0.4-GHz TO 4-GHz QUADRATURE MODULATOR

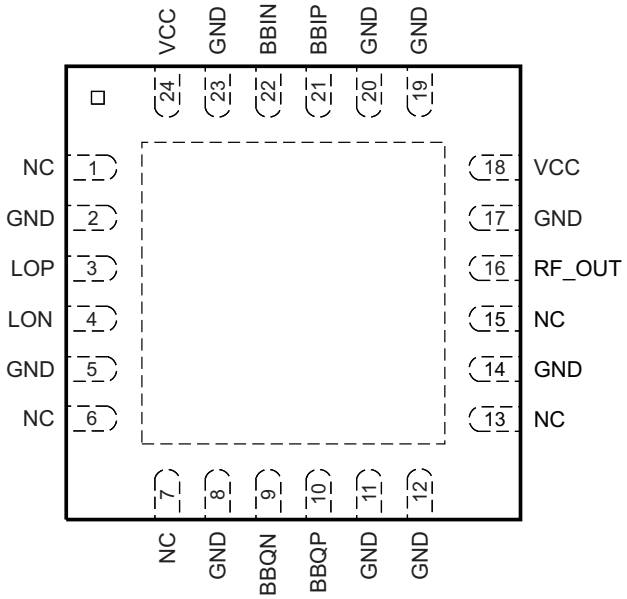
FEATURES

- **75-dBc Single-Carrier WCDMA ACPR at –11-dBm Channel Power**
- **Very Low Noise Floor: –163 dBm/Hz**
- **OPI3 of 23 dBm**
- **P1dB of 9 dBm**
- **Unadjusted Carrier Feedthrough of –40 dBm**
- **Unadjusted Side-Band Suppression of –40 dBc**
- **Single Supply: 4.5 V–5.5 V Operation**
- **Silicon Germanium Technology**
- **TRF370333 With 3.3-V CM at I, Q Baseband Inputs**
- **TRF370315 With 1.5-V CM at I, Q Baseband Inputs**

APPLICATIONS

- **Cellular Base Transceiver Station Transmit Channel**
- **CDMA: IS95, UMTS, CDMA2000, TD-SCDMA**
- **TDMA: GSM, IS-136, EDGE/UWC-136**
- **Wireless Local Loop**
- **Wireless MAN Wideband Transceivers**

RGE PACKAGE
(TOP VIEW)

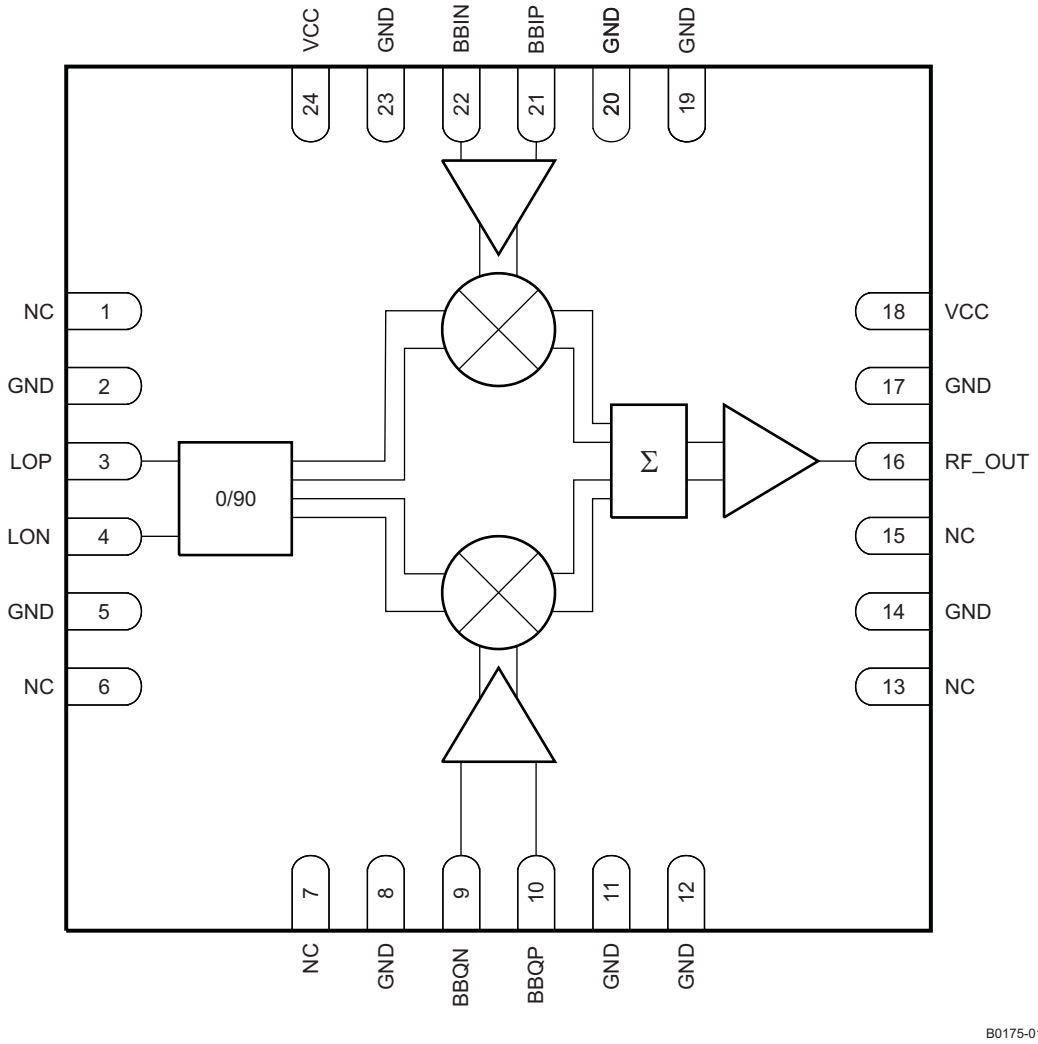


P0024-04

DESCRIPTION

The TRF3703 is a very low-noise direct quadrature modulator, capable of converting complex modulated signals from baseband or IF directly up to RF. The TRF3703 is ideal for high-performance direct RF modulation from 400 MHz up to 4 GHz. The modulator is implemented as a double-balanced mixer. The RF output block consists of a differential to single-ended converter and an RF amplifier capable of driving a single-ended 50-Ω load without any need of external components. The TRF3703 comes in two types, TRF370333 and TRF370315. The TRF370333 and TRF370315 devices have different common-mode voltage ratings at the I, Q baseband inputs. The TRF370333 requires a 3.3-V common-mode voltage, and the TRF370315 requires a 1.5-V common-mode voltage.

Functional Block Diagram



NOTE: NC = No connection

This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

DEVICE INFORMATION

TERMINAL FUNCTIONS

TERMINAL		I/O	DESCRIPTION
NAME	NO.		
BBIN	22	I	In-phase input
BBIP	21	I	In-phase input
BBQN	9	I	In-quadrature input
BBQP	10	I	In-quadrature input
GND	2, 5, 8,11, 12, 14, 17, 19, 20, 23	–	Ground
LON	4	I	Local oscillator input
LOP	3	I	Local oscillator input
NC	1, 6, 7, 13, 15	–	No connect
RF_OUT	16	O	RF output
VCC	18, 24	–	Power supply

ABSOLUTE MAXIMUM RATINGS(1)

over operating free-air temperature range (unless otherwise noted)

		VALUE(2)	UNIT
	Supply voltage range	–0.3 V to 6	V
	Digital I/O voltage range	–0.3 V to V_I + 0.3	V
T_J	Operating virtual junction temperature range	–40 to 150	°C
T_A	Operating ambient temperature range	–40 to 85	°C
T_{stg}	Storage temperature range	–65 to 150	°C

(1) Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
(2) All voltage values are with respect to network ground terminal.

RECOMMENDED OPERATING CONDITIONS

over operating free-air temperature range (unless otherwise noted)

		MIN	NOM	MAX	UNIT
V_{CC}	Power-supply voltage	4.5	5	5.5	V

THERMAL CHARACTERISTICS

PARAMETER		TEST CONDITIONS	VALUE	UNIT
$R_{\theta JA}$	Thermal resistance, junction-to-ambient	High-K board, still air	64.33	°C/W
$R_{\theta JC}$	Thermal resistance, junction-to-case		49.3	°C/W

ELECTRICAL CHARACTERISTICS

over operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
DC Parameters						
I_{CC}	Total supply current (1.5 V CM)	T_A = 25°C		195	205	mA
	Total supply current (3.3 V CM)	T_A = 25°C		210	235	
LO Input (50-Ω, Single-Ended)						
f_{LO}	LO frequency range		0.4		4	GHz
	LO input power		–5	0	12	dBm
	LO port return loss			15		dB
Baseband Inputs						
V_{CM}	I and Q input dc common voltage	TRF370333		3.3		V
		TRF370315		1.5		
BW	1-dB input frequency bandwidth		350			MHz
$Z_{I(single\ ended)}$	Input impedance, resistance	TRF370333		10		kΩ
	Input impedance, parallel capacitance			3		pF
	Input impedance, resistance	TRF370315		5		kΩ
	Input impedance, parallel capacitance			3		pF

ELECTRICAL CHARACTERISTICS

over recommended operating conditions, power supply = 5 V, T_A = 25°C, f_{LO} = **400 MHz** at 0 dBm (unless otherwise noted)

RF Output Parameters						
PARAMETER		**TEST CONDITIONS**	**MIN**	**TYP**	**MAX**	**UNIT**
G	Voltage gain	Output rms voltage over input I (or Q) rms voltage		–2.3		dB
P1dB	Output compression point			9.4		dBm
IP3	Output IP3		20	23		dBm
IP2	Output IP2	Measured at $f_{LO} + 2 \times f_{BB}$		62		dBm
	Carrier feedthrough	Unadjusted		–37		dBm
	Sideband suppression	Unadjusted		–39		dBc

ELECTRICAL CHARACTERISTICS

over recommended operating conditions, power supply = 5 V, T_A = 25°C, f_{LO} = **900 MHz** at 0 dBm (unless otherwise noted)

RF Output Parameters						
PARAMETER		**TEST CONDITIONS**	**MIN**	**TYP**	**MAX**	**UNIT**
G	Voltage gain	Output rms voltage over input I (or Q) rms voltage		–4.1		dB
P1dB	Output compression point			9		dBm
IP3	Output IP3		20	23		dBm
IP2	Output IP2	Measured at $f_{LO} + 2 \times f_{BB}$		63		dBm
	Carrier feedthrough	Unadjusted		–37		dBm
	Sideband suppression	Unadjusted		–42		dBc
	Output return loss			9		dB
	Output noise floor	DC only to BB inputs, 13 MHz offset from f_{LO}		–160.4		dBm/Hz
		1.8-MHz offset from f_{LO}; 1 CW tone; P_{out} = 0 dBm		–156.6		
		6-MHz offset from f_{LO}; 1 CW tone; P_{out} = 0 dBm		–158.5		
EVM	Error vector magnitude (rms)	1 EDGE signal, P_{out} = –5 dBm		0.59%		
		1 EDGE signal, P_{out} = 0 dBm		0.63%		
		1 EDGE signal, P_{out} = 0 dBm, 2nd harmonic of LO = –15 dBm, 3rd harmonic of LO = –33 dBm(1)		1%		

(1) The second- and third-harmonic tests were made independently at each frequency.

ELECTRICAL CHARACTERISTICS

over recommended operating conditions, power supply = 5 V, T_A = 25°C, f_{LO} = **1800 MHz** at 0 dBm (unless otherwise noted)

RF Output Parameters						
PARAMETER		**TEST CONDITIONS**	**MIN**	**TYP**	**MAX**	**UNIT**
G	Voltage gain	Output rms voltage over input I (or Q) rms voltage		–4.4		dB
P1dB	Output compression point			9.5		dBm
IP3	Output IP3		20	23		dBm
IP2	Output IP2	Measured at $f_{LO} + 2 \times f_{BB}$		55		dBm
	Carrier feedthrough	Unadjusted		–40		dBm
	Sideband suppression	Unadjusted		–47		dBc
	Output return loss			8		dB
	Output noise floor	DC only to BB inputs, 13 MHz offset from f_{LO}		–162.6		dBm/Hz
		1.8-MHz offset from f_{LO}; 1 CW tone; P_{out} = 0 dBm		–160		
		6-MHz offset from f_{LO}; 1 CW tone; P_{out} = 0 dBm		–159.4		
EVM	Error vector magnitude (rms)	1 EDGE signal, P_{out} = –5 dBm		0.66%		
		1 EDGE signal, P_{out} = 0 dBm		0.74%		
		1 EDGE signal, P_{out} = 0 dBm, 2nd harmonic of LO = –15.5 dBm, 3rd harmonic of LO = –30 dBm(1)		1%		

(1) The second- and third-harmonic tests were made independently at each frequency.

ELECTRICAL CHARACTERISTICS

over recommended operating conditions, power supply = 5 V, T_A = 25°C, f_{LO} = **2140 MHz** at 0 dBm (unless otherwise noted)

RF Output Parameters						
PARAMETER		**TEST CONDITIONS**	**MIN**	**TYP**	**MAX**	**UNIT**
G	Voltage gain	Output rms voltage over input I (or Q) rms voltage		–4.5		dB
P1dB	Output compression point			9.5		dBm
IP3	Output IP3		18	21		dBm
IP2	Output IP2	Measured at $f_{LO} + 2 \times f_{BB}$		58		dBm
	Carrier feedthrough	Unadjusted		–40		dBm
	Sideband suppression	Unadjusted		–47		dBc
	Output return loss			8.5		dB
	Output noise floor	20-MHz offset from f_{LO}; dc only to BB inputs		–163		dBm/Hz
		20-MHz offset from f_{LO}; 1 WCDMA signal; P_{in} = –20.5 dBVrms (I and Q input)		–162		
ACPR	Adjacent-channel power ratio	1 WCDMA signal; P_{out} = –13 dBm		–75.8		dBc
		1 WCDMA signal; P_{out} = –9 dBm		–72		
		4 WCDMA signals; P_{out} = –23 dBm per carrier		–68		
	Alternate-channel power ratio	1 WCDMA signal; P_{out} = –13 dBm		–79		dBc
		1 WCDMA signal; P_{out} = –9 dBm		–80.5		
		4 WCDMA signals; P_{out} = –23 dBm per carrier		–69		

ELECTRICAL CHARACTERISTICS

over recommended operating conditions, power supply = 5 V, T_A = 25°C, f_{LO} = **2500 MHz** at 0 dBm (unless otherwise noted)

RF Output Parameters						
PARAMETER		**TEST CONDITIONS**	**MIN**	**TYP**	**MAX**	**UNIT**
G	Voltage gain	Output rms voltage over input I (or Q) rms voltage		–4.4		dB
P1dB	Output compression point			9.5		dBm
IP3	Output IP3		18	21		dBm
IP2	Output IP2	Measured at $f_{LO} + 2 \times f_{BB}$		63		dBm
	Carrier feedthrough	Unadjusted		–38		dBm
	Sideband suppression	Unadjusted		–47		dBc

ELECTRICAL CHARACTERISTICS

over recommended operating conditions, power supply = 5 V, T_A = 25°C, f_{LO} = **3600 MHz** at 0 dBm (unless otherwise noted)

RF Output Parameters						
PARAMETER		**TEST CONDITIONS**	**MIN**	**TYP**	**MAX**	**UNIT**
G	Voltage gain	Output rms voltage over input I (or Q) rms voltage		–3.5		dB
P1dB	Output compression point			9.5		dBm
IP3	Output IP3		20	23		dBm
IP2	Output IP2	Measured at $f_{LO} + 2 \times f_{BB}$		63		dBm
	Carrier feedthrough	Unadjusted		–41		dBm
	Sideband suppression	Unadjusted		–45		dBc

ELECTRICAL CHARACTERISTICS

over recommended operating conditions, power supply = 5 V, T_A = 25°C, f_{LO} = **4000 MHz** at 0 dBm (unless otherwise noted)

RF Output Parameters						
PARAMETER		**TEST CONDITIONS**	**MIN**	**TYP**	**MAX**	**UNIT**
G	Voltage gain	Output rms voltage over input I (or Q) rms voltage		–4.5		dB
P1dB	Output compression point			9		dBm
IP3	Output IP3		19	22		dBm
IP2	Output IP2	Measured at $f_{LO} + 2 \times f_{BB}$		50		dBm
	Carrier feedthrough	Unadjusted		–37		dBm
	Sideband suppression	Unadjusted		–40		dBc

TYPICAL CHARACTERISTICS

P_{OUT} vs BASEBAND VOLTAGE

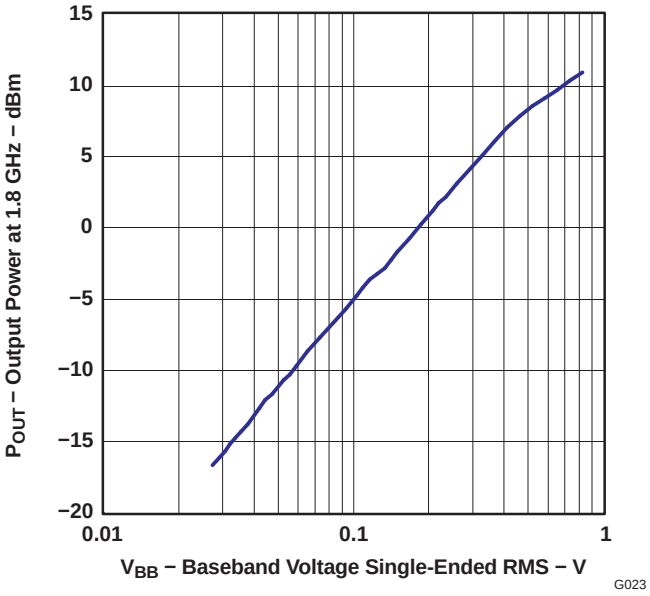


Figure 1.

P_{OUT} vs FREQUENCY AND TEMPERATURE

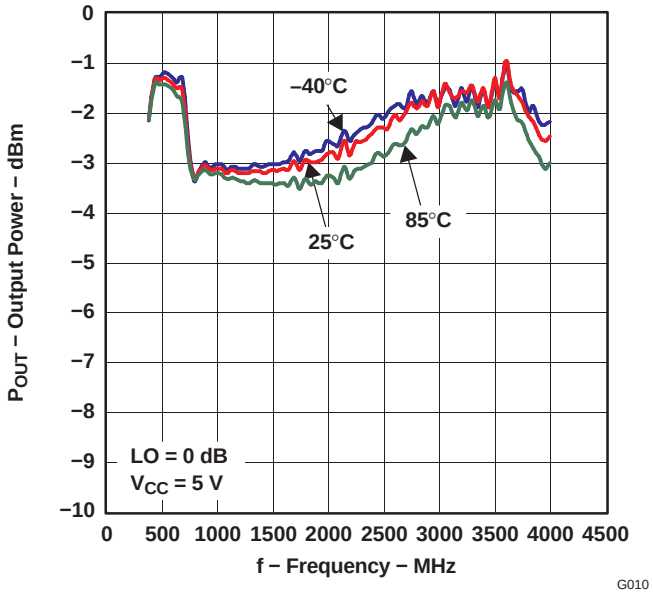


Figure 2.

P_{OUT} vs FREQUENCY AND SUPPLY VOLTAGE

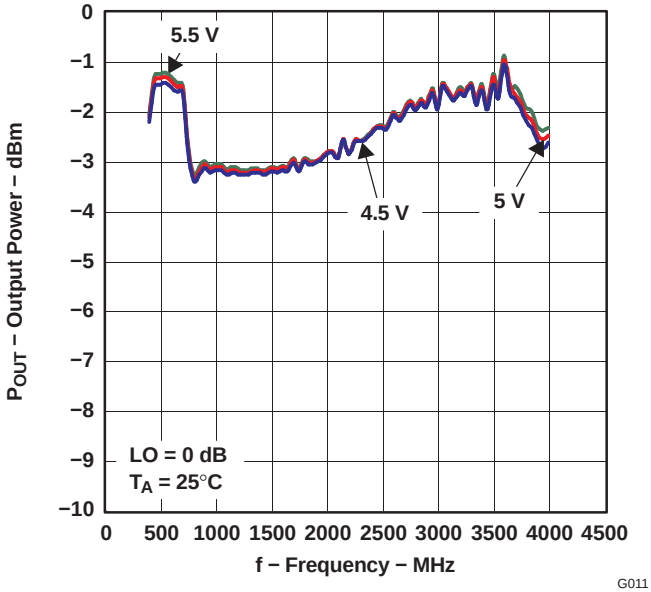


Figure 3.

P_{OUT} vs FREQUENCY AND LO POWER

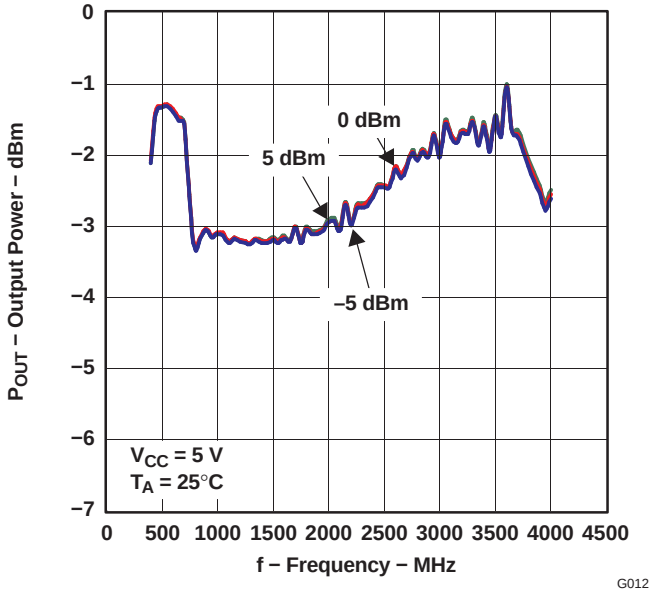


Figure 4.

TYPICAL CHARACTERISTICS (continued)

P1dB
vs
FREQUENCY AND TEMPERATURE

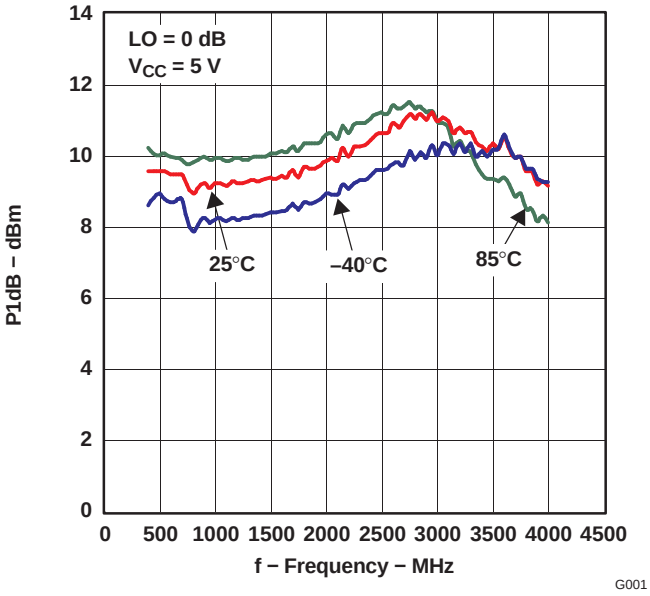


Figure 5.

P1dB
vs
FREQUENCY AND SUPPLY VOLTAGE

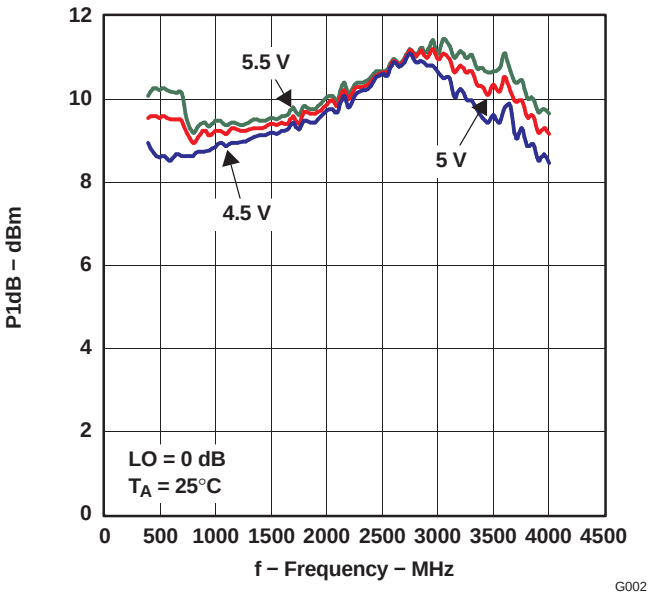


Figure 6.

P1dB
vs
FREQUENCY AND LO POWER

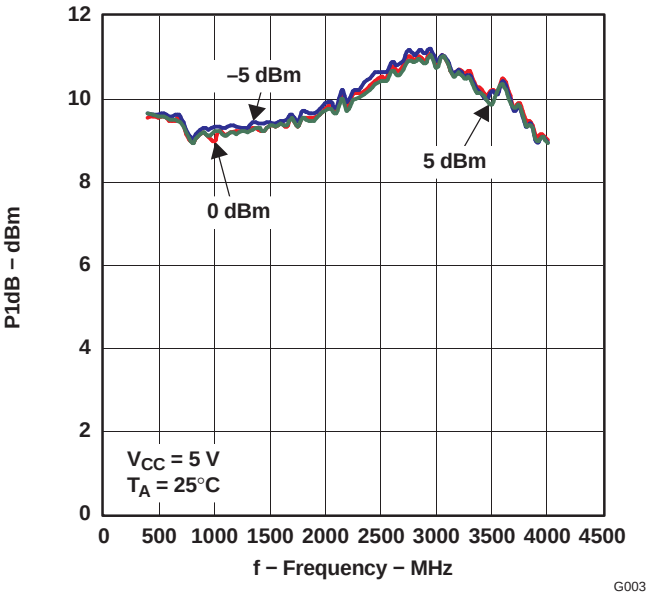


Figure 7.

OIP3
vs
FREQUENCY AND TEMPERATURE

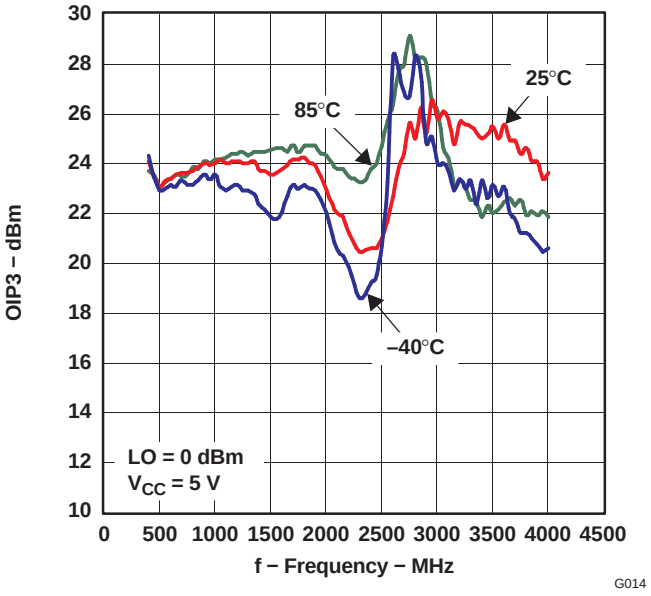


Figure 8.

TYPICAL CHARACTERISTICS (continued)

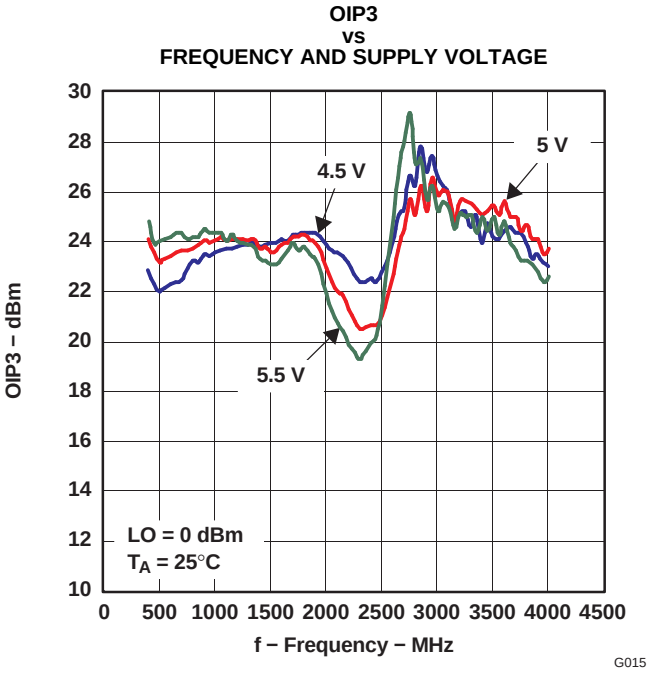


Figure 9.

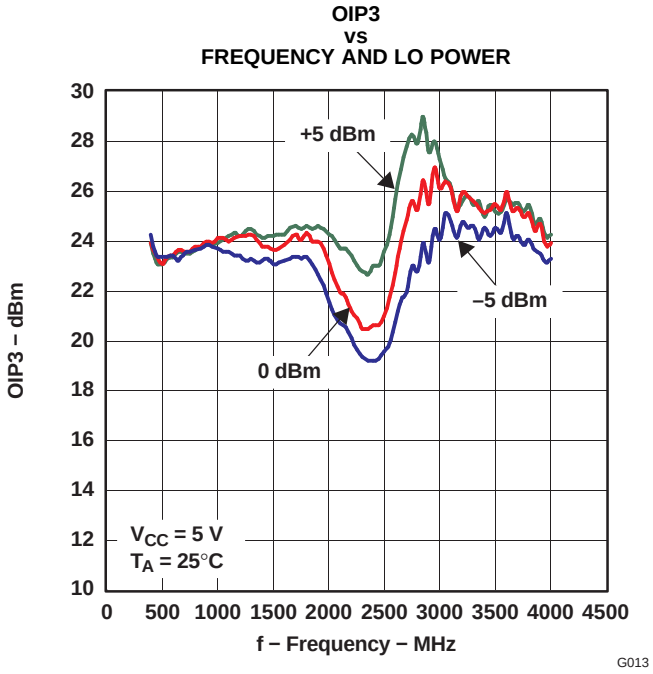


Figure 10.

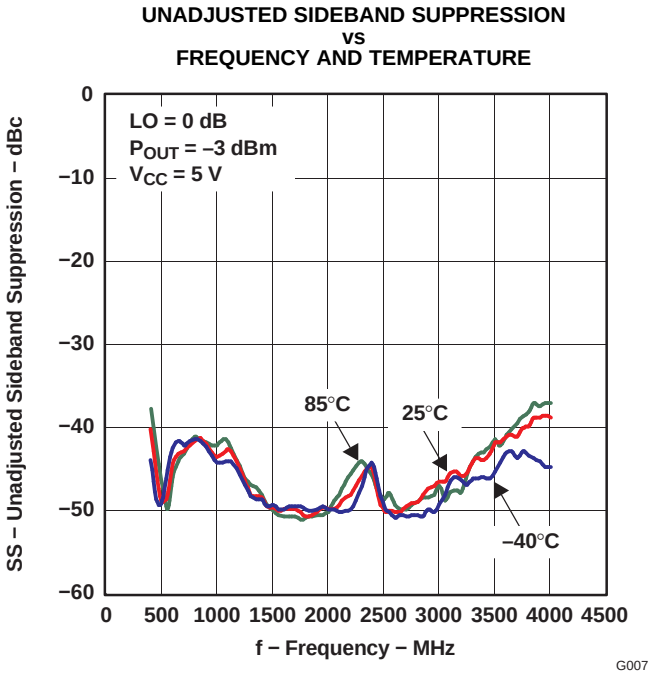


Figure 11.

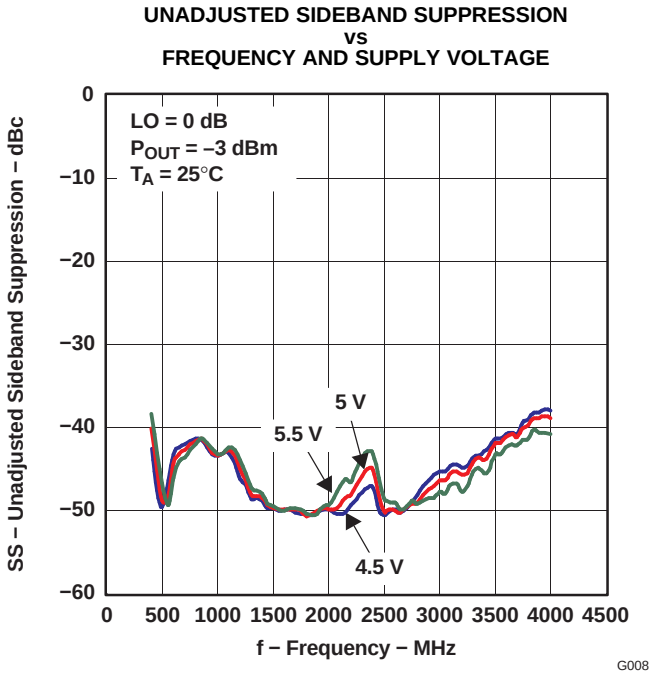


Figure 12.

TYPICAL CHARACTERISTICS (continued)

UNADJUSTED SIDEBAND SUPPRESSION
vs
FREQUENCY AND LO POWER

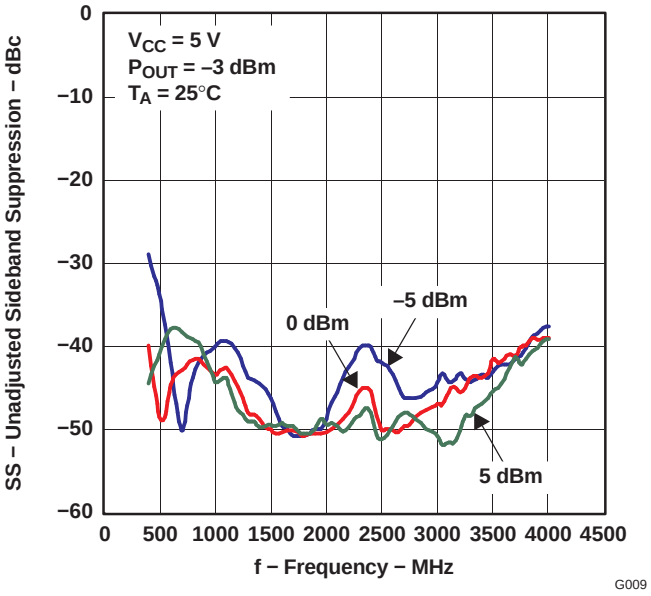


Figure 13.

ADJUSTED SIDEBAND SUPPRESSION
vs
FREQUENCY AND TEMPERATURE

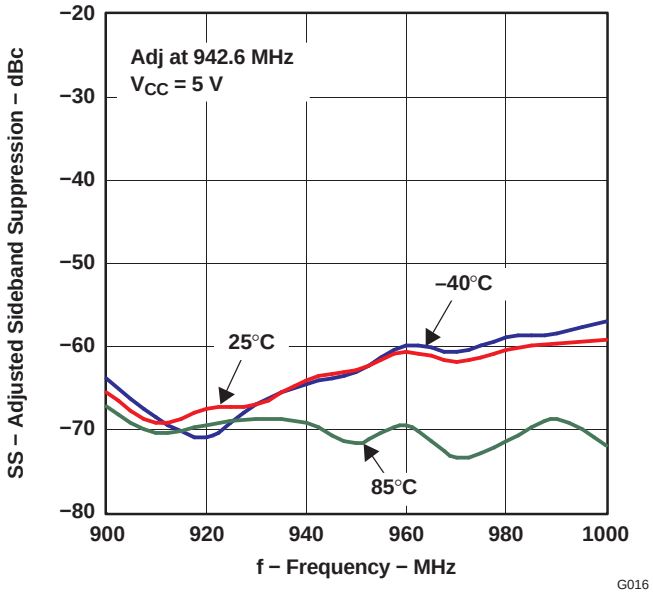


Figure 14.

ADJUSTED SIDEBAND SUPPRESSION
vs
FREQUENCY AND TEMPERATURE

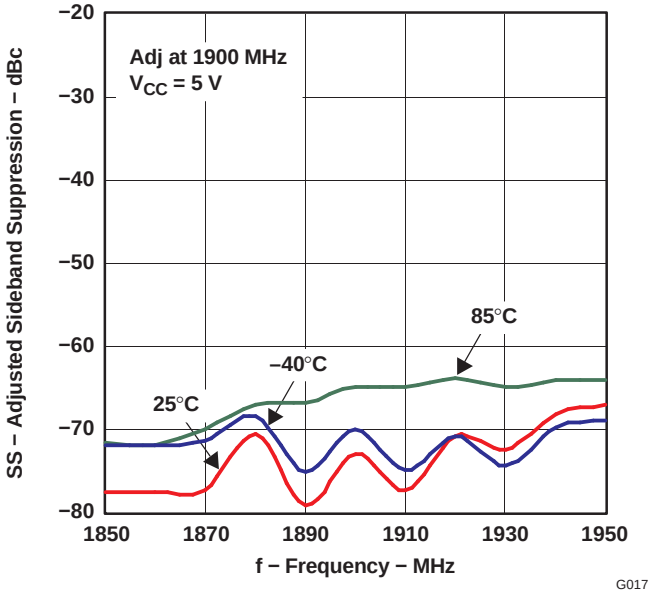


Figure 15.

ADJUSTED SIDEBAND SUPPRESSION
vs
FREQUENCY AND TEMPERATURE

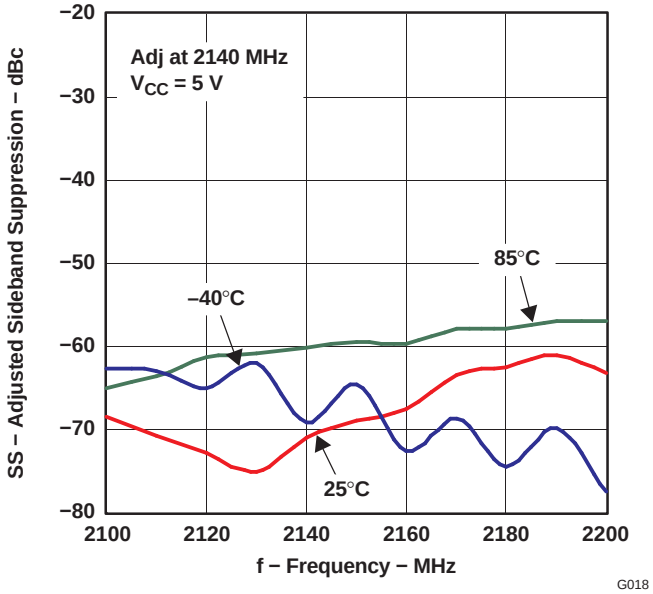


Figure 16.

TYPICAL CHARACTERISTICS (continued)

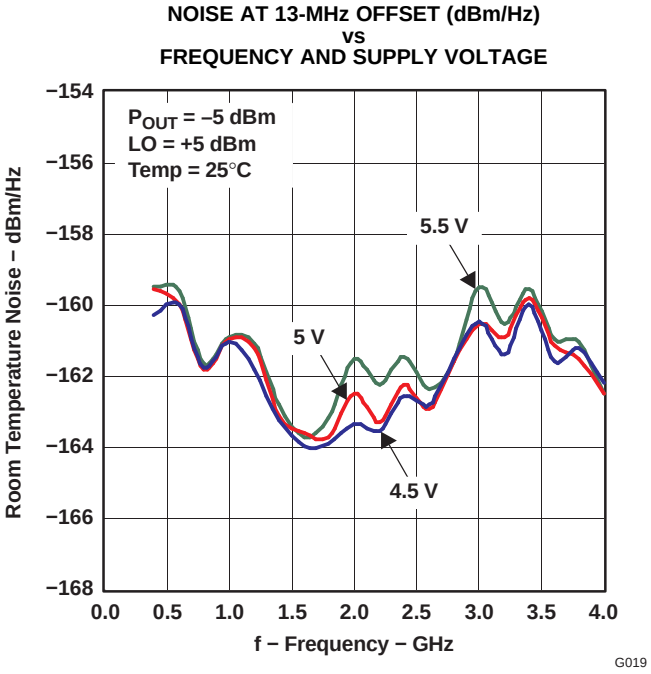


Figure 17.

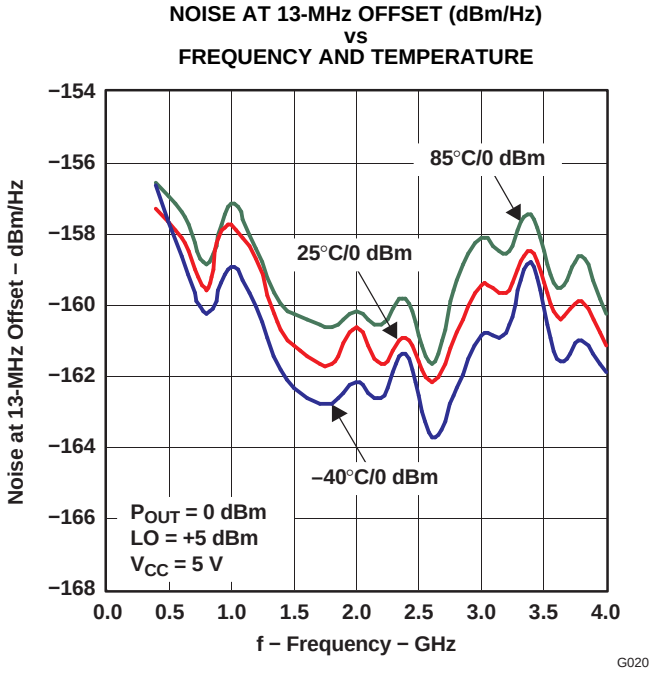


Figure 18.

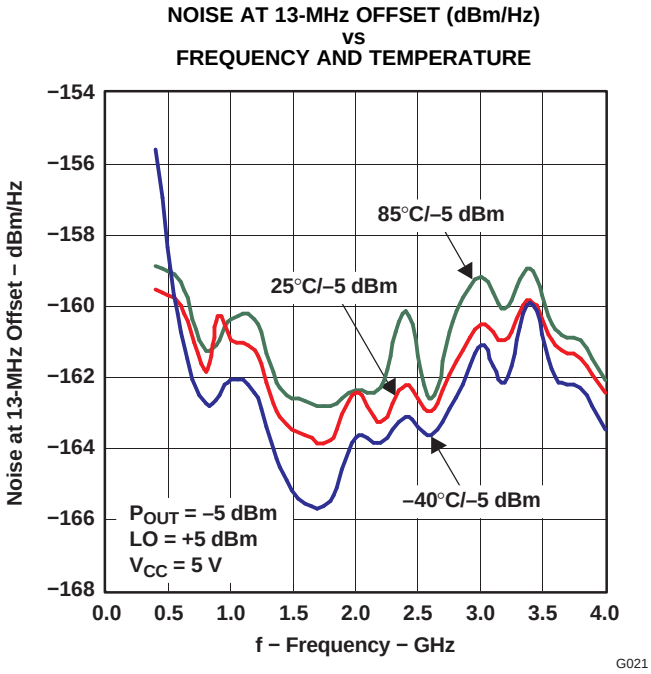


Figure 19.

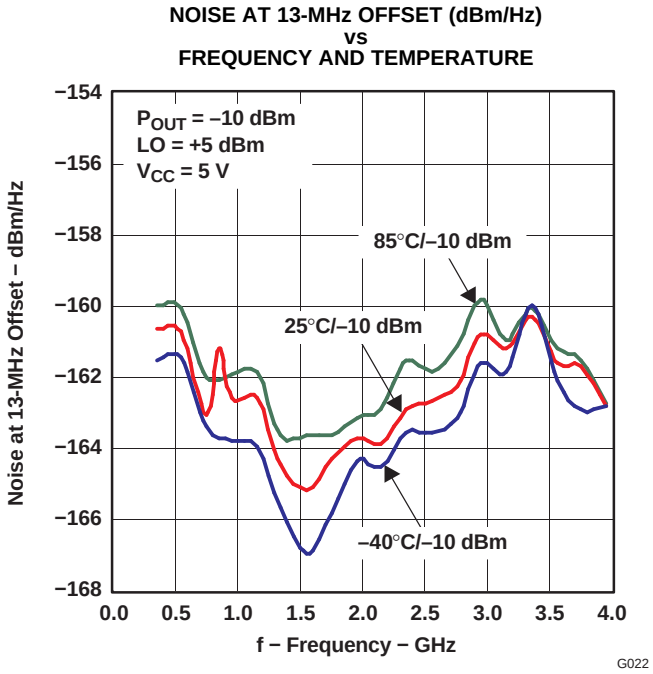


Figure 20.

TYPICAL CHARACTERISTICS (continued)

UNADJUSTED CARRIER FEEDTHROUGH
vs
SUPPLY VOLTAGE

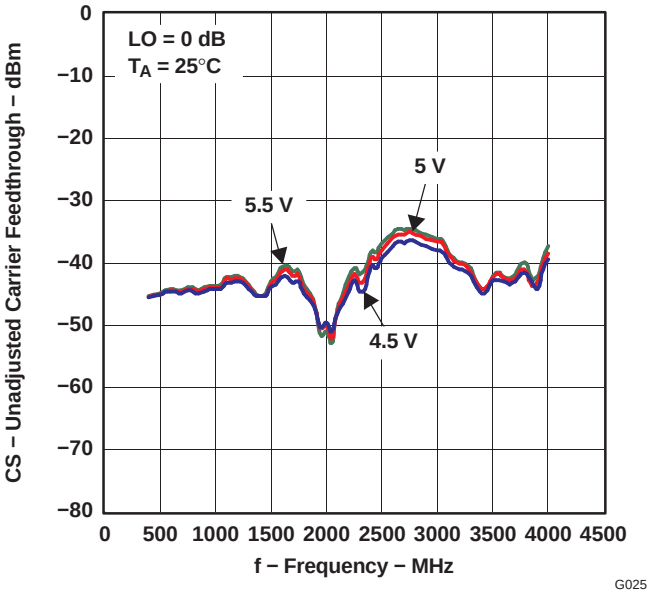


Figure 21.

UNADJUSTED CARRIER FEEDTHROUGH
vs
FREQUENCY AND TEMPERATURE

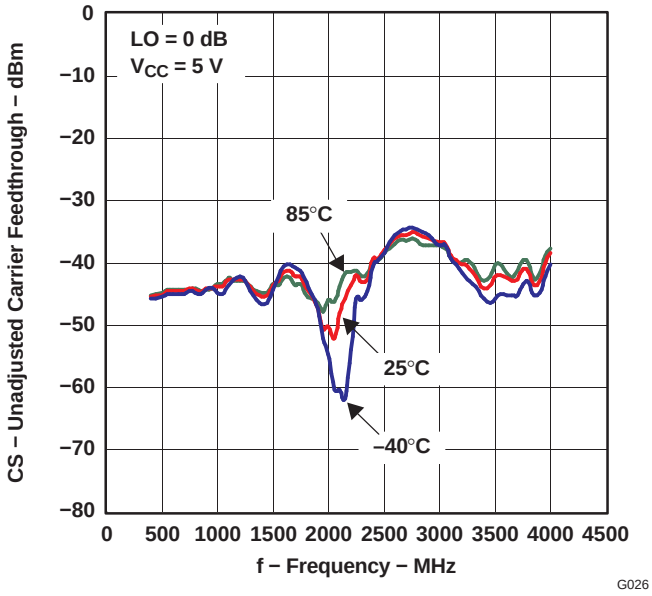


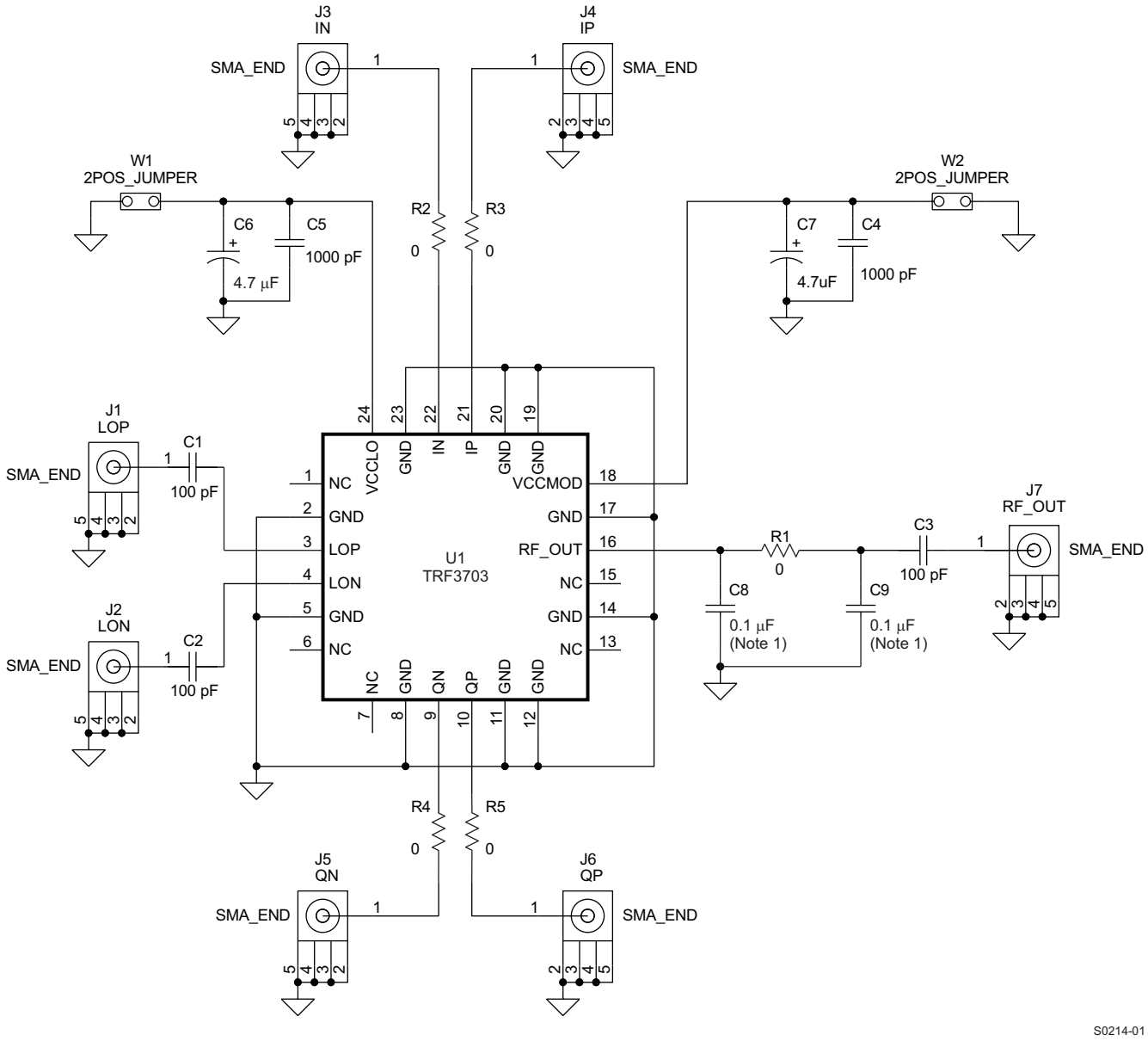
Figure 22.

APPLICATION INFORMATION AND EVALUATION BOARD

Basic Connections

- See Figure 23 for proper connection of the TRF3703 modulator.
- Connect a single power supply (4.5 V–5.5 V) to pins 18 and 24. These pins should be decoupled as shown on pins 4, 5, 6, and 7.
- Connect pins 2, 5, 8, 11, 12, 14, 17, 19, 20, and 23 to GND.
- Connect a single-ended LO source of desired frequency to LOP (amplitude between –5 dBm and 12 dBm). This should be ac-coupled through a 100-pF capacitor.
- Terminate the ac-coupled LON with 50 Ω to GND.
- Connect a baseband signal to pins 21 = I, 22 = $\overline{I}$, 10 = Q, and 9 = $\overline{Q}$.
- The differential baseband inputs should be set to the proper level, 3.3 V for the TRF370333 or 1.5 V for the TRF370315.
- RF_OUT, pin 16, can be fed to a spectrum analyzer set to the desired frequency, LO ± baseband signal. This pin should also be ac-coupled through a 100-pF capacitor.
- All NC pins can be left floating.

APPLICATION INFORMATION AND EVALUATION BOARD (continued)



(1) Do not install.

Figure 23. TRF3703 EVM Schematic

APPLICATION INFORMATION AND EVALUATION BOARD (continued)

Figure 24 shows the top view of the TRF3703 EVM board.

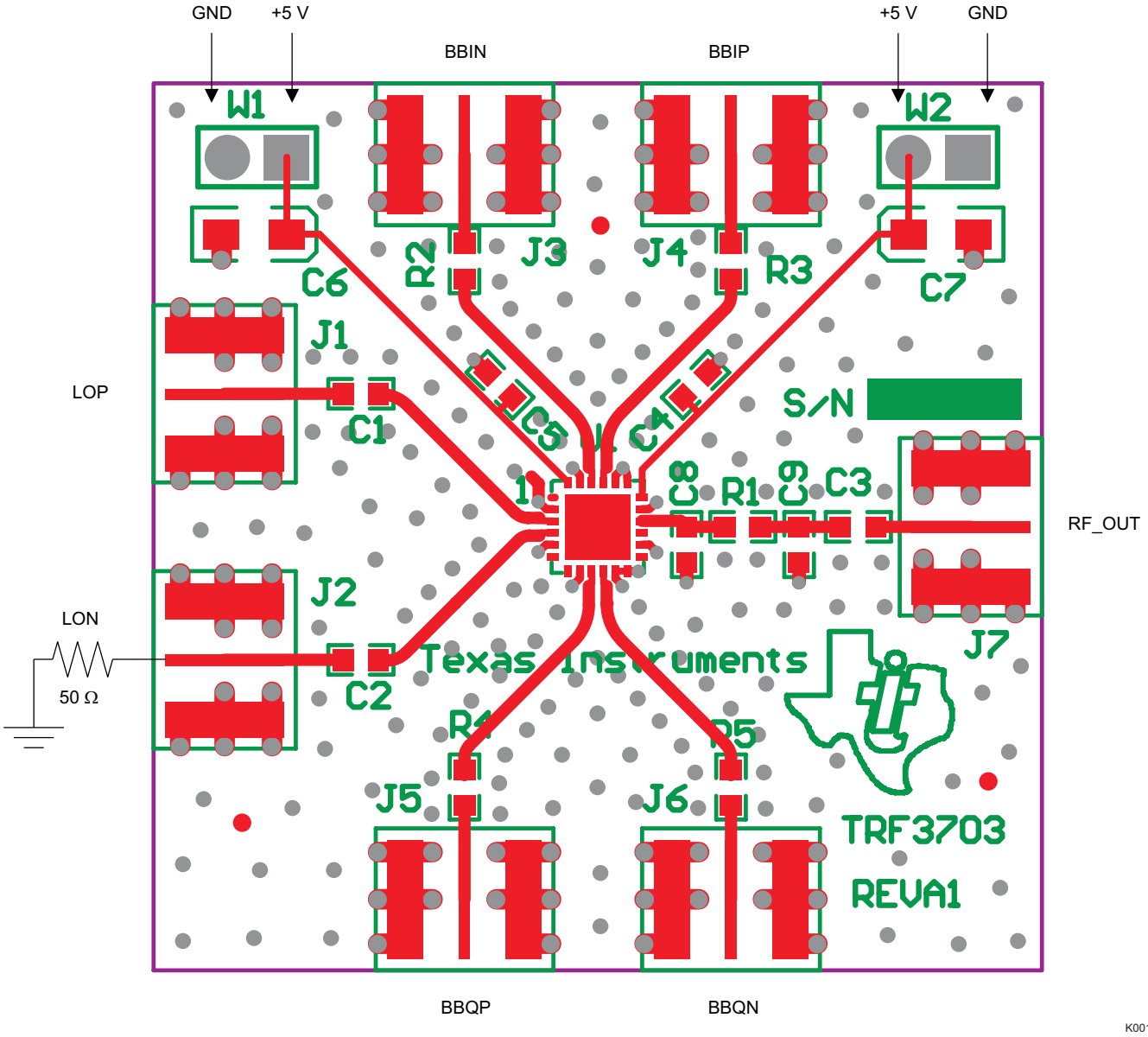


Figure 24. TRF3703 EVM Board Layout

Table 1. Bill of Materials for TRF3703 EVM

Value	Footprint	QTY	Part Number	Vendor	Digi-Key Number	REF DES	Not Installed
Tantalum 4.7-µF, 10-V, 10% capacitor	3216	2	T491A475K010AS	KEMET	399-1561-1-ND	C6, C7	
1000-pF, 50-V, 5% capacitor	603	2	ECJ-1VC1H102J	Panasonic	PCC2151CT-ND	C4, C5	
100-pF, 50-V, 5% capacitor	603	3	ECJ-1VC1H101J	Panasonic	PCC101ACVCT-ND	C1, C2, C3	
Capacitor	603	0					C8, C9

APPLICATION INFORMATION AND EVALUATION BOARD (continued)

Table 1. Bill of Materials for TRF3703 EVM (continued)

Value	Footprint	QTY	Part Number	Vendor	Digi-Key Number	REF DES	Not Installed
0-Ω resistor, 1/10-W, 5%	603	5	ERJ-3GEY0R00V	Panasonic	P0.0GCT-ND	R1, R2, R3, R4, R5	
TRF3703	24-QFN-PP-4X4MM	1		TI		U1	
SMA connectors	SMA_END_ SMALL	6	16F3627	Newark	142-0711-821	J1, J2, J3, J4, J5, J6, J7	
2POS_HEADER	2POS_JUMP	2	HTSW-150-07-L-S	SAMTEC	N/A	W1, W2	

GSM Applications

The TRF3703 is suited for GSM applications because of its high linearity and low noise level over the entire recommended operating range. It also has excellent EVM performance, which makes it ideal for the stringent GSM/EDGE applications.

WCDMA Applications

The TRF3703 is also optimized for WCDMA applications where both adjacent-channel power ratio (ACPR) and noise density are critically important. Using Texas instruments' DAC568X series of high-performance digital-to-analog converters as depicted in Figure 25, excellent ACPR levels were measured with one-, two-, and four-WCDMA carriers. See *Electrical Characteristics*, f_{LO} = 2140 MHz for exact ACPR values.

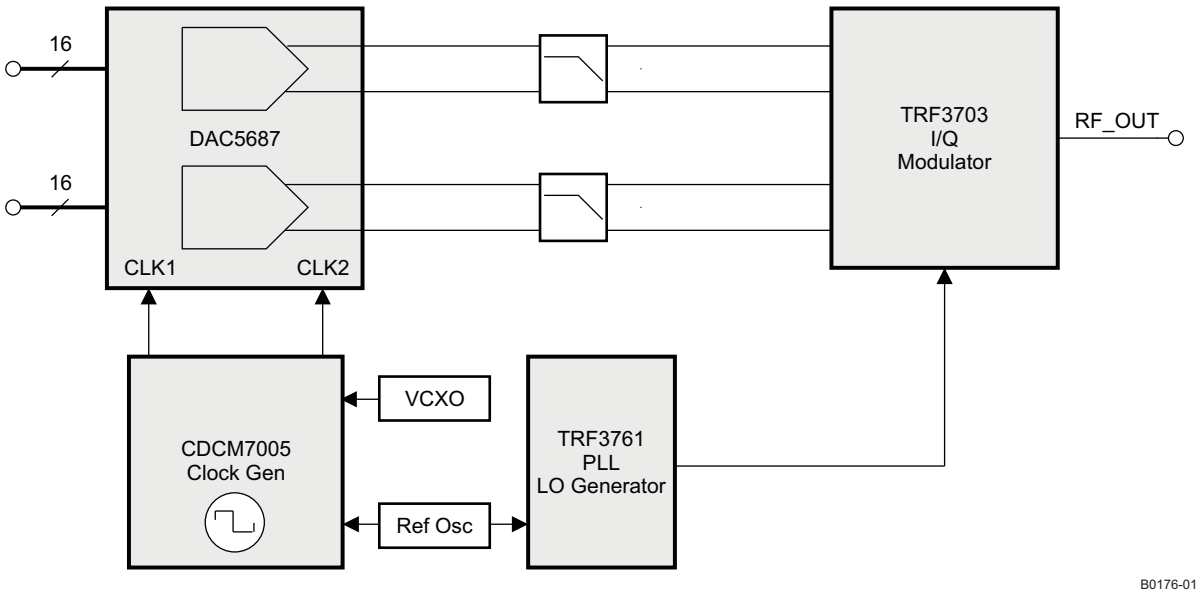


Figure 25. Typical Transmit Setup Block Diagram

DEFINITION OF SPECIFICATIONS

Unadjusted Carrier Feedthrough

This specification measures the amount by which the local oscillator component is attenuated in the output spectrum of the modulator relative to the carrier. This further assumes that the baseband inputs delivered to the pins of the TRF3703 are perfectly matched to have the same dc offset (VCM). This includes all four baseband inputs: I, $\overline{I}$, Q, and $\overline{Q}$. This is measured in dBm.

Adjusted (Optimized) Carrier Feedthrough

This differs from the unadjusted suppression number in that the baseband input dc offsets are iteratively adjusted around their theoretical value of VCM to yield the maximum suppression of the LO component in the output spectrum. This is measured in dBm.

Unadjusted Sideband Suppression

This specification measures the amount by which the unwanted sideband of the input signal is attenuated in the output of the modulator, relative to the wanted sideband. This further assumes that the baseband inputs delivered to the modulator input pins are perfectly matched in amplitude and are exactly 90° out of phase. This is measured in dBc.

Adjusted (Optimized) Sideband Suppression

This differs from the unadjusted sideband suppression in that the baseband inputs are iteratively adjusted around their theoretical values to maximize the amount of sideband suppression. This is measured in dBc.

Suppressions Over Temperature

This specification assumes that the user has gone though the optimization process for the suppression in question, and set the optimal settings for the I, Q inputs. This specification then measures the suppression when temperature conditions change after the initial calibration is done.

Figure 26 shows a simulated output and illustrates the respective definitions of various terms used in this data sheet. The graph assumes a baseband input of 50 kHz.

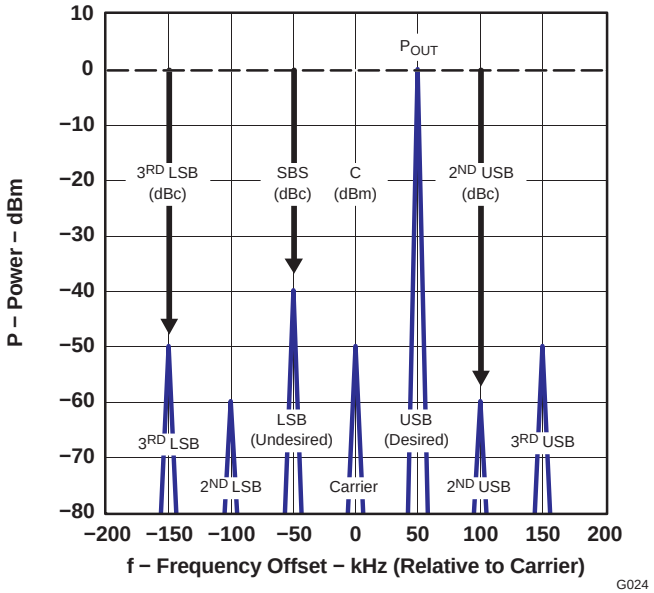


Figure 26. Graphical Illustration of Common Terms

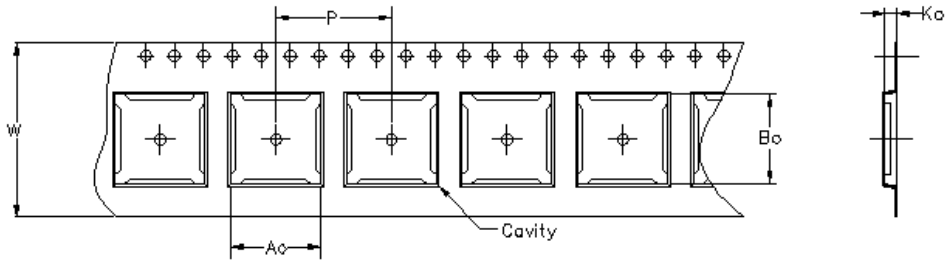
PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish	MSL Peak Temp (3)
TRF370315IRGER	ACTIVE	QFN	RGE	24	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR
TRF370315IRGET	ACTIVE	QFN	RGE	24	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR
TRF370333IRGER	ACTIVE	QFN	RGE	24	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR
TRF370333IRGERG4	ACTIVE	QFN	RGE	24	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR
TRF370333IRGET	ACTIVE	QFN	RGE	24	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR
TRF370333IRGETG4	ACTIVE	QFN	RGE	24	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR

(1) The marketing status values are defined as follows:
ACTIVE: Product device recommended for new designs.
LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.
NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.
PREVIEW: Device has been announced but is not in production. Samples may or may not be available.
OBSOLETE: TI has discontinued the production of the device.

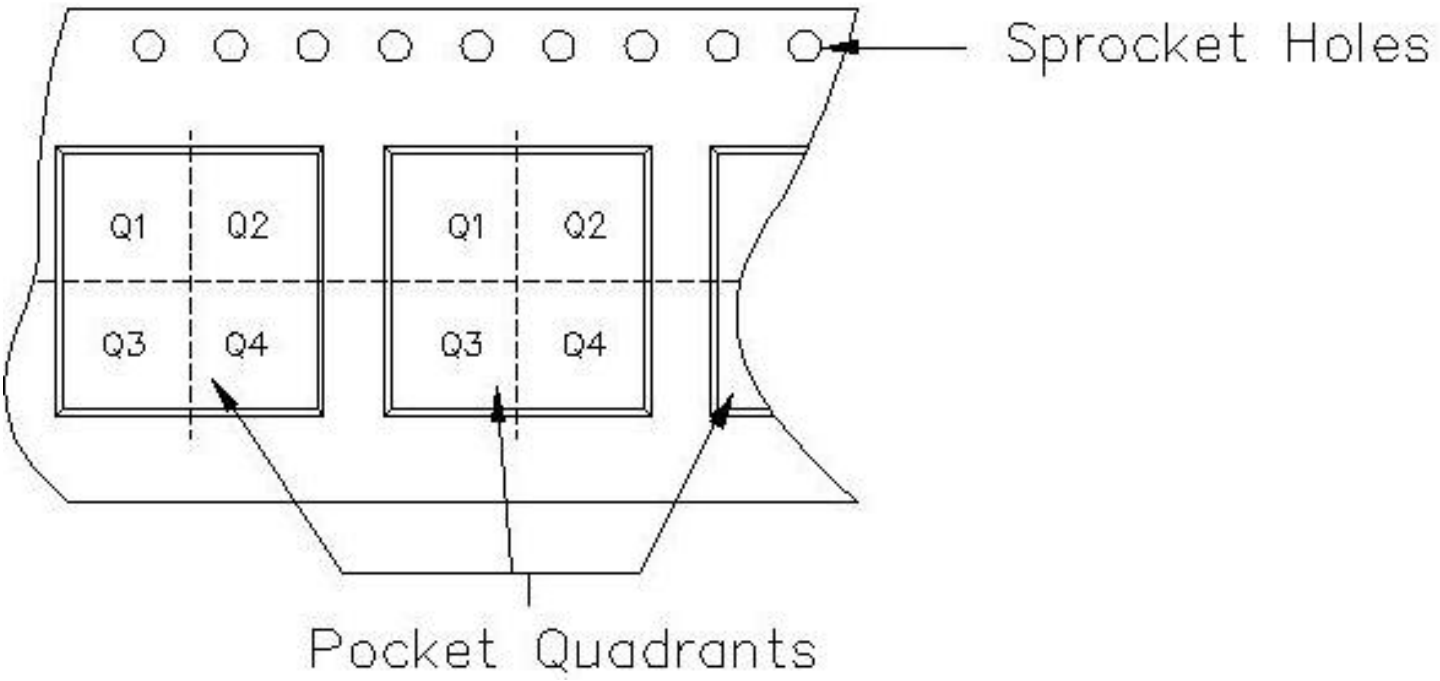
(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.
TBD: The Pb-Free/Green conversion plan has not been defined.
Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.
Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.
Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.



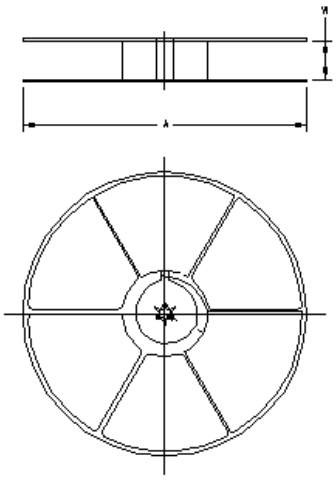
Carrier tape design is defined largely by the component lentgh, width, and thickness.

Ao = Dimension designed to accommodate the component width.
Bo = Dimension designed to accommodate the component length.
Ko = Dimension designed to accommodate the component thickness.
W = Overall width of the carrier tape.
P = Pitch between successive cavity centers.



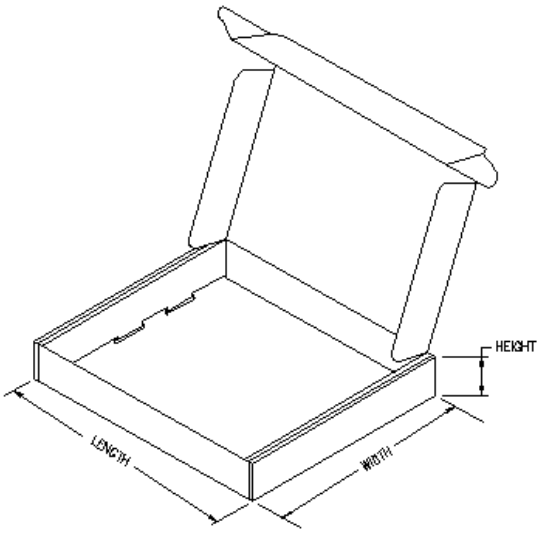
TAPE AND REEL INFORMATION

Device	Package	Pins	Site	Reel Diameter (mm)	Reel Width (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TRF370315IRGER	RGE	24	SITE 60	330	12	4.3	4.3	1.5	8	12	Q1
TRF370315IRGET	RGE	24	SITE 60	330	12	4.3	4.3	1.5	8	12	Q1
TRF370333IRGER	RGE	24	SITE 60	330	12	4.3	4.3	1.5	8	12	Q2
TRF370333IRGET	RGE	24	SITE 60	330	12	4.3	4.3	1.5	8	12	Q2

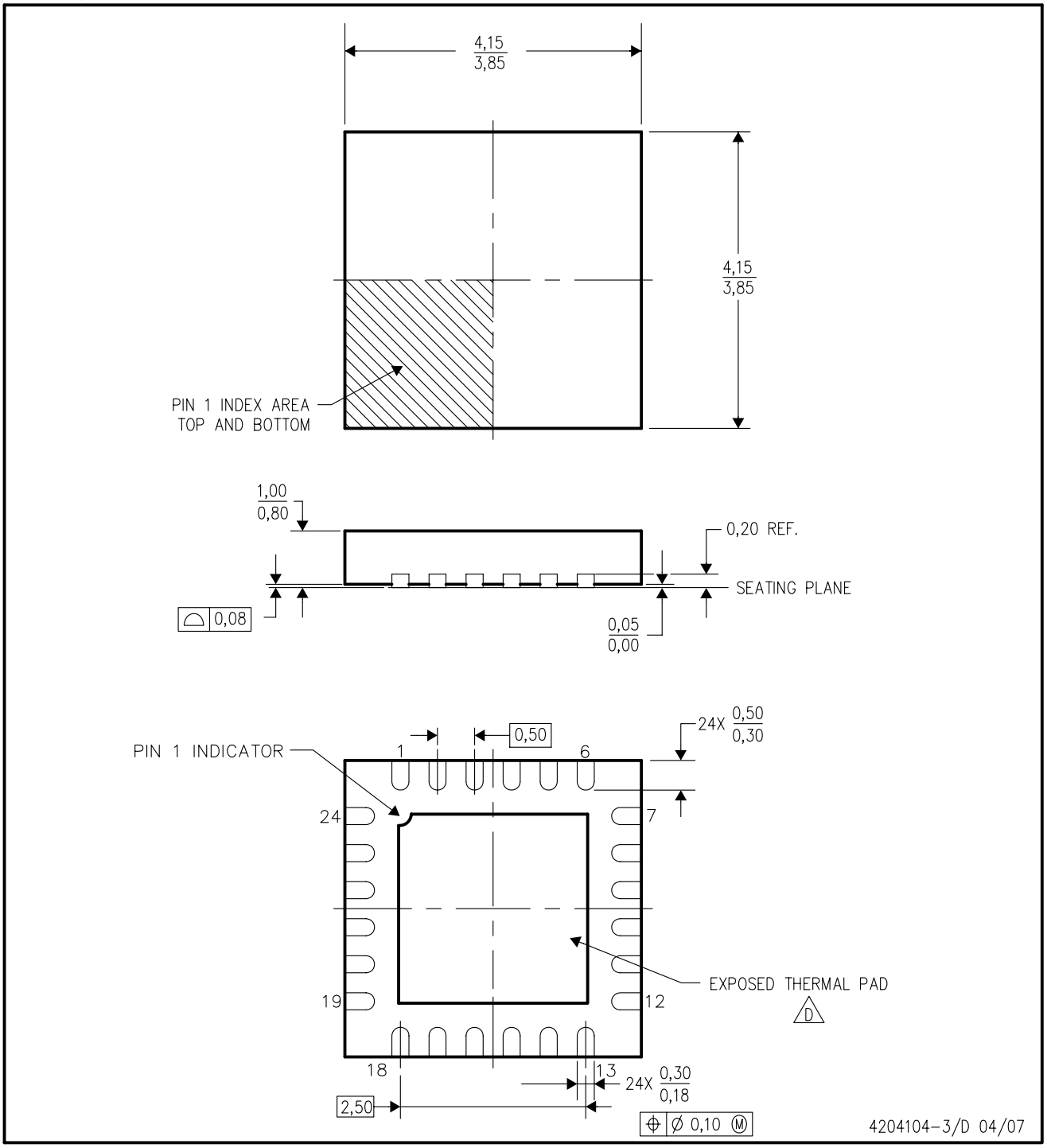


TAPE AND REEL BOX INFORMATION

Device	Package	Pins	Site	Length (mm)	Width (mm)	Height (mm)
TRF370315IRGER	RGE	24	SITE 60	342.9	336.6	20.64
TRF370315IRGET	RGE	24	SITE 60	407.0	336.6	97.0
TRF370333IRGER	RGE	24	SITE 60	342.9	336.6	20.64
TRF370333IRGET	RGE	24	SITE 60	342.9	336.6	20.64


HEIGHT
LENGTH
WIDTH

RGE (S–PQFP–N24) PIN 1 OPTION PLASTIC QUAD FLATPACK



NOTES:
A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M–1994.
B. This drawing is subject to change without notice.
C. Quad Flatpack, No–Leads (QFN) package configuration.
D. The package thermal pad must be soldered to the board for thermal and mechanical performance. See the Product Data Sheet for details regarding the exposed thermal pad dimensions.
E. Falls within JEDEC MO–220.

THERMAL PAD MECHANICAL DATA

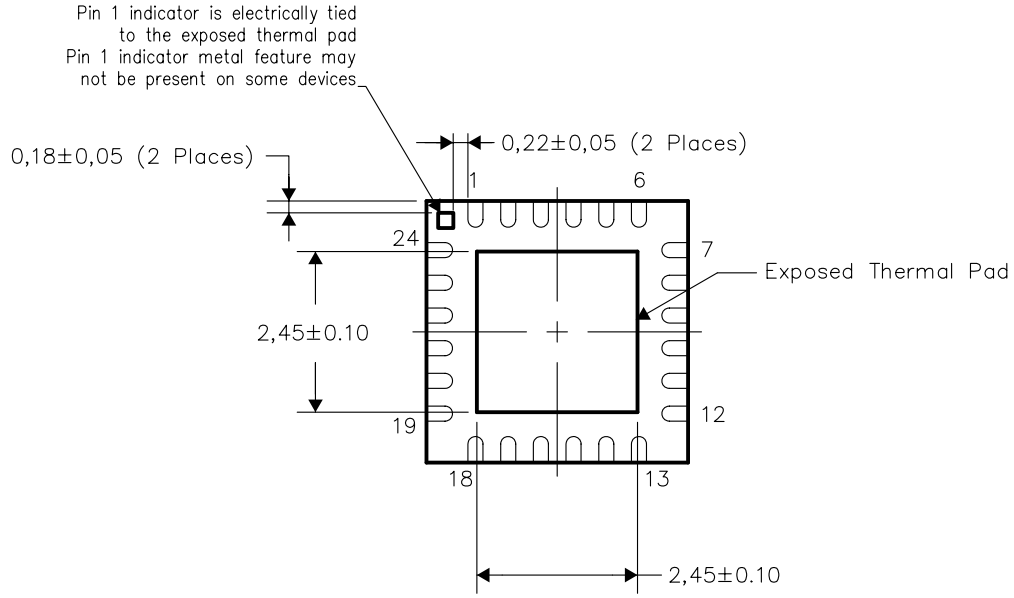
RGE (S–PQFP–N24)

THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to a ground or power plane (whichever is applicable), or alternatively, a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No–Lead (QFN) package and its advantages, refer to Application Report, Quad Flatpack No–Lead Logic Packages, Texas Instruments Literature No. SCBA017. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.

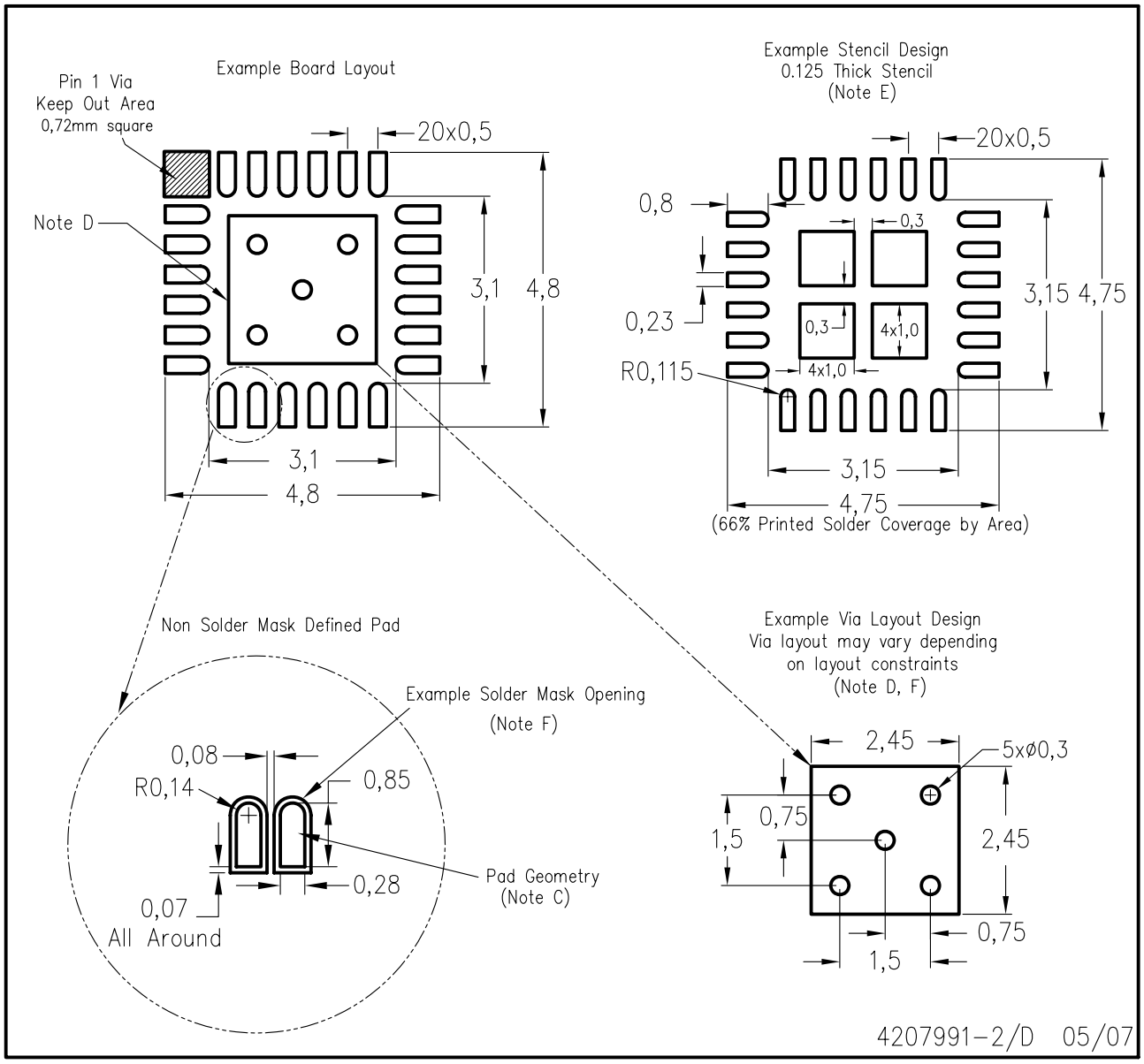


Bottom View

NOTE: All linear dimensions are in millimeters

Exposed Thermal Pad Dimensions

RGE (S–PQFP–N24)



NOTES:
A. All linear dimensions are in millimeters.
B. This drawing is subject to change without notice.
C. Publication IPC–7351 is recommended for alternate designs.
D. This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, Quad Flat–Pack Packages, Texas Instruments Literature No. SCBA017, SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com <http://www.ti.com>.
E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
F. Customers should contact their board fabrication site for recommended solder mask tolerances and via tenting recommendations for vias placed in the thermal pad.

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