CIRRUS LOGIC®

CS8415A

96 kHz Digital Audio Interface Receiver

Features

- Complete EIAJ CP1201, IEC-60958, AES3, S/PDIF compatible receiver
- +5 V Analog Supply (VA)
- +3 V to +5 V Digital Interface Supply (VL)
- 7:1 S/PDIF Input MUX
- Flexible 3-wire serial digital output port
- 8 kHz to 96 kHz sample frequency range
- Low jitter clock recovery
- Pin and microcontroller read access to Channel Status and User data
- Microcontroller and standalone modes
- Differential cable receiver
- On-chip Channel Status and User data buffer memories
- Auto-detection of compressed audio input streams
- Decodes CD Q sub-code
- OMCK System Clock Mode

General Description

The CS8415A is a monolithic CMOS device which receives and decodes one of 7 channels of audio data according to the IEC60958, S/PDIF, EIAJ CP1201, or AES3. The CS8415A has a serial digital audio output port and comprehensive control ability through a 4-wire microcontroller port. Channel status and user data are assembled in block sized buffers, making read access easy.

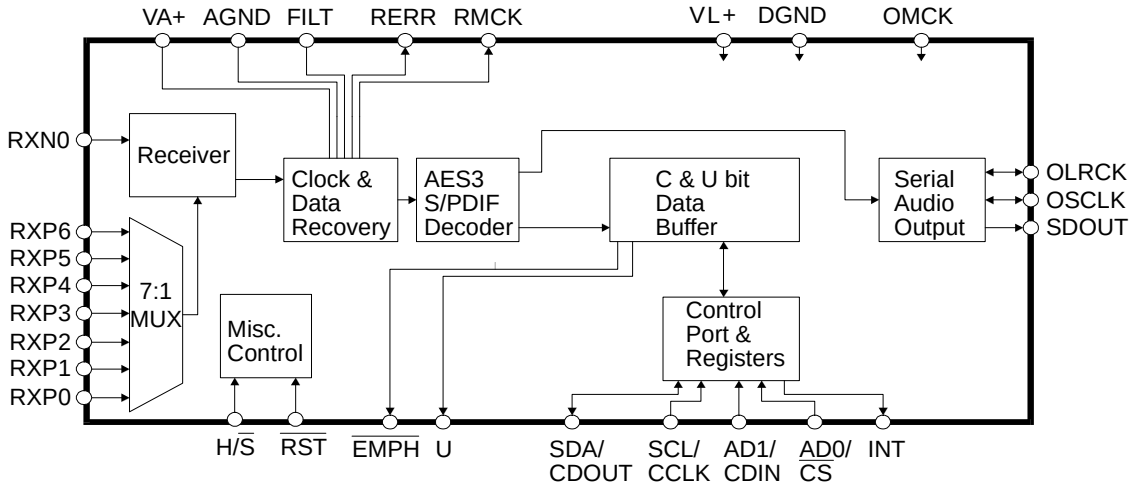
A low jitter clock recovery mechanism yields a very clean recovered clock from the incoming AES3 stream.

Stand-alone operation allows systems with no microcontroller to operate the CS8415A with dedicated output pins for channel status data.

Target applications include A/V receivers, CD-R, DVD receivers, multimedia speakers, digital mixing consoles, effects processors, set-top boxes, and computer and automotive audio systems.

ORDERING INFOMATION

CS8415A-CS	28-pin SOIC	-10 to +70°C
CS8415A-CZ	28-pin TSSOP	-10 to +70°C
CS8415A-IS	28-pin SOIC	-40 to +85°C
CS8415A-IZ	28-pin TSSOP	-40 to +85°C
CDB8415A	Evaluation Board	



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1. CHARACTERISTICS AND SPECIFICATIONS

(All Min/Max characteristics and specifications are guaranteed over the Specified Operating Conditions. Typical performance characteristics and specifications are derived from measurements taken at nominal supply voltages and T_A = 25 °C.)

SPECIFIED OPERATING CONDITIONS (AGND, DGND = 0 V, all voltages with respect to 0V)

Parameter		Symbol	Min	Typ	Max	Units
Power Supply Voltage		VA+	4.5	5.0	5.5	V
	(Note 1)	VL+	2.85	3.3 or 5.0	5.5	V
Ambient Operating Temperature:	'-CS' & '-CZ'	T_A	-10	-	+70	°C
	'-IS' & '-IZ'		-40	-	+85	

Notes: 1. I^2C protocol is supported only in 5V mode.

ABSOLUTE MAXIMUM RATINGS (AGND, DGND = 0 V; all voltages with respect to 0V. Operation beyond these limits may result in permanent damage to the device. Normal operation is not guaranteed at these extremes.)

Parameter		Symbol	Min	Max	Units
Power Supply Voltage		VL+,VA+	-	6.0	V
Input Current, Any Pin Except Supplies	(Note 2)	I_{in}	-	±10	mA
Input Voltage		V_{in}	-0.3	(VL+) + 0.3	V
Ambient Operating Temperature (power applied)		T_A	-55	125	°C
Storage Temperature		T_{stg}	-65	150	°C

Notes: 2. Transient currents of up to 100mA will not cause SCR latch-up.

DC ELECTRICAL CHARACTERISTICS (AGND = DGND = 0 V; all voltages with respect to 0 V.)

Parameters		Symbol	Min	Typ	Max	Units
Power-down Mode (Note 3)						
Supply Current in power down	VA+		-	20	-	µA
	VL+ = 3 V		-	60	-	µA
	VL+ = 5 V		-	60	-	µA
Normal Operation (Note 4)						
Supply Current at 48 kHz frame rate	VA+		-	6.3	-	mA
	VL+ = 3 V		-	30.1	-	mA
	VL+ = 5 V		-	46.5	-	mA
Supply Current at 96 kHz frame rate	VA+		-	6.6	-	mA
	VL+ = 3 V		-	44.8	-	mA
	VL+ = 5 V		-	76.6	-	mA

Notes: 3. Power Down Mode is defined as $\overline{RST}$ = LO with all clocks and data lines held static.

4. Normal operation is defined as $\overline{RST}$ = HI.

DIGITAL INPUT CHARACTERISTICS

Parameters	Symbol	Min	Typ	Max	Units
Input Leakage Current	I_{in}	-	±1	±10	µA
Differential Input Voltage, RXP0 to RXN0	V_{TH}	-	200	-	mV

DIGITAL INTERFACE SPECIFICATIONS (AGND = DGND = 0 V; all voltages with respect to 0 V.)

Parameters	Symbol	Min	Max	Units
High-Level Output Voltage (I_{OH} = -3.2 mA)	V_{OH}	(VL+) - 1.0	-	V
Low-Level Output Voltage (I_{OH} = 3.2 mA)	V_{OL}	-	0.4	V
High-Level Input Voltage, except RX_n	V_{IH}	2.0	(VL+) + 0.3	V
Low-Level Input Voltage, except RX_n (Note 5)	V_{IL}	-0.3	0.4/0.8	V

Notes: 5. At 5V mode, V_{IL} = 0.8V (Max), at 3V mode, V_{IL} =0.4V (Max).

SWITCHING CHARACTERISTICS

(Inputs: Logic 0 = 0V, Logic 1 = VL+; C_L = 20 pF)

Parameter	Symbol	Min	Typ	Max	Units
$\overline{RST}$ pin Low Pulse Width		200	-	-	µs
PLL Clock Recovery Sample Rate Range		8.0	-	108.0	kHz
RMCK output jitter		-	200	-	ps RMS
RMCK output duty cycle (Note 6)		40	50	60	%

Notes: 6. Cycle-to-cycle using 32-96kHz external PLL filter components.

SWITCHING CHARACTERISTICS - SERIAL AUDIO PORTS

(Inputs: Logic 0 = 0 V, Logic 1 = VL+; C_L = 20 pF)

Parameter		Symbol	Min	Typ	Max	Units
OSCLK Active Edge to SDOUT Output Valid	(Note 7)	t_{dpd}	-	-	20	ns
Master Mode						
RMCK to OSCLK active edge delay	(Note 7)	t_{smd}	0	-	10	ns
RMCK to OLRCK delay	(Note 8)	t_{lmd}	0	-	10	ns
OSCLK and OLRCK Duty Cycle			-	50	-	%
Slave Mode						
OSCLK Period	(Note 9)	t_{sckw}	36	-	-	ns
OSCLK Input Low Width		t_{sckl}	14	-	-	ns
OSCLK Input High Width		t_{sckh}	14	-	-	ns
OSCLK Active Edge to OLRCK Edge	(Note 7,8,10)	t_{lrckd}	20	-	-	ns
OLRCK Edge Setup Before OSCLK Active Edge	(Note 7,8,11)	t_{lrcks}	20	-	-	ns

Notes:
7. The active edges of OSCLK are programmable.
8. The polarity OLRCK is programmable.
9. No more than 128 SCLK per frame.
10. This delay is to prevent the previous OSCLK edge from being interpreted as the first one after OLRCK has changed.
11. This setup time ensures that this OSCLK edge is interpreted as the first one after OLRCK has changed.

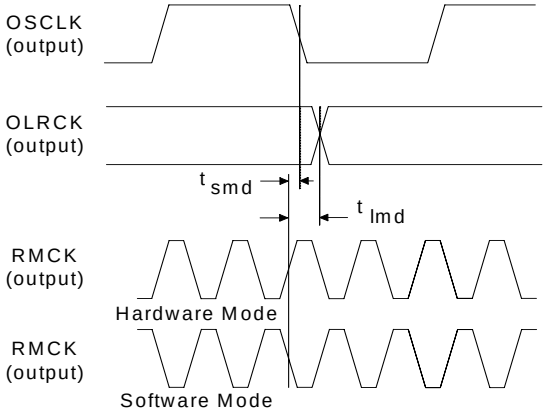


Figure 1. Audio Port Master Mode Timing

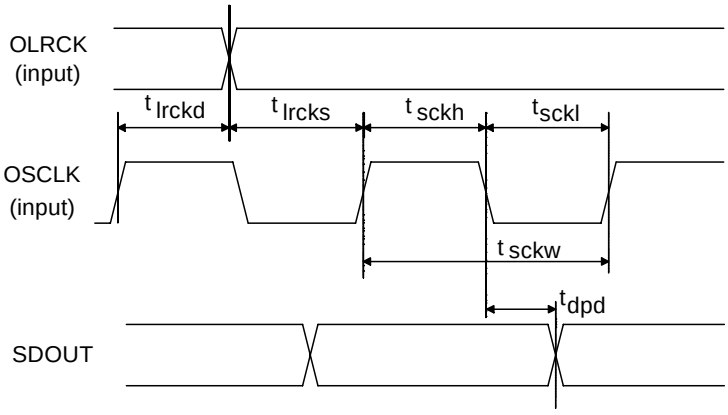


Figure 2. Audio Port Slave Mode and Data Input Timing

SWITCHING CHARACTERISTICS - CONTROL PORT - SPI MODE

(Inputs: Logic 0 = 0 V, Logic 1 = VL+; C_L = 20 pF)

Parameter		Symbol	Min	Typ	Max	Units
CCLK Clock Frequency	(Note 12)	f_{sck}	0	-	6.0	MHz
$\overline{CS}$ High Time Between Transmissions		t_{csh}	1.0	-	-	µs
$\overline{CS}$ Falling to CCLK Edge		t_{css}	20	-	-	ns
CCLK Low Time		t_{scl}	66	-	-	ns
CCLK High Time		t_{sch}	66	-	-	ns
CDIN to CCLK Rising Setup Time		t_{dsu}	40	-	-	ns
CCLK Rising to DATA Hold Time	(Note 13)	t_{dh}	15	-	-	ns
CCLK Falling to CDOUT Stable		t_{pd}	-	-	50	ns
Rise Time of CDOUT		t_{r1}	-	-	25	ns
Fall Time of CDOUT		t_{f1}	-	-	25	ns
Rise Time of CCLK and CDIN	(Note 14)	t_{r2}	-	-	100	ns
Fall Time of CCLK and CDIN	(Note 14)	t_{f2}	-	-	100	ns

Notes: 12. If Fs is lower than 46.875 kHz, the maximum CCLK frequency should be less than 128 Fs. This is dictated by the timing requirements necessary to access the Channel Status and User Bit buffer memory. Access to the control register file can be carried out at the full 6 MHz rate. The minimum allowable input sample rate is 8 kHz, so choosing CCLK to be less than or equal to 1.024 MHz should be safe for all possible conditions.

13. Data must be held for sufficient time to bridge the transition time of CCLK.

14. For f_{sck} <1 MHz.

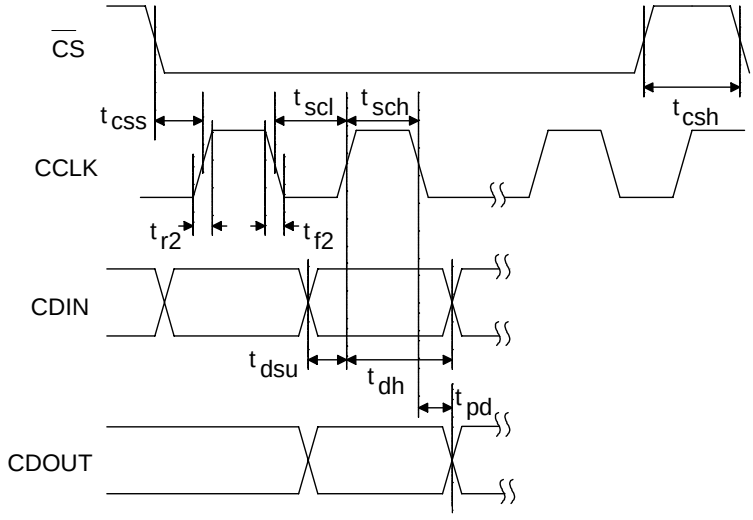


Figure 3. SPI Mode Timing

SWITCHING CHARACTERISTICS - CONTROL PORT - I²C MODE

(Note 15, Inputs: Logic 0 = 0 V, Logic 1 = VL+; C_L = 20 pF)

Parameter	Symbol	Min	Typ	Max	Units
SCL Clock Frequency	f_{scl}	-	-	100	kHz
Bus Free Time Between Transmissions	t_{buf}	4.7	-	-	µs
Start Condition Hold Time (prior to first clock pulse)	t_{hdst}	4.0	-	-	µs
Clock Low Time	t_{low}	4.7	-	-	µs
Clock High Time	t_{high}	4.0	-	-	µs
Setup Time for Repeated Start Condition	t_{sust}	4.7	-	-	µs
SDA Hold Time from SCL Falling (Note 16)	t_{hdd}	0	-	-	µs
SDA Setup Time to SCL Rising	t_{sud}	250	-	-	ns
Rise Time of Both SDA and SCL Lines	t_r	-	-	25	ns
Fall Time of Both SDA and SCL Lines	t_f	-	-	25	ns
Setup Time for Stop Condition	t_{susp}	4.7	-	-	µs

Notes: 15. I²C protocol is supported only in 5V mode.

16. Data must be held for sufficient time to bridge the 25 ns transition time of SCL.

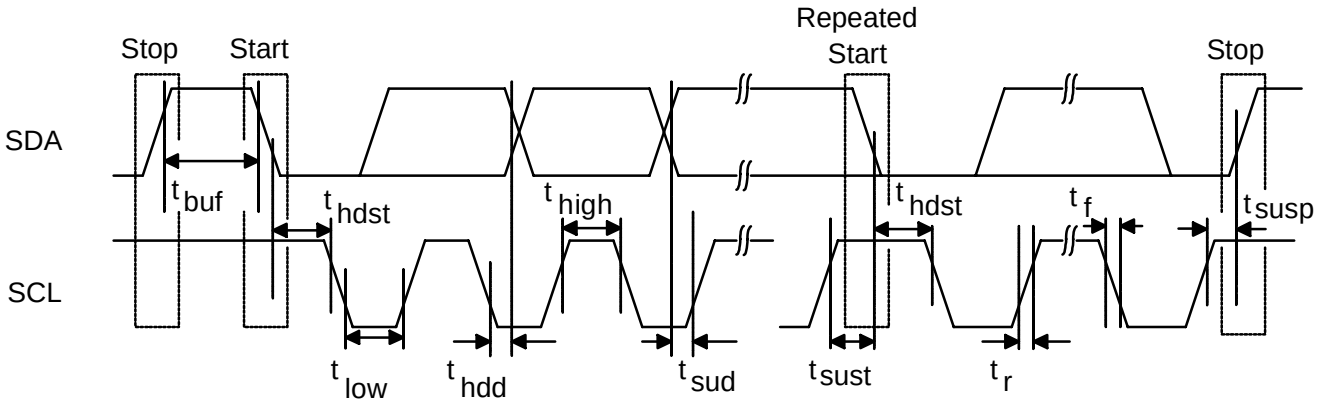
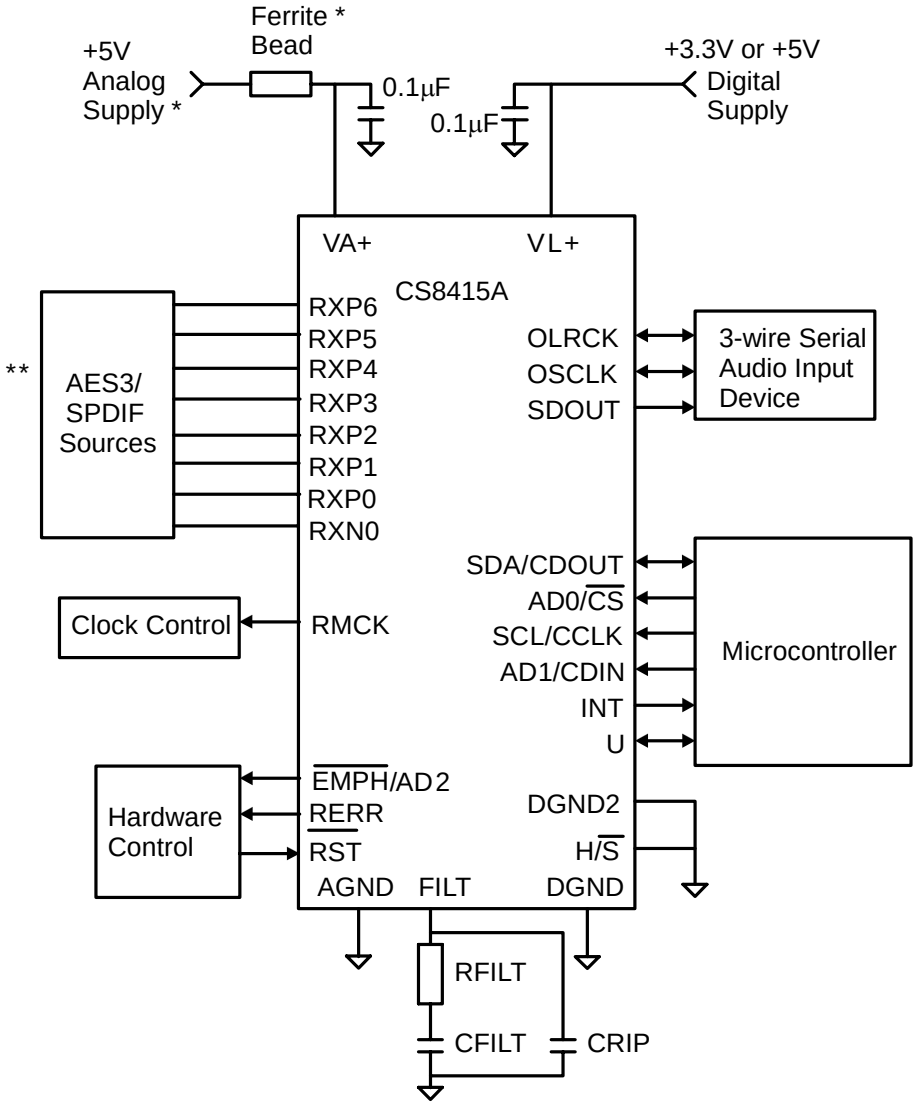


Figure 4. I²C Mode timing

2. TYPICAL CONNECTION DIAGRAM



* A separate analog supply is only necessary in applications where RMCK is used for a jitter sensitive task. For applications where RMCK is not used for a jitter sensitive task, connect VA+ to VD+ via a ferrite bead. Keep the decoupling capacitor between VA+ and AGND.

** Please see section 5.1 "7:1 S/PDIF Input Multiplexer" and Appendix A for typical input configurations and recommended input circuits.

Figure 5. Recommended Connection Diagram for Software Mode

3. GENERAL DESCRIPTION

The CS8415A is a monolithic CMOS device which receives and decodes audio data according to the AES3, IEC60958, S/PDIF, and EIAJ CP1201 interface standards.

Input data is either differential or single-ended. A low jitter clock is recovered from the incoming data using a PLL. The decoded audio data is output through a configurable, 3-wire output port. The channel status and user data are assembled in block sized buffers and may be accessed through an SPI or I^2C microcontroller port. For systems with no microcontroller, a stand alone mode allows direct access to channel status and user data output pins.

Target applications include AVR, CD-R, DAT, DVD, multimedia speakers, MD and VTR equipment, mixing consoles, digital audio transmission and receiving equipment, high quality D/A and A/D converters, effects processors, set-top TV boxes, and computer audio systems.

Figure 5 shows the supply and external connections to the CS8415A, when configured for operation with a microcontroller.

3.1 AES3 and S/PDIF Standards Documents

This data sheet assumes that the user is familiar with the AES3 and S/PDIF data formats. It is advisable to have current copies of the AES3 and IEC60958 specifications on hand for easy reference.

The latest AES3 standard is available from the Audio Engineering Society or ANSI at www.aes.org or www.ansi.org. Obtain the latest IEC60958 standard from ANSI or from the International Electrotechnical Commission at www.iec.ch. The latest EIAJ CP-1201 standard is available from the Japanese Electronics Bureau.

Cirrus Logic Application Note 22: *Overview of Digital Audio Interface Data Structures* contains a useful tutorial on digital audio specifications, but it should not be considered a substitute for the standards.

The paper *An Understanding and Implementation of the SCMS Serial Copy Management System for Digital Audio Transmission*, by Clifton Sanchez, is an excellent tutorial on SCMS. It is available from the AES as preprint 3518.

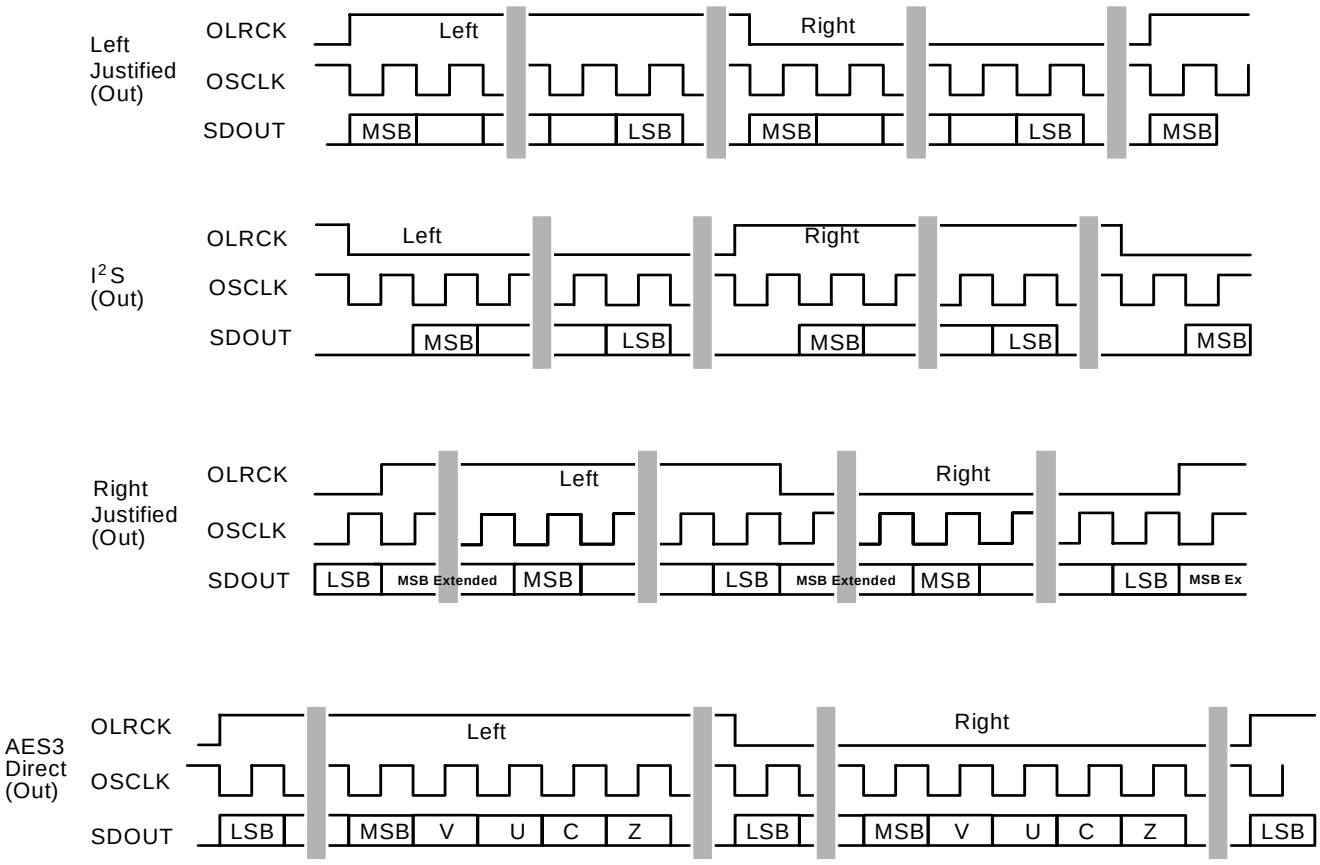
4. SERIAL AUDIO OUTPUT PORT

A 3-wire serial audio output port is provided. The port can be adjusted to suit the attached device setting the control registers. The following parameters are adjustable: master or slave, serial clock frequency, audio data resolution, left or right justification of the data relative to left/right clock, optional one-bit cell delay of the first data bit, the polarity of the bit clock and the polarity of the left/right clock. By setting the appropriate control bits, many formats are possible.

Figure 6 shows a selection of common output formats, along with the control bit settings. It should be noted that in right justified mode, the serial audio output data is "MSB extended". This means that in a sub-frame where the MSB of the data is '1', all bits preceding the MSB in the sub-frame will also be '1'. Conversely, in a sub-frame where the MSB of the data is '0', all bits preceding the MSB in the sub-frame will also be '0'.

A special AES3 direct output format is included, which allows the serial output port access to the V, U, and C bits embedded in the serial audio data stream. The P bit is replaced by a Z bit that marks the start of each block. The received channel status block start signal is only available in hardware mode, as the RCBL pin.

In master mode, the left/right clock and the serial bit clock are outputs, derived from the recovered RMCK clock. In slave mode, the left/right clock and the serial bit clock are inputs. The left/right clock must be synchronous to the appropriate master clock, but the serial bit clock can be asynchronous and discontinuous if required. By appropriate phasing of the left/right clock and control of the serial clocks, multiple CS8415A's can share one serial port. The left/right clock should be continuous, but the duty cycle can be less than the specified typical value of 50% if enough serial clocks are present in each phase to clock all the data bits. When in slave mode, the serial audio output port must not be set for right-justified data. When using the serial audio output port in slave mode with an OLRCK input which is asynchronous to the incoming AES3 data, then an interrupt bit(OSLIP) is provided to indicate when repeated or dropped samples occur.The CS8415A allows immediate mute of the serial audio output port audio data by the MUTESAO bit of Control Register 1.



	SOMS*	SOSF*	SORES[1:0]*	SOJUST*	SODEL*	SOSPOL*	SOLRPOL*
Left Justified	X	X	XX	0	0	0	0
I^2S	X	X	XX	0	1	0	1
Right Justified	1	X	XX	1	0	0	0
AES3 Direct	X	X	11	0	0	0	0

X = don't care to match format, but does need to be set to the desired setting

* See Serial Output Data Format Register Bit Descriptions for an explanation of the meaning of each bit

Figure 6. Serial Audio Output Example Formats

5. AES3 RECEIVER

The CS8415A includes an AES3 digital audio receiver. A comprehensive buffering scheme provides read access to the channel status and user data. This buffering scheme is described in Appendix B.

The AES3 receiver accepts and decodes audio and digital data according to the AES3, IEC60958 (S/PDIF), and EIAJ CP-1201 interface standards. The receiver consists of a differential input stage, driven through pins RXP0 and RXN0, a PLL based clock recovery circuit, and a decoder which separates the audio data from the channel status and user data.

External components are used to terminate and isolate the incoming data cables from the CS8415A. These components are detailed in Appendix A.

5.1 7:1 S/PDIF Input Multiplexer

The CS8415A employs a 7:1 S/PDIF Input Multiplexer to accommodate up to seven channels of input digital audio data. Digital audio data is single-ended and input through the RXP0-6 pins. When any portion of the multiplexer is implemented, unused RXP pins should be tied to ground, and RXN0 must be ac-coupled to ground. The multiplexer select line control is accessed through bits MUX2:0 in the Control 2 register. The multiplexer defaults to RXP0. Therefore, the default configuration is for a differential signal to be input through RXP0 & RXN0. Please see Appendix A for recommended input circuits.

5.2 OMCK System Clock Mode

A special clock switching mode is available that allows the clock that is input through the OMCK pin to be output through the RMCK pin. This feature is controlled by the SWCLK bit in register 1 of the control registers. When the PLL loses lock, the frequency of the VCO drops to 300 kHz. The clock switching mode allows the clock input through OMCK to be used as a clock in the system without any disruption when the PLL loses lock, for example, when the input is removed from the receiver. When SWCLK is enabled and this mode is implemented, RMCK is an output and is not bi-directional. This clock switching is done glitch free. Please note that internal circuitry associated with RMCK is not driven by OMCK. This means that OSCLK and OLRCK continue to be derived from the PLL and are not usable in this mode. This function is available only in software mode.

5.3 PLL, Jitter Attenuation, and Varispeed

Please see Appendix C for general description of the PLL, selection of recommended PLL filter components, and layout considerations. Figure 5 shows the recommended configuration of the two capacitors and one resistor that comprise the PLL filter.

5.4 Error Reporting and Hold Function

While decoding the incoming AES3 data stream, the CS8415A can identify several kinds of error, indicated in the Receiver Error register. The UNLOCK bit indicates whether the PLL is locked to the incoming AES3 data. The V bit reflects the current validity bit status. The CONF (confidence) bit is the logical OR of BIP and UNLOCK. The BIP (biphase) error bit indicates an error in incoming biphase coding. The PAR (parity) bit indicates a received parity error.

The error bits are "sticky": they are set on the first occurrence of the associated error and will remain set until the user reads the register through the control port. This enables the register to log all unmasked errors that occurred since the last time the register was read.

The Receiver Error Mask register allows masking of individual errors. The bits in this register serve as masks for the corresponding bits of the Receiver Error Register. If a mask bit is set to 1, the error is unmasked, which implies the following: its occurrence will be reported in the receiver error register, induce a pulse on RERR, invoke the occurrence of a RERR interrupt, and affect the current audio sample according to the status of the HOLD bits. The HOLD bits allow a choice of holding the previous sample, replacing the current sample with zero (mute), or not changing the current audio sample. If a mask bit is set to 0, the error is masked, which implies the following: its occurrence will not be reported in the receiver error register, will not induce a pulse on RERR or generate a RERR interrupt, and will not affect the current audio sample. The

QCRC and CCRC errors do not affect the current audio sample, even if unmasked.

5.5 Channel Status Data Handling

The first 2 bytes of the Channel Status block are decoded into the Receiver Channel Status register. The setting of the CHS bit in the Channel Status Data Buffer Control register determines whether the channel status decodes are from the A channel (CHS = 0) or B channel (CHS = 1).

The PRO (professional) bit is extracted directly. For consumer data, the COPY (copyright) bit is extracted, and the category code and L bits are decoded to determine SCMS status, indicated by the ORIG (original) bit. If the category code is set to General on the incoming AES3 stream, copyright will always be indicated even when the stream indicates no copyright. Finally, the AUDIO bit is extracted and used to set an AUDIO indicator, as described in the Non-Audio Auto-Detection section below.

If 50/15 µs pre-emphasis is detected, the state of the EMPH pin is adjusted accordingly.

The encoded channel status bits which indicate sample word length are decoded according to AES3-1992 or IEC 60958. Audio data routed to the serial audio output port is unaffected by the word length settings; all 24 bits are passed on as received.

Appendix A describes the overall handling of Channel Status and User data.

5.6 User Data Handling

The incoming user data is buffered in a user accessible buffer. Received User data may also be output to the U pin under the control of a control register bit. Depending on the clocking options selected, there may not be a clock available to qualify the U data output. Figure 7 illustrates the timing. If the incoming user data bits have been encoded as Q-channel subcode, the data is decoded and presented in 10 consecutive register locations. An interrupt may be enabled to indicate the decoding of a new Q-channel block, which may be read through the control port.

5.7 Non-Audio Auto-Detection

An AES3 data stream may be used to convey non-audio data, thus it is important to know whether the incoming AES3 data stream is digital audio or not. This information is typically conveyed in channel status bit 1 (AUDIO), which is extracted automatically by the CS8415A. However, certain non-audio sources, such as AC-3® or MPEG encoders, may not adhere to this convention, and the bit may not be properly set. The CS8415A AES3 receiver can detect such non-audio data. This is accomplished by looking for a 96-bit sync code, consisting of 0x0000, 0x0000, 0x0000, 0x0000, 0xF872, and 0x4E1F. When the sync code is detected, an internal AUTODETECT signal will be asserted. If no additional sync codes are detected within the next 4096 frames, AUTODETECT will be de-asserted until another sync code is detected. The AUDIO bit in the Receiver Channel Status register is the logical OR of AUTODETECT and the received channel status bit 1. If non-audio data is detected, the data is still processed exactly as if it were normal audio. It is up to the user to mute the outputs as required.

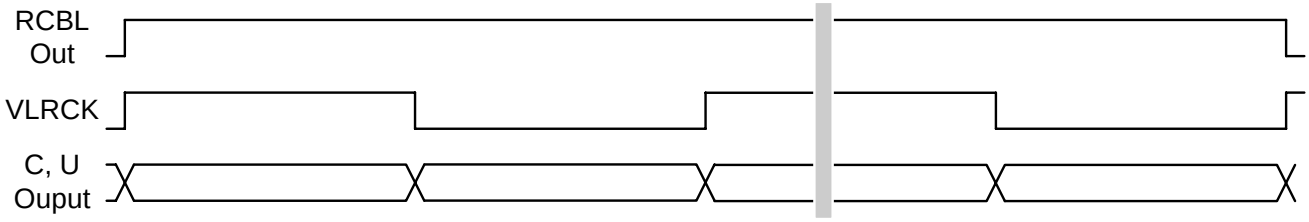
5.8 Mono Mode Operation

An AES3 stream may be used in more than one way to transmit 96 kHz sample rate data. One method is to double the frame rate of the current format. This results in a stereo signal with a sample rate of 96 kHz, carried over a single twisted pair cable. An alternate method is implemented using the 2 sub-frames in a 48 kHz frame rate AES3 signal to carry consecutive samples of a mono signal, resulting in a 96 kHz sample rate stream. This allows older equipment, whose AES3 transmitters and receivers are not rated for 96 kHz frame rate operation, to handle 96 kHz sample rate information. In this “mono mode”, 2 AES3 cables are needed for stereo data transfer. The CS8415A offers mono mode operation, controlled through the MMR control register bit.

The receiver mono mode effectively doubles Fs compared to the input frame rate. The clock output on the RMCK pin tracks Fs, and so is doubled in frequency compared to stereo mode. The receiver will run at a frame rate of Fs/2, and the serial audio output port will run at Fs. Sub-frame A data will be

routed to both the left and right data fields on SDOUT. Similarly, sub-frame B data will be routed to both the left and right data fields of the next word clock cycle of SDOUT.

Using mono mode is only necessary if the serial audio output port must run at 96 kHz. If the CS8415A is kept in normal stereo mode, and receives AES3 data arranged in mono mode, then the serial audio output port will run at 48 kHz, with left and right data fields representing consecutive audio samples.



RCBL and C output are only available in hardware mode.
RCBL goes high 2 frames after receipt of a Z pre-amble, and is high for 16 frames.
VLRCK is a virtual word clock, which may not exist, but is used to illustrate the CU timing.
VLRCK duty cycle is 50%. VLRCK frequency is always equal to the incoming frame rate.
If the serial audio output port is in master mode, VLRCK = OLRCK.
If the serial audio output port is in slave mode, then VLRCK needs to be externally created, if required.
C, U transitions are aligned within ±1% of VLRCK period to VLRCK edges

Figure 7. AES3 Receiver Timing for C & U pin output data

6. CONTROL PORT DESCRIPTION AND TIMING

The control port is used to access the registers, allowing the CS8415A to be configured for the desired operational modes and formats. In addition, Channel Status and User data may be read through the control port. The operation of the control port may be completely asynchronous with respect to the audio sample rates. However, to avoid potential interference problems, the control port pins should remain static if no operation is required.

The control port has 2 modes: SPI and I^2C, with the CS8415A acting as a slave device. SPI mode is selected if there is a high to low transition on the AD0/$\overline{CS}$ pin, after the $\overline{RST}$ pin has been brought high. I^2C mode is selected by connecting the AD0/$\overline{CS}$ pin to VL+ or DGND, thereby permanently selecting the desired AD0 bit address state.

6.1 SPI Mode

In SPI mode, $\overline{CS}$ is the CS8415A chip select signal, CCLK is the control port bit clock (input into the CS8415A from the microcontroller), CDIN is the input data line from the microcontroller, CDOUT is the output data line to the microcontroller. Data is clocked in on the rising edge of CCLK and out on the falling edge.

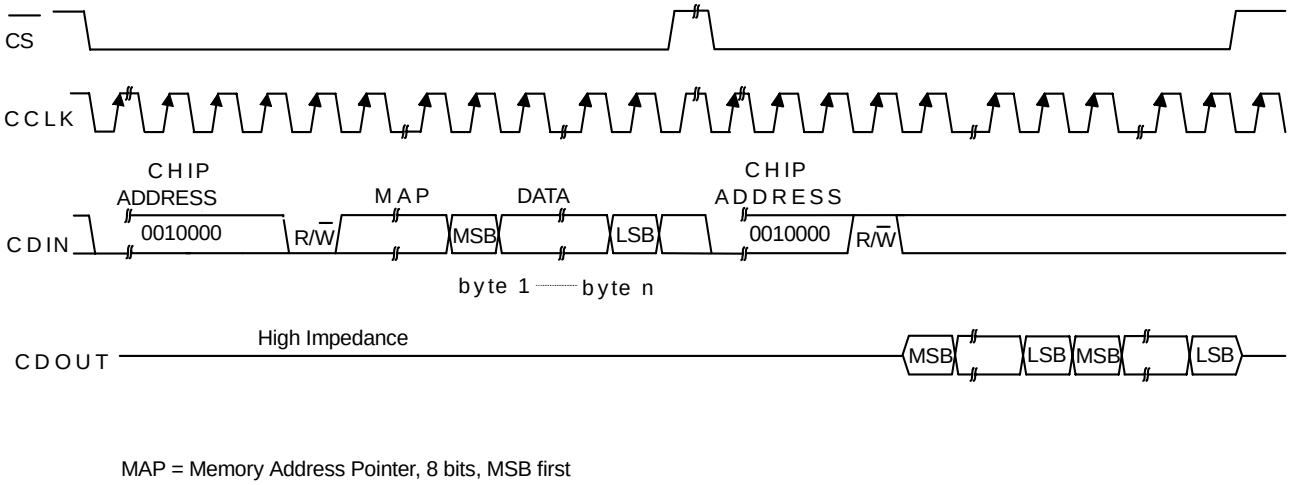
Figure 8 shows the operation of the control port in SPI mode. To write to a register, bring $\overline{CS}$ low. The first seven bits on CDIN form the chip address and must be 0010000. The eighth bit is a read/write indicator (R/$\overline{W}$), which should be low to write. The next eight bits form the Memory Address Pointer (MAP), which is set to the address of the register that is to be updated. The next eight bits are the data which will be placed into the register designated by the MAP. During writes, the CDOUT output stays in the Hi-Z state. It may be externally pulled high or low with a 47 KΩ resistor, if desired.

There is a MAP auto increment capability, enabled by the INCR bit in the MAP register. If INCR is a zero, the MAP will stay constant for successive read or writes. If INCR is set to a 1, the MAP will autoincrement after each byte is read or written, allowing block reads or writes of successive registers.

To read a register, the MAP has to be set to the correct address by executing a partial write cycle which finishes ($\overline{CS}$ high) immediately after the MAP byte. The MAP auto increment bit (INCR) may be set or not, as desired. To begin a read, bring $\overline{CS}$ low, send out the chip address and set the read/write bit (R/$\overline{W}$) high. The next falling edge of CCLK will clock out the MSB of the addressed register (CDOUT will leave the high impedance state). If the MAP auto increment bit is set to 1, the data for successive registers will appear consecutively.

6.2 I^2C Mode

In I^2C mode, SDA is a bidirectional data line. Data is clocked into and out of the part by the clock,



MAP = Memory Address Pointer, 8 bits, MSB first

Figure 8. Control Port Timing in SPI Mode

SCL, with the clock to data relationship as shown in Figure 9. There is no $\overline{CS}$ pin. Each individual CS8415A is given a unique address. Pins AD0 and AD1 form the two least significant bits of the chip address and should be connected to VL+ or DGND as desired. The $\overline{EMPH}$ pin is used to set the AD2 bit by connecting a resistor from the $\overline{EMPH}$ pin to VL+ or to DGND. The state of the pin is sensed while the CS8415A is being reset. The upper 4 bits of the 7-bit address field are fixed at 0010. To communicate with a CS8415A, the chip address field, which is the first byte sent to the CS8415A, should match 0010 followed by the settings of the $\overline{EMPH}$, AD1, and AD0. The eighth bit of the address is the R/$\overline{W}$ bit. If the operation is a write, the next byte is the Memory Address Pointer (MAP) which selects the register to be read or written. If the operation is a read, the contents of the register pointed to by the MAP will be output. Setting the auto increment bit in MAP allows successive reads or writes of consecutive registers. Each byte is separated by an acknowledge bit. The ACK bit is output from the CS8415A after each input byte is read, and is input to the CS8415A from the microcontroller after each transmitted byte. I^2C mode is supported only with VL+ in 5V mode.

6.3 Interrupts

The CS8415A has a comprehensive interrupt capability. The INT output pin is intended to drive the interrupt input pin on the host microcontroller. The INT pin may be set to be active low, active high or active low with no active pull-up transistor. This last mode is used for active low, wired-OR hook-ups, with multiple peripherals connected to the microcontroller interrupt input pin.

Many conditions can cause an interrupt, as listed in the interrupt status register descriptions. Each source may be masked off through mask register bits. In addition, each source may be set to rising edge, falling edge, or level sensitive. Combined with the option of level sensitive or edge sensitive modes within the microcontroller, many different configurations are possible, depending on the needs of the equipment designer.

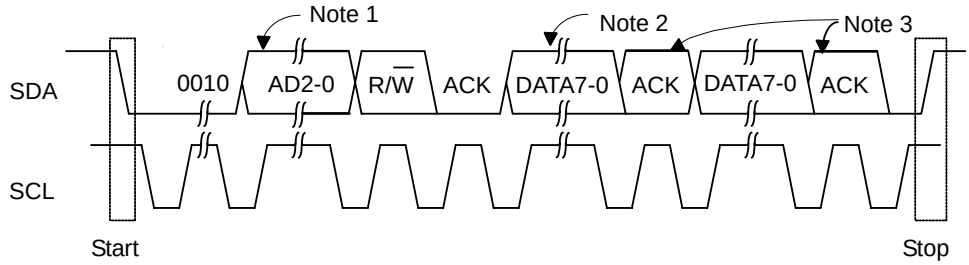


Figure 9. Control Port Timing in I^2C Mode

Notes:
1. AD2 is derived from a resistor attached to the $\overline{EMPH}$ pin. AD1 and AD0 are determined by the state of the corresponding pins.
2. If operation is a write, this byte contains the Memory Address Pointer, MAP.
3. If operation is a read, the last bit of the read should be NACK (high).

7. CONTROL PORT REGISTER SUMMARY

Addr (HEX)	Function	7	6	5	4	3	2	1	0
01	Control 1	SWCLK	0	MUTESAO	0	0	INT1	INT0	0
02	Control 2	0	HOLD1	HOLD0	RMCKF	MMR	MUX2	MUX1	MUX0
04	Clock Source Control	0	RUN	0	0	0	0	0	1
06	Serial Output Format	SOMS	SOSF	SORES1	SORES0	SOJUST	SODEL	SOSPOL	SOLRPOL
07	Interrupt 1 Status	0	OSLIP	0	0	0	DETC	0	RERR
08	Interrupt 2 Status	0	0	0	0	DETU	0	QCH	0
09	Interrupt 1 Mask	0	OSLIPM	0	0	0	DETCM	0	RERRM
0A	Interrupt 1 Mode (MSB)	0	OSLIP1	0	0	0	DETC1	0	RERR1
0B	Interrupt 1 Mode (LSB)	0	OSLIP0	0	0	0	DETC0	0	RERR0
0C	Interrupt 2 Mask	0	0	0	0	DETUM	0	QCHM	0
0D	Interrupt 2 Mode (MSB)	0	0	0	0	DETU1	0	QCH1	0
0E	Interrupt 2 Mode (LSB)	0	0	0	0	DETU0	0	QCH0	0
0F	Receiver CS Data	AUX3	AUX2	AUX1	AUX0	PRO	$\overline{\text{AUDIO}}$	COPY	ORIG
10	Receiver Errors	0	QCRC	CCRC	UNLOCK	V	CONF	BIP	PAR
11	Receiver Error Mask	0	QCRCM	CCRCM	UNLOCKM	VM	CONFM	BIPM	PARM
12	CS Data Buffer Control	0	0	BSEL	CBMR	DETCI	0	CAM	CHS
13	U Data Buffer Control	0	0	0	0	0	0	DETUI	0
14-1D	Q sub-code Data								
1E	OMCK/RMCK Ratio	ORR7	ORR6	ORR5	ORR4	ORR3	ORR2	ORR1	ORR0
20-37	C or U Data Buffer								
7F	ID and Version	ID3	ID2	ID1	ID0	VER3	VER2	VER1	VER0

Table 1. Control Register Map Summary

7.1 Memory Address Pointer (MAP)

7	6	5	4	3	2	1	0
INCR	MAP6	MAP5	MAP4	MAP3	MAP2	MAP1	MAP0

INCR - Auto Increment Address Control Bit

Default = '0'

0 - Disabled
1 - Enabled

MAP6:MAP0 - Register address

Note: Reserved registers must not be written to during normal operation. Some reserved registers are used for test modes, which can completely alter the normal operation of the CS8415A.

8. CONTROL PORT REGISTER BIT DEFINITIONS

8.1 Control 1(01h)

7	6	5	4	3	2	1	0
SWCLK	0	MUTESAO	0	0	INT1	INT0	0

SWCLK - Controls output of OMCK on RMCK when PLL loses lock

Default = '0'

0 - RMCK default function
1 - OMCK output on RMCK pin

MUTESAO - Mute control for the serial audio output port

Default = '0'

0 - Disabled
1 - Enabled

INT1:0 - Interrupt output pin (INT) control

Default = '00'

00 - Active high; high output indicates interrupt condition has occurred
01 - Active low, low output indicates an interrupt condition has occurred
10 - Open drain, active low. Requires an external pull-up resistor on the INT pin.
11 - Reserved

8.2 Control 2 (02h)

7	6	5	4	3	2	1	0
0	HOLD1	HOLD0	RMCKF	MMR	MUX2	MUX1	MUX0

HOLD1:0 - Determine how received audio sample is affected when a receiver error occurs

Default = '00'

00 - Hold the last valid audio sample
01 - Replace the current audio sample with 00 (mute)
10 - Do not change the received audio sample
11 - Reserved

RMCKF - Select recovered master clock output pin frequency.

Default = '0'

0 - RMCK is equal to 256 * Fs
1 - RMCK is equal to 128 * Fs

MMR - Select AES3 receiver mono or stereo operation

Default = '0'

0 - Normal stereo operation
1 - A and B subframes treated as consecutive samples of one channel of data.
Data is duplicated to both left and right parallel outputs of the AES receiver block.
The sample rate (Fs) is doubled compared to MMR=0

MUX2:0 - 7:1 S/PDIF Input Multiplexer Select Line Control

Default = '000'

000 - RXP0
001 - RXP1
010 - RXP2
011 - RXP3
100 - RXP4
101 - RXP5
110 - RXP6
111 - Reserved

8.3 Clock Source Control (04h)

7	6	5	4	3	2	1	0
0	RUN	0	0	0	0	0	0

This register configures the clock sources of various blocks. In conjunction with the Data Flow Control register, various Receiver/Transmitter/Transceiver modes may be selected.

RUN - Controls the internal clocks, allowing the CS8415A to be placed in a "powered down", low current consumption, state.

Default = '0'

0 - Internal clocks are stopped. Internal state machines are reset. The fully static control port is operational, allowing registers to be read or changed. Reading and writing the U and C data buffers is not possible. Power consumption is low.

1 - Normal part operation. This bit must be written to the 1 state to allow the CS8415A to begin operation. All input clocks should be stable in frequency and phase when RUN is set to 1.

8.4 Serial Audio Output Port Data Format (06h)

7	6	5	4	3	2	1	0
SOMS	SOSF	SORES1	SORES0	SOJUST	SODEL	SOSPOL	SOLRPOL

SOMS - Master/Slave Mode Selector

Default = '0'

0 - Serial audio output port is in slave mode
1 - Serial audio output port is in master mode

SOSF - OSCLK frequency (for master mode)

Default = '0'

0 - 64*Fs
1 - 128*Fs

SORES1:0 - Resolution of the output data on SDOUT

Default = '00'

00 - 24-bit resolution
01 - 20-bit resolution
10 - 16-bit resolution
11 - Direct copy of the received NRZ data from the AES3 receiver (including C, U, and V bits, the time slot normally occupied by the P bit is used to indicate the location of the block start,

SDOUT pin only, serial audio output port clock must be derived from the AES3 receiver recovered clock)

SOJUST - Justification of SDOUT data relative to OLRCK

Default = '0'

0 - Left-justified
1 - Right-justified (master mode only)

SODEL - Delay of SDOUT data relative to OLRCK, for left-justified data formats

Default = '0'

0 - MSB of SDOUT data occurs in the first OSCLK period after the OLRCK edge
1 - MSB of SDOUT data occurs in the second OSCLK period after the OLRCK edge

SOSPOL - OSCLK clock polarity

Default = '0'

0 - SDOUT sampled on rising edges of OSCLK
1 - SDOUT sampled on falling edges of OSCLK

SOLRPOL - OLRCK clock polarity

Default = '0'

0 - SDOUT data is for the left channel when OLRCK is high
1 - SDOUT data is for the right channel when OLRCK is high

8.5 Interrupt 1 Status (07h) (Read Only)

7	6	5	4	3	2	1	0
0	OSLIP	0	0	0	DETC	0	RERR

For all bits in this register, a "1" means the associated interrupt condition has occurred at least once since the register was last read. A "0" means the associated interrupt condition has NOT occurred since the last reading of the register. Reading the register resets all bits to 0, unless the interrupt mode is set to level and the interrupt source is still true. Status bits that are masked off in the associated mask register will always be "0" in this register. This register defaults to 00h.

OSLIP - Serial audio output port data slip interrupt

When the serial audio output port is in slave mode, and OLRCK is asynchronous to the port data source, This bit will go high every time a data sample is dropped or repeated.

DETC - D to E C-buffer transfer interrupt.

The source for this bit is true during the D to E buffer transfer in the C bit buffer management process.

RERR - A receiver error has occurred.

The Receiver Error register may be read to determine the nature of the error which caused the interrupt.

8.6 Interrupt 2 Status (08h) (Read Only)

7	6	5	4	3	2	1	0
0	0	0	0	DETU	0	QCH	0

For all bits in this register, a “1” means the associated interrupt condition has occurred at least once since the register was last read. A ”0” means the associated interrupt condition has NOT occurred since the last reading of the register. Reading the register resets all bits to 0, unless the interrupt mode is set to level and the interrupt source is still true. Status bits that are masked off in the associated mask register will always be “0” in this register. This register defaults to 00h.

DETU - D to E U-buffer transfer interrupt. (Block Mode only)

The source of this bit is true during the D to E buffer transfer in the U bit buffer management process.

QCH - A new block of Q-subcode data is available for reading.

The data must be completely read within 588 AES3 frames after the interrupt occurs to avoid corruption of the data by the next block.

8.7 Interrupt 1 Mask (09h)

7	6	5	4	3	2	1	0
0	OSLIPM	0	0	0	DETCM	0	RERRM

The bits of this register serve as a mask for the Interrupt 1 register. If a mask bit is set to 1, the error is unmasked, meaning that its occurrence will affect the INT pin and the status register. If a mask bit is set to 0, the error is masked, meaning that its occurrence will not affect the INT pin or the status register. The bit positions align with the corresponding bits in Interrupt 1 register. This register defaults to 00h.

8.8 Interrupt 1 Mode MSB (0Ah) and Interrupt 1 Mode LSB(0Bh)

7	6	5	4	3	2	1	0
0	OSLIP1	0	0	0	DETC1	0	RERR1
0	OSLIP0	0	0	0	DETC0	0	RERR0

The two Interrupt Mode registers form a 2-bit code for each Interrupt Register 1 function. There are three ways to set the INT pin active in accordance with the interrupt condition. In the Rising edge active mode, the INT pin becomes active on the arrival of the interrupt condition. In the Falling edge active mode, the INT pin becomes active on the removal of the interrupt condition. In Level active mode, the INT interrupt pin becomes active during the interrupt condition. Be aware that the active level(Actice High or Low) only depends on the INT[1:0] bits. These registers default to 00.

00 - Rising edge active
01 - Falling edge active
10 - Level active
11 - Reserved

8.9 Interrupt 2 Mask (0Ch)

7	6	5	4	3	2	1	0
0	0	0	0	DETUM	0	QCHM	0

The bits of this register serve as a mask for the Interrupt 2 register. If a mask bit is set to 1, the error is unmasked, meaning that its occurrence will affect the INT pin and the status register. If a mask bit is set to 0, the error is masked, meaning that its occurrence will not affect the INT pin or the status register. The bit positions align with the corre-

sponding bits in Interrupt 2 register. This register defaults to 00h.

8.10 Interrupt 2 Mode MSB (0Dh) and Interrupt 2 Mode LSB(0Eh)

7	6	5	4	3	2	1	0
0	0	0	0	DETU1	0	QCH1	0
0	0	0	0	DETU0	0	QCH0	0

The two Interrupt Mode registers form a 2-bit code for each Interrupt Register 1 function. There are three ways to set the INT pin active in accordance with the interrupt condition. In the Rising edge active mode, the INT pin becomes active on the arrival of the interrupt condition. In the Falling edge active mode, the INT pin becomes active on the removal of the interrupt condition. In Level active mode, the INT interrupt pin becomes active during the interrupt condition. Be aware that the active level(Actice High or Low) only depends on the INT[1:0] bits. These registers default to 00.

00 - Rising edge active
01 - Falling edge active
10 - Level active
11 - Reserved

8.11 Receiver Channel Status (0Fh) (Read Only)

7	6	5	4	3	2	1	0
AUX3	AUX2	AUX1	AUX0	PRO	AUDIO	COPY	ORIG

The bits in this register can be associated with either channel A or B of the received data. The desired channel is selected with the CHS bit of the Channel Status Data Buffer Control Register.

AUX3:0 - Incoming auxiliary data field width, as indicated by the incoming channel status bits, decoded according to IEC60958 and AES3.

0000 - Auxiliary data is not present
0001 - Auxiliary data is 1 bit long
0010 - Auxiliary data is 2 bits long
0011 - Auxiliary data is 3 bits long
0100 - Auxiliary data is 4 bits long
0101 - Auxiliary data is 5 bits long
0110 - Auxiliary data is 6 bits long
0111 - Auxiliary data is 7 bits long
1000 - Auxiliary data is 8 bits long
1001 - 1111 Reserved

PRO - Channel status block format indicator

0 - Received channel status block is in consumer format
1 - Received channel status block is in professional format

$\overline{\text{AUDIO}}$ - Audio indicator

0 - Received data is linearly coded PCM audio
1 - Received data is not linearly coded PCM audio

COPY - SCMS copyright indicator

0 - Copyright asserted
1 - Copyright not asserted

If the category code is set to General in the incoming AES3 stream, copyright will always be indicated by COPY, even when the stream indicates no copyright.

ORIG - SCMS generation indicator, decoded from the category code and the L bit.

0 - Received data is 1st generation or higher
1 - Received data is original

Note: COPY and ORIG will both be set to 1 if the incoming data is flagged as professional, or if the receiver is not in use.

8.12 Receiver Error (10h) (Read Only)

7	6	5	4	3	2	1	0
0	QCRC	CCRC	UNLOCK	V	CONF	BIP	PAR

This register contains the AES3 receiver and PLL status bits. Unmasked bits will go high on occurrence of the error, and will stay high until the register is read. Reading the register resets all bits to 0, unless the error source is still true. Bits that are masked off in the receiver error mask register will always be 0 in this register. This register defaults to 00h.

QCRC - Q-subcode data CRC error indicator. Updated on Q-subcode block boundaries

0 - No error
1 - Error

CCRC - Channel Status Block Cyclic Redundancy Check bit. Updated on CS block boundaries, valid in Pro mode

0 - No error
1 - Error

UNLOCK - PLL lock status bit. Updated on CS block boundaries.

0 - PLL locked
1 - PLL out of lock

V - Received AES3 Validity bit status. Updated on sub-frame boundaries.

0 - Data is valid and is normally linear coded PCM audio
1 - Data is invalid, or may be valid compressed audio

CONF - Confidence bit. Updated on sub-frame boundaries.

0 - No error
1 - Confidence error. This is the logical OR of BIP and UNLOCK.

BIP - Bi-phase error bit. Updated on sub-frame boundaries.

0 - No error
1 - Bi-phase error. This indicates an error in the received bi-phase coding.

PAR - Parity bit. Updated on sub-frame boundaries.

0 - No error
1 - Parity error

8.13 Receiver Error Mask (11h)

7	6	5	4	3	2	1	0
0	QCRCM	CCRCM	UNLOCKM	VM	CONFM	BIPM	PARM

The bits in this register serve as masks for the corresponding bits of the Receiver Error Register. If a mask bit is set to 1, the error is unmasked, meaning that its occurrence will appear in the receiver error register, will affect the RERR pin, will affect the RERR interrupt, and will affect the current audio sample according to the status of the HOLD bit. If a mask bit is set to 0, the error is masked, meaning that its occurrence will not appear in the receiver error register,

will not affect the RERR pin, will not affect the RERR interrupt, and will not affect the current audio sample. The CCRC and QCRC bits behave differently from the other bits: they do not affect the current audio sample even when unmasked. This register defaults to 00h.

8.14 Channel Status Data Buffer Control (12h)

7	6	5	4	3	2	1	0
0	0	BSEL	CBMR	DETCI	0	CAM	CHS

BSEL - Selects the data buffer register addresses to contain User data or Channel Status data

Default = '0'

0 - Data buffer address space contains Channel Status data
1 - Data buffer address space contains User data

CBMR - Control for the first 5 bytes of channel status "E" buffer

Default = '0'

0 - Allow D to E buffer transfers to overwrite the first 5 bytes of channel status data
1 - Prevent D to E buffer transfers from overwriting first 5 bytes of channel status data

DETCI - D to E C-data buffer transfer inhibit bit.

Default = '0'

0 - Allow C-data D to E buffer transfers
1 - Inhibit C-data D to E buffer transfers

CAM - C-data buffer control port access mode bit

Default = '0'

0 - One byte mode
1 - Two byte mode

CHS - Channel select bit

Default = '0'

0 - Channel A information is displayed at the $\overline{\text{EMPH}}$ pin and in the receiver channel status register. Channel A information is output during control port reads when CAM is set to 0 (One Byte Mode)
1 - Channel B information is displayed at the $\overline{\text{EMPH}}$ pin and in the receiver channel status register. Channel B information is output during control port reads when CAM is set to 0 (One Byte Mode)

8.15 User Data Buffer Control (13h)

7	6	5	4	3	2	1	0
0	0	0	0	0	0	DETUI	0

DETUI - D to E U-data buffer transfer inhibit bit (valid in block mode only).

Default = '0'

0 - Allow U-data D to E buffer transfers
1 - Inhibit U-data D to E buffer transfers

8.16 Q-Channel Subcode Bytes 0 to 9 (14h - 1Dh) (Read Only)

The following 10 registers contain the decoded Q-channel subcode data

7	6	5	4	3	2	1	0
CONTROL	CONTROL	CONTROL	CONTROL	ADDRESS	ADDRESS	ADDRESS	ADDRESS
TRACK	TRACK	TRACK	TRACK	TRACK	TRACK	TRACK	TRACK
INDEX	INDEX	INDEX	INDEX	INDEX	INDEX	INDEX	INDEX
MINUTE	MINUTE	MINUTE	MINUTE	MINUTE	MINUTE	MINUTE	MINUTE
SECOND	SECOND	SECOND	SECOND	SECOND	SECOND	SECOND	SECOND
FRAME	FRAME	FRAME	FRAME	FRAME	FRAME	FRAME	FRAME
ZERO	ZERO	ZERO	ZERO	ZERO	ZERO	ZERO	ZERO
ABS MINUTE	ABS MINUTE	ABS MINUTE	ABS MINUTE	ABS MINUTE	ABS MINUTE	ABS MINUTE	ABS MINUTE
ABS SECOND	ABS SECOND	ABS SECOND	ABS SECOND	ABS SECOND	ABS SECOND	ABS SECOND	ABS SECOND
ABS FRAME	ABS FRAME	ABS FRAME	ABS FRAME	ABS FRAME	ABS FRAME	ABS FRAME	ABS FRAME

Each byte is LSB first with respect to the 80 Q-subcode bits Q[79:0]. Thus bit 7 of address 14h is Q[0] while bit 0 of address 0Eh is Q[7]. Similarly bit 0 of address 1Dh corresponds to Q[79].

8.17 OMCK/RMCK Ratio (1Eh) (Read Only)

7	6	5	4	3	2	1	0
ORR7	ORR6	ORR5	ORR4	ORR3	ORR2	ORR1	ORR0

This register allows the calculation of the incoming sample rate by the host microcontroller from the equation ORR=Fso/Fsi. The Fso is determined by OMCK, whose frequency is assumed to be 256 Fso. ORR is represented as an unsigned 2-bit integer and a 6-bit fractional part. The value is meaningful only after the PLL has reached lock. For example, if the OMCK is 12.288MHz, Fso would be 48KHz (48KHz = 12.288MHz/256). Then if the input sample rate is also 48KHz, you would get 1.0 from the ORR register.(The value from the ORR register is hexadecimal, so the actual value you will get is 40h). If $F_{SO}/F_{SI} > 3\ ^{63}/_{64}$, ORR will saturate at the value FFh. Also, there is no hysteresis on ORR. Therefore a small amount of jitter on either clock can cause the LSB ORR[0] to oscillate.

ORR7:6 - Integer part of the ratio (Integer value=Integer(SRR[7:6]))

ORR5:0 - Fractional part of the ratio (Fraction value=Integer(SRR[5:0])/64)

8.18 C-bit or U-bit Data Buffer (20h - 37h)

Either channel status data buffer E or user data buffer E is accessible through these register addresses.

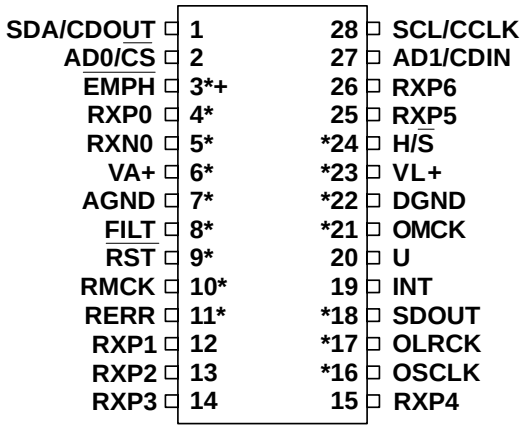
8.19 CS8415A I.D. and Version Register (7Fh) (Read Only)

7	6	5	4	3	2	1	0
ID3	ID2	ID1	ID0	VER3	VER2	VER1	VER0

ID3:0 - ID code for the CS8415A. Permanently set to 0100

VER3:0 - CS8415A revision level. Revision A is coded as 0001

9. PIN DESCRIPTION - SOFTWARE MODE



* Pins which remain the same function in all modes.
\+ Pins which require a pull up or pull down resistor to select the desired startup option.

SDA/CDOUT	1	**Serial Control Data I/O (I^2C) / Data Out (SPI)** (*Input/Output*) - In I^2C mode, SDA is the control I/O data line. SDA is open drain and requires an external pull-up resistor to VL+. In SPI mode, CDOUT is the output data from the control port interface on the CS8415A
AD0/$\overline{CS}$	2	**Address Bit 0 (I^2C) / Control Port Chip Select (SPI)** (*Input)* - A falling edge on this pin puts the CS8415A into SPI control port mode. With no falling edge, the CS8415A defaults to I^2C mode. In I^2C mode, AD0 is a chip address pin. In SPI mode, $\overline{CS}$ is used to enable the control port interface on the CS8415A
$\overline{EMPH}$	3	**Pre-Emphasis** (*Output*) - $\overline{EMPH}$ is low when the incoming Channel Status data indicates 50/15 μs pre-emphasis. $\overline{EMPH}$ is high when the Channel Status data indicates no pre-emphasis or indicates pre-emphasis other than 50/15 μs. This is also a start-up option pin, and requires a 47 kΩ resistor to either VL+ or DGND, which determines the AD2 address bit for the control port in I^2C mode
RXP0 **RXN0**	4 5	**AES3/SPDIF Receiver Port** (*Input*) - Differential line receiver inputs carrying AES3 data. RXP0 may be used as a single-ended input as part of 7:1 S/PDIF Input MUX. If RXP0 is used in MUX, RXN0 must be ac coupled to ground.
RXP1 **RXP2** **RXP3** **RXP4** **RXP5** **RXP6**	12 13 14 15 25 26	**Additional AES3/SPDIF Receiver Port** (*Input*) - Single-ended receiver inputs carrying AES3 or S/PDIF digital data. These inputs, along with RXP0, comprise the 7:1 S/PDIF Input Multiplexer and select line control is accessed using the MUX2:0 bits in the Control 2 register. Please note that any unused inputs should be tied to ground. See Appendix A for recommended input circuits.
VA+	6	**Positive Analog Power** (*Input*) - Positive supply for the analog section. Nominally +5 V. This supply should be as quiet as possible since noise on this pin will directly affect the jitter performance of the recovered clock
AGND	7	**Analog Ground** (*Input*) - Ground for the analog circuitry in the chip. AGND and DGND should be connected to a common ground area under the chip.
FILT	8	**PLL Loop Filter** (*Output*) - An RC network should be connected between this pin and ground. See “Appendix C: PLL Filter” on page 40 for recommended schematic and component values.
$\overline{RST}$	9	**Reset** (*Input*) - When $\overline{RST}$ is low, the CS8415A enters a low power mode and all internal states are reset. On initial power up, $\overline{RST}$ must be held low until the power supply is stable, and all input clocks are stable in frequency and phase. This is particularly true in hardware mode with multiple CS8415A devices where synchronization between devices is important

RMCK	10	**Input Section Recovered Master Clock** (*Output*) - Input section recovered master clock output when PLL is used. Frequency defaults to 256x the sample rate (Fs) and may be set to 128x.
RERR	11	**Receiver Error** (*Output*) - When high, indicates a problem with the operation of the AES3 receiver. The status of this pin is updated once per sub-frame of incoming AES3 data. Conditions that can cause RERR to go high are: validity, parity error, bi-phase coding error, confidence, QCRC and CCRC errors, as well as loss of lock in the PLL. Each condition may be optionally masked from affecting the RERR pin using the Receiver Error Mask Register. The RERR pin tracks the status of the unmasked errors: the pin goes high as soon as an unmasked error occurs and goes low immediately when all unmasked errors go away
OSCLK	16	**Serial Audio Output Bit Clock** (*Input*/*Output*) - Serial bit clock for audio data on the SDOUT pin
OLRCK	17	**Serial Audio Output Left/Right Clock** (*Input*/*Output*) - Word rate clock for the audio data on the SDOUT pin. Frequency will be the output sample rate (Fs)
SDOUT	18	**Serial Audio Output Data** (*Output*) - Audio data serial output pin
INT	19	**Interrupt** (*Output*) - Indicates errors and key events during the operation of the CS8415A. All bits affecting INT may be unmasked through bits in the control registers. The condition(s) that initiated interrupt are readable through a control register. The polarity of the INT output, as well as selection of a standard or open drain output, is set through a control register. Once set true, the INT pin goes false only after the interrupt status registers have been read and the interrupt status bits have returned to zero
U	20	**User Data** (*Output*) - Outputs User data from the AES3 receiver, see Figure 7 for timing information
OMCK	21	**System Clock** (*Input*) - When the OMCK System Clock Mode is enabled using the SWCLK bit in the Control 1 register, the clock signal input on this pin is output through RMCK. OMCK serves as reference signal for OMCK/RMCK ratio expressed in register 0x1E
DGND	22	**Digital Ground** (*Input*) - Ground for the digital circuitry in the chip. DGND and AGND should be connected to a common ground area under the chip.
VL+	23	**Positive Digital Power** (*Input*) - Positive supply for the digital section. Typically +3 to +5 V.
H/$\overline{S}$	24	**Hardware/Software Mode Control** (*Input*) - Determines the method of controlling the operation of the CS8415A, and the method of accessing CS and U data. In software mode, device control and CS and U data access is primarily through the control port, using a microcontroller. Hardware mode provides an alternate mode of operation and access to the CS and U data through dedicated pins. This pin should be permanently tied to VL+ or DGND
AD1/CDIN	27	**Address Bit 1 (I^2C) / Serial Control Data in (SPI)** (*Input*) - In I^2C mode, AD1 is a chip address pin. In SPI mode, CDIN is the input data line for the control port interface
SCL/CCLK	28	**Control Port Clock** (*Input*) - Serial control interface clock and is used to clock control data bits into and out of the CS8415A. In I^2C mode, SCL requires an external pull-up resistor to VL+

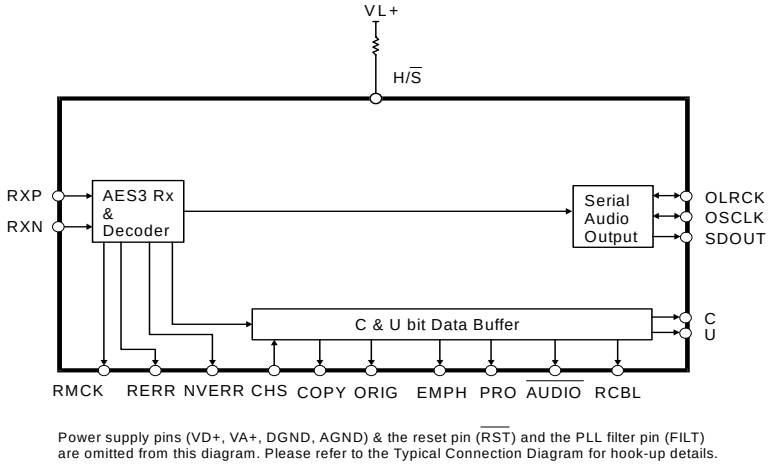
10. HARDWARE MODE

The CS8415A has a hardware mode which allows using the device without a microcontroller. Hardware mode is selected by connecting the H/$\overline{S}$ pin to VL+. Various pins change function in hardware mode, described in the hardware mode pin definition section.

Hardware mode data flow is shown in Figure 9. Audio data is input through the AES3 receiver, and routed to the serial audio output port. The PRO, COPY, ORIG, $\overline{EMPH}$, and $\overline{AUDIO}$ channel status bits are output on pins. The decoded C and U bits are also output, clocked at both edges of OLRCK (master mode only, see Figure 7).The current audio sample is passed unmodified to the serial audio output port if the validity bit is high, or a parity, biphase, or PLL lock error occurs.

10.1 Serial Audio Port Formats

In hardware mode, only a limited number of alternative serial audio port formats are available. Table 2 defines the equivalent software mode bit settings for each format. Start-up options are shown in Table 3, and allow choice of the serial audio output port as a master or slave, and the serial audio port format.



Power supply pins (VD+, VA+, DGND, AGND) & the reset pin ($\overline{RST}$) and the PLL filter pin (FILT) are omitted from this diagram. Please refer to the Typical Connection Diagram for hook-up details.

Figure 9. Hardware Mode

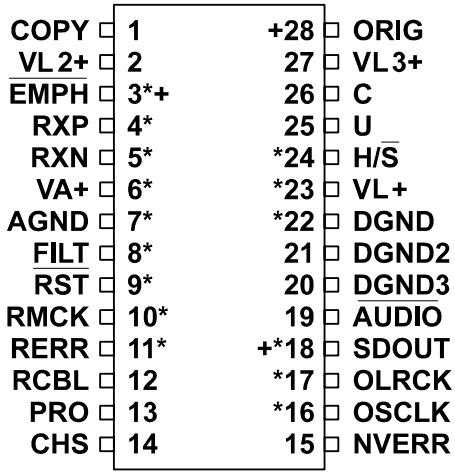
	SOSF	SORES1/0	SOJUST	SODEL	SOSPOL	SOLRPOL
OF1 - Left Justified	0	00	0	0	0	0
OF2 - I^2S 24-bit data	0	00	0	1	0	1
OF3 - Right Justified, master mode only	0	00	1	0	0	0
OF4 - Direct AES3 data	0	11	0	0	0	0

Table 2. Equivalent Software Mode Bit Definitions

SDOUT	ORIG	$\overline{EMPH}$	Function
LO	-	-	Serial Output Port is Slave
HI	-	-	Serial Output Port is Master
-	LO	LO	Left Justified
-	LO	HI	I2S 24-bit data
-	HI	LO	Right Justified
-	HI	HI	Direct AES3 data

Table 3. Hardware Mode Start-up Options

11. PIN DESCRIPTION - HARDWARE MODE



* Pins which remain the same function in all modes.
\+ Pins which require a pull up or pull down resistor to select the desired startup option.

COPY	1	**COPY Channel Status Bit** (*Output*) - Reflects the state of the Copyright Channel Status bit in the incoming AES3 data stream. If the category code is set to General, copyright will be indicated whatever the state of the Copyright bit.
VL2+ **VL+** **VL3+**	2 23 27	**Positive Digital Power** (*Input*) - Typically +3 to +5V.
$\overline{\text{EMPH}}$	3	**Pre-Emphasis** (*Output*) - $\overline{\text{EMPH}}$ is low when the incoming Channel Status data indicates 50/15 μs pre-emphasis. $\overline{\text{EMPH}}$ is high when the Channel Status data indicates no pre-emphasis or indicates pre-emphasis other than 50/15 μs. This pin is also a start-up option which, along with ORIG, determines the serial port format. A 47 kΩ resistor to either VL+ or DGND is required.
RXP0 **RXN0**	4 5	**AES3/SPDIF Receiver Port** (*Input*) - Differential line receiver inputs for the AES3 biphase encoded data. See Appendix A for recommended circuits.
VA+	6	**Positive Analog Power** (*Input*) - Nominally +5 V. This supply should be as quiet as possible since noise on this pin will directly affect the jitter performance of the recovered clock.
AGND	7	**Analog Ground** (*Input*) - Ground for the analog circuitry in the chip. AGND and DGND should be connected to a common ground area under the chip.
FILT	8	**PLL Loop Filter** (*Output*) - An RC network should be connected between this pin and ground. See “Appendix C: PLL Filter” on page 40 for recommended schematic and component values.
$\overline{\text{RST}}$	9	**Reset** (*Input*) - When $\overline{\text{RST}}$ is low, the CS8415A enters a low power mode and all internal states are reset. On initial power up, $\overline{\text{RST}}$ must be held low until the power supply is stable, and all input clocks are stable in frequency and phase. This is particularly true in hardware mode with multiple CS8415A devices where synchronization between devices is important.
RMCK	10	**Recovered Master Clock** (*Output*) - Recovered master clock output when PLL is locked to the incoming AES3 stream. Frequency is 256x the sample rate (Fs).
RERR	11	**Receiver Error** (*Output*) - When high, indicates an error condition in the AES3 receiver. The status of this pin is updated once per sub-frame of incoming AES3 data. Conditions that can cause RERR to go high are: validity bit high, parity error, bi-phase coding error, and loss of lock by the PLL.

RCBL	12	**Receiver Channel Status Block** (*Output*) -Indicates the beginning of a received channel status block. RCBL goes high two frames after the reception of a Z preamble, remains high for 16 frames while COPY, ORIG, AUDIO, $\overline{\text{EMPH}}$ and PRO are updated, and returns low for the remainder of the block. RCBL changes on rising edges of RMCK.
PRO	13	**PRO Channel Status Bit** (*Output*) - Reflects the state of the Professional/Consumer Channel Status bit in the incoming AES3 data stream. Low indicates Consumer and high indicates Professional.
CHS	14	**Channel Select (Input) -** Selects which sub-frame's channel status data is output on the $\overline{\text{EMPH}}$, COPY, ORIG, PRO and $\overline{\text{AUDIO}}$ pins. Channel A is selected when CHS is low, channel B is selected when CHS is high.
NVERR	15	**No Validity Receiver Error Indicator** (*Output*) - A high output indicates a problem with the operation of the AES3 receiver. The status of this pin is updated once per frame of incoming AES3 data. Conditions that cause NVERR to go high are: parity error, and bi-phase coding error, and loss of lock by the PLL.
OSCLK	16	**Serial Audio Output Bit Clock** (*Input*/*Output*) - Serial bit clock for audio data on the SDOUT pin.
OLRCK	17	**Serial Audio Output Left/Right Clock** (*Input*/*Output*) - Word rate clock for the audio data on the SDOUT pin. Frequency will be the output sample rate (Fs).
SDOUT	18	**Serial Audio Output Data** (*Output*) - Audio data serial output pin. This pin is also a start-up option which determines if the serial audio port is master or slave. A 47 kΩ resistor to either VL+ or DGND is required.
$\overline{\text{AUDIO}}$	19	**Audio Channel Status Bit** (*Output*) - Reflects the state of the audio/non audio Channel Status bit in the incoming AES3 data stream. When this bit is low a valid audio stream is indicated.
DGND3 **DGND2** **DGND**	20 21 22	**Digital Ground** (*Input*) - Ground for the digital circuitry in the chip. DGND and AGND should be connected to a common ground area under the chip.
H/$\overline{\text{S}}$	24	**Hardware/Software Mode Control** (*Input*) - Determines the method of controlling the operation of the CS8415A, and the method of accessing CS and U data. In software mode, device control and CS and U data access is primarily through the control port, using a microcontroller. Hardware mode provides an alternate mode of operation and access to the CS and U data through dedicated pins. This pin should be permanently tied to VL+ or DGND.
U	25	**User Data** (*Output*) - Outputs user data from the AES3 receiver, clocked by the rising and falling edges of OLRCK.
C	26	**Channel Status Data** (*Output*) - Outputs channel status data from the AES3 receiver, clocked by the rising and falling edges of OLRCK.
ORIG	28	**Original Channel Status** (*Output*) - SCMS generation indicator. This is decoded from the incoming category code and the L bit in the Channel Status bits. A low output indicates that the source of the audio data stream is a copy not an original. A high indicates that the audio data stream is original. This pin is also a start-up option which, along with $\overline{\text{EMPH}}$, determines the serial audio port format. A 47 kΩ resistor to either VL+ or DGND is required.

12. APPLICATIONS

12.1 Reset, Power Down and Start-up

When $\overline{\text{RST}}$ is low, the CS8415A enters a low power mode and all internal states are reset, including the control port and registers, and the outputs are muted. When $\overline{\text{RST}}$ is high, the control port becomes operational and the desired settings should be loaded into the control registers. Writing a 1 to the RUN bit will then cause the part to leave the low power state and begin operation. After the PLL has settled, the serial audio outputs will be enabled.

Some options within the CS8415A are controlled by a start-up mechanism. During the reset state, some of the output pins are reconfigured internally to be inputs. Immediately upon exiting the reset state, the level of these pins is sensed. The pins are then switched to be outputs. This mechanism allows output pins to be used to set alternative modes in the CS8415A by connecting a 47 KΩ resistor to between the pin and either VL+ (HI) or DGND (LO). For each mode, every start-up option select pin MUST have an external pull-up or pull-down resistor. In software mode, the only start-up option pin is $\overline{\text{EMPH}}$, which is used to set a chip address bit for the control port in I^2C mode. The hardware mode uses many start-up options, which are detailed in the hardware definition section at the end of this data sheet.

12.2 ID Code and Revision Code

The CS8415A has a register that contains a 4-bit code to indicate that the addressed device is a CS8415A. This is useful when other CS84XX family members are resident in the same system, allowing common software modules.

The CS8415A 4-bit revision code is also available. This allows the software driver for the CS8415A to identify which revision of the device is in a particular system, and modify its behavior accordingly. To allow for future revisions, it is strongly recommend that the revision code is read into a variable area within the microcontroller, and used wherever appropriate as revision details become known.

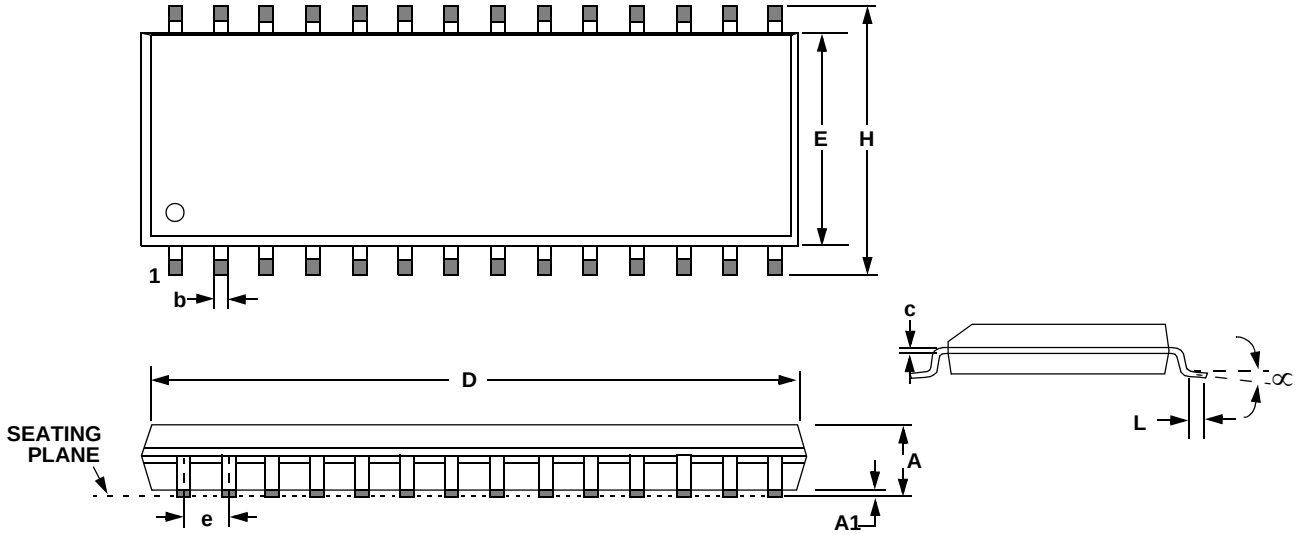
12.3 Power Supply, Grounding, and PCB layout

For most applications, the CS8415A can be operated from a single +5 V supply, following normal supply decoupling practices, see Figure 5. Note that the I^2C protocol is supported only in 5V mode. For applications where the recovered input clock, output on the RMCK pin, is required to be low jitter, then use a separate, quiet, analog +5 V supply for VA+, decoupled to AGND. In addition, a separate region of analog ground plane around the FILT, AGND, VA+, RXP0-6 and RXN0 pins is recommended.

Extensive use of power and ground planes, ground plane fill in unused areas and surface mount decoupling capacitors are recommended. Decoupling capacitors should be mounted on the same side of the board as the CS8415A to minimize inductance effects, and all decoupling capacitors should be as close to the CS8415A as possible.

13. PACKAGE DIMENSIONS

28L SOIC (300 MIL BODY) PACKAGE DRAWING

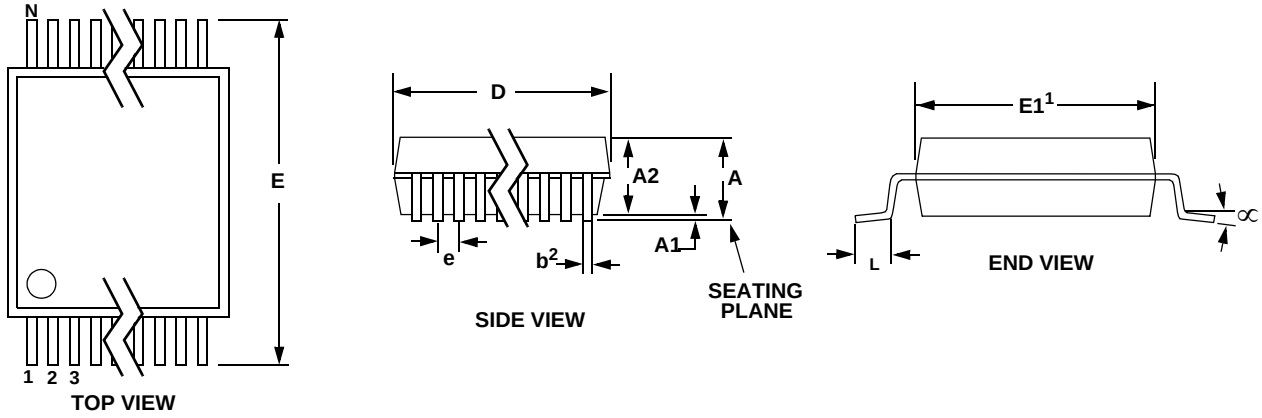


	INCHES			MILLIMETERS		
DIM	**MIN**	**NOM**	**MAX**	**MIN**	**NOM**	**MAX**
A	0.093	0.098	0.104	2.35	2.50	2.65
A1	0.004	0.008	0.012	0.10	0.20	0.30
b	0.013	0.017	0.020	0.33	0.42	0.51
C	0.009	0.011	0.013	0.23	0.28	0.32
D	0.697	0.705	0.713	17.70	17.90	18.10
E	0.291	0.295	0.299	7.40	7.50	7.60
e	0.040	0.050	0.060	1.02	1.27	1.52
H	0.394	0.407	0.419	10.00	10.34	10.65
L	0.016	0.026	0.050	0.40	0.65	1.27
∝	0°	4°	8°	0°	4°	8°

JEDEC #: MS-013

Controlling Dimension is Millimeters

28L TSSOP (4.4 mm BODY) PACKAGE DRAWING



	INCHES			MILLIMETERS			NOTE
DIM	**MIN**	**NOM**	**MAX**	**MIN**	**NOM**	**MAX**	
A	--	--	0.47	--	--	1.20	
A1	0.002	0.004	0.006	0.05	0.10	0.15	
A2	0.03150	0.035	0.04	0.80	0.90	1.00	
b	0.00748	0.0096	0.012	0.19	0.245	0.30	2,3
D	0.378 BSC	0.382 BSC	0.386 BSC	9.60 BSC	9.70 BSC	9.80 BSC	1
E	0.248	0.2519	0.256	6.30	6.40	6.50	
E1	0.169	0.1732	0.177	4.30	4.40	4.50	1
e	--	0.026 BSC	--	--	0.65 BSC	--	
L	0.020	0.024	0.029	0.50	0.60	0.75	
∝	0°	4°	8°	0°	4°	8°	

JEDEC #: MO-153
Controlling Dimension is Millimeters.

Notes: 1. “D” and “E1” are reference datums and do not included mold flash or protrusions, but do include mold mismatch and are measured at the parting line, mold flash or protrusions shall not exceed 0.20 mm per side.

2. Dimension “b” does not include dambar protrusion/intrusion. Allowable dambar protrusion shall be 0.13 mm total in excess of “b” dimension at maximum material condition. Dambar intrusion shall not reduce dimension “b” by more than 0.07 mm at least material condition.

3. These dimensions apply to the flat section of the lead between 0.10 and 0.25 mm from lead tips.

14. APPENDIX A: EXTERNAL AES3/SPDIF/IEC60958 RECEIVER COMPONENTS

14.1 AES3 Receiver External Components

The CS8415A AES3 receiver is designed to accept both the professional and consumer interfaces. The digital audio specifications for professional use call for a balanced receiver, using XLR connectors, with 110 Ω ±20% impedance. The XLR connector on the receiver should have female pins with a male shell. Since the receiver has a very high input impedance, a 110 Ω resistor should be placed across the receiver terminals to match the line impedance, as shown in Figure 10. Although transformers are not required by the AES, they are, however, strongly recommended.

If some isolation is desired without the use of transformers, a 0.01 μF capacitor should be placed in series with each input pin (RXP0 and RXN0) as shown in Figure 11. However, if a transformer is not used, high frequency energy could be coupled into the receiver, causing degradation in analog performance.

Figures 10 and 11 show an optional DC blocking capacitor (0.1 μF to 0.47 μF) in series with the cable input. This improves the robustness of the receiver, preventing the saturation of the transformer, or any DC current flow, if a DC voltage is present on the cable.

In the configuration of systems, it is important to avoid ground loops and DC current flowing down the shield of the cable that could result when boxes with different ground potentials are connected. Generally, it is good practice to ground the shield to the chassis of the transmitting unit, and connect the shield through a capacitor to chassis ground at the receiver. However, in some cases it is advantageous to have the ground of two boxes held to the same potential, and the cable shield might be depended upon to make that electrical connection. Generally, it may be a good idea to provide the option of grounding or capacitively coupling the shield to the chassis.

In the case of the consumer interface, the standards call for an unbalanced circuit having a receiver impedance of 75 Ω ±5%. The connector for the consumer interface is an RCA phono socket. The receiver circuit for the consumer interface is shown in Figure 12. Figure 13 shows an implementation of the Input S/PDIF Multiplexer using the consumer interface.

The circuit shown in Figure 14 may be used when external RS422 receivers, optical receivers or other TTL/CMOS logic outputs drive the CS8415A receiver section.

14.2 Isolating Transformer Requirements

Please refer to the application note AN134: *AES and SPDIF Recommended Transformers* for resources on transformer selection.

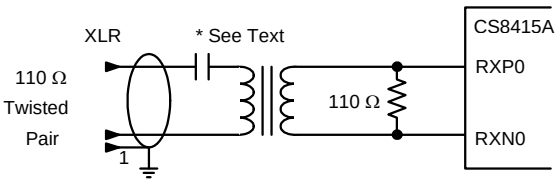


Figure 10. Professional Input Circuit

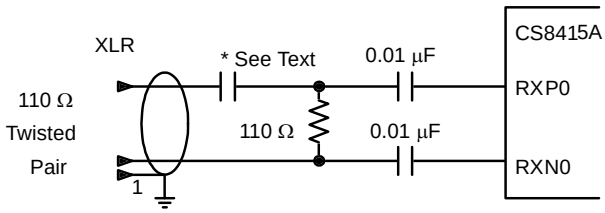


Figure 11. Transformerless Professional Input Circuit

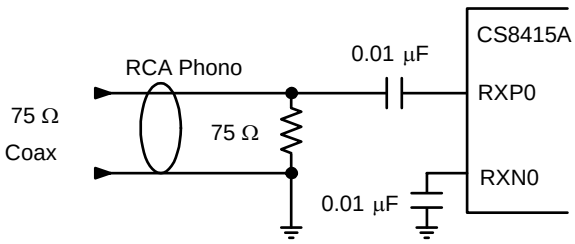


Figure 12. Consumer Input Circuit

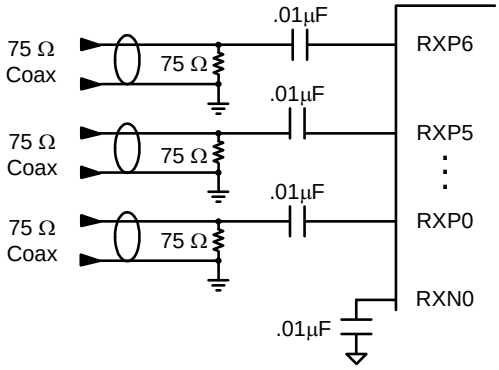


Figure 13. S/PDIF MUX Input Circuit

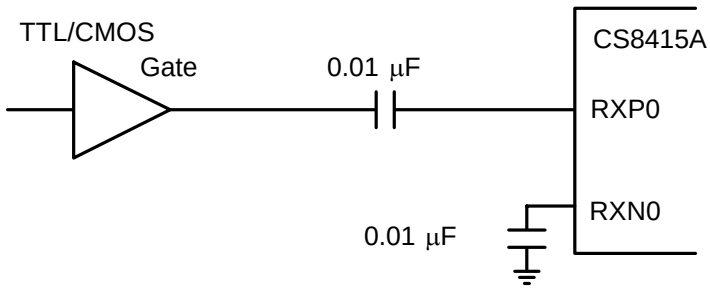


Figure 14. TTL/CMOS Input Circuit

15. APPENDIX B: CHANNEL STATUS AND USER DATA BUFFER MANAGEMENT

15.1 AES3 Channel Status (C) Bit Management

The CS8415A contains sufficient RAM to store a full block of C data for both A and B channels (192 x 2 = 384 bits), and also 384 bits of U information. The user may read from these buffer RAMs through the control port.

The buffering scheme involves 2 block-sized buffers, named D and E, as shown in Figure 15. The MSB of each byte represents the first bit in the serial C data stream. For example, the MSB of byte 0 (which is at control port address 20h) is the consumer/professional bit for channel status block A.

The first buffer (D) accepts incoming C data from the AES receiver. The 2nd buffer (E) accepts entire blocks of data from the D buffer. The E buffer is also accessible from the control port, allowing reading of the C data.

15.2 Accessing the E buffer

The user can monitor the incoming data by reading the E buffer, which is mapped into the register space of the CS8415A, through the control port.

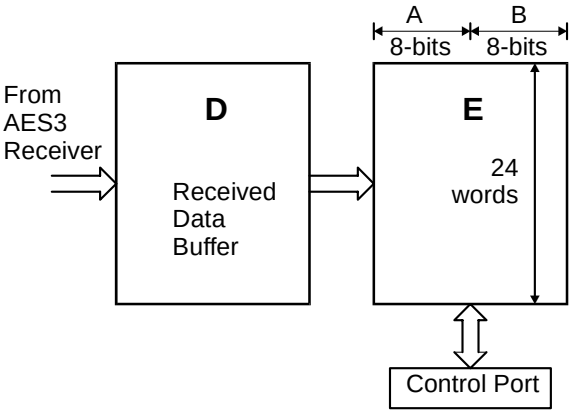


Figure 15. Channel Status Data Buffer Structure

The user can configure the interrupt enable register to cause interrupts to occur whenever D to E buffer transfers occur. This allows determination of the allowable time periods to interact with the E buffer.

Also provided is a D to E inhibit bit. This may be used whenever “long” control port interactions are occurring.

A flowchart for reading the E buffer is shown in Figure 16. Since a D to E interrupt just occurred after reading, there is a substantial time interval until the next D to E transfer (approximately 24 frames worth of time). This is usually plenty of time to access the E data without having to inhibit the next transfer.

15.2.1 Reserving the first 5 bytes in the E buffer

D to E buffer transfers periodically overwrite the data stored in the E buffer. The CS8415A has the capability of reserving the first 5 bytes of the E buffer for user writes only. When this capability is in use, internal D to E buffer transfers will NOT affect the first 5 bytes of the E buffer. Therefore, the user can set values in these first 5 E bytes once, and the settings will persist until the next user change. This mode is enabled using the Channel Status Data Buffer Control register.

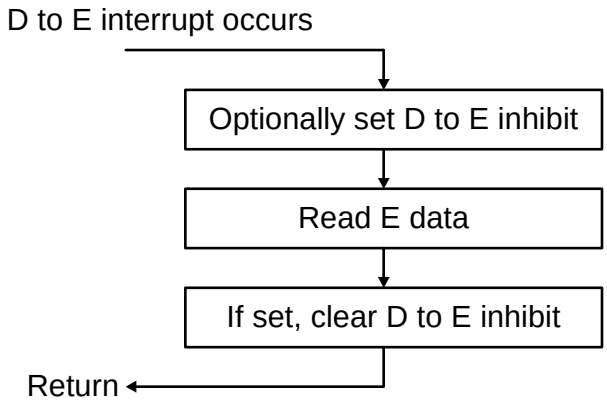


Figure 16. Flowchart for Reading the E Buffer

15.2.2 Serial Copy Management System (SCMS)

In software mode, the CS8415A allows read access to all the channel status bits. For consumer mode SCMS compliance, the host microcontroller needs to read and interpret the Category Code, Copy bit and L bit appropriately.

In hardware mode, the SCMS protocol can be followed by either using the COPY and ORIG output pins, or by using the C bit serial output pin. These options are documented in the hardware mode section of this data sheet.

15.2.3 Channel Status Data E Buffer Access

The E buffer is organized as 24 x 16-bit words. For each word the MS Byte is the A channel data, and the LS Byte is the B channel data (see Figure 15).

There are two methods of accessing this memory, known as one byte mode and two byte mode. The desired mode is selected by setting a control register bit.

15.2.3.1 One Byte mode

In many applications, the channel status blocks for the A and B channels will be identical. In this situation, if the user reads a byte from one of the channel's blocks, the corresponding byte for the other channel will be the same. One byte mode takes advantage of the often identical nature of A and B channel status data. When reading data in one byte mode, a single byte is returned, which can be from channel A or B data, depending on a register control bit.

One byte mode saves the user substantial control port access time, as it effectively accesses 2 bytes worth of information in 1 byte's worth of access time. If the control port's autoincrement addressing is used in combination with this mode, multi-byte accesses such as full-block reads can be done especially efficiently.

15.2.3.2 Two Byte mode

There are those applications in which the A and B channel status blocks will not be the same, and the user is interested in accessing both blocks. In these situations, two byte mode should be used to access the E buffer.

In this mode, a read will cause the CS8415A to output two bytes from its control port. The first byte out will represent the A channel status data, and the 2nd byte will represent the B channel status data.

15.3 AES3 User (U) Bit Management

Entire blocks of U data are buffered using a cascade of 2 block-sized RAMs to perform the buffering. The user has access to the second of these buffers, denoted the E buffer, through the control port. The U buffer access only operates in two byte mode, since there is no concept of A and B blocks for user data. The arrangement of the data is as followings:
Bit15[A7]Bit14[B7]Bit13[A6]Bit12[B6]...Bit1[A0]Bit 0[B0]. The arrangement of the data in the each byte is that the MSB is the first received bit and is the first transmitted bit. The first byte read is the first byte received, and the first byte sent is the first byte transmitted. If you read two bytes from the E buffer, you will get the following arrangement: A[7]B[7]A[6]B[6]....A[0]B[0].

16. APPENDIX C: PLL FILTER

16.1 General

An on-chip Phase Locked Loop (PLL) is used to recover the clock from the incoming data stream. Figure 17 is a simplified diagram of the PLL in these parts. When the PLL is locked to an AES3 input stream, it is updated at each preamble in the AES3 stream. This occurs at twice the sampling frequency, F_S. When the PLL is locked to ILRCK, it is updated at F_S so that the duty cycle of the input doesn't affect jitter.

There are some applications where low jitter in the recovered clock, presented on the RMCK pin, is important. For this reason, the PLL has been designed to have good jitter attenuation characteristics, as shown in Figure 19. In addition, the PLL has been designed to only use the preambles of the AES3 stream to provide lock update information to the PLL. This results in the PLL being immune to data dependent jitter affects because the AES3 preambles do not vary with the data.

The PLL has the ability to lock onto a wide range of input sample rates with no external component changes. If the sample rate of the input subsequently changes, for example in a varispeed application, the PLL will only track up to ±12.5% from the nominal center sample rate. The nominal center sample rate is the sample rate that the PLL first locks onto upon application of an AES3 data stream or after enabling the CS8415A clocks by setting the RUN control bit. If the 12.5% sample rate limit is exceeded, the PLL will return to its wide lock range mode and re-acquire a new nominal center sample rate.

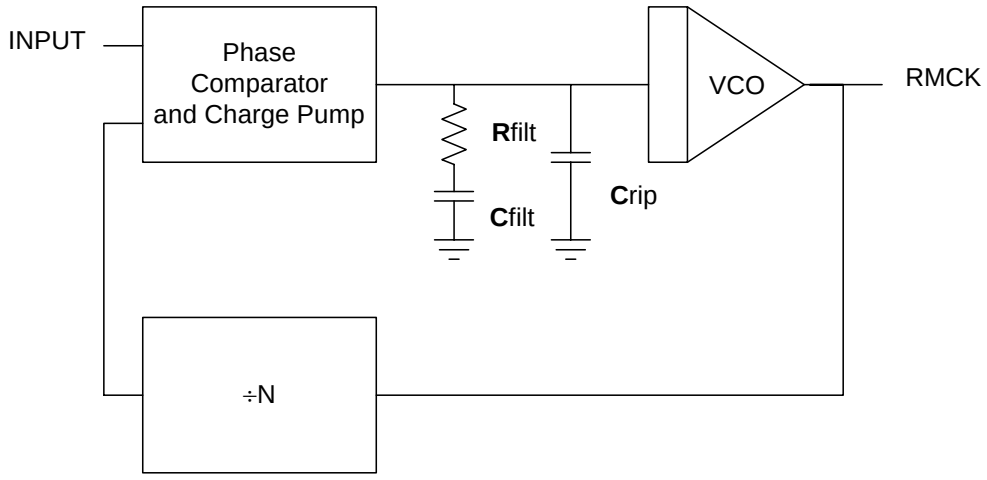


Figure 17. PLL Block Diagram

16.2 External Filter Components

16.2.1 General

The PLL behavior is affected by the external filter component values. Figure 5 shows the recommended configuration of the two capacitors and one resistor that comprise the PLL filter. In Table 4, the component values shown for the 32 to 96kHz range have the highest corner frequency jitter attenuation curve, takes the shortest time to lock, and offers the best output jitter performance. The component values shown for the 8 to 96kHz range allows the lowest input sample rate to be 8 kHz, and increases the lock time of the PLL. Lock times are worst case for an Fsi transition of 96 kHz.

16.2.2 Capacitor Selection

The type of capacitors used for the PLL filter can have a significant effect on receiver performance. Large or exotic film capacitors are not necessary as their leads and the required longer circuit board traces add undesirable inductance to the circuit. Surface mount ceramic capacitors are a good choice because their own inductance is low, and they can be mounted close to the FILT pin to minimize trace inductance. For C_{RIP}, a C0G or NPO dielectric is recommended, and for C_{FILT}, an X7R dielectric is preferred. Avoid capacitors with large temperature co-efficients, or capacitors with high dielectric constants, that are sensitive to shock and vibration. These include the Z5U and Y5V dielectrics.

16.2.3 Circuit Board Layout

Board layout and capacitor choice affect each other and determine the performance of the PLL. Figure 18 contains a suggested layout for the PLL filter components and for bypassing the analog supply voltage. The 0.1 µF bypass capacitor is in a 1206 form factor. R_{FILT} and the other three capacitors are in an 0805 form factor. The traces are on the top surface of the board with the IC so that there is no via inductance. The traces themselves are short to minimize the inductance in the filter path. The VA+ and AGND traces extend back to their origin and are shown only in truncated form in the drawing.

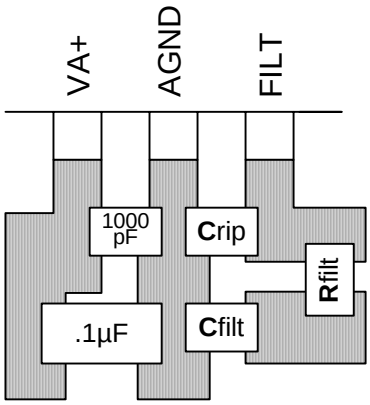


Figure 18. Recommended Layout Example

16.2.4 Component Value Selection

The external PLL component values are listed in Table 4.

Fs Range	R_{FILT} (kΩ)	C_{FILT} (μF)	C_{RIP} (nF)	PLL Lock Time (ms)
8 - 96	0.909	1.8	33	56
32 - 96	3.0	0.047	2.2	60

Table 4. External PLL Component Values

16.2.5 Jitter Attenuation

Shown in Figure 19 is the jitter attenuation plot when used with the 36-96kHz PLL component values. The AES3 and IEC60958-4 specifications state a maximum of 2dB jitter gain or peaking.

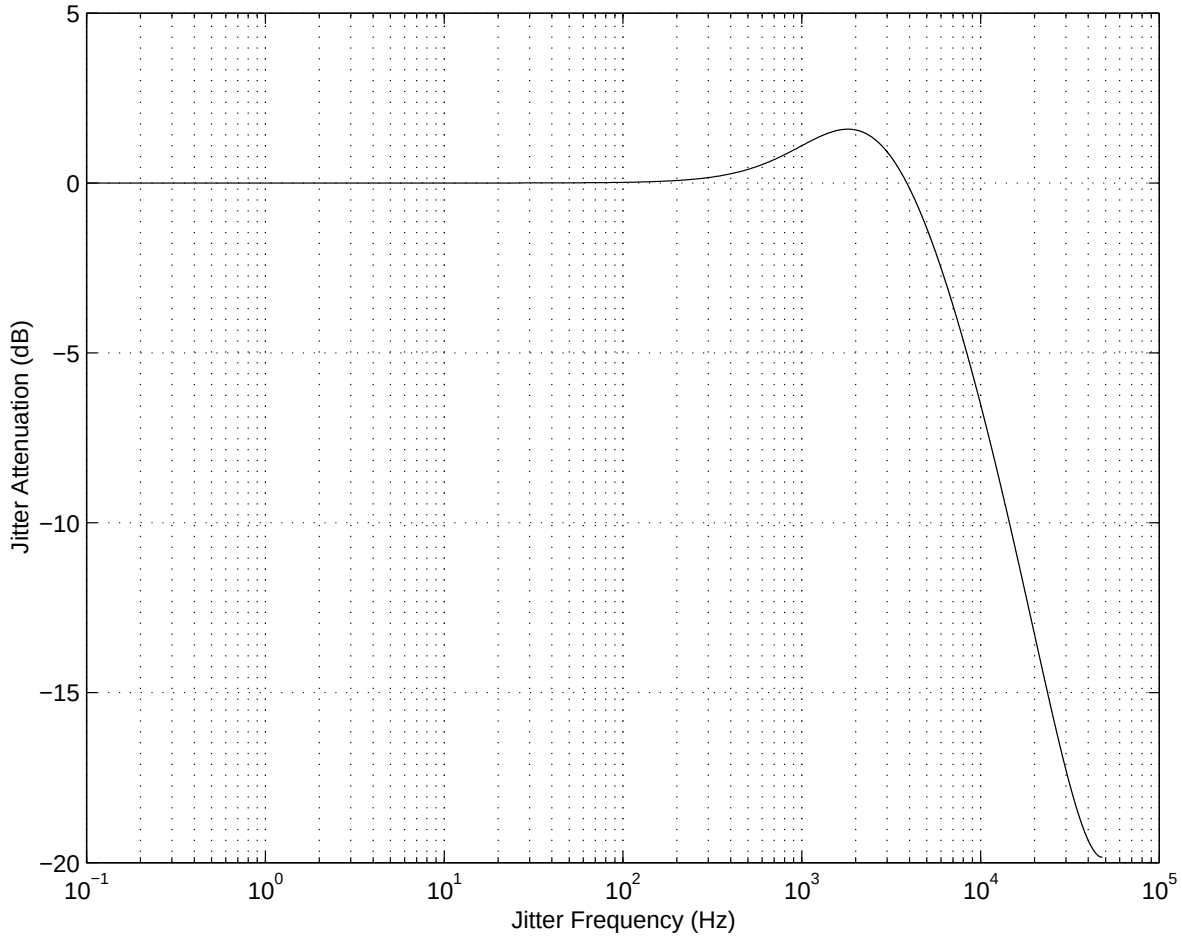


Figure 19. Jitter Attenuation Characteristics of PLL with Fs=32 to 96 kHz Filter Components