

Normally – OFF Silicon Carbide Junction Transistor

 V_{DS} = 600 V $R_{DS(ON)}$ = 25 m Ω $I_{D (Tc = 25^{\circ}C)}$ = 100 A $h_{FE (Tc = 25^{\circ}C)}$ = 105

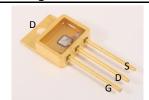
Features

- 225°C maximum operating temperature
- · Gate Oxide Free SiC Switch
- Exceptional Safe Operating Area
- Excellent Gain Linearity
- Temperature Independent Switching Performance
- Low Output Capacitance
- Positive Temperature Coefficient of RDS,ON
- Suitable for Connecting an Anti-parallel Diode

Advantages

- Compatible with Si MOSFET/IGBT Gate Drive ICs
- > 20 µs Short-Circuit Withstand Capability
- Lowest-in-class Conduction Losses
- High Circuit Efficiency
- Minimal Input Signal Distortion
- High Amplifier Bandwidth

Package





TO-258

Applications

- Down Hole Oil Drilling
- Geothermal Instrumentation
- Solenoid Actuators
- General Purpose High-Temperature Switching
- Amplifiers
- Solar Inverters
- Switched-Mode Power Supply (SMPS)
- Power Factor Correction (PFC)

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Section I: Absolute Maximum Ratings

Parameter	Symbol	Conditions	Value	Unit	Notes
Drain – Source Voltage	V_{DS}	$V_{GS} = 0 V$	600	V	
Continuous Drain Current	I _D	$T_J = 225$ °C, $T_C = 25$ °C	100	Α	Fig. 21
Continuous Gate Current	I_{GM}		3.5	Α	
Turn-Off Safe Operating Area	RBSOA	$T_J = 225$ °C, $I_G = 3.5$ A, Clamped Inductive Load	$I_{D,max} = 50$ $ V_{DS} \le V_{DSmax}$	Α	Fig. 19
Short Circuit Safe Operating Area	SCSOA	T_J = 225°C, I_G = 3.5 A, V_{DS} = 400 V, Non Repetitive	>20	μs	
Reverse Gate – Source Voltage	V _{SG}	·	30	V	
Reverse Drain – Source Voltage	V_{SD}		25	V	
Power Dissipation	P _{tot}	$T_J = 225$ °C, $T_C = 25$ °C	769	W	Fig. 16
Operating and Storage Temperature	T _{stg}		-55 to 225	°C	



Section II: Static Electrical Characteristics

Donomoton	Come le ed	Canditions	Value		I Incid	Madaa	
Parameter	Symbol	Conditions	Min.	Typical	Max.	Unit	Notes
A: On State							
Drain – Source On Resistance	R _{DS(ON)}	$\begin{split} I_D &= 50 \text{ A, } T_j = 25 \text{ °C} \\ I_D &= 50 \text{ A, } T_j = 125 \text{ °C} \\ I_D &= 50 \text{ A, } T_j = 175 \text{ °C} \\ I_D &= 50 \text{ A, } T_j = 225 \text{ °C} \end{split}$		25 39 43 62		mΩ	Fig. 5
Gate – Source Saturation Voltage	$V_{GS,SAT}$	$I_D = 50 \text{ A}, I_D/I_G = 40, T_j = 25 \text{ °C}$ $I_D = 50 \text{ A}, I_D/I_G = 30, T_j = 175 \text{ °C}$		3.42 3.23		V	Fig. 7
DC Current Gain	h _{FE}	$V_{DS} = 5 \text{ V}, I_{D} = 50 \text{ A}, T_{j} = 25 \text{ °C}$ $V_{DS} = 5 \text{ V}, I_{D} = 50 \text{ A}, T_{j} = 125 \text{ °C}$ $V_{DS} = 5 \text{ V}, I_{D} = 50 \text{ A}, T_{j} = 175 \text{ °C}$ $V_{DS} = 5 \text{ V}, I_{D} = 50 \text{ A}, T_{j} = 225 \text{ °C}$		105 77 71 69		-	Fig. 5
B: Off State							
Drain Leakage Current	I _{DSS}	$\begin{array}{l} V_R = 600 \text{ V}, V_{GS} = 0 \text{ V}, T_j = 25 \text{ °C} \\ V_R = 600 \text{ V}, V_{GS} = 0 \text{ V}, T_j = 125 \text{ °C} \\ V_R = 600 \text{ V}, V_{GS} = 0 \text{ V}, T_j = 225 \text{ °C} \end{array}$		10 50 100		μΑ	Fig. 8
Gate Leakage Current	I _{SG}	$V_{SG} = 20 \text{ V}, T_j = 25 \text{ °C}$		20		nA	
C: Thermal							
Thermal resistance, junction - case	R_{thJC}			0.26		°C/W	Fig. 20

Section III: Dynamic Electrical Characteristics

Parameter	Cumbal	ol Conditions		Value		- Unit	Notes
Parameter	Symbol	Conditions	Min.	Typical	Max.	Unit	notes
A: Capacitance and Gate Charg	е						
Input Capacitance	C _{iss}	V _{GS} = 0 V, V _D = 100 V, <i>f</i> = 1 MHz		6450		pF	Fig. Error! Reference source not found.
Reverse Transfer/Output Capacitance	C _{rss} /C _{oss}	V _D = 100 V, <i>f</i> = 1 MHz		420		pF	Fig. Error! Reference source not found.
Output Capacitance Stored Energy	E _{oss}	$V_{GS} = 0 \text{ V}, V_{D} = 400 \text{ V}, f = 1 \text{ MHz}$		17.4		μJ	Fig. Error! Reference source not found.
Effective Output Capacitance, time related	$C_{oss,tr}$	I_D = constant, V_{GS} = 0 V, V_{DS} = 0400 V		390		pF	
Effective Output Capacitance, energy related	C _{oss,er}	V _{GS} = 0 V, V _{DS} = 0400 V		284		pF	
Gate-Source Charge	Q_{GS}	V _{GS} = -53 V		55		nC	
Gate-Drain Charge	Q_{GD}	V _{GS} = 0 V, V _{DS} = 0400 V		156		nC	
Gate Charge - Total	Q_G			211		nC	
B: Switching ¹							
Internal Gate Resistance – zero bias	R _{G(INT-ZERO)}	$f = 1 \text{ MHz}, V_{AC} = 50 \text{ mV}, V_{DS} = 0 \text{ V}, V_{GS} = 0 \text{ V}, T_i = 225 ^{\circ}\text{C}$		0.9		Ω	
Internal Gate Resistance – ON	$R_{G(INT-ON)}$	$V_{GS} > 2.5 \text{ V}, V_{DS} = 0 \text{ V}, T_j = 225 {}^{\circ}\text{C}$		0.09		Ω	
Turn On Delay Time	t _{d(on)}			25		ns	
Fall Time, V _{DS}	t _f	$\begin{split} T_{J} &= 25 ^{\circ}\text{C}, \text{V}_{DS} = 400 \text{ V}, \\ I_{D} &= 50 \text{ A}, \text{ Resistive Load} \\ \text{Refer to Section V for additional} \\ \text{driving information}. \end{split}$		44		ns	Fig. Error! Reference source not found., Error! Reference source not found.
Turn Off Delay Time	$t_{d(off)}$	<u>=</u>		40		ns	



GA50JT06-258

Rise Time, V _{DS}	tr		33	ns	Fig. Error! Reference source not found., Error! Reference source not found.
Turn On Delay Time	$t_{d(on)}$		19	ns	
Fall Time, V_{DS}	t _f	$T_i = 225 ^{\circ}\text{C}, V_{DS} = 400 \text{V},$	43	ns	Fig. Error! Reference source not found.
Turn Off Delay Time	$t_{d(off)}$	I _D = 50 A, Resistive Load	89	ns	
Rise Time, V _{DS}	t _r		27	ns	Fig. Error! Reference source not found.
Turn-On Energy Per Pulse	E _{on}	T _i = 25 °C, V _{DS} = 400 V,	690	μJ	Fig. Error! Reference source not found., Error! Reference source not found.
Turn-Off Energy Per Pulse	E _{off}	I _D = 50 A, Inductive Load Refer to Section V.	359	μJ	Fig. Error! Reference source not found., Error! Reference source not found.
Total Switching Energy	E_tot		1049	μJ	
Turn-On Energy Per Pulse	E_{on}	T. 00500 V. 105V	758	μJ	Fig. Error! Reference source not found.
Turn-Off Energy Per Pulse	E _{off}	$T_J = 225$ °C, $V_{DS} = 400$ V, $I_D = 50$ A, Inductive Load	337	μJ	Fig. Error! Reference source not found.
Total Switching Energy	E_tot		1095	μJ	

 $^{^{\}rm 1}$ – All times are relative to the Drain-Source Voltage $\rm V_{\rm DS}$



Section IV: Figures

A: Static Characteristics

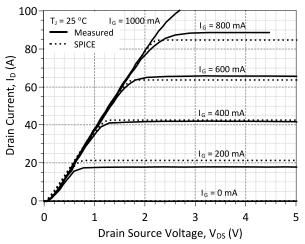


Figure 1: Typical Output Characteristics at 25 °C

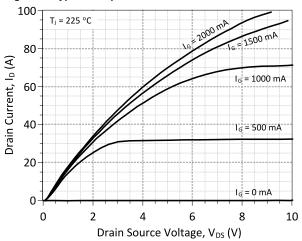


Figure 3: Typical Output Characteristics at 225 °C

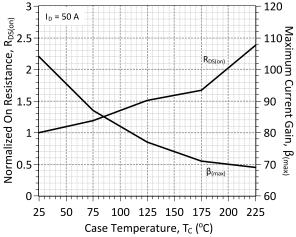


Figure 5: Normalized On-Resistance and Current Gain vs. Temperature

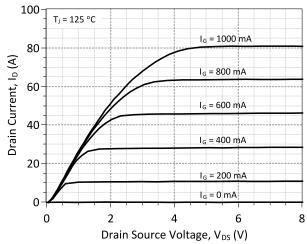


Figure 2: Typical Output Characteristics at 125 °C

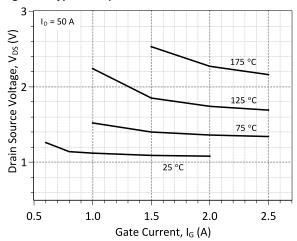


Figure 4: Drain-Source Voltage vs. Gate Current

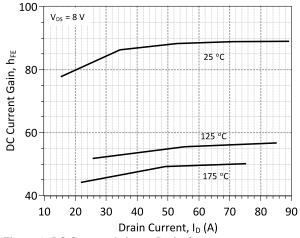


Figure 6: DC Current Gain vs. Drain Current



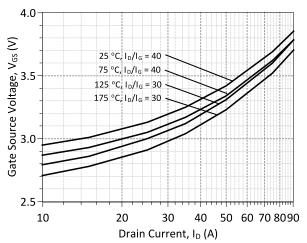


Figure 7: Typical Gate - Source Saturation Voltage

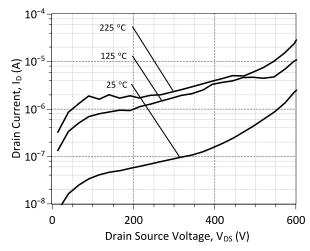


Figure 8: Typical Blocking Characteristics

B: Dynamic Characteristics

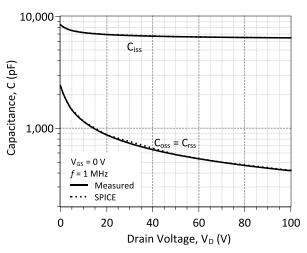


Figure 9: Input, Output, and Reverse Transfer Capacitance

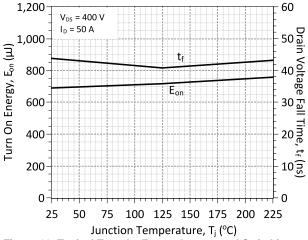


Figure 11: Typical Turn On Energy Losses and Switching Times vs. Temperature

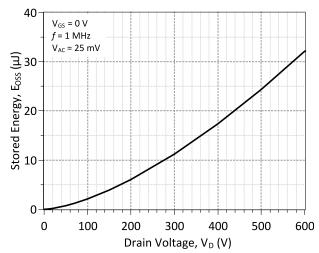


Figure 10: Output Capacitance Stored Energy

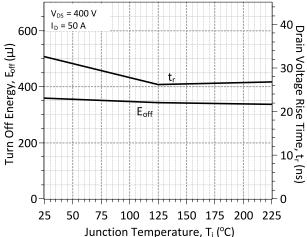


Figure 12: Typical Turn Off Energy Losses and Switching Times vs. Temperature





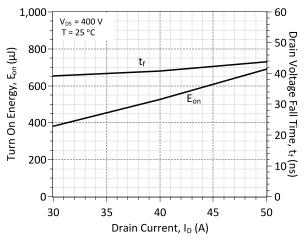


Figure 13: Typical Turn On Energy Losses and Switching Times vs. Drain Current

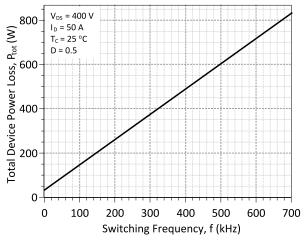


Figure 15: Typical Hard Switched Device Power Loss vs. Switching Frequency ²

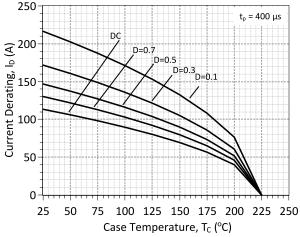


Figure 17: Drain Current Derating vs. Temperature

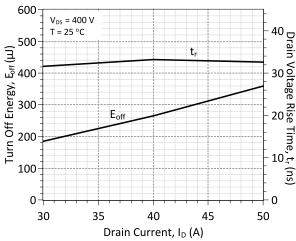


Figure 14: Typical Turn Off Energy Losses and Switching Times vs. Drain Current

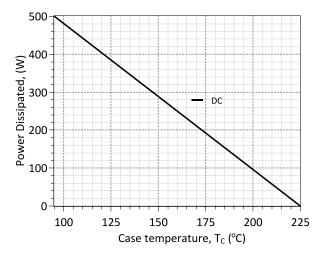


Figure 16: Power Derating Curve

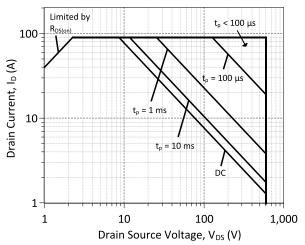


Figure 18: Forward Bias Safe Operating Area at T_c= 25 °C

² – Representative values based on device conduction and switching loss. Actual losses will depend on gate drive conditions, device load, and circuit topology.

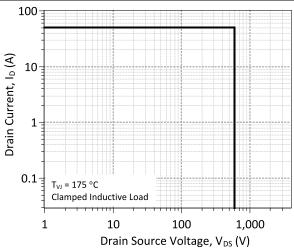


Figure 19: Turn-Off Safe Operating Area

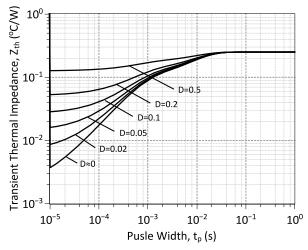


Figure 20: Transient Thermal Impedance

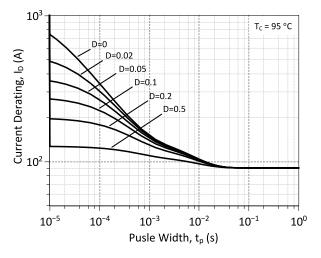


Figure 21: Drain Current Derating vs. Pulse Width



Section V: Driving the GA50JT06-247

The GA50JT06-247 is a current controlled SiC transistor which requires a positive gate current for turn-on and to remain in on-state. It may be driven by different drive topologies depending on the intended application.

Table 1: Estimated Power Consumption and switching frequencies for various Gate Drive topologies.

Drive Topology	Gate Drive Power Consumption	Switching Frequency
Simple TTL	High	Low
Constant Current	Medium	Medium
High Speed – Boost Capacitor	Medium	High
High Speed – Boost Inductor	Low	High
Proportional	Lowest	Medium
Pulsed Power	Medium	N/A

A: Simple TTL Drive

The GA50JT06-247 may be driven by 5 V TTL logic by using a simple current amplification stage. The current amplifier output current must meet or exceed the steady state gate current, $I_{G,steady}$, required to operate the GA50JT06-247. An external gate resistor R_G , shown in the Figure 22 topology, sets $I_{G,steady}$ to the required level which is dependent on the SJT drain current I_D and DC current gain h_{FE} , R_G may be calculated from the equation below. The values of h_{FE} and $V_{GS,sat}$ may be read from Figure 6 and Figure 7, respectively. $V_{EC,sat}$ can be taken from the PNP datasheet, a partial list of high-temperature PNP and NPN transistors options is given below. High-temperature MOSFETs may also be used in the topology.

$$R_{G,max} = \frac{\left(5.0 \ V - V_{EC,sat}(PNP) - V_{GS,sat}(SJT)\right) * h_{FE}(T,I_D)}{I_D * 1.5}$$

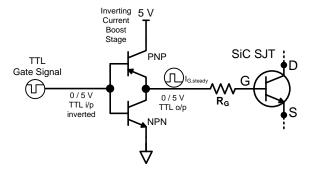


Figure 22: Simple TTL Gate Drive Topology

Table 2: Partial List of High-Temperature BJTs for TTL Gate Driving

BJT Part Number	Туре	T _{j,max} (°C)
PHPT60603PY	PNP	175
PHPT60603NY	NPN	175
2N2222	NPN	200
2N6730	PNP	200
2N2905	PNP	200
2N5883	PNP	200
2N5885	NPN	200



B: High Speed Driving

For ultra high speed GA50JT06-247 switching (t_n , t_l < 20 ns) while maintaining low gate drive losses the supplied gate current should include a positive current peak during turn-on, a negative voltage peak during turn-off, and continuous gate current I_G to remain on.

An SJT is rapidly switched from its blocking state to on-state, when the necessary gate charge for turn-on, Q_G , is supplied by a burst of high gate current until the gate-source capacitance, C_{GS} , and gate-drain capacitance, C_{GD} , are fully charged. Ideally, the burst should terminate when the drain voltage has fallen to its on-state value in order to avoid unnecessary drive losses. A negative voltage peak is recommended for the turn-off transition in order to ensure that the gate current is not being supplied under high dV/dt due to the Miller effect. While satisfactory turn off can be achieved with $V_{GS} = 0$ V, a negative V_{GS} value may be used in order to speed up the turn-off transition.

B:1: High Speed, Low Loss Drive with Boost Capacitor

The GA50JT06-247 may be driven using a High Speed, Low Loss Drive with Boost Capacitor topology in which multiple voltage levels, a gate resistor, and a gate capacitor are used to provide current peaks at turn-on and turn-off for fast switching and a continuous gate current while in on-state. As shown in Figure 23, in this topology two gate driver ICs are utilized. An external gate resistor R_G is driven by a low voltage driver to supply the continuous gate current throughout on-state.and a gate capacitor C_G is driven at a higher voltage level to supply a high current peak at turn-on and turn-off. A 3 kV isolated evaluation gate drive board (GA03IDDJT30-FR4) from GeneSiC Semiconductor utilizing this topology is commercially available for high and low-side driving, its datasheet provides additional details about this drive topology.

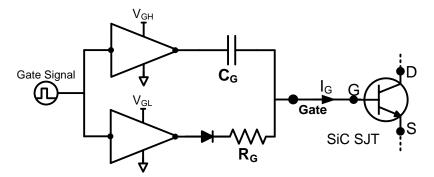


Figure 23: High Speed, Low Loss Drive with Boost Capacitor Topology

B:2: High Speed, Low Loss Drive with Boost Inductor

A High Speed, Low-Loss Driver with Boost Inductor is also capable of driving the GA50JT06-247 at high-speed. It utilizes a gate drive inductor instead of a capacitor to provide the high-current gate current pulses $I_{G,on}$ and $I_{G,off}$. During operation, inductor L is charged to a specified $I_{G,on}$ current value then made to discharge I_L into the SJT gate pin using logic control of S_1 , S_2 , S_3 , and S_4 , as shown in Figure 24. After turn on, while the device remains on the necessary steady state gate current $I_{G,steady}$ is supplied from source V_{CC} through R_G . Please refer to the article "A current-source concept for fast and efficient driving of silicon carbide transistors" by Dr. Jacek Rąbkowski for additional information on this driving topology.³

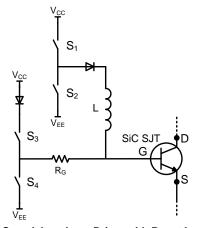


Figure 24: High Speed, Low-Loss Driver with Boost Inductor Topology

^{3 -} Archives of Electrical Engineering. Volume 62, Issue 2, Pages 333-343, ISSN (Print) 0004-0746, DOI: 10.2478/aee-2013-0026, June 2013



C: Proportional Gate Current Driving

A proportional gate drive topology may be beneficial for applications in which the GA50JT06-247 will operate over a wide range of drain current conditions to lower the gate drive power consumption. A proportional gate driver relies on instantaneous drain current I_D feedback to vary the steady state gate current $I_{G,steady}$ supplied to the GA50JT06-247.

C:1: Voltage Controlled Proportional Driver

A voltage controlled proportional driver relies on a gate drive integrated circuit to detect the GA50JT06-247 drain-source voltage V_{DS} during on-state to sense I_D . The integrated circuit will then increase or decrease I_G in response to I_D . This allows I_G and gate drive power consumption to reduce while I_D is low or for I_G to increase when I_D increases. A high voltage diode connected between the drain and sense protects the integrated circuit from high-voltage when blocking. A simplified version of this topology is shown in Figure 25. Additional information will be available in the future at http://www.genesicsemi.com/references/product-notes/.

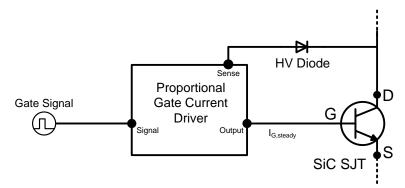


Figure 25: Simplified Voltage Controlled Proportional Driver

C:2: Current Controlled Proportional Driver

The current controlled proportional driver relies on a low-loss transformer in the drain or source path to provide feedback of the GA50JT06-247 drain current during on-state to supply $I_{G,steady}$ into the gate. $I_{G,steady}$ will increase or decrease in response to I_D at a fixed forced current gain which is set be the turns ratio of the transformer, $h_{force} = I_D / I_G = N_2 / N_1$. GA50JT06-247 is initially tuned-on using a gate current pulse supplied into an RC drive circuit to allow I_D current to begin flowing. This topology allows $I_{G,steady}$ and the gate drive power consumption to reduce while I_D is relatively low or for $I_{G,steady}$ to increase when I_D increases. A simplified version of this topology is shown in Figure 26. Additional information will be available in the future at http://www.genesicsemi.com/references/product-notes/.

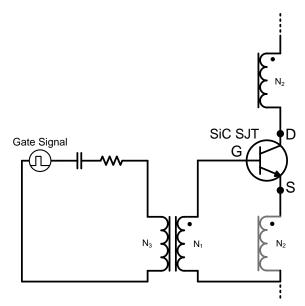
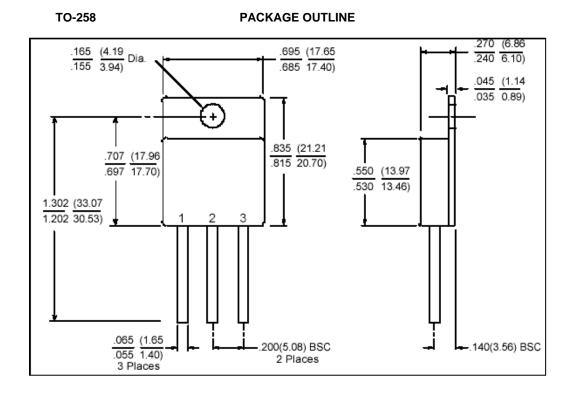


Figure 26: Simplified Current Controlled Proportional Driver



Section VI: Package Dimensions



- 1. CONTROLLED DIMENSION IS INCH. DIMENSION IN BRACKET IS MILLIMETER.
 2. DIMENSIONS DO NOT INCLUDE END FLASH, MOLD FLASH, MATERIAL PROTRUSIONS

Revision History						
Date	Revision	Comments	Supersedes			
2014/12/12	5	Updated Electrical Characteristics				
2014/08/23	4	Updated Electrical Characteristics				
2014/04/10	3	Updated Electrical Characteristics				
2014/02/05	2	Updated Electrical Characteristics				
2013/12/19	1	Updated Gate Drive Section				
2013/12/05	0	Initial release				

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Section VII: SPICE Model Parameters

This is a secure document. Please copy this code from the SPICE model PDF file on our website (http://www.genesicsemi.com/images/hit_sic/sjt/GA50JT06-258_SPICE.pdf) into LTSPICE (version 4) software for simulation of the GA50JT06-258.

```
MODEL OF GeneSiC Semiconductor Inc.
       $Revision: 1.3
       $Date: 12-DEC-2014
       GeneSiC Semiconductor Inc.
       43670 Trade Center Place Ste. 155
       Dulles, VA 20166
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* These models are provided "AS IS, WHERE IS, AND WITH NO WARRANTY
* OF ANY KIND EITHER EXPRESSED OR IMPLIED, INCLUDING BUT NOT LIMITED
* TO ANY IMPLIED WARRANTIES OF MERCHANTABILITY AND FITNESS FOR A
* PARTICULAR PURPOSE."
^{\star} Models accurate up to 2 times rated drain current.
.model GA50JT06 NPN
+ IS
              5.00E-47
+ ISE
              1.26E-26
+ EG
              3.23
+ BF
              106
+ BR
              0.55
              9000
+ IKF
+ NF
              1
+ NE
              2
              0.9
+ RB
+ IRB
             0.002
+ RBM
              0.09
+ RE
              0.01
+ RC
              0.013
              2.3989E-9
+ CJC
+ VJC
              2.8346223
              0.4846
+ MJC
+ CJE
              6.026E-09
+ VJE
              3.17915435
+ MJE
              0.52951635
+ XTI
              -1.2
+ XTB
              7.00E-3
+ TRC1
+ VCEO
              600
+ ICRATING
             100
+ MFG
              GeneSiC_Semiconductor
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^{*} End of GA50JT06 SPICE Model