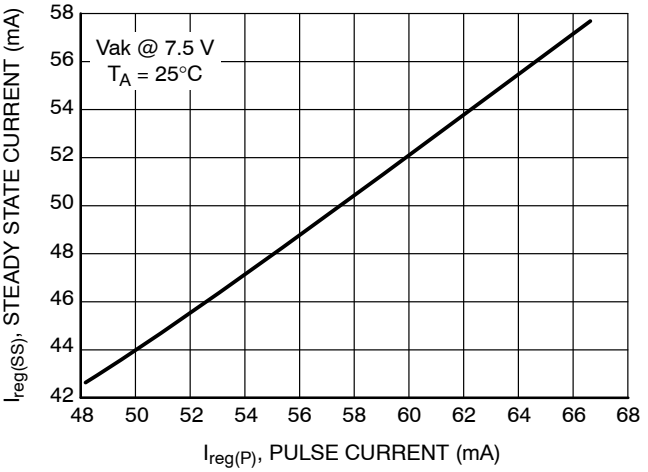


Figure 4. Steady State Current vs. Pulse Current Testing

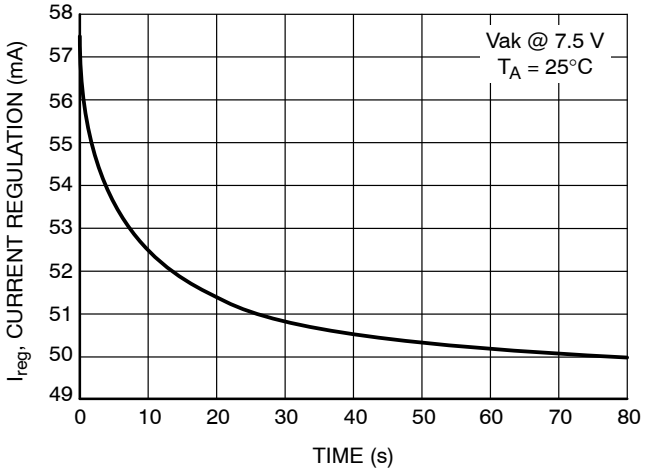


Figure 5. Current Regulation vs. Time

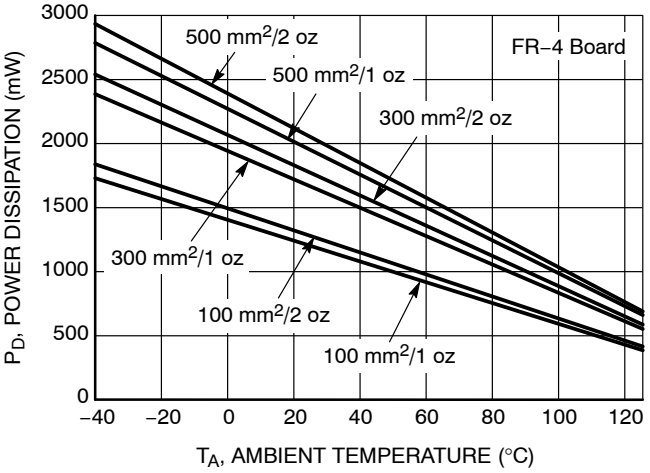


Figure 6. Power Dissipation vs. Ambient Temperature @ T_J = 175°C: Small Footprint

TYPICAL PERFORMANCE CURVES

(Minimum FR−4 @ 100 mm^2, 1 oz. Copper Trace, Still Air)

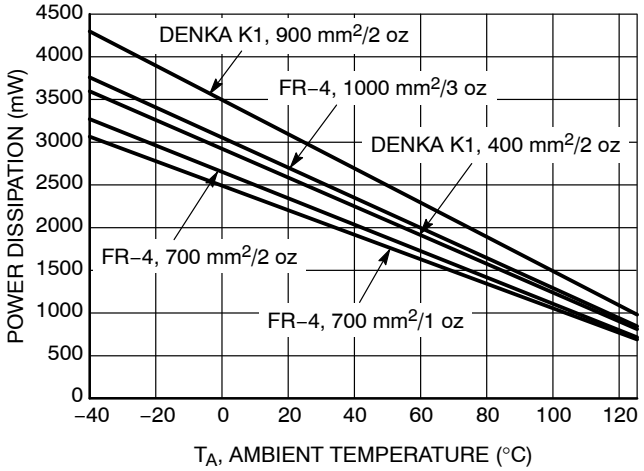


Figure 7. Power Dissipation vs. Ambient Temperature @ T_J = 175°C: Large Footprint

APPLICATIONS INFORMATION

The CCR is a self biased transistor designed to regulate the current through itself and any devices in series with it. The device has a slight negative temperature coefficient, as shown in Figure 2 – Tri Temp. (i.e. if the temperature increases the current will decrease). This negative temperature coefficient will protect the LEDS by reducing the current as temperature rises.

The CCR turns on immediately and is typically at 20% of regulation with only 0.5 V across it.

The device is capable of handling voltage for short durations of up to 120 V so long as the die temperature does not exceed 175°C. The determination will depend on the thermal pad it is mounted on, the ambient temperature, the pulse duration, pulse shape and repetition.

AC Applications

The CCR is a DC device; however, it can be used with full wave rectified AC as shown in application notes AND8433/D and AND8492/D and design notes DN05013/D and DN06065/D. Figure 8 shows the basic circuit configuration.

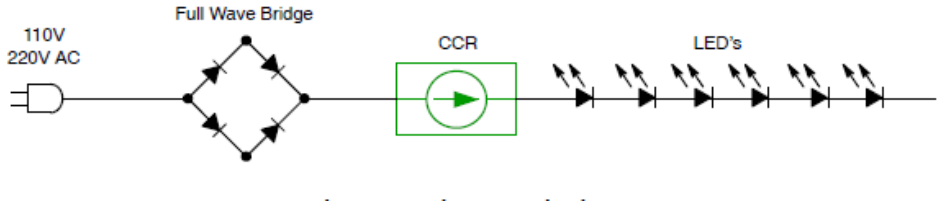


Figure 8. Basic AC Application

Single LED String

The CCR can be placed in series with LEDs as a High Side or a Low Side Driver. The number of the LEDs can vary from one to an unlimited number. The designer needs to calculate the maximum voltage across the CCR by taking the maximum input voltage less the voltage across the LED string (Figures 9 and 10).

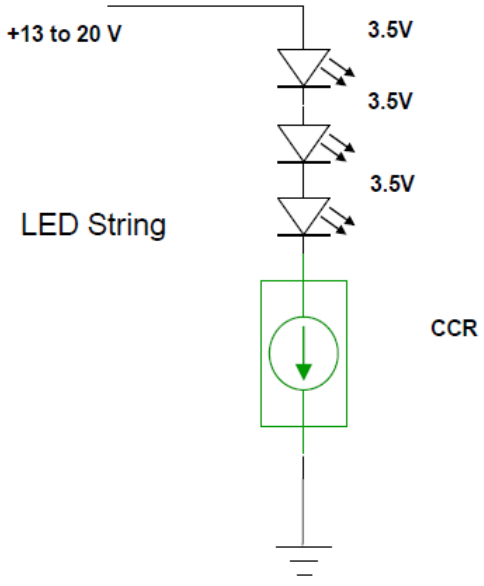


Figure 9.

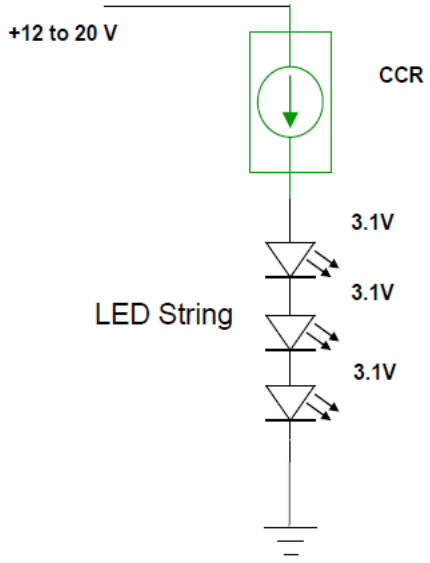


Figure 10.

Higher Current LED Strings

Two or more fixed current CCRs can be connected in parallel. The current through them is additive (Figure 11).

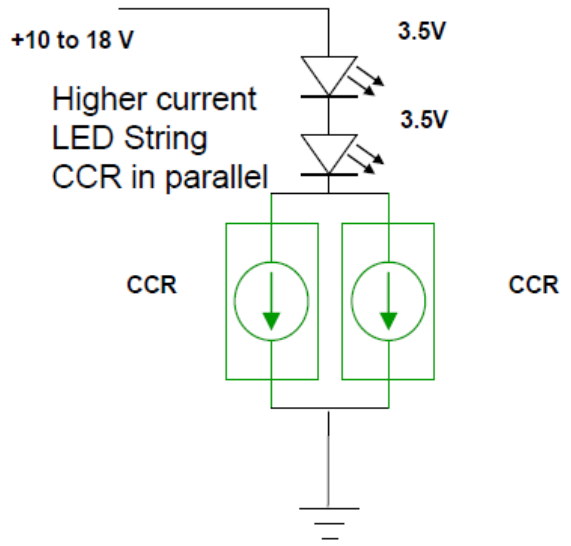


Figure 11.

Other Currents

The adjustable CCR can be placed in parallel with any other CCR to obtain a desired current. The adjustable CCR provides the ability to adjust the current as LED efficiency increases to obtain the same light output (Figure 12).

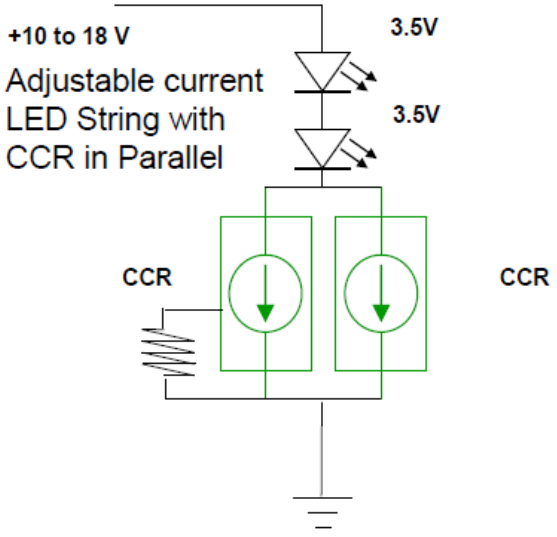


Figure 12.

Dimming using PWM

The dimming of an LED string can be easily achieved by placing a BJT in series with the CCR (Figure 13).

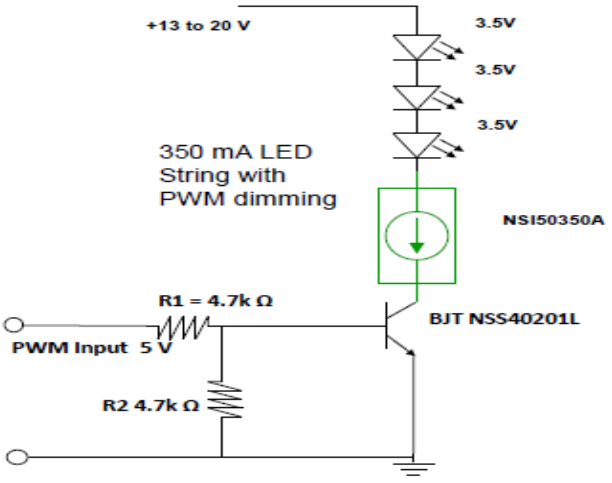


Figure 13.

The method of pulsing the current through the LEDs is known as Pulse Width Modulation (PWM) and has become the preferred method of changing the light level. LEDs being a silicon device, turn on and off rapidly in response to the current through them being turned on and off. The switching time is in the order of 100 nanoseconds, this equates to a maximum frequency of 10 Mhz, and applications will typically operate from a 100 Hz to 100 kHz. Below 100 Hz the human eye will detect a flicker from the light emitted from the LEDs. Between 500 Hz and 20 kHz the circuit may generate audible sound. Dimming is achieved by turning the LEDs on and off for a portion of a single cycle. This on/off cycle is called the Duty cycle (D) and is expressed by the amount of time the LEDs are on (Ton) divided by the total time of an on/off cycle (Ts) (Figure 14).

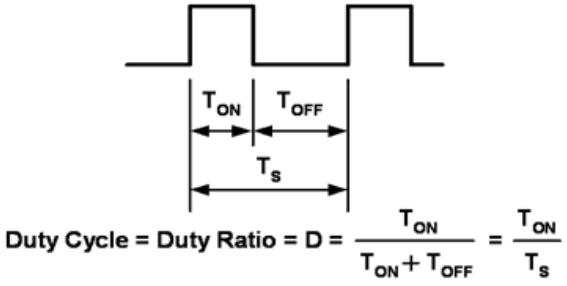


Figure 14.

The current through the LEDs is constant during the period they are turned on resulting in the light being consistent with no shift in chromaticity (color). The brightness is in proportion to the percentage of time that the LEDs are turned on.

Figure 15 is a typical response of Luminance vs Duty Cycle.

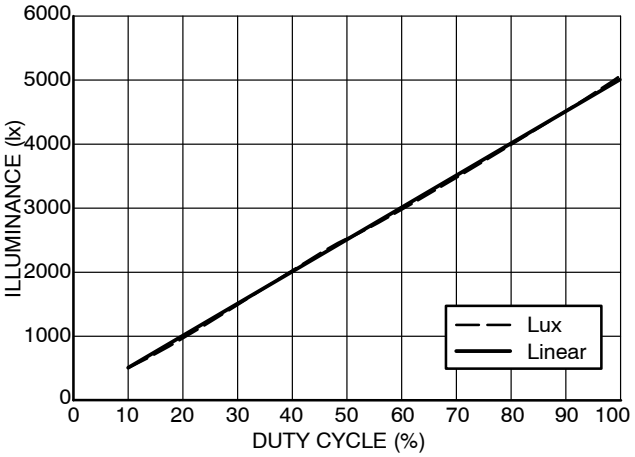


Figure 15. Luminous Emmitance vs. Duty Cycle

Reducing EMI

Designers creating circuits switching medium to high currents need to be concerned about Electromagnetic Interference (EMI). The LEDs and the CCR switch extremely fast, less than 100 nanoseconds. To help eliminate EMI, a capacitor can be added to the circuit across R2. (Figure 13) This will cause the slope on the rising and falling edge on the current through the circuit to be extended. The slope of the CCR on/off current can be controlled by the values of R1 and C1.

The selected delay / slope will impact the frequency that is selected to operate the dimming circuit. The longer the delay, the lower the frequency will be. The delay time should not be less than a 10:1 ratio of the minimum on time. The frequency is also impacted by the resolution and dimming steps that are required. With a delay of 1.5 microseconds on the rise and the fall edges, the minimum on time would be 30 microseconds. If the design called for a resolution of 100 dimming steps, then a total duty cycle time (Ts) of 3 milliseconds or a frequency of 333 Hz will be required.

Thermal Considerations

As power in the CCR increases, it might become necessary to provide some thermal relief. The maximum power dissipation supported by the device is dependent upon board design and layout. Mounting pad configuration on the PCB, the board material, and the ambient temperature affect the rate of junction temperature rise for the part. When the device has good thermal conductivity through the PCB, the junction temperature will be relatively low with high power applications. The maximum dissipation the device can handle is given by:

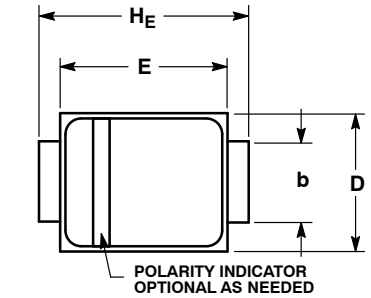
$$P_{D(MAX)} = \frac{T_{J(MAX)} - T_A}{R_{\theta JA}}$$

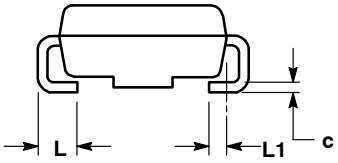
Referring to the thermal table on page 2 the appropriate $R_{\theta JA}$ for the circuit board can be selected.

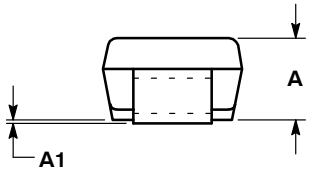
NSIC2050BT3G

PACKAGE DIMENSIONS

SMB
CASE 403A−03
ISSUE H



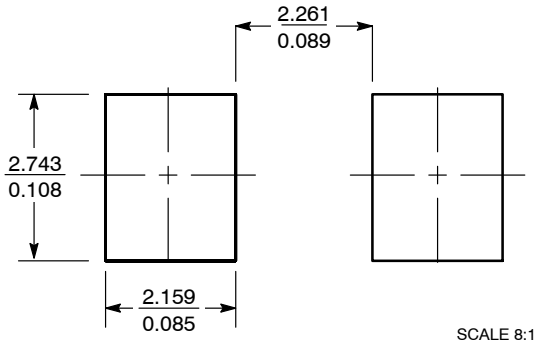




NOTES:
1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
2. CONTROLLING DIMENSION: INCH.
3. D DIMENSION SHALL BE MEASURED WITHIN DIMENSION P.

	MILLIMETERS			INCHES		
DIM	MIN	NOM	MAX	MIN	NOM	MAX
A	1.90	2.20	2.28	0.075	0.087	0.090
A1	0.05	0.10	0.19	0.002	0.004	0.007
b	1.96	2.03	2.20	0.077	0.080	0.087
c	0.15	0.23	0.31	0.006	0.009	0.012
D	3.30	3.56	3.95	0.130	0.140	0.156
E	4.06	4.32	4.60	0.160	0.170	0.181
H_E	5.21	5.44	5.60	0.205	0.214	0.220
L	0.76	1.02	1.60	0.030	0.040	0.063
L1	0.51 REF			0.020 REF		

SOLDERING FOOTPRINT*



*For additional information on our Pb−Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

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