# DS42MB200 Dual 4.25 Gbps 2:1/1:2 CML Mux/Buffer with Transmit Pre-Emphasis and Receive Equalization

**Check for Samples: DS42MB200**

## FEATURES

- **1– 4.25 Gbps Fully Differential Data Paths**
- **Fixed Input Equalization**
- **Programmable Output Pre-emphasis**
- **Independent Switch and Line Side Pre-emphasis Controls**
- **Programmable Switch-side Loopback Mode**
- **On-chip Terminations**
- **+3.3V Dupply**
- **ESD Rating HBM 6 kV**
- **Lead-less WQFN-48 Package (7mmx7mmx0.8mm, 0.5mm Pitch)**
- **–40°C to +85°C Operating Temperature Range**

## APPLICATIONS

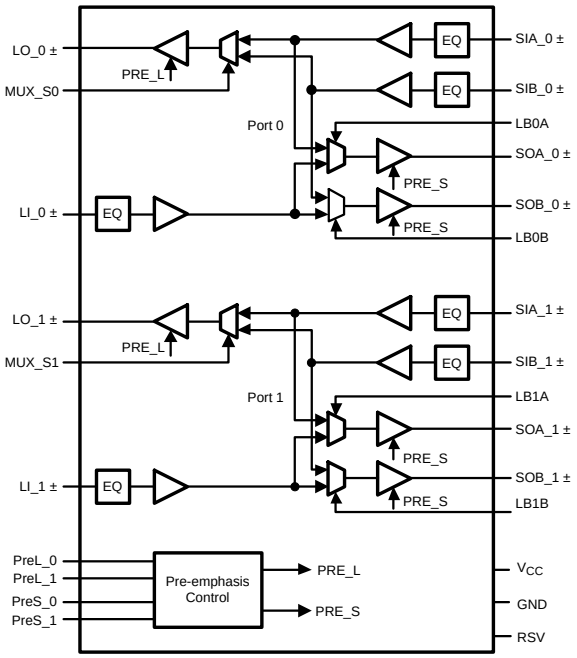
- **Backplane Driver or Cable Driver**
- **Redundancy and Signal Conditioning Applications**
- **XAUI**

## DESCRIPTION

The DS42MB200 is a dual signal conditioning 2:1 multiplexer and 1:2 fan-out buffer designed for use in backplane redundancy applications. Signal conditioning features include input equalization and programmable output pre-emphasis that enable data communication in FR4 backplanes up to 4.25 Gbps. Each input stage has a fixed equalizer to reduce ISI distortion from board traces.

All output drivers have 4 selectable steps of pre-emphasis to compensate for transmission losses from long FR4 backplanes and reduce deterministic jitter. The pre-emphasis levels can be independently controlled for the line-side and switch-side drivers. The internal loopback paths from switch-side input to switch-side output enable at-speed system testing. All receiver inputs are internally terminated with 100Ω differential terminating resistors. All driver outputs are internally terminated with 50Ω to $V_{CC}$.

### Functional Block Diagram



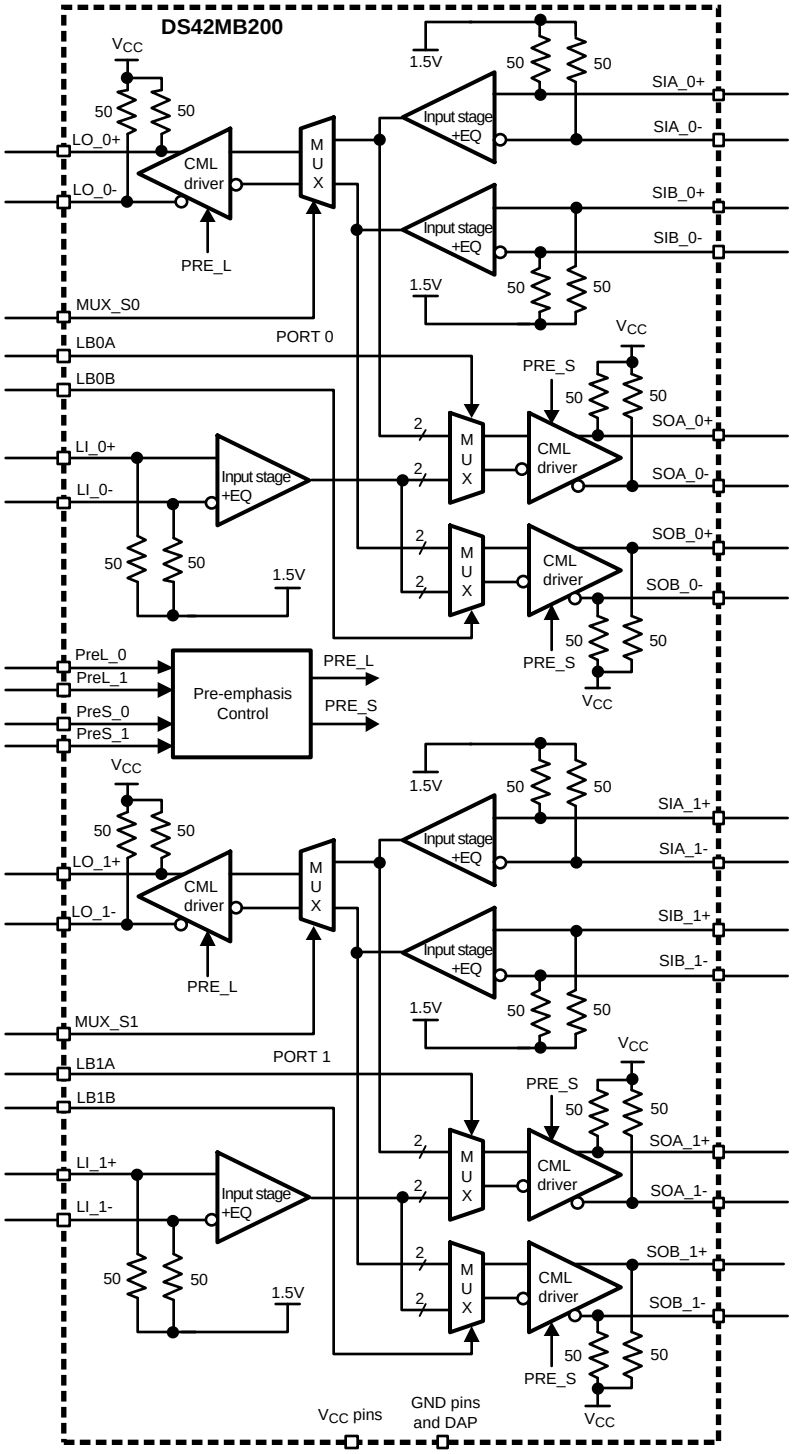

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## Simplified Block Diagram





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## Connection Diagram

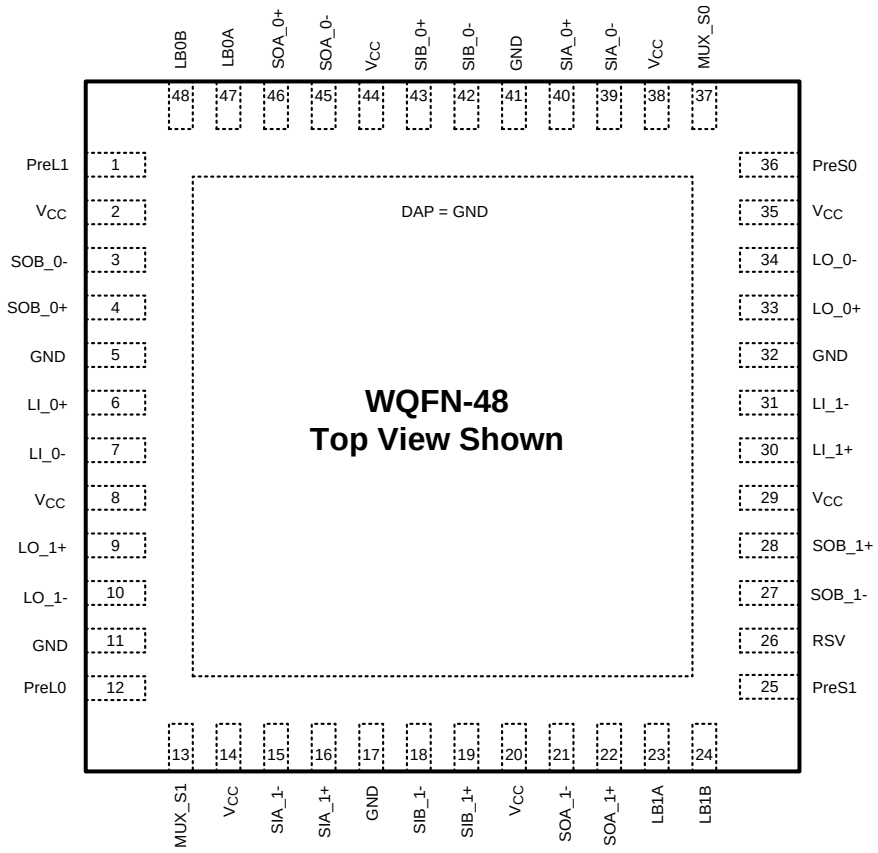



**Figure 1. See Package Number NJU0048D**

### PIN DESCRIPTIONS

| Pin Name | Pin Number | I/O | Description |
|---|---|---|---|
| **LINE SIDE HIGH SPEED DIFFERENTIAL IO's** | | | |
| LI_0+<br>LI_0− | 6<br>7 | I | Inverting and non-inverting differential inputs of port_0 at the line side. LI_0+ and LI_0− have an internal 50Ω connected to an internal reference voltage. See Figure 7. |
| LO_0+<br>LO_0− | 33<br>34 | O | Inverting and non-inverting differential outputs of port_0 at the line side. LO_0+ and LO_0− have an internal 50Ω connected to $V_{CC}$. |
| LI_1+<br>LI_1− | 30<br>31 | I | Inverting and non-inverting differential inputs of port_1 at the line side. LI_1+ and LI_1− have an internal 50Ω connected to an internal reference voltage. See Figure 7. |
| LO_1+<br>LO_1− | 9<br>10 | O | Inverting and non-inverting differential outputs of port_1 at the line side. LO_1+ and LO_1− have an internal 50Ω connected to $V_{CC}$. |
| **SWITCH SIDE HIGH SPEED DIFFERENTIAL IO's** | | | |
| SOA_0+<br>SOA_0− | 46<br>45 | O | Inverting and non-inverting differential outputs of mux_0 at the switch_A side. SOA_0+ and SOA_0− have an internal 50Ω connected to $V_{CC}$. |
| SOB_0+<br>SOB_0− | 4<br>3 | O | Inverting and non-inverting differential outputs of mux_0 at the switch_B side. SOB_0+ and SOB_0− have an internal 50Ω connected to $V_{CC}$. |
| SIA_0+<br>SIA_0− | 40<br>39 | I | Inverting and non-inverting differential inputs to the mux_0 at the switch_A side. SIA_0+ and SIA_0− have an internal 50Ω connected to an internal reference voltage. See Figure 7. |
| SIB_0+<br>SIB_0− | 43<br>42 | I | Inverting and non-inverting differential inputs to the mux_0 at the switch_B side. SIB_0+ and SIB_0− have an internal 50Ω connected to an internal reference voltage. See Figure 7. |
| SOA_1+<br>SOA_1− | 22<br>21 | O | Inverting and non-inverting differential outputs of mux_1 at the switch_A side. SOA_1+ and SOA_1− have an internal 50Ω connected to $V_{CC}$. |
| SOB_1+<br>SOB_1− | 28<br>27 | O | Inverting and non-inverting differential outputs of mux_1 at the switch_B side. SOB_1+ and SOB_1− have an internal 50Ω connected to $V_{CC}$. |

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**PIN DESCRIPTIONS (continued)**

| Pin Name | Pin Number | I/O | Description |
|---|---|---|---|
| SIA_1+<br>SIA_1− | 16<br>15 | I | Inverting and non-inverting differential inputs to the mux_1 at the switch_A side. SIA_1+ and SIA_1− have an internal 50Ω connected to an internal reference voltage. See Figure 7. |
| SIB_1+<br>SIB_1− | 19<br>18 | I | Inverting and non-inverting differential inputs to the mux_1 at the switch_B side. SIB_1+ and SIB_1− have an internal 50Ω connected to an internal reference voltage. See Figure 7. |
| **CONTROL (3.3V LVCMOS)** | | | |
| MUX_S0 | 37 | I | A logic low at MUX_S0 selects mux_0 to switch B. MUX_S0 is internally pulled high. Default state for mux_0 is switch A. |
| MUX_S1 | 13 | I | A logic low at MUX_S1 selects mux_1 to switch B. MUX_S1 is internally pulled high. Default state for mux_1 is switch A. |
| PREL_0<br>PREL_1 | 12<br>1 | I | PREL_0 and PREL_1 select the output pre-emphasis of the line side drivers (LO_0± and LO_1±). PREL_0 and PREL_1 are internally pulled high. See Table 3 for line side pre-emphasis levels. |
| PRES_0<br>PRES_1 | 36<br>25 | I | PRES_0 and PRES_1 select the output pre-emphasis of the switch side drivers (SOA_0±, SOB_0±, SOA_1± and SOB_1±). PRES_0 and PRES_1 are internally pulled high. See Table 4 for switch side pre-emphasis levels. |
| LB0A | 47 | I | A logic low at LB0A enables the internal loopback path from SIA_0± to SOA_0±. LB0A is internally pulled high. |
| LB0B | 48 | I | A logic low at LB0B enables the internal loopback path from SIB_0± to SOB_0±. LB0B is internally pulled high. |
| LB1A | 23 | I | A logic low at LB1A enables the internal loopback path from SIA_1± to SOA_1±. LB1A is internally pulled high. |
| LB1B | 24 | I | A logic low at LB1B enables the internal loopback path from SIB_1± to SOB_1±. LB1B is internally pulled high. |
| RSV | 26 | I | Reserve pin to support factory testing. This pin can be left open, or tied to GND, or tied to GND through an external pull-down resistor. |
| **POWER** | | | |
| $V_{CC}$ | 2, 8, 14, 20, 29, 35, 38, 44 | P | $V_{CC}$ = 3.3V ± 5%.<br>Each $V_{CC}$ pin should be connected to the $V_{CC}$ plane through a low inductance path, typically with a via located as close as possible to the landing pad of the $V_{CC}$ pin.<br>It is recommended to have a 0.01 µF or 0.1 µF, X7R, size-0402 bypass capacitor from each $V_{CC}$ pin to ground plane. |
| GND | 5, 11, 17, 32, 41 | P | Ground reference. Each ground pin should be connected to the ground plane through a low inductance path, typically with a via located as close as possible to the landing pad of the GND pin. |
| GND | DAP | P | Die Attach Pad (DAP) is the metal contact at the bottom side, located at the center of the WQFN-48 package. It should be connected to the GND plane with at least 4 via to lower the ground impedance and improve the thermal performance of the package. |

## Functional Description

The DS42MB200 is a signal conditioning 2:1 multiplexer and a 1:2 buffer designed to support port redundancy up to 4.25 Gbps. Each input stage has a fixed equalizer that provides equalization to compensate about 5 dB of transmission loss from a short backplane trace (about 10 inches backplane). The output driver has pre-emphasis (driver-side equalization) to compensate the transmission loss of the backplane that it is driving. The driver conditions the output signal such that the lower frequency and higher frequency pulses reach approximately the same amplitude at the end of the backplane, and minimize the deterministic jitter caused by the amplitude disparity. The DS42MB200 provides 4 steps of user-selectable pre-emphasis ranging from 0, -3, -6 and –9 dB to handle different lengths of backplane. Figure 1 shows a driver pre-emphasis waveform. The pre-emphasis duration is 200ps nominal, corresponds to 0.8 bit-width at 4 Gbps. The pre-emphasis levels of switch-side and line-side can be individually programmed.

The high speed inputs are self-biased to about 1.5V and are designed for AC coupling allowing the DS42MB200 to be directly inserted into the datapath without any limitation. The ideal AC coupling capacitor value is often based on the lowest frequency component embedded within the serial link. A typical AC coupling capacitor value ranges between 100 and 1000nF, some specifications with scrambled data may require a larger coupling capacitor for optimal performance. To reduce unwanted parasitics around and within the AC coupling capacitor, a body size of 0402 is recommended. *Figure 5* shows the AC coupling capacitor placement in an AC test circuit. The inputs are compatible to most AC coupling differential signals such as LVDS, LVPECL and CML. See Figure 7 for details.


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**Table 1. LOGIC TABLE FOR MULTIPLEX CONTROLS**

| **MUX_S0** | **Mux Function** |
|---|---|
| 0 | MUX_0 select switch_B input, SIB_0±. |
| 1 (default) | MUX_0 select switch_A input, SIA_0±. |
| **MUX_S1** | **Mux Function** |
| 0 | MUX_1 select switch_B input, SIB_1±. |
| 1 (default) | MUX_1 select switch_A input, SIA_0±. |

**Table 2. LOGIC TABLE FOR LOOPBACK Controls**

| **LB0A** | **Loopback Function** |
|---|---|
| 0 | Enable loopback from SIA_0± to SOA_0±. |
| 1 (default) | Normal mode. Loopback disabled. |
| **LB0B** | **Loopback Function** |
| 0 | Enable loopback from SIB_0± to SOB_0±. |
| 1 (default) | Normal mode. Loopback disabled. |
| **LB1A** | **Loopback Function** |
| 0 | Enable loopback from SIA_1± to SOA_1±. |
| 1 (default) | Normal mode. Loopback disabled. |
| **LB1B** | **Loopback Function** |
| 0 | Enable loopback from SIB_1± to SOB_1±. |
| 1 (default) | Normal mode. Loopback disabled. |

**Table 3. LINE-SIDE PRE-EMPHASIS CONTROLS**

| **PreL_[1:0]** | **Pre-Emphasis Level in $mV_{PP}$ (VODB)** | **De-Emphasis Level in $mV_{PP}$ (VODPE)** | **Pre-Emphasis in dB (VODPE/VODB)** | **Typical FR4 board trace** |
|---|---|---|---|---|
| 0 0 | 1200 | 1200 | 0 | 10 inches |
| 0 1 | 1200 | 850 | −3 | 20 inches |
| 1 0 | 1200 | 600 | −6 | 30 inches |
| 1 1 (default) | 1200 | 426 | −9 | 40 inches |

**Table 4. SWITCH-SIDE PRE-EMPHASIS CONTROLS**

| **PreS_[1:0]** | **Pre-Emphasis Level in $mV_{PP}$ (VODB)** | **De-Emphasis Level in $mV_{PP}$ (VODPE)** | **Pre-Emphasis in dB (VODPE/VODB)** | **Typical FR4 board trace** |
|---|---|---|---|---|
| 0 0 | 1200 | 1200 | 0 | 10 inches |
| 0 1 | 1200 | 850 | −3 | 20 inches |
| 1 0 | 1200 | 600 | −6 | 30 inches |
| 1 1 (default) | 1200 | 426 | −9 | 40 inches |

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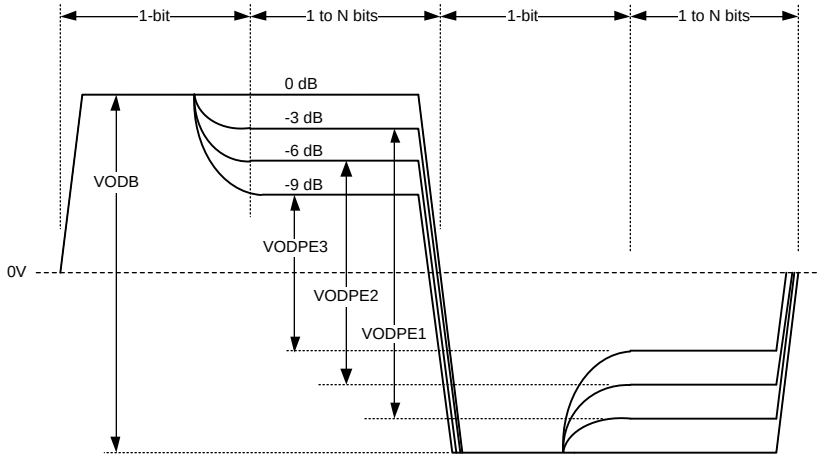



**Figure 2. Driver Pre-Emphasis Differential Waveform (showing all 4 pre-emphasis steps)**

These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

## Absolute Maximum Ratings (1)(2)

| | |
|---|---|
| Supply Voltage ($V_{CC}$) | −0.3V to 4V |
| CMOS/TTL Input Voltage | −0.3V to ($V_{CC}$ +0.3V) |
| CML Input/Output Voltage | −0.3V to ($V_{CC}$ +0.3V) |
| Junction Temperature | +125°C |
| Storage Temperature | −65°C to +150°C |
| Lead Temperature (Soldering, 4 sec.) | +260°C |
| Thermal Resistance, $\theta_{JA}$ | 33.7°C/W |
| Thermal Resistance, $\theta_{JC\text{-}top}$ | 20.7°C/W |
| Thermal Resistance, $\theta_{JC\text{-}bottom}$ | 5.8°C/W |
| Thermal Resistance,$\Phi_{JB}$ | 18.2°C/W |
| ESD Rating HBM, 1.5 kΩ, 100 pF | 6 kV |
| ESD Rating Machine Model | 250V |

(1) "Absolute Maximum Ratings" are the ratings beyond which the safety of the device cannot be verified. They are not meant to imply that the device should be operated at these limits.
(2) If Military/Aerospace specified devices are required, please contact the Texas Instruments Sales Office/ Distributors for availability and specifications.

## Recommended Operating Ratings

| | Min | Typ | Max | Units |
|---|---|---|---|---|
| Supply Voltage ($V_{CC}$-GND) | 3.135 | 3.3 | 3.465 | V |
| Supply Noise Amplitude (10 Hz to 2 GHz) | | | 20 | $mV_{PP}$ |
| Ambient Temperature | -40 | | 85 | °C |
| Case Temperature | | | 100 | °C |


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## Electrical Characteristics

Over recommended operating supply and temperature ranges unless otherwise specified.

| Symbol | Parameter | Conditions | Min | Typ(1) | Max | Units |
|---|---|---|---|---|---|---|
| **LVCMOS DC SPECIFICATIONS** | | | | | | |
| $V_{IH}$ | High Level Input Voltage | | 2.0 | | $V_{CC}$ +0.3 | V |
| $V_{IL}$ | Low Level Input Voltage | | −0.3 | | 0.8 | V |
| $I_{IH}$ | High Level Input Current | $V_{IN} = V_{CC}$ | −10 | | 10 | µA |
| $I_{IL}$ | Low Level Input Current | $V_{IN}$ = GND | 75 | 94 | 124 | µA |
| $R_{PU}$ | Pull-High Resistance | | | 35 | | kΩ |
| **RECEIVER SPECIFICATIONS** | | | | | | |
| $V_{ID}$ | Differential Input Voltage Range | AC Coupled Differential Signal<br>Below 1.25 Gbps<br>At 1.25 Gbps–3.125 Gbps<br>Above 3.125 Gbps<br>This parameter is not production tested. | <br>100<br>100<br>100 | | <br>1750<br>1560<br>1200 | <br>$mV_{P-P}$<br>$mV_{P-P}$<br>$mV_{P-P}$ |
| $V_{ICM}$ | Common Mode Voltage at Receiver Inputs | Measured at receiver inputs reference to ground. | | 1.3 | | V |
| $R_{ITD}$ | Input Differential Termination | On-chip differential termination between IN+ or IN−.(2) | 84 | 100 | 116 | Ω |
| **DRIVER SPECIFICATIONS** | | | | | | |
| VODB | Output Differential Voltage Swing without Pre-Emphasis | $R_L$ = 100Ω ±1%<br>PRES_1=PRES_0=0<br>PREL_1=PREL_0=0<br>Driver pre-emphasis disabled.<br>Running K28.7 pattern at 4.25 Gbps.(3)<br>See Figure 6 for test circuit. | 1000 | 1200 | 1400 | $mV_{P-P}$ |
| $V_{PE}$ | Output Pre-Emphasis Voltage Ratio 20*log(VODPE/VODB) | $R_L$ = 100Ω ±1%<br>Running K28.7 pattern at 4.25 Gbps(3)<br>PREx_[1:0]=00<br>PREx_[1:0]=01<br>PREx_[1:0]=10<br>PREx_[1:0]=11<br>x=S for switch side pre-emphasis control<br>x=L for line side pre-emphasis control<br>See Figure 2 on waveform.<br>See Figure 6 for test circuit. | | <br>0<br>−3<br>−6<br>−9 | | <br>dB<br>dB<br>dB<br>dB |
| $t_{PE}$ | Pre-Emphasis Width(4) | Tested at −9 dB pre-emphasis level, PREx[1:0]=11<br>x=S for switch side pre-emphasis control<br>x=L for line side pre-emphasis control<br>See Figure 5 on measurement condition. | 125 | 200 | 250 | ps |
| $R_{OTSE}$ | Output Termination | On-chip termination from OUT+ or OUT− to $V_{CC}$ | 42 | 50 | 58 | Ω |
| $R_{OTD}$ | Output Differential Termination | On-chip differential termination between OUT+ and OUT− | | 100 | | Ω |
| $\Delta R_{OTSE}$ | Mis-Match in Output Termination Resistors | Mis-match in output terminations at OUT+ and OUT− | | | 5 | % |
| $V_{OCM}$ | Output Common Mode Voltage | | 2.4 | | 2.9 | V |

(1) Typical parameters measured at $V_{CC}$ = 3.3V, $T_A$ = 25°C. They are for reference purposes and are not production-tested.
(2) IN+ and IN− are generic names refer to one of the many pairs of complimentary inputs of the DS42MB200. OUT+ and OUT− are generic names refer to one of the many pairs of the complimentary outputs of the DS42MB200. Differential input voltage $V_{ID}$ is defined as |IN+–IN−|. Differential output voltage $V_{OD}$ is defined as |OUT+–OUT−|.
(3) K28.7 pattern is a 10-bit repeating pattern of K28.7 code group {001111 1000}K28.5 pattern is a 20-bit repeating pattern of +K28.5 and −K28.5 code groups {110000 0101 001111 1010}
(4) Specified by desigh and characterization using statistical analysis.

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## Electrical Characteristics (continued)

Over recommended operating supply and temperature ranges unless otherwise specified.

| Symbol | Parameter | Conditions | Min | Typ(1) | Max | Units |
|---|---|---|---|---|---|---|
| **POWER DISSIPATION** | | | | | | |
| $P_D$ | Power Dissipation | $V_{DD}$ = 3.465V<br>All outputs terminated by 100Ω ±1%.<br>PREL_[1:0]=0, PRES_[1:0]=0<br>Running PRBS $2^7$-1 pattern at 4.25 Gbps | | | 1 | W |
| **AC CHARACTERISTICS** | | | | | | |
| $t_R$ | Differential Low to High Transition Time | Measured with a clock-like pattern at 100 MHz, between 20% and 80% of the differential output voltage. Pre-emphasis disabled.<br>Transition time is measured with fixture as shown in Figure 6, adjusted to reflect the transition time at the output pins. | | 80 | | ps |
| $t_F$ | Differential High to Low Transition Time | | | 80 | | ps |
| $t_{PLH}$ | Differential Low to High Propagation Delay | Measured at 50% differential voltage from input to output. | | 0.5 | 2 | ns |
| $t_{PHL}$ | Differential High to Low Propagation Delay | | | 0.5 | 2 | ns |
| $t_{SKP}$ | Pulse Skew(5) | $\|t_{PHL}–t_{PLH}\|$ | | | 20 | ps |
| $t_{SKO}$ | Output Skew(6)(5) | Difference in propagation delay among data paths in the same device. | | | 200 | ps |
| $t_{SKPP}$ | Part-to-Part Skew(5) | Difference in propagation delay between the same output from devices operating under identical condition. | | | 500 | ps |
| $t_{SM}$ | Mux Switch Time | Measured from $V_{IH}$ or $V_{IL}$ of the mux-control or loopback control to 50% of the valid differential output. | | 1.8 | 6 | ns |
| RJ | Device Random Jitter(7)(5) | See Figure 6 for test circuit.<br>Alternating-1-0 pattern.<br>Pre-emphasis disabled.<br>At 1.25 Gbps<br>At 4.25 Gbps | | | <br><br><br>2<br>2 | <br><br><br>psrms<br>psrms |
| DJ | Device Deterministic Jitter(8)(5) | See Figure 6 for test circuit.<br>Pre-emphasis disabled.<br>At 4.25 Gbps, PRBS7 pattern for DS42MB200@ –40° to 85°C | | | <br><br>35 | <br><br>pspp |
| $DR_{MAX}$ | Maximum Data Rate(5) | Tested with alternating-1-0 pattern | 4.25 | | | Gbps |

(5) Specified by desigh and characterization using statistical analysis.

(6) $t_{SKO}$ is the magnitude difference in the propagation delays among data paths between switch A and switch B of the same port and similar data paths between port 0 and port 1. An example is the output skew among data paths from SIA_0± to LO_0±, SIB_0± to LO_0±, SIA_1± to LO_1± and SIB_1± to LO_1±. Another example is the output skew among data paths from LI_0± to SOA_0±, LI_0± to SOB_0±, LI_1± to SOA_1± and LI_1± to SOB_1±. $t_{SKO}$ also refers to the delay skew of the loopback paths of the same port and between similar data paths between port 0 and port 1. An example is the output skew among data paths SIA_0± to SOA_0±, SIB_0± to SOB_0±, SIA_1± to SOA_1± and SIB_1± to SOB_1±.

(7) Device output random jitter is a measurement of the random jitter contribution from the device. It is derived by the equation sqrt($RJ_{OUT}^2$–$RJ_{IN}^2$), where $RJ_{OUT}$ is the random jitter measured at the output of the device in psrms, $RJ_{IN}$ is the random jitter of the pattern generator driving the device.

(8) Device output deterministic jitter is a measurement of the deterministic jitter contribution from the device. It is derived by the equation ($DJ_{OUT}$–$DJ_{IN}$), where $DJ_{OUT}$ is the peak-to-peak deterministic jitter measured at the output of the device in pspp, $DJ_{IN}$ is the peak-to-peak deterministic jitter of the pattern generator driving the device.


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## Timing Diagrams

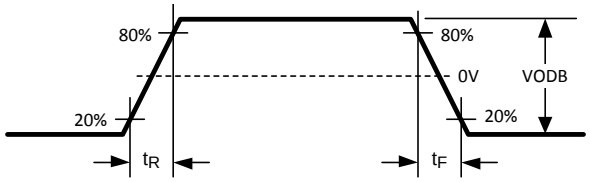



**Figure 3. Driver Output Transition Time**

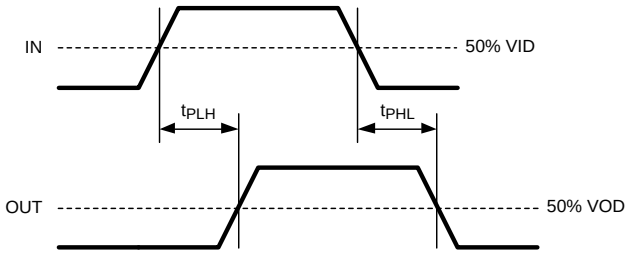



**Figure 4. Propagation Delay from input to output**

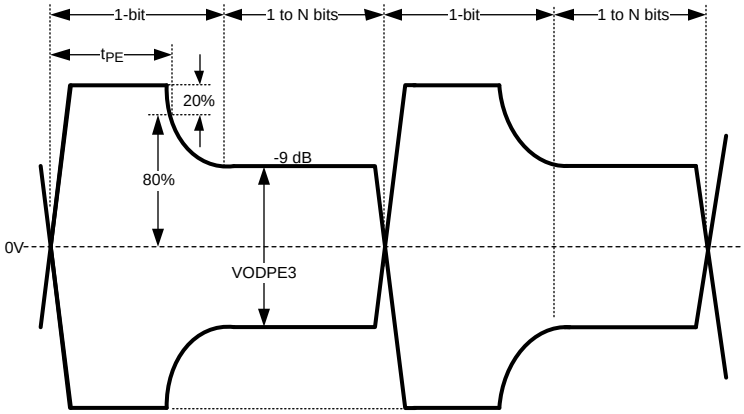



**Figure 5. Test condition for output pre-emphasis duration**

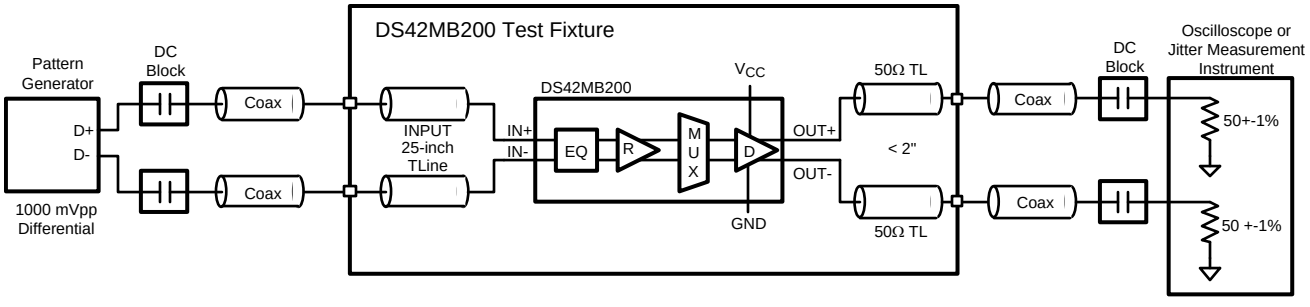



**Figure 6. AC Test Circuit**

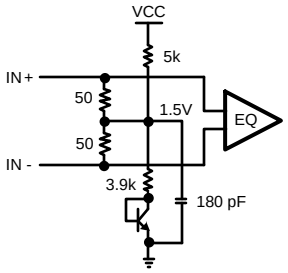



**Figure 7. Receiver Input Termination and Biasing Circuit**

## APPLICATIONS INFORMATION

The DS42MB200 input equalizer provides equalization to compensate about 5 dB of transmission loss from a short backplane transmission line. For characterization purposes, a 25-inch FR4 coupled micro-strip board trace is used in place of the short backplane link. The 25-inch microstrip board trace has approximately 5 dB of attenuation between 375 MHz and 1.875 GHz, representing closely the transmission loss of the short backplane transmission line. The 25-inch microstrip is connected between the pattern generator and the differential inputs of the DS42MB200 for AC measurements.

| Trace Length | Finished Trace Width W | Separation between Traces | Dielectric Height H | Dielectric Constant $\varepsilon_R$ | Loss Tangent |
|---|---|---|---|---|---|
| 25 inches | 8.5 mil | 11.5 mil | 6 mil | 3.8 | 0.022 |

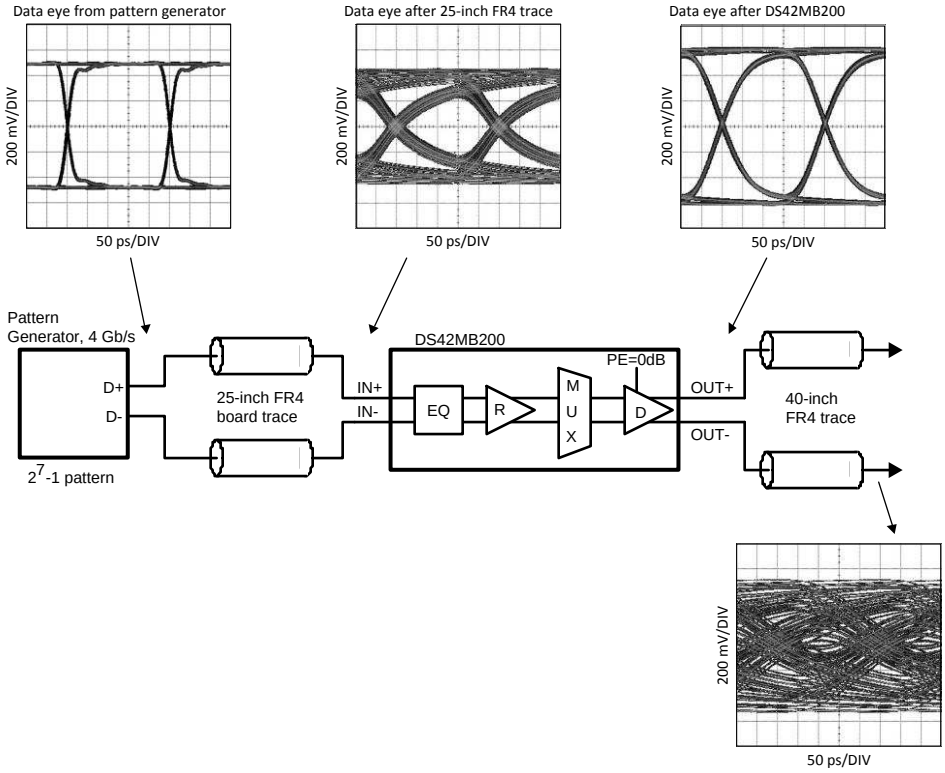



**Figure 8. Data input and output eye patterns with driver set to 0 dB pre-emphasis**


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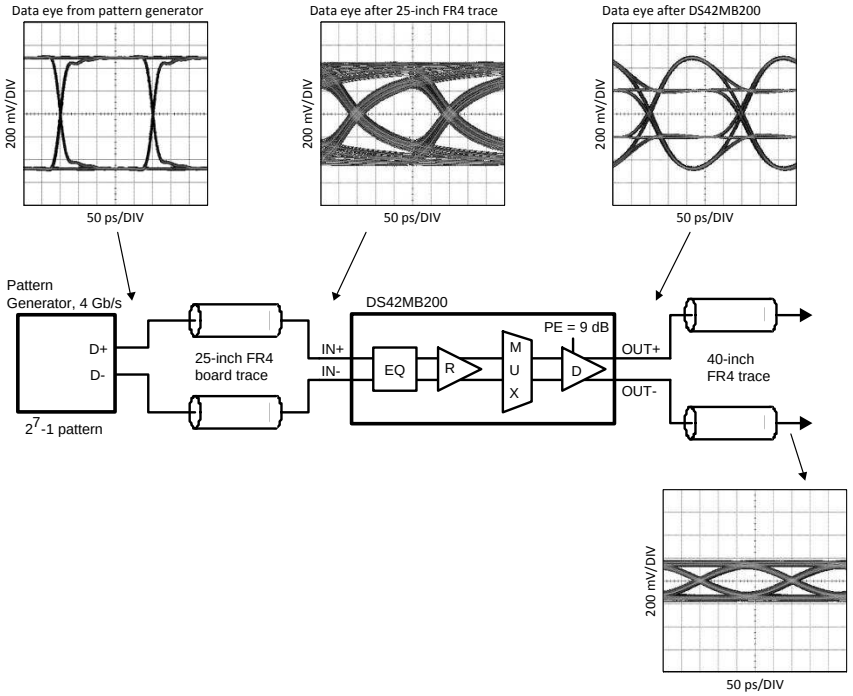



**Figure 9. Data input and output eye patterns with driver set to 9dB pre-emphasis**

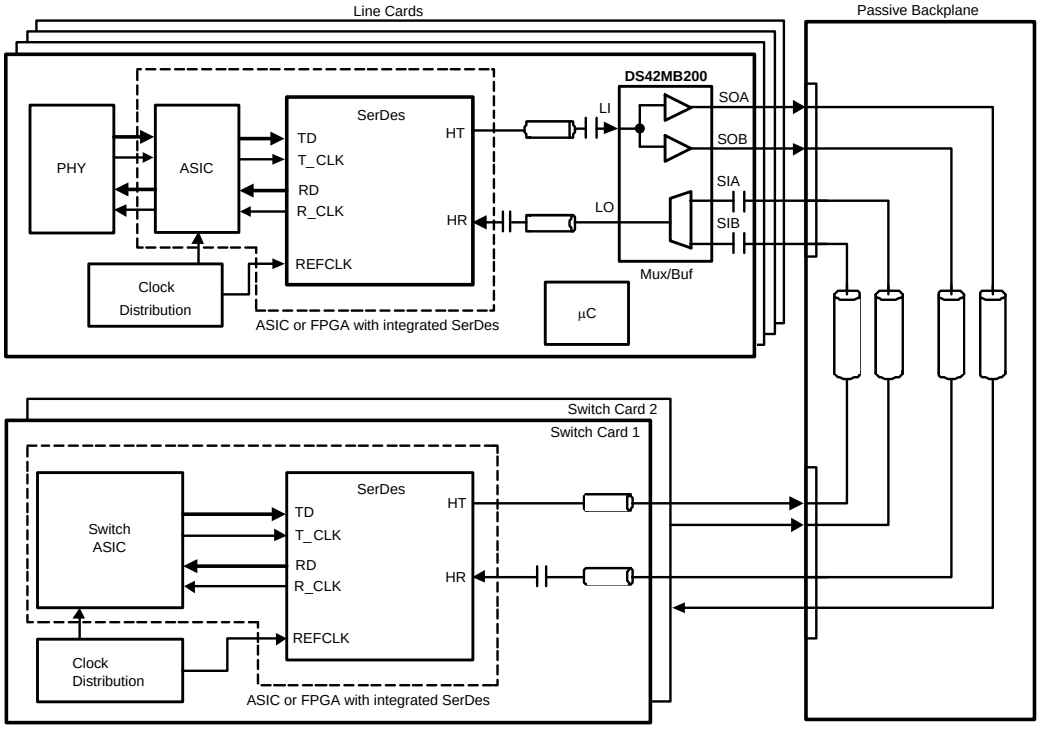



**Figure 10. Application diagram (showing data paths of port 0)**

## REVISION HISTORY

| Changes from Revision F (April 2013) to Revision G | Page |
|---|---|
| • Changed layout of National Data Sheet to TI format | 11 |


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## PACKAGING INFORMATION

| Orderable Device | Status (1) | Package Type | Package Drawing | Pins | Package Qty | Eco Plan (2) | Lead/Ball Finish (6) | MSL Peak Temp (3) | Op Temp (°C) | Device Marking (4/5) | Samples |
|---|---|---|---|---|---|---|---|---|---|---|---|
| DS42MB200TSQ/NOPB | ACTIVE | WQFN | NJU | 48 | 250 | Green (RoHS & no Sb/Br) | SN | Level-3-260C-168 HR | -40 to 85 | 42MB200 | Samples |

(1) The marketing status values are defined as follows:
**ACTIVE:** Product device recommended for new designs.
**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.
**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.
**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.
**OBSOLETE:** TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".
**RoHS Exempt:** TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.
**Green:** TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

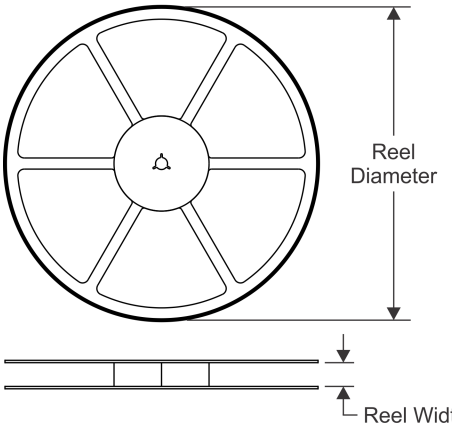
(6) Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

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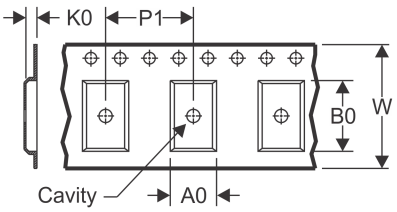
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## TAPE AND REEL INFORMATION

**REEL DIMENSIONS**

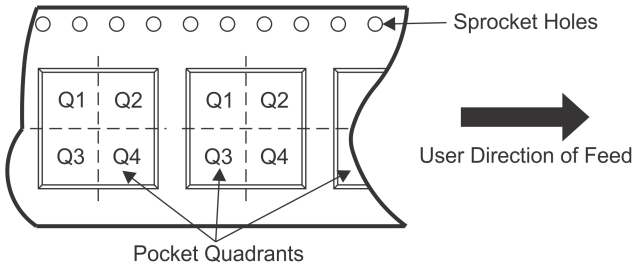



**TAPE DIMENSIONS**




| | |
|---|---|
| A0 | Dimension designed to accommodate the component width |
| B0 | Dimension designed to accommodate the component length |
| K0 | Dimension designed to accommodate the component thickness |
| W | Overall width of the carrier tape |
| P1 | Pitch between successive cavity centers |

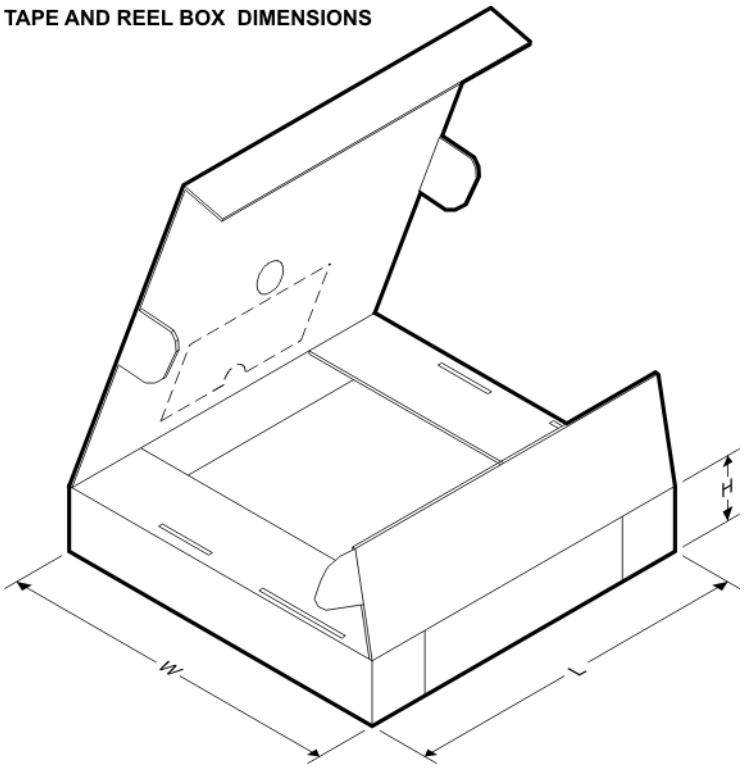
**QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE**




*All dimensions are nominal

| Device | Package Type | Package Drawing | Pins | SPQ | Reel Diameter (mm) | Reel Width W1 (mm) | A0 (mm) | B0 (mm) | K0 (mm) | P1 (mm) | W (mm) | Pin1 Quadrant |
|---|---|---|---|---|---|---|---|---|---|---|---|---|
| DS42MB200TSQ/NOPB | WQFN | NJU | 48 | 250 | 178.0 | 16.4 | 7.3 | 7.3 | 1.3 | 12.0 | 16.0 | Q1 |

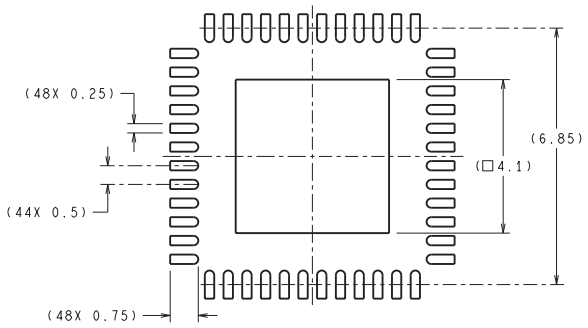
TAPE AND REEL BOX DIMENSIONS




*All dimensions are nominal

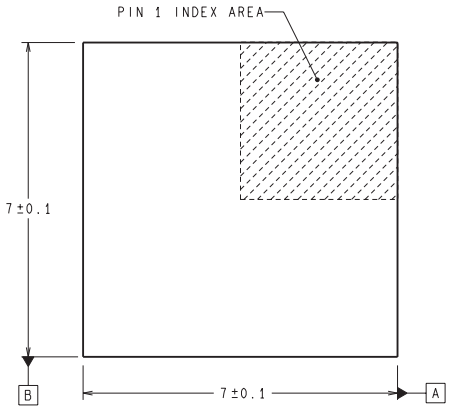
| Device | Package Type | Package Drawing | Pins | SPQ | Length (mm) | Width (mm) | Height (mm) |
|---|---|---|---|---|---|---|---|
| DS42MB200TSQ/NOPB | WQFN | NJU | 48 | 250 | 210.0 | 185.0 | 35.0 |

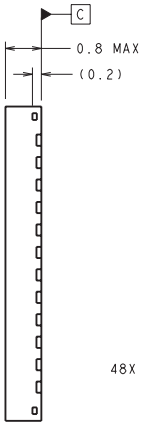
# NJU0048D

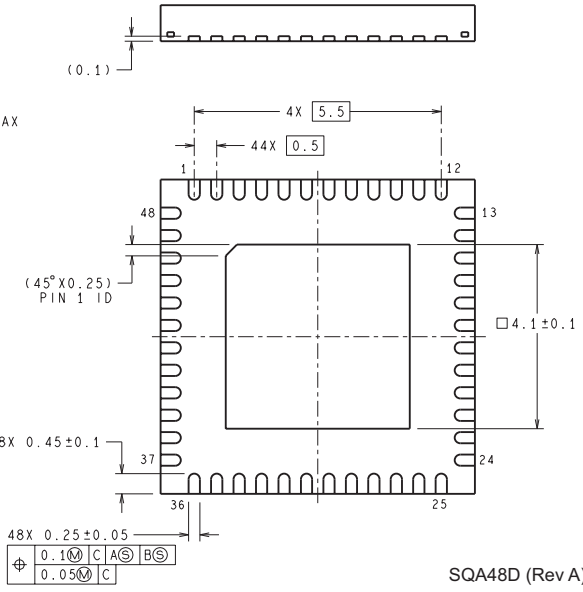



RECOMMENDED LAND PATTERN

DIMENSIONS ARE IN MILLIMETERS
DIMENSIONS IN ( ) FOR REFERENCE ONLY




SQA48D (Rev A)

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