# C8051F350/1/2/3

## 8 k ISP Flash MCU Family

### Analog Peripherals
- **24 or 16-Bit ADC**
  - No missing codes
  - 0.0015% nonlinearity
  - Programmable conversion rates up to 1 ksps
  - 8-Input multiplexer
  - 1x to 128x PGA
  - Built-in temperature sensor
- **Two 8-Bit Current Output DACs**
- **Comparator**
  - Programmable hysteresis and response time
  - Configurable as interrupt or reset source
  - Low current (0.4 µA)

### On-chip Debug
- On-chip debug circuitry facilitates full speed, non-intrusive in-system debug (No emulator required)
- Provides breakpoints, single stepping, inspect/modify memory and registers
- Superior performance to emulation systems using ICE-Chips, target pods, and sockets
- Low Cost, **Complete** Development Kit

### Supply Voltage 2.7 to 3.6 V
- Typical operating current: 5.8 mA @ 25 MHz; 11 µA @ 32 kHz
- Typical stop mode current: 0.1 µA

### Temperature Range: –40 to +85 °C

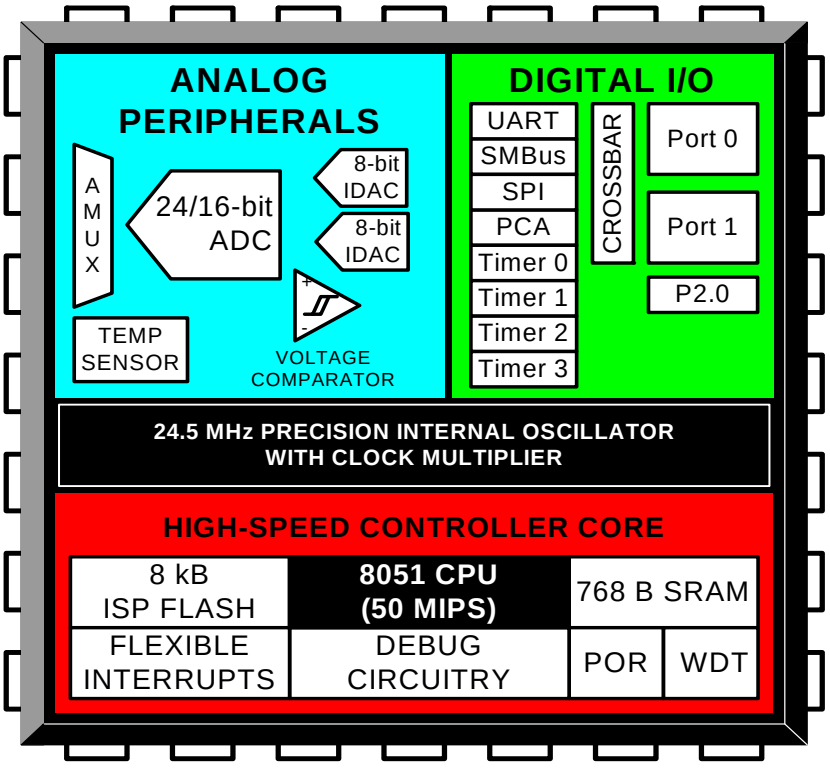
### High Speed 8051 µC Core
- Pipelined Instruction architecture; executes 70% of instructions in 1 or 2 system clocks
- Up to 50 MIPS throughput
- Expanded interrupt handler

### Memory
- 768 Bytes (256 + 512) On-Chip RAM
- 8 kB Flash; In-system programmable in 512-byte Sectors

### Digital Peripherals
- 17 Port I/O; All 5 V tolerant with high sink current
- Enhanced UART, SMBus™, and SPI™ Serial Ports
- Four general purpose 16-bit counter/timers
- 16-bit programmable counter array (PCA) with three capture/compare modules
- Real time clock mode using PCA or timer and external clock source

### Clock Sources
- Internal Oscillator: 24.5 MHz with ± 2% accuracy supports UART operation
- External Oscillator: Crystal, RC, C, or clock (1 or 2 pin modes)
- Clock multiplier to achieve 50 MHz internal clock
- Can switch between clock sources on-the-fly

### 28-Pin QFN or 32-Pin LQFP Package
- 5 x 5 mm PCB footprint with 28-QFN





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C2 Register Definition 25.3. REVID: C2 Revision ID . . . . . 230
C2 Register Definition 25.4. FPCTL: C2 Flash Programming Control . . . . . 230
C2 Register Definition 25.5. FPDAT: C2 Flash Programming Data . . . . . 230

## NOTES:

# 1. System Overview

C8051F350/1/2/3 devices are fully integrated mixed-signal System-on-a-Chip MCUs. Highlighted features are listed below. Refer to Table 1.1 for specific product feature selection.

- High-speed pipelined 8051-compatible microcontroller core (up to 50 MIPS)
- In-system, full-speed, non-intrusive debug interface (on-chip)
- 24 or 16-bit single-ended/differential ADC with analog multiplexer
- Two 8-bit Current Output DACs
- Precision programmable 24.5 MHz internal oscillator
- 8 kB of on-chip Flash memory
- 768 bytes of on-chip RAM
- SMBus/I2C, Enhanced UART, and SPI serial interfaces implemented in hardware
- Four general-purpose 16-bit timers
- Programmable counter/timer array (PCA) with three capture/compare modules and watchdog timer function
- On-chip power-on reset, $V_{DD}$ monitor, and temperature sensor
- On-chip voltage comparator
- 17 Port I/O (5 V tolerant)

With on-chip power-on reset, $V_{DD}$ monitor, watchdog timer, and clock oscillator, the C8051F350/1/2/3 devices are truly stand-alone System-on-a-Chip solutions. The Flash memory can be reprogrammed even in-circuit, providing non-volatile data storage, and also allowing field upgrades of the 8051 firmware. User software has complete control of all peripherals, and may individually shut down any or all peripherals for power savings.

The on-chip Silicon Labs 2-Wire (C2) Development Interface allows non-intrusive (uses no on-chip resources), full speed, in-circuit debugging using the production MCU installed in the final application. This debug logic supports inspection and modification of memory and registers, setting breakpoints, single stepping, run and halt commands. All analog and digital peripherals are fully functional while debugging using C2. The two C2 interface pins can be shared with user functions, allowing in-system debugging without occupying package pins.

Each device is specified for 2.7 to 3.6 V operation over the industrial temperature range (–45 to +85 °C). The Port I/O and /RST pins are tolerant of input signals up to 5 V. The C8051F350/1/2/3 are available in 28-pin QFN (also referred to as MLP or MLF) or 32-pin LQFP packaging, as shown in Figure 1.1 through Figure 1.4.

**Table 1.1. Product Selection Guide**

| Ordering Part Number | MIPS (Peak) | Flash Memory | RAM | Calibrated Internal 24.5 MHz Oscillator | Clock Multiplier | SMBus/I2C | SPI | UART | Timers (16-bit) | Programmable Counter Array | Digital Port I/Os | 24-bit ADC | 16-bit ADC | Two 8-bit Current Output DACs | Internal Voltage Reference | Temperature Sensor | Analog Comparator | Lead-free (RoHS Compliant) | Package |
|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|
| C8051F350-GQ | 50 | 8 kB | 768 | ✓ | ✓ | ✓ | ✓ | ✓ | 4 | ✓ | 17 | ✓ | — | ✓ | ✓ | ✓ | ✓ | ✓ | LQFP-32 |
| C8051F351-GM | 50 | 8 kB | 768 | ✓ | ✓ | ✓ | ✓ | ✓ | 4 | ✓ | 17 | ✓ | — | ✓ | ✓ | ✓ | ✓ | ✓ | QFN-28 |
| C8051F352-GQ | 50 | 8 kB | 768 | ✓ | ✓ | ✓ | ✓ | ✓ | 4 | ✓ | 17 | — | ✓ | ✓ | ✓ | ✓ | ✓ | ✓ | LQFP-32 |
| C8051F353-GM | 50 | 8 kB | 768 | ✓ | ✓ | ✓ | ✓ | ✓ | 4 | ✓ | 17 | — | ✓ | ✓ | ✓ | ✓ | ✓ | ✓ | QFN-28 |

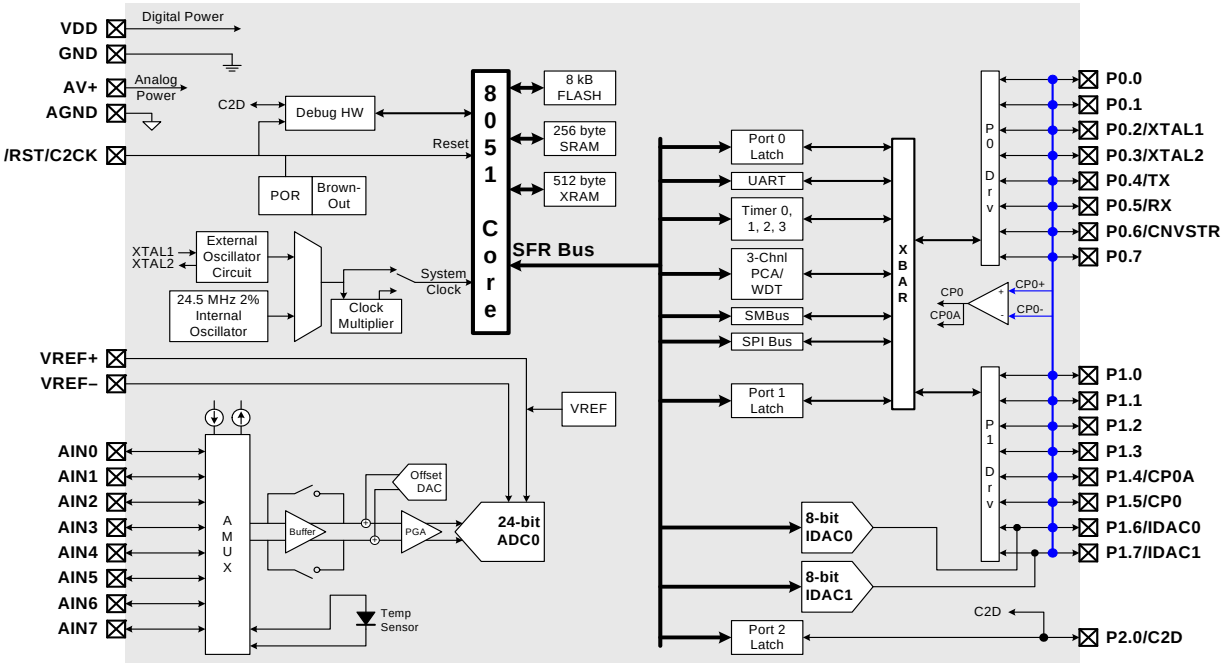



**Figure 1.1. C8051F350 Block Diagram**

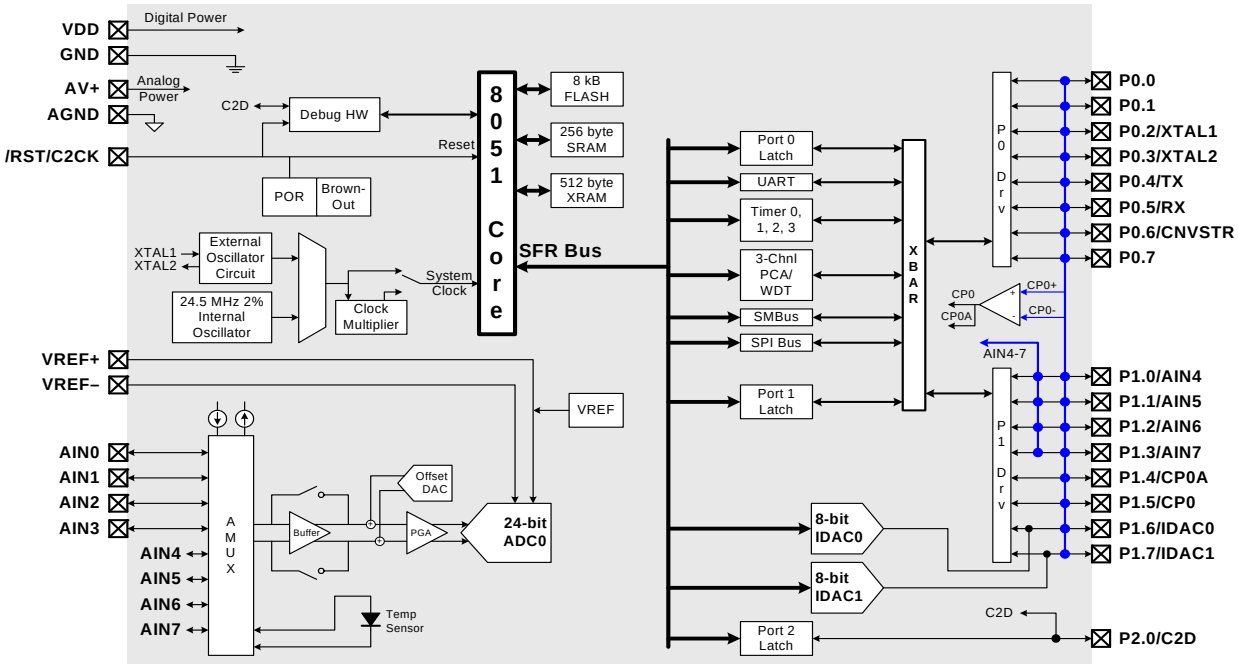



**Figure 1.2. C8051F351 Block Diagram**

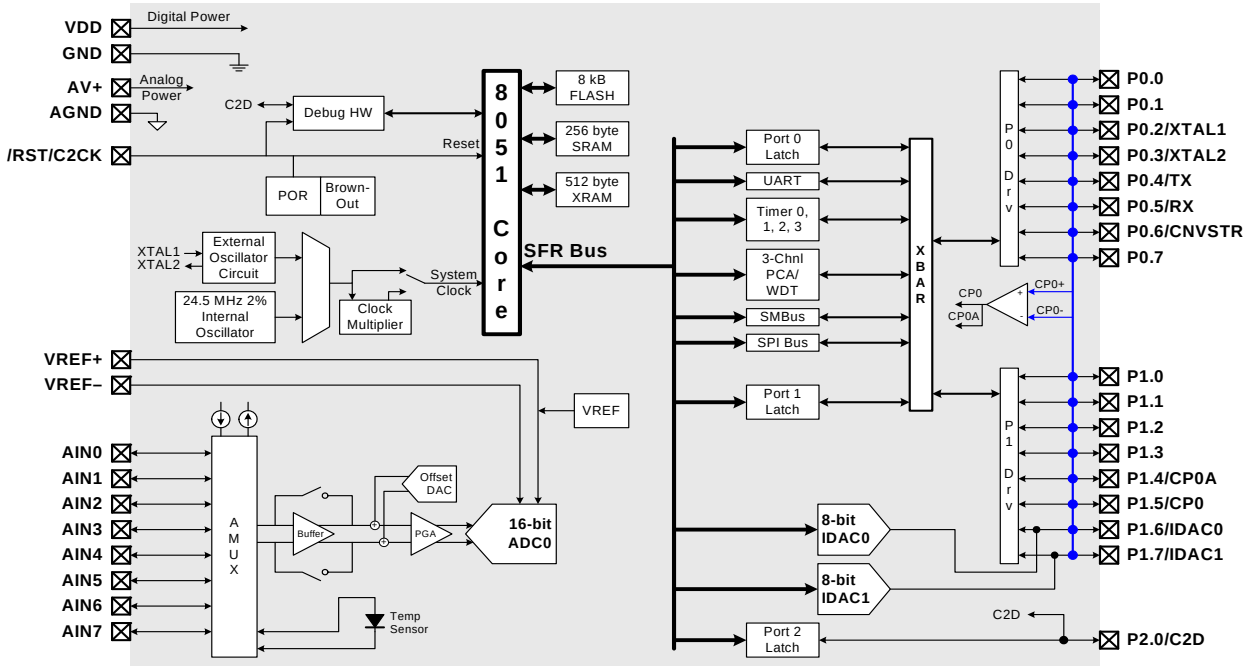



**Figure 1.3. C8051F352 Block Diagram**

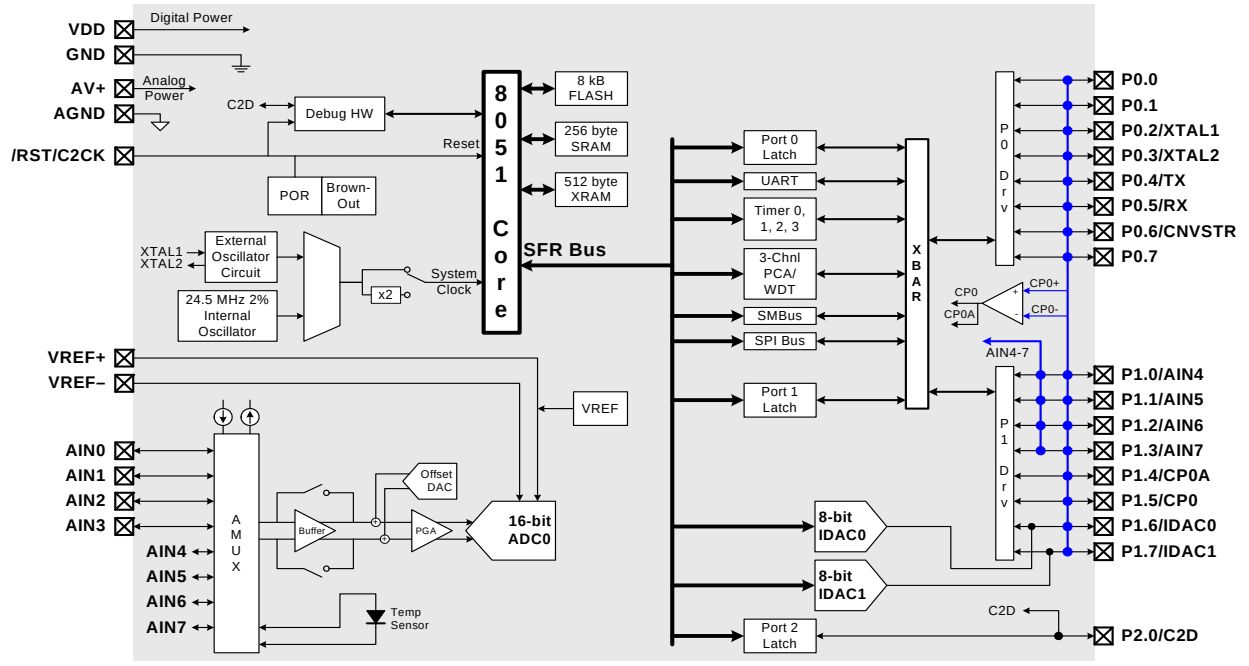



**Figure 1.4. C8051F353 Block Diagram**

## 1.1. CIP-51™ Microcontroller

### 1.1.1. Fully 8051 Compatible Instruction Set

The C8051F35x devices use Silicon Labs' proprietary CIP-51 microcontroller core. The CIP-51 is fully compatible with the MCS-51™ instruction set. Standard 803x/805x assemblers and compilers can be used to develop software. The C8051F35x family has a superset of all the peripherals included with a standard 8052.

### 1.1.2. Improved Throughput

The CIP-51 employs a pipelined architecture that greatly increases its instruction throughput over the standard 8051 architecture. In a standard 8051, all instructions except for MUL and DIV take 12 or 24 system clock cycles to execute, and usually have a maximum system clock of 12 to 24 MHz. By contrast, the CIP-51 core executes 70% of its instructions in one or two system clock cycles, with no instructions taking more than eight system clock cycles.

With the CIP-51's system clock running at 50 MHz, it has a peak throughput of 50 MIPS. The CIP-51 has a total of 109 instructions. The table below shows the total number of instructions that require each execution time.

| Clocks to Execute | 1 | 2 | 2/3 | 3 | 3/4 | 4 | 4/5 | 5 | 8 |
|---|---|---|---|---|---|---|---|---|---|
| Number of Instructions | 26 | 50 | 5 | 14 | 7 | 3 | 1 | 2 | 1 |

### 1.1.3. Additional Features

The C8051F350/1/2/3 SoC family includes several key enhancements to the CIP-51 core and peripherals to improve performance and ease of use in end applications.

An extended interrupt handler allows the numerous analog and digital peripherals to operate independently of the controller core and interrupt the controller only when necessary. By requiring less intervention from the microcontroller core, an interrupt-driven system is more efficient and allows for easier implementation of multi-tasking, real-time systems.

Eight reset sources are available: power-on reset circuitry (POR), an on-chip $V_{DD}$ monitor, a Watchdog Timer, a Missing Clock Detector, a voltage level detection from Comparator0, a forced software reset, an external reset pin, and an illegal Flash access protection circuit. Each reset source except for the POR, Reset Input Pin, or Flash error may be disabled by the user in software. The WDT may be permanently enabled in software after a power-on reset during MCU initialization.

The internal oscillator is factory calibrated to 24.5 MHz ±2%. An external oscillator drive circuit is also included, allowing an external crystal, ceramic resonator, capacitor, RC, or CMOS clock source to generate the system clock. A clock multiplier allows for operation at up to 50 MHz. An external oscillator can also be extremely useful in low power applications, allowing the MCU to run from a slow (power saving) source, while periodically switching to the fast internal oscillator as needed.

## 1.2. On-Chip Debug Circuitry

The C8051F350/1/2/3 devices include on-chip Silicon Labs 2-Wire (C2) debug circuitry that provides non-intrusive, full speed, in-circuit debugging of the production part *installed in the end application.*

Silicon Labs' debugging system supports inspection and modification of memory and registers, breakpoints, and single stepping. No additional target RAM, program memory, timers, or communications channels are required. All the digital and analog peripherals are functional and work correctly while debugging. All the peripherals (except for the ADC and SMBus) are stalled when the MCU is halted, during single stepping, or at a breakpoint in order to keep them synchronized.

The C8051F350DK development kit provides all the hardware and software necessary to develop application code and perform in-circuit debugging with the C8051F35x MCUs. The kit includes software with a developer's studio and debugger, a C2 debug adapter, a target application board with the associated MCU installed, and the required cables and wall-mount power supply. The development kit requires a computer with Windows 98 SE or later installed.

The Silicon Labs IDE interface is a vastly superior developing and debugging configuration, compared to standard MCU emulators that use on-board "ICE Chips" and require the MCU in the application board to be socketed. Silicon Labs' debug paradigm increases ease of use and preserves the performance of the precision analog peripherals.

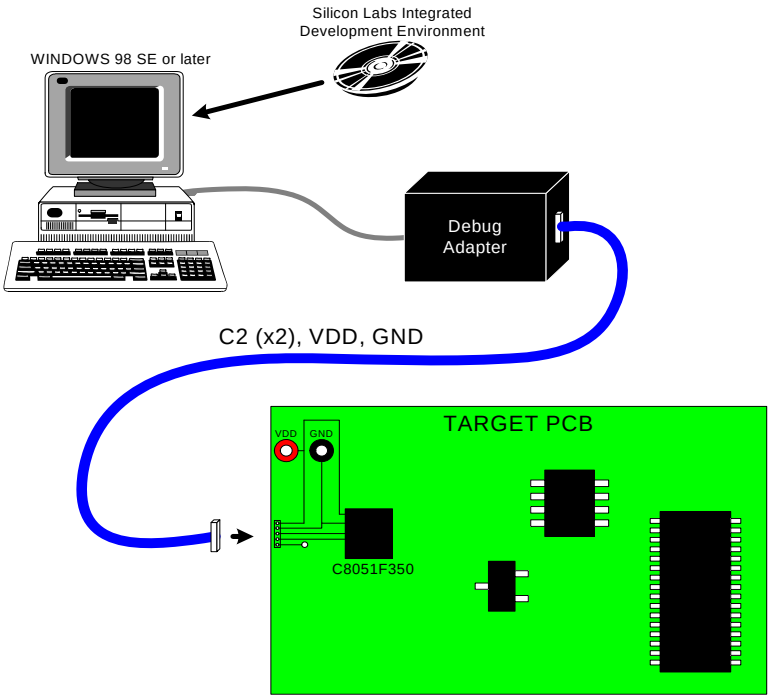



**Figure 1.5. Development/In-System Debug Diagram**

## 1.3. On-Chip Memory

The CIP-51 has a standard 8051 program and data address configuration. It includes 256 bytes of data RAM, with the upper 128 bytes dual-mapped. Indirect addressing accesses the upper 128 bytes of general purpose RAM, and direct addressing accesses the 128 byte SFR address space. The lower 128 bytes of RAM are accessible via direct and indirect addressing. The first 32 bytes are addressable as four banks of general purpose registers, and the next 16 bytes can be byte addressable or bit addressable.

Program memory consists of 8 kB bytes of Flash. This memory may be reprogrammed in-system in 512 byte sectors, and requires no special off-chip programming voltage.

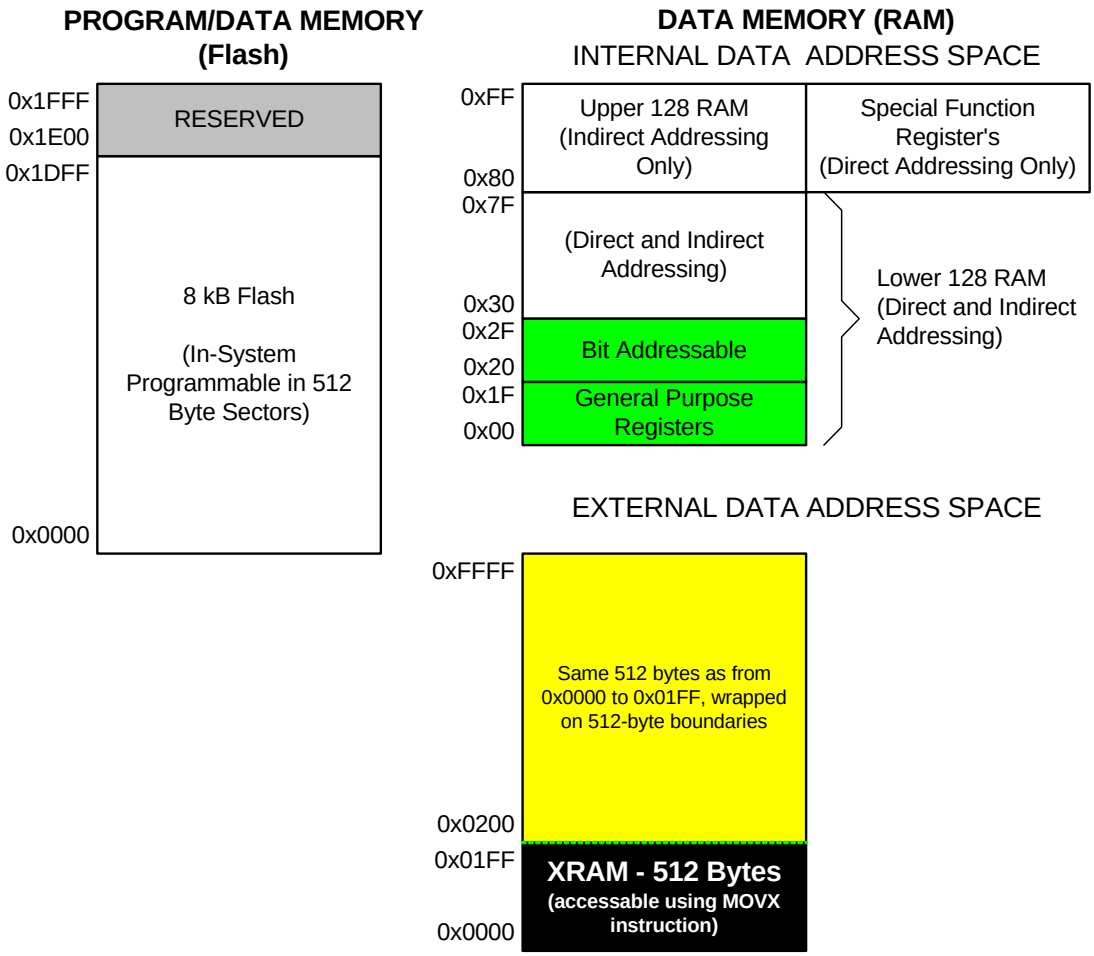



**Figure 1.6. Memory Map**

## 1.4. 24 or 16-Bit Analog to Digital Converter (ADC0)

The C8051F350/1/2/3 include a fully-differential, 24-bit (C8051F350/1) or 16-bit (C8051F352/3) Sigma-Delta Analog to Digital Converter (ADC) with on-chip calibration capabiliites. Two separate decimation filters can be programmed for throughputs of up to 1 kHz. An internal 2.5 V reference is available, or a differential external reference can be used for ratiometric measurements. A Programmable Gain Amplifier (PGA) is included, with eight gain settings up to 128x. An analog front-end multiplexer connects the differential inputs to eight external pins, the internal temperature sensor, or AGND. The on-chip input buffers can be used to provide a high input impedance for direct connection to sensitive transducers. An 8-bit offset DAC allows for correction of large input offset voltages.

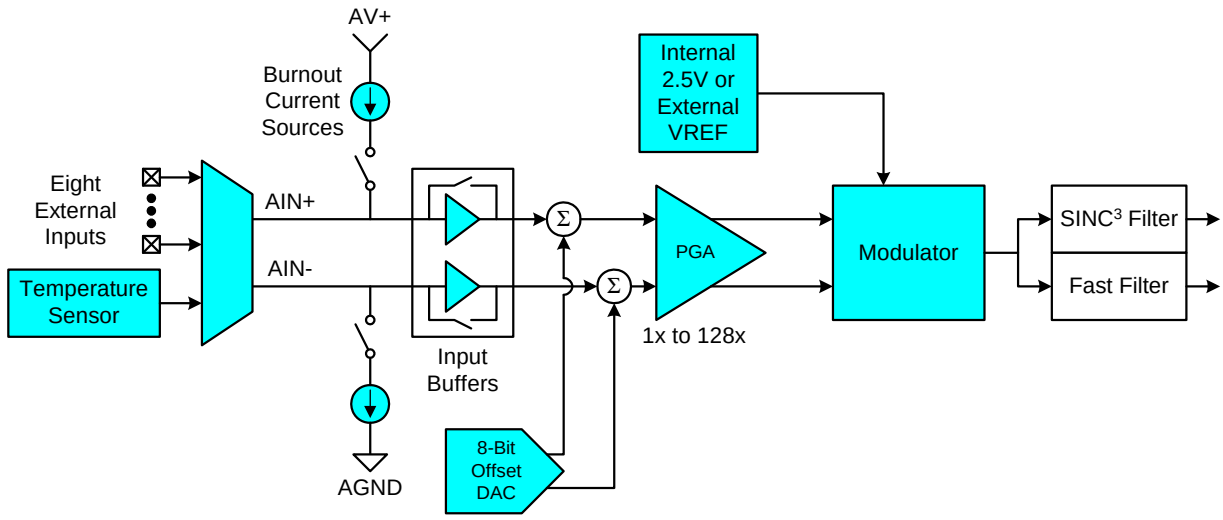



**Figure 1.7. ADC0 Block Diagram**

## 1.5. Two 8-bit Current-Mode DACs

The C8051F350/1/2/3 devices include two 8-bit current-mode Digital-to-Analog Converters (IDACs). The maximum current output of the IDACs can be adjusted for four different current settings; 0.25 mA, 0.5 mA, 1 mA, and 2 mA. A flexible output update mechanism allows for seamless full-scale changes, and supports jitter-free updates for waveform generation. IDAC updates can be performed on-demand, scheduled on a Timer overflow, or synchronized with an external signal. Figure 1.8 shows a block diagram of the IDAC circuitry.

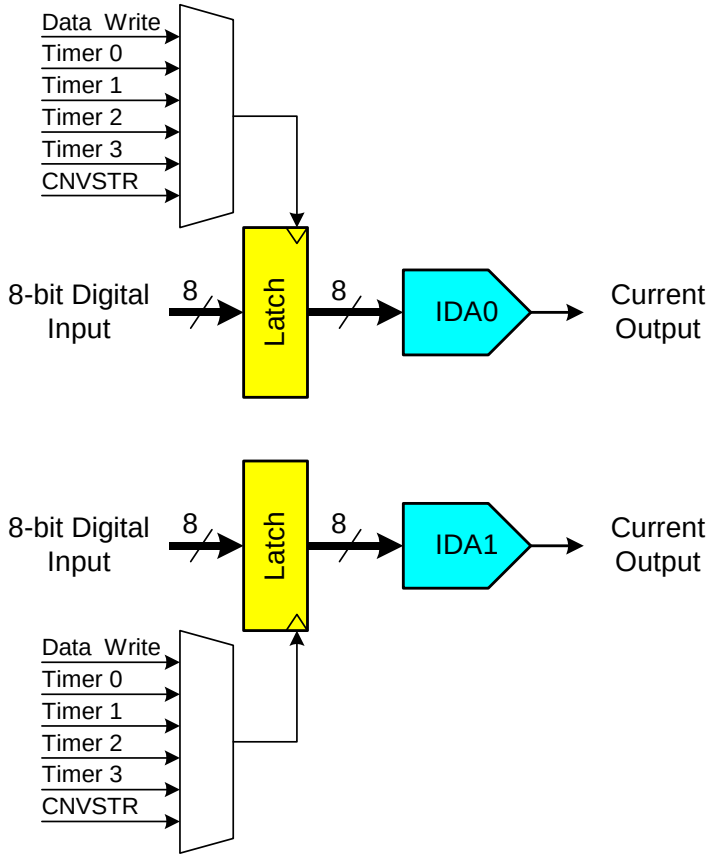



**Figure 1.8. IDAC Block Diagram**

## 1.6. Programmable Comparator

C8051F350/1/2/3 devices include a software-configurable voltage comparator with an input multiplexer. The Comparator offers programmable response time and hysteresis and two outputs that are optionally available at the Port pins: a synchronous “latched” output (CP0), or an asynchronous “raw” output (CP0A). Comparator interrupts may be generated on rising, falling, or both edges. When in IDLE mode, these interrupts may be used as a “wake-up” source for the processor. Comparator0 may also be configured as a reset source. A block diagram of the Comparator is shown in Figure 1.9.

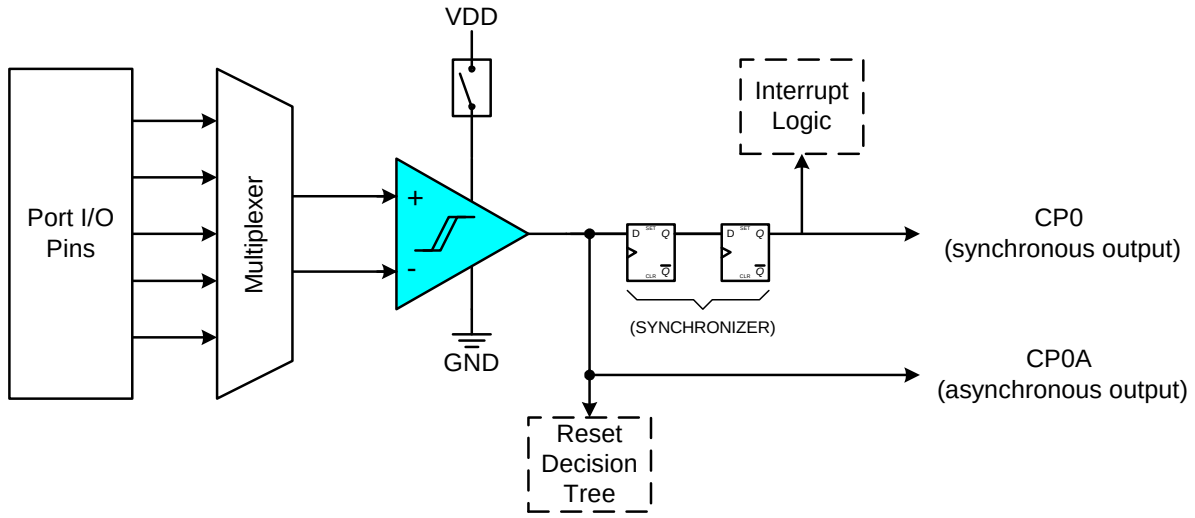



**Figure 1.9. Comparator0 Block Diagram**

## 1.7. Serial Ports

The C8051F350/1/2/3 Family includes an SMBus/I2C interface, a full-duplex UART with enhanced baud rate configuration, and an Enhanced SPI interface. Each of the serial buses is fully implemented in hardware and makes extensive use of the CIP-51's interrupts, thus requiring very little CPU intervention.

## 1.8. Port Input/Output

C8051F350/1/2/3 devices include 17 I/O pins. Port pins are organized as two byte-wide ports and one 1-bit port. The port pins behave like typical 8051 ports with a few enhancements. Each port pin can be configured as a digital or analog I/O pin. Pins selected as digital I/O can be configured for push-pull or open-drain operation. The “weak pull-ups” that are fixed on typical 8051 devices may be globally disabled to save power.

The Digital Crossbar allows mapping of internal digital system resources to port I/O pins. On-chip conter/timers, serial buses, hardware interrupts, and other digital signals can be configured to appear on the port pins using the Crossbar control resgiters. This allows the user to select the exact mix of general-purpose port I/O, digital, and analog resources needed for the application.

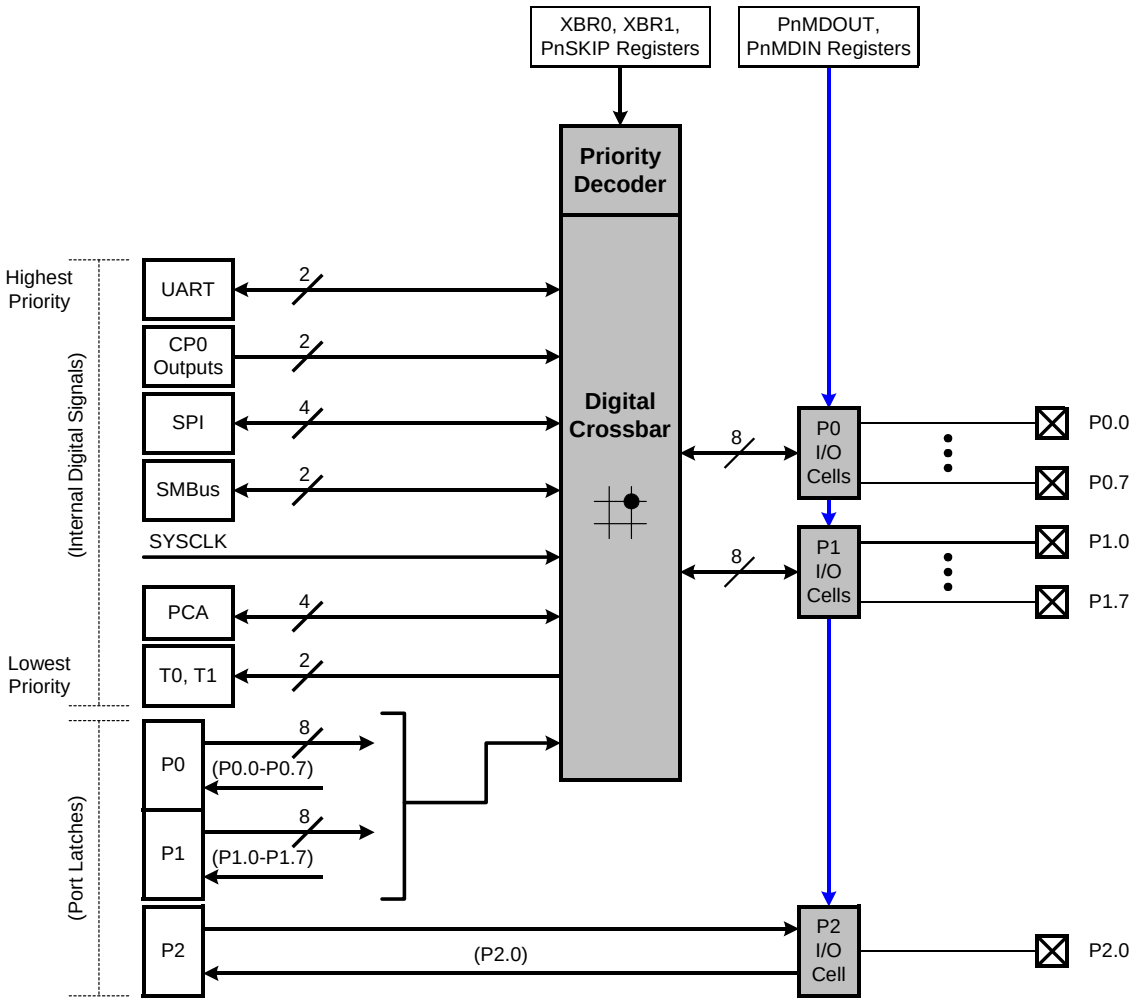



**Figure 1.10. Port I/O Functional Block Diagram**

## 1.9. Programmable Counter Array

The Programmable Counter Array (PCA0) provides enhanced timer functionality while requiring less CPU intervention than the standard 8051 counter/timers. The PCA consists of a dedicated 16-bit counter/timer and three 16-bit capture/compare modules. The counter/timer is driven by a programmable timebase that can select between six sources: system clock, system clock divided by four, system clock divided by twelve, the external oscillator clock source divided by 8, Timer 0 overflow, or an external clock signal on the External Clock nput (ECI) input pin.

Each capture/compare module may be configured to operate independently in one of six modes: Edge-Triggered Capture, Software Timer, High-Speed Output, Frequency Output, 8-Bit PWM, or 16-Bit PWM. Additionally, PCA Module 2 may be used as a watchdog timer (WDT), and is enabled in this mode following a system reset. The PCA Capture/Compare Module I/O and the External Clock Input may be routed to Port I/O using the digital crossbar.

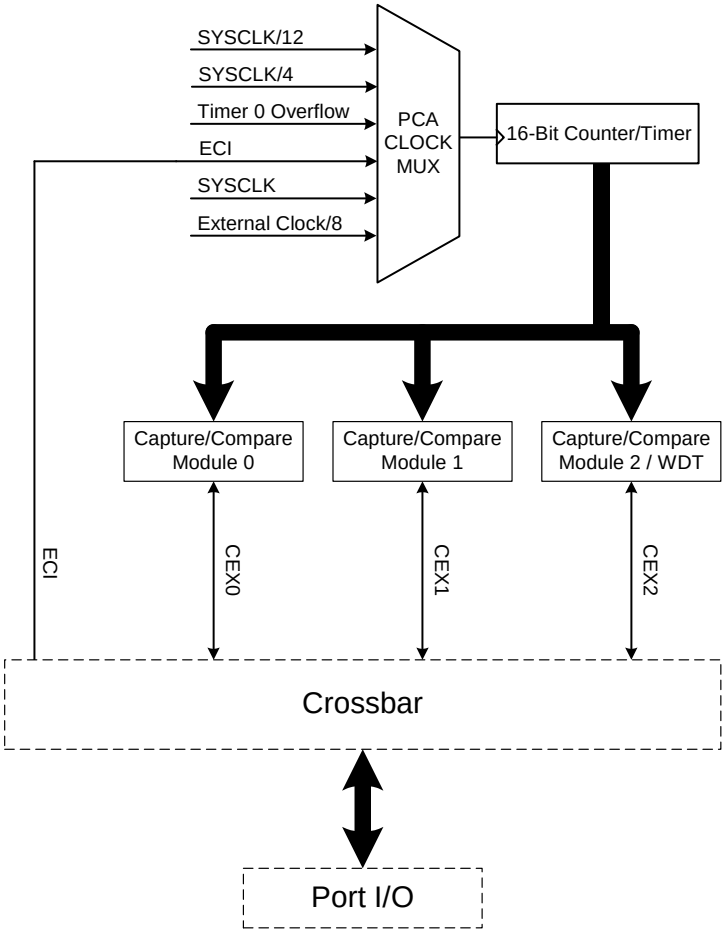



**Figure 1.11. PCA Block Diagram**

# 2. Absolute Maximum Ratings

## Table 2.1. Absolute Maximum Ratings

| Parameter | Min | Typ | Max | Units |
|---|---|---|---|---|
| Ambient temperature under bias | –55 | — | 125 | °C |
| Storage Temperature | –65 | — | 150 | °C |
| Voltage on AIN0.0–AIN0.7, VREF+, and VREF– with respect to DGND | –0.3 | — | $V_{DD}$ + 0.3 | V |
| Voltage on any Port 0, 1, or 2 Pin or /RST with respect to DGND | –0.3 | — | 5.8 | V |
| Voltage on $V_{DD}$ with respect to DGND | –0.3 | — | 4.2 | V |
| Voltage on AV+ with respect to AGND | –0.3 | — | 4.2 | V |
| Maximum output current sunk by any Port 0, 1, or 2 pin | — | — | 100 | mA |
| Maximum output current sunk by any other I/O pin | — | — | 50 | mA |
| Maximum output current sourced by any Port 0, 1, or 2 pin | — | — | 100 | mA |
| Maximum output current sourced by any other I/O pin | — | — | 50 | mA |
| Maximum Total current through $V_{DD}$, AV+, DGND, and AGND | — | — | 500 | mA |

**Note:** Stresses above those listed under “Absolute Maximum Ratings” may cause permanent damage to the device. This is a stress rating only and functional operation of the devices at those or any other conditions above those indicated in the operation listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

# 3. Global DC Electrical Characteristics

## Table 3.1. Global DC Electrical Characteristics

–40 to +85 °C, 25 MHz System Clock unless otherwise specified.

| Parameter | Conditions | Min | Typ | Max | Units |
|---|---|---|---|---|---|
| Analog Supply Voltage[1] | | 2.7 | 3.0 | 3.6 | V |
| Analog Supply Current | Internal REF, ADC, IDACs, Comparators all active | — | 0.75 | 1.3 | mA |
| Analog Supply Current with analog sub-systems inactive | Internal REF, ADC, IDACs, Comparators all disabled, oscillator disabled | — | < 1 | — | µA |
| Analog-to-Digital Supply Delta ($\|V_{DD} – AV+\|$) | | — | — | 0.5 | V |
| Digital Supply Voltage | | 2.7 | 3.0 | 3.6 | V |
| Digital Supply Current with CPU active | $V_{DD}$ = 2.7 V; SYSCLK = 25 MHz<br>$V_{DD}$ = 2.7 V; SYSCLK = 50 MHz<br>$V_{DD}$ = 3.3 V; SYSCLK = 25 MHz<br>$V_{DD}$ = 3.3 V; SYSCLK = 50 MHz | —<br>—<br>—<br>— | 9.9<br>17.8<br>13.6<br>24.9 | 11.3<br>20.0<br>15.5<br>27.1 | mA<br>mA<br>mA<br>mA |
| Digital Supply Current with CPU inactive (not accessing Flash) | $V_{DD}$ = 2.7 V; SYSCLK = 25 MHz<br>$V_{DD}$ = 2.7 V; SYSCLK = 50 MHz<br>$V_{DD}$ = 3.3 V; SYSCLK = 25 MHz<br>$V_{DD}$ = 3.3 V; SYSCLK = 50 MHz | —<br>—<br>—<br>— | 5.7<br>11.1<br>7.5<br>15.0 | 6.6<br>12.7<br>8.5<br>16.5 | mA<br>mA<br>mA<br>mA |
| Digital Supply Current (shutdown) | Oscillator not running | — | < 0.1 | — | µA |
| Digital Supply RAM Data Retention Voltage | | — | 1.5 | — | V |
| SYSCLK (System Clock) [2,3] | | 0 | — | 50 | MHz |
| Specified Operating Temperature Range | | –40 | — | +85 | °C |

**Notes:**

1. Analog Supply AV+ must be greater than 1 V for $V_{DD}$ monitor to operate.
2. SYSCLK is the internal device clock. For operational speeds in excess of 25 MHz, SYSCLK must be derived from the internal clock multiplier.
3. SYSCLK must be at least 32 kHz to enable debugging.

# 4. Pinout and Package Definitions

**Table 4.1. Pin Definitions for the C8051F350/1/2/3**

| Name | Pin Numbers | | Type | Description |
|---|---|---|---|---|
| | ‘F350 ‘F352 | ‘F351 ‘F353 | | |
| $V_{DD}$ | 21 | 17 | Power | Digital Supply Voltage. Must be tied to +2.7 V to +3.6 V power. |
| DGND | 22 | 18 | Ground | Digital Ground. Must be tied to Ground. |
| AV+ | 10 | 6 | Power | Analog Supply Voltage. Must be tied to +2.7 V to +3.6 V power. |
| AGND | 9 | 5 | Ground | Analog Ground. Must be tied to Ground. |
| /RST<br><br><br><br><br>C2CK | 12 | 8 | D I/O<br><br><br><br><br>D I/O | Device Reset. Open-drain output of internal POR or $V_{DD}$ monitor. An external source can initiate a system reset by driving this pin low for at least 15 µs. A 1kΩ pull-up to $V_{DD}$ is recommended. See Reset Sources Section.<br><br>Clock signal for the C2 Debug Interface. |
| P2.0/<br><br>C2D | 11 | 7 | D I/O<br><br>D I/O | Port 2.0. See Port I/O Section for a complete description.<br><br>Bi-directional data signal for the C2 Debug Interface. |
| P0.0 | 13 | 9 | D I/O or A In | Port 0.0. See Port I/O Section for a complete description. |
| P0.1 | 14 | 10 | D I/O or A In | Port 0.1. See Port I/O Section for a complete description. |
| P0.2/<br><br><br>XTAL1 | 15 | 11 | D I/O or A In<br><br><br>A In | Port 0.2. See Port I/O Section for a complete description.<br><br><br>This pin is the external oscillator return for a crystal or resonator. See Oscillator Section. |
| P0.3/<br><br>XTAL2 | 16 | 12 | D I/O<br><br>A I/O or D In | Port 0.3. See Port I/O Section for a complete description.<br><br>This pin is the excitation driver for an external crystal or resonator, or an external clock input for CMOS, capacitor, or RC oscillator configurations. See Oscillator Section. |
| P0.4 | 17 | 13 | D I/O or A In | Port 0.4. See Port I/O Section for a complete description. |
| P0.5 | 18 | 14 | D I/O or A In | Port 0.5. See Port I/O Section for a complete description. |

**Table 4.1. Pin Definitions for the C8051F350/1/2/3 (Continued)**

| Name | Pin Numbers | | Type | Description |
|---|---|---|---|---|
| | 'F350 'F352 | 'F351 'F353 | | |
| P0.6/ | 19 | 15 | D I/O or A In | Port 0.6. See Port I/O Section for a complete description. |
| CNVSTR | | | D In | External Convert Start Input for IDACs (See IDAC Section for complete description). |
| P0.7 | 20 | 16 | D I/O or A In | Port 0.7. See Port I/O Section for a complete description. |
| P1.0/ | 23 | 19 | D I/O or A In | Port 1.0. See Port I/O Section for a complete description. |
| AIN0.4 | | | A In | ADC0 Input Channel 4 (C8051F351/3 - See ADC0 Section for complete description). |
| P1.1/ | 24 | 20 | D I/O or A In | Port 1.1. See Port I/O Section for a complete description. |
| AIN0.5 | | | A In | ADC0 Input Channel 5 (C8051F351/3 - See ADC0 Section for complete description). |
| P1.2/ | 25 | 21 | D I/O or A In | Port 1.2. See Port I/O Section for a complete description. |
| AIN0.6 | | | A In | ADC0 Input Channel 6 (C8051F351/3 - See ADC0 Section for complete description). |
| P1.3/ | 26 | 22 | D I/O or A In | Port 1.3. See Port I/O Section for a complete description. |
| AIN0.7 | | | A In | ADC0 Input Channel 7 (C8051F351/3 - See ADC0 Section for complete description). |
| P1.4 | 27 | 23 | D I/O or A In | Port 1.4. See Port I/O Section for a complete description. |
| P1.5 | 28 | 24 | D I/O or A In | Port 1.5. See Port I/O Section for a complete description. |
| P1.6/ | 29 | 25 | D I/O or A In | Port 1.6. See Port I/O Section for a complete description. |
| IDA0 | | | A Out | IDAC0 Output (See IDAC Section for complete description). |
| P1.7/ | 30 | 26 | D I/O or A In | Port 1.7. See Port I/O Section for a complete description. |
| IDA1 | | | A Out | IDAC1 Output (See IDAC Section for complete description). |

**Table 4.1. Pin Definitions for the C8051F350/1/2/3 (Continued)**

| Name | Pin Numbers | | Type | Description |
|---|---|---|---|---|
| | ‘F350 ‘F352 | ‘F351 ‘F353 | | |
| AIN0.0 | 1 | 1 | A In | ADC0 Input Channel 0 (See ADC0 Section for complete description). |
| AIN0.1 | 2 | 2 | A In | ADC0 Input Channel 1 (See ADC0 Section for complete description). |
| AIN0.2 | 3 | 3 | A In | ADC0 Input Channel 2(See ADC0 Section for complete description). |
| AIN0.3 | 4 | 4 | A In | ADC0 Input Channel 3 (See ADC0 Section for complete description). |
| AIN0.4 | 5 | — | A In | ADC0 Input Channel 4 (C8051F350/2 - See ADC0 Section for complete description). |
| AIN0.5 | 6 | — | A In | ADC0 Input Channel 5 (C8051F350/2 - See ADC0 Section for complete description). |
| AIN0.6 | 7 | — | A In | ADC0 Input Channel 6 (C8051F350/2 - See ADC0 Section for complete description). |
| AIN0.7 | 8 | — | A In | ADC0 Input Channel 7 (C8051F350/2 - See ADC0 Section for complete description). |
| VREF+ | 31 | 27 | A I/O | VREF Positive Voltage Pin (See VREF Section for complete description). |
| VREF– | 32 | 28 | A I/O | VREF Negative Voltage Pin (See VREF Section for complete description). |

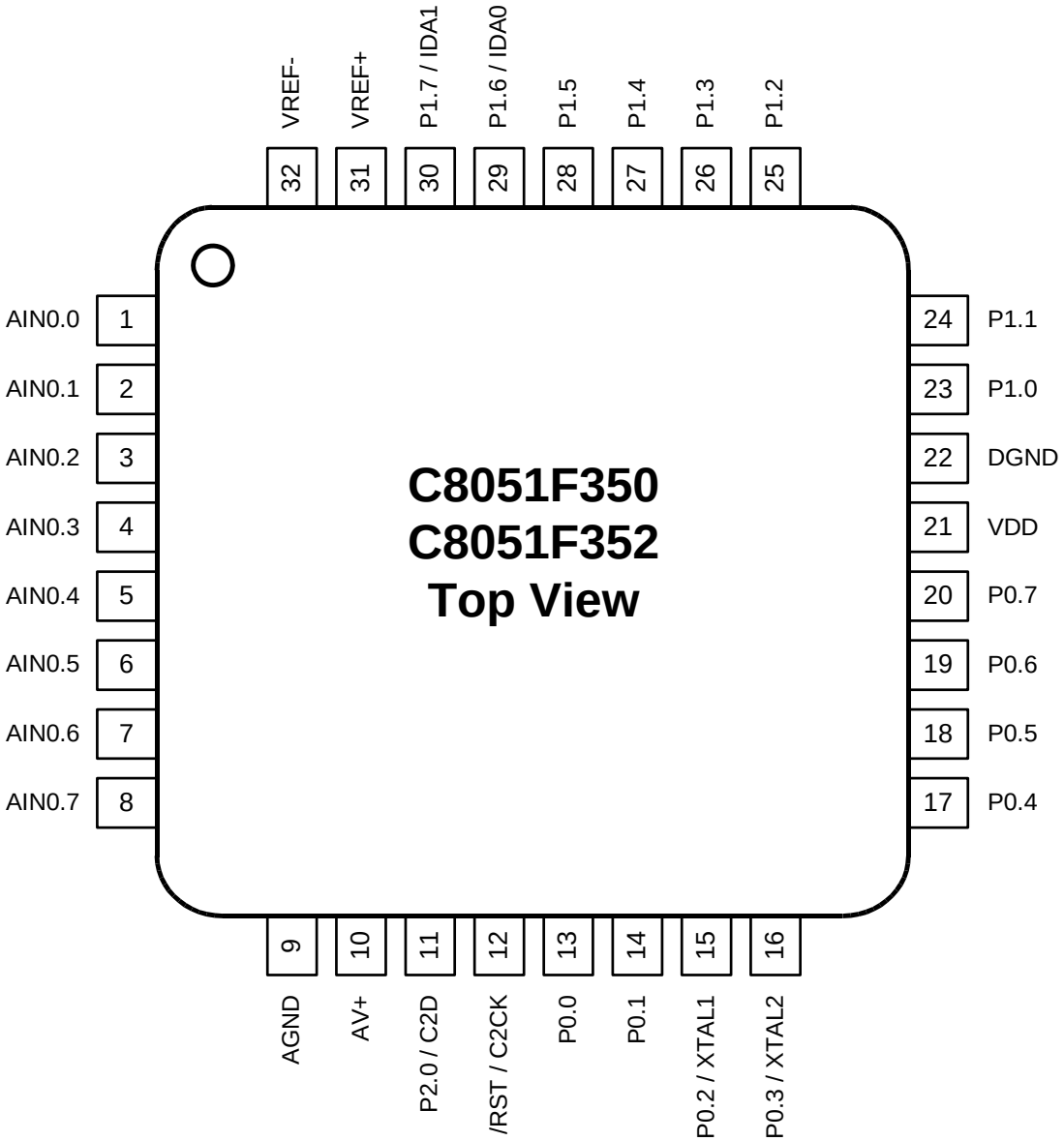



**Figure 4.1. LQFP-32 Pinout Diagram (Top View)**

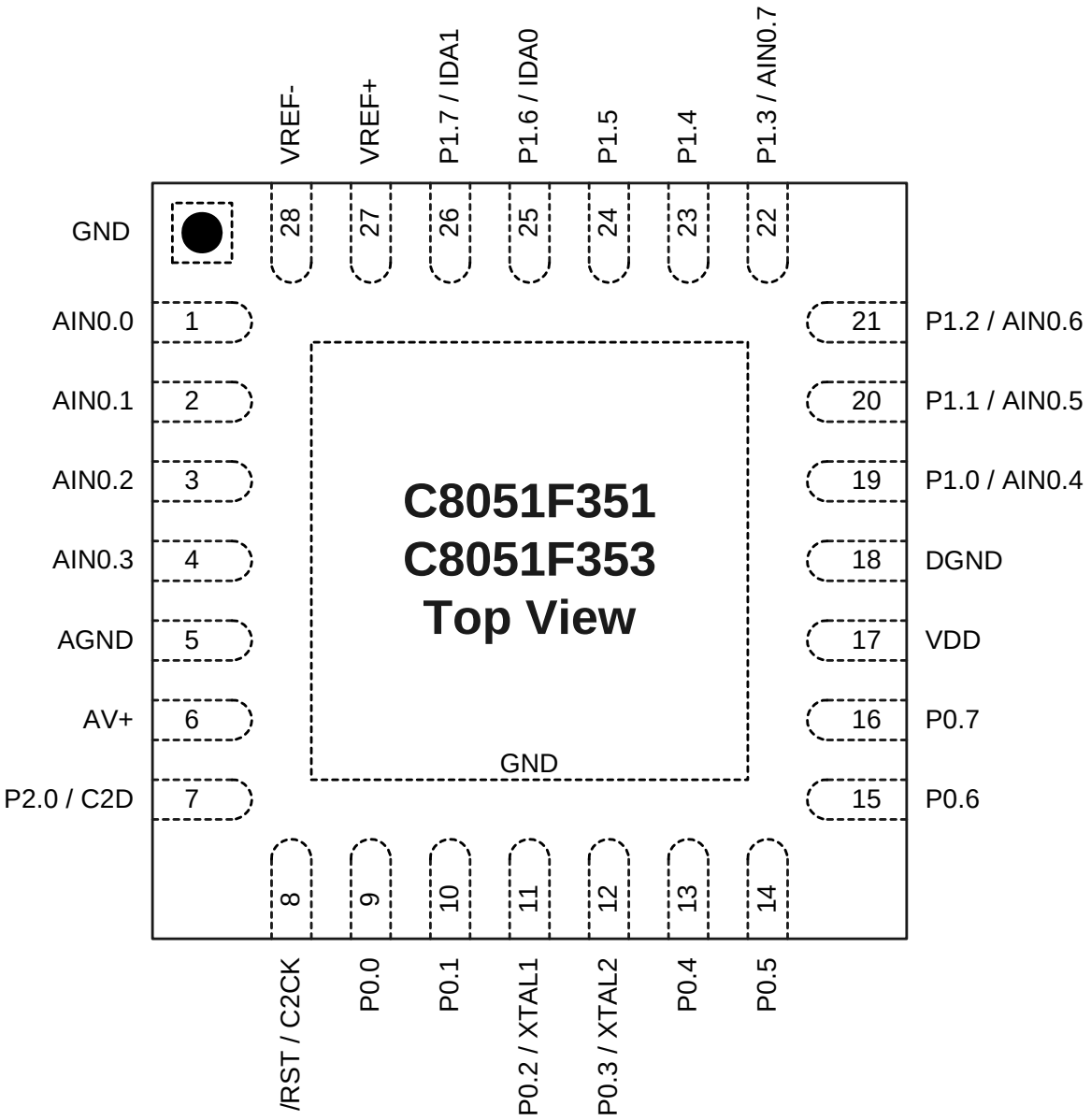



**Figure 4.2. QFN-28 Pinout Diagram (Top View)**

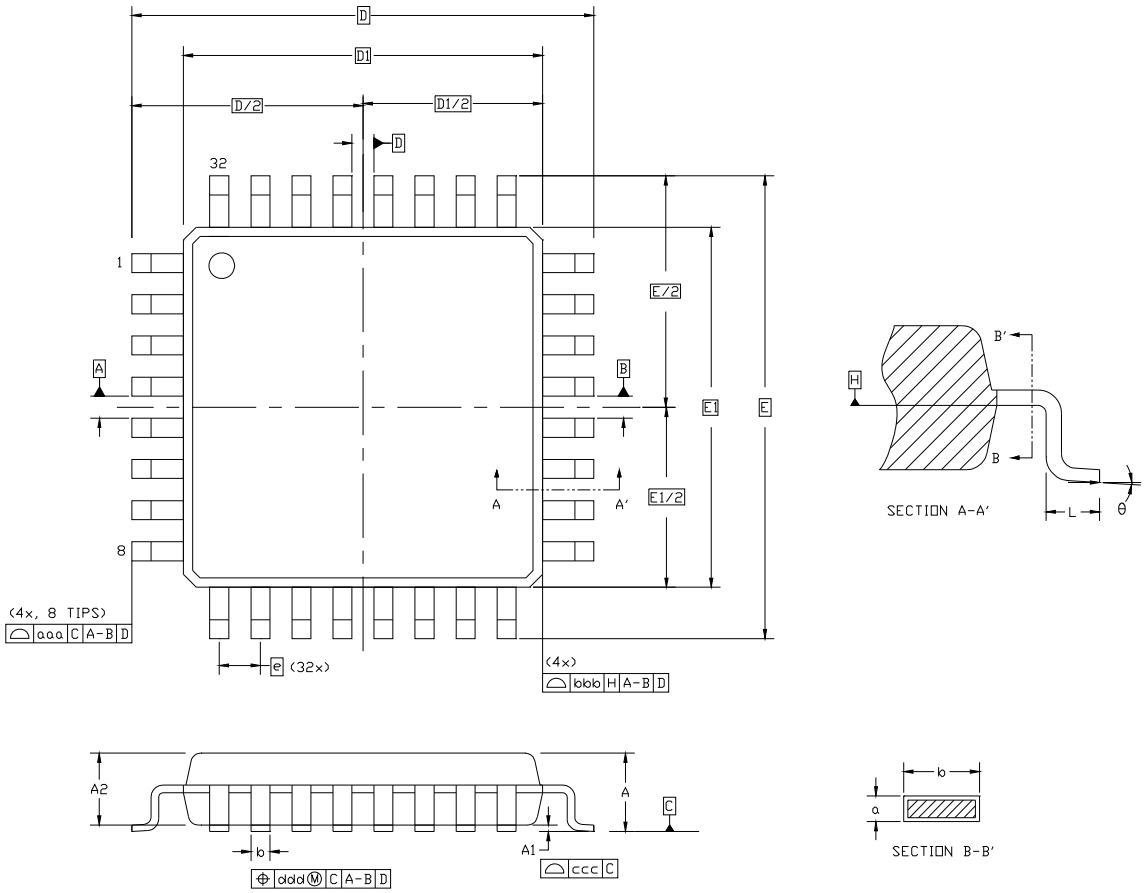



**Figure 4.3. LQFP-32 Package Diagram**

**Table 4.2. LQFP-32 Package Dimensions**

| | MM | | |
|---|---|---|---|
| | **MIN** | **TYP** | **MAX** |
| A | — | — | 1.60 |
| A1 | 0.05 | — | 0.15 |
| A2 | 1.35 | 1.40 | 1.45 |
| b | 0.30 | 0.37 | 0.45 |
| c | 0.09 | — | 0.20 |
| D | — | 9.00 | — |
| D1 | — | 7.00 | — |
| e | — | 0.80 | — |
| E | — | 9.00 | — |
| E1 | — | 7.00 | — |
| L | 0.45 | 0.60 | 0.75 |

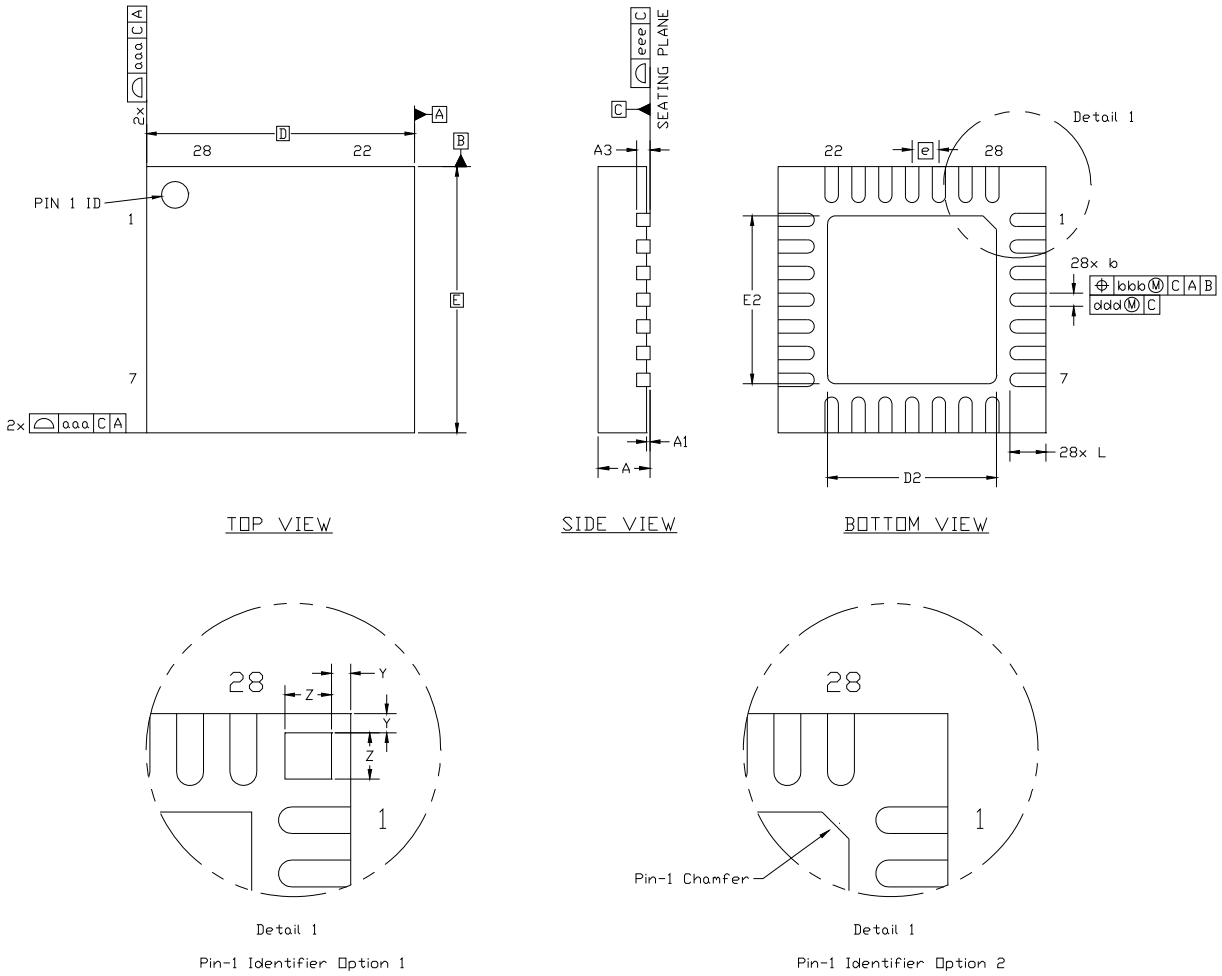



**Figure 4.4. QFN-28 Package Drawing**

**Table 4.3. QFN-28 Package Dimensions**

| | MM | | |
|---|---|---|---|
| | **MIN** | **TYP** | **MAX** |
| A | 0.80 | 0.90 | 1.00 |
| A1 | 0.03 | 0.07 | 0.11 |
| A3 | | 0.25 REF | |
| b | 0.18 | 0.25 | 0.30 |
| D | | 5.00 BSC. | |
| D2 | 2.90 | 3.15 | 3.35 |
| e | | 0.50 BSC. | |
| E | | 5.00 BSC. | |
| E2 | 2.90 | 3.15 | 3.35 |
| L | 0.45 | 0.55 | 0.65 |

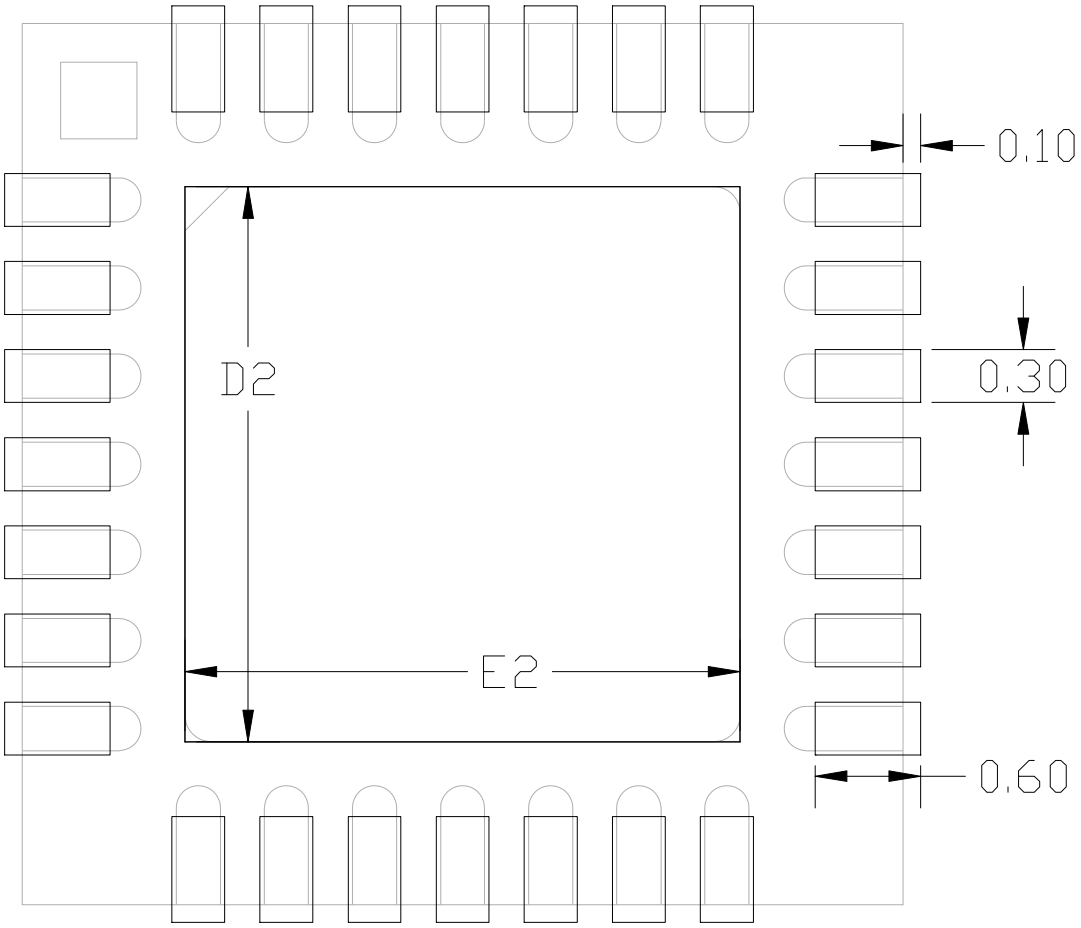



**Figure 4.5. Typical QFN-28 Landing Diagram**

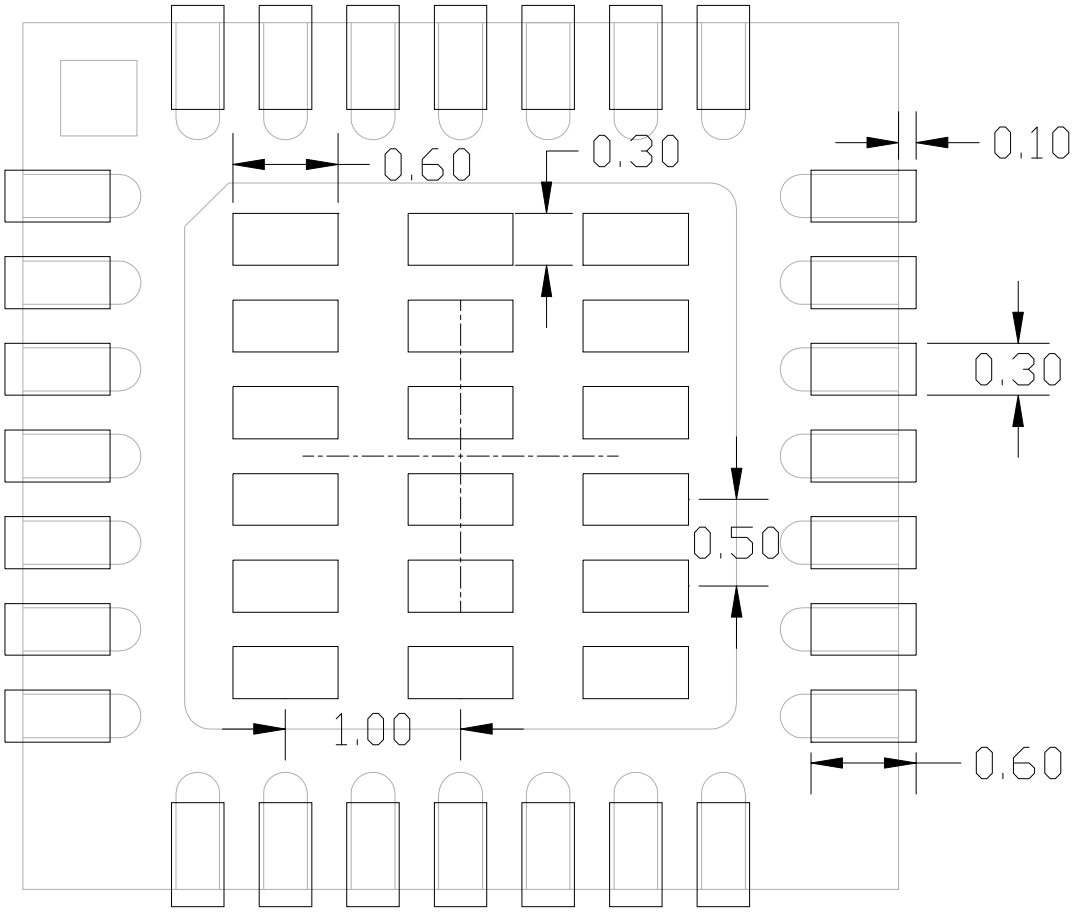



**Figure 4.6. Typical QFN-28 Solder Paste Diagram**

## NOTES:

# 5. 24 or 16-Bit Analog to Digital Converter (ADC0)

The C8051F350/1/2/3 include a fully-differential, 24-bit (C8051F350/1) or 16-bit (C8051F352/3) Sigma-Delta Analog to Digital Converter (ADC) with on-chip calibration capabiliites. Two separate decimation filters can be programmed for throughputs of up to 1 kHz. An internal reference is available, or a differential external reference can be used for ratiometric measurements. A Programmable Gain Amplifier (PGA) is included, with eight gain settings up to 128x. The on-chip input buffers can be used to provide a high input impedance for direct connection to sensitive transducers. An 8-bit offset DAC allows for correction of large input offset voltages.

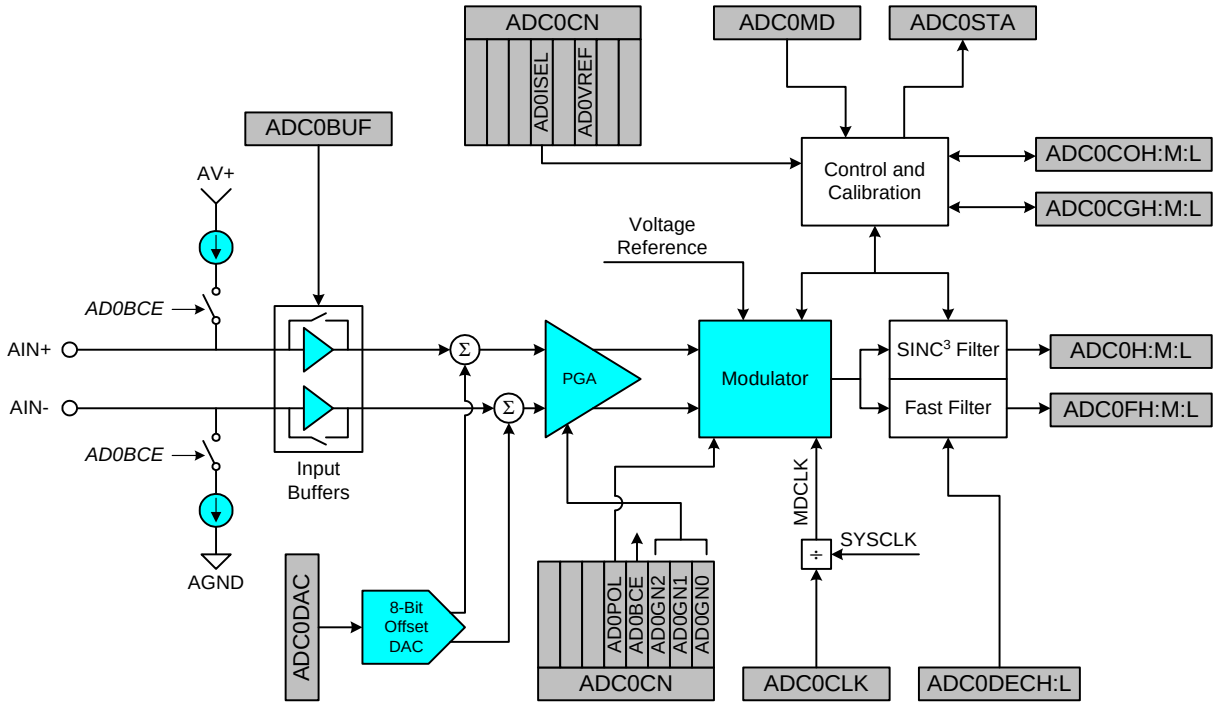



**Figure 5.1. ADC0 Block Diagram**

## 5.1. Configuration

ADC0 is enabled by setting the AD0EN bit in register ADC0MD (SFR Definition 5.3) to ‘1’. When the ADC is disabled, it is placed in a low-power shutdown mode with all clocks turned off, to minimize unnecessary power consumption. The ADC will retain all of its settings in shutdown mode, with the exception of the AD0SM bits, which are reset to 000b (Idle Mode).

### 5.1.1. Voltage Reference Selection

The ADC’s voltage reference is selected using the AD0VREF bit in register ADC0CF (SFR Definition 5.2). When set to ‘1’, the ADC uses an external voltage reference source. When cleared to ‘0’, the internal reference is used. A more detailed description of the voltage reference options can be found in Section “7. Voltage Reference’ on page 73.

### 5.1.2. Analog Inputs

The ADC’s analog inputs are connected to external device pins or internal voltages as described in Section “5.6. Analog Multiplexer’ on page 59. They can be configured as either single-ended (one independent input measured with respect to AGND) or differential (two independent inputs measured with respect to each other). For accurate measurements, the ADC inputs must remain within the input range specifications found in Table 5.3. To prevent damage to the device, all external ADC inputs must also remain within the Absolute Maximum ratings for the input pin, given in Table 2.1.

#### 5.1.2.1. Programmable Gain Amplifier

A programmable gain amplifier (PGA) provides amplification settings of 1, 2, 4, 8, 16, 32, 64, and 128 for the ADC inputs. The PGA gain setting is controlled by the AD0GN bits in register ADC0CN (SFR Definition 5.1).

#### 5.1.2.2. Input Buffers

Independent input buffers are included for AIN+ and AIN–, as shown in Figure 5.2. Each input has a set of two buffers that can be used to minimize the input current of the ADC for sensitive measurements. The “low” input buffer can be used when the absolute pin input voltage is in the lower half of the supply range. The “high” input buffer on each pin can be used when the absolute pin input voltage is in the upper half of the supply range. See Table 5.3 for the input buffer range specifications. The input buffers can also be bypassed, for a direct connection to the PGA inputs. The ADC input buffers are controlled with the ADC0BUF register (SFR Definition 5.8).

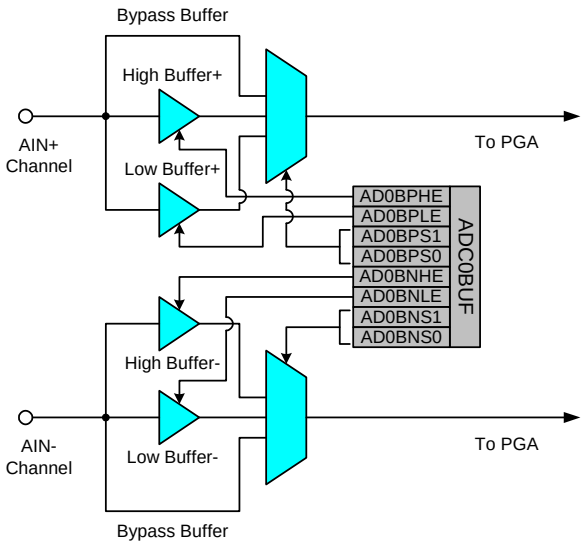



**Figure 5.2. ADC0 Buffer Control**

### 5.1.3. Modulator Clock

The ADC0CLK register (SFR Definition 5.4) holds the Modulator Clock (MDCLK) divisor value. The modulator clock determines the switching frequency for the ADC sampling capacitors. Optimal performance will be achieved when the MDCLK frequency is equal to 2.4576 MHz. The modulator samples the input at a rate of MDCLK / 128.

### 5.1.4. Decimation Ratio

The decimation ratio of the ADC filters is selected by the DECI[10:0] bits in the ADC0DECH and ADC0DECL registers (SFR Definition 5.5 and SFR Definition 5.6, respectively). The decimation ratio is equal to 1 + DECI[10:0]. The decimation ratio determines how many modulator samples are used to generate a single output word. The ADC output word rate is equal to the modulator sampling rate divided by the decimation ratio. For more information on how the ADC output word rate is derived, see SFR Definition 5.4 and SFR Definition 5.6. Higher decimation ratios will produce lower-noise results over a longer conversion period. The minimum decimation ratio is 20. **When using the fast filter output, the decimation ratio must be set to a multiple of 8.**

## 5.2. Calibrating the ADC

ADC0 can be calibrated in-system for both gain and offset, using internal or system calibration modes. To ensure calibration accuracy, offset calibrations must be performed prior to gain calibrations. It is not necessary to perform both internal and system calibrations, as a system calibration will also compensate for any internal error sources.

Offset calibration is a single-point measurement that sets which input voltage produces a zero at the ADC output. When performing an offset calibration, any deviation from zero in the measurement is stored in the offset register. The offset value is subtracted from all conversions as they take place.

Gain calibration is a two-point measurement that sets the slope of the ADC transfer function. When performed, a gain calibration takes only a single measurement, which is assumed to be the desired full-scale value in the ADC transfer function. The offset calibration value is used as the other point in the gain calibration measurement, so that a gain factor can be calculated. After offset correction, conversions are multiplied by the gain factor.

Calibrations are initiated by writing the ADC System Mode bits (AD0SM) to one of the calibration options. During a calibration, the AD0CBSY bit is set to ‘1’. Upon completion of a calibration the the AD0SM bits will return to Idle mode, the AD0CBSY bit will be cleared to ‘0’, the AD0CALC bit will be set to ‘1’, and an ADC interrupt will be generated. Calibration results are also written to the appropriate calibration registers when the calibration is complete.

### 5.2.1. Internal Calibration

Internal calibration is performed without requiring a specific voltage on the ADC input pins. Internal calibrations can be performed in three different ways: offset only, gain only, or full (offset and gain). A full internal calibration consists of an internal offset calibration followed by an internal gain calibration. If offset and gain calibrations are performed independently, offset calibration must be performed prior to gain calibration. During an internal offset calibration, the ADC inputs are connected internally to AGND. For an internal gain calibration, the ADC inputs are connected internally to a full-scale Voltage that is equal to the selected Voltage reference divided by the PGA gain.

### 5.2.2. System Calibration

System calibration is performed using voltages which are applied to the ADC inputs. There are two system calibration options: offset calibration and gain calibration. For accurate calibration results, offset calibration must be performed prior to gain calibration. During a system offset calibration, the ADC inputs should be connected to a “zero” value. During a system gain calibration, the ADC inputs should be connected to the positive full-scale value for the current PGA gain setting.

### 5.2.3. Calibration Coefficient Storage

The calibration results for offset and gain are each 24-bits long. The calibration results are stored in SFRs that are both readable and writeable from software. This enables factory calibrations, as well as manual modification of the offset and gain parameters. The offset calibration results are stored as a two’s complement, 24-bit number in the ADC0COH, ADC0COM, and ADC0COL registers. The mapping of the offset register is shown in Figure 5.3. The gain calibration results are stored as a fixed-point, 24-bit number in the ADC0CGH, ADC0CGM, and ADC0CGL registers. The mapping of the gain register is shown in Figure 5.4.

The offset calibration value adjusts the zero point of the ADC's transfer function. It is stored as a two's complement, 24-bit number. An offset calibration which results in a full-scale positive (0x7FFFFF) or full-scale negative (0x800000) result will cause an ADC error condition.

The Offset Calibration results are stored in registers ADC0COH, ADC0COM, and ADC0COL. The weighting of the bits in the offset register (in LSBs) are shown below:

| 24-bit ADC (C8051F350/1) | | | | | | | | | | | | | | | | | | | | | | | |
|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|
| ADC0COH | | | | | | | | ADC0COM | | | | | | | | ADC0COL | | | | | | | |
| MSB | 22 | 21 | 20 | 19 | 18 | 17 | 16 | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | LSB |
| $-2^{23}$ | $2^{22}$ | $2^{21}$ | $2^{20}$ | $2^{19}$ | $2^{18}$ | $2^{17}$ | $2^{16}$ | $2^{15}$ | $2^{14}$ | $2^{13}$ | $2^{12}$ | $2^{11}$ | $2^{10}$ | $2^{9}$ | $2^{8}$ | $2^{7}$ | $2^{6}$ | $2^{5}$ | $2^{4}$ | $2^{3}$ | $2^{2}$ | $2^{1}$ | $2^{0}$ |

| 16-bit ADC (C8051F352/3) | | | | | | | | | | | | | | | | | | | | | | | |
|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|
| ADC0COH | | | | | | | | ADC0COM | | | | | | | | ADC0COL | | | | | | | |
| MSB | 22 | 21 | 20 | 19 | 18 | 17 | 16 | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | LSB |
| $-2^{15}$ | $2^{14}$ | $2^{13}$ | $2^{12}$ | $2^{11}$ | $2^{10}$ | $2^{9}$ | $2^{8}$ | $2^{7}$ | $2^{6}$ | $2^{5}$ | $2^{4}$ | $2^{3}$ | $2^{2}$ | $2^{1}$ | $2^{0}$ | $2^{-1}$ | $2^{-2}$ | $2^{-3}$ | $2^{-4}$ | $2^{-5}$ | $2^{-6}$ | $2^{-7}$ | $2^{-8}$ |

**Figure 5.3. ADC0 Offset Calibration Register Coding**

The gain calibration value adjusts the slope of the ADC's transfer function. The gain calibration regsiter can range from 0 to $2 - 2^{-23}$. A gain calibration which results in either of these extremes will cause an ADC error condition.

The Gain Calibration results are stored in registers ADC0CGH, ADC0CGM, and ADC0CGL, as follows:

| ADC0CGH | | | | | | | | ADC0CGM | | | | | | | | ADC0CGL | | | | | | | |
|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|
| MSB | 22 | 21 | 20 | 19 | 18 | 17 | 16 | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | LSB |
| $2^{0}$ | $2^{-1}$ | $2^{-2}$ | $2^{-3}$ | $2^{-4}$ | $2^{-5}$ | $2^{-6}$ | $2^{-7}$ | $2^{-8}$ | $2^{-9}$ | $2^{-10}$ | $2^{-11}$ | $2^{-12}$ | $2^{-13}$ | $2^{-14}$ | $2^{-15}$ | $2^{-16}$ | $2^{-17}$ | $2^{-18}$ | $2^{-19}$ | $2^{-20}$ | $2^{-21}$ | $2^{-22}$ | $2^{-23}$ |

Example Decoding for Gain Register setting of 0x940000 (10010100 00000000 00000000b):

Slope Adjustment = $2^{0} + 2^{-3} + 2^{-5} = 1.0 + 0.125 + 0.03125 = 1.15625$

**Figure 5.4. ADC0 Gain Calibration Register Coding**

## 5.3. Performing Conversions

The ADC offers two conversion modes: Single Conversion, and Continuous Conversion. In single conversion mode, a single conversion result is produced for each of the filters (SINC3 and Fast). In continuous conversion mode, the ADC will perform back-to-back conversions until the ADC mode is changed. Procedures for single and continuous conversion modes are detailed in the sections below.

### 5.3.1. Single Conversions

A single conversion is initiated by writing the ADC System Mode bits (AD0SM) to the “Single Conversion” option. Single conversion mode instructs the ADC to gather enough information to produce a result for the filter that is selected by the AD0ISEL bit. During the conversion, the AD0BUSY flag will be set to ‘1’. The Fast filter results will be available after one period of the ADC’s conversion cycle (determined by the modulator clock and the decimation ratio). The SINC3 filter results will be available after three periods of the ADC’s conversion cycle. The AD0ISEL bit in register ADC0CF determines when the end-of-conversion interrupt will occur, and return the ADC to Idle mode. If the AD0ISEL bit is set to ‘1’, the AD0INT bit will be set to ‘1’ when the Fast filter results are available. If the AD0ISEL bit is cleared to ‘0’, the AD0INT bit will be set to ‘1’ when the SINC3 filter results are available. The AD0SM bits will return to idle mode and the AD0BUSY bit will be cleared to ‘0’ when the selected filter is finished. When using the SINC3 filter, a valid result will also be output by the Fast filter. When using the Fast filter in single-conversion mode, the SINC3 filter results will not be accurate.

### 5.3.2. Continuous Conversions

Continuous conversions are initiated by writing the ADC System Mode bits (AD0SM) to the “Continuous Conversion” option. In continous conversion mode, the ADC will start a new conversion as soon as each conversion is completed. During the conversions, the AD0BUSY flag will be set to ‘1’. The Fast filter results will be available after one period of the ADC’s conversion cycle, and on every conversion cycle thereafter (determined by the modulator clock and the decimation ratio). The first SINC3 filter result will be available after three periods of the ADC’s conversion cycle, and subsequent SINC3 conversion results will be available at the end of every conversion cycle thereafter. The AD0ISEL bit in register ADC0CF determines when the end-of-conversion interrupts will occur. If the AD0ISEL bit is cleared to ‘0’, the AD0INT bit will be set to ‘1’ when SINC3 filter results are available. If the AD0ISEL bit is set to ‘1’, the AD0INT bit will be set to ‘1’ when Fast filter results are available. Regardless of the setting of the AD0ISEL bit, both filters will update their results registers when new results are available. To stop conversions and exit from continuous conversion mode, the AD0SM bits should be written to Idle mode.

### 5.3.3. ADC Output

The ADC’s two filters each have their own output data registers. The SINC3 filter results are stored in the ADC0H, ADC0M, and ADC0L registers, while the Fast filter results are stored in the ADC0FH, ADC0FM, and ADC0FL registers. The ADC output can be configured for Unipolar or Bipolar mode using the AD0POL bit in register ADC0CN. Decoding of the ADC output words are shown in Table 5.1 and Table 5.2. The SINC3 filter uses information from the past three conversion cycles to produce an ADC output. The Fast filter uses information from only the current conversion cycle to produce an ADC output. The fast filter reacts more quickly to changes on the analog input, while the SINC3 filter produces lower-noise results.

**Table 5.1. ADC0 Unipolar Output Word Coding (AD0POL = 0)**

| Input Voltage* (AIN+ – AIN–) | 24-bit Output Word (C8051F350/1) | 16-bit Output Word (C8051F352/3) |
|---|---|---|
| VREF – 1 LSB | 0xFFFFFF | 0xFFFF |
| VREF / 2 | 0x800000 | 0x8000 |
| +1 LSB | 0x000001 | 0x0001 |
| 0 | 0x000000 | 0x0000 |
| ***Note:** Input Voltage is voltage at ADC inputs after amplification by the PGA. | | |

**Table 5.2. ADC0 Bipolar Output Word Coding (AD0POL = 1)**

| Input Voltage* (AIN+ – AIN–) | 24-bit Output Word (C8051F350/1) | 16-bit Output Word (C8051F352/3) |
|---|---|---|
| VREF – 1 LSB | 0x7FFFFF | 0x7FFF |
| VREF / 2 | 0x400000 | 0x4000 |
| +1 LSB | 0x000001 | 0x0001 |
| 0 | 0x000000 | 0x0000 |
| –1 LSB | 0xFFFFFF | 0xFFFF |
| –VREF / 2 | 0xC00000 | 0xC000 |
| –VREF | 0x800000 | 0x8000 |
| ***Note:** Input Voltage is voltage at ADC inputs after amplification by the PGA. | | |

### 5.3.1. Error Conditions

Any errors during a conversion or calibration are indicated using bits in the ADC0STA register. The AD0S3C flag will be set to ‘1’ if there is a SINC3 filter clip during the conversion. Likewise, the AD0FFC flag will be set to ‘1’ if there is a Fast filter clip during the conversion. A filter clip occurs whenever an internal filter register overflows during a conversion. The AD0OVR flag will be set to ‘1’ if an ADC overrun condition occurs. An overrun occurs if the end of a conversion is reached while the AD0INT flag is still set to ‘1’ from the previous conversion. If the data registers have not been read, the new data values will be updated, and the previous conversion will be lost. The general AD0ERR flag indicates that an AD0S3C, AD0FFC, or AD0OVR error condition has occurred, or that a calibration resulted in a value that was beyond the limits of the offset or gain register. The data output registers are updated at the end of every conversion regardless of whether or not an error occurs.

## 5.4. Offset DAC

An 8-bit offset DAC is included, which can be used for offset correction up to approximately ±1/2 of the ADC’s input range on any PGA gain setting. The ADC0DAC register (SFR Definition 5.7) controls the offset DAC voltage. The register is decoded as a signed binary word. The MSB (bit 7) determines the sign of the DAC magnitude (0 = positive, 1 = negative), and the remaining seven bits (bits 6–0) determine the magnitude. Each LSB of the offset DAC is equivalent to approximately 0.4% of the ADC’s input span. A write to the ADC0DAC register initiates a change on the offset DAC output.

## 5.5. Burnout Current Sources

The burnout current sources can be used to detect an open circuit or short circuit at the ADC inputs. The burnout current sources are enabled by setting the AD0BCE bit in register ADC0CN to ‘1’ (SFR Definition 5.1). The positive-channel burnout current source sources approximately 2 µA on AIN+, and the negative-channel burnout current sinks approximately 2 µA on AIN–. If an open circuit exists between AIN+ and AIN– when the burnout current sources are enabled, the ADC will read a full scale positive value. If a short-circuit exists between AIN+ and AIN– when the burnout current sources are enabled, the ADC will read a value near zero. The burnout current sources should be disabled during normal ADC measurements.

## SFR Definition 5.1. ADC0CN: ADC0 Control

| R | R | R | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| — | — | — | AD0POL | AD0BCE | AD0GN | | | 00010000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xF4

Bits 7–5: Unused: Read = 000b, Write = don't care.
Bit 4: AD0POL: ADC0 Polarity.
0: ADC operates in Unipolar mode (straight binary result).
1: ADC operates in Bipolar mode (2's compliment result).
Bit 3: AD0BCE: ADC0 Burnout Current Source Enable.
0: ADC Burnout current sources disabled.
1: ADC Burnout current sources enabled.
Bits 2:0 AD0GN: ADC0 Programmable Gain Setting.
000: PGA Gain = 1.
001: PGA Gain = 2.
010: PGA Gain = 4.
011: PGA Gain = 8.
100: PGA Gain = 16.
101: PGA Gain = 32.
110: PGA Gain = 64.
111: PGA Gain = 128.

**This SFR can only be modified when ADC0 is in IDLE mode.**

## SFR Definition 5.2. ADC0CF: ADC0 Configuration

| R | R | R | R/W | R | R/W | R | R | Reset Value |
|---|---|---|---|---|---|---|---|---|
| — | — | — | AD0ISEL | — | AD0VREF | — | — | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xFB

Bits 7–5: Unused: Read = 000b, Write = don't care.
Bit 4: AD0ISEL: ADC0 Interrupt Source Select.
This bit selects which filter completion will set the AD0INT interrupt flag.
0: SINC3 filter.
1: FAST filter.
Bit 3: Unused: Read = 0b, Write = don't care.
Bit 2: AD0VREF: ADC0 VREF Source Select.
0: ADC0 uses the internal VREF (2.5 V). Setting this bit to '0' enables the internal Voltage Reference.
1: ADC0 uses an external VREF.
Bits 1–0: Unused: Read = 00b, Write = don't care.

**This SFR can only be modified when ADC0 is in IDLE mode.**

## SFR Definition 5.3. ADC0MD: ADC0 Mode

| R/W | R | R/W | R/W | R | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| AD0EN | — | Reserved | Reserved | — | AD0SM | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xF3

Bit 7: AD0EN: ADC0 Enable Bit.
0: ADC0 Disabled. ADC is in low-power shutdown.
1: ADC0 Enabled. ADC is active and ready to perform calibrations or conversions.
Note: Disabling the ADC automatically resets the AD0SM bits back to the "Idle" state.

Bit 6: Unused: Read = 0b, Write = don't care.

Bits 5–4: RESERVED: Must Write to 00b.

Bit 3: Unused: Read = 0b, Write = don't care.

Bits 2–0: AD0SM: ADC0 System Mode Select.
These bits define the operating mode for the ADC. They are used to initiate all ADC conversion and calibration cycles.
000: Idle
001: Full Internal Calibration (offset and gain).
010: Single Conversion.
011: Continuous Conversion.
100: Internal Offset Calibration.
101: Internal Gain Calibration.
110: System Offset Calibration.
111: System Gain Calibration.

**Note: Any system mode change by the user during a conversion or calibration will terminate the operation, and corrupt the result. To write to many of the other ADC registers, the AD0SM bits must be set to IDLE mode (000b).**

**Note: During an ADC conversion or calibration, the AD0SM bits may return intermediate values if they are read. It is not recommended to use these bits as indicator of the ADC status. Only the ADC0STA register should be used as indicator of the ADC status.**

## SFR Definition 5.4. ADC0CLK: ADC0 Modulator Clock Divisor

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| ADC0CLK | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xF7

Bits 7–0: ADC0CLK: ADC0 Modulator Clock Divisor.
This register establishes the Modulator Clock (MDCLK), by dividing down the system clock (SYSCLK). The input signal is sampled by the modulator at a frequency of MDCLK / 128. For optimal performance, the divider should be chosen such that the modulator clock is equal to 2.4576 MHz (modulator sampling rate = 19.2 kHz).

The system clock is divided according to the equation:

**MDCLK** = SYSCLK / (ADC0CLK + 1)

**Note: The Modulator Sampling Rate is not the ADC Output Word Rate. See Section 5.1.4 for details.**

## SFR Definition 5.5. ADC0DECH: ADC0 Decimation Ratio Register High Byte

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| — | — | — | — | — | DECI10 | DECI9 | DECI8 | 00000111 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0x9B

Bits 7–3: Unused: Read = 00000b, Write = don't care.
Bits 2–0: DECI[10:8]: ADC0 Decimation Ratio Register, Bits 10–8.
This register contains the high bits of the 11-bit ADC Decimation Ratio. The decimation ratio determines the output word rate of ADC0, based on the Modulator Clock (MDCLK). See the ADC0DECL register description for more information.

**This SFR can only be modified when ADC0 is in IDLE mode.**

## SFR Definition 5.6. ADC0DECL: ADC0 Decimation Ratio Register Low Byte

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| DECI7 | DECI6 | DECI5 | DECI4 | DECI3 | DECI2 | DECI1 | DECI0 | 11111111 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0x9A

Bits 7–0: DECI[7:0]: ADC0 Decimation Ratio Register, Bits 7–0.
This register contains the low byte of the 11-bit ADC Decimation Ratio. The decimation ratio determines the number of modulator input samples used to generate a single output word from the ADC.

The ADC0 decimation ratio is defined as:

**Decimation Ratio** = DECI[10:0] + 1

The corresponding sampling period and output word rate of ADC0 is:

**ADC0 Conversion Period** = [(DECI[10:0] + 1) * 128] / MDCLK

**ADC0 Output Word Rate** = MDCLK / [128 * (DECI[10:0] + 1)]

The minimum decimation ratio setting is 20. Any register setting below 19 will automatically be interpreted as 19.
**Important: When using the fast filter, the decimation ratio must be divisible by 8 (DECI[2:0] = 111b)**.

**This SFR can only be modified when ADC0 is in IDLE mode.**

## SFR Definition 5.7. ADC0DAC: ADC0 Offset DAC

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| ADC0DAC | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xBF

Bits 7–0: ADC0DAC: ADC0 PGA Offset DAC Magnitude.
This register determines the ADC0 Offset DAC Magnitude. The value in the offset DAC is a signed-magnitude representation. Bit 7 represents the sign value (0 = positive, 1 = negative), while Bits 6–0 represent the magnitude.

**This SFR can only be modified when ADC0 is in IDLE mode.**

## SFR Definition 5.8. ADC0BUF: ADC0 Input Buffer Control

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| AD0BPHE | AD0BPLE | AD0BPS | | AD0BNHE | AD0BNLE | AD0BNS | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xBD

Bit 7: AD0BPHE: Positive Channel High Buffer Enable.
0: Positive Channel High Input Buffer Disabled.
1: Positive Channel High Input Buffer Enabled.

Bit 6: AD0BPLE: Positive Channel Low Enable.
0: Positive Channel Low Input Buffer Disabled.
1: Positive Channel Low Input Buffer Enabled.

Bits 5–4: AD0BPS: Positive Channel Input Selection.
00 = Bypass Input Buffer (default).
01 = Select Low Input Buffer Range.
10 = Select High Input Buffer Range.
11 = Reserved.

Bit 3: AD0BNHE: Negative Channel High Buffer Enable.
0: Negative Channel High Input Buffer Disabled.
1: Negative Channel High Input Buffer Enabled.

Bit 2: AD0BNLE: Negative Channel Low Enable.
0: Negative Channel Low Input Buffer Disabled.
1: Negative Channel Low Input Buffer Enabled.

Bits 1–0: AD0BNS: Negative Channel Input Selection.
00 = Bypass Input Buffer (default).
01 = Select Low Input Buffer Range.
10 = Select High Input Buffer Range.
11 = Reserved.

**This SFR can only be modified when ADC0 is in IDLE mode.**

## SFR Definition 5.9. ADC0STA: ADC0 Status

| R | R | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| AD0BUSY | AD0CBSY | AD0INT | AD0S3C | AD0FFC | AD0CALC | AD0ERR | AD0OVR | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | Bit Addressable |

SFR Address: 0xE8

Bit 7: AD0BUSY: ADC0 Conversion In-Progress Flag.
0: ADC0 is not performing conversions.
1: ADC0 conversion in progress.

Bit 6: AD0CBSY: ADC0 Calibration In-Progress Flag.
0: ADC0 is not performing a calibration.
1: ADC0 calibration in progress.

Bit 5: AD0INT: ADC0 Conversion Complete Interrupt Flag.
This flag must be cleared in software.
0: ADC0 has not completed a conversion since the last time this flag was cleared.
1: ADC0 conversion is complete.

Bit 4: AD0S3C: ADC0 SINC3 Filter Clip Flag.
This error flag indicates that a clip has occurred in the SINC3 filter during the conversion process.
0: ADC0 SINC3 filter clip did not occur.
1: ADC0 SINC3 filter clip occurred during conversion.

Bit 3: AD0FFC: ADC0 Fast Filter Clip Flag.
This error flag indicates that a clip has occurred in the fast filter during the conversion process.
0: ADC0 fast filter clip did not occur.
1: ADC0 fast filter clip occurred during conversion.

Bit 2: AD0CALC: ADC0 Calibration Complete Flag.
0: ADC0 calibration not complete.
1: ADC0 calibration complete.

Bit 1: AD0ERR: ADC0 Error Flag
This bit is set by hardware under the following conditions:
1) A conversion cycle produced an AD0OVR, AD0S3C, or AD0FFC error.
2) A calibration cycle produced a result that is beyond the limits of the offset or gain register.
0: ADC0 error did not occur.
1: ADC0 error occurred.

Bit 0: AD0OVR: ADC0 Overrun Flag
This error flag indicates an overrun condition.
0: ADC0 Overrun did not occur.
1: ADC0 Overrun occurred.

### SFR Definition 5.10. ADC0COH: ADC0 Offset Calibration Register High Byte

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| OCAL23 | OCAL22 | OCAL21 | OCAL20 | OCAL19 | OCAL18 | OCAL17 | OCAL16 | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xBC

Bits 7–0: OCAL[23:16]: ADC0 Offset Calibration Register High Byte.
This register contains the high byte of the 24-bit ADC Offset Calibration Value.

### SFR Definition 5.11. ADC0COM: ADC0 Offset Calibration Register Middle Byte

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| OCAL15 | OCAL14 | OCAL13 | OCAL12 | OCAL11 | OCAL10 | OCAL9 | OCAL8 | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xBB

Bits 7–0: OCAL[15:8]: ADC0 Offset Calibration Register Middle Byte.
This register contains the middle byte of the 24-bit ADC Offset Calibration Value.

### SFR Definition 5.12. ADC0COL: ADC0 Offset Calibration Register Low Byte

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| OCAL7 | OCAL6 | OCAL5 | OCAL4 | OCAL3 | OCAL2 | OCAL1 | OCAL0 | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xBA

Bits 7–0: OCAL[7:0]: ADC0 Offset Calibration Register Low Byte.
This register contains the low byte of the 24-bit ADC Offset Calibration Value.

### SFR Definition 5.13. ADC0CGH: ADC0 Gain Calibration Register High Byte

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| GCAL23 | GCAL22 | GCAL21 | GCAL20 | GCAL19 | GCAL18 | GCAL17 | GCAL16 | Variable |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | 10000000 |

SFR Address: 0xAD

Bits 7–0: GCAL[23:16]: ADC0 Gain Calibration Register High Byte.
This register contains the high byte of the 24-bit ADC Gain Calibration Value.

### SFR Definition 5.14. ADC0CGM: ADC0 Gain Calibration Register Middle Byte

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| GCAL15 | GCAL14 | GCAL13 | GCAL12 | GCAL11 | GCAL10 | GCAL9 | GCAL8 | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xAC

Bits 7–0: GCAL[15:8]: ADC0 Gain Calibration Register Middle Byte.
This register contains the middle byte of the 24-bit ADC Gain Calibration Value.

### SFR Definition 5.15. ADC0CGL: ADC0 Gain Calibration Register Low Byte

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| GCAL7 | GCAL6 | GCAL5 | GCAL4 | GCAL3 | GCAL2 | GCAL1 | GCAL0 | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xAB

Bits 7–0: GCAL[7:0]: ADC0 Gain Calibration Register Low Byte.
This register contains the low byte of the 24-bit ADC Gain Calibration Value.

## SFR Definition 5.16. ADC0H: ADC0 Conversion Register (SINC3 Filter) High Byte

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| ADC0H | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xC5

Bits 7–0: ADC0H: ADC0 Conversion Register (SINC3 Filter) High Byte.
**C8051F350/1**: This register contains bits 23–16 of the 24-bit ADC SINC3 filter conversion result.
**C8051F352/3**: This register contains bits 15–8 of the 16-bit ADC SINC3 filter conversion result.

## SFR Definition 5.17. ADC0M: ADC0 Conversion Register (SINC3 Filter) Middle Byte

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| ADC0M | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xC4

Bits 7–0: ADC0M: ADC0 Conversion Register (SINC3 Filter) Middle Byte.
**C8051F350/1**: This register contains bits 15–8 of the 24-bit ADC SINC3 filter conversion result.
**C8051F352/3**: This register contains bits 7–0 of the 16-bit ADC SINC3 filter conversion result.

## SFR Definition 5.18. ADC0L: ADC0 Conversion Register (SINC3 Filter) Low Byte

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| ADC0L | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xC3

Bits 7–0: ADC0L: ADC0 Conversion Register (SINC3 Filter) Low Byte.
**C8051F350/1**: This register contains bits 7–0 of the 24-bit ADC SINC3 filter conversion result.
**C8051F352/3**: This register contains all zeros (00000000b).

## SFR Definition 5.19. ADC0FH: ADC0 Conversion Register (Fast Filter) High Byte

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| ADC0FH | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xFE

Bits 7–0: ADC0FH: ADC0 Conversion Register (Fast Filter) High Byte.
**C8051F350/1**: This register contains bits 23–16 of the 24-bit ADC fast filter conversion result.
**C8051F352/3**: This register contains bits 15–8 of the 16-bit ADC fast filter conversion result.

## SFR Definition 5.20. ADC0FM: ADC0 Conversion Register (Fast Filter) Middle Byte

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| ADC0FM | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xFD

Bits 7–0: ADC0FM: ADC0 Conversion Register (Fast Filter) Middle Byte.
**C8051F350/1**: This register contains bits 15–8 of the 24-bit ADC fast filter conversion result.
**C8051F352/3**: This register contains bits 7–0 of the 16-bit ADC fast filter conversion result.

## SFR Definition 5.21. ADC0FL: ADC0 Conversion Register (Fast Filter) Low Byte

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| ADC0FL | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xFC

Bits 7–0: ADC0FL: ADC0 Conversion Register (Fast Filter) Low Byte.
**C8051F350/1**: This register contains bits 7–0 of the 24-bit ADC fast filter conversion result.
**C8051F352/3**: This register contains all zeros (00000000b).

## 5.6. Analog Multiplexer

ADC0 includes analog multiplexer circuitry with independent selection capability for the AIN+ and AIN– inputs. Each input can be connected to one of ten possible input sources: AIN0.0 though AIN0.7, AGND, or the on-chip temperature sensor circuitry (Figure 5.5). The ADC0MUX register (SFR Definition 5.22) controls the input mux selection for both input channels. The multiplexer configuration allows for measurement of single-ended or differential signals. A single-ended measurement can be performed by connecting one of the ADC inputs to AGND. Additionally, the temperature sensor can be measured in single-ended or differential mode. The temperature sensor is automatically enabled when it is selected with the ADC multiplexer. See Section “8. Temperature Sensor’ on page 77 for more details on the temperature sensor.

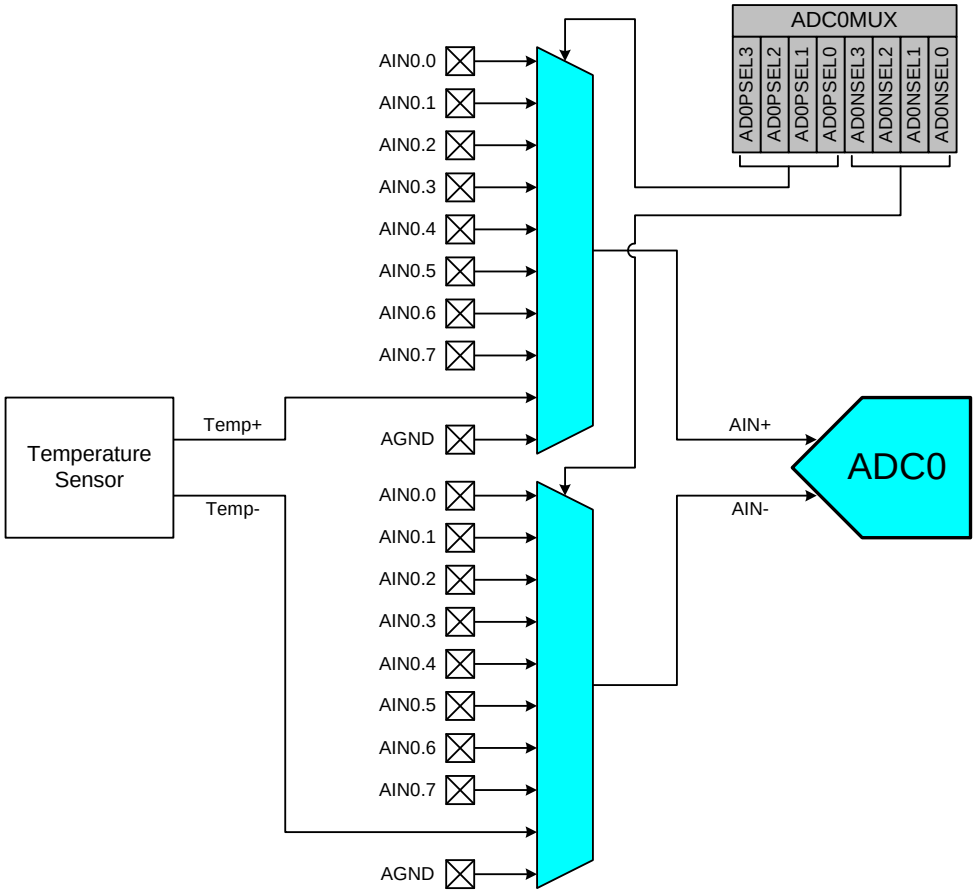



**Figure 5.5. ADC0 Multiplexer Connections**

## SFR Definition 5.22. ADC0MUX: ADC0 Analog Multiplexer Control

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| AD0PSEL | | | | AD0NSEL | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xC6

Bits 7–4: AD0PSEL: ADC0 Positive Multiplexer Channel Select.
0000 = AIN0.0
0001 = AIN0.1
0010 = AIN0.2
0011 = AIN0.3
0100 = AIN0.4
0101 = AIN0.5
0110 = AIN0.6
0111 = AIN0.7
1111 = Temperature Sensor
All Other Settings = AGND

Bits 3–0: AD0NSEL: ADC0 Negative Multiplexer Channel Select.
0000 = AIN0.0
0001 = AIN0.1
0010 = AIN0.2
0011 = AIN0.3
0100 = AIN0.4
0101 = AIN0.5
0110 = AIN0.6
0111 = AIN0.7
1111 = Temperature Sensor
All Other Settings = AGND

**This SFR should only be modified when ADC0 is in IDLE mode.**

## Table 5.3. ADC0 Electrical Characteristics

$V_{DD}$ = AV+ = 3.0 V, VREF = 2.5 V External, PGA Gain = 1, MDCLK = 2.4576 MHz,
Decimation Ratio = 1920, –40 to +85 °C unless otherwise noted.

| Parameter | Conditions | Min | Typ | Max | Units |
|---|---|---|---|---|---|
| **24-bit ADC (C8051F350/1)** | | | | | |
| Resolution | | | 24 | | bits |
| No Missing Codes | | | 24 | | bits |
| **16-bit ADC (C8051F352/3)** | | | | | |
| Resolution | | | 16 | | bits |
| No Missing Codes | | | 16 | | bits |
| **All Devices** | | | | | |
| Integral Nonlinearity | | — | — | ±15 | ppm FS |
| Offset Error (Calibrated) | | — | ±5 | — | ppm |
| Offset Drift vs. Temperature | | — | 10 | — | nV/ °C |
| Gain Error (Calibrated) | | — | ±0.002 | — | % |
| Gain Drift vs. Temperature | | — | ±0.5 | — | ppm/ °C |
| Modulator Clock (MDCLK) | | — | 2.4576 | — | MHz |
| Modulator Sampling Frequency | | | MDCLK/128 | | Hz |
| Output Word Rate | | — | — | 1000 | sps |
| **Analog Inputs** | | | | | |
| Analog Input Voltage Range (AIN+ – AIN–) | PGA Gain = 1, Bipolar<br>PGA Gain = 1, Unipolar | –VREF<br>0 | —<br>— | +VREF<br>+VREF | V |
| Absolute Voltage on AIN+ or AIN– pin with respect to AGND | Input Buffers OFF | 0 | — | AV+ | V |
| Input Current | Input Buffer ON | — | ± 1.5 | 30 | nA |
| Input Impedance | Input Buffer OFF, Gain = 1 | — | 7 | — | MΩ |
| Common Mode Rejection Ratio | DC<br>50/60 Hz | 95 | 110<br>100 | —<br>— | dB<br>dB |
| **Input Buffers** | | | | | |
| High Buffer Input Range with respect to AGND | PGA Gain = 1, 2, 4, or 8<br>PGA Gain = 16<br>PGA Gain = 32<br>PGA Gain = 64 or 128 | 1.4<br>1.45<br>1.5<br>1.6 | —<br>—<br>—<br>— | AV+ – 0.1<br>AV+ – 0.15<br>AV+ – 0.2<br>AV+ – 0.25 | V<br>V<br>V<br>V |
| Low Buffer Input Range with respect to AGND | PGA Gain = 1, 2, 4, or 8<br>PGA Gain = 16<br>PGA Gain = 32<br>PGA Gain = 64 or 128 | 0.1<br>0.15<br>0.2<br>0.25 | —<br>—<br>—<br>— | AV+ – 1.4<br>AV+ – 1.45<br>AV+ – 1.5<br>AV+ – 1.6 | V<br>V<br>V<br>V |
| **Burnout Current Sources** | | | | | |
| Positive (AIN+) Channel Current | VREF = 2.5 V | 0.9 | 2 | 2.9 | µA |
| Negative (AIN–) Channel Current | VREF = 2.5 V | –0.9 | –2 | –2.9 | µA |

## Table 5.3. ADC0 Electrical Characteristics (Continued)

$V_{DD}$ = AV+ = 3.0 V, VREF = 2.5 V External, PGA Gain = 1, MDCLK = 2.4576 MHz,
Decimation Ratio = 1920, –40 to +85 °C unless otherwise noted.

| Parameter | Conditions | Min | Typ | Max | Units |
|---|---|---|---|---|---|
| **Power Specifications** | | | | | |
| AV+ Supply Current to ADC0 | | — | 230 | 650 | µA |
| AV+ Supply Current to Input Buffers (Each Enabled Buffer) | | — | 90 | 125 | µA |
| Power Supply Rejection | | 80 | — | — | dB |

## Table 5.4. ADC0 SINC3 Filter Typical RMS Noise (µV)

| Decimation Ratio | Output Word Rate* | PGA Gain Setting | | | | | | | |
|---|---|---|---|---|---|---|---|---|---|
| | | 1 | 2 | 4 | 8 | 16 | 32 | 64 | 128 |
| 1920 | 10 Hz | 2.38 | 1.23 | 0.68 | 0.41 | 0.24 | 0.16 | 0.12 | 0.11 |
| 768 | 25 Hz | 3.90 | 2.04 | 1.14 | 0.68 | 0.44 | 0.33 | 0.28 | 0.27 |
| 640 | 30 Hz | 4.50 | 2.39 | 1.31 | 0.81 | 0.54 | 0.42 | 0.36 | 0.36 |
| 384 | 50 Hz | 6.00 | 3.21 | 1.86 | 1.20 | 0.86 | 0.73 | 0.66 | 0.66 |
| 320 | 60 Hz | 7.26 | 3.96 | 2.32 | 1.51 | 1.11 | 0.97 | 0.89 | 0.89 |
| 192 | 100 Hz | 13.1 | 7.11 | 4.24 | 2.85 | 2.16 | 1.91 | 1.79 | 1.77 |
| 80 | 240 Hz | 93.2 | 47.7 | 24.8 | 13.9 | 9.34 | 7.61 | 6.97 | 6.67 |
| 40 | 480 Hz | 537 | 267 | 135 | 69.5 | 38.8 | 25.7 | 20.9 | 18.9 |
| 20 | 960 Hz | 2974 | 1586 | 771 | 379 | 196 | 108 | 70.0 | 45.4 |

***Note:** Output Word Rate assuming Modulator Clock frequency = 2.4576 MHz (sampling clock frequency = 19.2 kHz)

## Table 5.5. ADC0 SINC3 Filter Effective Resolution[1] in Unipolar Mode (bits)

| Decimation Ratio | Output Word Rate[2] | PGA Gain Setting | | | | | | | |
|---|---|---|---|---|---|---|---|---|---|
| | | 1 | 2 | 4 | 8 | 16 | 32 | 64 | 128 |
| 1920 | 10 Hz | 20.00 | 19.95 | 19.81 | 19.54 | 19.31 | 18.90 | 18.31 | 17.44 |
| 768 | 25 Hz | 19.29 | 19.22 | 19.06 | 18.81 | 18.44 | 17.85 | 17.09 | 16.14 |
| 640 | 30 Hz | 19.08 | 19.00 | 18.86 | 18.56 | 18.14 | 17.51 | 16.73 | 15.73 |
| 384 | 50 Hz | 18.67 | 18.57 | 18.36 | 17.99 | 17.47 | 16.71 | 15.85 | 14.85 |
| 320 | 60 Hz | 18.39 | 18.27 | 18.04 | 17.66 | 17.10 | 16.30 | 15.42 | 14.42 |
| 192 | 100 Hz | 17.54 | 17.42 | 17.17 | 16.74 | 16.14 | 15.32 | 14.41 | 13.43 |
| 80 | 240 Hz | 14.71 | 14.68 | 14.62 | 14.46 | 14.03 | 13.33 | 12.45 | 11.52 |
| 40 | 480 Hz | 12.18 | 12.19 | 12.18 | 12.13 | 11.98 | 11.57 | 10.87 | 10.01 |
| 20 | 960 Hz | 9.72 | 9.62 | 9.66 | 9.69 | 9.64 | 9.50 | 9.12 | 8.75 |

**Notes:**

1. *Effective resolution* = $\log_2\left(\frac{FullInputRange(V)}{RMS\ Noise(V)}\right)$

   where *Full Input Range* = $\frac{V_{REF}}{PGA\ Gain}$ in Unipolar mode and *RMS Noise* is obtained from Table 5.4.
2. Output Word Rate assuming Modular Clock frequency = 2.4576 MHz (sampling clock frequency = 19.2 kHz)

## Table 5.6. ADC0 SINC3 Filter Flicker-Free (Noise-Free) Resolution[1] in Unipolar Mode (bits)

| Decimation Ratio | Output Word Rate[2] | PGA Gain Setting | | | | | | | |
|---|---|---|---|---|---|---|---|---|---|
| | | 1 | 2 | 4 | 8 | 16 | 32 | 64 | 128 |
| 1920 | 10 Hz | 17.28 | 17.23 | 17.09 | 16.82 | 16.59 | 16.17 | 15.59 | 14.72 |
| 768 | 25 Hz | 16.57 | 16.50 | 16.34 | 16.09 | 15.72 | 15.13 | 14.37 | 13.42 |
| 640 | 30 Hz | 16.36 | 16.27 | 16.14 | 15.84 | 15.42 | 14.78 | 14.00 | 13.00 |
| 384 | 50 Hz | 15.95 | 15.85 | 15.64 | 15.27 | 14.75 | 13.99 | 13.13 | 12.13 |
| 320 | 60 Hz | 15.67 | 15.55 | 15.32 | 14.94 | 14.38 | 13.57 | 12.70 | 11.70 |
| 192 | 100 Hz | 14.82 | 14.70 | 14.45 | 14.02 | 13.42 | 12.60 | 11.69 | 10.71 |
| 80 | 240 Hz | 11.99 | 11.96 | 11.90 | 11.73 | 11.31 | 10.60 | 9.73 | 8.79 |
| 40 | 480 Hz | 9.46 | 9.47 | 9.45 | 9.41 | 9.25 | 8.85 | 8.15 | 7.29 |
| 20 | 960 Hz | 6.99 | 6.90 | 6.94 | 6.96 | 6.92 | 6.78 | 6.40 | 6.03 |

**Notes:**

1. *Flicker-free (Noise-free) Resolution* = $\log_2\left(\frac{FullInputRange(V)}{6.6 \times RMS\ Noise(V)}\right)$

   where *Full Input Range* = $\frac{V_{REF}}{PGA\ Gain}$ in Unipolar mode and *RMS Noise* is obtained from Table 5.4.
2. Output Word Rate assuming Modular Clock frequency = 2.4576 MHz (sampling clock frequency = 19.2 kHz)

### Table 5.7. ADC0 Fast Filter Typical RMS Noise (µV)

| Decimation Ratio | Output Word Rate* | PGA Gain Setting | | | | | | | |
|---|---|---|---|---|---|---|---|---|---|
| | | 1 | 2 | 4 | 8 | 16 | 32 | 64 | 128 |
| 1920 | 10 Hz | 4.84 | 2.68 | 1.55 | 1.03 | 0.75 | 0.61 | 0.56 | 0.58 |
| 768 | 25 Hz | 17.92 | 9.77 | 5.85 | 3.72 | 2.79 | 2.45 | 2.28 | 2.21 |
| 640 | 30 Hz | 29.98 | 14.84 | 7.81 | 5.39 | 3.89 | 3.27 | 3.19 | 3.03 |
| 384 | 50 Hz | 103.93 | 48.53 | 25.71 | 14.07 | 9.24 | 7.17 | 6.45 | 6.06 |
| 320 | 60 Hz | 171.12 | 89.87 | 42.99 | 23.05 | 13.81 | 10.33 | 9.00 | 8.52 |
| 192 | 100 Hz | 550.29 | 305.55 | 140.58 | 72.90 | 40.97 | 25.52 | 19.96 | 17.68 |

***Note:** Output Word Rate assuming Modulator Clock frequency = 2.4576 MHz (sampling clock frequency = 19.2 kHz)

### Table 5.8. ADC0 Fast Filter Effective Resolution[1] in Unipolar Mode (bits)

| Decimation Ratio | Output Word Rate[2] | PGA Gain Setting | | | | | | | |
|---|---|---|---|---|---|---|---|---|---|
| | | 1 | 2 | 4 | 8 | 16 | 32 | 64 | 128 |
| 1920 | 10 Hz | 18.98 | 18.83 | 18.62 | 18.21 | 17.67 | 16.97 | 16.09 | 15.04 |
| 768 | 25 Hz | 17.09 | 16.97 | 16.71 | 16.36 | 15.77 | 14.96 | 14.06 | 13.11 |
| 640 | 30 Hz | 16.35 | 16.36 | 16.29 | 15.82 | 15.29 | 14.54 | 13.58 | 12.65 |
| 384 | 50 Hz | 14.55 | 14.65 | 14.57 | 14.44 | 14.05 | 13.41 | 12.56 | 11.65 |
| 320 | 60 Hz | 13.83 | 13.76 | 13.83 | 13.73 | 13.47 | 12.88 | 12.08 | 11.16 |
| 192 | 100 Hz | 12.15 | 12.00 | 12.12 | 12.07 | 11.90 | 11.58 | 10.93 | 10.11 |

**Notes:**

1. *Effective Resolution* = $\log_2\left(\frac{FullInputRange(V)}{RMS\ Noise(V)}\right)$

   where *Full Input Range* = $\frac{V_{REF}}{PGA\ Gain}$ in Unipolar mode and *RMS Noise* is obtained from Table 5.7.
2. Output Word Rate assuming Modular Clock frequency = 2.4576 MHz (sampling clock frequency = 19.2 kHz)

**Table 5.9. ADC0 Fast Filter Flicker-Free (Noise-Free) Resolution[1] in Unipolar Mode (bits)**

| Decimation Ratio | Output Word Rate[2] | PGA Gain Setting | | | | | | | |
|---|---|---|---|---|---|---|---|---|---|
| | | 1 | 2 | 4 | 8 | 16 | 32 | 64 | 128 |
| 1920 | 10 Hz | 16.26 | 16.11 | 15.90 | 15.49 | 14.95 | 14.24 | 13.37 | 12.32 |
| 768 | 25 Hz | 14.37 | 14.24 | 13.98 | 13.64 | 13.05 | 12.24 | 11.34 | 10.39 |
| 640 | 30 Hz | 13.63 | 13.64 | 13.57 | 13.10 | 12.57 | 11.82 | 10.86 | 9.93 |
| 384 | 50 Hz | 11.83 | 11.93 | 11.85 | 11.72 | 11.32 | 10.69 | 9.84 | 8.93 |
| 320 | 60 Hz | 11.11 | 11.04 | 11.11 | 11.00 | 10.74 | 10.16 | 9.36 | 8.44 |
| 192 | 100 Hz | 9.43 | 9.28 | 9.40 | 9.34 | 9.17 | 8.86 | 8.21 | 7.39 |

**Notes:**

**1.** *Flicker-free (Noise-free) Resolution* = $\log_2\left(\frac{FullInputRange(V)}{6.6 \times RMS\ Noise(V)}\right)$

where *Full Input Range* = $\frac{V_{REF}}{PGA\ Gain}$ in Unipolar mode and *RMS Noise* is obtained from Table 5.7.

**2.** Output Word Rate assuming Modular Clock frequency = 2.4576 MHz (sampling clock frequency = 19.2 kHz)

## NOTES:

## 6. 8-Bit Current Mode DACS (IDA0 and IDA1)

The C8051F350/1/2/3 devices include two 8-bit current-mode Digital-to-Analog Converters (IDACs). The maximum current output of the IDACs can be adjusted for four different current settings; 0.25 mA, 0.5 mA, 1 mA, and 2 mA. The IDACs can be individually enabled or disabled using the enable bits in the corresponding IDAC Control Register (IDA0CN or IDA1CN). An internal bandgap bias generator is used to generate a reference current for the IDACs whenever they are enabled. IDAC updates can be performed on-demand, scheduled on a Timer overflow, or synchronized with an external pin edge. Figure 6.1 shows a block diagram of the IDAC circuitry.

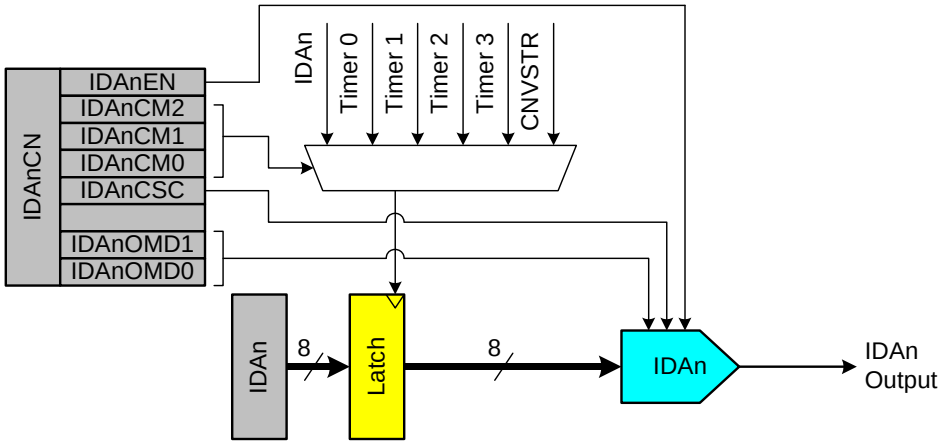



**Figure 6.1. IDAC Functional Block Diagram**

## 6.1. IDAC Output Scheduling

A flexible output update mechanism allows for seamless full-scale changes and supports jitter-free updates for waveform generation. Three update modes are provided, allowing IDAC output updates on a write to the IDAC's data register, on a Timer overflow, or on an external pin edge.

### 6.1.1. Update Output On-Demand

In its default mode (IDAnCN.[6:4] = '111') the IDAC output is updated "on-demand" with a write to the data register (IDAn). In this mode, data is immediately latched into the IDAC after a write to its data register.

### 6.1.2. Update Output Based on Timer Overflow

The IDAC output update can be scheduled on a Timer overflow. This feature is useful in systems where the IDAC is used to generate a waveform of a defined sampling rate, by eliminating the effects of variable interrupt latency and instruction execution on the timing of the IDAC output. When the IDAnCM bits (IDAnCN.[6:4]) are set to '000', '001', '010' or '011', writes to the IDAC data register (IDAn) are held until an associated Timer overflow event (Timer 0, Timer 1, Timer 2 or Timer 3, respectively) occurs, at which time the data register contents are copied to the IDAC input latch, allowing the IDAC output to change to the new value.

### 6.1.3. Update Output Based on CNVSTR Edge

The IDAC output can also be configured to update on a rising edge, falling edge, or both edges of the external CNVSTR signal. When the IDAnCM bits (IDAnCN.[6:4]) are set to '100', '101', or '110', writes to the IDAC data register (IDAn) are held until an edge occurs on the CNVSTR input pin. The particular setting of the IDAnCM bits determines whether the IDAC output is updated on rising, falling, or both edges of CNVSTR. When a corresponding edge occurs, the data register contents are copied to the IDAC input latch, allowing the IDAC output to change to the new value.

## 6.2. IDAC Output Mapping

The data word mapping for the IDAC is shown in Figure 6.2. The full-scale output current of the IDAC is selected using the IDAnOMD bits (IDAnCN[1:0]). By default, the IDAC is set to a full-scale output current of 0.25 mA. The IDAnOMD bits can also be configured to provide full-scale output currents of 0.5 mA, 1 mA, or 2 mA.

| IDAn Data Word (D7 – D0) | Output Current vs IDAnOMD bit setting | | | |
|---|---|---|---|---|
| | '11' (2 mA) | '10' (1 mA) | '01' (0.5 mA) | '00' (0.25 mA) |
| 0x00 | 0 mA | 0 mA | 0 mA | 0 mA |
| 0x01 | 1/256 x 2 mA | 1/256 x 1 mA | 1/256 x 0.5 mA | 1/256 x 0.25 mA |
| 0x80 | 128/256 x 2 mA | 128/256 x 1 mA | 128/256 x 0.5 mA | 128/256 x 0.25 mA |
| 0xFF | 255/256 x 2 mA | 255/256 x 1 mA | 255/256 x 0.5 mA | 255/256 x 0.25 mA |

**Figure 6.2. IDAC Data Word Mapping**

## SFR Definition 6.1. IDA0CN: IDA0 Control

| R/W | R/W | R/W | R/W | R/W | R | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| IDA0EN | IDA0CM | | | IDA0CSC | — | IDA0OMD | | 01110000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xB9

Bit 7: IDA0EN: IDA0 Enable.
0: IDA0 Disabled.
1: IDA0 Enabled.

Bits 6–4: IDA0CM[2:0]: IDA0 Update Source Select bits.
000: DAC output updates on Timer 0 overflow.
001: DAC output updates on Timer 1 overflow.
010: DAC output updates on Timer 2 overflow.
011: DAC output updates on Timer 3 overflow.
100: DAC output updates on rising edge of CNVSTR.
101: DAC output updates on falling edge of CNVSTR.
110: DAC output updates on any edge of CNVSTR.
111: DAC output updates on write to IDA0.

Bit 3: IDA0CSC: IDA0 Constant Supply Current.
0: Current draw on $V_{DD}$ is dependent on IDA0 Output Word.
1: Current draw on $V_{DD}$ is independent of IDA0 Output Word.

Bit 2: Unused. Read = 0b, Write = Don't Care.

Bits 1:0: IDA0OMD[1:0]: IDA0 Output Mode Select bits.
00: 0.25 mA full-scale output current.
01: 0.5 mA full-scale output current.
10: 1.0 mA full-scale output current.
11: 2.0 mA full-scale output current.

## SFR Definition 6.2. IDA0: IDA0 Data Word

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0x96

Bits 7–0: IDA0 Data Word Bits.
Bits 7–0 hold the 8-bit IDA0 Data Word.

## SFR Definition 6.3. IDA1CN: IDA1 Control

| R/W | R/W | R/W | R/W | R/W | R | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| IDA1EN | IDA1CM | | | IDA1CSC | — | IDA1OMD | | 01110000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xD7

Bit 7: IDA1EN: IDA1 Enable.
0: IDA1 Disabled.
1: IDA1 Enabled.

Bits 6–4: IDA1CM[2:0]: IDA1 Update Source Select bits.
000: DAC output updates on Timer 0 overflow.
001: DAC output updates on Timer 1 overflow.
010: DAC output updates on Timer 2 overflow.
011: DAC output updates on Timer 3 overflow.
100: DAC output updates on rising edge of CNVSTR.
101: DAC output updates on falling edge of CNVSTR.
110: DAC output updates on any edge of CNVSTR.
111: DAC output updates on write to IDA1.

Bit 3: IDA1CSC: IDA1 Constant Supply Current.
0: Current draw on $V_{DD}$ is dependent on IDA1 Output Word.
1: Current draw on $V_{DD}$ is independent of IDA1 Output Word.

Bit 2: Unused. Read = 0b, Write = Don't Care.

Bits 1:0: IDA1OMD[1:0]: IDA1 Output Mode Select bits.
00: 0.25 mA full-scale output current.
01: 0.5 mA full-scale output current.
10: 1.0 mA full-scale output current.
11: 2.0 mA full-scale output current.

## SFR Definition 6.4. IDA1: IDA1 Data Word

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xDD

Bits 7–0: IDA1 Data Word Bits.
Bits 7–0 hold the 8-bit IDA1 Data Word.

## 6.3. IDAC External Pin Connections

The IDA0 output is connected to P1.6, and the IDA1 output is connected to P1.7. When the enable bit for an IDAC (IDAnEN) is set to ‘0’, the IDAC output behaves as a normal GPIO pin. When the enable bit is set to ‘1’, the digital output drivers and weak pull-up for the IDAC pin are automatically disabled, and the pin is connected to the IDAC output. When using the IDACs, the IDAC pins should be skipped in the Crossbar by setting the corresponding PnSKIP bits to a ‘1’. Figure 6.3 shows the pin connections for IDA0 and IDA1.

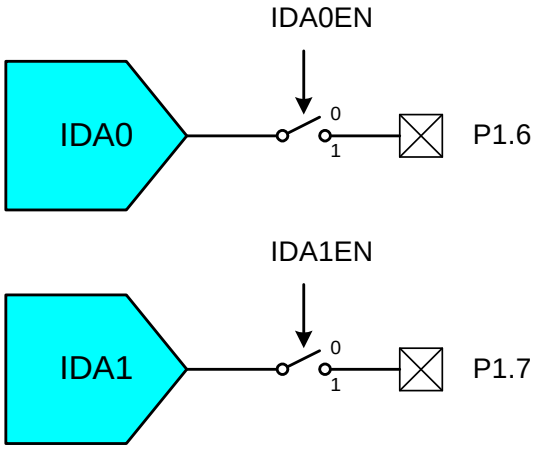



**Figure 6.3. IDAC Pin Connections**

## Table 6.1. IDAC Electrical Characteristics

–40 to +85 °C, $V_{DD}$ = 3.0 V Full-scale output current set to 2 mA unless otherwise specified.

| Parameter | Conditions | Min | Typ | Max | Units |
|---|---|---|---|---|---|
| **Static Performance** | | | | | |
| Resolution | | | 8 | | bits |
| Integral Nonlinearity | | — | ±0.5 | — | LSB |
| Differential Nonlinearity | Guaranteed Monotonic | — | ±0.5 | ±1 | LSB |
| Output Compliance Range | | — | — | $V_{DD}$ – 1.2 | V |
| Output Noise | $I_{OUT}$ = 2 mA; $R_{LOAD}$ = 100 Ω | — | 2 | — | nA/rtHz |
| Offset Error | | — | 0 | — | LSB |
| Full Scale Output | 2 mA Full Scale Setting | 1.6 | 2.0 | 2.4 | mA |
| Gain-Error Tempco | | — | 4 | — | µA/°C |
| VDD Power Supply Rejection Ratio | | — | 30 | — | µA/V |
| Output Capacitance | | — | 2 | — | pF |
| **Dynamic Performance** | | | | | |
| Output Settling Time to 1/2 LSB | | — | 5 | — | µs |
| Startup Time | | — | 5 | — | µs |
| Gain Variation From 2 mA range | 1 mA Full Scale Output Current<br>0.5 mA Full Scale Output Current<br>0.25 mA Full Scale Output Current | —<br>—<br>— | ± 1<br>± 1<br>± 1 | —<br>—<br>— | %<br>%<br>% |
| **Power Consumption** | | | | | |
| Power Supply Current (IDAnCSC = 0) | $I_{OUT}$ = Current sourced from IDAC | — | $I_{OUT}$ + 85 | $I_{OUT}$ + 130 | µA |
| Power Supply Current (IDAnCSC = 1) | 2 mA Full Scale Output Current<br>1 mA Full Scale Output Current<br>0.5 mA Full Scale Output Current<br>0.25 mA Full Scale Output Current | —<br>—<br>—<br>— | 2085<br>1085<br>585<br>335 | 2130<br>1130<br>630<br>380 | µA<br>µA<br>µA<br>µA |

# 7. Voltage Reference

There are two voltage reference options for the C8051F350/1/2/3 ADCs: the internal 2.5 V reference voltage, or an external reference voltage (see Figure 7.1). The AD0VREF bit in the ADC0CF register selects the reference source.

The internal voltage reference circuit consists of a 1.25 V, temperature stable bandgap voltage reference generator and a gain-of-two output buffer amplifier, to produce a 2.5 V voltage reference. When the internal voltage reference is used, it is driven out on the VREF+ pin. In this configuration, the VREF– must be connected to the AGND pin external to the device. See Section “24. Revision Specific Behavior’ on page 227 for more information. The internal voltage reference is enabled by setting the AD0EN bit in register ADC0MD to ‘1’ and clearing the AD0VREF bit in register ADC0CF to ‘0’ (See Section “5. 24 or 16-Bit Analog to Digital Converter (ADC0)’ on page 41). Electrical specifications for the internal voltage reference and bias generators are given in Table 7.1.

The internal oscillator bias generator is automatically enabled whenever the internal oscillator is enabled. For power requirement characterization, the BIASE bit in register REF0CN can also be used to enable the internal oscillator bias generator, when the oscillator itself is not enabled. Likewise, the REFBE bit in register REF0CN can be used to enable the internal bandgap generator, which is used by the ADC, the IDACs, the Clock Multiplier, and the internal Voltage Reference. The internal reference bias generator is automatically enabled whenever a peripheral requires it. See SFR Definition 7.1 for the REF0CN register description.

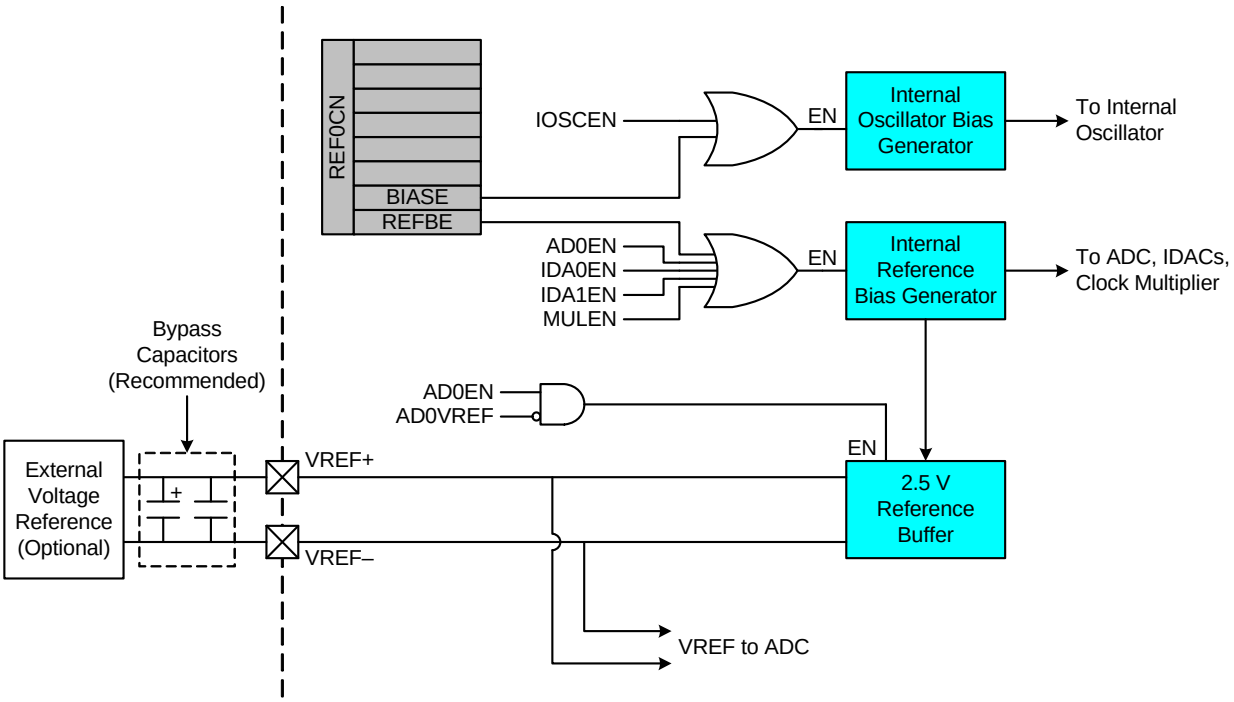



**Figure 7.1. Reference Circuitry Block Diagram**

## SFR Definition 7.1. REF0CN: Reference Control

| R | R | R | R | R | R | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| — | — | — | — | — | — | BIASE | REFBE | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xD1

NOTE: Modification of this register is not necessary in most applications. The appropriate circuitry is enabled when it is needed by a peripheral.

Bits7–2: Unused. Read = 000000b; Write = don't care.
Bit1: BIASE: Internal Oscillator Bias Enable.
This bit is ORed with the Internal Oscillator Enable bit to enable the internal oscillator bias generator.
0: Internal Oscillator Bias enable determined by Internal Oscillator Enable bit.
1: Internal Oscillator Bias Generator On.
Bit0: REFBE: Internal Reference Bias Enable Bit.
This bit is ORed with the Enable bits for ADC0, IDAC0, IDAC1, and the Clock Multiplier to enable the internal bandgap generator.
0: Internal Reference Bias enable determined by individual component.
1: Internal Reference Bias enabled.

## Table 7.1. Voltage Reference Electrical Characteristics

$V_{DD}$ = 3.0 V; –40 to +85 °C unless otherwise specified.

| Parameter | Conditions | Min | Typ | Max | Units |
|---|---|---|---|---|---|
| **Internal Reference** | | | | | |
| Output Voltage | 25 °C ambient | 2.35 | 2.45 | 2.50 | V |
| VREF Short-Circuit Current | To AGND<br>To AV+ | —<br>— | 9<br>20 | —<br>— | mA<br>µA |
| VREF Temperature Coefficient | | — | 15 | — | ppm/°C |
| Load Regulation | Load = 0 to 200 µA to AGND | — | 0.5 | — | ppm/µA |
| VREF Turn-on Time 1 (0.01%) | 4.7 µF tantalum, 0.1 µF ceramic bypass capacitors | — | 3.9 | — | ms |
| VREF Turn-on Time 2 (0.01%) | 0.1 µF ceramic bypass capacitor | — | 400 | — | µs |
| VREF Turn-on Time 3 (0.01%) | no bypass capacitor | — | 3 | — | µs |
| Power Supply Rejection | | — | 50 | — | dB |
| **External Reference** | | | | | |
| Input Voltage Range (VREF+ – VREF–) | | 1 | 2.5 | AV+ | V |
| Voltage on VREF+ or VREF– pin with respect to AGND | | 0 | — | AV+ | V |
| Input Current | VREF = 2.5 V | — | 2 | — | µA |
| Common Mode Rejection Ratio | | — | 120 | — | dB |
| **Power Specifications** | | | | | |
| Internal Reference Bias and Band Gap Generator | | — | 106 | 135 | µA |

## NOTES:

## 8. Temperature Sensor

The temperature sensor system consists of two diodes with different temperature properties and two constant current sources. The two channels are connected to the ADC inputs internally, using the ADC's analog multiplexer. The temperature sensor system can be used in single-ended or differential mode to measure the temperature of the C8051F350/1/2/3. Single channel measurements produce more output voltage per degree C, but are not as linear as differential measurements. See Table 8.1 for temperature sensor electrical characteristics.

The temperature sensor channels are automatically enabled when they are selected by the ADC multiplexer. To use the temperature sensor for a single-channel measurement, the ADC multiplexer should be configured with one channel connected to the temperature sensor, and the other connected to AGND. For a differential measurement, the temperature sensor should be selected for both ADC channels.

The transfer functions for single-channel and differential measurements are shown in Figure 8.2 and Figure 8.3, respectively. For slope and offset values, refer to Table 8.1.

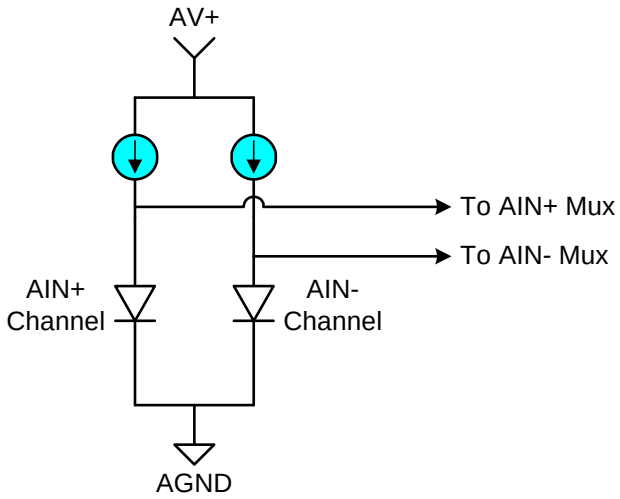



**Figure 8.1. Temperature Sensor Block Diagram**

### Table 8.1. Temperature Sensor Electrical Characteristics

$V_{DD}$ = 3.0 V; –40 to +85 °C unless otherwise specified.

| Parameter | Conditions | Min | Typ | Max | Units |
|---|---|---|---|---|---|
| Linearity | Single Channel Measurement<br>Differential Measurement | — | ±0.4<br>±0.01 | — | °C<br>°C |
| Offset | AIN+ Channel Measurement, Temp = 0 °C<br>Differential Measurement, Temp = 0 °C | — | 757<br>54.3 | — | mV<br>mV |
| Offset Error* | | — | ±1 | — | mV |
| Slope | AIN+ Channel Measurement<br>Differential Measurement | — | –1.73<br>205 | — | mV/°C<br>uV/°C |
| Slope Error* | | — | ±6.6 | — | µV/°C |
| AV+ Supply Current | Single Channel Measurement<br>Differential Measurement | — | 10<br>20 | — | µA<br>µA |
| ***Note:** Represents one standard deviation from the mean. | | | | | |

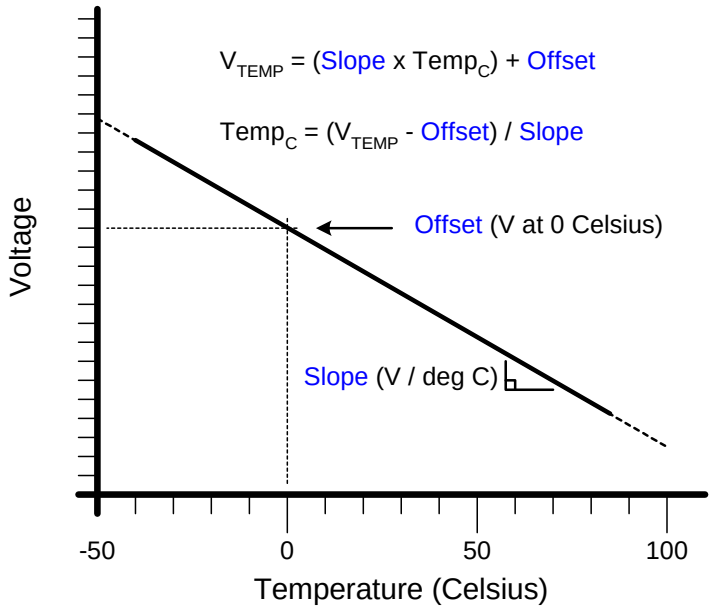



**Figure 8.2. Single Channel Transfer Function**

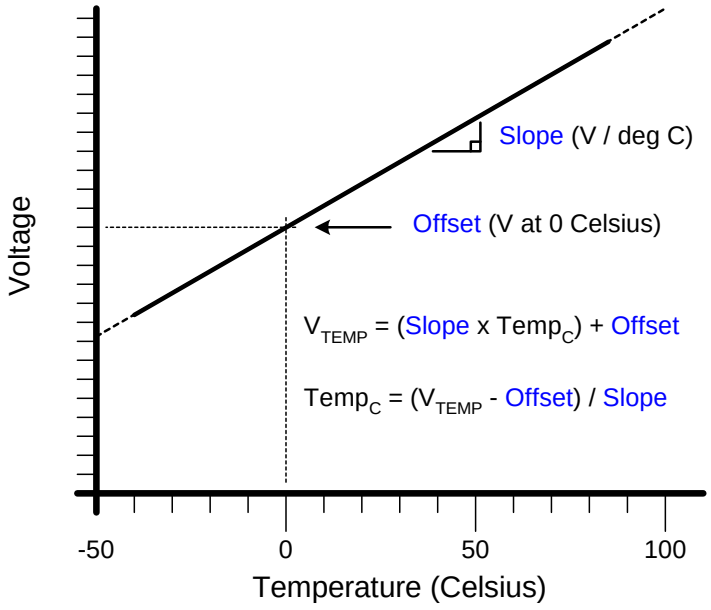



**Figure 8.3. Differential Transfer Function**

# 9. Comparator0

C8051F350/1/2/3 devices include an on-chip programmable voltage comparator, Comparator0, shown in Figure 9.1.

The Comparator offers programmable response time and hysteresis and two outputs that are optionally available at the Port pins: a synchronous “latched” output (CP0), or an asynchronous “raw” output (CP0A). The asynchronous CP0A signal is available even when the system clock is not active. This allows the Comparator to operate and generate an output with the device in STOP mode. When assigned to a Port pin, the Comparator output may be configured as open drain or push-pull (see Section “18.2. Port I/O Initialization’ on page 141). Comparator0 may also be used as a reset source (see Section “14.5. Comparator0 Reset’ on page 118).

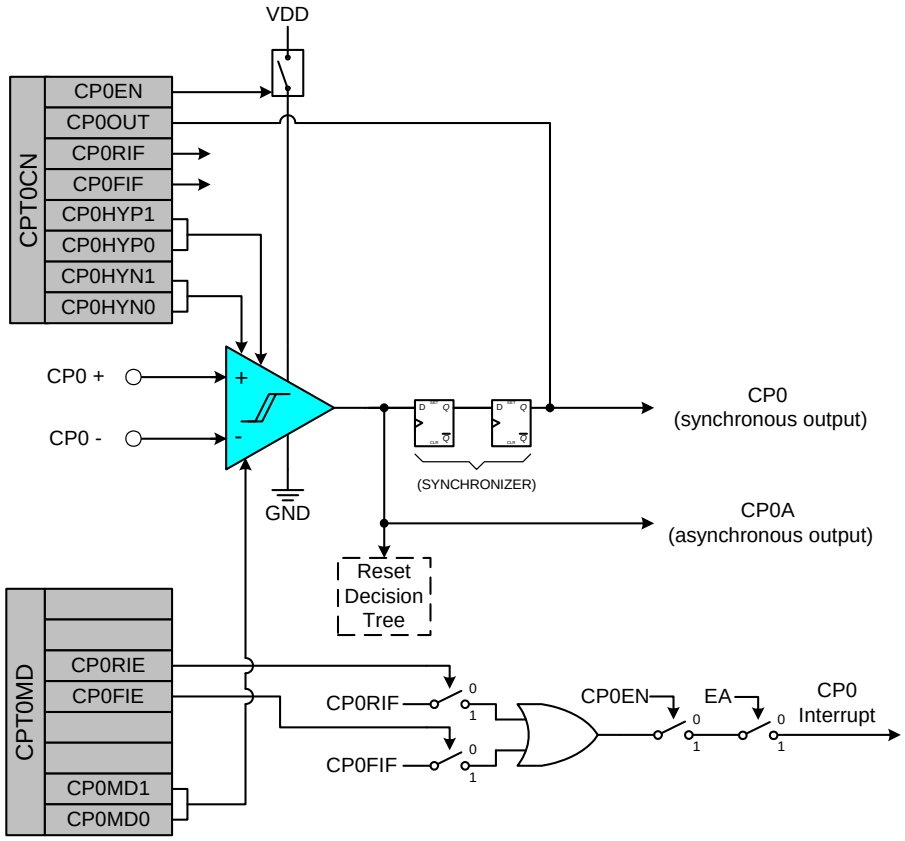



**Figure 9.1. Comparator0 Functional Block Diagram**

The Comparator output can be polled in software, used as an interrupt source, and/or routed to a Port pin. When routed to a Port pin, the Comparator output is available asynchronous or synchronous to the system clock; the asynchronous output is available even in STOP mode (with no system clock active). When disabled, the Comparator output (if assigned to a Port I/O pin) defaults to the logic low state, and its supply current falls to less than 100 nA. Comparator inputs can be externally driven from –0.25 V to ($V_{DD}$) + 0.25 V without damage or upset. The complete Comparator electrical specifications are given in Table 9.1.

The Comparator response time may be configured in software via the CPT0MD register (see SFR Definition 9.2). Selecting a longer response time reduces the Comparator supply current. See Table 9.1 for complete timing and power consumption specifications.

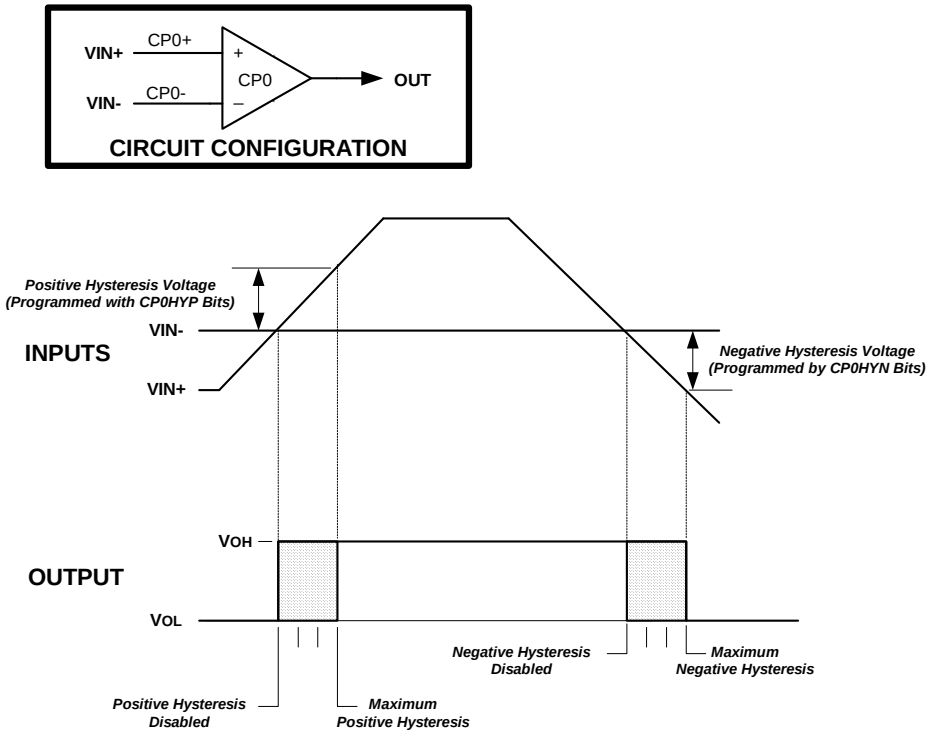



**Figure 9.2. Comparator Hysteresis Plot**

The Comparator hysteresis is software-programmable via its Comparator Control register CPT0CN. The user can program both the amount of hysteresis voltage (referred to the input voltage) and the positive and negative-going symmetry of this hysteresis around the threshold voltage.

The Comparator hysteresis is programmed using Bits3–0 in the Comparator Control Register CPT0CN (shown in SFR Definition 9.1). The amount of negative hysteresis voltage is determined by the settings of the CP0HYN bits. As shown in Figure 9.2, settings of 20, 10 or 5 mV of negative hysteresis can be programmed, or negative hysteresis can be disabled. In a similar way, the amount of positive hysteresis is determined by the setting the CP0HYP bits.

Comparator interrupts can be generated on both rising-edge and falling-edge output transitions. (For Interrupt enable and priority control, see Section “12. Interrupt Handler’ on page 105). The CP0FIF flag is set to logic 1 upon a Comparator falling-edge occurrence, and the CP0RIF flag is set to logic 1 upon the Comparator rising-edge occurrence. Once set, these bits remain set until cleared by software. The Comparator rising-edge interrupt mask is enabled by setting CP0RIE to a logic 1. The Comparator0 falling-edge interrupt mask is enabled by setting CP0FIE to a logic 1.

The output state of the Comparator can be obtained at any time by reading the CP0OUT bit. The Comparator is enabled by setting the CP0EN bit to logic 1, and is disabled by clearing this bit to logic 0.

Note that false rising edges and falling edges can be detected when the comparator is first powered on or if changes are made to the hysteresis or response time control bits. Therefore, it is recommended that the rising-edge and falling-edge flags be explicitly cleared to logic 0 a short time after the comparator is enabled or its mode bits have been changed. This Power Up Time is specified in Table 9.1 on page 85.

## SFR Definition 9.1. CPT0CN: Comparator0 Control

| R/W | R | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| CP0EN | CP0OUT | CP0RIF | CP0FIF | CP0HYP1 | CP0HYP0 | CP0HYN1 | CP0HYN0 | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0x9C

Bit7: CP0EN: Comparator0 Enable Bit.
0: Comparator0 Disabled.
1: Comparator0 Enabled.

Bit6: CP0OUT: Comparator0 Output State Flag.
0: Voltage on CP0+ < CP0–.
1: Voltage on CP0+ > CP0–.

Bit5: CP0RIF: Comparator0 Rising-Edge Flag. Must be cleared by software.
0: No Comparator0 Rising Edge has occurred since this flag was last cleared.
1: Comparator0 Rising Edge has occurred.

Bit4: CP0FIF: Comparator0 Falling-Edge Flag. Must be cleared by software.
0: No Comparator0 Falling-Edge has occurred since this flag was last cleared.
1: Comparator0 Falling-Edge has occurred.

Bits3–2: CP0HYP1–0: Comparator0 Positive Hysteresis Control Bits.
00: Positive Hysteresis Disabled.
01: Positive Hysteresis = 5 mV.
10: Positive Hysteresis = 10 mV.
11: Positive Hysteresis = 20 mV.

Bits1–0: CP0HYN1–0: Comparator0 Negative Hysteresis Control Bits.
00: Negative Hysteresis Disabled.
01: Negative Hysteresis = 5 mV.
10: Negative Hysteresis = 10 mV.
11: Negative Hysteresis = 20 mV.

## SFR Definition 9.2. CPT0MD: Comparator0 Mode Selection

| R | R | R/W | R/W | R | R | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| — | — | CP0RIE | CP0FIE | — | — | CP0MD1 | CP0MD0 | 00000010 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0x9D

Bits7–6: UNUSED. Read = 00b, Write = don't care.
Bit5: CP0RIE: Comparator0 Rising-Edge Interrupt Enable.
0: Comparator0 Rising-edge interrupt disabled.
1: Comparator0 Rising-edge interrupt enabled.
Bit4: CP0FIE: Comparator0 Falling-Edge Interrupt Enable.
0: Comparator0 Falling-edge interrupt disabled.
1: Comparator0 Falling-edge interrupt enabled.
Bits3–2: UNUSED. Read = 00b, Write = don't care.
Bits1–0: CP0MD1–CP0MD0: Comparator0 Mode Select
These bits select the response time for Comparator0.

| Mode | CP0MD1 | CP0MD0 | Notes |
|---|---|---|---|
| 0 | 0 | 0 | Fastest Response Time |
| 1 | 0 | 1 | — |
| 2 | 1 | 0 | — |
| 3 | 1 | 1 | Lowest Power Consumption |

## 9.1. Comparator0 Inputs and Outputs

Figure 9.3 shows the external pin connections for the comparator. The positive and negative inputs to the comparator can each be routed to one of eight different pins using the comparator mux. Comparator outputs can optionally be routed to port pins using the Crossbar circuitry.

The comparator inputs (CP0+ and CP0–) are selected in the CPT0MX register (SFR Definition 9.3). The CMX0P1–CMX0P0 bits select the comparator's positive input; the CMX0N1–CMX0N0 bits select the comparator's negative input. **Important Note About Comparator Inputs:** The Port pins selected as comparator inputs should be configured as analog inputs in their associated Port configuration register, and configured to be skipped by the Crossbar.

Two versions of the comparator output can be routed to port pins, using the Port I/O Crossbar. The raw (asynchronous) comparator output CP0A is enabled using bit 5 in the XBR0 register, and will be available at P1.4. The CP0 output (synchronized to SYSCLK) is available at P1.5 when it is enabled with bit 4 in the XBR0 register.

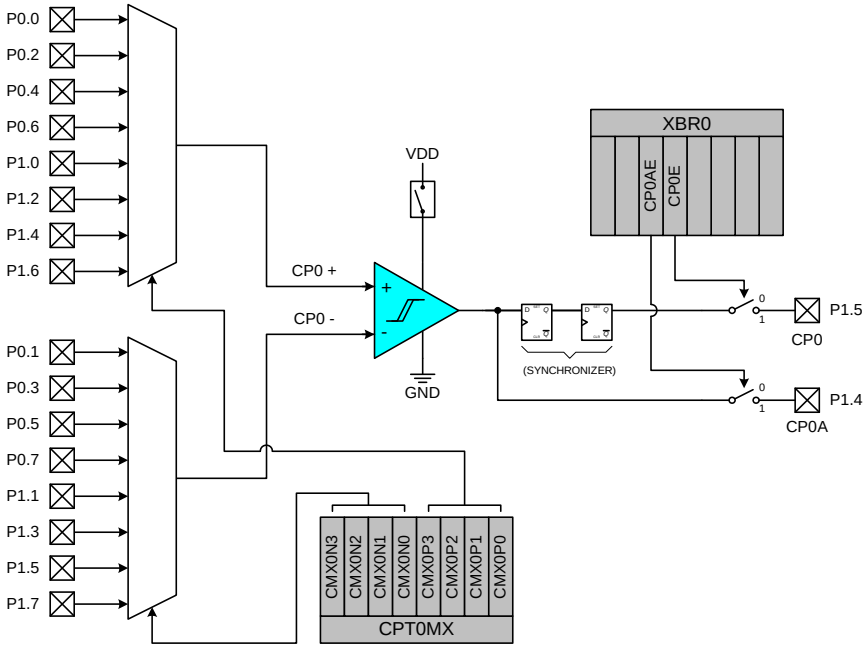



**Figure 9.3. Comparator Pin Connections**

## SFR Definition 9.3. CPT0MX: Comparator0 MUX Selection

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| CMX0N3 | CMX0N2 | CMX0N1 | CMX0N0 | CMX0P3 | CMX0P2 | CMX0P1 | CMX0P0 | 11111111 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0x9F

Bits7–4: CMX0N3–CMX0N0: Comparator0 Negative Input MUX Select.
These bits select which Port pin is used as the Comparator0 negative input.

| CMX0N3 | CMX0N2 | CMX0N1 | CMX0N0 | Negative Input |
|---|---|---|---|---|
| 0 | 0 | 0 | 0 | P0.1 |
| 0 | 0 | 0 | 1 | P0.3 |
| 0 | 0 | 1 | 0 | P0.5 |
| 0 | 0 | 1 | 1 | P0.7 |
| 0 | 1 | 0 | 0 | P1.1 |
| 0 | 1 | 0 | 1 | P1.3 |
| 0 | 1 | 1 | 0 | P1.5 |
| 0 | 1 | 1 | 1 | P1.7 |
| 1 | x | x | x | None |

Bits3–0: CMX0P3–CMX0P0: Comparator0 Positive Input MUX Select.
These bits select which Port pin is used as the Comparator0 positive input.

| CMX0P3 | CMX0P2 | CMX0P1 | CMX0P0 | Positive Input |
|---|---|---|---|---|
| 0 | 0 | 0 | 0 | P0.0 |
| 0 | 0 | 0 | 1 | P0.2 |
| 0 | 0 | 1 | 0 | P0.4 |
| 0 | 0 | 1 | 1 | P0.6 |
| 0 | 1 | 0 | 0 | P1.0 |
| 0 | 1 | 0 | 1 | P1.2 |
| 0 | 1 | 1 | 0 | P1.4 |
| 0 | 1 | 1 | 1 | P1.6 |
| 1 | x | x | x | None |

## Table 9.1. Comparator Electrical Characteristics

VDD = 3.0 V, –40 to +85 °C unless otherwise noted.

| Parameter | Conditions | Min | Typ | Max | Units |
|---|---|---|---|---|---|
| Response Time: Mode 0, Vcm* = 1.5 V | CP0+ – CP0– = 100 mV | — | 100 | — | ns |
| | CP0+ – CP0– = –100 mV | — | 250 | — | ns |
| Response Time: Mode 1, Vcm* = 1.5 V | CP0+ – CP0– = 100 mV | — | 175 | — | ns |
| | CP0+ – CP0– = –100 mV | — | 500 | — | ns |
| Response Time: Mode 2, Vcm* = 1.5 V | CP0+ – CP0– = 100 mV | — | 320 | — | ns |
| | CP0+ – CP0– = –100 mV | — | 1100 | — | ns |
| Response Time: Mode 3, Vcm* = 1.5 V | CP0+ – CP0– = 100 mV | — | 1050 | — | ns |
| | CP0+ – CP0– = –100 mV | — | 5200 | — | ns |
| Common-Mode Rejection Ratio | | — | 1.5 | 4 | mV/V |
| Positive Hysteresis 1 | CP0HYP1-0 = 00 | — | 0 | 1 | mV |
| Positive Hysteresis 2 | CP0HYP1-0 = 01 | 2 | 5 | 10 | mV |
| Positive Hysteresis 3 | CP0HYP1-0 = 10 | 7 | 10 | 20 | mV |
| Positive Hysteresis 4 | CP0HYP1-0 = 11 | 15 | 20 | 30 | mV |
| Negative Hysteresis 1 | CP0HYN1-0 = 00 | | 0 | 1 | mV |
| Negative Hysteresis 2 | CP0HYN1-0 = 01 | 2 | 5 | 10 | mV |
| Negative Hysteresis 3 | CP0HYN1-0 = 10 | 7 | 10 | 20 | mV |
| Negative Hysteresis 4 | CP0HYN1-0 = 11 | 15 | 20 | 30 | mV |
| Inverting or Non-Inverting Input Voltage Range | | –0.25 | — | VDD + 0.25 | V |
| Input Capacitance | | — | 4 | — | pF |
| Input Bias Current | | — | 0.001 | — | nA |
| Input Offset Voltage | | –5 | — | +5 | mV |
| **Power Supply** | | | | | |
| Power Supply Rejection | | — | 0.1 | — | mV/V |
| Power-up Time | | — | 10 | — | µs |
| Supply Current at DC | Mode 0 | — | 9.6 | 20.0 | µA |
| | Mode 1 | — | 3.8 | 10.0 | µA |
| | Mode 2 | — | 1.6 | 5.0 | µA |
| | Mode 3 | — | 0.3 | 2.0 | µA |

***Note:** Vcm is the common-mode voltage on CP0+ and CP0–.

## NOTES:

## 10. CIP-51 Microcontroller

The MCU system controller core is the CIP-51 microcontroller. The CIP-51 is fully compatible with the MCS-51™ instruction set. Standard 803x/805x assemblers and compilers can be used to develop software. The C8051F35x family has a superset of all the peripherals included with a standard 8051. See Section “1. System Overview’ on page 17 for more information about the available peripherals. The CIP-51 includes on-chip debug hardware which interfaces directly with the analog and digital subsystems, providing a complete data acquisition or control-system solution in a single integrated circuit.

The CIP-51 Microcontroller core implements the standard 8051 organization and peripherals as well as additional custom peripherals and functions to extend its capability (see Figure 10.1 for a block diagram). The CIP-51 core includes the following features:

- Fully Compatible with MCS-51 Instruction Set
- 50 MIPS Peak Throughput
- 256 Bytes of Internal RAM
- Extended Interrupt Handler
- Reset Input
- Power Management Modes
- Integrated Debug Logic

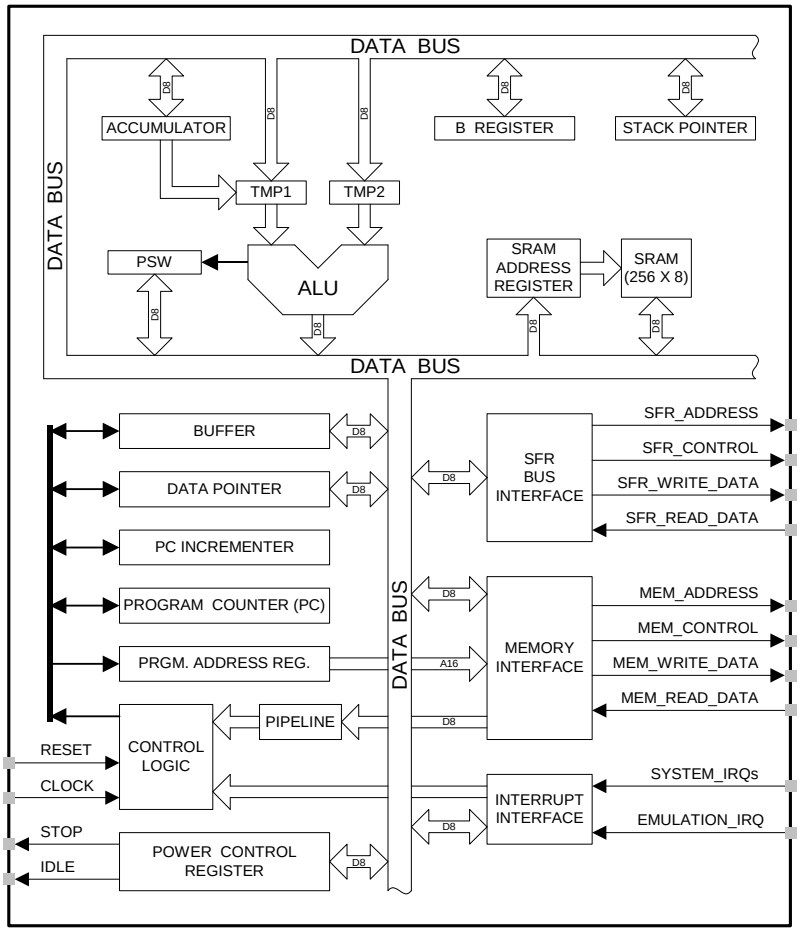



**Figure 10.1. CIP-51 Block Diagram**

**Performance**

The CIP-51 employs a pipelined architecture that greatly increases its instruction throughput over the standard 8051 architecture. In a standard 8051, all instructions except for MUL and DIV take 12 or 24 system clock cycles to execute, and usually have a maximum system clock of 12 MHz. By contrast, the CIP-51 core executes 70% of its instructions in one or two system clock cycles, with no instructions taking more than eight system clock cycles.

With the CIP-51's system clock running at 50 MHz, it has a peak throughput of 50 MIPS. The CIP-51 has a total of 109 instructions. The table below shows the total number of instructions that require each execution time.

| Clocks to Execute | 1 | 2 | 2/3 | 3 | 3/4 | 4 | 4/5 | 5 | 8 |
|---|---|---|---|---|---|---|---|---|---|
| Number of Instructions | 26 | 50 | 5 | 14 | 7 | 3 | 1 | 2 | 1 |

**Programming and Debugging Support**

In-system programming of the Flash program memory and communication with on-chip debug support logic is accomplished via the Cygnal 2-Wire (C2) interface. Note that the re-programmable Flash can also be read and changed a single byte at a time by the application software using the MOVC and MOVX instructions. This feature allows program memory to be used for non-volatile data storage as well as updating program code under software control.

The on-chip debug support logic facilitates full speed in-circuit debugging, allowing the setting of hardware breakpoints, starting, stopping and single stepping through program execution (including interrupt service routines), examination of the program's call stack, and reading/writing the contents of registers and memory. This method of on-chip debugging is completely non-intrusive, requiring no RAM, Stack, timers, or other on-chip resources.

The CIP-51 is supported by development tools from Silicon Labs and third party vendors. Silicon Labs provides an integrated development environment (IDE) including editor, macro assembler, debugger and programmer. The IDE's debugger and programmer interface to the CIP-51 via the on-chip debug logic to provide fast and efficient in-system device programming and debugging. Third party macro assemblers and C compilers are also available.

## 10.1. Instruction Set

The instruction set of the CIP-51 System Controller is fully compatible with the standard MCS-51™ instruction set. Standard 8051 development tools can be used to develop software for the CIP-51. All CIP-51 instructions are the binary and functional equivalent of their MCS-51™ counterparts, including opcodes, addressing modes and effect on PSW flags. However, instruction timing is different than that of the standard 8051.

### 10.1.1. Instruction and CPU Timing

In many 8051 implementations, a distinction is made between machine cycles and clock cycles, with machine cycles varying from 2 to 12 clock cycles in length. However, the CIP-51 implementation is based solely on clock cycle timing. All instruction timings are specified in terms of clock cycles.

Due to the pipelined architecture of the CIP-51, most instructions execute in the same number of clock cycles as there are program bytes in the instruction. Conditional branch instructions take one less clock cycle to complete when the branch is not taken as opposed to when the branch is taken. Table 10.1 is the CIP-51 Instruction Set Summary, which includes the mnemonic, number of bytes, and number of clock cycles for each instruction.

### 10.1.2. MOVX Instruction and Program Memory

The MOVX instruction is typically used to access data stored in XDATA memory space. In the CIP-51, the MOVX instruction can also be used to write or erase on-chip program memory space implemented as re-programmable Flash memory. The Flash access feature provides a mechanism for the CIP-51 to update program code and use the program memory space for non-volatile data storage. Refer to Section “15. Flash Memory’ on page 121 for further details.

**Table 10.1. CIP-51 Instruction Set Summary**

| Mnemonic | Description | Bytes | Clock Cycles |
|---|---|---|---|
| **Arithmetic Operations** | | | |
| ADD A, Rn | Add register to A | 1 | 1 |
| ADD A, direct | Add direct byte to A | 2 | 2 |
| ADD A, @Ri | Add indirect RAM to A | 1 | 2 |
| ADD A, #data | Add immediate to A | 2 | 2 |
| ADDC A, Rn | Add register to A with carry | 1 | 1 |
| ADDC A, direct | Add direct byte to A with carry | 2 | 2 |
| ADDC A, @Ri | Add indirect RAM to A with carry | 1 | 2 |
| ADDC A, #data | Add immediate to A with carry | 2 | 2 |
| SUBB A, Rn | Subtract register from A with borrow | 1 | 1 |
| SUBB A, direct | Subtract direct byte from A with borrow | 2 | 2 |
| SUBB A, @Ri | Subtract indirect RAM from A with borrow | 1 | 2 |
| SUBB A, #data | Subtract immediate from A with borrow | 2 | 2 |
| INC A | Increment A | 1 | 1 |
| INC Rn | Increment register | 1 | 1 |
| INC direct | Increment direct byte | 2 | 2 |
| INC @Ri | Increment indirect RAM | 1 | 2 |
| DEC A | Decrement A | 1 | 1 |
| DEC Rn | Decrement register | 1 | 1 |

**Table 10.1. CIP-51 Instruction Set Summary (Continued)**

| Mnemonic | Description | Bytes | Clock Cycles |
|---|---|---|---|
| DEC direct | Decrement direct byte | 2 | 2 |
| DEC @Ri | Decrement indirect RAM | 1 | 2 |
| INC DPTR | Increment Data Pointer | 1 | 1 |
| MUL AB | Multiply A and B | 1 | 4 |
| DIV AB | Divide A by B | 1 | 8 |
| DA A | Decimal adjust A | 1 | 1 |
| **Logical Operations** | | | |
| ANL A, Rn | AND Register to A | 1 | 1 |
| ANL A, direct | AND direct byte to A | 2 | 2 |
| ANL A, @Ri | AND indirect RAM to A | 1 | 2 |
| ANL A, #data | AND immediate to A | 2 | 2 |
| ANL direct, A | AND A to direct byte | 2 | 2 |
| ANL direct, #data | AND immediate to direct byte | 3 | 3 |
| ORL A, Rn | OR Register to A | 1 | 1 |
| ORL A, direct | OR direct byte to A | 2 | 2 |
| ORL A, @Ri | OR indirect RAM to A | 1 | 2 |
| ORL A, #data | OR immediate to A | 2 | 2 |
| ORL direct, A | OR A to direct byte | 2 | 2 |
| ORL direct, #data | OR immediate to direct byte | 3 | 3 |
| XRL A, Rn | Exclusive-OR Register to A | 1 | 1 |
| XRL A, direct | Exclusive-OR direct byte to A | 2 | 2 |
| XRL A, @Ri | Exclusive-OR indirect RAM to A | 1 | 2 |
| XRL A, #data | Exclusive-OR immediate to A | 2 | 2 |
| XRL direct, A | Exclusive-OR A to direct byte | 2 | 2 |
| XRL direct, #data | Exclusive-OR immediate to direct byte | 3 | 3 |
| CLR A | Clear A | 1 | 1 |
| CPL A | Complement A | 1 | 1 |
| RL A | Rotate A left | 1 | 1 |
| RLC A | Rotate A left through Carry | 1 | 1 |
| RR A | Rotate A right | 1 | 1 |
| RRC A | Rotate A right through Carry | 1 | 1 |
| SWAP A | Swap nibbles of A | 1 | 1 |
| **Data Transfer** | | | |
| MOV A, Rn | Move Register to A | 1 | 1 |
| MOV A, direct | Move direct byte to A | 2 | 2 |
| MOV A, @Ri | Move indirect RAM to A | 1 | 2 |
| MOV A, #data | Move immediate to A | 2 | 2 |
| MOV Rn, A | Move A to Register | 1 | 1 |
| MOV Rn, direct | Move direct byte to Register | 2 | 2 |
| MOV Rn, #data | Move immediate to Register | 2 | 2 |
| MOV direct, A | Move A to direct byte | 2 | 2 |
| MOV direct, Rn | Move Register to direct byte | 2 | 2 |
| MOV direct, direct | Move direct byte to direct byte | 3 | 3 |
| MOV direct, @Ri | Move indirect RAM to direct byte | 2 | 2 |
| MOV direct, #data | Move immediate to direct byte | 3 | 3 |

**Table 10.1. CIP-51 Instruction Set Summary (Continued)**

| Mnemonic | Description | Bytes | Clock Cycles |
|---|---|---|---|
| MOV @Ri, A | Move A to indirect RAM | 1 | 2 |
| MOV @Ri, direct | Move direct byte to indirect RAM | 2 | 2 |
| MOV @Ri, #data | Move immediate to indirect RAM | 2 | 2 |
| MOV DPTR, #data16 | Load DPTR with 16-bit constant | 3 | 3 |
| MOVC A, @A+DPTR | Move code byte relative DPTR to A | 1 | 3 |
| MOVC A, @A+PC | Move code byte relative PC to A | 1 | 3 |
| MOVX A, @Ri | Move external data (8-bit address) to A | 1 | 3 |
| MOVX @Ri, A | Move A to external data (8-bit address) | 1 | 3 |
| MOVX A, @DPTR | Move external data (16-bit address) to A | 1 | 3 |
| MOVX @DPTR, A | Move A to external data (16-bit address) | 1 | 3 |
| PUSH direct | Push direct byte onto stack | 2 | 2 |
| POP direct | Pop direct byte from stack | 2 | 2 |
| XCH A, Rn | Exchange Register with A | 1 | 1 |
| XCH A, direct | Exchange direct byte with A | 2 | 2 |
| XCH A, @Ri | Exchange indirect RAM with A | 1 | 2 |
| XCHD A, @Ri | Exchange low nibble of indirect RAM with A | 1 | 2 |
| **Boolean Manipulation** | | | |
| CLR C | Clear Carry | 1 | 1 |
| CLR bit | Clear direct bit | 2 | 2 |
| SETB C | Set Carry | 1 | 1 |
| SETB bit | Set direct bit | 2 | 2 |
| CPL C | Complement Carry | 1 | 1 |
| CPL bit | Complement direct bit | 2 | 2 |
| ANL C, bit | AND direct bit to Carry | 2 | 2 |
| ANL C, /bit | AND complement of direct bit to Carry | 2 | 2 |
| ORL C, bit | OR direct bit to carry | 2 | 2 |
| ORL C, /bit | OR complement of direct bit to Carry | 2 | 2 |
| MOV C, bit | Move direct bit to Carry | 2 | 2 |
| MOV bit, C | Move Carry to direct bit | 2 | 2 |
| JC rel | Jump if Carry is set | 2 | 2/3 |
| JNC rel | Jump if Carry is not set | 2 | 2/3 |
| JB bit, rel | Jump if direct bit is set | 3 | 3/4 |
| JNB bit, rel | Jump if direct bit is not set | 3 | 3/4 |
| JBC bit, rel | Jump if direct bit is set and clear bit | 3 | 3/4 |
| **Program Branching** | | | |
| ACALL addr11 | Absolute subroutine call | 2 | 3 |
| LCALL addr16 | Long subroutine call | 3 | 4 |
| RET | Return from subroutine | 1 | 5 |
| RETI | Return from interrupt | 1 | 5 |
| AJMP addr11 | Absolute jump | 2 | 3 |
| LJMP addr16 | Long jump | 3 | 4 |
| SJMP rel | Short jump (relative address) | 2 | 3 |
| JMP @A+DPTR | Jump indirect relative to DPTR | 1 | 3 |
| JZ rel | Jump if A equals zero | 2 | 2/3 |
| JNZ rel | Jump if A does not equal zero | 2 | 2/3 |

**Table 10.1. CIP-51 Instruction Set Summary (Continued)**

| Mnemonic | Description | Bytes | Clock Cycles |
|---|---|---|---|
| CJNE A, direct, rel | Compare direct byte to A and jump if not equal | 3 | 3/4 |
| CJNE A, #data, rel | Compare immediate to A and jump if not equal | 3 | 3/4 |
| CJNE Rn, #data, rel | Compare immediate to Register and jump if not equal | 3 | 3/4 |
| CJNE @Ri, #data, rel | Compare immediate to indirect and jump if not equal | 3 | 4/5 |
| DJNZ Rn, rel | Decrement Register and jump if not zero | 2 | 2/3 |
| DJNZ direct, rel | Decrement direct byte and jump if not zero | 3 | 3/4 |
| NOP | No operation | 1 | 1 |

**Notes on Registers, Operands and Addressing Modes:**

**Rn** - Register R0–R7 of the currently selected register bank.

**@Ri** - Data RAM location addressed indirectly through R0 or R1.

**rel** - 8-bit, signed (two's complement) offset relative to the first byte of the following instruction. Used by SJMP and all conditional jumps.

**direct** - 8-bit internal data location's address. This could be a direct-access Data RAM location (0x00–0x7F) or an SFR (0x80–0xFF).

**#data** - 8-bit constant

**#data16** - 16-bit constant

**bit** - Direct-accessed bit in Data RAM or SFR

**addr11** - 11-bit destination address used by ACALL and AJMP. The destination must be within the same 2 kB page of program memory as the first byte of the following instruction.

**addr16** - 16-bit destination address used by LCALL and LJMP. The destination may be anywhere within the 8 kB program memory space.

There is one unused opcode (0xA5) that performs the same function as NOP.
All mnemonics copyrighted © Intel Corporation 1980.

## 10.2. Register Descriptions

Following are descriptions of SFRs related to the operation of the CIP-51 System Controller. Reserved bits should not be set to logic l. Future product versions may use these bits to implement new features in which case the reset value of the bit will be logic 0, selecting the feature's default state. Detailed descriptions of the remaining SFRs are included in the sections of the datasheet associated with their corresponding system function.

### SFR Definition 10.1. SP: Stack Pointer

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| | | | | | | | | 00000111 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0x81

Bits7–0: SP: Stack Pointer.
The Stack Pointer holds the location of the top of the stack. The stack pointer is incremented before every PUSH operation. The SP register defaults to 0x07 after reset.

### SFR Definition 10.2. DPL: Data Pointer Low Byte

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0x82

Bits7–0: DPL: Data Pointer Low.
The DPL register is the low byte of the 16-bit DPTR. DPTR is used to access indirectly addressed XRAM and Flash memory.

### SFR Definition 10.3. DPH: Data Pointer High Byte

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0x83

Bits7–0: DPH: Data Pointer High.
The DPH register is the high byte of the 16-bit DPTR. DPTR is used to access indirectly addressed XRAM and Flash memory.

## SFR Definition 10.4. PSW: Program Status Word

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R | Reset Value |
|---|---|---|---|---|---|---|---|---|
| CY | AC | F0 | RS1 | RS0 | OV | F1 | PARITY | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | Bit Addressable |

SFR Address: 0xD0

Bit7: CY: Carry Flag.
This bit is set when the last arithmetic operation resulted in a carry (addition) or a borrow (subtraction). It is cleared to 0 by all other arithmetic operations.

Bit6: AC: Auxiliary Carry Flag
This bit is set when the last arithmetic operation resulted in a carry into (addition) or a borrow from (subtraction) the high order nibble. It is cleared to 0 by all other arithmetic operations.

Bit5: F0: User Flag 0.
This is a bit-addressable, general purpose flag for use under software control.

Bits4–3: RS1–RS0: Register Bank Select.
These bits select which register bank is used during register accesses.

| RS1 | RS0 | Register Bank | Address |
|---|---|---|---|
| 0 | 0 | 0 | 0x00 – 0x07 |
| 0 | 1 | 1 | 0x08 – 0x0F |
| 1 | 0 | 2 | 0x10 – 0x17 |
| 1 | 1 | 3 | 0x18 – 0x1F |

Bit2: OV: Overflow Flag.
This bit is set to 1 under the following circumstances:
- An ADD, ADDC, or SUBB instruction causes a sign-change overflow.
- A MUL instruction results in an overflow (result is greater than 255).
- A DIV instruction causes a divide-by-zero condition.

The OV bit is cleared to 0 by the ADD, ADDC, SUBB, MUL, and DIV instructions in all other cases.

Bit1: F1: User Flag 1.
This is a bit-addressable, general purpose flag for use under software control.

Bit0: PARITY: Parity Flag.
This bit is set to 1 if the sum of the eight bits in the accumulator is odd and cleared if the sum is even.

## SFR Definition 10.5. ACC: Accumulator

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| ACC.7 | ACC.6 | ACC.5 | ACC.4 | ACC.3 | ACC.2 | ACC.1 | ACC.0 | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | Bit Addressable |

SFR Address: 0xE0

Bits7–0: ACC: Accumulator.
This register is the accumulator for arithmetic operations.

## SFR Definition 10.6. B: B Register

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| B.7 | B.6 | B.5 | B.4 | B.3 | B.2 | B.1 | B.0 | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | Bit Addressable |

SFR Address: 0xF0

Bits7–0: B: B Register.
This register serves as a second accumulator for certain arithmetic operations.

## 10.3. Power Management Modes

The CIP-51 core has two software programmable power management modes: Idle and Stop. Idle mode halts the CPU while leaving the peripherals and internal clocks active. In Stop mode, the CPU is halted, all interrupts and timers (except the Missing Clock Detector) are inactive, and the internal oscillator is stopped (analog peripherals remain in their selected states; the external oscillator is not effected). Since clocks are running in Idle mode, power consumption is dependent upon the system clock frequency and the number of peripherals left in active mode before entering Idle. Stop mode consumes the least power. SFR Definition 10.7 describes the Power Control Register (PCON) used to control the CIP-51's power management modes.

Although the CIP-51 has Idle and Stop modes built in (as with any standard 8051 architecture), power management of the entire MCU is better accomplished by enabling/disabling individual peripherals as needed. Each analog peripheral can be disabled when not in use and placed in low power mode. Digital peripherals, such as timers or serial buses, draw little power when they are not in use. Turning off the oscillators lowers power consumption considerably; however a reset is required to restart the MCU.

### 10.3.1. Idle Mode

Setting the Idle Mode Select bit (PCON.0) causes the CIP-51 to halt the CPU and enter Idle mode as soon as the instruction that sets the bit completes execution. All internal registers and memory maintain their original data. All analog and digital peripherals can remain active during Idle mode.

Idle mode is terminated when an enabled interrupt is asserted or a reset occurs. The assertion of an enabled interrupt will cause the Idle Mode Selection bit (PCON.0) to be cleared and the CPU to resume operation. The pending interrupt will be serviced and the next instruction to be executed after the return from interrupt (RETI) will be the instruction immediately following the one that set the Idle Mode Select bit. If Idle mode is terminated by an internal or external reset, the CIP-51 performs a normal reset sequence and begins program execution at address 0x0000.

If enabled, the Watchdog Timer (WDT) will eventually cause an internal watchdog reset and thereby terminate the Idle mode. This feature protects the system from an unintended permanent shutdown in the event of an inadvertent write to the PCON register. If this behavior is not desired, the WDT may be disabled by software prior to entering the Idle mode if the WDT was initially configured to allow this operation. This provides the opportunity for additional power savings, allowing the system to remain in the Idle mode indefinitely, waiting for an external stimulus to wake up the system.

### 10.3.2. Stop Mode

Setting the Stop Mode Select bit (PCON.1) causes the CIP-51 to enter Stop mode as soon as the instruction that sets the bit completes execution. In Stop mode the internal oscillator, CPU, and all digital peripherals are stopped; the state of the external oscillator circuit is not affected. Each analog peripheral (including the external oscillator circuit) may be shut down individually prior to entering Stop Mode. Stop mode can only be terminated by an internal or external reset. On reset, the CIP-51 performs the normal reset sequence and begins program execution at address 0x0000.

If enabled, the Missing Clock Detector will cause an internal reset and thereby terminate the Stop mode. The Missing Clock Detector should be disabled if the CPU is to be put to in STOP mode for longer than the MCD timeout period of 100 µs.

## SFR Definition 10.7. PCON: Power Control

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| — | — | — | — | — | — | STOP | IDLE | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0x87

Bits7–3: Reserved.

Bit1: STOP: STOP Mode Select.
Writing a ‘1’ to this bit will place the CIP-51 into STOP mode. This bit will always read ‘0’.
1: CIP-51 forced into power-down mode. (Turns off internal oscillator).

Bit0: IDLE: IDLE Mode Select.
Writing a ‘1’ to this bit will place the CIP-51 into IDLE mode. This bit will always read ‘0’.
1: CIP-51 forced into IDLE mode. (Shuts off clock to CPU, but clock to Timers, Interrupts, and all peripherals remain active.)

## NOTES:

## 13. Prefetch Engine

The C8051F350/1/2/3 family of devices incorporate a 2-byte prefetch engine. Because the access time of the Flash memory is 40 ns, and the minimum instruction time is 20 ns, the prefetch engine is necessary for full-speed code execution. Instructions are read from Flash memory two bytes at a time by the prefetch engine, and given to the CIP-51 processor core to execute. When running linear code (code without any jumps or branches), the prefetch engine allows instructions to be executed at full speed. When a code branch occurs, the processor may be stalled for up to two clock cycles while the next set of code bytes is retrieved from Flash memory. The FLRT bit (FLSCL.4) determines how many clock cycles are used to read each set of two code bytes from Flash. When operating from a system clock of 25 MHz or less, the FLRT bit should be set to '0' so that the prefetch engine takes only one clock cycle for each read. When operating with a system clock of greater than 25 MHz (up to 50 MHz), the FLRT bit should be set to '1', so that each prefetch code read lasts for two clock cycles.

### SFR Definition 13.1. PFE0CN: Prefetch Engine Control

| R | R | R/W | R | R | R | R | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| | | PFEN | | | | | FLBWE | 00100000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xE3

Bits 7–6: Unused. Read = 00b; Write = Don't Care
Bit 5: PFEN: Prefetch Enable.
This bit enables the prefetch engine.
0: Prefetch engine is disabled.
1: Prefetch engine is enabled.
Bits 4–1: Unused. Read = 0000b; Write = Don't Care
Bit 0: FLBWE: Flash Block Write Enable.
This bit allows block writes to Flash memory from software.
0: Each byte of a software Flash write is written individually.
1: Flash bytes are written in groups of two.

## NOTES:

# 14. Reset Sources

Reset circuitry allows the controller to be easily placed in a predefined default condition. On entry to this reset state, the following occur:

- CIP-51 halts program execution
- Special Function Registers (SFRs) are initialized to their defined reset values
- External Port pins are forced to a known state
- Interrupts and timers are disabled.

All SFRs are reset to the predefined values noted in the SFR detailed descriptions. The contents of internal data memory are unaffected during a reset; any previously stored data is preserved. However, since the stack pointer SFR is reset, the stack is effectively lost, even though the data on the stack is not altered.

The Port I/O latches are reset to 0xFF (all logic ones) in open-drain mode. Weak pull-ups are enabled during and after the reset. For $V_{DD}$ Monitor and power-on resets, the /RST pin is driven low until the device exits the reset state.

On exit from the reset state, the program counter (PC) is reset, and the system clock defaults to the internal oscillator. Refer to Section “17. Oscillators’ on page 129 for information on selecting and configuring the system clock source. The Watchdog Timer is enabled with the system clock divided by 12 as its clock source (Section “23.3. Watchdog Timer Mode’ on page 220 details the use of the Watchdog Timer). Program execution begins at location 0x0000.

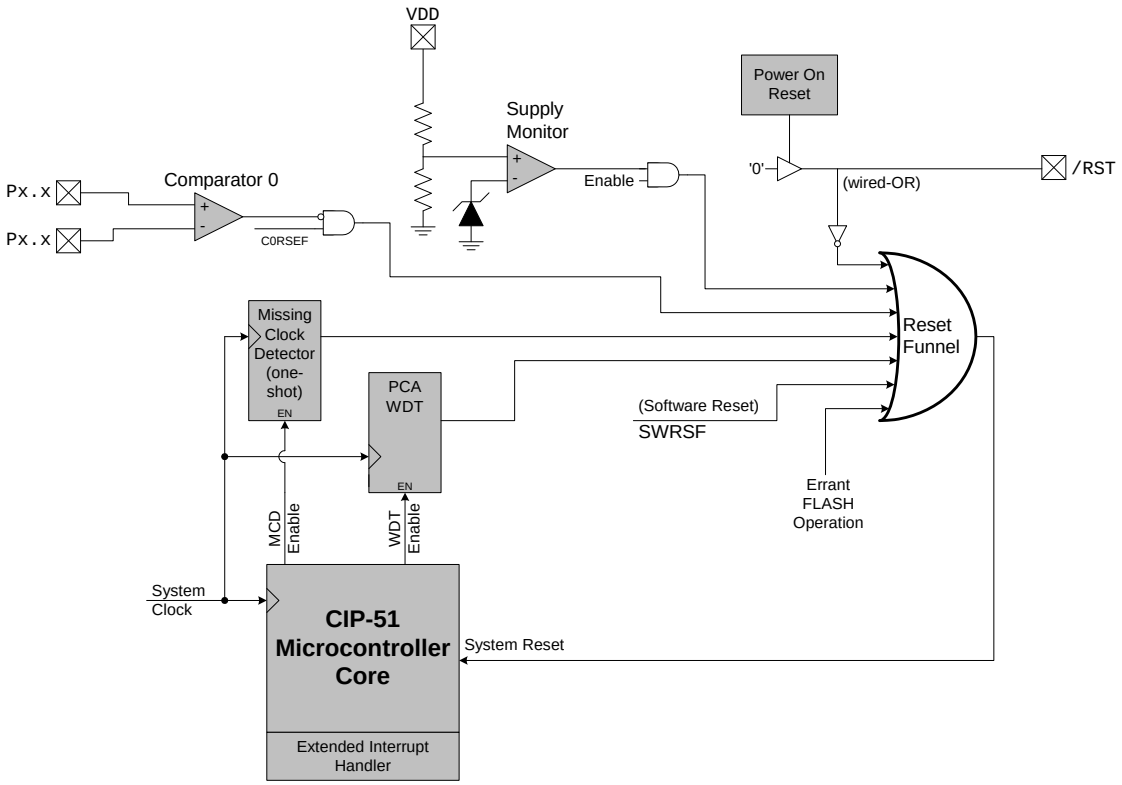



**Figure 14.1. Reset Sources**

## 14.1. Power-On Reset

During power-up, the device is held in a reset state and the /RST pin is driven low until $V_{DD}$ settles above $V_{RST}$. An additional delay occurs before the device is released from reset; the delay decreases as the $V_{DD}$ ramp time increases ($V_{DD}$ ramp time is defined as how fast $V_{DD}$ ramps from 0 V to $V_{RST}$). Figure 14.2. plots the power-on and $V_{DD}$ monitor reset timing. For valid ramp times (less than 1 ms), the power-on reset delay ($T_{PORDelay}$) is typically less than 0.3 ms.

**Note:** The maximum $V_{DD}$ ramp time is 1 ms; slower ramp times may cause the device to be released from reset before $V_{DD}$ reaches the $V_{RST}$ level.

On exit from a power-on reset, the PORSF flag (RSTSRC.1) is set by hardware to logic 1. When PORSF is set, all of the other reset flags in the RSTSRC Register are indeterminate (PORSF is cleared by all other resets). Since all resets cause program execution to begin at the same location (0x0000) software can read the PORSF flag to determine if a power-up was the cause of reset. The contents of internal data memory should be assumed to be undefined after a power-on reset. The $V_{DD}$ monitor is enabled following a power-on reset.

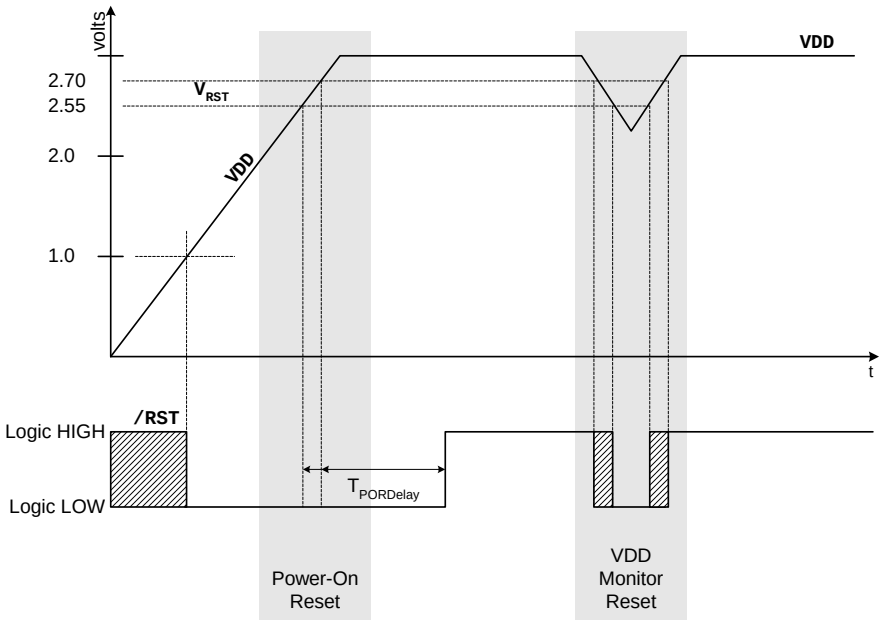



**Figure 14.2. Power-On and $V_{DD}$ Monitor Reset Timing**

## 14.2. Power-Fail Reset / $V_{DD}$ Monitor

When a power-down transition or power irregularity causes $V_{DD}$ to drop below $V_{RST}$, the power supply monitor will drive the /RST pin low and hold the CIP-51 in a reset state (see Figure 14.2). When $V_{DD}$ returns to a level above $V_{RST}$, the CIP-51 will be released from the reset state. Note that even though internal data memory contents are not altered by the power-fail reset, it is impossible to determine if $V_{DD}$ dropped below the level required for data retention. If the PORSF flag reads '1', the data may no longer be valid. The $V_{DD}$ monitor is enabled and selected as a reset source after power-on resets; however its defined state (enabled/disabled) is not altered by any other reset source. For example, if the $V_{DD}$ monitor is disabled by software, and a software reset is performed, the $V_{DD}$ monitor will still be disabled after the reset. **To protect the integrity of Flash contents, it is strongly recommended that the $V_{DD}$ monitor remain enabled and selected as a reset source if software contains routines which erase or write Flash memory.**

The $V_{DD}$ monitor must be enabled before it is selected as a reset source. Selecting the $V_{DD}$ monitor as a reset source before it is enabled and stabilized may cause a system reset. The procedure for re-enabling the $V_{DD}$ monitor and configuring the $V_{DD}$ monitor as a reset source is shown below:

Step 1. Enable the $V_{DD}$ monitor (VDMEN bit in VDM0CN = '1').
Step 2. Wait for the $V_{DD}$ monitor to stabilize (see Table 14.1 for the $V_{DD}$ Monitor turn-on time).
**Note: This delay should be omitted if software contains routines which erase or write Flash memory.**
Step 3. Select the $V_{DD}$ monitor as a reset source (PORSF bit in RSTSRC = '1').

See Figure 14.2 for $V_{DD}$ monitor timing; note that the reset delay is not incurred after a $V_{DD}$ monitor reset. See Table 14.1 for complete electrical characteristics of the $V_{DD}$ monitor.

### SFR Definition 14.1. VDM0CN: $V_{DD}$ Monitor Control

| R/W | R | R | R | R | R | R | R | Reset Value |
|---|---|---|---|---|---|---|---|---|
| VDMEN | $V_{DD}$STAT | Reserved | Reserved | Reserved | Reserved | Reserved | Reserved | Variable |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xFF

Bit7: VDMEN: $V_{DD}$ Monitor Enable.
This bit is turns the $V_{DD}$ monitor circuit on/off. The $V_{DD}$ Monitor cannot generate system resets until it is also selected as a reset source in register RSTSRC (SFR Definition 14.2). The $V_{DD}$ Monitor must be allowed to stabilize before it is selected as a reset source. **Selecting the $V_{DD}$ monitor as a reset source before it has stabilized may generate a system reset.** See Table 14.1 for the minimum $V_{DD}$ Monitor turn-on time.
0: $V_{DD}$ Monitor Disabled.
1: $V_{DD}$ Monitor Enabled (default).

Bit6: $V_{DD}$ STAT: $V_{DD}$ Status.
This bit indicates the current power supply status ($V_{DD}$ Monitor output).
0: $V_{DD}$ is at or below the $V_{DD}$ monitor threshold.
1: $V_{DD}$ is above the $V_{DD}$ monitor threshold.

Bits5–0: Reserved. Read = Variable. Write = don't care.

## 14.3. External Reset

The external /RST pin provides a means for external circuitry to force the device into a reset state. Asserting an active-low signal on the /RST pin generates a reset; an external pull-up and/or decoupling of the /RST pin may be necessary to avoid erroneous noise-induced resets. See Table 14.1 for complete /RST pin specifications. The PINRSF flag (RSTSRC.0) is set on exit from an external reset.

## 14.4. Missing Clock Detector Reset

The Missing Clock Detector (MCD) is a one-shot circuit that is triggered by the system clock. If the system clock remains high or low for more than 100 µs, the one-shot will time out and generate a reset. After a MCD reset, the MCDRSF flag (RSTSRC.2) will read ‘1’, signifying the MCD as the reset source; otherwise, this bit reads ‘0’. Writing a ‘1’ to the MCDRSF bit enables the Missing Clock Detector; writing a ‘0’ disables it. The state of the /RST pin is unaffected by this reset.

## 14.5. Comparator0 Reset

Comparator0 can be configured as a reset source by writing a ‘1’ to the C0RSEF flag (RSTSRC.5). Comparator0 should be enabled and allowed to settle prior to writing to C0RSEF to prevent any turn-on chatter on the output from generating an unwanted reset. The Comparator0 reset is active-low: if the non-inverting input voltage (on CP0+) is less than the inverting input voltage (on CP0–), the device is put into the reset state. After a Comparator0 reset, the C0RSEF flag (RSTSRC.5) will read ‘1’ signifying Comparator0 as the reset source; otherwise, this bit reads ‘0’. The state of the /RST pin is unaffected by this reset.

## 14.6. PCA Watchdog Timer Reset

The programmable Watchdog Timer (WDT) function of the Programmable Counter Array (PCA) can be used to prevent software from running out of control during a system malfunction. The PCA WDT function can be enabled or disabled by software as described in Section “23.3. Watchdog Timer Mode’ on page 220; the WDT is enabled and clocked by SYSCLK / 12 following any reset. If a system malfunction prevents user software from updating the WDT, a reset is generated and the WDTRSF bit (RSTSRC.5) is set to ‘1’. The state of the /RST pin is unaffected by this reset.

## 14.7. Flash Error Reset

If a Flash read/write/erase or program read targets an illegal address, a system reset is generated. This may occur due to any of the following:

- A Flash write or erase is attempted above user code space. This occurs when PSWE is set to ‘1’ and a MOVX write operation targets an address above address 0x1DFF.
- A Flash read is attempted above user code space. This occurs when a MOVC operation targets an address above address 0x1DFF.
- A Program read is attempted above user code space. This occurs when user code attempts to branch to an address above 0x1DFF.
- A Flash read, write or erase attempt is restricted due to a Flash security setting (see Section “15.3. Security Options’ on page 123).

The FERROR bit (RSTSRC.6) is set following a Flash error reset. The state of the /RST pin is unaffected by this reset.

## 14.8. Software Reset

Software may force a reset by writing a ‘1’ to the SWRSF bit (RSTSRC.4). The SWRSF bit will read ‘1’ following a software forced reset. The state of the /RST pin is unaffected by this reset.

## SFR Definition 14.2. RSTSRC: Reset Source

| R | R | R/W | R/W | R | R/W | R/W | R | Reset Value |
|---|---|---|---|---|---|---|---|---|
| — | FERROR | C0RSEF | SWRSF | WDTRSF | MCDRSF | PORSF | PINRSF | Variable |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xEF

Bit7: UNUSED. Read = 0. Write = don't care.

Bit6: FERROR: Flash Error Indicator.
0: Source of last reset was not a Flash read/write/erase error.
1: Source of last reset was a Flash read/write/erase error.

Bit5: C0RSEF: Comparator0 Reset Enable and Flag.
0: **Read:** Source of last reset was not Comparator0. **Write:** Comparator0 is not a reset source.
1: **Read:** Source of last reset was Comparator0. **Write:** Comparator0 is a reset source (active-low).

Bit4: SWRSF: Software Reset Force and Flag.
0: **Read:** Source of last reset was not a write to the SWRSF bit. **Write:** No Effect.
1: **Read:** Source of last was a write to the SWRSF bit. **Write:** Forces a system reset.

Bit3: WDTRSF: Watchdog Timer Reset Flag.
0: Source of last reset was not a WDT timeout.
1: Source of last reset was a WDT timeout.

Bit2: MCDRSF: Missing Clock Detector Flag.
0: **Read:** Source of last reset was not a Missing Clock Detector timeout. **Write:** Missing Clock Detector disabled.
1: **Read:** Source of last reset was a Missing Clock Detector timeout. **Write:** Missing Clock Detector enabled; triggers a reset if a missing clock condition is detected.

Bit1: PORSF: Power-On Reset Force and Flag.
This bit is set anytime a power-on reset occurs. Writing this bit enables/disables the $V_{DD}$ monitor as a reset source. **Note: writing '1' to this bit before the $V_{DD}$ monitor is enabled and stabilized may cause a system reset.** See register VDM0CN (SFR Definition 14.1)
0: **Read:** Last reset was not a power-on or $V_{DD}$ monitor reset. **Write:** $V_{DD}$ monitor is not a reset source.
1: **Read:** Last reset was a power-on or $V_{DD}$ monitor reset; all other reset flags indeterminate. **Write:** $V_{DD}$ monitor is a reset source.

Bit0: PINRSF: HW Pin Reset Flag.
0: Source of last reset was not /RST pin.
1: Source of last reset was /RST pin.

## Table 14.1. Reset Electrical Characteristics

–40 to +85 °C unless otherwise specified.

| Parameter | Conditions | Min | Typ | Max | Units |
|---|---|---|---|---|---|
| /RST Output Low Voltage | $I_{OL}$ = 8.5 mA, $V_{DD}$ = 2.7 to 3.6 V | — | — | 0.6 | V |
| /RST Input High Voltage | | 0.7 x $V_{DD}$ | — | — | V |
| /RST Input Low Voltage | | — | — | 0.3 x $V_{DD}$ | |
| /RST Input Pullup Current | /RST = 0.0 V | — | 25 | 40 | µA |
| $V_{DD}$ Monitor Threshold ($V_{RST}$) | | 2.40 | 2.55 | 2.70 | V |
| Missing Clock Detector Timeout | Time from last system clock rising edge to reset initiation | 100 | 220 | 600 | µs |
| Reset Time Delay | Delay between release of any reset source and code execution at location 0x0000 | 5.0 | — | — | µs |
| Minimum /RST Low Time to Generate a System Reset | | 15 | — | — | µs |
| $V_{DD}$ Monitor Turn-on Time | | 100 | — | — | µs |
| $V_{DD}$ Monitor Supply Current | | — | 20 | 50 | µA |
| $V_{DD}$ Ramp Time | $V_{DD}$ = 0 V to $V_{RST}$ | — | — | 1 | ms |

# 15. Flash Memory

On-chip, re-programmable Flash memory is included for program code and non-volatile data storage. The Flash memory can be programmed in-system through the C2 interface or by software using the MOVX instruction. Once cleared to logic 0, a Flash bit must be erased to set it back to logic 1. Flash bytes would typically be erased (set to 0xFF) before being reprogrammed. The write and erase operations are automatically timed by hardware for proper execution; data polling to determine the end of the write/erase operation is not required. Code execution is stalled during a Flash write/erase operation. Refer to Table 15.1 for complete Flash memory electrical characteristics.

## 15.1. Programming The Flash Memory

The simplest means of programming the Flash memory is through the C2 interface using programming tools provided by Silicon Labs or a third party vendor. This is the only means for programming a non-initialized device. For details on the C2 commands to program Flash memory, see Section “25. C2 Interface’ on page 229.

**To ensure the integrity of Flash contents, it is strongly recommended that the on-chip $V_{DD}$ Monitor be enabled in any system that includes code that writes and/or erases Flash memory from software.**

### 15.1.1. Flash Lock and Key Functions

Flash writes and erases by user software are protected with a lock and key function. The Flash Lock and Key Register (FLKEY) must be written with the correct key codes, in sequence, before Flash operations may be performed. The key codes are: 0xA5, 0xF1. The timing does not matter, but the codes must be written in order. If the key codes are written out of order, or the wrong codes are written, Flash writes and erases will be disabled until the next system reset. Flash writes and erases will also be disabled if a Flash write or erase is attempted before the key codes have been written properly. The Flash lock resets after each write or erase; the key codes must be written again before a following Flash operation can be performed. The FLKEY register is detailed in SFR Definition 15.2.

### 15.1.2. Flash Erase Procedure

The Flash memory can be programmed by software using the MOVX write instruction with the address and data byte to be programmed provided as normal operands. Before writing to Flash memory using MOVX, Flash write operations must be enabled by: (1) setting the PSWE Program Store Write Enable bit (PSCTL.0) to logic 1 (this directs the MOVX writes to target Flash memory); and (2) Writing the Flash key codes in sequence to the Flash Lock register (FLKEY). The PSWE bit remains set until cleared by software.

A write to Flash memory can clear bits to logic 0 but cannot set them; only an erase operation can set bits to logic 1 in Flash. **A byte location to be programmed should be erased before a new value is written.** The Flash memory is organized in 512-byte pages. The erase operation applies to an entire page (setting all bytes in the page to 0xFF). To erase an entire 512-byte page, perform the following steps:

Step 1. Disable interrupts (recommended).
Step 2. Set thePSEE bit (register PSCTL).
Step 3. Set the PSWE bit (register PSCTL).
Step 4. Write the first key code to FLKEY: 0xA5.
Step 5. Write the second key code to FLKEY: 0xF1.
Step 6. Using the MOVX instruction, write a data byte to any location within the 512-byte page to be erased.
Step 7. Clear the PSWE and PSEE bits.
Step 8. Re-enable interrupts.

### 15.1.3. Flash Write Procedure

Bytes in Flash memory can be written one byte at a time, or in groups of two. The FLBWE bit in register PFE0CN (SFR Definition 13.1) controls whether a single byte or a block of two bytes is written to Flash during a write operation. When FLBWE is cleared to ‘0’, the Flash will be written one byte at a time. When FLBWE is set to ‘1’, the Flash will be written in two-byte blocks. Block writes are performed in the same amount of time as single-byte writes, which can save time when storing large amounts of data to Flash memory.

During a single-byte write to Flash, bytes are written individually, and a Flash write will be performed after each MOVX write instruction. The recommended procedure for writing Flash in single bytes is:

Step 1. Disable interrupts.
Step 2. Clear the FLBWE bit (register PFE0CN) to select single-byte write mode.
Step 3. Set the PSWE bit (register PSCTL).
Step 4. Clear the PSEE bit (register PSCTL).
Step 5. Write the first key code to FLKEY: 0xA5.
Step 6. Write the second key code to FLKEY: 0xF1.
Step 7. Using the MOVX instruction, write a single data byte to the desired location within the 512-byte sector.
Step 8. Clear the PSWE bit.
Step 9. Re-enable interrupts.

Steps 5–7 must be repeated for each byte to be written.

For block Flash writes, the Flash write procedure is only performed after the last byte of each block is written with the MOVX write instruction. A Flash write block is two bytes long, from even addresses to odd addresses. Writes must be performed sequentially (i.e. addresses ending in 0b and 1b must be written in order). The Flash write will be performed following the MOVX write that targets the address ending in 1b. If a byte in the block does not need to be updated in Flash, it should be written to 0xFF. The recommended procedure for writing Flash in blocks is:

Step 1. Disable interrupts.
Step 2. Set the FLBWE bit (resgister PFE0CN) to select block write mode.
Step 3. Set the PSWE bit (register PSCTL).
Step 4. Clear the PSEE bit (register PSCTL).
Step 5. Write the first key code to FLKEY: 0xA5.
Step 6. Write the second key code to FLKEY: 0xF1.
Step 7. Using the MOVX instruction, write the first data byte to the even block location (ending in 0b).
Step 8. Write the first key code to FLKEY: 0xA5.
Step 9. Write the second key code to FLKEY: 0xF1.
Step 10. Using the MOVX instruction, write the second data byte to the odd block location (ending in 1b).
Step 11. Clear the PSWE bit.
Step 12. Re-enable interrupts.

Steps 5–10 must be repeated for each block to be written.

## 15.2. Non-volatile Data Storage

The Flash memory can be used for non-volatile data storage as well as program code. This allows data such as calibration coefficients to be calculated and stored at run time. Data is written using the MOVX write instruction and read using the MOVC instruction. Note: MOVX read instructions always target XRAM.

## 15.3. Security Options

The CIP-51 provides security options to protect the Flash memory from inadvertent modification by software as well as to prevent the viewing of proprietary program code and constants. The Program Store Write Enable (bit PSWE in register PSCTL) and the Program Store Erase Enable (bit PSEE in register PSCTL) bits protect the Flash memory from accidental modification by software. PSWE must be explicitly set to '1' before software can modify the Flash memory; both PSWE and PSEE must be set to '1' before software can erase Flash memory. Additional security features prevent proprietary program code and data constants from being read or altered across the C2 interface.

A Security Lock Byte located at the last byte of Flash user space offers protection of the Flash program memory from access (reads, writes, or erases) by unprotected code or the C2 interface. The Flash security mechanism allows the user to lock *n* 512-byte Flash pages, starting at page 0 (addresses 0x0000 to 0x01FF), where *n* is the 1's complement number represented by the Security Lock Byte. Note that the page containing the Flash Security Lock Byte is locked when any other Flash pages are locked. See example below.

| | |
|---|---|
| Security Lock Byte: | 11111101b |
| 1's Complement: | 00000010b |
| Flash pages locked: | 3 (First two Flash pages + Lock Byte Page) |
| Addresses locked: | 0x0000 to 0x03FF (first two Flash pages) and 0x1C00 to 0x1DFF (Lock Byte Page) |

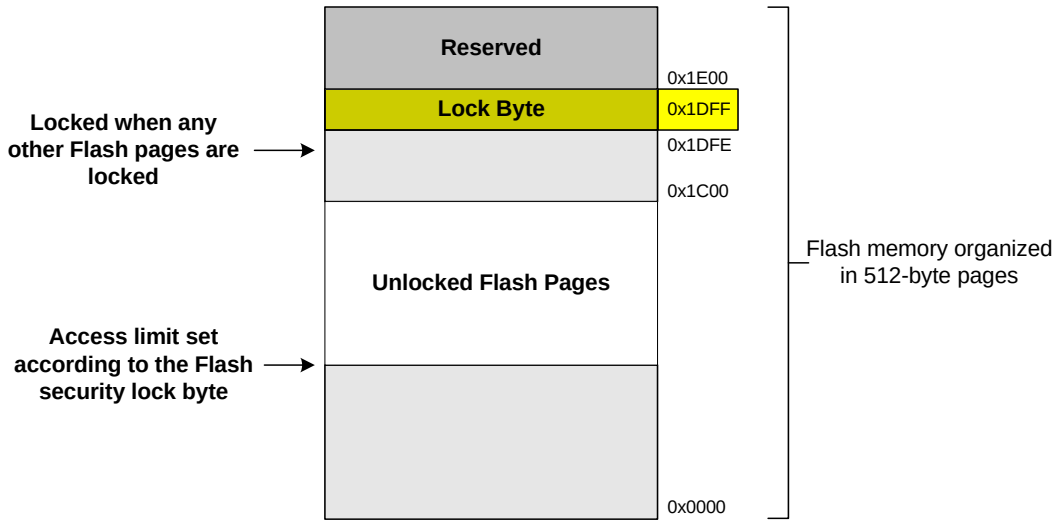



**Figure 15.1. Flash Memory Map**

The level of Flash security depends on the Flash access method. The three Flash access methods that can be restricted are reads, writes, and erases from the C2 debug interface, user firmware executing on unlocked pages, and user firmware executing on locked pages.

Accessing Flash from the C2 debug interface:

1. Any unlocked page may be read, written, or erased.
2. Locked pages cannot be read, written, or erased.
3. The page containing the Lock Byte may be read, written, or erased if it is unlocked.
4. Reading the contents of the Lock Byte is always permitted only if no pages are locked.
5. Locking additional pages (changing ‘1’s to ‘0’s in the Lock Byte) is not permitted.
6. Unlocking Flash pages (changing ‘0’s to ‘1’s in the Lock Byte) requires the C2 Device Erase command, which erases all Flash pages including the page containing the Lock Byte and the Lock Byte itself.
7. The Reserved Area cannot be read, written, or erased.

Accessing Flash from user firmware executing from an unlocked page:

1. Any unlocked page except the page containing the Lock Byte may be read, written, or erased.
2. Locked pages cannot be read, written, or erased. An erase attempt on the page containing the Lock Byte will result in a Flash Error device reset.
3. The page containing the Lock Byte cannot be erased. It may be read or written only if it is unlocked. An erase attempt on the page containing the Lock Byte will result in a Flash Error device reset.
4. Reading the contents of the Lock Byte is always permitted.
5. Locking additional pages (changing ‘1’s to ‘0’s in the Lock Byte) is not permitted.
6. Unlocking Flash pages (changing ‘0’s to ‘1’s in the Lock Byte) is not permitted.
7. The Reserved Area cannot be read, written, or erased. Any attempt to access the reserved area, or any other locked page, will result in a Flash Error device reset.

Accessing Flash from user firmware executing from a locked page:

1. Any unlocked page except the page containing the Lock Byte may be read, written, or erased.
2. Any locked page except the page containing the Lock Byte may be read, written, or erased. An erase attempt on the page containing the Lock Byte will result in a Flash Error device reset.
3. The page containing the Lock Byte cannot be erased. It may only be read or written. An erase attempt on the page containing the Lock Byte will result in a Flash Error device reset.
4. Reading the contents of the Lock Byte is always permitted.
5. Locking additional pages (changing ‘1’s to ‘0’s in the Lock Byte) is not permitted.
6. Unlocking Flash pages (changing ‘0’s to ‘1’s in the Lock Byte) is not permitted.
7. The Reserved Area cannot be read, written, or erased. Any attempt to access the reserved area, or any other locked page, will result in a Flash Error device reset.

## SFR Definition 15.1. PSCTL: Program Store R/W Control

| R | R | R | R | R | R | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| — | — | — | — | — | — | PSEE | PSWE | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0x8F

Bits7–2: UNUSED: Read = 000000b, Write = don't care.

Bit1: PSEE: Program Store Erase Enable
Setting this bit (in combination with PSWE) allows an entire page of Flash program memory to be erased. If this bit is logic 1 and Flash writes are enabled (PSWE is logic 1), a write to Flash memory using the MOVX instruction will erase the entire page that contains the location addressed by the MOVX instruction. The value of the data byte written does not matter.
0: Flash program memory erasure disabled.
1: Flash program memory erasure enabled.

Bit0: PSWE: Program Store Write Enable
Setting this bit allows writing a byte of data to the Flash program memory using the MOVX write instruction. The Flash location should be erased before writing data.
0: Writes to Flash program memory disabled.
1: Writes to Flash program memory enabled; the MOVX write instruction targets Flash memory.

## SFR Definition 15.2. FLKEY: Flash Lock and Key

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xB7

Bits7–0: FLKEY: Flash Lock and Key Register
Write:
This register provides a lock and key function for Flash erasures and writes. Flash writes and erases are enabled by writing 0xA5 followed by 0xF1 to the FLKEY register. Flash writes and erases are automatically disabled after the next write or erase is complete. If any writes to FLKEY are performed incorrectly, or if a Flash write or erase operation is attempted while these operations are disabled, the Flash will be permanently locked from writes or erasures until the next device reset. If an application never writes to Flash, it can intentionally lock the Flash by writing a non-0xA5 value to FLKEY from software.
Read:
When read, bits 1–0 indicate the current Flash lock state.
00: Flash is write/erase locked.
01: The first key code has been written (0xA5).
10: Flash is unlocked (writes/erases allowed).
11: Flash writes/erases disabled until the next reset.

## SFR Definition 15.3. FLSCL: Flash Scale

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| Reserved | Reserved | Reserved | FLRT | Reserved | Reserved | Reserved | Reserved | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xB6

Bits7–5: RESERVED. Read = 000b. Must Write 000b.
Bit 4: FLRT: Flash Read Time.
This bit should be programmed to the smallest allowed value, according to the system clock speed.
0: SYSCLK ≤ 25 MHz.
1: SYSCLK ≤ 50 MHz.
Bits3–0: RESERVED. Read = 0000b. Must Write 0000b.

## Table 15.1. Flash Electrical Characteristics

$V_{DD}$ = 2.7 to 3.6 V; –40 to +85 ºC unless otherwise specified.

| Parameter | Conditions | Min | Typ | Max | Units |
|---|---|---|---|---|---|
| Flash Size | C8051F350/1/2/3 | 8192* | — | — | bytes |
| Endurance | | 20 k | 100 k | — | Erase/Write |
| Erase Cycle Time | 50 MHz System Clock | 10 | 15 | 20 | ms |
| Write Cycle Time | 50 MHz System Clock | 40 | 55 | 70 | µs |
| ***Note:** 512 bytes at addresses 0x1E00 to 0x1FFF are reserved. | | | | | |

# 16. External RAM

The C8051F350/1/2/3 devices include 512 bytes of RAM mapped into the external data memory space. All of these address locations may be accessed using the external move instruction (MOVX) and the data pointer (DPTR), or using MOVX indirect addressing mode. If the MOVX instruction is used with an 8-bit address operand (such as @R1), then the high byte of the 16-bit address is provided by the External Memory Interface Control Register (EMI0CN as shown in SFR Definition 16.1). Note: the MOVX instruction is also used for writes to the Flash memory. See Section “15. Flash Memory’ on page 121 for details. The MOVX instruction accesses XRAM by default.

For a 16-bit MOVX operation (@DPTR), the upper 7-bits of the 16-bit external data memory address word are "don't cares". As a result, the 512-byte RAM is mapped modulo style over the entire 64 k external data memory address range. For example, the XRAM byte at address 0x0000 is shadowed at addresses 0x0200, 0x0400, 0x0600, 0x0800, etc. This is a useful feature when performing a linear memory fill, as the address pointer doesn't have to be reset when reaching the RAM block boundary.

### SFR Definition 16.1. EMI0CN: External Memory Interface Control

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| — | — | — | — | — | — | — | PGSEL | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xAA

Bits 7–1: UNUSED. Read = 0000000b. Write = don’t care.

Bit 0: PGSEL: XRAM Page Select.
The EMI0CN register provides the high byte of the 16-bit external data memory address when using an 8-bit MOVX command, effectively selecting a 256-byte page of RAM. Since the upper (unused) bits of the register are always zero, the PGSEL determines which page of XRAM is accessed.

For Example: If EMI0CN = 0x01, addresses 0x0100 through 0x01FF will be accessed.

## NOTES:

# 11. Memory Organization and SFRs

The memory organization of the C8051F350/1/2/3 is similar to that of a standard 8051. There are two separate memory spaces: program memory and data memory. Program and data memory share the same address space but are accessed via different instruction types. The memory map is shown in Figure 11.1.

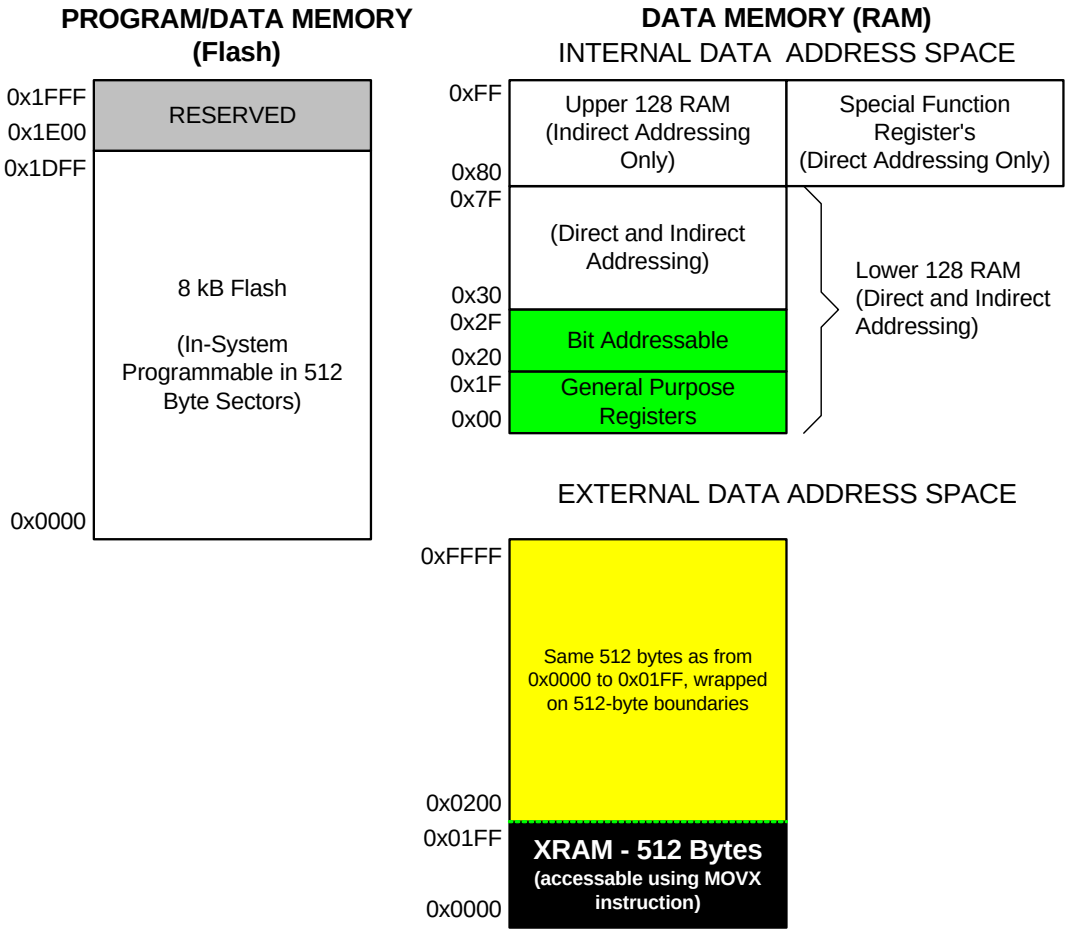



**Figure 11.1. Memory Map**

## 11.1. Program Memory

The CIP-51 core has a 64 kB program memory space. The C8051F350/1/2/3 implements 8 kB of this program memory space as in-system, re-programmable Flash memory, organized in a contiguous block from addresses 0x0000 to 0x1DFF. Addresses above 0x1DFF are reserved.

Program memory is normally assumed to be read-only. However, the C8051F350/1/2/3 can write to program memory by setting the Program Store Write Enable bit (PSCTL.0) and using the MOVX write instruction. This feature provides a mechanism for updates to program code and use of the program memory space for non-volatile data storage. Refer to Section “15. Flash Memory’ on page 121 for further details.

## 11.2. Data Memory

The C8051F350/1/2/3 includes 256 bytes of internal RAM mapped into the data memory space from 0x00 through 0xFF. The lower 128 bytes of data memory are used for general purpose registers and scratch pad memory. Either direct or indirect addressing may be used to access the lower 128 bytes of data memory. Locations 0x00 through 0x1F are addressable as four banks of general purpose registers, each bank consisting of eight byte-wide registers. The next 16 bytes, locations 0x20 through 0x2F, may either be addressed as bytes or as 128 bit locations accessible with the direct addressing mode.

The upper 128 bytes of data memory are accessible only by indirect addressing. This region occupies the same address space as the Special Function Registers (SFRs) but is physically separate from the SFR space. The addressing mode used by an instruction when accessing locations above 0x7F determines whether the CPU accesses the upper 128 bytes of data memory space or the SFRs. Instructions that use direct addressing will access the SFR space. Instructions using indirect addressing above 0x7F access the upper 128 bytes of data memory. Figure 11.1 illustrates the data memory organization of the C8051F350/1/2/3.

The C8051F35x family also includes 512 bytes of on-chip RAM mapped into the external memory (XDATA) space. This RAM can be accessed using the CIP-51 core's MOVX instruction. More information on the XRAM memory can be found in Section "16. External RAM' on page 127.

## 11.3. General Purpose Registers

The lower 32 bytes of data memory (locations 0x00 through 0x1F) may be addressed as four banks of general-purpose registers. Each bank consists of eight byte-wide registers designated R0 through R7. Only one of these banks may be enabled at a time. Two bits in the program status word, RS0 (PSW.3) and RS1 (PSW.4), select the active register bank (see description of the PSW in SFR Definition 10.4). This allows fast context switching when entering subroutines and interrupt service routines. Indirect addressing modes use registers R0 and R1 as index registers.

## 11.4. Bit Addressable Locations

In addition to direct access to data memory organized as bytes, the sixteen data memory locations at 0x20 through 0x2F are also accessible as 128 individually addressable bits. Each bit has a bit address from 0x00 to 0x7F. Bit 0 of the byte at 0x20 has bit address 0x00 while bit 7 of the byte at 0x20 has bit address 0x07. Bit 7 of the byte at 0x2F has bit address 0x7F. A bit access is distinguished from a full byte access by the type of instruction used (bit source or destination operands as opposed to a byte source or destination).

The MCS-51™ assembly language allows an alternate notation for bit addressing of the form XX.B where XX is the byte address and B is the bit position within the byte. For example, the instruction:

```
MOV    C, 22.3h
```

moves the Boolean value at 0x13 (bit 3 of the byte at location 0x22) into the Carry flag.

## 11.5. Stack

A programmer's stack can be located anywhere in the 256-byte data memory. The stack area is designated using the Stack Pointer (SP, 0x81) SFR. The SP will point to the last location used. The next value pushed on the stack is placed at SP+1 and then SP is incremented. A reset initializes the stack pointer to location 0x07. Therefore, the first value pushed on the stack is placed at location 0x08, which is also the first register (R0) of register bank 1. Thus, if more than one register bank is to be used, the SP should be initialized to a location in the data memory not being used for data storage. The stack depth can extend up to 256 bytes.

## 11.6. Special Function Registers

The direct-access data memory locations from 0x80 to 0xFF constitute the special function registers (SFRs). The SFRs provide control and data exchange with the CIP-51's resources and peripherals. The CIP-51 duplicates the SFRs found in a typical 8051 implementation as well as implementing additional SFRs used to configure and access the sub-systems unique to the MCU. This allows the addition of new functionality while retaining compatibility with the MCS-51™ instruction set. Table 11.1 lists the SFRs implemented in the CIP-51 System Controller.

The SFR registers are accessed anytime the direct addressing mode is used to access memory locations from 0x80 to 0xFF. SFRs with addresses ending in 0x0 or 0x8 (e.g. P0, TCON, IE, etc.) are bit-addressable as well as byte-addressable. All other SFRs are byte-addressable only. Unoccupied addresses in the SFR space are reserved for future use. Accessing these areas will have an indeterminate effect and should be avoided. Refer to the corresponding pages of the datasheet, as indicated in Table 11.2, for a detailed description of each register.

**Table 11.1. Special Function Register (SFR) Memory Map**

| | | | | | | | | |
|---|---|---|---|---|---|---|---|---|
| F8 | SPI0CN | PCA0L | PCA0H | ADC0CF | ADC0FL | ADC0FM | ADC0FH | VDM0CN |
| F0 | B | P0MDIN | P1MDIN | ADC0MD | ADC0CN | | EIP1 | ADC0CLK |
| E8 | ADC0STA | PCA0CPL0 | PCA0CPH0 | PCA0CPL1 | PCA0CPH1 | PCA0CPL2 | PCA0CPH2 | RSTSRC |
| E0 | ACC | XBR0 | XBR1 | PFE0CN | IT01CF | | EIE1 | |
| D8 | PCA0CN | PCA0MD | PCA0CPM0 | PCA0CPM1 | PCA0CPM2 | IDA1 | | |
| D0 | PSW | REF0CN | | | P0SKIP | P1SKIP | | IDA1CN |
| C8 | TMR2CN | | TMR2RLL | TMR2RLH | TMR2L | TMR2H | | |
| C0 | SMB0CN | SMB0CF | SMB0DAT | ADC0L | ADC0M | ADC0H | ADC0MUX | |
| B8 | IP | IDA0CN | ADC0COL | ADC0COM | ADC0COH | ADC0BUF | CKMUL | ADC0DAC |
| B0 | | OSCXCN | OSCICN | OSCICL | | | FLSCL | FLKEY |
| A8 | IE | CLKSEL | EMI0CN | ADC0CGL | ADC0CGM | ADC0CGH | | |
| A0 | P2 | SPI0CFG | SPI0CKR | SPI0DAT | P0MDOUT | P1MDOUT | P2MDOUT | |
| 98 | SCON0 | SBUF0 | ADC0DECL | ADC0DECH | CPT0CN | CPT0MD | | CPT0MX |
| 90 | P1 | TMR3CN | TMR3RLL | TMR3RLH | TMR3L | TMR3H | IDA0 | |
| 88 | TCON | TMOD | TL0 | TL1 | TH0 | TH1 | CKCON | PSCTL |
| 80 | P0 | SP | DPL | DPH | | | | PCON |
| | 0(8) (bit addressable) | 1(9) | 2(A) | 3(B) | 4(C) | 5(D) | 6(E) | 7(F) |

## Table 11.2. Special Function Registers

SFRs are listed in alphabetical order. All undefined SFR locations are reserved

| Register | Address | Description | Page |
|---|---|---|---|
| ACC | 0xE0 | Accumulator | 95 |
| ADC0BUF | 0xBD | ADC0 Buffer Control | 53 |
| ADC0CF | 0xFB | ADC0 Configuration | 49 |
| ADC0CGH | 0xAD | ADC0 Gain Calibration High | 56 |
| ADC0CGL | 0xAB | ADC0 Gain Calibration Low | 56 |
| ADC0CGM | 0xAC | ADC0 Gain Calibration Middle | 56 |
| ADC0CLK | 0xF7 | ADC0 Clock | 51 |
| ADC0CN | 0xF4 | ADC0 Control | 48 |
| ADC0COH | 0xBC | ADC0 Offset Calibration High | 55 |
| ADC0COL | 0xBA | ADC0 Offset Calibration Low | 55 |
| ADC0COM | 0xBB | ADC0 Offset Calibration Middle | 55 |
| ADC0DAC | 0xBF | ADC0 Offset DAC | 52 |
| ADC0DECH | 0x9B | ADC0 Decimation High | 51 |
| ADC0DECL | 0x9A | ADC0 Decimation Low | 52 |
| ADC0FH | 0xFE | ADC0 Fast Filter Output High | 58 |
| ADC0FL | 0xFC | ADC0 Fast Filter Output Low | 58 |
| ADC0FM | 0xFD | ADC0 Fast Filter Output Middle | 58 |
| ADC0H | 0xC5 | ADC0 Output High | 57 |
| ADC0L | 0xC3 | ADC0 Output Low | 57 |
| ADC0M | 0xC4 | ADC0 Output Middle | 57 |
| ADC0MD | 0xF3 | ADC0 Mode | 50 |
| ADC0MUX | 0xC6 | ADC0 Multiplexer | 60 |
| ADC0STA | 0xE8 | ADC0 Status | 54 |
| B | 0xF0 | B Register | 95 |
| CKCON | 0x8E | Clock Control | 201 |
| CLKMUL | 0xBE | Clock Multiplier | 135 |
| CLKSEL | 0xA9 | Clock Select | 136 |
| CPT0CN | 0x9C | Comparator0 Control | 81 |
| CPT0MD | 0x9D | Comparator0 Mode Selection | 82 |
| CPT0MX | 0x9F | Comparator0 MUX Selection | 84 |
| DPH | 0x83 | Data Pointer High | 93 |
| DPL | 0x82 | Data Pointer Low | 93 |
| EIE1 | 0xE6 | Extended Interrupt Enable 1 | 109 |
| EIP1 | 0xF6 | Extended Interrupt Priority 1 | 110 |
| EMI0CN | 0xAA | External Memory Interface Control | 127 |
| FLKEY | 0xB7 | Flash Lock and Key | 125 |
| FLSCL | 0xB6 | Flash Scale | 126 |
| IDA0 | 0x96 | Current Mode DAC0 Low | 69 |
| IDA0CN | 0xB9 | Current Mode DAC0 Control | 69 |
| IDA1 | 0xDD | Current Mode DAC1 Low | 70 |
| IDA1CN | 0xD7 | Current Mode DAC1 Control | 70 |
| IE | 0xA8 | Interrupt Enable | 107 |
| IP | 0xB8 | Interrupt Priority | 108 |
| IT01CF | 0xE4 | INT0/INT1 Configuration | 112 |

## Table 11.2. Special Function Registers (Continued)

SFRs are listed in alphabetical order. All undefined SFR locations are reserved

| Register | Address | Description | Page |
|---|---|---|---|
| OSCICL | 0xB3 | Internal Oscillator Calibration | 130 |
| OSCICN | 0xB2 | Internal Oscillator Control | 130 |
| OSCXCN | 0xB1 | External Oscillator Control | 134 |
| P0 | 0x80 | Port 0 Latch | 145 |
| P0MDIN | 0xF1 | Port 0 Input Mode Configuration | 145 |
| P0MDOUT | 0xA4 | Port 0 Output Mode Configuration | 146 |
| P0SKIP | 0xD4 | Port 0 Skip | 146 |
| P1 | 0x90 | Port 1 Latch | 147 |
| P1MDIN | 0xF2 | Port 1 Input Mode Configuration | 147 |
| P1MDOUT | 0xA5 | Port 1 Output Mode Configuration | 148 |
| P1SKIP | 0xD5 | Port 1 Skip | 148 |
| P2 | 0xA0 | Port 2 Latch | 149 |
| P2MDOUT | 0xA6 | Port 2 Output Mode Configuration | 149 |
| PCA0CN | 0xD8 | PCA Control | 222 |
| PCA0CPH0 | 0xEA | PCA Capture 0 High | 226 |
| PCA0CPH1 | 0xEC | PCA Capture 1 High | 226 |
| PCA0CPH2 | 0xEE | PCA Capture 2 High | 226 |
| PCA0CPL0 | 0xE9 | PCA Capture 0 Low | 226 |
| PCA0CPL1 | 0xEB | PCA Capture 1 Low | 226 |
| PCA0CPL2 | 0xED | PCA Capture 2 Low | 226 |
| PCA0CPM0 | 0xDA | PCA Module 0 Mode | 224 |
| PCA0CPM1 | 0xDB | PCA Module 1 Mode | 224 |
| PCA0CPM2 | 0xDC | PCA Module 2 Mode | 224 |
| PCA0H | 0xFA | PCA Counter High | 225 |
| PCA0L | 0xF9 | PCA Counter Low | 225 |
| PCA0MD | 0xD9 | PCA Mode | 223 |
| PCON | 0x87 | Power Control | 97 |
| PFE0CN | 0xE3 | Prefetch Engine Control | 113 |
| PSCTL | 0x8F | Program Store R/W Control | 125 |
| PSW | 0xD0 | Program Status Word | 94 |
| REF0CN | 0xD1 | Voltage Reference Control | 74 |
| RSTSRC | 0xEF | Reset Source Configuration/Status | 119 |
| SBUF0 | 0x99 | UART0 Data Buffer | 177 |
| SCON0 | 0x98 | UART0 Control | 176 |
| SMB0CF | 0xC1 | SMBus Configuration | 158 |
| SMB0CN | 0xC0 | SMBus Control | 160 |
| SMB0DAT | 0xC2 | SMBus Data | 162 |
| SP | 0x81 | Stack Pointer | 93 |
| SPI0CFG | 0xA1 | SPI Configuration | 187 |
| SPI0CKR | 0xA2 | SPI Clock Rate Control | 189 |
| SPI0CN | 0xF8 | SPI Control | 188 |
| SPI0DAT | 0xA3 | SPI Data | 190 |
| TCON | 0x88 | Timer/Counter Control | 199 |
| TH0 | 0x8C | Timer/Counter 0 High | 202 |
| TH1 | 0x8D | Timer/Counter 1 High | 202 |

**Table 11.2. Special Function Registers (Continued)**

SFRs are listed in alphabetical order. All undefined SFR locations are reserved

| Register | Address | Description | Page |
|---|---|---|---|
| TL0 | 0x8A | Timer/Counter 0 Low | 202 |
| TL1 | 0x8B | Timer/Counter 1 Low | 202 |
| TMOD | 0x89 | Timer/Counter Mode | 200 |
| TMR2CN | 0xC8 | Timer/Counter 2 Control | 205 |
| TMR2H | 0xCD | Timer/Counter 2 High | 206 |
| TMR2L | 0xCC | Timer/Counter 2 Low | 206 |
| TMR2RLH | 0xCB | Timer/Counter 2 Reload High | 206 |
| TMR2RLL | 0xCA | Timer/Counter 2 Reload Low | 206 |
| TMR3CN | 0x91 | Timer/Counter 3Control | 209 |
| TMR3H | 0x95 | Timer/Counter 3 High | 210 |
| TMR3L | 0x94 | Timer/Counter 3 Low | 210 |
| TMR3RLH | 0x93 | Timer/Counter 3 Reload High | 210 |
| TMR3RLL | 0x92 | Timer/Counter 3 Reload Low | 210 |
| VDM0CN | 0xFF | $V_{DD}$ Monitor Control | 117 |
| XBR0 | 0xE1 | Port I/O Crossbar Control 0 | 142 |
| XBR1 | 0xE2 | Port I/O Crossbar Control 1 | 143 |

# 12. Interrupt Handler

The C8051F35x family includes an extended interrupt system supporting a total of 12 interrupt sources with two priority levels. The allocation of interrupt sources between on-chip peripherals and external input pins varies according to the specific version of the device. Each interrupt source has one or more associated interrupt-pending flag(s) located in an SFR. When a peripheral or external source meets a valid interrupt condition, the associated interrupt-pending flag is set to logic 1.

If interrupts are enabled for the source, an interrupt request is generated when the interrupt-pending flag is set. As soon as execution of the current instruction is complete, the CPU generates an LCALL to a predetermined address to begin execution of an interrupt service routine (ISR). Each ISR must end with an RETI instruction, which returns program execution to the next instruction that would have been executed if the interrupt request had not occurred. If interrupts are not enabled, the interrupt-pending flag is ignored by the hardware and program execution continues as normal. (The interrupt-pending flag is set to logic 1 regardless of the interrupt's enable/disable state.)

Each interrupt source can be individually enabled or disabled through the use of an associated interrupt enable bit in the Interrupt Enable and Extended Interrupt Enable SFRs. However, interrupts must first be globally enabled by setting the EA bit (IE.7) to logic 1 before the individual interrupt enables are recognized. Setting the EA bit to logic 0 disables all interrupt sources regardless of the individual interrupt-enable settings. Note that interrupts which occur when the EA bit is set to logic 0 will be held in a pending state, and will not be serviced until the EA bit is set back to logic 1.

Some interrupt-pending flags are automatically cleared by the hardware when the CPU vectors to the ISR. However, most are not cleared by the hardware and must be cleared by software before returning from the ISR. If an interrupt-pending flag remains set after the CPU completes the return-from-interrupt (RETI) instruction, a new interrupt request will be generated immediately and the CPU will re-enter the ISR after the completion of the next instruction.

## 12.1. MCU Interrupt Sources and Vectors

The MCUs support 12 interrupt sources. Software can simulate an interrupt by setting any interrupt-pending flag to logic 1. If interrupts are enabled for the flag, an interrupt request will be generated and the CPU will vector to the ISR address associated with the interrupt-pending flag. MCU interrupt sources, associated vector addresses, priority order and control bits are summarized in Table 12.1 on page 106. Refer to the datasheet section associated with a particular on-chip peripheral for information regarding valid interrupt conditions for the peripheral and the behavior of its interrupt-pending flag(s).

## 12.2. Interrupt Priorities

Each interrupt source can be individually programmed to one of two priority levels: low or high. A low priority interrupt service routine can be preempted by a high priority interrupt. A high priority interrupt cannot be preempted. Each interrupt has an associated interrupt priority bit in an SFR (IP or EIP1) used to configure its priority level. Low priority is the default. If two interrupts are recognized simultaneously, the interrupt with the higher priority is serviced first. If both interrupts have the same priority level, a fixed priority order is used to arbitrate, given in Table 12.1.

## 12.3. Interrupt Latency

Interrupt response time depends on the state of the CPU when the interrupt occurs. Pending interrupts are sampled and priority decoded each system clock cycle. Therefore, the fastest possible response time is 5 system clock cycles: 1 clock cycle to detect the interrupt and 4 clock cycles to complete the LCALL to the ISR. If an interrupt is pending when a RETI is executed, a single instruction is executed before an LCALL

is made to service the pending interrupt. Therefore, the maximum response time for an interrupt (when no other interrupt is currently being serviced or the new interrupt is of greater priority) occurs when the CPU is performing an RETI instruction followed by a DIV as the next instruction. In this case, the response time is 18 system clock cycles: 1 clock cycle to detect the interrupt, 5 clock cycles to execute the RETI, 8 clock cycles to complete the DIV instruction and 4 clock cycles to execute the LCALL to the ISR. If the CPU is executing an ISR for an interrupt with equal or higher priority, the new interrupt will not be serviced until the current ISR completes, including the RETI and following instruction.

**Table 12.1. Interrupt Summary**

| Interrupt Source | Interrupt Vector | Priority Order | Pending Flag | Bit addressable? | Cleared by HW? | Enable Flag | Priority Control |
|---|---|---|---|---|---|---|---|
| Reset | 0x0000 | Top | None | N/A | N/A | Always Enabled | Always Highest |
| External Interrupt 0 (/INT0) | 0x0003 | 0 | IE0 (TCON.1) | Y | Y | EX0 (IE.0) | PX0 (IP.0) |
| Timer 0 Overflow | 0x000B | 1 | TF0 (TCON.5) | Y | Y | ET0 (IE.1) | PT0 (IP.1) |
| External Interrupt 1 (/INT1) | 0x0013 | 2 | IE1 (TCON.3) | Y | Y | EX1 (IE.2) | PX1 (IP.2) |
| Timer 1 Overflow | 0x001B | 3 | TF1 (TCON.7) | Y | Y | ET1 (IE.3) | PT1 (IP.3) |
| UART0 | 0x0023 | 4 | RI0 (SCON0.0)<br>TI0 (SCON0.1) | Y | N | ES0 (IE.4) | PS0 (IP.4) |
| Timer 2 Overflow | 0x002B | 5 | TF2H (TMR2CN.7)<br>TF2L (TMR2CN.6) | Y | N | ET2 (IE.5) | PT2 (IP.5) |
| SPI0 | 0x0033 | 6 | SPIF (SPI0CN.7)<br>WCOL (SPI0CN.6)<br>MODF (SPI0CN.5)<br>RXOVRN (SPI0CN.4) | Y | N | ESPI0 (IE.6) | PSPI0 (IP.6) |
| SMB0 | 0x003B | 7 | SI (SMB0CN.0) | Y | N | ESMB0 (EIE1.0) | PSMB0 (EIP1.0) |
| RESERVED | 0x0043 | 8 | N/A | N/A | N/A | N/A | N/A |
| RESERVED | 0x004B | 9 | N/A | N/A | N/A | N/A | N/A |
| ADC0 | 0x0053 | 10 | AD0INT (ADC0STA.5) | Y | N | EADC0 (EIE1.3) | PADC0 (EIP1.3) |
| Programmable Counter Array | 0x005B | 11 | CF (PCA0CN.7)<br>CCFn (PCA0CN.n) | Y | N | EPCA0 (EIE1.4) | PPCA0 (EIP1.4) |
| Comparator0 | 0x0063 | 12 | CP0FIF (CPT0CN.4)<br>CP0RIF (CPT0CN.5) | N | N | ECP0 (EIE1.5) | PCP0 (EIP1.5) |
| RESERVED | 0x006B | 13 | N/A | N/A | N/A | N/A | N/A |
| Timer 3 Overflow | 0x0073 | 14 | TF3H (TMR3CN.7)<br>TF3L (TMR3CN.6) | N | N | ET3 (EIE1.7) | PT3 (EIP1.7) |

## 12.4. Interrupt Register Descriptions

The SFRs used to enable the interrupt sources and set their priority level are described below. Refer to the datasheet section associated with a particular on-chip peripheral for information regarding valid interrupt conditions for the peripheral and the behavior of its interrupt-pending flag(s).

### SFR Definition 12.1. IE: Interrupt Enable

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| EA | ESPI0 | ET2 | ES0 | ET1 | EX1 | ET0 | EX0 | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | Bit Addressable |

SFR Address: 0xA8

Bit 7: EA: Enable All Interrupts.
This bit globally enables/disables all interrupts. It overrides the individual interrupt mask settings.
0: Disable all interrupt sources.
1: Enable each interrupt according to its individual mask setting.

Bit 6: ESPI0: Enable Serial Peripheral Interface (SPI0) Interrupt.
This bit sets the masking of the SPI0 interrupts.
0: Disable all SPI0 interrupts.
1: Enable interrupt requests generated by SPI0.

Bit 5: ET2: Enable Timer 2 Interrupt.
This bit sets the masking of the Timer 2 interrupt.
0: Disable Timer 2 interrupt.
1: Enable interrupt requests generated by the TF2L or TF2H flags.

Bit 4: ES0: Enable UART0 Interrupt.
This bit sets the masking of the UART0 interrupt.
0: Disable UART0 interrupt.
1: Enable UART0 interrupt.

Bit 3: ET1: Enable Timer 1 Interrupt.
This bit sets the masking of the Timer 1 interrupt.
0: Disable all Timer 1 interrupt.
1: Enable interrupt requests generated by the TF1 flag.

Bit 2: EX1: Enable External Interrupt 1.
This bit sets the masking of External Interrupt 1.
0: Disable external interrupt 1.
1: Enable interrupt requests generated by the /INT1 input.

Bit 1: ET0: Enable Timer 0 Interrupt.
This bit sets the masking of the Timer 0 interrupt.
0: Disable all Timer 0 interrupt.
1: Enable interrupt requests generated by the TF0 flag.

Bit 0: EX0: Enable External Interrupt 0.
This bit sets the masking of External Interrupt 0.
0: Disable external interrupt 0.
1: Enable interrupt requests generated by the /INT0 input.

## SFR Definition 12.2. IP: Interrupt Priority

| R | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| — | PSPI0 | PT2 | PS0 | PT1 | PX1 | PT0 | PX0 | 10000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | Bit Addressable |

SFR Address: 0xB8

Bit 7: UNUSED. Read = 1, Write = don't care.
Bit 6: PSPI0: Serial Peripheral Interface (SPI0) Interrupt Priority Control.
This bit sets the priority of the SPI0 interrupt.
0: SPI0 interrupt set to low priority level.
1: SPI0 interrupt set to high priority level.
Bit 5: PT2: Timer 2 Interrupt Priority Control.
This bit sets the priority of the Timer 2 interrupt.
0: Timer 2 interrupt set to low priority level.
1: Timer 2 interrupt set to high priority level.
Bit 4: PS0: UART0 Interrupt Priority Control.
This bit sets the priority of the UART0 interrupt.
0: UART0 interrupt set to low priority level.
1: UART0 interrupt set to high priority level.
Bit 3: PT1: Timer 1 Interrupt Priority Control.
This bit sets the priority of the Timer 1 interrupt.
0: Timer 1 interrupt set to low priority level.
1: Timer 1 interrupt set to high priority level.
Bit 2: PX1: External Interrupt 1 Priority Control.
This bit sets the priority of the External Interrupt 1 interrupt.
0: External Interrupt 1 set to low priority level.
1: External Interrupt 1 set to high priority level.
Bit 1: PT0: Timer 0 Interrupt Priority Control.
This bit sets the priority of the Timer 0 interrupt.
0: Timer 0 interrupt set to low priority level.
1: Timer 0 interrupt set to high priority level.
Bit 0: PX0: External Interrupt 0 Priority Control.
This bit sets the priority of the External Interrupt 0 interrupt.
0: External Interrupt 0 set to low priority level.
1: External Interrupt 0 set to high priority level.

## SFR Definition 12.3. EIE1: Extended Interrupt Enable 1

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| ET3 | Reserved | ECP0 | EPCA0 | EADC0 | Reserved | Reserved | ESMB0 | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xE6

Bit 7: ET3: Enable Timer 3 Interrupt.
This bit sets the masking of the Timer 3 interrupt.
0: Disable Timer 3 interrupts.
1: Enable interrupt requests generated by the TF3L or TF3H flags.

Bit 6: RESERVED. Read = 0. Must Write 0.

Bit 5: ECP0: Enable Comparator0 (CP0) Interrupt.
This bit sets the masking of the CP0 interrupt.
0: Disable CP0 interrupts.
1: Enable interrupt requests generated by the CP0RIF or CP0FIF flags.

Bit 4: EPCA0: Enable Programmable Counter Array (PCA0) Interrupt.
This bit sets the masking of the PCA0 interrupts.
0: Disable all PCA0 interrupts.
1: Enable interrupt requests generated by PCA0.

Bit 3: EADC0: Enable ADC0 Conversion Complete Interrupt.
This bit sets the masking of the ADC0 Conversion Complete interrupt.
0: Disable ADC0 Conversion Complete interrupt.
1: Enable interrupt requests generated by the AD0INT flag.

Bits 2–1: RESERVED. Read = 00. Must Write 00.

Bit 0: ESMB0: Enable SMBus (SMB0) Interrupt.
This bit sets the masking of the SMB0 interrupt.
0: Disable all SMB0 interrupts.
1: Enable interrupt requests generated by SMB0.

## SFR Definition 12.4. EIP1: Extended Interrupt Priority 1

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| PT3 | Reserved | PCP0 | PPCA0 | PADC0 | Reserved | Reserved | PSMB0 | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xF6

Bit 7: PT3: Timer 3 Interrupt Priority Control.
This bit sets the priority of the Timer 3 interrupt.
0: Timer 3 interrupts set to low priority level.
1: Timer 3 interrupts set to high priority level.

Bit 6: RESERVED. Read = 0. Must Write 0.

Bit 5: PCP0: Comparator0 (CP0) Interrupt Priority Control.
This bit sets the priority of the CP0 interrupt.
0: CP0 interrupt set to low priority level.
1: CP0 interrupt set to high priority level.

Bit 4: PPCA0: Programmable Counter Array (PCA0) Interrupt Priority Control.
This bit sets the priority of the PCA0 interrupt.
0: PCA0 interrupt set to low priority level.
1: PCA0 interrupt set to high priority level.

Bit 3: PADC0 ADC0 Conversion Complete Interrupt Priority Control.
This bit sets the priority of the ADC0 Conversion Complete interrupt.
0: ADC0 Conversion Complete interrupt set to low priority level.
1: ADC0 Conversion Complete interrupt set to high priority level.

Bits 2–1: RESERVED. Read = 00. Must Write 00.

Bit 0: PSMB0: SMBus (SMB0) Interrupt Priority Control.
This bit sets the priority of the SMB0 interrupt.
0: SMB0 interrupt set to low priority level.
1: SMB0 interrupt set to high priority level.

## 12.5. External Interrupts

The /INT0 and /INT1 external interrupt sources are configurable as active high or low, edge or level sensitive. The IN0PL (/INT0 Polarity) and IN1PL (/INT1 Polarity) bits in the IT01CF register select active high or active low; the IT0 and IT1 bits in TCON (Section “22.1. Timer 0 and Timer 1’ on page 195) select level or edge sensitive. The table below lists the possible configurations.

| IT0 | IN0PL | /INT0 Interrupt |
|---|---|---|
| 1 | 0 | Active low, edge sensitive |
| 1 | 1 | Active high, edge sensitive |
| 0 | 0 | Active low, level sensitive |
| 0 | 1 | Active high, level sensitive |

| IT1 | IN1PL | /INT1 Interrupt |
|---|---|---|
| 1 | 0 | Active low, edge sensitive |
| 1 | 1 | Active high, edge sensitive |
| 0 | 0 | Active low, level sensitive |
| 0 | 1 | Active high, level sensitive |

/INT0 and /INT1 are assigned to Port pins as defined in the IT01CF register (see SFR Definition 12.5). Note that /INT0 and /INT0 Port pin assignments are independent of any Crossbar assignments. /INT0 and /INT1 will monitor their assigned Port pins without disturbing the peripheral that was assigned the Port pin via the Crossbar. To assign a Port pin only to /INT0 and/or /INT1, configure the Crossbar to skip the selected pin(s). This is accomplished by setting the associated bit in register XBR0 (see Section “18.1. Priority Crossbar Decoder’ on page 139 for complete details on configuring the Crossbar).

IE0 (TCON.1) and IE1 (TCON.3) serve as the interrupt-pending flags for the /INT0 and /INT1 external interrupts, respectively. If an /INT0 or /INT1 external interrupt is configured as edge-sensitive, the corresponding interrupt-pending flag is automatically cleared by the hardware when the CPU vectors to the ISR. When configured as level sensitive, the interrupt-pending flag remains logic 1 while the input is active as defined by the corresponding polarity bit (IN0PL or IN1PL); the flag remains logic 0 while the input is inactive. The external interrupt source must hold the input active until the interrupt request is recognized. It must then deactivate the interrupt request before execution of the ISR completes or another interrupt request will be generated.

## SFR Definition 12.5. IT01CF: INT0/INT1 Configuration

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| IN1PL | IN1SL2 | IN1SL1 | IN1SL0 | IN0PL | IN0SL2 | IN0SL1 | IN0SL0 | 00000001 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xE4

**Note:** Refer to SFR Definition 22.1 for INT0/1 edge- or level-sensitive interrupt selection.

Bit 7: IN1PL: /INT1 Polarity
0: /INT1 input is active low.
1: /INT1 input is active high.

Bits 6–4: IN1SL2–0: /INT1 Port Pin Selection Bits
These bits select which Port pin is assigned to /INT1. Note that this pin assignment is independent of the Crossbar; /INT1 will monitor the assigned Port pin without disturbing the peripheral that has been assigned the Port pin via the Crossbar. The Crossbar will not assign the Port pin to a peripheral if it is configured to skip the selected pin (accomplished by setting to '1' the corresponding bit in register P0SKIP).

| IN1SL2–0 | /INT1 Port Pin |
|---|---|
| 000 | P0.0 |
| 001 | P0.1 |
| 010 | P0.2 |
| 011 | P0.3 |
| 100 | P0.4 |
| 101 | P0.5 |
| 110 | P0.6 |
| 111 | P0.7 |

Bit 3: IN0PL: /INT0 Polarity
0: /INT0 interrupt is active low.
1: /INT0 interrupt is active high.

Bits 2–0: INT0SL2–0: /INT0 Port Pin Selection Bits
These bits select which Port pin is assigned to /INT0. Note that this pin assignment is independent of the Crossbar. /INT0 will monitor the assigned Port pin without disturbing the peripheral that has been assigned the Port pin via the Crossbar. The Crossbar will not assign the Port pin to a peripheral if it is configured to skip the selected pin (accomplished by setting to '1' the corresponding bit in register P0SKIP).

| IN0SL2–0 | /INT0 Port Pin |
|---|---|
| 000 | P0.0 |
| 001 | P0.1 |
| 010 | P0.2 |
| 011 | P0.3 |
| 100 | P0.4 |
| 101 | P0.5 |
| 110 | P0.6 |
| 111 | P0.7 |

# 17. Oscillators

C8051F350/1/2/3 devices include a programmable internal oscillator, an external oscillator drive circuit, and a clock multiplier. The internal oscillator can be enabled/disabled and calibrated using the OSCICN and OSCICL registers, as shown in Figure 17.1. The system clock (SYSCLK) can be derived from the internal oscillator, external oscillator circuit, or the clock multiplier. The clock multiplier can produce three possible outputs: Internal Oscillator x 2, External Oscillator x 2, or External Oscillator x 4. Oscillator electrical specifications are given in Table 17.1 on page 136.

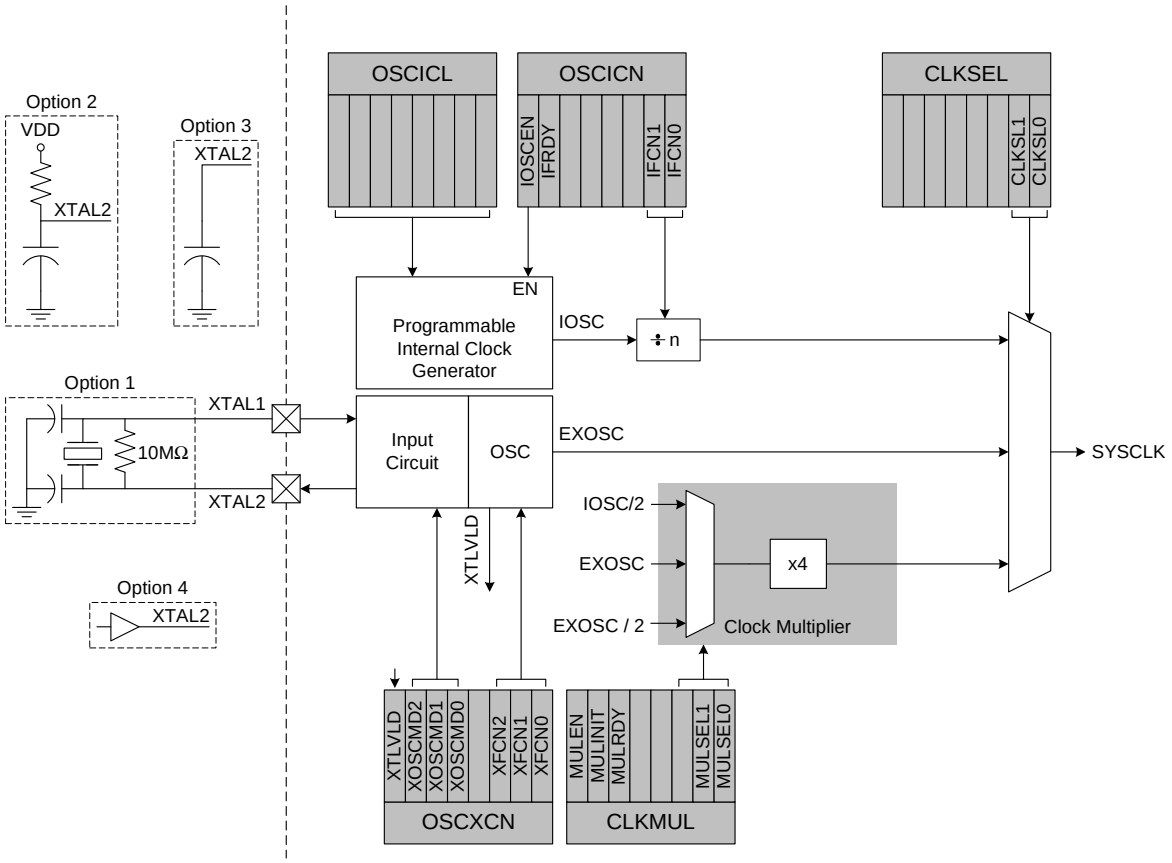



**Figure 17.1. Oscillator Diagram**

## 17.1. Programmable Internal Oscillator

All C8051F350/1/2/3 devices include a programmable internal oscillator that defaults as the system clock after a system reset. The internal oscillator period can be programmed via the OSCICL register, shown in SFR Definition 17.2. On C8051F350/1/2/3 devices, OSCICL is factory calibrated to obtain a 24.5 MHz frequency.

Electrical specifications for the precision internal oscillator are given in Table 17.1 on page 136. Note that the system clock may be derived from the programmed internal oscillator divided by 1, 2, 4, or 8, as defined by the IFCN bits in register OSCICN. The divide value defaults to 8 following a reset.

## SFR Definition 17.1. OSCICN: Internal Oscillator Control

| R/W | R | R | R | R | R | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| IOSCEN | IFRDY | — | — | — | — | IFCN1 | IFCN0 | 11000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xB2

Bit7: IOSCEN: Internal Oscillator Enable Bit.
0: Internal Oscillator Disabled.
1: Internal Oscillator Enabled.
Bit6: IFRDY: Internal Oscillator Frequency Ready Flag.
0: Internal Oscillator is not running at programmed frequency.
1: Internal Oscillator is running at programmed frequency.
Bits5–2: UNUSED. Read = 0000b, Write = don't care.
Bits1–0: IFCN1–0: Internal Oscillator Frequency Control Bits.
00: SYSCLK derived from Internal Oscillator divided by 8.
01: SYSCLK derived from Internal Oscillator divided by 4.
10: SYSCLK derived from Internal Oscillator divided by 2.
11: SYSCLK derived from Internal Oscillator divided by 1.

## SFR Definition 17.2. OSCICL: Internal Oscillator Calibration

| R | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| | | | | | | | | Variable |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xB3

Bit7: UNUSED. Read = 0. Write = don’t care.
Bits 6–0: OSCICL: Internal Oscillator Calibration Register.
This register determines the internal oscillator period. On C8051F350/1/2/3 devices, the reset value is factory calibrated to generate an internal oscillator frequency of 24.5 MHz.

## 17.2. External Oscillator Drive Circuit

The external oscillator circuit may drive an external crystal, ceramic resonator, capacitor, or RC network. A CMOS clock may also provide a clock input. For a crystal or ceramic resonator configuration, the crystal/resonator must be wired across the XTAL1 and XTAL2 pins as shown in Option 1 of Figure 17.1. A 10 MΩ resistor also must be wired across the XTAL1 and XTAL2 pins for the crystal/resonator configuration. In RC, capacitor, or CMOS clock configuration, the clock source should be wired to the XTAL2 pin as shown in Option 2, 3, or 4 of Figure 17.1. The type of external oscillator must be selected in the OSCXCN register, and the frequency control bits (XFCN) must be selected appropriately (see SFR Definition 17.3)

**Important Note on External Oscillator Usage:** Port pins must be configured when using the external oscillator circuit. When the external oscillator drive circuit is enabled in crystal/resonator mode, Port pins P0.2 and P0.3 are used as XTAL1 and XTAL2 respectively. When the external oscillator drive circuit is enabled in capacitor, RC, or CMOS clock mode, Port pin P0.3 is used as XTAL2. The Port I/O Crossbar should be configured to skip the Port pins used by the oscillator circuit; see Section “18.1. Priority Crossbar Decoder’ on page 139 for Crossbar configuration. Additionally, when using the external oscillator circuit in crystal/resonator, capacitor, or RC mode, the associated Port pins should be configured as analog inputs. In CMOS clock mode, the associated pin should be configured as a **digital input**. See Section “18.2. Port I/O Initialization’ on page 141 for details on Port input mode selection.

### 17.2.1. Clocking Timers Directly Through the External Oscillator

The external oscillator source divided by eight is a clock option for the timers (Section “22. Timers’ on page 195) and the Programmable Counter Array (PCA) (Section “23. Programmable Counter Array’ on page 211). When the external oscillator is used to clock these peripherals, but is not used as the system clock, the external oscillator frequency must be less than or equal to the system clock frequency. In this configuration, the clock supplied to the peripheral (external oscillator / 8) is synchronized with the system clock; the jitter associated with this synchronization is limited to ±0.5 system clock cycles.

### 17.2.2. External Crystal Example

If a crystal or ceramic resonator is used as an external oscillator source for the MCU, the circuit should be configured as shown in Figure 17.1, Option 1. The External Oscillator Frequency Control value (XFCN) should be chosen from the Crystal column of the table in SFR Definition 17.3 (OSCXCN register). For example, a 12 MHz crystal requires an XFCN setting of 111b.

When the crystal oscillator is first enabled, the oscillator amplitude detection circuit requires a settling time to achieve proper bias. Introducing a delay of 1 ms between enabling the oscillator and checking the XTLVLD bit will prevent a premature switch to the external oscillator as the system clock. Switching to the external oscillator before the crystal oscillator has stabilized can result in unpredictable behavior. The recommended procedure is:

Step 1. Force the XTAL1 and XTAL2 pins low by writing 0’s to the port latch.
Step 2. Configure XTAL1 and XTAL2 as analog inputs.
Step 3. Enable the external oscillator.
Step 4. Wait at least 1 ms.
Step 5. Poll for XTLVLD => ‘1’.
Step 6. Switch the system clock to the external oscillator.

Note: Tuning-fork crystals may require additional settling time before XTLVLD returns a valid result.

The capacitors shown in the external crystal configuration provide the load capacitance required by the crystal for correct oscillation. These capacitors are “in series” as seen by the crystal and “in parallel” with the stray capacitance of the XTAL1 and XTAL2 pins.

Note: The load capacitance depends upon the crystal and the manufacturer. Please refer to the crystal datasheet when completing these calculations.

For example, a tuning-fork crystal of 32.768 kHz with a recommended load capacitance of 12.5 pF should use the configuration shown in Figure 17.1, Option 1. The total value of the capacitors and the stray capacitance of the XTAL pins should equal 25 pF. With a stray capacitance of 3 pF per pin, the 22 pF capacitors yield an equivalent capacitance of 12.5 pF across the crystal, as shown in Figure 17.2.

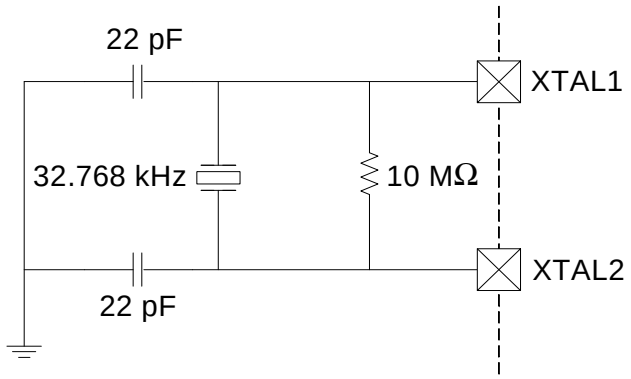



**Figure 17.2. 32.768 kHz External Crystal Example**

**Important note on external crystals:** Crystal oscillator circuits are quite sensitive to PCB layout. The crystal should be placed as close as possible to the XTAL pins on the device. The traces should be as short as possible and shielded with ground plane from any other traces which could introduce noise or interference.

### 17.2.3. External RC Example

If an RC network is used as an external oscillator source for the MCU, the circuit should be configured as shown in Figure 17.1, Option 2. The capacitor should be no greater than 100 pF; however for very small capacitors, the total capacitance may be dominated by parasitic capacitance in the PCB layout. To determine the required External Oscillator Frequency Control value (XFCN) in the OSCXCN Register, first select the RC network value to produce the desired frequency of oscillation. If the frequency desired is 100 kHz, let R = 246 kΩ and C = 50 pF:

f = 1.23( $10^3$ ) / RC = 1.23 ( $10^3$ ) / [ 246 * 50 ] = 0.1 MHz = 100 kHz

Referring to the table in SFR Definition 17.3, the required XFCN setting is 010b. Programming XFCN to a higher setting in RC mode will improve frequency accuracy at an increased external oscillator supply current.

### 17.2.4. External Capacitor Example

If a capacitor is used as an external oscillator for the MCU, the circuit should be configured as shown in Figure 17.1, Option 3. The capacitor should be no greater than 100 pF; however for very small capacitors, the total capacitance may be dominated by parasitic capacitance in the PCB layout. To determine the required External Oscillator Frequency Control value (XFCN) in the OSCXCN Register, select the capacitor to be used and find the frequency of oscillation from the equations below. Assume $V_{DD}$ = 3.0 V and f = 150 kHz:

f = KF / (C x $V_{DD}$)
0.150 MHz = KF / (C x 3.0)

Since the frequency of roughly 150 kHz is desired, select the K Factor from the table in SFR Definition 17.3 as KF = 22:

0.150 MHz = 22 / (C x 3.0)
C x 3.0 = 22 / 0.150 MHz

C = 146.6 / 3.0 pF = 48.8 pF

Therefore, the XFCN value to use in this example is 011b and C = 50 pF.

## SFR Definition 17.3. OSCXCN: External Oscillator Control

| R | R/W | R/W | R/W | R | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| XTLVLD | XOSCMD2 | XOSCMD1 | XOSCMD0 | — | XFCN2 | XFCN1 | XFCN0 | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xB1

Bit7: XTLVLD: Crystal Oscillator Valid Flag.
(Read only when XOSCMD = 11x.)
0: Crystal Oscillator is unused or not yet stable.
1: Crystal Oscillator is running and stable.

Bits6–4: XOSCMD2–0: External Oscillator Mode Bits.
00x: External Oscillator circuit off.
010: External CMOS Clock Mode.
011: External CMOS Clock Mode with divide by 2 stage.
100: RC Oscillator Mode.
101: Capacitor Oscillator Mode.
110: Crystal Oscillator Mode.
111: Crystal Oscillator Mode with divide by 2 stage.

Bit3: RESERVED. Read = 0, Write = don't care.

Bits2–0: XFCN2–0: External Oscillator Frequency Control Bits.
000–111: See table below:

| XFCN | Crystal (XOSCMD = 11x) | RC (XOSCMD = 10x) | C (XOSCMD = 10x) |
|---|---|---|---|
| 000 | f ≤ 32 kHz | f ≤ 25 kHz | K Factor = 0.87 |
| 001 | 32 kHz < f ≤ 84 kHz | 25 kHz < f ≤ 50 kHz | K Factor = 2.6 |
| 010 | 84 kHz < f ≤ 225 kHz | 50 kHz < f ≤ 100 kHz | K Factor = 7.7 |
| 011 | 225 kHz < f ≤ 590 kHz | 100 kHz < f ≤ 200 kHz | K Factor = 22 |
| 100 | 590 kHz < f ≤ 1.5 MHz | 200 kHz < f ≤ 400 kHz | K Factor = 65 |
| 101 | 1.5 MHz < f ≤ 4 MHz | 400 kHz < f ≤ 800 kHz | K Factor = 180 |
| 110 | 4 MHz < f ≤ 10 MHz | 800 kHz < f ≤ 1.6 MHz | K Factor = 664 |
| 111 | 10 MHz < f ≤ 30 MHz | 1.6 MHz < f ≤ 3.2 MHz | K Factor = 1590 |

**CRYSTAL MODE** (Circuit from Figure 17.1, Option 1; XOSCMD = 11x)
Choose XFCN value to match crystal or resonator frequency.

**RC MODE** (Circuit from Figure 17.1, Option 2; XOSCMD = 10x)
Choose XFCN value to match frequency range:
**$f = 1.23(10^3) / (R * C)$**, where
f = frequency of clock in MHz
C = capacitor value in pF
R = Pull-up resistor value in kΩ

**C MODE** (Circuit from Figure 17.1, Option 3; XOSCMD = 10x)
Choose K Factor (KF) for the oscillation frequency desired:
**$f = KF / (C * V_{DD})$**, where
f = frequency of clock in MHz
C = capacitor value the XTAL2 pin in pF
$V_{DD}$ = Power Supply on MCU in volts

## 17.3. Clock Multiplier

The Clock Multiplier generates an output clock which is 4 times the input clock frequency. The Clock Multiplier's input can be selected from the external oscillator, or 1/2 the internal or external oscillators. This produces three possible outputs: Internal Oscillator x 2, External Oscillator x 2, or External Oscillator x 4. See Section "17.4. System Clock Selection' on page 136 for details on system clock selection.

The Clock Multiplier is configured via the CLKMUL register (SFR Definition 17.4). The procedure for configuring and enabling the Clock Multiplier is as follows:

1. Reset the Multiplier by writing 0x00 to register CLKMUL.
2. Select the Multiplier input source via the MULSEL bits.
3. Enable the Multiplier with the MULEN bit (CLKMUL | = 0x80).
4. Delay for >5 µs.
5. Initialize the Multiplier with the MULINIT bit (CLKMUL | = 0xC0).
6. Poll for MULRDY => '1'.

**Important Note: When using an external oscillator as the input to the Clock Multiplier, the external source must be enabled and stable before the Multiplier is initialized. See Section "17.4. System Clock Selection' on page 136 for details on selecting an external oscillator source.**

### SFR Definition 17.4. CLKMUL: Clock Multiplier Control

| R/W | R/W | R | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| MULEN | MULINIT | MULRDY | — | — | — | MULSEL | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xBE

Bit7: MULEN: Clock Multiplier Enable
0: Clock Multiplier disabled.
1: Clock Multiplier enabled.

Bit6: MULINIT: Clock Multiplier Initialize
This bit should be a '0' when the Clock Multiplier is enabled. Once enabled, writing a '1' to this bit will initialize the Clock Multiplier. The MULRDY bit reads '1' when the Clock Multiplier is stabilized.

Bit5: MULRDY: Clock Multiplier Ready
This read-only bit indicates the status of the Clock Multiplier.
0: Clock Multiplier not ready.
1: Clock Multiplier ready (locked).

Bits4–2: Unused. Read = 000b; Write = don't care.

Bits1–0: MULSEL: Clock Multiplier Input Select
These bits select the clock supplied to the Clock Multiplier.

| MULSEL | Selected Input Clock | Clock Multipler Output |
|---|---|---|
| 00 | Internal Oscillator / 2 | Internal Oscillator x 2 |
| 01 | External Oscillator | External Oscillator x 4 |
| 10 | External Oscillator / 2 | External Oscillator x 2 |
| 11 | RESERVED | RESERVED |

## 17.4. System Clock Selection

The internal oscillator requires little start-up time and may be selected as the system clock immediately following the OSCICN write that enables the internal oscillator. External crystals and ceramic resonators typically require a start-up time before they are settled and ready for use. The Crystal Valid Flag (XTLVLD in register OSCXCN) is set to '1' by hardware when the external oscillator is settled. **To avoid reading a false XTLVLD, in crystal mode software should delay at least 1 ms between enabling the external oscillator and checking XTLVLD.** RC and C modes typically require no startup time.

The CLKSL[1:0] bits in register CLKSEL select which oscillator source is used as the system clock. CLKSL[1:0] must be set to 01b for the system clock to run from the external oscillator; however the external oscillator may still clock certain peripherals (timers, PCA) when the internal oscillator is selected as the system clock. The system clock may be switched on-the-fly between the internal oscillator, external oscillator, and Clock Multiplier so long as the selected clock source is enabled and has settled.

### SFR Definition 17.5. CLKSEL: Clock Select

| R | R | R | R | R | R | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| — | — | — | — | — | — | CLKSL | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xA9

Bits7–2: Unused. Read = 000000b; Write = don't care.
Bits1–0: CLKSL1–0: System Clock Select
These bits select the system clock source.

| CLKSL | Selected Clock |
|---|---|
| 00 | Internal Oscillator (as determined by the IFCN bits in register OSCICN) |
| 01 | External Oscillator |
| 10 | Clock Multiplier |
| 11 | RESERVED |

### Table 17.1. Oscillator Electrical Characteristics

–40 to +85 °C unless otherwise specified.

| Parameter | Conditions | Min | Typ | Max | Units |
|---|---|---|---|---|---|
| Internal Oscillator Frequency | Reset Frequency | 24 | 24.5 | 25 | MHz |
| Internal Oscillator Supply Current (from $V_{DD}$) | OSCICN.7 = 1 | — | 450 | — | µA |

## 18. Port Input/Output

Digital and analog resources are available through 17 I/O pins. Port pins are organized as two byte-wide Ports and one 1-bit Port. Each of the Port pins can be defined as general-purpose I/O (GPIO) or analog input/output; Port pins P0.0 - P1.7 can be assigned to one of the internal digital resources as shown in Figure 18.3. The designer has complete control over which functions are assigned, limited only by the number of physical I/O pins. This resource assignment flexibility is achieved through the use of a Priority Crossbar Decoder. Note that the state of a Port I/O pin can always be read in the corresponding Port latch, regardless of the Crossbar settings.

The Crossbar assigns the selected internal digital resources to the I/O pins based on the Priority Decoder (Figure 18.3 and Figure 18.4). The registers XBR0 and XBR1, defined in SFR Definition 18.1 and SFR Definition 18.2, are used to select internal digital functions.

All Port I/Os are 5 V tolerant (refer to Figure 18.2 for the Port cell circuit). The Port I/O cells are configured as either push-pull or open-drain in the Port Output Mode registers (PnMDOUT, where n = 0,1,2). Complete Electrical Specifications for Port I/O are given in Table 18.1 on page 150.

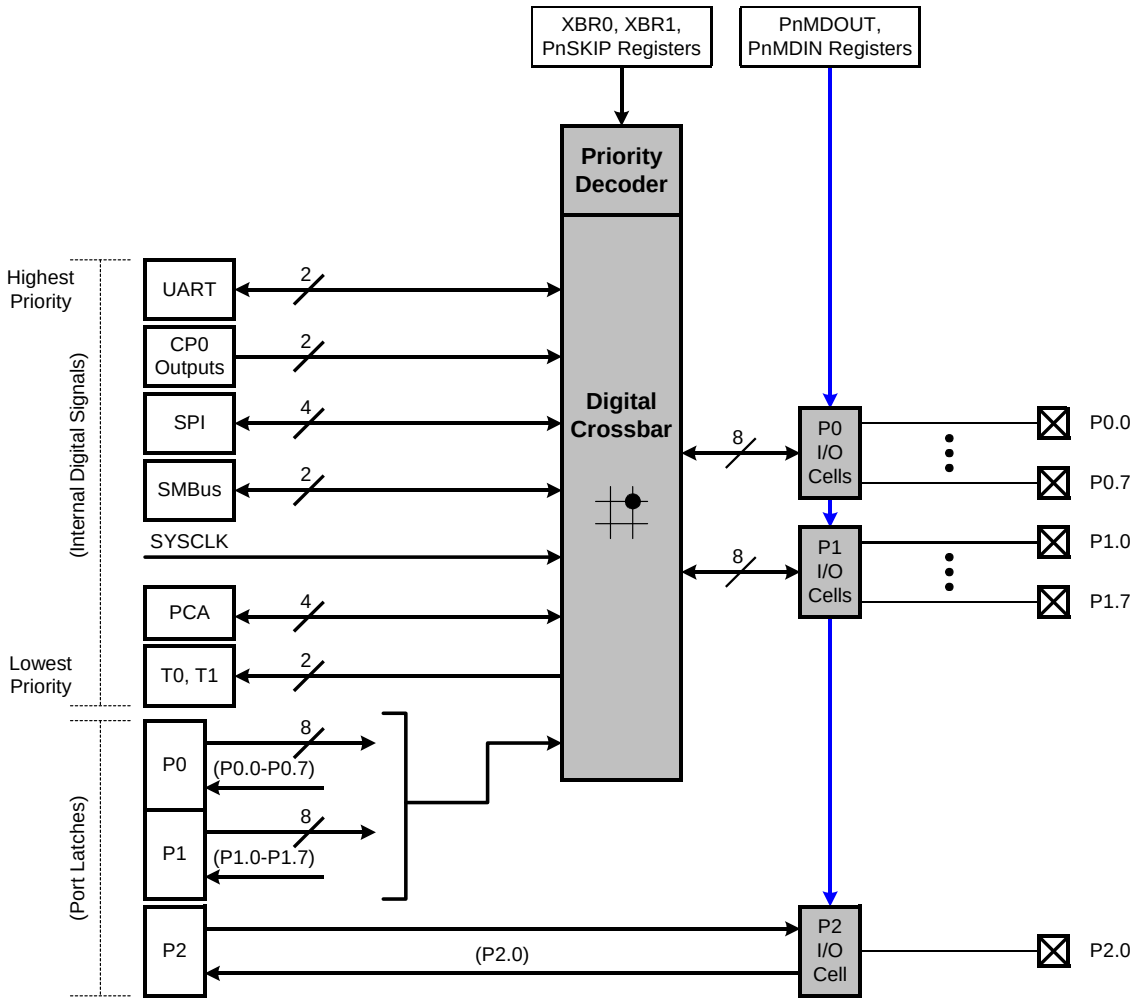



**Figure 18.1. Port I/O Functional Block Diagram**

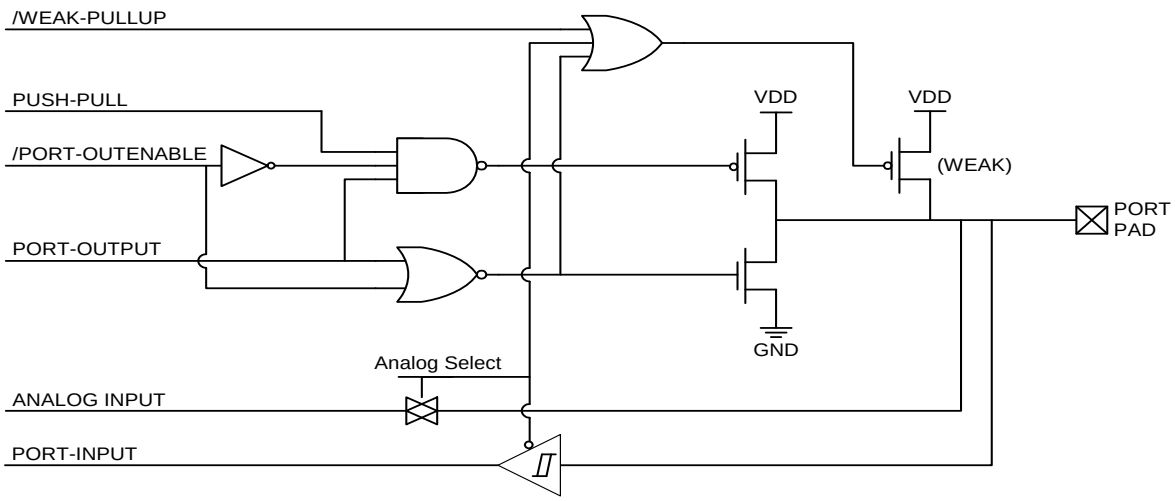



**Figure 18.2. Port I/O Cell Block Diagram**

## 18.1. Priority Crossbar Decoder

The Priority Crossbar Decoder (Figure 18.3) assigns a priority to each I/O function, starting at the top with UART0. When a digital resource is selected, the least-significant unassigned Port pin is assigned to that resource (excluding UART0, which will be assigned to pins P0.4 and P0.5, and the Comparator0 outputs, which will be assigned to P1.4 and P1.5). If a Port pin is assigned, the Crossbar skips that pin when assigning the next selected resource. Additionally, the Crossbar will skip Port pins whose associated bits in the PnSKIP registers are set. The PnSKIP registers allow software to skip Port pins that are to be used for analog input, dedicated functions, or GPIO.

**Important Note on Crossbar Configuration:** If a Port pin is claimed by a peripheral without use of the Crossbar, its corresponding PnSKIP bit should be set. This applies to P0.3 and/or P0.2 for the external oscillator, P0.6 for the external CNVSTR signal, P1.6 for IDA0, P1.7 for IDA1, and any selected ADC or comparator inputs. The Crossbar skips selected pins as if they were already assigned, and moves to the next unassigned pin. Figure 18.3 shows the Crossbar Decoder priority with no Port pins skipped (P0SKIP, P1SKIP = 0x00); Figure 18.4 shows the Crossbar Decoder priority with the XTAL1 (P0.2) and XTAL2 (P0.3) pins skipped (P0SKIP = 0x0C).

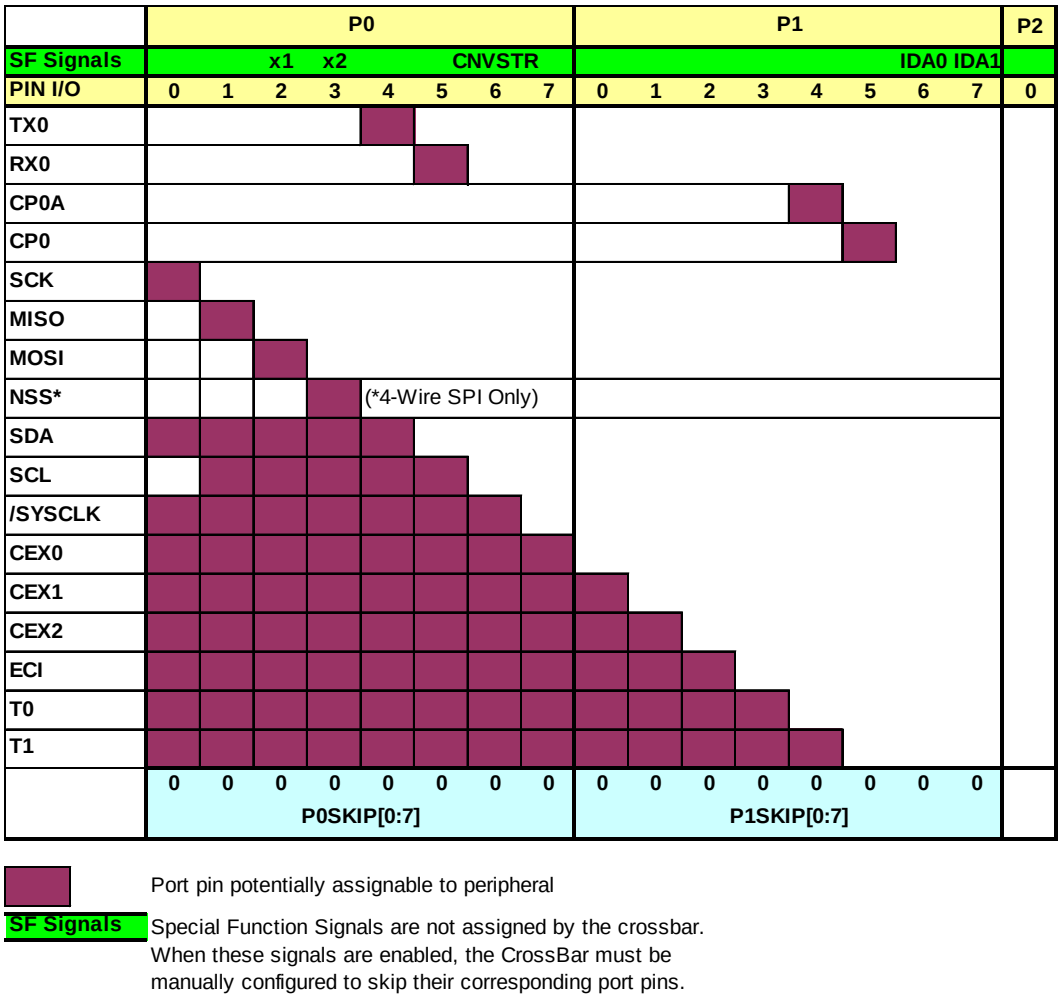



**Figure 18.3. Crossbar Priority Decoder with No Pins Skipped**

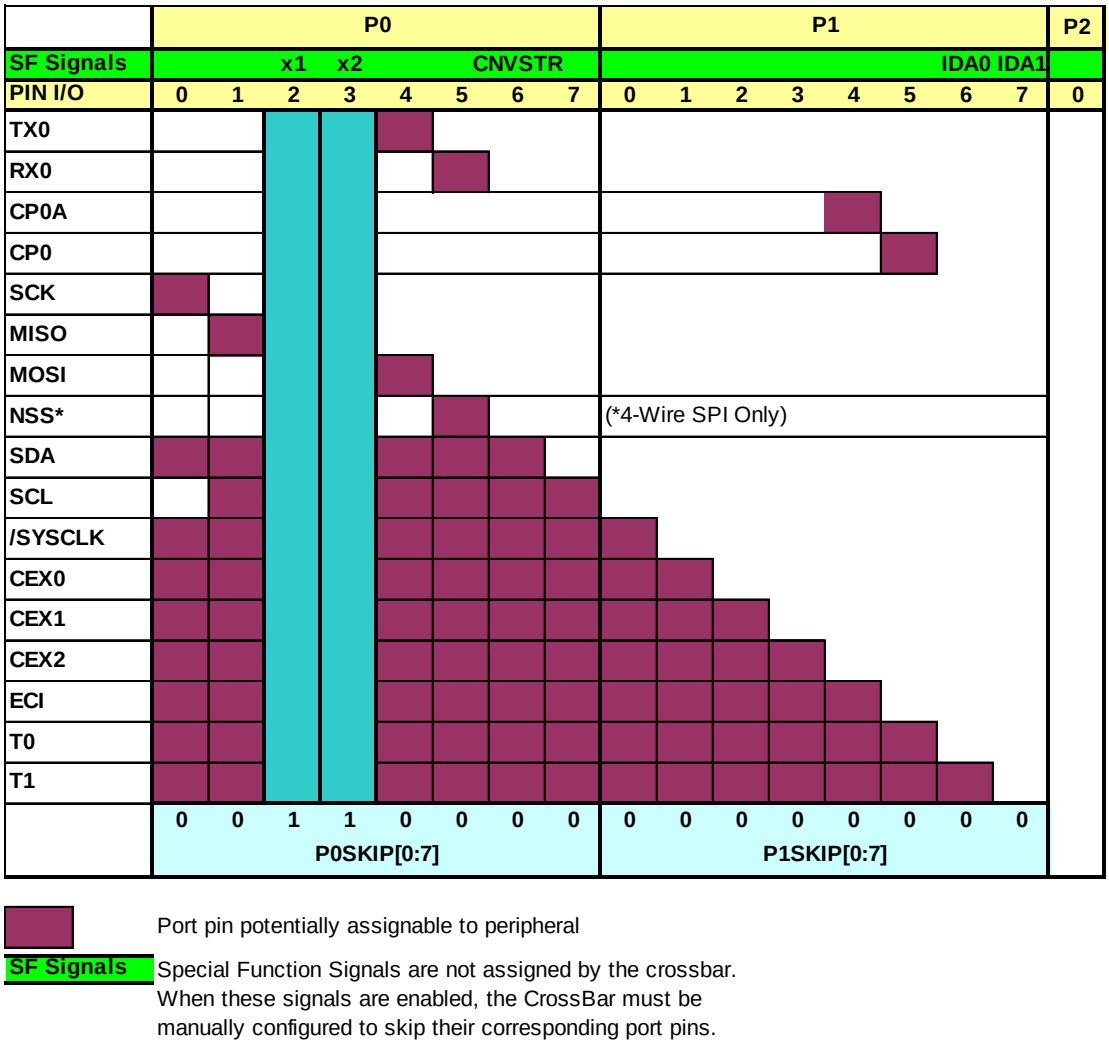



**Figure 18.4. Crossbar Priority Decoder with Crystal Pins Skipped**

Registers XBR0 and XBR1 are used to assign the digital I/O resources to the physical I/O Port pins. Note that when the SMBus is selected, the Crossbar assigns both pins associated with the SMBus (SDA and SCL); when the UART is selected, the Crossbar assigns both pins associated with the UART (TX and RX). UART0 pin assignments are fixed for bootloading purposes: UART TX0 is always assigned to P0.4; UART RX0 is always assigned to P0.5. Comparator outputs are also fixed: CP0A will appear only on P1.4, CP0 will appear only on P1.5. Standard Port I/Os appear contiguously after the prioritized functions have been assigned.

**Important Note:** The SPI can be operated in either 3-wire or 4-wire modes, pending the state of the NSSMD1–NSSMD0 bits in register SPI0CN. According to the SPI mode, the NSS signal may or may not be routed to a Port pin.

## 18.2. Port I/O Initialization

Port I/O initialization consists of the following steps:

Step 1. Select the input mode (analog or digital) for all Port pins, using the Port Input Mode register (PnMDIN).
Step 2. Select the output mode (open-drain or push-pull) for all Port pins, using the Port Output Mode register (PnMDOUT).
Step 3. Select any pins to be skipped by the I/O Crossbar using the Port Skip registers (PnSKIP).
Step 4. Assign Port pins to desired peripherals.
Step 5. Enable the Crossbar (XBARE = ‘1’).

All Port pins must be configured as either analog or digital inputs. Any pins to be used as Comparator or ADC inputs should be configured as an analog inputs. When a pin is configured as an analog input, its weak pull-up, digital driver, and digital receiver are disabled. This process saves power and reduces noise on the analog input. Pins configured as digital inputs may still be used by analog peripherals; however this practice is not recommended.

Additionally, all analog input pins should be configured to be skipped by the Crossbar (accomplished by setting the associated bits in PnSKIP). Port input mode is set in the PnMDIN register, where a ‘1’ indicates a digital input, and a ‘0’ indicates an analog input. All pins default to digital inputs on reset. See SFR Definition 18.4 for the PnMDIN register details.

The output driver characteristics of the I/O pins are defined using the Port Output Mode registers (PnMDOUT). Each Port Output driver can be configured as either open drain or push-pull. This selection is required even for the digital resources selected in the XBRn registers, and is not automatic. The only exception to this is the SMBus (SDA, SCL) pins, which are configured as open-drain regardless of the PnMDOUT settings. When the WEAKPUD bit in XBR1 is ‘0’, a weak pull-up is enabled for all Port I/O configured as open-drain. WEAKPUD does not affect the push-pull Port I/O. Furthermore, the weak pull-up is turned off on an output that is driving a ‘0’ to avoid unnecessary power dissipation.

Registers XBR0 and XBR1 must be loaded with the appropriate values to select the digital I/O functions required by the design. Setting the XBARE bit in XBR1 to ‘1’ enables the Crossbar. Until the Crossbar is enabled, the external pins remain as standard Port I/O (in input mode), regardless of the XBRn Register settings. For given XBRn Register settings, one can determine the I/O pin-out using the Priority Decode Table.

The Crossbar must be enabled to use Port pins as standard Port I/O in output mode. Port output drivers are disabled while the Crossbar is disabled.

## SFR Definition 18.1. XBR0: Port I/O Crossbar Register 0

| R | R | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| — | — | CP0AE | CP0E | SYSCKE | SMB0E | SPI0E | URT0E | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xE1

Bits7–6: UNUSED. Read = 00b, Write = don't care.
Bit5: CP0AE: Comparator0 Asynchronous Output Enable
0: Asynchronous CP0 unavailable at Port pin.
1: Asynchronous CP0 routed to Port pin P1.4.
Bit4: CP0E: Comparator0 Output Enable
0: CP0 unavailable at Port pin.
1: CP0 routed to Port pin P1.5.
Bit3: SYSCKE: /SYSCLK Output Enable
0: /SYSCLK unavailable at Port pin.
1: /SYSCLK output routed to Port pin.
Bit2: SMB0E: SMBus I/O Enable
0: SMBus I/O unavailable at Port pins.
1: SMBus I/O routed to Port pins.
Bit1: SPI0E: SPI I/O Enable
0: SPI I/O unavailable at Port pins.
1: SPI I/O routed to Port pins. Note that the SPI can be assigned either 3 or 4 GPIO pins.
Bit0: URT0E: UART I/O Output Enable
0: UART I/O unavailable at Port pin.
1: UART TX0, RX0 routed to Port pins P0.4 and P0.5.

## SFR Definition 18.2. XBR1: Port I/O Crossbar Register 1

| R/W | R/W | R/W | R/W | R/W | R | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| WEAKPUD | XBARE | T1E | T0E | ECIE | — | PCA0ME | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xE2

Bit7: WEAKPUD: Port I/O Weak Pull-up Disable.
0: Weak Pull-ups enabled (except for Ports whose I/O are configured as analog input).
1: Weak Pull-ups disabled.

Bit6: XBARE: Crossbar Enable.
0: Crossbar disabled.
1: Crossbar enabled.

Bit5: T1E: T1 Enable
0: T1 unavailable at Port pin.
1: T1 routed to Port pin.

Bit4: T0E: T0 Enable
0: T0 unavailable at Port pin.
1: T0 routed to Port pin.

Bit3: ECIE: PCA0 External Counter Input Enable
0: ECI unavailable at Port pin.
1: ECI routed to Port pin.

Bit2: Unused. Read = 0b. Write = don't care.

Bits1–0: PCA0ME: PCA Module I/O Enable Bits.
00: All PCA I/O unavailable at Port pins.
01: CEX0 routed to Port pin.
10: CEX0, CEX1 routed to Port pins.
11: CEX0, CEX1, CEX2 routed to Port pins.

## 18.3. General Purpose Port I/O

Port pins that remain unassigned by the Crossbar and are not used by analog peripherals can be used for general purpose I/O. Ports P0–P2 are accessed through corresponding special function registers (SFRs) that are both byte addressable and bit addressable. When writing to a Port, the value written to the SFR is latched to maintain the output data value at each pin. When reading, the logic levels of the Port's input pins are returned regardless of the XBRn settings (i.e., even when the pin is assigned to another signal by the Crossbar, the Port register can always read its corresponding Port I/O pin). The exception to this is the execution of the read-modify-write instructions that target a Port Latch register as the destination. The read-modify-write instructions when operating on a Port SFR are the following: ANL, ORL, XRL, JBC, CPL, INC, DEC, DJNZ and MOV, CLR or SETB, when the destination is an individual bit in a Port SFR. For these instructions, the value of the register (not the pin) is read, modified, and written back to the SFR.

## SFR Definition 18.3. P0: Port0

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| P0.7 | P0.6 | P0.5 | P0.4 | P0.3 | P0.2 | P0.1 | P0.0 | 11111111 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | Bit Addressable |

SFR Address: 0x80

Bits7–0: P0.[7:0]
Write - Output appears on I/O pins per Crossbar Registers.
0: Logic Low Output.
1: Logic High Output (high impedance if corresponding P0MDOUT.n bit = 0).
Read - Always reads '0' if selected as analog input in register P0MDIN. Directly reads Port pin when configured as digital input.
0: P0.n pin is logic low.
1: P0.n pin is logic high.

## SFR Definition 18.4. P0MDIN: Port0 Input Mode

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| | | | | | | | | 11111111 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xF1

Bits7–0: Analog Input Configuration Bits for P0.7–P0.0 (respectively).
Port pins configured as analog inputs have their weak pull-up, digital driver, and digital receiver disabled.
0: Corresponding P0.n pin is configured as an analog input.
1: Corresponding P0.n pin is not configured as an analog input.

### SFR Definition 18.5. P0MDOUT: Port0 Output Mode

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xA4

Bits7–0: Output Configuration Bits for P0.7–P0.0 (respectively): ignored if corresponding bit in register P0MDIN is logic 0.
0: Corresponding P0.n Output is open-drain.
1: Corresponding P0.n Output is push-pull.

(Note: When SDA and SCL appear on any of the Port I/O, each are open-drain regardless of the value of P0MDOUT).

### SFR Definition 18.6. P0SKIP: Port0 Skip

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xD4

Bits7–0: P0SKIP[7:0]: Port0 Crossbar Skip Enable Bits.
These bits select Port pins to be skipped by the Crossbar Decoder. Port pins used as analog inputs (for ADC or Comparator) or used as special functions (VREF input, external oscillator circuit, CNVSTR input) should be skipped by the Crossbar.
0: Corresponding P0.n pin is not skipped by the Crossbar.
1: Corresponding P0.n pin is skipped by the Crossbar.

## SFR Definition 18.7. P1: Port1

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| P1.7 | P1.6 | P1.5 | P1.4 | P1.3 | P1.2 | P1.1 | P1.0 | 11111111 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | Bit Addressable |

SFR Address: 0x90

Bits7–0: P1.[7:0]
Write - Output appears on I/O pins per Crossbar Registers.
0: Logic Low Output.
1: Logic High Output (high impedance if corresponding P1MDOUT.n bit = 0).
Read - Always reads '0' if selected as analog input in register P1MDIN. Directly reads Port pin when configured as digital input.
0: P1.n pin is logic low.
1: P1.n pin is logic high.

## SFR Definition 18.8. P1MDIN: Port1 Input Mode

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| | | | | | | | | 11111111 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xF2

Bits7–0: Analog Input Configuration Bits for P1.7–P1.0 (respectively).
Port pins configured as analog inputs have their weak pull-up, digital driver, and digital receiver disabled.
0: Corresponding P1.n pin is configured as an analog input.
1: Corresponding P1.n pin is not configured as an analog input.

### SFR Definition 18.9. P1MDOUT: Port1 Output Mode

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xA5

Bits7–0: Output Configuration Bits for P1.7–P1.0 (respectively): ignored if corresponding bit in register P1MDIN is logic 0.
0: Corresponding P1.n Output is open-drain.
1: Corresponding P1.n Output is push-pull.

### SFR Definition 18.10. P1SKIP: Port1 Skip

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xD5

Bits7–0: P1SKIP[7:0]: Port1 Crossbar Skip Enable Bits.
These bits select Port pins to be skipped by the Crossbar Decoder. Port pins used as analog inputs (for ADC or Comparator) or used as special functions (VREF input, external oscillator circuit, CNVSTR input) should be skipped by the Crossbar.
0: Corresponding P1.n pin is not skipped by the Crossbar.
1: Corresponding P1.n pin is skipped by the Crossbar.

## SFR Definition 18.11. P2: Port2

| R | R | R | R | R | R | R | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| — | — | — | — | — | — | — | P2.0 | 00000001 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | Bit Addressable |

SFR Address: 0xA0

Bits7–1: Unused. Read = 0000000b. Write = don't care.
Bit0: P2.0
Write - Output appears on I/O pins per Crossbar Registers.
0: Logic Low Output.
1: Logic High Output (high impedance if corresponding P2MDOUT.n bit = 0).
Read - Directly reads Port pin.
0: P2.n pin is logic low.
1: P2.n pin is logic high.

## SFR Definition 18.12. P2MDOUT: Port2 Output Mode

| R | R | R | R | R | R | R | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| — | — | — | — | — | — | — | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xA6

Bits7–1: Unused. Read = 0000000b. Write = don't care.
Bit0: Output Configuration Bit for P2.0.
0: P2.0 Output is open-drain.
1: P2.0 Output is push-pull.

### Table 18.1. Port I/O DC Electrical Characteristics

$V_{DD}$ = 2.7 to 3.6 V, –40 to +85 °C unless otherwise specified.

| Parameters | Conditions | Min | Typ | Max | Units |
|---|---|---|---|---|---|
| Output High Voltage | $I_{OH}$ = –3 mA, Port I/O push-pull<br>$I_{OH}$ = –10 µA, Port I/O push-pull<br>$I_{OH}$ = –10 mA, Port I/O push-pull | $V_{DD}$ – 0.7<br>$V_{DD}$ – 0.1<br>— | —<br>—<br>$V_{DD}$ – 0.8 | —<br>—<br>— | V |
| Output Low Voltage | $I_{OL}$ = 8.5 mA<br>$I_{OL}$ = 10 µA<br>$I_{OL}$ = 25 mA | —<br>—<br>— | —<br>—<br>1.0 | 0.6<br>0.1<br>— | V |
| Input High Voltage | | 2.0 | — | — | V |
| Input Low Voltage | | — | — | 0.8 | V |
| Input Leakage Current | Weak Pull-up Off<br>Weak Pull-up On, $V_{IN}$ = 0 V | —<br>— | <br>25 | ±1<br>50 | µA |

## 19. SMBus

The SMBus I/O interface is a two-wire, bi-directional serial bus. The SMBus is compliant with the System Management Bus Specification, version 1.1, and compatible with the I2C serial bus. Reads and writes to the interface by the system controller are byte oriented with the SMBus interface autonomously controlling the serial transfer of the data. Data can be transferred at up to 1/20th of the system clock as a master or slave (this can be faster than allowed by the SMBus specification, depending on the system clock used). A method of extending the clock-low duration is available to accommodate devices with different speed capabilities on the same bus.

The SMBus interface may operate as a master and/or slave, and may function on a bus with multiple masters. The SMBus provides control of SDA (serial data), SCL (serial clock) generation and synchronization, arbitration logic, and START/STOP control and generation. Three SFRs are associated with the SMBus: SMB0CF configures the SMBus; SMB0CN controls the status of the SMBus; and SMB0DAT is the data register, used for both transmitting and receiving SMBus data and slave addresses.

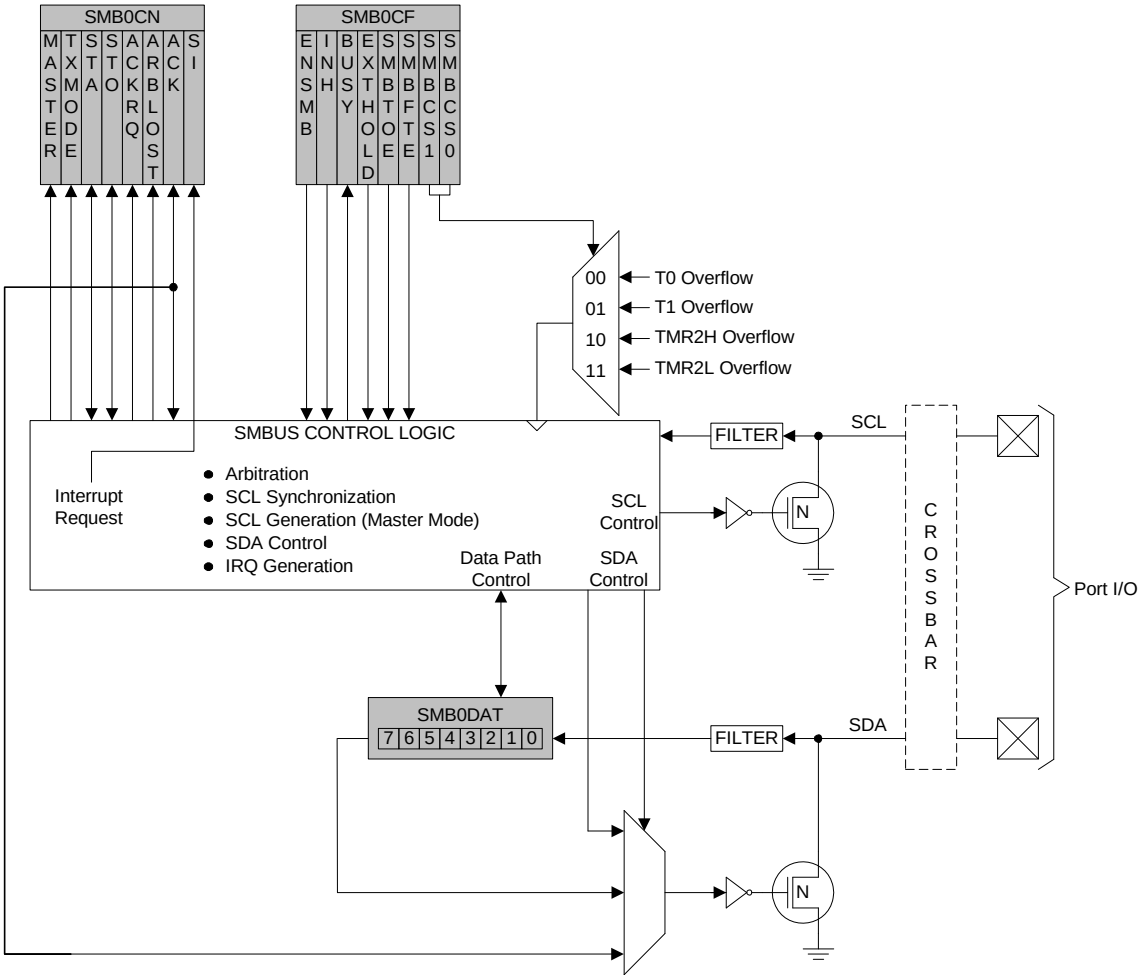



**Figure 19.1. SMBus Block Diagram**

## 19.1. Supporting Documents

It is assumed the reader is familiar with or has access to the following supporting documents:

1. The I2C-Bus and How to Use It (including specifications), Philips Semiconductor.
2. The I2C-Bus Specification -- Version 2.0, Philips Semiconductor.
3. System Management Bus Specification -- Version 1.1, SBS Implementers Forum.

## 19.2. SMBus Configuration

Figure 19.2 shows a typical SMBus configuration. The SMBus specification allows any recessive voltage between 3.0 V and 5.0 V; different devices on the bus may operate at different voltage levels. The bi-directional SCL (serial clock) and SDA (serial data) lines must be connected to a positive power supply voltage through a pull-up resistor or similar circuit. Every device connected to the bus must have an open-drain or open-collector output for both the SCL and SDA lines, so that both are pulled high (recessive state) when the bus is free. The maximum number of devices on the bus is limited only by the requirement that the rise and fall times on the bus not exceed 300 ns and 1000 ns, respectively.

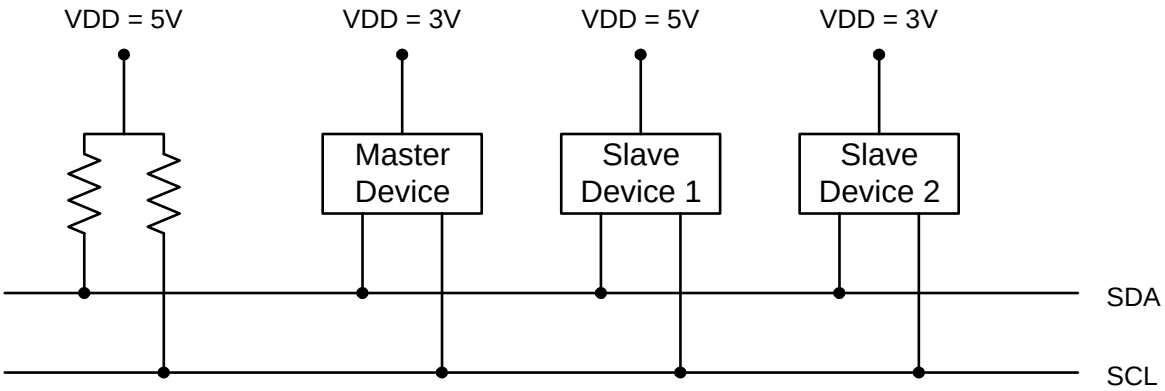



**Figure 19.2. Typical SMBus Configuration**

## 19.3. SMBus Operation

Two types of data transfers are possible: data transfers from a master transmitter to an addressed slave receiver (WRITE), and data transfers from an addressed slave transmitter to a master receiver (READ). The master device initiates both types of data transfers and provides the serial clock pulses on SCL. The SMBus interface may operate as a master or a slave, and multiple master devices on the same bus are supported. If two or more masters attempt to initiate a data transfer simultaneously, an arbitration scheme is employed with a single master always winning the arbitration. Note that it is not necessary to specify one device as the Master in a system; any device who transmits a START and a slave address becomes the master for the duration of that transfer.

A typical SMBus transaction consists of a START condition followed by an address byte (Bits7–1: 7-bit slave address; Bit0: R/W direction bit), one or more bytes of data, and a STOP condition. Each byte that is received (by a master or slave) must be acknowledged (ACK) with a low SDA during a high SCL (see Figure 19.3). If the receiving device does not ACK, the transmitting device will read a NACK (not acknowledge), which is a high SDA during a high SCL.

The direction bit (R/W) occupies the least-significant bit position of the address byte. The direction bit is set to logic 1 to indicate a "READ" operation and cleared to logic 0 to indicate a "WRITE" operation.

All transactions are initiated by a master, with one or more addressed slave devices as the target. The master generates the START condition and then transmits the slave address and direction bit. If the transaction is a WRITE operation from the master to the slave, the master transmits the data a byte at a time waiting for an ACK from the slave at the end of each byte. For READ operations, the slave transmits the data waiting for an ACK from the master at the end of each byte. At the end of the data transfer, the master generates a STOP condition to terminate the transaction and free the bus. Figure 19.3 illustrates a typical SMBus transaction.

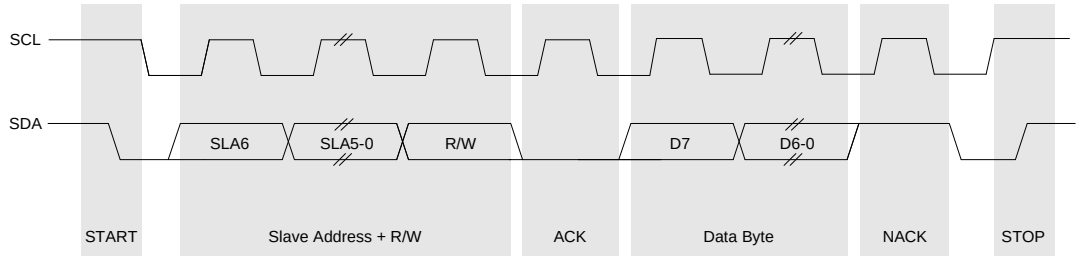



**Figure 19.3. SMBus Transaction**

### 19.3.1. Arbitration

A master may start a transfer only if the bus is free. The bus is free after a STOP condition or after the SCL and SDA lines remain high for a specified time (see Section “19.3.4. SCL High (SMBus Free) Timeout’ on page 154). In the event that two or more devices attempt to begin a transfer at the same time, an arbitration scheme is employed to force one master to give up the bus. The master devices continue transmitting until one attempts a HIGH while the other transmits a LOW. Since the bus is open-drain, the bus will be pulled LOW. The master attempting the HIGH will detect a LOW SDA and lose the arbitration. The winning master continues its transmission without interruption; the losing master becomes a slave and receives the rest of the transfer if addressed. This arbitration scheme is non-destructive: one device always wins, and no data is lost.

### 19.3.2. Clock Low Extension

SMBus provides a clock synchronization mechanism, similar to I2C, which allows devices with different speed capabilities to coexist on the bus. A clock-low extension is used during a transfer in order to allow slower slave devices to communicate with faster masters. The slave may temporarily hold the SCL line LOW to extend the clock low period, effectively decreasing the serial clock frequency.

### 19.3.3. SCL Low Timeout

If the SCL line is held low by a slave device on the bus, no further communication is possible. Furthermore, the master cannot force the SCL line high to correct the error condition. To solve this problem, the SMBus protocol specifies that devices participating in a transfer must detect any clock cycle held low longer than 25 ms as a “timeout” condition. Devices that have detected the timeout condition must reset the communication no later than 10 ms after detecting the timeout condition.

When the SMBTOE bit in SMB0CF is set, Timer 3 is used to detect SCL low timeouts. Timer 3 is forced to reload when SCL is high, and allowed to count when SCL is low. With Timer 3 enabled and configured to overflow after 25 ms (and SMBTOE set), the Timer 3 interrupt service routine can be used to reset (disable and re-enable) the SMBus in the event of an SCL low timeout.

### 19.3.4. SCL High (SMBus Free) Timeout

The SMBus specification stipulates that if the SCL and SDA lines remain high for more that 50 µs, the bus is designated as free. When the SMBFTE bit in SMB0CF is set, the bus will be considered free if SCL and SDA remain high for more than 10 SMBus clock source periods. If the SMBus is waiting to generate a Master START, the START will be generated following this timeout. Note that a clock source is required for free timeout detection, even in a slave-only implementation.

## 19.4. Using the SMBus

The SMBus can operate in both Master and Slave modes. The interface provides timing and shifting control for serial transfers; higher level protocol is determined by user software. The SMBus interface provides the following application-independent features:

- Byte-wise serial data transfers
- Clock signal generation on SCL (Master Mode only) and SDA data synchronization
- Timeout/bus error recognition, as defined by the SMB0CF configuration register
- START/STOP timing, detection, and generation
- Bus arbitration
- Interrupt generation
- Status information

SMBus interrupts are generated for each data byte or slave address that is transferred. When transmitting, this interrupt is generated after the ACK cycle so that software may read the received ACK value; when receiving data, this interrupt is generated before the ACK cycle so that software may define the outgoing ACK value. See Section “19.5. SMBus Transfer Modes’ on page 163 for more details on transmission sequences.

Interrupts are also generated to indicate the beginning of a transfer when a master (START generated), or the end of a transfer when a slave (STOP detected). Software should read the SMB0CN (SMBus Control register) to find the cause of the SMBus interrupt. The SMB0CN register is described in Section “19.4.2. SMB0CN Control Register’ on page 159; Table 19.4 provides a quick SMB0CN decoding reference.

SMBus configuration options include:

- Timeout detection (SCL Low Timeout and/or Bus Free Timeout)
- SDA setup and hold time extensions
- Slave event enable/disable
- Clock source selection

These options are selected in the SMB0CF register, as described in Section “19.4.1. SMBus Configuration Register’ on page 156.

### 19.4.1. SMBus Configuration Register

The SMBus Configuration register (SMB0CF) is used to enable the SMBus Master and/or Slave modes, select the SMBus clock source, and select the SMBus timing and timeout options. When the ENSMB bit is set, the SMBus is enabled for all master and slave events. Slave events may be disabled by setting the INH bit. With slave events inhibited, the SMBus interface will still monitor the SCL and SDA pins; however, the interface will NACK all received addresses and will not generate any slave interrupts. When the INH bit is set, all slave events will be inhibited following the next START (interrupts will continue for the duration of the current transfer).

**Table 19.1. SMBus Clock Source Selection**

| SMBCS1 | SMBCS0 | SMBus Clock Source |
|---|---|---|
| 0 | 0 | Timer 0 Overflow |
| 0 | 1 | Timer 1 Overflow |
| 1 | 0 | Timer 2 High Byte Overflow |
| 1 | 1 | Timer 2 Low Byte Overflow |

The SMBCS1–0 bits select the SMBus clock source, which is used only when operating as a master or when the Free Timeout detection is enabled. When operating as a master, overflows from the selected source determine the absolute minimum SCL low and high times as defined in Equation 19.1. Note that the selected clock source may be shared by other peripherals so long as the timer is left running at all times. For example, Timer 1 overflows may generate the SMBus and UART baud rates simultaneously. Timer configuration is covered in Section “22. Timers’ on page 195.

$$T_{HighMin} = T_{LowMin} = \frac{1}{f_{ClockSourceOverflow}}$$

**Equation 19.1. Minimum SCL High and Low Times**

The selected clock source should be configured to establish the minimum SCL High and Low times as per Equation 19.1. When the interface is operating as a master (and SCL is not driven or extended by any other devices on the bus), the typical SMBus bit rate is approximated by Equation 19.2.

$$BitRate = \frac{f_{ClockSourceOverflow}}{3}$$

**Equation 19.2. Typical SMBus Bit Rate**

Figure 19.4 shows the typical SCL generation described by Equation 19.2. Notice that $T_{HIGH}$ is typically twice as large as $T_{LOW}$. The actual SCL output may vary due to other devices on the bus (SCL may be extended low by slower slave devices, or driven low by contending master devices). The bit rate when operating as a master will never exceed the limits defined by equation Equation 19.1.

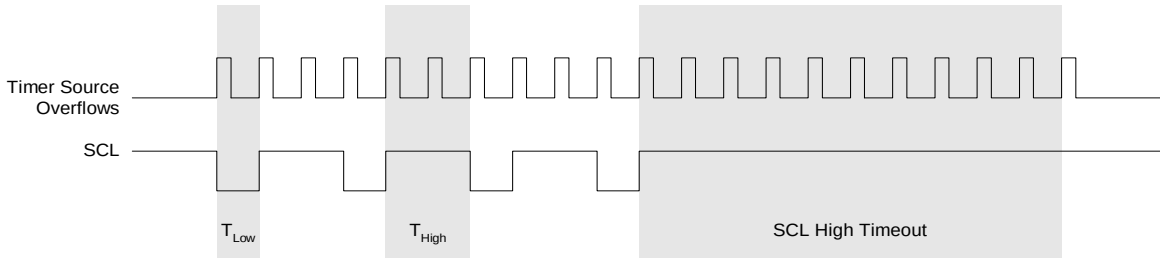



**Figure 19.4. Typical SMBus SCL Generation**

Setting the EXTHOLD bit extends the minimum setup and hold times for the SDA line. The minimum SDA setup time defines the absolute minimum time that SDA is stable before SCL transitions from low-to-high. The minimum SDA hold time defines the absolute minimum time that the current SDA value remains stable after SCL transitions from high-to-low. EXTHOLD should be set so that the minimum setup and hold times meet the SMBus Specification requirements of 250 ns and 300 ns, respectively. Table 19.2 shows the minimum setup and hold times for the two EXTHOLD settings. Setup and hold time extensions are typically necessary when SYSCLK is above 10 MHz.

**Table 19.2. Minimum SDA Setup and Hold Times**

| EXTHOLD | Minimum SDA Setup Time | Minimum SDA Hold Time |
|---|---|---|
| 0 | $T_{low}$ – 4 system clocks<br>OR<br>1 system clock + s/w delay* | 3 system clocks |
| 1 | 11 system clocks | 12 system clocks |
| ***Note:** Setup Time for ACK bit transmissions and the MSB of all data transfers. The s/w delay occurs between the time SMB0DAT or ACK is written and when SI is cleared. Note that if SI is cleared in the same write that defines the outgoing ACK value, s/w delay is zero. | | |

With the SMBTOE bit set, Timer 3 should be configured to overflow after 25 ms in order to detect SCL low timeouts (see Section “19.3.3. SCL Low Timeout’ on page 154). The SMBus interface will force Timer 3 to reload while SCL is high, and allow Timer 3 to count when SCL is low. The Timer 3 interrupt service routine should be used to reset SMBus communication by disabling and re-enabling the SMBus.

SMBus Free Timeout detection can be enabled by setting the SMBFTE bit. When this bit is set, the bus will be considered free if SDA and SCL remain high for more than 10 SMBus clock source periods (see Figure 19.4). When a Free Timeout is detected, the interface will respond as if a STOP was detected (an interrupt will be generated, and STO will be set).

## SFR Definition 19.1. SMB0CF: SMBus Clock/Configuration

| R/W | R/W | R | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| ENSMB | INH | BUSY | EXTHOLD | SMBTOE | SMBFTE | SMBCS1 | SMBCS0 | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xC1

Bit7: ENSMB: SMBus Enable.
This bit enables/disables the SMBus interface. When enabled, the interface constantly monitors the SDA and SCL pins.
0: SMBus interface disabled.
1: SMBus interface enabled.

Bit6: INH: SMBus Slave Inhibit.
When this bit is set to logic 1, the SMBus does not generate an interrupt when slave events occur. This effectively removes the SMBus slave from the bus. Master Mode interrupts are not affected.
0: SMBus Slave Mode enabled.
1: SMBus Slave Mode inhibited.

Bit5: BUSY: SMBus Busy Indicator.
This bit is set to logic 1 by hardware when a transfer is in progress. It is cleared to logic 0 when a STOP or free-timeout is sensed.

Bit4: EXTHOLD: SMBus Setup and Hold Time Extension Enable.
This bit controls the SDA setup and hold times according to Table 19.2.
0: SDA Extended Setup and Hold Times disabled.
1: SDA Extended Setup and Hold Times enabled.

Bit3: SMBTOE: SMBus SCL Timeout Detection Enable.
This bit enables SCL low timeout detection. If set to logic 1, the SMBus forces Timer 3 to reload while SCL is high and allows Timer 3 to count when SCL goes low. If Timer 3 is configured in split mode (T3SPLIT is set), only the high byte of Timer 3 is held in reload while SCL is high. Timer 3 should be programmed to generate interrupts at 25 ms, and the Timer 3 interrupt service routine should reset SMBus communication.

Bit2: SMBFTE: SMBus Free Timeout Detection Enable.
When this bit is set to logic 1, the bus will be considered free if SCL and SDA remain high for more than 10 SMBus clock source periods.

Bits1–0: SMBCS1–SMBCS0: SMBus Clock Source Selection.
These two bits select the SMBus clock source, which is used to generate the SMBus bit rate. The selected device should be configured according to Equation 19.1.

| SMBCS1 | SMBCS0 | SMBus Clock Source |
|---|---|---|
| 0 | 0 | Timer 0 Overflow |
| 0 | 1 | Timer 1 Overflow |
| 1 | 0 | Timer 2 High Byte Overflow |
| 1 | 1 | Timer 2 Low Byte Overflow |

### 19.4.2. SMB0CN Control Register

SMB0CN is used to control the interface and to provide status information (see SFR Definition 19.2). The higher four bits of SMB0CN (MASTER, TXMODE, STA, and STO) form a status vector that can be used to jump to service routines. MASTER and TXMODE indicate the master/slave state and transmit/receive modes, respectively.

STA and STO indicate that a START and/or STOP has been detected or generated since the last SMBus interrupt. STA and STO are also used to generate START and STOP conditions when operating as a master. Writing a ‘1’ to STA will cause the SMBus interface to enter Master Mode and generate a START when the bus becomes free (STA is not cleared by hardware after the START is generated). Writing a ‘1’ to STO while in Master Mode will cause the interface to generate a STOP and end the current transfer after the next ACK cycle. If STO and STA are both set (while in Master Mode), a STOP followed by a START will be generated.

As a receiver, writing the ACK bit defines the outgoing ACK value; as a transmitter, reading the ACK bit indicates the value received on the last ACK cycle. ACKRQ is set each time a byte is received, indicating that an outgoing ACK value is needed. When ACKRQ is set, software should write the desired outgoing value to the ACK bit before clearing SI. A NACK will be generated if software does not write the ACK bit before clearing SI. SDA will reflect the defined ACK value immediately following a write to the ACK bit; however SCL will remain low until SI is cleared. If a received slave address is not acknowledged, further slave events will be ignored until the next START is detected.

The ARBLOST bit indicates that the interface has lost an arbitration. This may occur anytime the interface is transmitting (master or slave). A lost arbitration while operating as a slave indicates a bus error condition. ARBLOST is cleared by hardware each time SI is cleared.

The SI bit (SMBus Interrupt Flag) is set at the beginning and end of each transfer, after each byte frame, or when an arbitration is lost; see Table 19.3 for more details.

**Important note about the SI bit:** The SMBus interface is stalled while SI is set; thus SCL is held low, and the bus is stalled until software clears SI.

Table 19.3 lists all sources for hardware changes to the SMB0CN bits. Refer to Table 19.4 for SMBus status decoding using the SMB0CN register.

## SFR Definition 19.2. SMB0CN: SMBus Control

| R | R | R/W | R/W | R | R | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| MASTER | TXMODE | STA | STO | ACKRQ | ARBLOST | ACK | SI | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | Bit Addressable |

SFR Address: 0xC0

Bit7: MASTER: SMBus Master/Slave Indicator.
This read-only bit indicates when the SMBus is operating as a master.
0: SMBus operating in Slave Mode.
1: SMBus operating in Master Mode.

Bit6: TXMODE: SMBus Transmit Mode Indicator.
This read-only bit indicates when the SMBus is operating as a transmitter.
0: SMBus in Receiver Mode.
1: SMBus in Transmitter Mode.

Bit5: STA: SMBus Start Flag.
Write:
0: No Start generated.
1: When operating as a master, a START condition is transmitted if the bus is free (If the bus is not free, the START is transmitted after a STOP is received or a timeout is detected). If STA is set by software as an active Master, a repeated START will be generated after the next ACK cycle.
Read:
0: No Start or repeated Start detected.
1: Start or repeated Start detected.

Bit4: STO: SMBus Stop Flag.
Write:
0: No STOP condition is transmitted.
1: Setting STO to logic 1 causes a STOP condition to be transmitted after the next ACK cycle. When the STOP condition is generated, hardware clears STO to logic 0. If both STA and STO are set, a STOP condition is transmitted followed by a START condition.
Read:
0: No Stop condition detected.
1: Stop condition detected (if in Slave Mode) or pending (if in Master Mode).

Bit3: ACKRQ: SMBus Acknowledge Request
This read-only bit is set to logic 1 when the SMBus has received a byte and needs the ACK bit to be written with the correct ACK response value.

Bit2: ARBLOST: SMBus Arbitration Lost Indicator.
This read-only bit is set to logic 1 when the SMBus loses arbitration while operating as a transmitter. A lost arbitration while a slave indicates a bus error condition.

Bit1: ACK: SMBus Acknowledge Flag.
This bit defines the out-going ACK level and records incoming ACK levels. It should be written each time a byte is received (when ACKRQ=1), or read after each byte is transmitted.
0: A "not acknowledge" has been received (if in Transmitter Mode) OR will be transmitted (if in Receiver Mode).
1: An "acknowledge" has been received (if in Transmitter Mode) OR will be transmitted (if in Receiver Mode).

Bit0: SI: SMBus Interrupt Flag.
This bit is set by hardware under the conditions listed in Table 19.3. SI must be cleared by software. While SI is set, SCL is held low and the SMBus is stalled.

**Table 19.3. Sources for Hardware Changes to SMB0CN**

| Bit | Set by Hardware When: | Cleared by Hardware When: |
|---|---|---|
| MASTER | • A START is generated. | • A STOP is generated.<br>• Arbitration is lost. |
| TXMODE | • START is generated.<br>• SMB0DAT is written before the start of an SMBus frame. | • A START is detected.<br>• Arbitration is lost.<br>• SMB0DAT is not written before the start of an SMBus frame. |
| STA | • A START followed by an address byte is received. | • Must be cleared by software. |
| STO | • A STOP is detected while addressed as a slave.<br>• Arbitration is lost due to a detected STOP. | • A pending STOP is generated. |
| ACKRQ | • A byte has been received and an ACK response value is needed. | • After each ACK cycle. |
| ARBLOST | • A repeated START is detected as a MASTER when STA is low (unwanted repeated START).<br>• SCL is sensed low while attempting to generate a STOP or repeated START condition.<br>• SDA is sensed low while transmitting a ‘1’ (excluding ACK bits). | • Each time SI is cleared. |
| ACK | • The incoming ACK value is low (ACKNOWLEDGE). | • The incoming ACK value is high (NOT ACKNOWLEDGE). |
| SI | • A START has been generated.<br>• Lost arbitration.<br>• A byte has been transmitted and an ACK/NACK received.<br>• A byte has been received.<br>• A START or repeated START followed by a slave address + R/W has been received.<br>• A STOP has been received. | • Must be cleared by software. |

### 19.4.3. Data Register

The SMBus Data register SMB0DAT holds a byte of serial data to be transmitted or one that has just been received. Software may safely read or write to the data register when the SI flag is set. Software should not attempt to access the SMB0DAT register when the SMBus is enabled and the SI flag is cleared to logic 0, as the interface may be in the process of shifting a byte of data into or out of the register.

Data in SMB0DAT is always shifted out MSB first. After a byte has been received, the first bit of received data is located at the MSB of SMB0DAT. While data is being shifted out, data on the bus is simultaneously being shifted in. SMB0DAT always contains the last data byte present on the bus. In the event of lost arbitration, the transition from master transmitter to slave receiver is made with the correct data or address in SMB0DAT.

**SFR Definition 19.3. SMB0DAT: SMBus Data**

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xC2

Bits7–0: SMB0DAT: SMBus Data.
The SMB0DAT register contains a byte of data to be transmitted on the SMBus serial interface or a byte that has just been received on the SMBus serial interface. The CPU can read from or write to this register whenever the SI serial interrupt flag (SMB0CN.0) is set to logic 1. The serial data in the register remains stable as long as the SI flag is set. When the SI flag is not set, the system may be in the process of shifting data in/out and the CPU should not attempt to access this register.

## 19.5. SMBus Transfer Modes

The SMBus interface may be configured to operate as master and/or slave. At any particular time, it will be operating in one of the following four modes: Master Transmitter, Master Receiver, Slave Transmitter, or Slave Receiver. The SMBus interface enters Master Mode any time a START is generated, and remains in Master Mode until it loses an arbitration or generates a STOP. An SMBus interrupt is generated at the end of all SMBus byte frames; however, note that the interrupt is generated before the ACK cycle when operating as a receiver, and after the ACK cycle when operating as a transmitter.

### 19.5.1. Master Transmitter Mode

Serial data is transmitted on SDA while the serial clock is output on SCL. The SMBus interface generates the START condition and transmits the first byte containing the address of the target slave and the data direction bit. In this case the data direction bit (R/W) will be logic 0 (WRITE). The master then transmits one or more bytes of serial data. After each byte is transmitted, an acknowledge bit is generated by the slave. The transfer is ended when the STO bit is set and a STOP is generated. Note that the interface will switch to Master Receiver Mode if SMB0DAT is not written following a Master Transmitter interrupt. Figure 19.5 shows a typical Master Transmitter sequence. Two transmit data bytes are shown, though any number of bytes may be transmitted. Notice that the ‘data byte transferred’ interrupts occur **after** the ACK cycle in this mode.

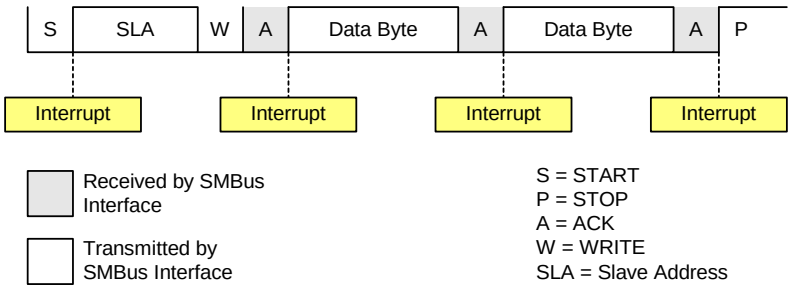



**Figure 19.5. Typical Master Transmitter Sequence**

### 19.5.2. Master Receiver Mode

Serial data is received on SDA while the serial clock is output on SCL. The SMBus interface generates the START condition and transmits the first byte containing the address of the target slave and the data direction bit. In this case the data direction bit (R/W) will be logic 1 (READ). Serial data is then received from the slave on SDA while the SMBus outputs the serial clock. The slave transmits one or more bytes of serial data. After each byte is received, ACKRQ is set to ‘1’ and an interrupt is generated. Software must write the ACK bit (SMB0CN.1) to define the outgoing acknowledge value (Note: writing a ‘1’ to the ACK bit generates an ACK; writing a ‘0’ generates a NACK). Software should write a ‘0’ to the ACK bit after the last byte is received, to transmit a NACK. The interface exits Master Receiver Mode after the STO bit is set and a STOP is generated. Note that the interface will switch to Master Transmitter Mode if SMB0DAT is written while an active Master Receiver. Figure 19.6 shows a typical Master Receiver sequence. Two received data bytes are shown, though any number of bytes may be received. Notice that the ‘data byte transferred’ interrupts occur **before** the ACK cycle in this mode.

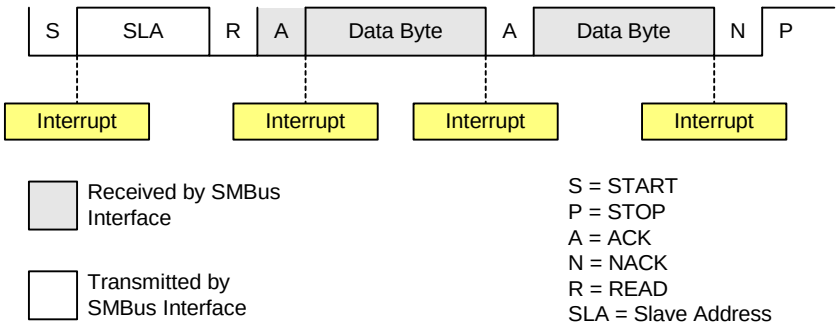



**Figure 19.6. Typical Master Receiver Sequence**

### 19.5.3. Slave Receiver Mode

Serial data is received on SDA and the clock is received on SCL. When slave events are enabled (INH = 0), the interface enters Slave Receiver Mode when a START followed by a slave address and direction bit (WRITE in this case) is received. Upon entering Slave Receiver Mode, an interrupt is generated and the ACKRQ bit is set. Software responds to the received slave address with an ACK, or ignores the received slave address with a NACK. If the received slave address is ignored, slave interrupts will be inhibited until the next START is detected. If the received slave address is acknowledged, zero or more data bytes are received. Software must write the ACK bit after each received byte to ACK or NACK the received byte. The interface exits Slave Receiver Mode after receiving a STOP. Note that the interface will switch to Slave Transmitter Mode if SMB0DAT is written while an active Slave Receiver. Figure 19.7 shows a typical Slave Receiver sequence. Two received data bytes are shown, though any number of bytes may be received. Notice that the 'data byte transferred' interrupts occur **before** the ACK cycle in this mode.

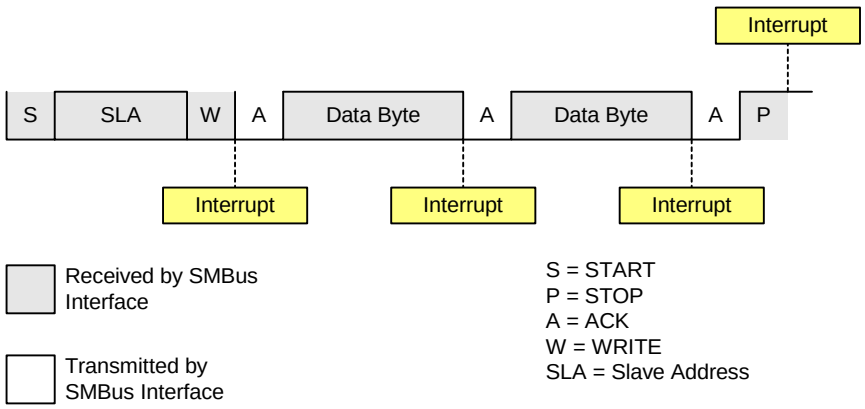



**Figure 19.7. Typical Slave Receiver Sequence**

### 19.5.4. Slave Transmitter Mode

Serial data is transmitted on SDA and the clock is received on SCL. When slave events are enabled (INH = 0), the interface enters Slave Receiver Mode (to receive the slave address) when a START followed by a slave address and direction bit (READ in this case) is received. Upon entering Slave Transmitter Mode, an interrupt is generated and the ACKRQ bit is set. Software responds to the received slave address with an ACK, or ignores the received slave address with a NACK. If the received slave address is ignored, slave interrupts will be inhibited until a START is detected. If the received slave address is acknowledged, data should be written to SMB0DAT to be transmitted. The interface enters Slave Transmitter Mode, and transmits one or more bytes of data. After each byte is transmitted, the master sends an acknowledge bit; if the acknowledge bit is an ACK, SMB0DAT should be written with the next data byte. If the acknowledge bit is a NACK, SMB0DAT should not be written to before SI is cleared (Note: an error condition may be generated if SMB0DAT is written following a received NACK while in Slave Transmitter Mode). The interface exits Slave Transmitter Mode after receiving a STOP. Note that the interface will switch to Slave Receiver Mode if SMB0DAT is not written following a Slave Transmitter interrupt. Figure 19.8 shows a typical Slave Transmitter sequence. Two transmitted data bytes are shown, though any number of bytes may be transmitted. Notice that the 'data byte transferred' interrupts occur **after** the ACK cycle in this mode.

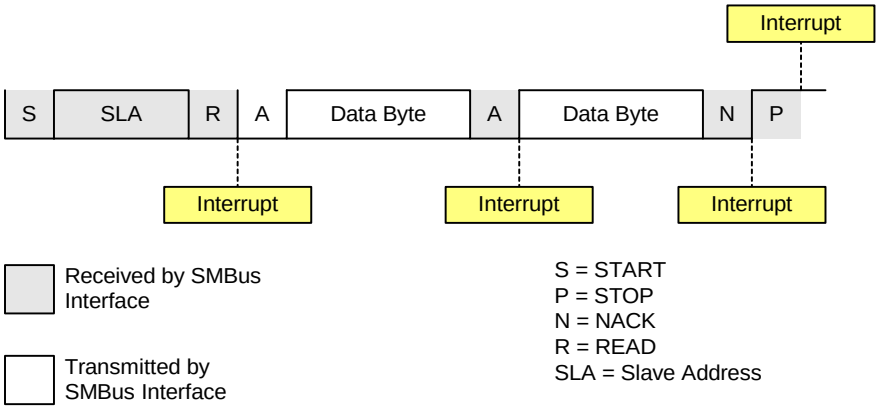



**Figure 19.8. Typical Slave Transmitter Sequence**

## 19.6. SMBus Status Decoding

The current SMBus status can be easily decoded using the SMB0CN register. In the table below, STATUS VECTOR refers to the four upper bits of SMB0CN: MASTER, TXMODE, STA, and STO. Note that the shown response options are only the typical responses; application-specific procedures are allowed as long as they conform to the SMBus specification. Highlighted responses are allowed but do not conform to the SMBus specification.

**Table 19.4. SMBus Status Decoding**

| Mode | Values Read | | | | Current SMbus State | Typical Response Options | Values Written | | |
|---|---|---|---|---|---|---|---|---|---|
| | Status Vector | ACKRQ | ARBLOST | ACK | | | STA | STO | ACK |
| Master Transmitter | 1110 | 0 | 0 | X | A master START was generated. | Load slave address + R/W into SMB0DAT. | 0 | 0 | X |
| | 1100 | 0 | 0 | 0 | A master data or address byte was transmitted; NACK received. | Set STA to restart transfer. | 1 | 0 | X |
| | | | | | | Abort transfer. | 0 | 1 | X |
| | | 0 | 0 | 1 | A master data or address byte was transmitted; ACK received. | Load next data byte into SMB0DAT. | 0 | 0 | X |
| | | | | | | End transfer with STOP. | 0 | 1 | X |
| | | | | | | End transfer with STOP and start another transfer. | 1 | 1 | X |
| | | | | | | Send repeated START. | 1 | 0 | X |
| | | | | | | Switch to Master Receiver Mode (clear SI without writing new data to SMB0DAT). | 0 | 0 | X |

**Table 19.4. SMBus Status Decoding (Continued)**

| Mode | Values Read | | | | Current SMbus State | Typical Response Options | Values Written | | |
|---|---|---|---|---|---|---|---|---|---|
| | Status Vector | ACKRQ | ARBLOST | ACK | | | STA | STO | ACK |
| Master Receiver | 1000 | 1 | 0 | X | A master data byte was received; ACK requested. | Acknowledge received byte; Read SMB0DAT. | 0 | 0 | 1 |
| | | | | | | Send NACK to indicate last byte, and send STOP. | 0 | 1 | 0 |
| | | | | | | Send NACK to indicate last byte, and send STOP followed by START. | 1 | 1 | 0 |
| | | | | | | Send ACK followed by repeated START. | 1 | 0 | 1 |
| | | | | | | Send NACK to indicate last byte, and send repeated START. | 1 | 0 | 0 |
| | | | | | | Send ACK and switch to Master Transmitter Mode (write to SMB0DAT before clearing SI). | 0 | 0 | 1 |
| | | | | | | Send NACK and switch to Master Transmitter Mode (write to SMB0DAT before clearing SI). | 0 | 0 | 0 |
| Slave Transmitter | 0100 | 0 | 0 | 0 | A slave byte was transmitted; NACK received. | No action required (expecting STOP condition). | 0 | 0 | X |
| | | 0 | 0 | 1 | A slave byte was transmitted; ACK received. | Load SMB0DAT with next data byte to transmit. | 0 | 0 | X |
| | | 0 | 1 | X | A Slave byte was transmitted; error detected. | No action required (expecting Master to end transfer). | 0 | 0 | X |
| | 0101 | 0 | X | X | A illegal STOP or bus error was detected while a Slave Transmission was in progress. | Clear STO. | 0 | 0 | X |

**Table 19.4. SMBus Status Decoding (Continued)**

| Mode | Values Read | | | | Current SMbus State | Typical Response Options | Values Written | | |
|---|---|---|---|---|---|---|---|---|---|
| | Status Vector | ACKRQ | ARBLOST | ACK | | | STA | STO | ACK |
| Slave Receiver | 0010 | 1 | 0 | X | A slave address was received; ACK requested. | Acknowledge received address. | 0 | 0 | 1 |
| | | | | | | Do not acknowledge received address. | 0 | 0 | 0 |
| | | 1 | 1 | X | Lost arbitration as master; slave address received; ACK requested. | Acknowledge received address. | 0 | 0 | 1 |
| | | | | | | Do not acknowledge received address. | 0 | 0 | 0 |
| | | | | | | Reschedule failed transfer; do not acknowledge received address. | 1 | 0 | 0 |
| | 0010 | 0 | 1 | X | Lost arbitration while attempting a repeated START. | Abort failed transfer. | 0 | 0 | X |
| | | | | | | Reschedule failed transfer. | 1 | 0 | X |
| | 0001 | 1 | 1 | X | Lost arbitration while attempting a STOP. | No action required (transfer complete/aborted). | 0 | 0 | 0 |
| | | 0 | 0 | X | A STOP was detected while addressed as a Slave Transmitter or Slave Receiver. | Clear STO. | 0 | 0 | X |
| | | 0 | 1 | X | Lost arbitration due to a detected STOP. | Abort transfer. | 0 | 0 | X |
| | | | | | | Reschedule failed transfer. | 1 | 0 | X |
| | 0000 | 1 | 0 | X | A slave byte was received; ACK requested. | Acknowledge received byte; Read SMB0DAT. | 0 | 0 | 1 |
| | | | | | | Do not acknowledge received byte. | 0 | 0 | 0 |
| | | 1 | 1 | X | Lost arbitration while transmitting a data byte as master. | Abort failed transfer. | 0 | 0 | 0 |
| | | | | | | Reschedule failed transfer. | 1 | 0 | 0 |

## NOTES:

# 20. UART0

UART0 is an asynchronous, full duplex serial port offering modes 1 and 3 of the standard 8051 UART. Enhanced baud rate support allows a wide range of clock sources to generate standard baud rates (details in Section “20.1. Enhanced Baud Rate Generation’ on page 172). Received data buffering allows UART0 to start reception of a second incoming data byte before software has finished reading the previous data byte.

UART0 has two associated SFRs: Serial Control Register 0 (SCON0) and Serial Data Buffer 0 (SBUF0). The single SBUF0 location provides access to both transmit and receive registers. **Writes to SBUF0 always access the Transmit register. Reads of SBUF0 always access the buffered Receive register; it is not possible to read data from the Transmit register.**

With UART0 interrupts enabled, an interrupt is generated each time a transmit is completed (TI0 is set in SCON0), or a data byte has been received (RI0 is set in SCON0). The UART0 interrupt flags are not cleared by hardware when the CPU vectors to the interrupt service routine. They must be cleared manually by software, allowing software to determine the cause of the UART0 interrupt (transmit complete or receive complete).

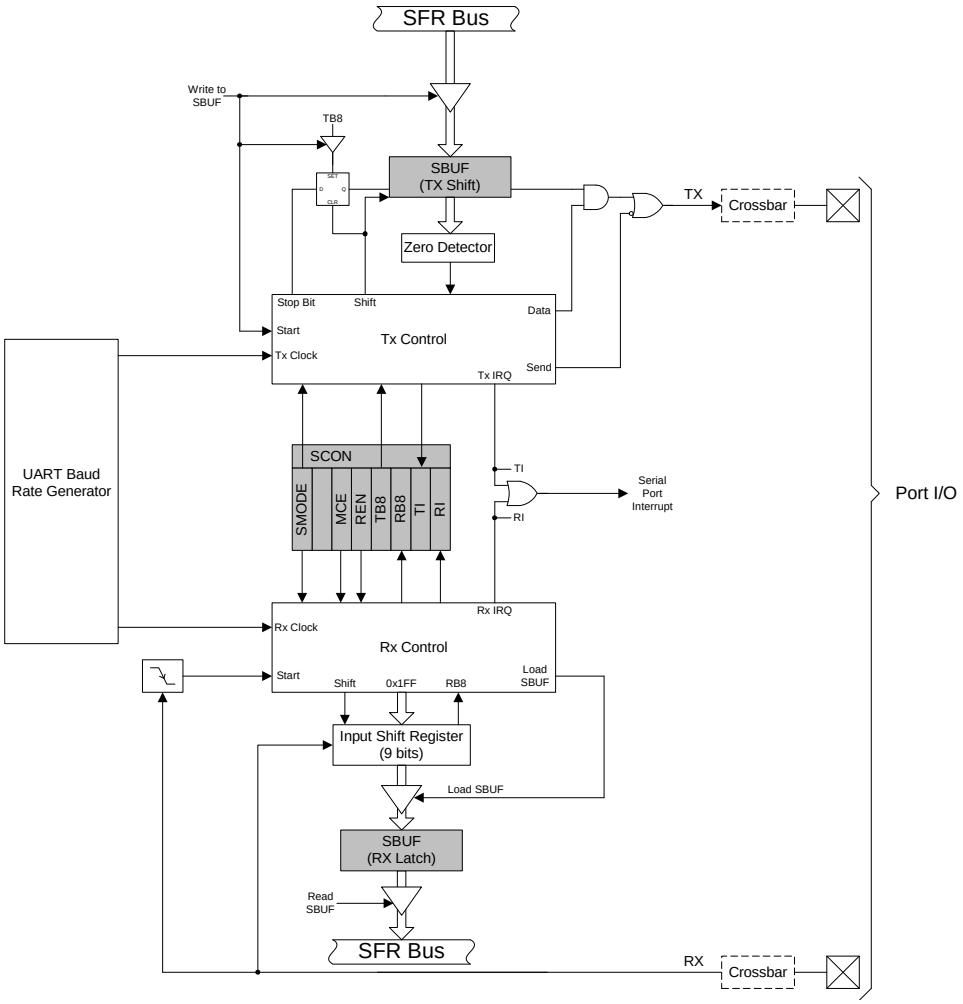



**Figure 20.1. UART0 Block Diagram**

## 20.1. Enhanced Baud Rate Generation

The UART0 baud rate is generated by Timer 1 in 8-bit auto-reload mode. The TX clock is generated by TL1; the RX clock is generated by a copy of TL1 (shown as RX Timer in Figure 20.2), which is not user-accessible. Both TX and RX Timer overflows are divided by two to generate the TX and RX baud rates. The RX Timer runs when Timer 1 is enabled, and uses the same reload value (TH1). However, an RX Timer reload is forced when a START condition is detected on the RX pin. This allows a receive to begin any time a START is detected, independent of the TX Timer state.

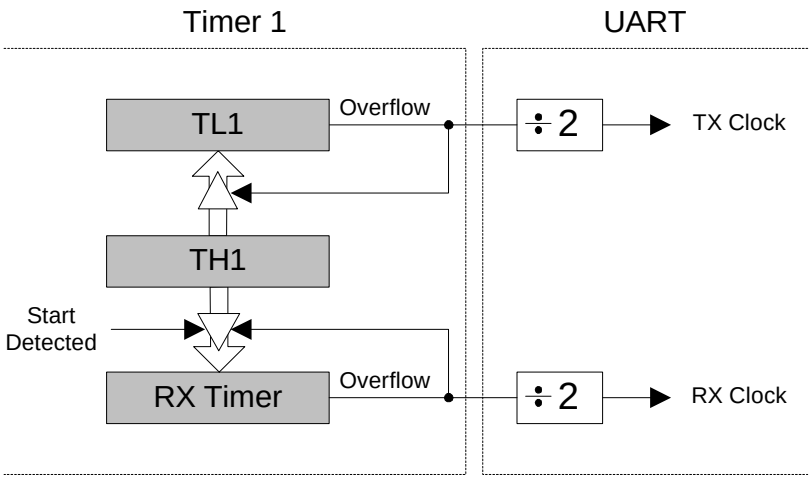



**Figure 20.2. UART0 Baud Rate Logic**

Timer 1 should be configured for Mode 2, 8-bit auto-reload (see Section “22.1.3. Mode 2: 8-bit Counter/Timer with Auto-Reload’ on page 197). The Timer 1 reload value should be set so that overflows will occur at two times the desired UART baud rate frequency. Note that Timer 1 may be clocked by one of six sources: SYSCLK, SYSCLK / 4, SYSCLK / 12, SYSCLK / 48, the external oscillator clock / 8, or an external input T1. The UART0 baud rate is determined by Equation 20.1-A and Equation 20.1-B.

$$\text{A)}\quad \text{UartBaudRate} = \frac{1}{2} \times \text{T1_Overflow_Rate}$$

$$\text{B)}\quad \text{T1_Overflow_Rate} = \frac{\text{T1}_{\text{CLK}}}{256 - \text{TH1}}$$

**Equation 20.1. UART0 Baud Rate**

Where $T1_{CLK}$ is the frequency of the clock supplied to Timer 1, and *T1H* is the high byte of Timer 1 (reload value). Timer 1 clock frequency is selected as described in Section “22. Timers’ on page 195. A quick reference for typical baud rates and system clock frequencies is given in Table 20.1 through Table 20.6. Note that the internal oscillator may still generate the system clock when the external oscillator is driving Timer 1.

## 20.2. Operational Modes

UART0 provides standard asynchronous, full duplex communication. The UART mode (8-bit or 9-bit) is selected by the S0MODE bit (SCON0.7). Typical UART connection options are shown below.

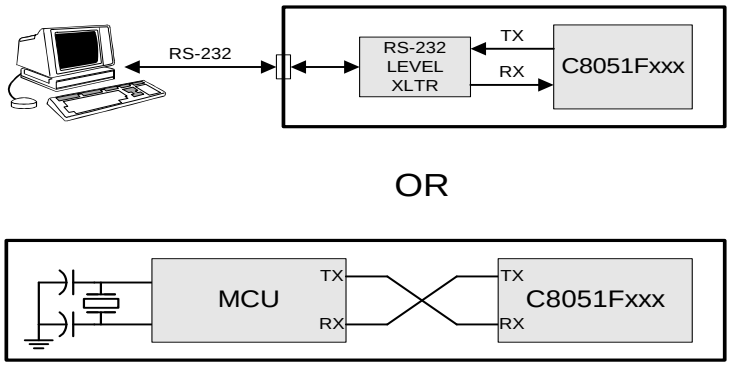



**Figure 20.3. UART Interconnect Diagram**

### 20.2.1. 8-Bit UART

8-Bit UART mode uses a total of 10 bits per data byte: one start bit, eight data bits (LSB first), and one stop bit. Data are transmitted LSB first from the TX0 pin and received at the RX0 pin. On receive, the eight data bits are stored in SBUF0 and the stop bit goes into RB80 (SCON0.2).

Data transmission begins when software writes a data byte to the SBUF0 register. The TI0 Transmit Interrupt Flag (SCON0.1) is set at the end of the transmission (the beginning of the stop-bit time). Data reception can begin any time after the REN0 Receive Enable bit (SCON0.4) is set to logic 1. After the stop bit is received, the data byte will be loaded into the SBUF0 receive register if the following conditions are met: RI0 must be logic 0, and if MCE0 is logic 1, the stop bit must be logic 1. In the event of a receive data overrun, the first received 8 bits are latched into the SBUF0 receive register and the following overrun data bits are lost.

If these conditions are met, the eight bits of data is stored in SBUF0, the stop bit is stored in RB80 and the RI0 flag is set. If these conditions are not met, SBUF0 and RB80 will not be loaded and the RI0 flag will not be set. An interrupt will occur if enabled when either TI0 or RI0 is set.

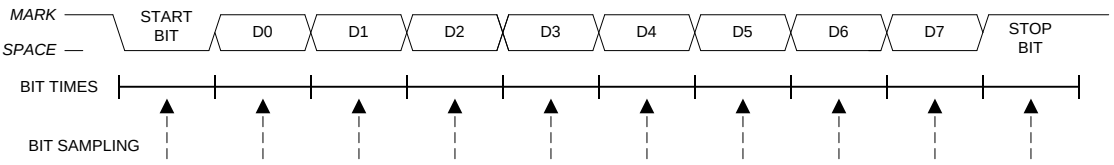



**Figure 20.4. 8-Bit UART Timing Diagram**

### 20.2.2. 9-Bit UART

9-bit UART mode uses a total of eleven bits per data byte: a start bit, 8 data bits (LSB first), a programmable ninth data bit, and a stop bit. The state of the ninth transmit data bit is determined by the value in TB80 (SCON0.3), which is assigned by user software. It can be assigned the value of the parity flag (bit P in register PSW) for error detection, or used in multiprocessor communications. On receive, the ninth data bit goes into RB80 (SCON0.2) and the stop bit is ignored.

Data transmission begins when an instruction writes a data byte to the SBUF0 register. The TI0 Transmit Interrupt Flag (SCON0.1) is set at the end of the transmission (the beginning of the stop-bit time). Data reception can begin any time after the REN0 Receive Enable bit (SCON0.4) is set to '1'. After the stop bit is received, the data byte will be loaded into the SBUF0 receive register if the following conditions are met: (1) RI0 must be logic 0, and (2) if MCE0 is logic 1, the 9th bit must be logic 1 (when MCE0 is logic 0, the state of the ninth data bit is unimportant). If these conditions are met, the eight bits of data are stored in SBUF0, the ninth bit is stored in RB80, and the RI0 flag is set to '1'. If the above conditions are not met, SBUF0 and RB80 will not be loaded and the RI0 flag will not be set to '1'. A UART0 interrupt will occur if enabled when either TI0 or RI0 is set to '1'.

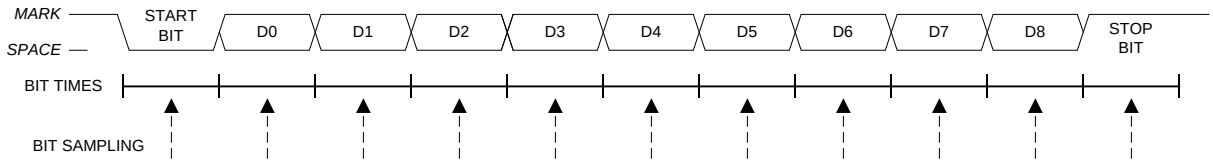



**Figure 20.5. 9-Bit UART Timing Diagram**

## 20.3. Multiprocessor Communications

9-Bit UART mode supports multiprocessor communication between a master processor and one or more slave processors by special use of the ninth data bit. When a master processor wants to transmit to one or more slaves, it first sends an address byte to select the target(s). An address byte differs from a data byte in that its ninth bit is logic 1; in a data byte, the ninth bit is always set to logic 0.

Setting the MCE0 bit (SCON0.5) of a slave processor configures its UART such that when a stop bit is received, the UART will generate an interrupt only if the ninth bit is logic 1 (RB80 = 1) signifying an address byte has been received. In the UART interrupt handler, software will compare the received address with the slave's own assigned 8-bit address. If the addresses match, the slave will clear its MCE0 bit to enable interrupts on the reception of the following data byte(s). Slaves that weren't addressed leave their MCE0 bits set and do not generate interrupts on the reception of the following data bytes, thereby ignoring the data. Once the entire message is received, the addressed slave resets its MCE0 bit to ignore all transmissions until it receives the next address byte.

Multiple addresses can be assigned to a single slave and/or a single address can be assigned to multiple slaves, thereby enabling "broadcast" transmissions to more than one slave simultaneously. The master processor can be configured to receive all transmissions or a protocol can be implemented such that the master/slave role is temporarily reversed to enable half-duplex transmission between the original master and slave(s).

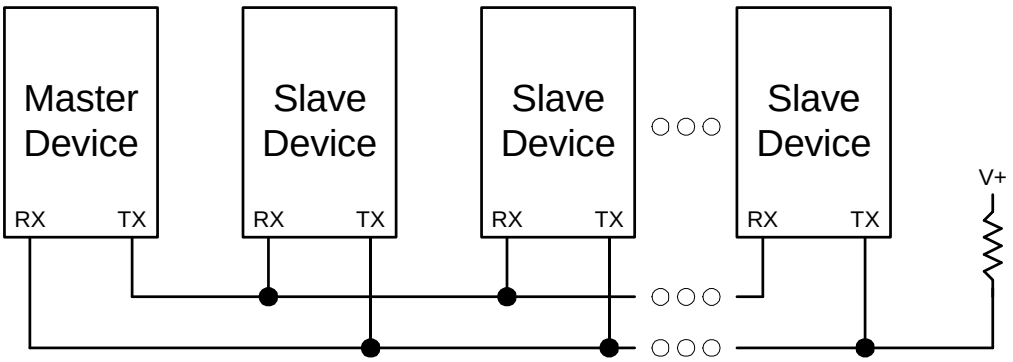



**Figure 20.6. UART Multi-Processor Mode Interconnect Diagram**

## SFR Definition 20.1. SCON0: Serial Port 0 Control

| R/W | R | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| S0MODE | — | MCE0 | REN0 | TB80 | RB80 | TI0 | RI0 | 01000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | Bit Addressable |

SFR Address: 0x98

Bit7: S0MODE: Serial Port 0 Operation Mode.
This bit selects the UART0 Operation Mode.
0: 8-bit UART with Variable Baud Rate.
1: 9-bit UART with Variable Baud Rate.

Bit6: UNUSED. Read = 1b. Write = don't care.

Bit5: MCE0: Multiprocessor Communication Enable.
The function of this bit is dependent on the Serial Port 0 Operation Mode.
S0MODE = 0: Checks for valid stop bit.
0: Logic level of stop bit is ignored.
1: RI0 will only be activated if stop bit is logic level 1.
S0MODE = 1: Multiprocessor Communications Enable.
0: Logic level of ninth bit is ignored.
1: RI0 is set and an interrupt is generated only when the ninth bit is logic 1.

Bit4: REN0: Receive Enable.
This bit enables/disables the UART receiver.
0: UART0 reception disabled.
1: UART0 reception enabled.

Bit3: TB80: Ninth Transmission Bit.
The logic level of this bit will be assigned to the ninth transmission bit in 9-bit UART Mode. It is not used in 8-bit UART Mode. Set or cleared by software as required.

Bit2: RB80: Ninth Receive Bit.
RB80 is assigned the value of the STOP bit in Mode 0; it is assigned the value of the 9th data bit in Mode 1.

Bit1: TI0: Transmit Interrupt Flag.
Set by hardware when a byte of data has been transmitted by UART0 (after the 8th bit in 8-bit UART Mode, or at the beginning of the STOP bit in 9-bit UART Mode). When the UART0 interrupt is enabled, setting this bit causes the CPU to vector to the UART0 interrupt service routine. This bit must be cleared manually by software.

Bit0: RI0: Receive Interrupt Flag.
Set to '1' by hardware when a byte of data has been received by UART0 (set at the STOP bit sampling time). When the UART0 interrupt is enabled, setting this bit to '1' causes the CPU to vector to the UART0 interrupt service routine. This bit must be cleared manually by software.

## SFR Definition 20.2. SBUF0: Serial (UART0) Port Data Buffer

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0x99

Bits7–0: SBUF0[7:0]: Serial Data Buffer Bits 7–0 (MSB–LSB)
This SFR accesses two registers; a transmit shift register and a receive latch register. When data is written to SBUF0, it goes to the transmit shift register and is held for serial transmission. Writing a byte to SBUF0 initiates the transmission. A read of SBUF0 returns the contents of the receive latch.

**Table 20.1. Timer Settings for Standard Baud Rates Using the Internal Oscillator**

| | Frequency: 24.5 MHz | | | | | | |
|---|---|---|---|---|---|---|---|
| | Target Baud Rate (bps) | Baud Rate % Error | Oscillator Divide Factor | Timer Clock Source | SCA1–SCA0 (pre-scale select)* | T1M* | Timer 1 Reload Value (hex) |
| SYSCLK from Internal Osc. | 230400 | –0.32% | 106 | SYSCLK | XX | 1 | 0xCB |
| | 115200 | –0.32% | 212 | SYSCLK | XX | 1 | 0x96 |
| | 57600 | 0.15% | 426 | SYSCLK | XX | 1 | 0x2B |
| | 28800 | –0.32% | 848 | SYSCLK / 4 | 01 | 0 | 0x96 |
| | 14400 | 0.15% | 1704 | SYSCLK / 12 | 00 | 0 | 0xB9 |
| | 9600 | –0.32% | 2544 | SYSCLK / 12 | 00 | 0 | 0x96 |
| | 2400 | –0.32% | 10176 | SYSCLK / 48 | 10 | 0 | 0x96 |
| | 1200 | 0.15% | 20448 | SYSCLK / 48 | 10 | 0 | 0x2B |

X = Don't care

***Note:** SCA1–SCA0 and T1M bit definitions can be found in Section 22.1.

**Table 20.2. Timer Settings for Standard Baud Rates Using an External 25.0 MHz Oscillator**

| | Frequency: 25.0 MHz | | | | | | |
|---|---|---|---|---|---|---|---|
| | Target Baud Rate (bps) | Baud Rate % Error | Oscillator Divide Factor | Timer Clock Source | SCA1–SCA0 (pre-scale select)* | T1M* | Timer 1 Reload Value (hex) |
| SYSCLK from External Osc. | 230400 | –0.47% | 108 | SYSCLK | XX | 1 | 0xCA |
| | 115200 | 0.45% | 218 | SYSCLK | XX | 1 | 0x93 |
| | 57600 | –0.01% | 434 | SYSCLK | XX | 1 | 0x27 |
| | 28800 | 0.45% | 872 | SYSCLK / 4 | 01 | 0 | 0x93 |
| | 14400 | –0.01% | 1736 | SYSCLK / 4 | 01 | 0 | 0x27 |
| | 9600 | 0.15% | 2608 | EXTCLK / 8 | 11 | 0 | 0x5D |
| | 2400 | 0.45% | 10464 | SYSCLK / 48 | 10 | 0 | 0x93 |
| | 1200 | –0.01% | 20832 | SYSCLK / 48 | 10 | 0 | 0x27 |
| SYSCLK from Internal Osc. | 57600 | –0.47% | 432 | EXTCLK / 8 | 11 | 0 | 0xE5 |
| | 28800 | –0.47% | 864 | EXTCLK / 8 | 11 | 0 | 0xCA |
| | 14400 | 0.45% | 1744 | EXTCLK / 8 | 11 | 0 | 0x93 |
| | 9600 | 0.15% | 2608 | EXTCLK / 8 | 11 | 0 | 0x5D |

X = Don't care

***Note:** SCA1–SCA0 and T1M bit definitions can be found in Section 22.1.

**Table 20.3. Timer Settings for Standard Baud Rates Using an External 22.1184 MHz Oscillator**

| | Frequency: 22.1184 MHz | | | | | | |
|---|---|---|---|---|---|---|---|
| | Target Baud Rate (bps) | Baud Rate % Error | Oscilla-tor Divide Factor | Timer Clock Source | SCA1–SCA0 (pre-scale select)* | T1M* | Timer 1 Reload Value (hex) |
| SYSCLK from External Osc. | 230400 | 0.00% | 96 | SYSCLK | XX | 1 | 0xD0 |
| | 115200 | 0.00% | 192 | SYSCLK | XX | 1 | 0xA0 |
| | 57600 | 0.00% | 384 | SYSCLK | XX | 1 | 0x40 |
| | 28800 | 0.00% | 768 | SYSCLK / 12 | 00 | 0 | 0xE0 |
| | 14400 | 0.00% | 1536 | SYSCLK / 12 | 00 | 0 | 0xC0 |
| | 9600 | 0.00% | 2304 | SYSCLK / 12 | 00 | 0 | 0xA0 |
| | 2400 | 0.00% | 9216 | SYSCLK / 48 | 10 | 0 | 0xA0 |
| | 1200 | 0.00% | 18432 | SYSCLK / 48 | 10 | 0 | 0x40 |
| SYSCLK from Internal Osc. | 230400 | 0.00% | 96 | EXTCLK / 8 | 11 | 0 | 0xFA |
| | 115200 | 0.00% | 192 | EXTCLK / 8 | 11 | 0 | 0xF4 |
| | 57600 | 0.00% | 384 | EXTCLK / 8 | 11 | 0 | 0xE8 |
| | 28800 | 0.00% | 768 | EXTCLK / 8 | 11 | 0 | 0xD0 |
| | 14400 | 0.00% | 1536 | EXTCLK / 8 | 11 | 0 | 0xA0 |
| | 9600 | 0.00% | 2304 | EXTCLK / 8 | 11 | 0 | 0x70 |

X = Don't care

***Note:** SCA1–SCA0 and T1M bit definitions can be found in Section 22.1.

**Table 20.4. Timer Settings for Standard Baud Rates Using an External 18.432 MHz Oscillator**

| | Frequency: 18.432 MHz | | | | | | |
|---|---|---|---|---|---|---|---|
| | Target Baud Rate (bps) | Baud Rate % Error | Oscilla-tor Divide Factor | Timer Clock Source | SCA1–SCA0 (pre-scale select)* | T1M* | Timer 1 Reload Value (hex) |
| SYSCLK from External Osc. | 230400 | 0.00% | 80 | SYSCLK | XX | 1 | 0xD8 |
| | 115200 | 0.00% | 160 | SYSCLK | XX | 1 | 0xB0 |
| | 57600 | 0.00% | 320 | SYSCLK | XX | 1 | 0x60 |
| | 28800 | 0.00% | 640 | SYSCLK / 4 | 01 | 0 | 0xB0 |
| | 14400 | 0.00% | 1280 | SYSCLK / 4 | 01 | 0 | 0x60 |
| | 9600 | 0.00% | 1920 | SYSCLK / 12 | 00 | 0 | 0xB0 |
| | 2400 | 0.00% | 7680 | SYSCLK / 48 | 10 | 0 | 0xB0 |
| | 1200 | 0.00% | 15360 | SYSCLK / 48 | 10 | 0 | 0x60 |
| SYSCLK from Internal Osc. | 230400 | 0.00% | 80 | EXTCLK / 8 | 11 | 0 | 0xFB |
| | 115200 | 0.00% | 160 | EXTCLK / 8 | 11 | 0 | 0xF6 |
| | 57600 | 0.00% | 320 | EXTCLK / 8 | 11 | 0 | 0xEC |
| | 28800 | 0.00% | 640 | EXTCLK / 8 | 11 | 0 | 0xD8 |
| | 14400 | 0.00% | 1280 | EXTCLK / 8 | 11 | 0 | 0xB0 |
| | 9600 | 0.00% | 1920 | EXTCLK / 8 | 11 | 0 | 0x88 |

X = Don't care

***Note:** SCA1–SCA0 and T1M bit definitions can be found in Section 22.1.

**Table 20.5. Timer Settings for Standard Baud Rates Using an External 11.0592 MHz Oscillator**

| | Frequency: 11.0592 MHz | | | | | | |
|---|---|---|---|---|---|---|---|
| | Target Baud Rate (bps) | Baud Rate % Error | Oscilla-tor Divide Factor | Timer Clock Source | SCA1–SCA0 (pre-scale select)* | T1M* | Timer 1 Reload Value (hex) |
| SYSCLK from External Osc. | 230400 | 0.00% | 48 | SYSCLK | XX | 1 | 0xE8 |
| | 115200 | 0.00% | 96 | SYSCLK | XX | 1 | 0xD0 |
| | 57600 | 0.00% | 192 | SYSCLK | XX | 1 | 0xA0 |
| | 28800 | 0.00% | 384 | SYSCLK | XX | 1 | 0x40 |
| | 14400 | 0.00% | 768 | SYSCLK / 12 | 00 | 0 | 0xE0 |
| | 9600 | 0.00% | 1152 | SYSCLK / 12 | 00 | 0 | 0xD0 |
| | 2400 | 0.00% | 4608 | SYSCLK / 12 | 00 | 0 | 0x40 |
| | 1200 | 0.00% | 9216 | SYSCLK / 48 | 10 | 0 | 0xA0 |
| SYSCLK from Internal Osc. | 230400 | 0.00% | 48 | EXTCLK / 8 | 11 | 0 | 0xFD |
| | 115200 | 0.00% | 96 | EXTCLK / 8 | 11 | 0 | 0xFA |
| | 57600 | 0.00% | 192 | EXTCLK / 8 | 11 | 0 | 0xF4 |
| | 28800 | 0.00% | 384 | EXTCLK / 8 | 11 | 0 | 0xE8 |
| | 14400 | 0.00% | 768 | EXTCLK / 8 | 11 | 0 | 0xD0 |
| | 9600 | 0.00% | 1152 | EXTCLK / 8 | 11 | 0 | 0xB8 |

X = Don't care

***Note:** SCA1–SCA0 and T1M bit definitions can be found in Section 22.1.

**Table 20.6. Timer Settings for Standard Baud Rates Using an External 3.6864 MHz Oscillator**

| | Frequency: 3.6864 MHz | | | | | | |
|---|---|---|---|---|---|---|---|
| | Target Baud Rate (bps) | Baud Rate% Error | Oscilla-tor Divide Factor | Timer Clock Source | SCA1–SCA0 (pre-scale select)* | T1M* | Timer 1 Reload Value (hex) |
| SYSCLK from External Osc. | 230400 | 0.00% | 16 | SYSCLK | XX | 1 | 0xF8 |
| | 115200 | 0.00% | 32 | SYSCLK | XX | 1 | 0xF0 |
| | 57600 | 0.00% | 64 | SYSCLK | XX | 1 | 0xE0 |
| | 28800 | 0.00% | 128 | SYSCLK | XX | 1 | 0xC0 |
| | 14400 | 0.00% | 256 | SYSCLK | XX | 1 | 0x80 |
| | 9600 | 0.00% | 384 | SYSCLK | XX | 1 | 0x40 |
| | 2400 | 0.00% | 1536 | SYSCLK / 12 | 00 | 0 | 0xC0 |
| | 1200 | 0.00% | 3072 | SYSCLK / 12 | 00 | 0 | 0x80 |
| SYSCLK from Internal Osc. | 230400 | 0.00% | 16 | EXTCLK / 8 | 11 | 0 | 0xFF |
| | 115200 | 0.00% | 32 | EXTCLK / 8 | 11 | 0 | 0xFE |
| | 57600 | 0.00% | 64 | EXTCLK / 8 | 11 | 0 | 0xFC |
| | 28800 | 0.00% | 128 | EXTCLK / 8 | 11 | 0 | 0xF8 |
| | 14400 | 0.00% | 256 | EXTCLK / 8 | 11 | 0 | 0xF0 |
| | 9600 | 0.00% | 384 | EXTCLK / 8 | 11 | 0 | 0xE8 |

X = Don't care

***Note:** SCA1–SCA0 and T1M bit definitions can be found in Section 22.1.

# 21. Serial Peripheral Interface (SPI0)

The Serial Peripheral Interface (SPI0) provides access to a flexible, full-duplex synchronous serial bus. SPI0 can operate as a master or slave device in both 3-wire or 4-wire modes, and supports multiple masters and slaves on a single SPI bus. The slave-select (NSS) signal can be configured as an input to select SPI0 in slave mode, or to disable Master Mode operation in a multi-master environment, avoiding contention on the SPI bus when more than one master attempts simultaneous data transfers. NSS can also be configured as a chip-select output in master mode, or disabled for 3-wire operation. Additional general purpose port I/O pins can be used to select multiple slave devices in master mode.

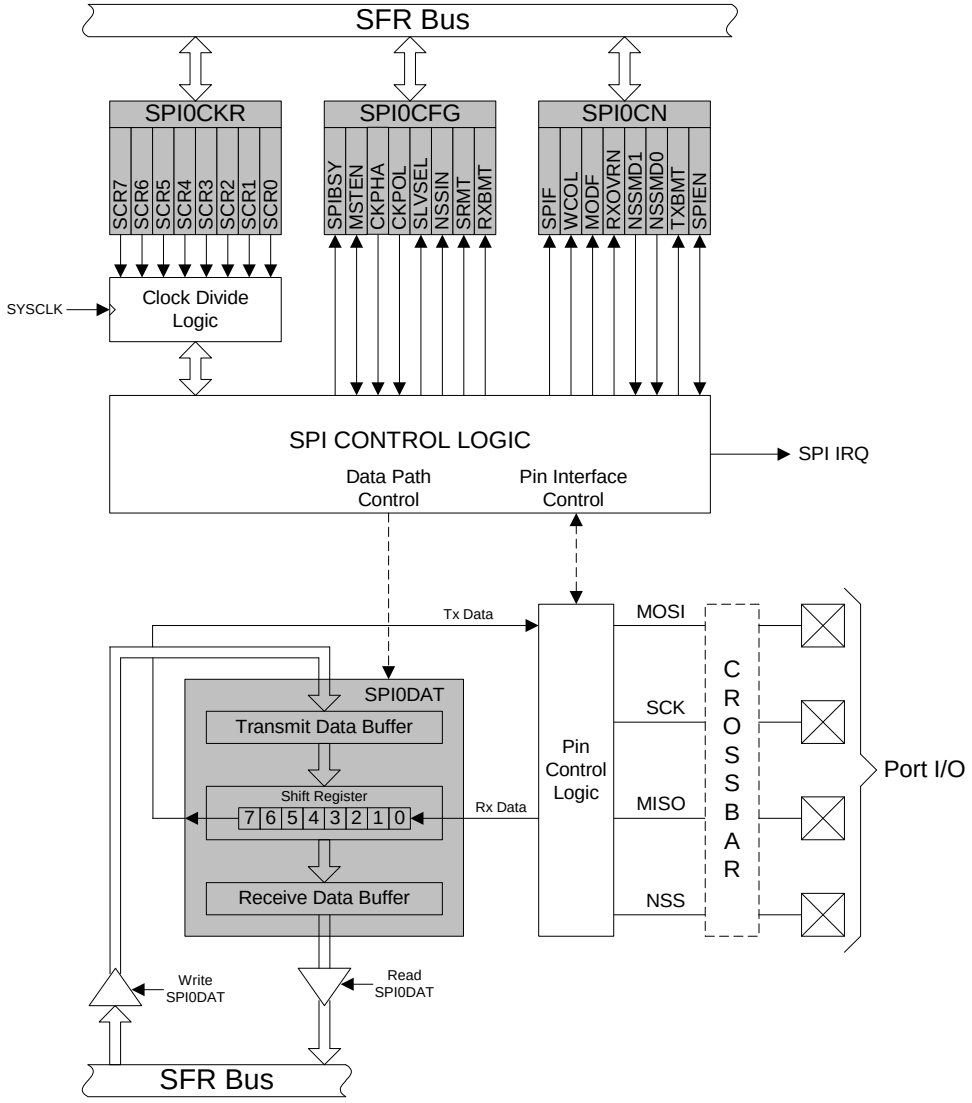



**Figure 21.1. SPI Block Diagram**

## 21.1. Signal Descriptions

The four signals used by SPI0 (MOSI, MISO, SCK, NSS) are described below.

### 21.1.1. Master Out, Slave In (MOSI)

The master-out, slave-in (MOSI) signal is an output from a master device and an input to slave devices. It is used to serially transfer data from the master to the slave. This signal is an output when SPI0 is operating as a master and an input when SPI0 is operating as a slave. Data is transferred most-significant bit first. When configured as a master, MOSI is driven by the MSB of the shift register in both 3- and 4-wire mode.

### 21.1.2. Master In, Slave Out (MISO)

The master-in, slave-out (MISO) signal is an output from a slave device and an input to the master device. It is used to serially transfer data from the slave to the master. This signal is an input when SPI0 is operating as a master and an output when SPI0 is operating as a slave. Data is transferred most-significant bit first. The MISO pin is placed in a high-impedance state when the SPI module is disabled and when the SPI operates in 4-wire mode as a slave that is not selected. When acting as a slave in 3-wire mode, MISO is always driven by the MSB of the shift register.

### 21.1.3. Serial Clock (SCK)

The serial clock (SCK) signal is an output from the master device and an input to slave devices. It is used to synchronize the transfer of data between the master and slave on the MOSI and MISO lines. SPI0 generates this signal when operating as a master. The SCK signal is ignored by a SPI slave when the slave is not selected (NSS = 1) in 4-wire slave mode.

### 21.1.4. Slave Select (NSS)

The function of the slave-select (NSS) signal is dependent on the setting of the NSSMD1 and NSSMD0 bits in the SPI0CN register. There are three possible modes that can be selected with these bits:

1. NSSMD[1:0] = 00: 3-Wire Master or 3-Wire Slave Mode: SPI0 operates in 3-wire mode, and NSS is disabled. When operating as a slave device, SPI0 is always selected in 3-wire mode. Since no select signal is present, SPI0 must be the only slave on the bus in 3-wire mode. This is intended for point-to-point communication between a master and one slave.
2. NSSMD[1:0] = 01: 4-Wire Slave or Multi-Master Mode: SPI0 operates in 4-wire mode, and NSS is enabled as an input. When operating as a slave, NSS selects the SPI0 device. When operating as a master, a 1-to-0 transition of the NSS signal disables the master function of SPI0 so that multiple master devices can be used on the same SPI bus.
3. NSSMD[1:0] = 1x: 4-Wire Master Mode: SPI0 operates in 4-wire mode, and NSS is enabled as an output. The setting of NSSMD0 determines what logic level the NSS pin will output. This configuration should only be used when operating SPI0 as a master device.

See Figure 21.2, Figure 21.3, and Figure 21.4 for typical connection diagrams of the various operational modes. **Note that the setting of NSSMD bits affects the pinout of the device.** When in 3-wire master or 3-wire slave mode, the NSS pin will not be mapped by the crossbar. In all other modes, the NSS signal will be mapped to a pin on the device. See Section “18. Port Input/Output’ on page 137 for general purpose port I/O and crossbar information.

## 21.2. SPI0 Master Mode Operation

A SPI master device initiates all data transfers on a SPI bus. SPI0 is placed in master mode by setting the Master Enable flag (MSTEN, SPI0CN.6). Writing a byte of data to the SPI0 data register (SPI0DAT) when in master mode writes to the transmit buffer. If the SPI shift register is empty, the byte in the transmit buffer is moved to the shift register, and a data transfer begins. The SPI0 master immediately shifts out the data serially on the MOSI line while providing the serial clock on SCK. The SPIF (SPI0CN.7) flag is set to logic 1 at the end of the transfer. If interrupts are enabled, an interrupt request is generated when the SPIF flag is set. While the SPI0 master transfers data to a slave on the MOSI line, the addressed SPI slave device simultaneously transfers data to the SPI master on the MISO line in a full-duplex operation. Therefore, the SPIF flag serves as both a transmit-complete and receive-data-ready flag. The data byte received from the slave is transferred MSB-first into the master's shift register. When a byte is fully shifted into the register, it is moved to the receive buffer where it can be read by the processor by reading SPI0DAT.

When configured as a master, SPI0 can operate in one of three different modes: multi-master mode, 3-wire single-master mode, and 4-wire single-master mode. The default, multi-master mode is active when NSSMD1 (SPI0CN.3) = 0 and NSSMD0 (SPI0CN.2) = 1. In this mode, NSS is an input to the device, and is used to disable the master SPI0 when another master is accessing the bus. When NSS is pulled low in this mode, MSTEN (SPI0CN.6) and SPIEN (SPI0CN.0) are set to 0 to disable the SPI master device, and a Mode Fault is generated (MODF, SPI0CN.5 = 1). Mode Fault will generate an interrupt if enabled. SPI0 must be manually re-enabled in software under these circumstances. In multi-master systems, devices will typically default to being slave devices while they are not acting as the system master device. In multi-master mode, slave devices can be addressed individually (if needed) using general-purpose I/O pins. Figure 21.2 shows a connection diagram between two master devices in multiple-master mode.

3-wire single-master mode is active when NSSMD1 (SPI0CN.3) = 0 and NSSMD0 (SPI0CN.2) = 0. In this mode, NSS is not used, and is not mapped to an external port pin through the crossbar. Any slave devices that must be addressed in this mode should be selected using general-purpose I/O pins. Figure 21.3 shows a connection diagram between a master device in 3-wire master mode and a slave device.

4-wire single-master mode is active when NSSMD1 (SPI0CN.3) = 1. In this mode, NSS is configured as an output pin, and can be used as a slave-select signal for a single SPI device. In this mode, the output value of NSS is controlled (in software) with the bit NSSMD0 (SPI0CN.2). Additional slave devices can be addressed using general-purpose I/O pins. Figure 21.4 shows a connection diagram for a master device in 4-wire master mode and two slave devices.

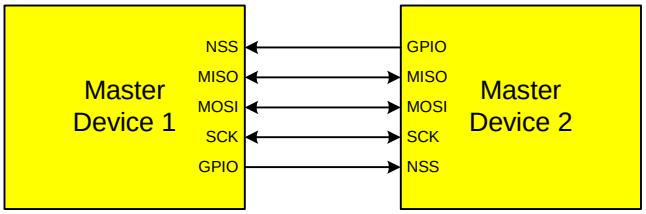



**Figure 21.2. Multiple-Master Mode Connection Diagram**

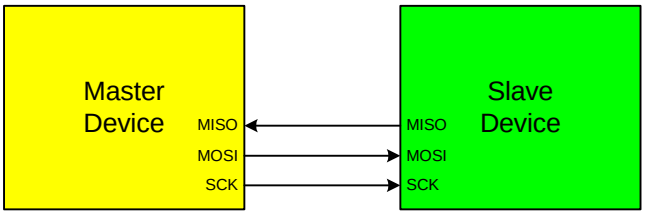



**Figure 21.3. 3-Wire Single Master and Slave Mode Connection Diagram**

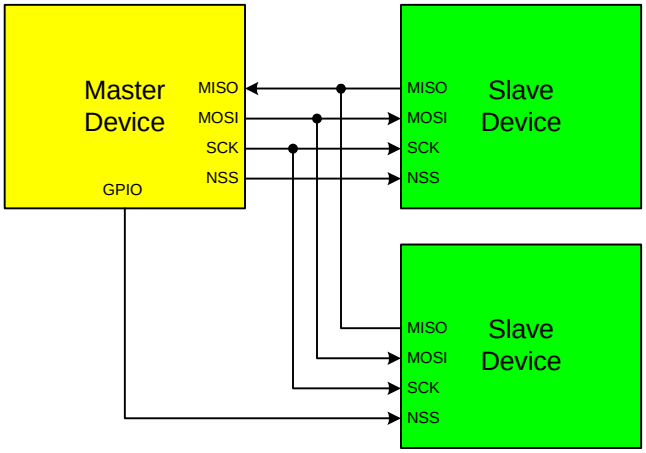



**Figure 21.4. 4-Wire Single Master and Slave Mode Connection Diagram**

## 21.3. SPI0 Slave Mode Operation

When SPI0 is enabled and not configured as a master, it will operate as a SPI slave. As a slave, bytes are shifted in through the MOSI pin and out through the MISO pin by a master device controlling the SCK signal. A bit counter in the SPI0 logic counts SCK edges. When 8 bits have been shifted into the shift register, the SPIF flag is set to logic 1, and the byte is copied into the receive buffer. Data is read from the receive buffer by reading SPI0DAT. A slave device cannot initiate transfers. Data to be transferred to the master device is pre-loaded into the shift register by writing to SPI0DAT. Writes to SPI0DAT are double-buffered, and are placed in the transmit buffer first. If the shift register is empty, the contents of the transmit buffer will immediately be transferred into the shift register. When the shift register already contains data, the SPI will load the shift register with the transmit buffer's contents after the last SCK edge of the next (or current) SPI transfer.

When configured as a slave, SPI0 can be configured for 4-wire or 3-wire operation. The default, 4-wire slave mode, is active when NSSMD1 (SPI0CN.3) = 0 and NSSMD0 (SPI0CN.2) = 1. In 4-wire mode, the NSS signal is routed to a port pin and configured as a digital input. SPI0 is enabled when NSS is logic 0, and disabled when NSS is logic 1. The bit counter is reset on a falling edge of NSS. Note that the NSS signal must be driven low at least 2 system clocks before the first active edge of SCK for each byte transfer. Figure 21.4 shows a connection diagram between two slave devices in 4-wire slave mode and a master device.

3-wire slave mode is active when NSSMD1 (SPI0CN.3) = 0 and NSSMD0 (SPI0CN.2) = 0. NSS is not used in this mode, and is not mapped to an external port pin through the crossbar. Since there is not a way of uniquely addressing the device in 3-wire slave mode, SPI0 must be the only slave device present on the bus. It is important to note that in 3-wire slave mode there is no external means of resetting the bit counter that determines when a full byte has been received. The bit counter can only be reset by disabling and re-enabling SPI0 with the SPIEN bit. Figure 21.3 shows a connection diagram between a slave device in 3-wire slave mode and a master device.

## 21.4. SPI0 Interrupt Sources

When SPI0 interrupts are enabled, the following four flags will generate an interrupt when they are set to logic 1:

*Note that all of the following bits must be cleared by software.*

1. The SPI Interrupt Flag, SPIF (SPI0CN.7) is set to logic 1 at the end of each byte transfer. This flag can occur in all SPI0 modes.
2. The Write Collision Flag, WCOL (SPI0CN.6) is set to logic 1 if a write to SPI0DAT is attempted when the transmit buffer has not been emptied to the SPI shift register. When this occurs, the write to SPI0DAT will be ignored, and the transmit buffer will not be written.This flag can occur in all SPI0 modes.
3. The Mode Fault Flag MODF (SPI0CN.5) is set to logic 1 when SPI0 is configured as a master, and for multi-master mode and the NSS pin is pulled low. When a Mode Fault occurs, the MSTEN and SPIEN bits in SPI0CN are set to logic 0 to disable SPI0 and allow another master device to access the bus.
4. The Receive Overrun Flag RXOVRN (SPI0CN.4) is set to logic 1 when configured as a slave, and a transfer is completed while the receive buffer still holds an unread byte from a previous transfer. The new byte is not transferred to the receive buffer, allowing the previously received data byte to be read. The data byte which caused the overrun is lost.

## 21.5. Serial Clock Timing

Four combinations of serial clock phase and polarity can be selected using the clock control bits in the SPI0 Configuration Register (SPI0CFG). The CKPHA bit (SPI0CFG.5) selects one of two clock phases (edge used to latch the data). The CKPOL bit (SPI0CFG.4) selects between an active-high or active-low clock. Both master and slave devices must be configured to use the same clock phase and polarity. SPI0 should be disabled (by clearing the SPIEN bit, SPI0CN.0) when changing the clock phase or polarity. The clock and data line relationships are shown in Figure 21.5.

The SPI0 Clock Rate Register (SPI0CKR) as shown in SFR Definition 21.3 controls the master mode serial clock frequency. This register is ignored when operating in slave mode. When the SPI is configured as a master, the maximum data transfer rate (bits/sec) is one-half the system clock frequency or 12.5 MHz, whichever is slower. When the SPI is configured as a slave, the maximum data transfer rate (bits/sec) for full-duplex operation is 1/10 the system clock frequency, provided that the master issues SCK, NSS (in 4-wire slave mode), and the serial input data synchronously with the slave's system clock. If the master issues SCK, NSS, and the serial input data asynchronously, the maximum data transfer rate (bits/sec) must be less than 1/10 the system clock frequency. In the special case where the master only wants to transmit data to the slave and does not need to receive data from the slave (i.e. half-duplex operation), the SPI slave can receive data at a maximum data transfer rate (bits/sec) of 1/4 the system clock frequency. This is provided that the master issues SCK, NSS, and the serial input data synchronously with the slave's system clock.

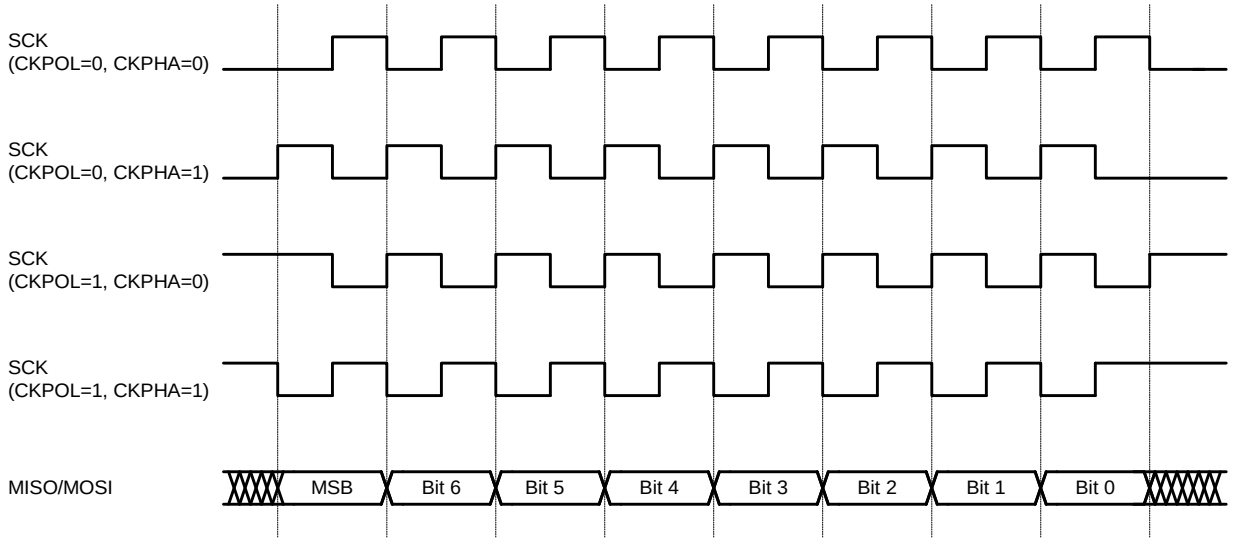



**Figure 21.5. Data/Clock Timing Relationship**

## 21.6. SPI Special Function Registers

SPI0 is accessed and controlled through four special function registers in the system controller: SPI0CN Control Register, SPI0DAT Data Register, SPI0CFG Configuration Register, and SPI0CKR Clock Rate Register. The four special function registers related to the operation of the SPI0 Bus are described in the following figures.

## SFR Definition 21.1. SPI0CFG: SPI0 Configuration

| R | R/W | R/W | R/W | R | R | R | R | Reset Value |
|---|---|---|---|---|---|---|---|---|
| SPIBSY | MSTEN | CKPHA | CKPOL | SLVSEL | NSSIN | SRMT | RXBMT | 00000111 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xA1

Bit 7: SPIBSY: SPI Busy (read only).
This bit is set to logic 1 when a SPI transfer is in progress (Master or Slave Mode).

Bit 6: MSTEN: Master Mode Enable.
0: Disable master mode. Operate in slave mode.
1: Enable master mode. Operate as a master.

Bit 5: CKPHA: SPI0 Clock Phase.
This bit controls the SPI0 clock phase.
0: Data centered on first edge of SCK period.*
1: Data centered on second edge of SCK period.*

Bit 4: CKPOL: SPI0 Clock Polarity.
This bit controls the SPI0 clock polarity.
0: SCK line low in idle state.
1: SCK line high in idle state.

Bit 3: SLVSEL: Slave Selected Flag (read only).
This bit is set to logic 1 whenever the NSS pin is low indicating SPI0 is the selected slave. It is cleared to logic 0 when NSS is high (slave not selected). This bit does not indicate the instantaneous value at the NSS pin, but rather a de-glitched version of the pin input.

Bit 2: NSSIN: NSS Instantaneous Pin Input (read only).
This bit mimics the instantaneous value that is present on the NSS port pin at the time that the register is read. This input is not de-glitched.

Bit 1: SRMT: Shift Register Empty (Valid in Slave Mode, read only).
This bit will be set to logic 1 when all data has been transferred in/out of the shift register, and there is no new information available to read from the transmit buffer or write to the receive buffer. It returns to logic 0 when a data byte is transferred to the shift register from the transmit buffer or by a transition on SCK.
NOTE: SRMT = 1 when in Master Mode.

Bit 0: RXBMT: Receive Buffer Empty (Valid in Slave Mode, read only).
This bit will be set to logic 1 when the receive buffer has been read and contains no new information. If there is new information available in the receive buffer that has not been read, this bit will return to logic 0.
NOTE: RXBMT = 1 when in Master Mode.

***Note:** See Table 21.1 for timing parameters.

## SFR Definition 21.2. SPI0CN: SPI0 Control

| R/W | R/W | R/W | R/W | R/W | R/W | R | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| SPIF | WCOL | MODF | RXOVRN | NSSMD1 | NSSMD0 | TXBMT | SPIEN | 00000110 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | Bit Addressable |

SFR Address: 0xF8

Bit 7: SPIF: SPI0 Interrupt Flag.
This bit is set to logic 1 by hardware at the end of a data transfer. If interrupts are enabled, setting this bit causes the CPU to vector to the SPI0 interrupt service routine. This bit is not automatically cleared by hardware. It must be cleared by software.

Bit 6: WCOL: Write Collision Flag.
This bit is set to logic 1 by hardware (and generates a SPI0 interrupt) to indicate a write to the SPI0 data register was attempted while a data transfer was in progress. This bit is not automatically cleared by hardware. It must be cleared by software.

Bit 5: MODF: Mode Fault Flag.
This bit is set to logic 1 by hardware (and generates a SPI0 interrupt) when a master mode collision is detected (NSS is low, MSTEN = 1, and NSSMD[1:0] = 01). This bit is not automatically cleared by hardware. It must be cleared by software.

Bit 4: RXOVRN: Receive Overrun Flag (Slave Mode only).
This bit is set to logic 1 by hardware (and generates a SPI0 interrupt) when the receive buffer still holds unread data from a previous transfer and the last bit of the current transfer is shifted into the SPI0 shift register. This bit is not automatically cleared by hardware. It must be cleared by software.

Bits 3–2: NSSMD1–NSSMD0: Slave Select Mode.
Selects between the following NSS operation modes:
(See Section “21.2. SPI0 Master Mode Operation’ on page 183 and Section “21.3. SPI0 Slave Mode Operation’ on page 185).
00: 3-Wire Slave or 3-wire Master Mode. NSS signal is not routed to a port pin.
01: 4-Wire Slave or Multi-Master Mode (Default). NSS is always an input to the device.
1x: 4-Wire Single-Master Mode. NSS signal is mapped as an output from the device and will assume the value of NSSMD0.

Bit 1: TXBMT: Transmit Buffer Empty.
This bit will be set to logic 0 when new data has been written to the transmit buffer. When data in the transmit buffer is transferred to the SPI shift register, this bit will be set to logic 1, indicating that it is safe to write a new byte to the transmit buffer.

Bit 0: SPIEN: SPI0 Enable.
This bit enables/disables the SPI.
0: SPI disabled.
1: SPI enabled.

## SFR Definition 21.3. SPI0CKR: SPI0 Clock Rate

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| SCR7 | SCR6 | SCR5 | SCR4 | SCR3 | SCR2 | SCR1 | SCR0 | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xA2

Bits 7–0: SCR7–SCR0: SPI0 Clock Rate.
These bits determine the frequency of the SCK output when the SPI0 module is configured for master mode operation. The SCK clock frequency is a divided version of the system clock, and is given in the following equation, where *SYSCLK* is the system clock frequency and *SPI0CKR* is the 8-bit value held in the SPI0CKR register.

$$f_{SCK} = \frac{SYSCLK}{2 \times (SPI0CKR + 1)}$$

for 0 <= SPI0CKR <= 255

Example: If SYSCLK = 2 MHz and SPI0CKR = 0x04,

$$f_{SCK} = \frac{2000000}{2 \times (4 + 1)}$$

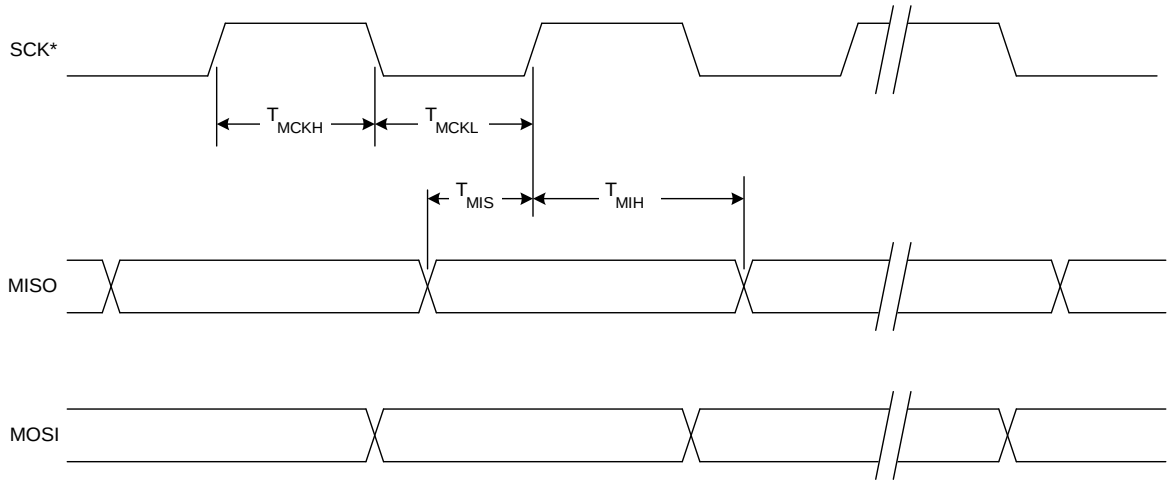
$$f_{SCK} = 200kHz$$

## SFR Definition 21.4. SPI0DAT: SPI0 Data

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

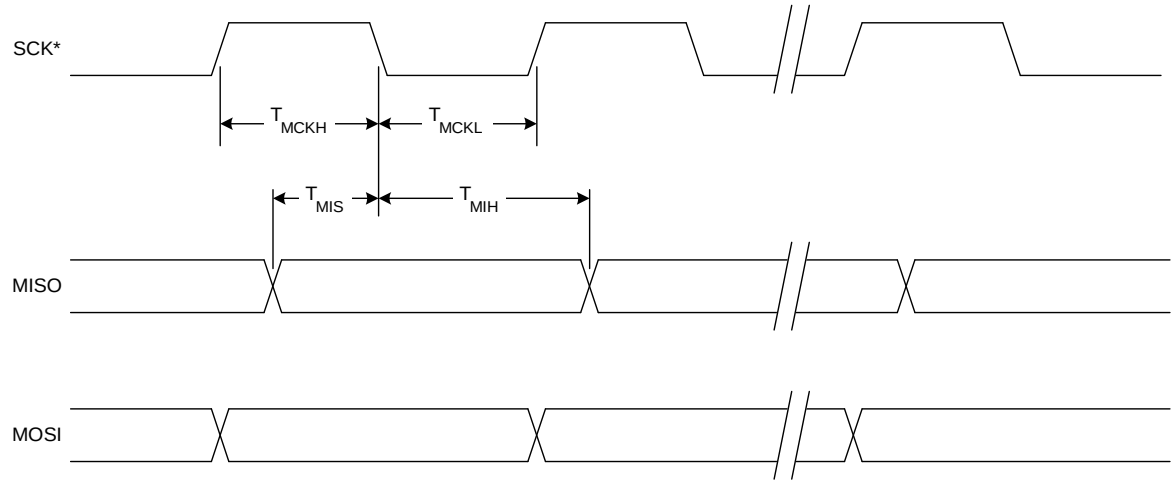
SFR Address: 0xA3

Bits 7–0: SPI0DAT: SPI0 Transmit and Receive Data.
The SPI0DAT register is used to transmit and receive SPI0 data. Writing data to SPI0DAT places the data into the transmit buffer and initiates a transfer when in Master Mode. A read of SPI0DAT returns the contents of the receive buffer.



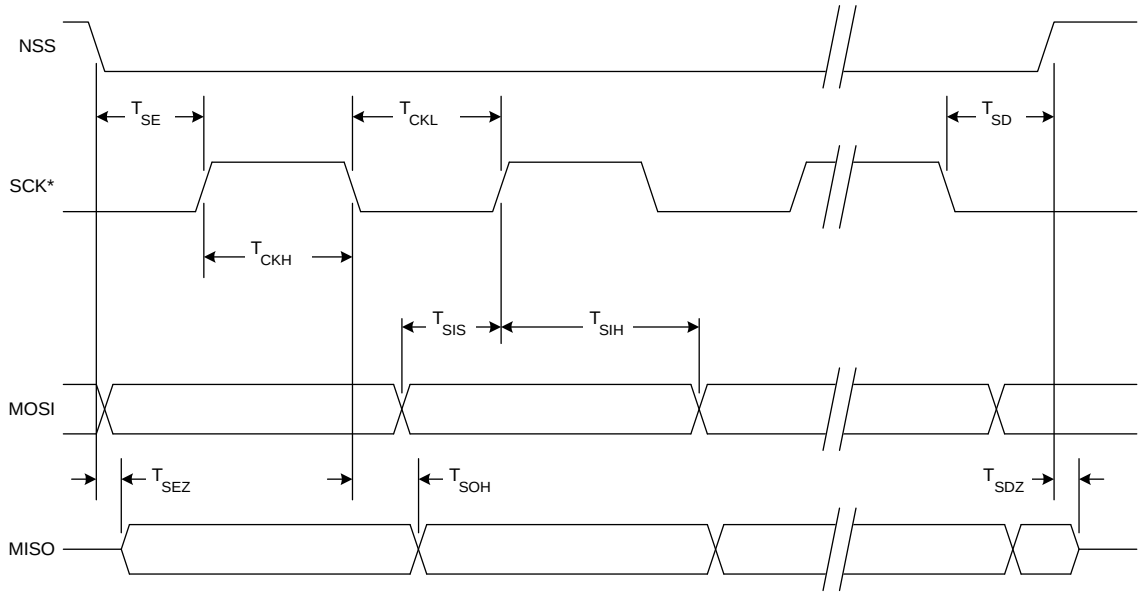

* SCK is shown for CKPOL = 0. SCK is the opposite polarity for CKPOL = 1.

**Figure 21.6. SPI Master Timing (CKPHA = 0)**



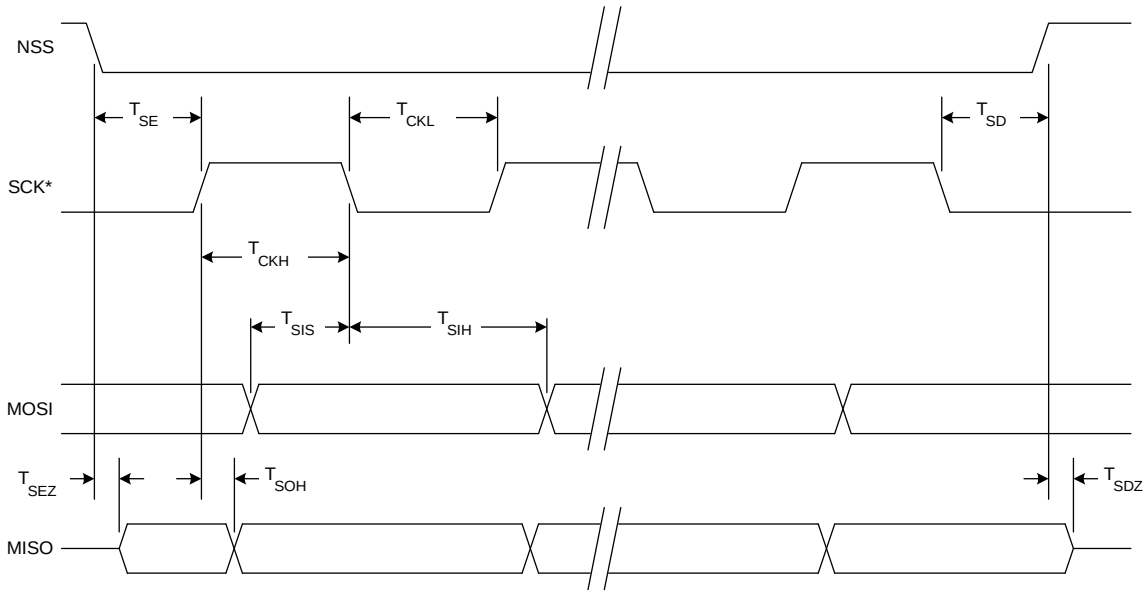

* SCK is shown for CKPOL = 0. SCK is the opposite polarity for CKPOL = 1.

**Figure 21.7. SPI Master Timing (CKPHA = 1)**




* SCK is shown for CKPOL = 0. SCK is the opposite polarity for CKPOL = 1.

**Figure 21.8. SPI Slave Timing (CKPHA = 0)**




* SCK is shown for CKPOL = 0. SCK is the opposite polarity for CKPOL = 1.

**Figure 21.9. SPI Slave Timing (CKPHA = 1)**

**Table 21.1. SPI Slave Timing Parameters**

| Parameter | Description | Min | Max | Units |
|---|---|---|---|---|
| **Master Mode Timing*** (See Figure 21.6 and Figure 21.7) | | | | |
| $T_{MCKH}$ | SCK High Time | 1 x $T_{SYSCLK}$ | — | ns |
| $T_{MCKL}$ | SCK Low Time | 1 x $T_{SYSCLK}$ | — | ns |
| $T_{MIS}$ | MISO Valid to SCK Sample Edge | 20 | — | ns |
| $T_{MIH}$ | SCK Sample Edge to MISO Change | 0 | — | ns |
| **Slave Mode Timing*** (See Figure 21.8 and Figure 21.9) | | | | |
| $T_{SE}$ | NSS Falling to First SCK Edge | 2 x $T_{SYSCLK}$ | — | ns |
| $T_{SD}$ | Last SCK Edge to NSS Rising | 2 x $T_{SYSCLK}$ | — | ns |
| $T_{SEZ}$ | NSS Falling to MISO Valid | — | 4 x $T_{SYSCLK}$ | ns |
| $T_{SDZ}$ | NSS Rising to MISO High-Z | — | 4 x $T_{SYSCLK}$ | ns |
| $T_{CKH}$ | SCK High Time | 5 x $T_{SYSCLK}$ | — | ns |
| $T_{CKL}$ | SCK Low Time | 5 x $T_{SYSCLK}$ | — | ns |
| $T_{SIS}$ | MOSI Valid to SCK Sample Edge | 2 x $T_{SYSCLK}$ | — | ns |
| $T_{SIH}$ | SCK Sample Edge to MOSI Change | 2 x $T_{SYSCLK}$ | — | ns |
| $T_{SOH}$ | SCK Shift Edge to MISO Change | — | 4 x $T_{SYSCLK}$ | ns |
| ***Note:** $T_{SYSCLK}$ is equal to one period of the device system clock (SYSCLK) in ns. | | | | |

## NOTES:

# 22. Timers

Each MCU includes four counter/timers: two are 16-bit counter/timers compatible with those found in the standard 8051, and two are 16-bit auto-reload timer for use with other device peripherals or for general purpose use. These timers can be used to measure time intervals, count external events and generate periodic interrupt requests. Timer 0 and Timer 1 are nearly identical and have four primary modes of operation. Timer 2 and Timer 3 offer 16-bit and split 8-bit timer functionality with auto-reload.

| Timer 0 and Timer 1 Modes: | Timer 2 Modes: | Timer 3 Modes: |
|---|---|---|
| 13-bit counter/timer | 16-bit timer with auto-reload | 16-bit timer with auto-reload |
| 16-bit counter/timer | | |
| 8-bit counter/timer with auto-reload | Two 8-bit timers with auto-reload | Two 8-bit timers with auto-reload |
| Two 8-bit counter/timers (Timer 0 only) | | |

Timers 0 and 1 may be clocked by one of five sources, determined by the Timer Mode Select bits (T1M–T0M) and the Clock Scale bits (SCA1–SCA0). The Clock Scale bits define a pre-scaled clock from which Timer 0 and/or Timer 1 may be clocked (See SFR Definition 22.3 for pre-scaled clock selection).

Timer 0/1 may then be configured to use this pre-scaled clock signal or the system clock. Timer 2 and Timer 3 may be clocked by the system clock, the system clock divided by 12, or the external oscillator clock source divided by 8.

Timer 0 and Timer 1 may also be operated as counters. When functioning as a counter, a counter/timer register is incremented on each high-to-low transition at the selected input pin (T0 or T1). Events with a frequency of up to one-fourth the system clock's frequency can be counted. The input signal need not be periodic, but it must be held at a given level for at least two full system clock cycles to ensure the level is properly sampled.

## 22.1. Timer 0 and Timer 1

Each timer is implemented as a 16-bit register accessed as two separate bytes: a low byte (TL0 or TL1) and a high byte (TH0 or TH1). The Counter/Timer Control register (TCON) is used to enable Timer 0 and Timer 1 as well as indicate status. Timer 0 interrupts can be enabled by setting the ET0 bit in the IE register (Section “12.4. Interrupt Register Descriptions’ on page 107); Timer 1 interrupts can be enabled by setting the ET1 bit in the IE register (Section 12.4). Both counter/timers operate in one of four primary modes selected by setting the Mode Select bits T1M1–T0M0 in the Counter/Timer Mode register (TMOD). Each timer can be configured independently. Each operating mode is described below.

### 22.1.1. Mode 0: 13-bit Counter/Timer

Timer 0 and Timer 1 operate as 13-bit counter/timers in Mode 0. The following describes the configuration and operation of Timer 0. However, both timers operate identically, and Timer 1 is configured in the same manner as described for Timer 0.

The TH0 register holds the eight MSBs of the 13-bit counter/timer. TL0 holds the five LSBs in bit positions TL0.4–TL0.0. The three upper bits of TL0 (TL0.7–TL0.5) are indeterminate and should be masked out or ignored when reading. As the 13-bit timer register increments and overflows from 0x1FFF (all ones) to 0x0000, the timer overflow flag TF0 (TCON.5) is set and an interrupt will occur if Timer 0 interrupts are enabled.

The C/T0 bit (TMOD.2) selects the counter/timer's clock source. When C/T0 is set to logic 1, high-to-low transitions at the selected Timer 0 input pin (T0) increment the timer register (Refer to Section “18.1. Priority Crossbar Decoder’ on page 139 for information on selecting and configuring external I/O pins). Clearing C/T selects the clock defined by the T0M bit (CKCON.3). When T0M is set, Timer 0 is clocked by the system clock. When T0M is cleared, Timer 0 is clocked by the source selected by the Clock Scale bits in CKCON (see SFR Definition 22.3).

Setting the TR0 bit (TCON.4) enables the timer when either GATE0 (TMOD.3) is logic 0 or the input signal /INT0 is active as defined by bit IN0PL in register IT01CF (see SFR Definition 12.5). Setting GATE0 to ‘1’ allows the timer to be controlled by the external input signal /INT0 (see Section “12.4. Interrupt Register Descriptions’ on page 107), facilitating pulse width measurements.

| TR0 | GATE0 | /INT0 | Counter/Timer |
|---|---|---|---|
| 0 | X | X | Disabled |
| 1 | 0 | X | Enabled |
| 1 | 1 | 0 | Disabled |
| 1 | 1 | 1 | Enabled |
| X = Don't Care | | | |

Setting TR0 does not force the timer to reset. The timer registers should be loaded with the desired initial value before the timer is enabled.

TL1 and TH1 form the 13-bit register for Timer 1 in the same manner as described above for TL0 and TH0. Timer 1 is configured and controlled using the relevant TCON and TMOD bits just as with Timer 0. The input signal /INT1 is used with Timer 1; the /INT1 polarity is defined by bit IN1PL in register IT01CF (see SFR Definition 12.5).

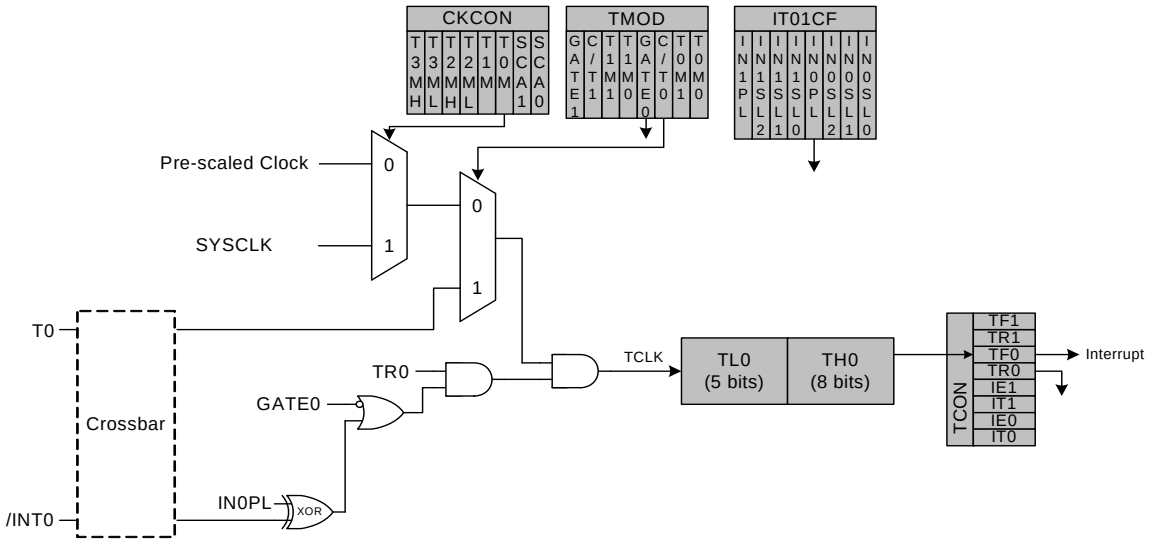



**Figure 22.1. T0 Mode 0 Block Diagram**

### 22.1.2. Mode 1: 16-bit Counter/Timer

Mode 1 operation is the same as Mode 0, except that the counter/timer registers use all 16 bits. The counter/timers are enabled and configured in Mode 1 in the same manner as for Mode 0.

### 22.1.3. Mode 2: 8-bit Counter/Timer with Auto-Reload

Mode 2 configures Timer 0 and Timer 1 to operate as 8-bit counter/timers with automatic reload of the start value. TL0 holds the count and TH0 holds the reload value. When the counter in TL0 overflows from all ones to 0x00, the timer overflow flag TF0 (TCON.5) is set and the counter in TL0 is reloaded from TH0. If Timer 0 interrupts are enabled, an interrupt will occur when the TF0 flag is set. The reload value in TH0 is not changed. TL0 must be initialized to the desired value before enabling the timer for the first count to be correct. When in Mode 2, Timer 1 operates identically to Timer 0.

Both counter/timers are enabled and configured in Mode 2 in the same manner as Mode 0. Setting the TR0 bit (TCON.4) enables the timer when either GATE0 (TMOD.3) is logic 0 or when the input signal /INT0 is active as defined by bit IN0PL in register IT01CF (see Section “12.5. External Interrupts’ on page 111 for details on the external input signals /INT0 and /INT1).

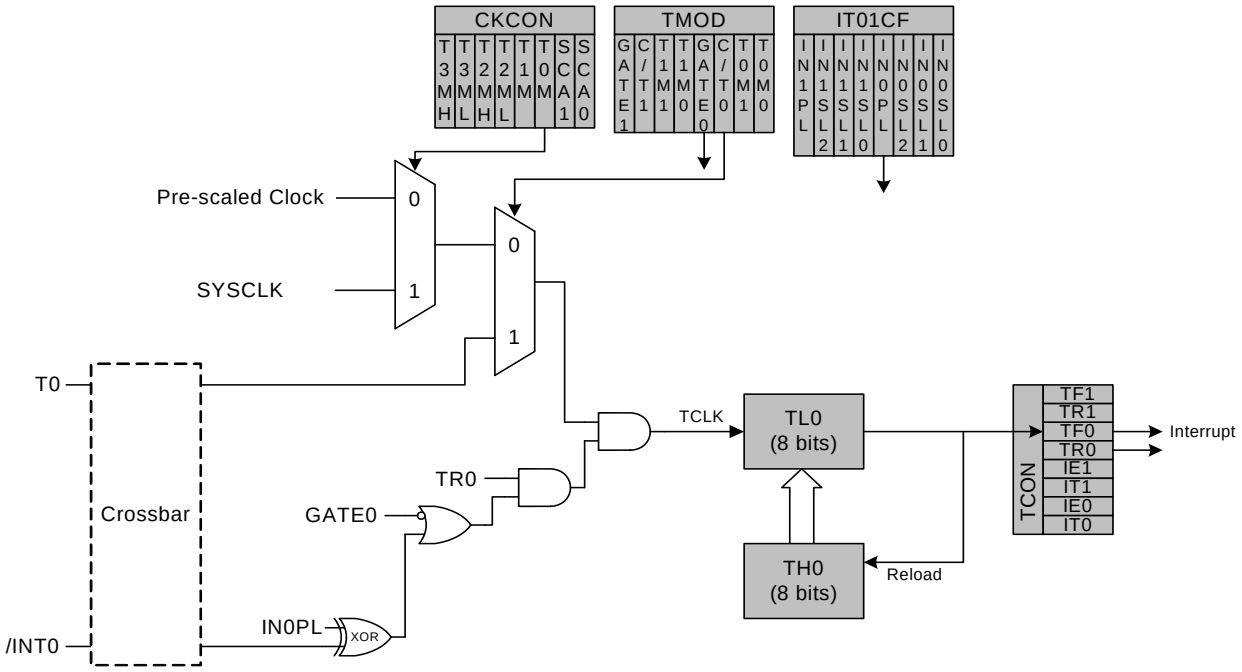



**Figure 22.2. T0 Mode 2 Block Diagram**

### 22.1.4. Mode 3: Two 8-bit Counter/Timers (Timer 0 Only)

In Mode 3, Timer 0 is configured as two separate 8-bit counter/timers held in TL0 and TH0. The counter/timer in TL0 is controlled using the Timer 0 control/status bits in TCON and TMOD: TR0, C/T0, GATE0 and TF0. TL0 can use either the system clock or an external input signal as its timebase. The TH0 register is restricted to a timer function sourced by the system clock or prescaled clock. TH0 is enabled using the Timer 1 run control bit TR1. TH0 sets the Timer 1 overflow flag TF1 on overflow and thus controls the Timer 1 interrupt.

Timer 1 is inactive in Mode 3. When Timer 0 is operating in Mode 3, Timer 1 can be operated in Modes 0, 1 or 2, but cannot be clocked by external signals nor set the TF1 flag and generate an interrupt. However, the Timer 1 overflow can be used to generate baud rates for the SMBus and UART. While Timer 0 is operating in Mode 3, Timer 1 run control is handled through its mode settings. To run Timer 1 while Timer 0 is in Mode 3, set the Timer 1 Mode as 0, 1, or 2. To disable Timer 1, configure it for Mode 3.

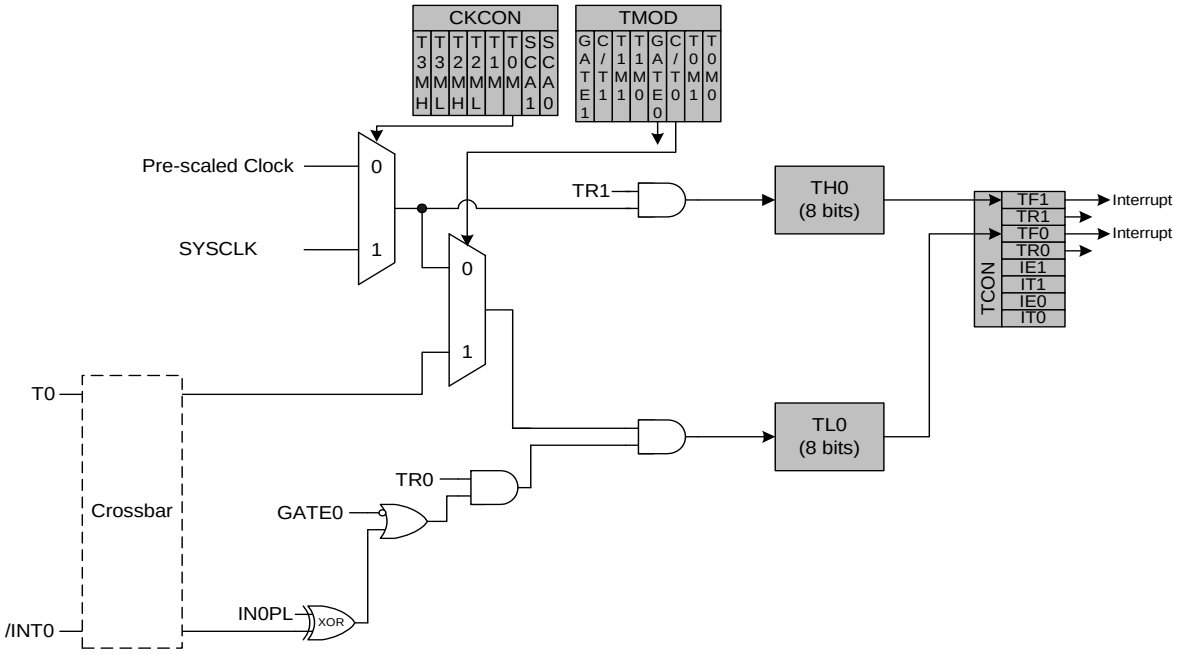



**Figure 22.3. T0 Mode 3 Block Diagram**

## SFR Definition 22.1. TCON: Timer Contro

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| TF1 | TR1 | TF0 | TR0 | IE1 | IT1 | IE0 | IT0 | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | Bit Addressable |

SFR Address: 0x88

Bit7: TF1: Timer 1 Overflow Flag.
Set by hardware when Timer 1 overflows. This flag can be cleared by software but is automatically cleared when the CPU vectors to the Timer 1 interrupt service routine.
0: No Timer 1 overflow detected.
1: Timer 1 has overflowed.

Bit6: TR1: Timer 1 Run Control.
0: Timer 1 disabled.
1: Timer 1 enabled.

Bit5: TF0: Timer 0 Overflow Flag.
Set by hardware when Timer 0 overflows. This flag can be cleared by software but is automatically cleared when the CPU vectors to the Timer 0 interrupt service routine.
0: No Timer 0 overflow detected.
1: Timer 0 has overflowed.

Bit4: TR0: Timer 0 Run Control.
0: Timer 0 disabled.
1: Timer 0 enabled.

Bit3: IE1: External Interrupt 1.
This flag is set by hardware when an edge/level of type defined by IT1 is detected. It can be cleared by software but is automatically cleared when the CPU vectors to the External Interrupt 1 service routine if IT1 = 1. When IT1 = 0, this flag is set to ‘1’ when /INT1 is active as defined by bit IN1PL in register IT01CF (see SFR Definition 12.5).

Bit2: IT1: Interrupt 1 Type Select.
This bit selects whether the configured /INT1 interrupt will be edge or level sensitive. /INT1 is configured active low or high by the IN1PL bit in the IT01CF register (see SFR Definition 12.5).
0: /INT1 is level triggered.
1: /INT1 is edge triggered.

Bit1: IE0: External Interrupt 0.
This flag is set by hardware when an edge/level of type defined by IT0 is detected. It can be cleared by software but is automatically cleared when the CPU vectors to the External Interrupt 0 service routine if IT0 = 1. When IT0 = 0, this flag is set to ‘1’ when /INT0 is active as defined by bit IN0PL in register IT01CF (see SFR Definition 12.5).

Bit0: IT0: Interrupt 0 Type Select.
This bit selects whether the configured /INT0 interrupt will be edge or level sensitive. /INT0 is configured active low or high by the IN0PL bit in register IT01CF (see SFR Definition 12.5).
0: /INT0 is level triggered.
1: /INT0 is edge triggered.

## SFR Definition 22.2. TMOD: Timer Mode

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| GATE1 | C/T1 | T1M1 | T1M0 | GATE0 | C/T0 | T0M1 | T0M0 | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0x89

Bit7: GATE1: Timer 1 Gate Control.
0: Timer 1 enabled when TR1 = 1 irrespective of /INT1 logic level.
1: Timer 1 enabled only when TR1 = 1 AND /INT1 is active as defined by bit IN1PL in register IT01CF (see SFR Definition 12.5).

Bit6: C/T1: Counter/Timer 1 Select.
0: Timer Function: Timer 1 incremented by clock defined by T1M bit (CKCON.4).
1: Counter Function: Timer 1 incremented by high-to-low transitions on external input pin (T1).

Bits5–4: T1M1–T1M0: Timer 1 Mode Select.
These bits select the Timer 1 operation mode.

| T1M1 | T1M0 | Mode |
|---|---|---|
| 0 | 0 | Mode 0: 13-bit counter/timer |
| 0 | 1 | Mode 1: 16-bit counter/timer |
| 1 | 0 | Mode 2: 8-bit counter/timer with auto-reload |
| 1 | 1 | Mode 3: Timer 1 inactive |

Bit3: GATE0: Timer 0 Gate Control.
0: Timer 0 enabled when TR0 = 1 irrespective of /INT0 logic level.
1: Timer 0 enabled only when TR0 = 1 AND /INT0 is active as defined by bit IN0PL in register IT01CF (see SFR Definition 12.5).

Bit2: C/T0: Counter/Timer Select.
0: Timer Function: Timer 0 incremented by clock defined by T0M bit (CKCON.3).
1: Counter Function: Timer 0 incremented by high-to-low transitions on external input pin (T0).

Bits1–0: T0M1–T0M0: Timer 0 Mode Select.
These bits select the Timer 0 operation mode.

| T0M1 | T0M0 | Mode |
|---|---|---|
| 0 | 0 | Mode 0: 13-bit counter/timer |
| 0 | 1 | Mode 1: 16-bit counter/timer |
| 1 | 0 | Mode 2: 8-bit counter/timer with auto-reload |
| 1 | 1 | Mode 3: Two 8-bit counter/timers |

## SFR Definition 22.3. CKCON: Clock Control

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| T3MH | T3ML | T2MH | T2ML | T1M | T0M | SCA1 | SCA0 | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0x8E

Bit7: T3MH: Timer 3 High Byte Clock Select.
This bit selects the clock supplied to the Timer 3 high byte if Timer 3 is configured in split 8-bit timer mode. T3MH is ignored if Timer 3 is in any other mode.
0: Timer 3 high byte uses the clock defined by the T3XCLK bit in TMR3CN.
1: Timer 3 high byte uses the system clock.

Bit6: T3ML: Timer 3 Low Byte Clock Select.
This bit selects the clock supplied to Timer 3. If Timer 3 is configured in split 8-bit timer mode, this bit selects the clock supplied to the lower 8-bit timer.
0: Timer 3 low byte uses the clock defined by the T3XCLK bit in TMR3CN.
1: Timer 3 low byte uses the system clock.

Bit5: T2MH: Timer 2 High Byte Clock Select.
This bit selects the clock supplied to the Timer 2 high byte if Timer 2 is configured in split 8-bit timer mode. T2MH is ignored if Timer 2 is in any other mode.
0: Timer 2 high byte uses the clock defined by the T2XCLK bit in TMR2CN.
1: Timer 2 high byte uses the system clock.

Bit4: T2ML: Timer 2 Low Byte Clock Select.
This bit selects the clock supplied to Timer 2. If Timer 2 is configured in split 8-bit timer mode, this bit selects the clock supplied to the lower 8-bit timer.
0: Timer 2 low byte uses the clock defined by the T2XCLK bit in TMR2CN.
1: Timer 2 low byte uses the system clock.

Bit3: T1M: Timer 1 Clock Select.
This select the clock source supplied to Timer 1. T1M is ignored when C/T1 is set to logic 1.
0: Timer 1 uses the clock defined by the prescale bits, SCA1–SCA0.
1: Timer 1 uses the system clock.

Bit2: T0M: Timer 0 Clock Select.
This bit selects the clock source supplied to Timer 0. T0M is ignored when C/T0 is set to logic 1.
0: Counter/Timer 0 uses the clock defined by the prescale bits, SCA1–SCA0.
1: Counter/Timer 0 uses the system clock.

Bits1–0: SCA1–SCA0: Timer 0/1 Prescale Bits.
These bits control the division of the clock supplied to Timer 0 and Timer 1 if configured to use prescaled clock inputs.

| SCA1 | SCA0 | Prescaled Clock |
|---|---|---|
| 0 | 0 | System clock divided by 12 |
| 0 | 1 | System clock divided by 4 |
| 1 | 0 | System clock divided by 48 |
| 1 | 1 | External clock divided by 8 |

**Note:** External clock divided by 8 is synchronized with the system clock.

### SFR Definition 22.4. TL0: Timer 0 Low Byte

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0x8A

Bits 7–0: TL0: Timer 0 Low Byte.
The TL0 register is the low byte of the 16-bit Timer 0.

### SFR Definition 22.5. TL1: Timer 1 Low Byte

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0x8B

Bits 7–0: TL1: Timer 1 Low Byte.
The TL1 register is the low byte of the 16-bit Timer 1.

### SFR Definition 22.6. TH0: Timer 0 High Byte

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0x8C

Bits 7–0: TH0: Timer 0 High Byte.
The TH0 register is the high byte of the 16-bit Timer 0.

### SFR Definition 22.7. TH1: Timer 1 High Byte

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0x8D

Bits 7–0: TH1: Timer 1 High Byte.
The TH1 register is the high byte of the 16-bit Timer 1.

## 22.2. Timer 2

Timer 2 is a 16-bit timer formed by two 8-bit SFRs: TMR2L (low byte) and TMR2H (high byte). Timer 2 may operate in 16-bit auto-reload mode or (split) 8-bit auto-reload mode. The T2SPLIT bit (TMR2CN.3) defines the Timer 2 operation mode.

Timer 2 may be clocked by the system clock, the system clock divided by 12, or the external oscillator source divided by 8. The external clock mode is ideal for real-time clock (RTC) functionality, where the internal oscillator drives the system clock while Timer 2 (and/or the PCA) is clocked by an external precision oscillator. Note that the external oscillator source divided by 8 is synchronized with the system clock.

### 22.2.1. 16-bit Timer with Auto-Reload

When T2SPLIT (TMR2CN.3) is zero, Timer 2 operates as a 16-bit timer with auto-reload. Timer 2 can be clocked by SYSCLK, SYSCLK divided by 12, or the external oscillator clock source divided by 8. As the 16-bit timer register increments and overflows from 0xFFFF to 0x0000, the 16-bit value in the Timer 2 reload registers (TMR2RLH and TMR2RLL) is loaded into the Timer 2 register as shown in Figure 22.4, and the Timer 2 High Byte Overflow Flag (TMR2CN.7) is set. If Timer 2 interrupts are enabled (if IE.5 is set), an interrupt will be generated on each Timer 2 overflow. Additionally, if Timer 2 interrupts are enabled and the TF2LEN bit is set (TMR2CN.5), an interrupt will be generated each time the lower 8 bits (TMR2L) overflow from 0xFF to 0x00.

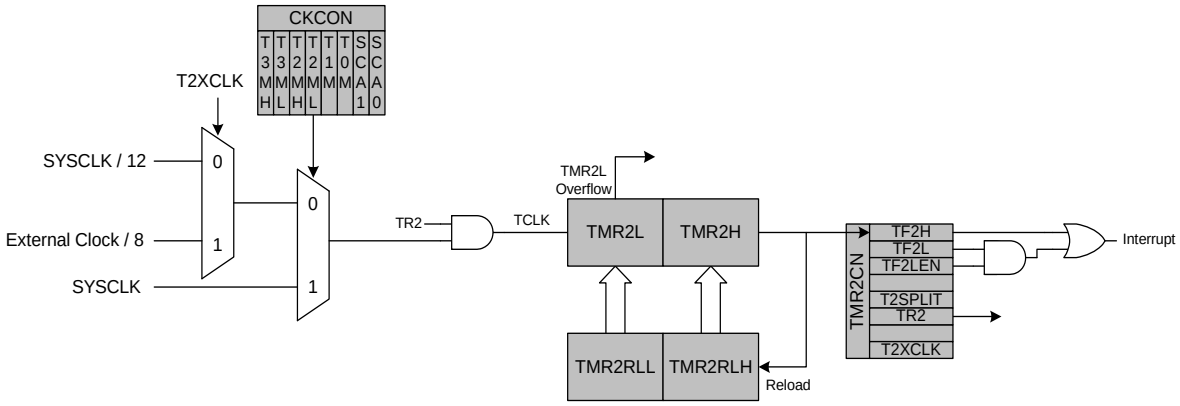



**Figure 22.4. Timer 2 16-Bit Mode Block Diagram**

### 22.2.2. 8-bit Timers with Auto-Reload

When T2SPLIT is set, Timer 2 operates as two 8-bit timers (TMR2H and TMR2L). Both 8-bit timers operate in auto-reload mode as shown in Figure 22.5. TMR2RLL holds the reload value for TMR2L; TMR2RLH holds the reload value for TMR2H. The TR2 bit in TMR2CN handles the run control for TMR2H. TMR2L is always running when configured for 8-bit Mode.

Each 8-bit timer may be configured to use SYSCLK, SYSCLK divided by 12, or the external oscillator clock source divided by 8. The Timer 2 Clock Select bits (T2MH and T2ML in CKCON) select either SYSCLK or the clock defined by the Timer 2 External Clock Select bit (T2XCLK in TMR2CN), as follows:

| T2MH | T2XCLK | TMR2H Clock Source |
|---|---|---|
| 0 | 0 | SYSCLK / 12 |
| 0 | 1 | External Clock / 8 |
| 1 | X | SYSCLK |

| T2ML | T2XCLK | TMR2L Clock Source |
|---|---|---|
| 0 | 0 | SYSCLK / 12 |
| 0 | 1 | External Clock / 8 |
| 1 | X | SYSCLK |

The TF2H bit is set when TMR2H overflows from 0xFF to 0x00; the TF2L bit is set when TMR2L overflows from 0xFF to 0x00. When Timer 2 interrupts are enabled (IE.5), an interrupt is generated each time TMR2H overflows. If Timer 2 interrupts are enabled and TF2LEN (TMR2CN.5) is set, an interrupt is generated each time either TMR2L or TMR2H overflows. When TF2LEN is enabled, software must check the TF2H and TF2L flags to determine the source of the Timer 2 interrupt. The TF2H and TF2L interrupt flags are not cleared by hardware and must be manually cleared by software.

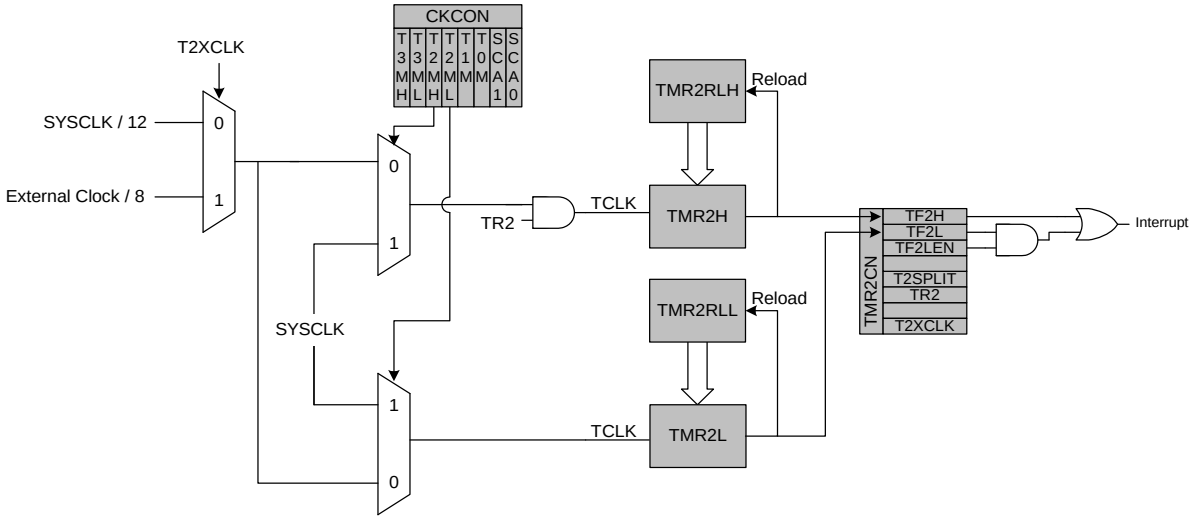



**Figure 22.5. Timer 2 8-Bit Mode Block Diagram**

## SFR Definition 22.8. TMR2CN: Timer 2 Control

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| TF2H | TF2L | TF2LEN | — | T2SPLIT | TR2 | — | T2XCLK | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | Bit Addressable |

SFR Address: 0xC8

Bit7: TF2H: Timer 2 High Byte Overflow Flag.
Set by hardware when the Timer 2 high byte overflows from 0xFF to 0x00. In 16 bit mode, this will occur when Timer 2 overflows from 0xFFFF to 0x0000. When the Timer 2 interrupt is enabled, setting this bit causes the CPU to vector to the Timer 2 interrupt service routine. TF2H is not automatically cleared by hardware and must be cleared by software.

Bit6: TF2L: Timer 2 Low Byte Overflow Flag.
Set by hardware when the Timer 2 low byte overflows from 0xFF to 0x00. When this bit is set, an interrupt will be generated if TF2LEN is set and Timer 2 interrupts are enabled. TF2L will set when the low byte overflows regardless of the Timer 2 mode. This bit is not automatically cleared by hardware.

Bit5: TF2LEN: Timer 2 Low Byte Interrupt Enable.
This bit enables/disables Timer 2 Low Byte interrupts. If TF2LEN is set and Timer 2 interrupts are enabled, an interrupt will be generated when the low byte of Timer 2 overflows. This bit should be cleared when operating Timer 2 in 16-bit mode.
0: Timer 2 Low Byte interrupts disabled.
1: Timer 2 Low Byte interrupts enabled.

Bit4: UNUSED. Read = 0b. Write = don't care.

Bit3: T2SPLIT: Timer 2 Split Mode Enable.
When this bit is set, Timer 2 operates as two 8-bit timers with auto-reload.
0: Timer 2 operates in 16-bit auto-reload mode.
1: Timer 2 operates as two 8-bit auto-reload timers.

Bit2: TR2: Timer 2 Run Control.
This bit enables/disables Timer 2. In 8-bit mode, this bit enables/disables TMR2H only; TMR2L is always enabled in this mode.
0: Timer 2 disabled.
1: Timer 2 enabled.

Bit1: UNUSED. Read = 0b. Write = don't care.

Bit0: T2XCLK: Timer 2 External Clock Select.
This bit selects the external clock source for Timer 2. If Timer 2 is in 8-bit mode, this bit selects the external oscillator clock source for both timer bytes. However, the Timer 2 Clock Select bits (T2MH and T2ML in register CKCON) may still be used to select between the external clock and the system clock for either timer.
0: Timer 2 external clock selection is the system clock divided by 12.
1: Timer 2 external clock selection is the external clock divided by 8. Note that the external oscillator source divided by 8 is synchronized with the system clock.

### SFR Definition 22.9. TMR2RLL: Timer 2 Reload Register Low Byte

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xCA

Bits 7–0: TMR2RLL: Timer 2 Reload Register Low Byte.
TMR2RLL holds the low byte of the reload value for Timer 2.

### SFR Definition 22.10. TMR2RLH: Timer 2 Reload Register High Byte

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xCB

Bits 7–0: TMR2RLH: Timer 2 Reload Register High Byte.
The TMR2RLH holds the high byte of the reload value for Timer 2.

### SFR Definition 22.11. TMR2L: Timer 2 Low Byte

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xCC

Bits 7–0: TMR2L: Timer 2 Low Byte.
In 16-bit mode, the TMR2L register contains the low byte of the 16-bit Timer 2. In 8-bit mode, TMR2L contains the 8-bit low byte timer value.

### SFR Definition 22.12. TMR2H Timer 2 High Byte

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xCD

Bits 7–0: TMR2H: Timer 2 High Byte.
In 16-bit mode, the TMR2H register contains the high byte of the 16-bit Timer 2. In 8-bit mode, TMR2H contains the 8-bit high byte timer value.

## 22.3. Timer 3

Timer 3 is a 16-bit timer formed by two 8-bit SFRs: TMR3L (low byte) and TMR3H (high byte). Timer 3 may operate in 16-bit auto-reload mode or (split) 8-bit auto-reload mode. The T3SPLIT bit (TMR3CN.3) defines the Timer 3 operation mode.

Timer 3 may be clocked by the system clock, the system clock divided by 12, or the external oscillator source divided by 8. The external clock mode is ideal for real-time clock (RTC) functionality, where the internal oscillator drives the system clock while Timer 3 (and/or the PCA) is clocked by an external precision oscillator. Note that the external oscillator source divided by 8 is synchronized with the system clock.

### 22.3.1. 16-bit Timer with Auto-Reload

When T3SPLIT (TMR3CN.3) is zero, Timer 3 operates as a 16-bit timer with auto-reload. Timer 3 can be clocked by SYSCLK, SYSCLK divided by 12, or the external oscillator clock source divided by 8. As the 16-bit timer register increments and overflows from 0xFFFF to 0x0000, the 16-bit value in the Timer 3 reload registers (TMR3RLH and TM3RLL) is loaded into the Timer 3 register as shown in Figure 22.6, and the Timer 3 High Byte Overflow Flag (TMR3CN.7) is set. If Timer 3 interrupts are enabled, an interrupt will be generated each Timer 3 overflow. Additionally, if Timer 3 interrupts are enabled and the TF3LEN bit is set (TMR3CN.5), an interrupt will be generated each time the lower 8 bits (TMR3L) overflow from 0xFF to 0x00.

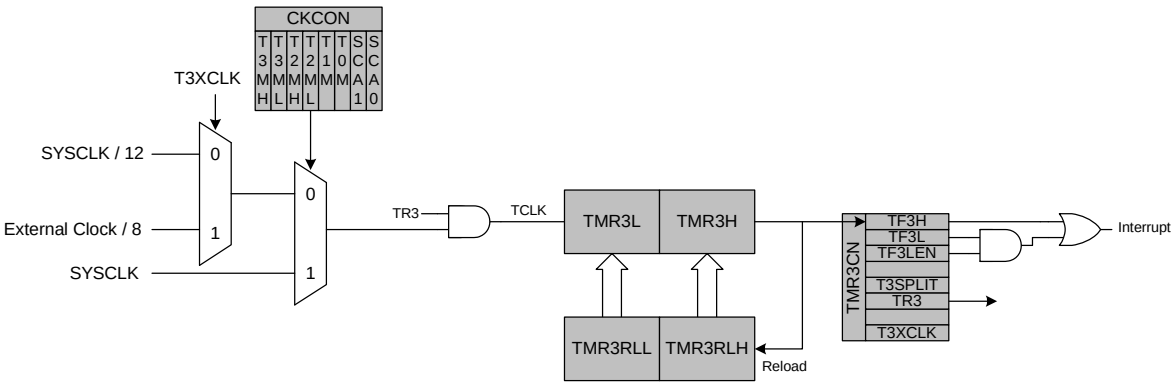



**Figure 22.6. Timer 3 16-Bit Mode Block Diagram**

### 22.3.2. 8-bit Timers with Auto-Reload

When T3SPLIT is set, Timer 3 operates as two 8-bit timers (TMR3H and TMR3L). Both 8-bit timers operate in auto-reload mode as shown in Figure 22.7. TMR3RLL holds the reload value for TMR3L; TMR3RLH holds the reload value for TMR3H. The TR3 bit in TMR3CN handles the run control for TMR3H. TMR3L is always running when configured for 8-bit Mode.

Each 8-bit timer may be configured to use SYSCLK, SYSCLK divided by 12, or the external oscillator clock source divided by 8. The Timer 3 Clock Select bits (T3MH and T3ML in CKCON) select either SYSCLK or the clock defined by the Timer 3 External Clock Select bit (T3XCLK in TMR3CN), as follows:

| T3MH | T3XCLK | TMR3H Clock Source |
|---|---|---|
| 0 | 0 | SYSCLK / 12 |
| 0 | 1 | External Clock / 8 |
| 1 | X | SYSCLK |

| T3ML | T3XCLK | TMR3L Clock Source |
|---|---|---|
| 0 | 0 | SYSCLK / 12 |
| 0 | 1 | External Clock / 8 |
| 1 | X | SYSCLK |

The TF3H bit is set when TMR3H overflows from 0xFF to 0x00; the TF3L bit is set when TMR3L overflows from 0xFF to 0x00. When Timer 3 interrupts are enabled (IE.5), an interrupt is generated each time TMR3H overflows. If Timer 3 interrupts are enabled and TF3LEN (TMR3CN.5) is set, an interrupt is generated each time either TMR3L or TMR3H overflows. When TF3LEN is enabled, software must check the TF3H and TF3L flags to determine the source of the Timer 3 interrupt. The TF3H and TF3L interrupt flags are not cleared by hardware and must be manually cleared by software.

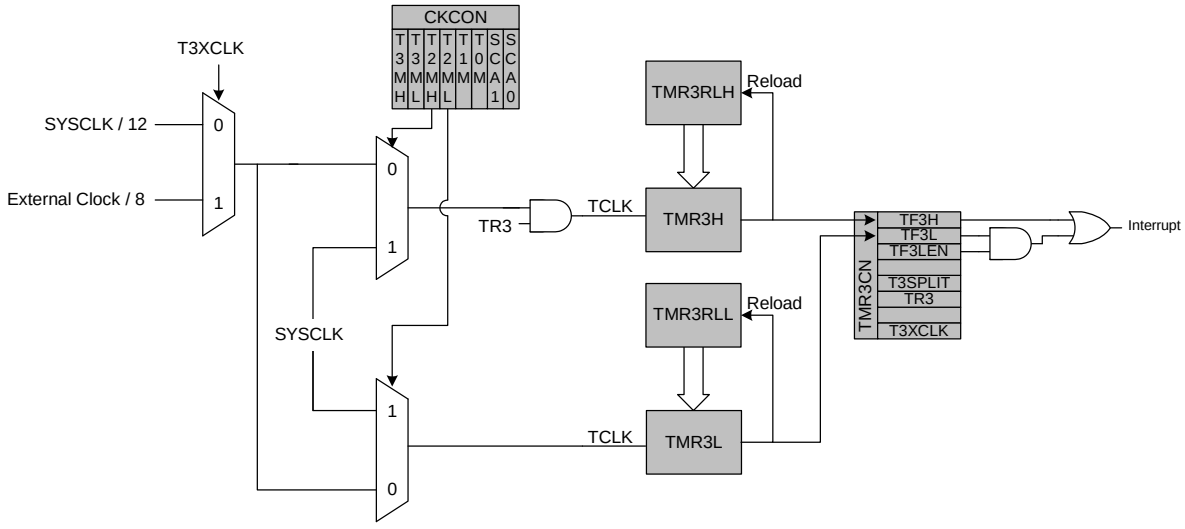



**Figure 22.7. Timer 3 8-Bit Mode Block Diagram**

## SFR Definition 22.13. TMR3CN: Timer 3 Control

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| TF3H | TF3L | TF3LEN | — | T3SPLIT | TR3 | — | T3XCLK | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0x91

Bit7: TF3H: Timer 3 High Byte Overflow Flag.
Set by hardware when the Timer 3 high byte overflows from 0xFF to 0x00. In 16 bit mode, this will occur when Timer 3 overflows from 0xFFFF to 0x0000. When the Timer 3 interrupt is enabled, setting this bit causes the CPU to vector to the Timer 3 interrupt service routine. TF3H is not automatically cleared by hardware and must be cleared by software.

Bit6: TF3L: Timer 3 Low Byte Overflow Flag.
Set by hardware when the Timer 3 low byte overflows from 0xFF to 0x00. When this bit is set, an interrupt will be generated if TF3LEN is set and Timer 3 interrupts are enabled. TF3L will set when the low byte overflows regardless of the Timer 3 mode. This bit is not automatically cleared by hardware.

Bit5: TF3LEN: Timer 3 Low Byte Interrupt Enable.
This bit enables/disables Timer 3 Low Byte interrupts. If TF3LEN is set and Timer 3 interrupts are enabled, an interrupt will be generated when the low byte of Timer 3 overflows. This bit should be cleared when operating Timer 3 in 16-bit mode.
0: Timer 3 Low Byte interrupts disabled.
1: Timer 3 Low Byte interrupts enabled.

Bit4: UNUSED. Read = 0b. Write = don't care.

Bit3: T3SPLIT: Timer 3 Split Mode Enable.
When this bit is set, Timer 3 operates as two 8-bit timers with auto-reload.
0: Timer 3 operates in 16-bit auto-reload mode.
1: Timer 3 operates as two 8-bit auto-reload timers.

Bit2: TR3: Timer 3 Run Control.
This bit enables/disables Timer 3. In 8-bit mode, this bit enables/disables TMR3H only; TMR3L is always enabled in this mode.
0: Timer 3 disabled.
1: Timer 3 enabled.

Bit1: UNUSED. Read = 0b. Write = don't care.

Bit0: T3XCLK: Timer 3 External Clock Select.
This bit selects the external clock source for Timer 3. If Timer 3 is in 8-bit mode, this bit selects the external oscillator clock source for both timer bytes. However, the Timer 3 Clock Select bits (T3MH and T3ML in register CKCON) may still be used to select between the external clock and the system clock for either timer.
0: Timer 3 external clock selection is the system clock divided by 12.
1: Timer 3 external clock selection is the external clock divided by 8. Note that the external oscillator source divided by 8 is synchronized with the system clock.

### SFR Definition 22.14. TMR3RLL: Timer 3 Reload Register Low Byte

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0x92

Bits 7–0: TMR3RLL: Timer 3 Reload Register Low Byte.
TMR3RLL holds the low byte of the reload value for Timer 3.

### SFR Definition 22.15. TMR3RLH: Timer 3 Reload Register High Byte

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0x93

Bits 7–0: TMR3RLH: Timer 3 Reload Register High Byte.
The TMR3RLH holds the high byte of the reload value for Timer 3.

### SFR Definition 22.16. TMR3L: Timer 3 Low Byte

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0x94

Bits 7–0: TMR3L: Timer 3 Low Byte.
In 16-bit mode, the TMR3L register contains the low byte of the 16-bit Timer 3. In 8-bit mode, TMR3L contains the 8-bit low byte timer value.

### SFR Definition 22.17. TMR3H Timer 3 High Byte

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0x95

Bits 7–0: TMR3H: Timer 3 High Byte.
In 16-bit mode, the TMR3H register contains the high byte of the 16-bit Timer 3. In 8-bit mode, TMR3H contains the 8-bit high byte timer value.

# 23. Programmable Counter Array

The Programmable Counter Array (PCA0) provides enhanced timer functionality while requiring less CPU intervention than the standard 8051 counter/timers. The PCA consists of a dedicated 16-bit counter/timer and three 16-bit capture/compare modules. Each capture/compare module has its own associated I/O line (CEXn) which is routed through the Crossbar to Port I/O when enabled (See Section “18.1. Priority Crossbar Decoder’ on page 139 for details on configuring the Crossbar). The counter/timer is driven by a programmable timebase that can select between six sources: system clock, system clock divided by four, system clock divided by twelve, the external oscillator clock source divided by 8, Timer 0 overflow, or an external clock signal on the ECI input pin. Each capture/compare module may be configured to operate independently in one of six modes: Edge-Triggered Capture, Software Timer, High-Speed Output, Frequency Output, 8-Bit PWM, or 16-Bit PWM (each mode is described in Section “23.2. Capture/Compare Modules’ on page 213). The external oscillator clock option is ideal for real-time clock (RTC) functionality, allowing the PCA to be clocked by a precision external oscillator while the internal oscillator drives the system clock. The PCA is configured and controlled through the system controller's Special Function Registers. The PCA block diagram is shown in Figure 23.1

**Important Note:** The PCA Module 2 may be used as a watchdog timer (WDT), and is enabled in this mode following a system reset. **Access to certain PCA registers is restricted while WDT mode is enabled**. See Section “23.3. Watchdog Timer Mode’ on page 220 for details.

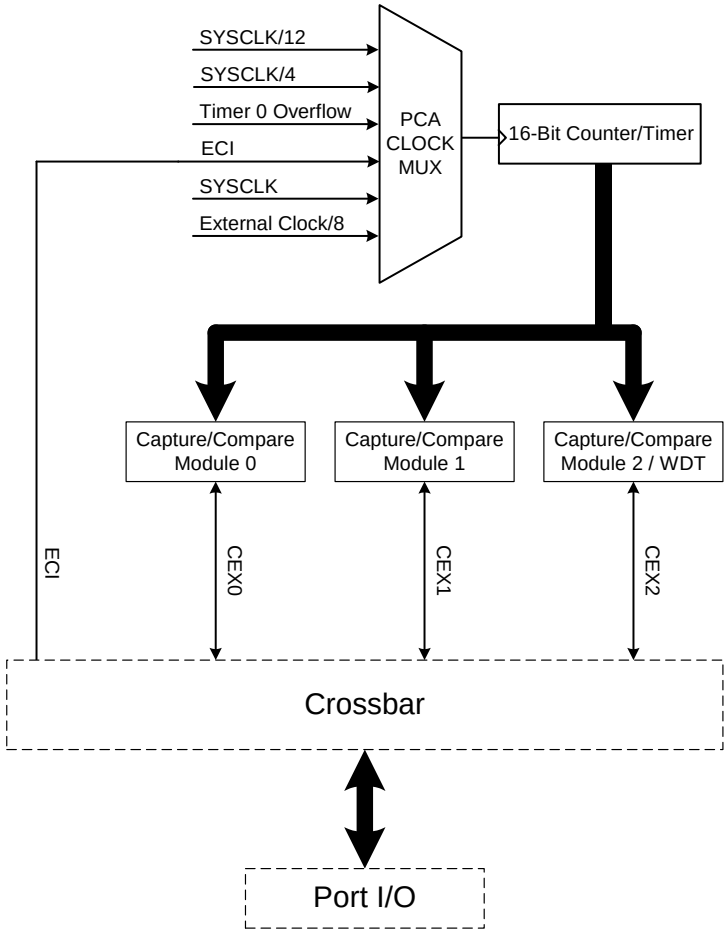



**Figure 23.1. PCA Block Diagram**

## 23.1. PCA Counter/Timer

The 16-bit PCA counter/timer consists of two 8-bit SFRs: PCA0L and PCA0H. PCA0H is the high byte (MSB) of the 16-bit counter/timer and PCA0L is the low byte (LSB). Reading PCA0L automatically latches the value of PCA0H into a “snapshot” register; the following PCA0H read accesses this “snapshot” register. **Reading the PCA0L Register first guarantees an accurate reading of the entire 16-bit PCA0 counter.** Reading PCA0H or PCA0L does not disturb the counter operation. The CPS2–CPS0 bits in the PCA0MD register select the timebase for the counter/timer as shown in Table 23.1.

When the counter/timer overflows from 0xFFFF to 0x0000, the Counter Overflow Flag (CF) in PCA0MD is set to logic 1 and an interrupt request is generated if CF interrupts are enabled. Setting the ECF bit in PCA0MD to logic 1 enables the CF flag to generate an interrupt request. The CF bit is not automatically cleared by hardware when the CPU vectors to the interrupt service routine, and must be cleared by software (Note: PCA0 interrupts must be globally enabled before CF interrupts are recognized. PCA0 interrupts are globally enabled by setting the EA bit (IE.7) and the EPCA0 bit in EIE1 to logic 1). Clearing the CIDL bit in the PCA0MD register allows the PCA to continue normal operation while the CPU is in Idle mode.

**Table 23.1. PCA Timebase Input Options**

| CPS2 | CPS1 | CPS0 | Timebase |
|---|---|---|---|
| 0 | 0 | 0 | System clock divided by 12 |
| 0 | 0 | 1 | System clock divided by 4 |
| 0 | 1 | 0 | Timer 0 overflow |
| 0 | 1 | 1 | High-to-low transitions on ECI (max rate = system clock divided by 4) |
| 1 | 0 | 0 | System clock |
| 1 | 0 | 1 | External oscillator source divided by 8* |
| ***Note:** External oscillator source divided by 8 is synchronized with the system clock. | | | |

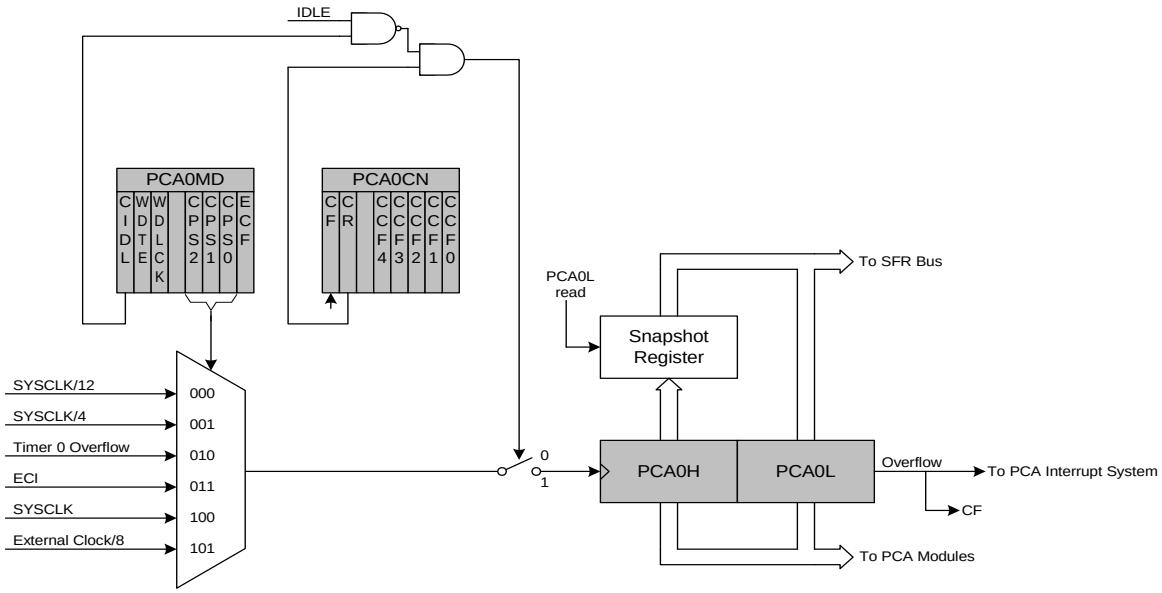



**Figure 23.2. PCA Counter/Timer Block Diagram**

## 23.2. Capture/Compare Modules

Each module can be configured to operate independently in one of six operation modes: Edge-triggered Capture, Software Timer, High Speed Output, Frequency Output, 8-Bit Pulse Width Modulator, or 16-Bit Pulse Width Modulator. Each module has Special Function Registers (SFRs) associated with it in the CIP-51 system controller. These registers are used to exchange data with a module and configure the module's mode of operation.

Table 23.2 summarizes the bit settings in the PCA0CPMn registers used to select the PCA capture/compare module's operating modes. Setting the ECCFn bit in a PCA0CPMn register enables the module's CCFn interrupt. Note: PCA0 interrupts must be globally enabled before individual CCFn interrupts are recognized. PCA0 interrupts are globally enabled by setting the EA bit and the EPCA0 bit to logic 1. See Figure 23.3 for details on the PCA interrupt configuration.

**Table 23.2. PCA0CPM Register Settings for PCA Capture/Compare Modules**

| PWM16 | ECOM | CAPP | CAPN | MAT | TOG | PWM | ECCF | Operation Mode |
|---|---|---|---|---|---|---|---|---|
| X | X | 1 | 0 | 0 | 0 | 0 | X | Capture triggered by positive edge on CEXn |
| X | X | 0 | 1 | 0 | 0 | 0 | X | Capture triggered by negative edge on CEXn |
| X | X | 1 | 1 | 0 | 0 | 0 | X | Capture triggered by transition on CEXn |
| X | 1 | 0 | 0 | 1 | 0 | 0 | X | Software Timer |
| X | 1 | 0 | 0 | 1 | 1 | 0 | X | High Speed Output |
| X | 1 | 0 | 0 | X | 1 | 1 | X | Frequency Output |
| 0 | 1 | 0 | 0 | X | 0 | 1 | X | 8-Bit Pulse Width Modulator |
| 1 | 1 | 0 | 0 | X | 0 | 1 | X | 16-Bit Pulse Width Modulator |
| X = Don't Care | | | | | | | | |

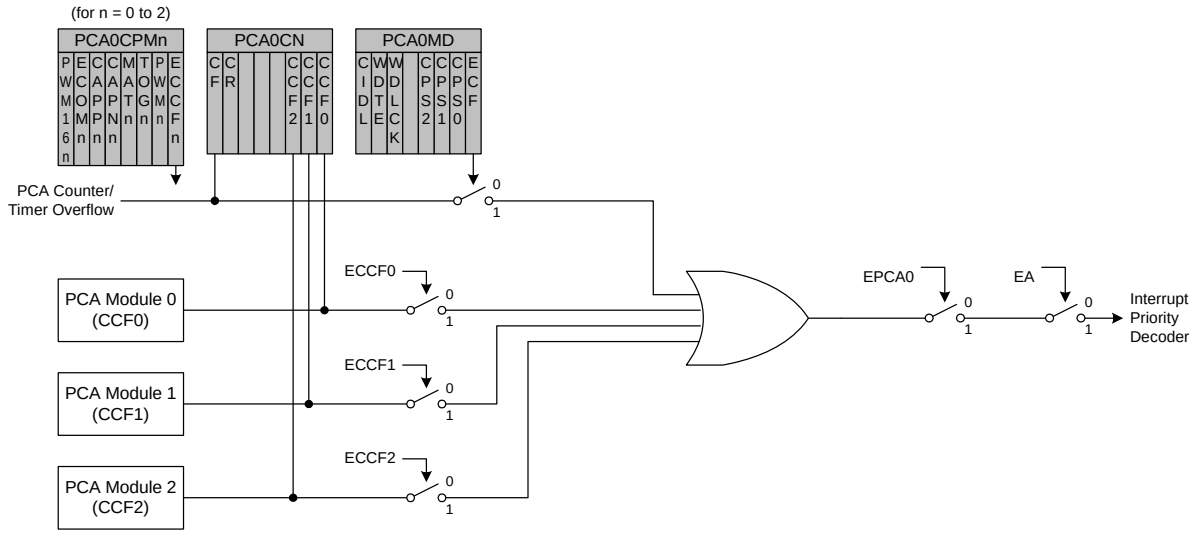



**Figure 23.3. PCA Interrupt Block Diagram**

### 23.2.1. Edge-triggered Capture Mode

In this mode, a valid transition on the CEXn pin causes the PCA to capture the value of the PCA counter/timer and load it into the corresponding module's 16-bit capture/compare register (PCA0CPLn and PCA0CPHn). The CAPPn and CAPNn bits in the PCA0CPMn register are used to select the type of transition that triggers the capture: low-to-high transition (positive edge), high-to-low transition (negative edge), or either transition (positive or negative edge). When a capture occurs, the Capture/Compare Flag (CCFn) in PCA0CN is set to logic 1 and an interrupt request is generated if CCF interrupts are enabled. The CCFn bit is not automatically cleared by hardware when the CPU vectors to the interrupt service routine, and must be cleared by software. If both CAPPn and CAPNn bits are set to logic 1, then the state of the Port pin associated with CEXn can be read directly to determine whether a rising-edge or falling-edge caused the capture.

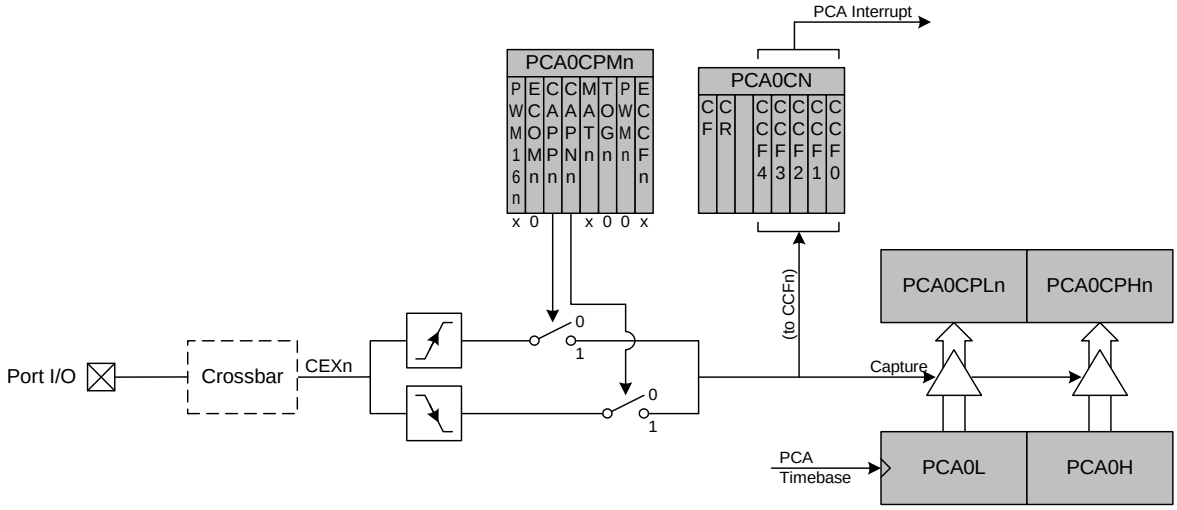



**Figure 23.4. PCA Capture Mode Diagram**

Note: The CEXn input signal must remain high or low for at least 2 system clock cycles to be recognized by the hardware.

### 23.2.2. Software Timer (Compare) Mode

In Software Timer mode, the PCA counter/timer value is compared to the module's 16-bit capture/compare register (PCA0CPHn and PCA0CPLn). When a match occurs, the Capture/Compare Flag (CCFn) in PCA0CN is set to logic 1 and an interrupt request is generated if CCF interrupts are enabled. The CCFn bit is not automatically cleared by hardware when the CPU vectors to the interrupt service routine, and must be cleared by software. Setting the ECOMn and MATn bits in the PCA0CPMn register enables Software Timer mode.

**Important Note About Capture/Compare Registers**: When writing a 16-bit value to the PCA0 Capture/Compare registers, the low byte should always be written first. Writing to PCA0CPLn clears the ECOMn bit to '0'; writing to PCA0CPHn sets ECOMn to '1'.

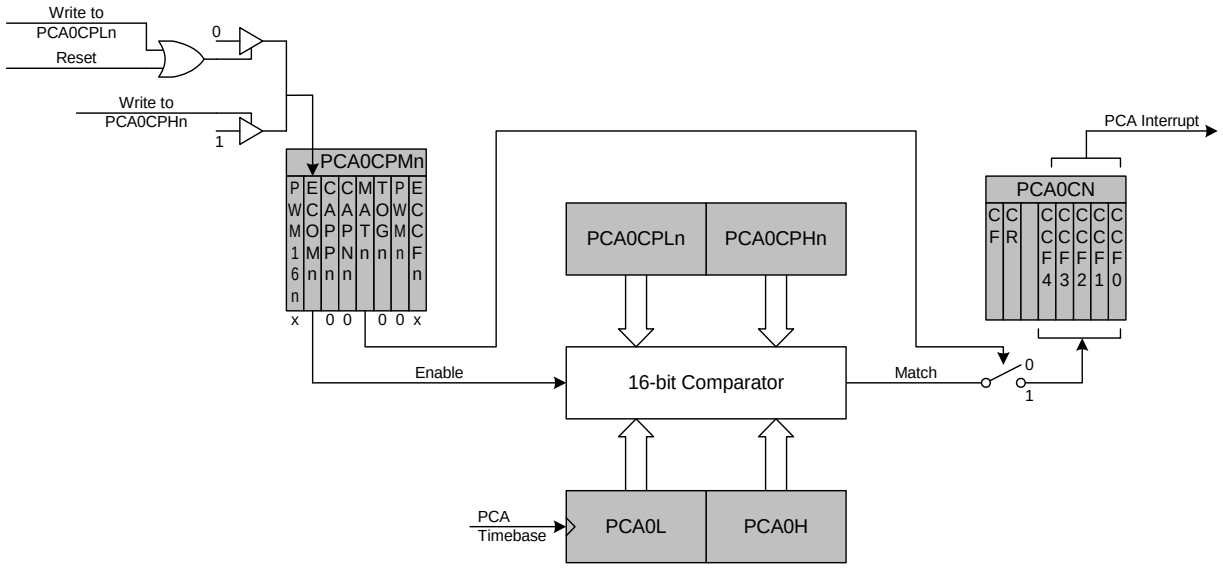



**Figure 23.5. PCA Software Timer Mode Diagram**

### 23.2.3. High Speed Output Mode

In High Speed Output mode, a module's associated CEXn pin is toggled each time a match occurs between the PCA Counter and the module's 16-bit capture/compare register (PCA0CPHn and PCA0CPLn) Setting the TOGn, MATn, and ECOMn bits in the PCA0CPMn register enables the High-Speed Output mode.

**Important Note About Capture/Compare Registers**: When writing a 16-bit value to the PCA0 Capture/Compare registers, the low byte should always be written first. Writing to PCA0CPLn clears the ECOMn bit to '0'; writing to PCA0CPHn sets ECOMn to '1'.

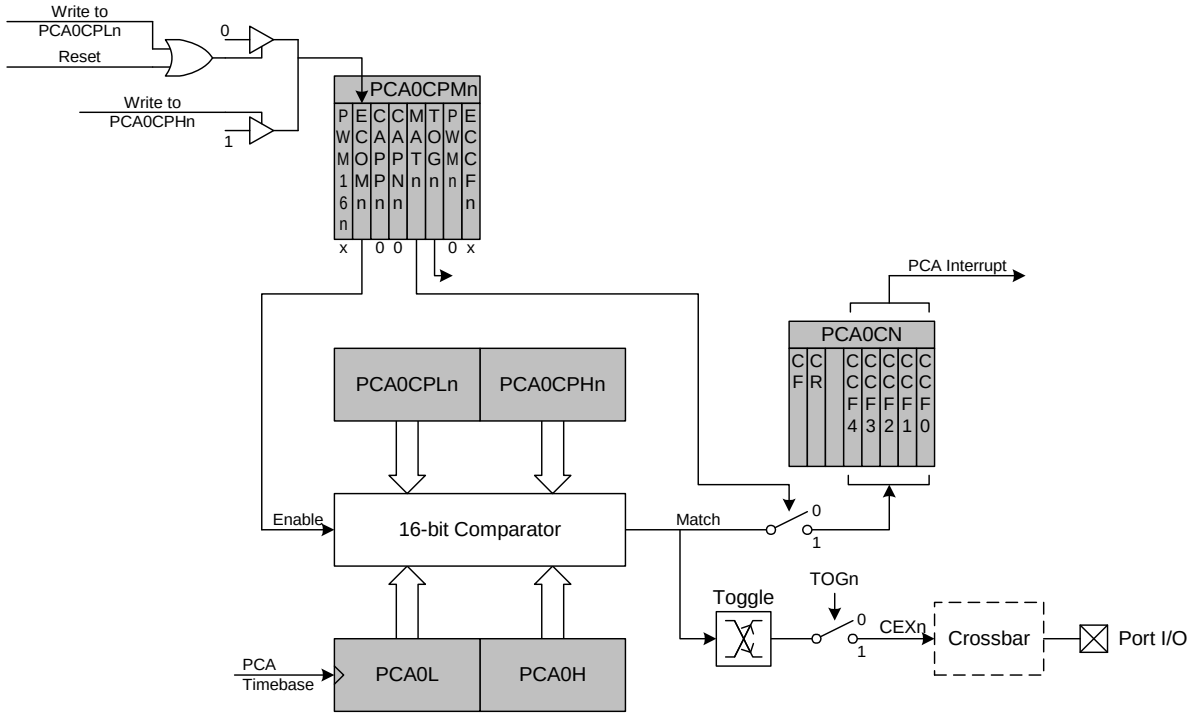



**Figure 23.6. PCA High Speed Output Mode Diagram**

### 23.2.4. Frequency Output Mode

Frequency Output Mode produces a programmable-frequency square wave on the module's associated CEXn pin. The capture/compare module high byte holds the number of PCA clocks to count before the output is toggled. The frequency of the square wave is then defined by Equation 23.1.

$$F_{CEXn} = \frac{F_{PCA}}{2 \times PCA0CPHn}$$

Note: A value of 0x00 in the PCA0CPHn register is equal to 256 for this equation.

**Equation 23.1. Square Wave Frequency Output**

Where $F_{PCA}$ is the frequency of the clock selected by the CPS2–0 bits in the PCA mode register, PCA0MD. The lower byte of the capture/compare module is compared to the PCA counter low byte; on a match, CEXn is toggled and the offset held in the high byte is added to the matched value in PCA0CPLn. Frequency Output Mode is enabled by setting the ECOMn, TOGn, and PWMn bits in the PCA0CPMn register.

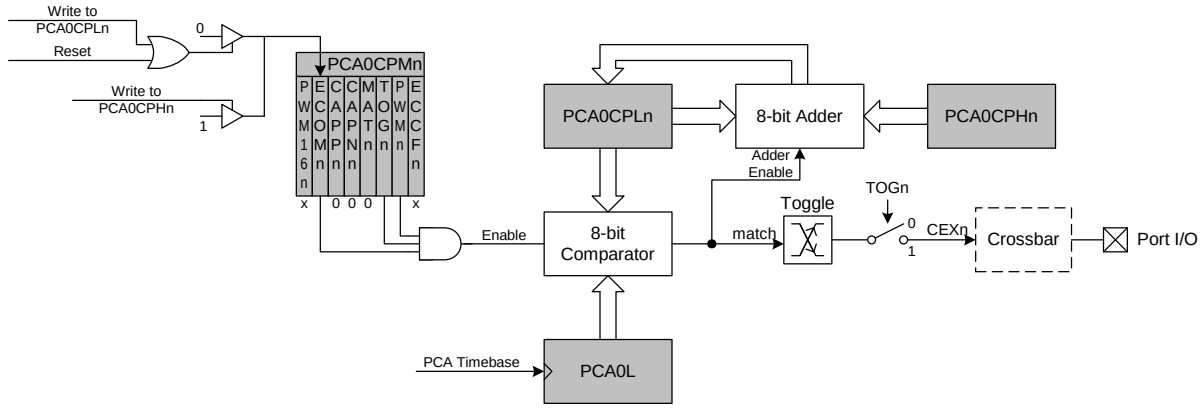



**Figure 23.7. PCA Frequency Output Mode**

### 23.2.5. 8-Bit Pulse Width Modulator Mode

Each module can be used independently to generate a pulse width modulated (PWM) output on its associated CEXn pin. The frequency of the output is dependent on the timebase for the PCA counter/timer. The duty cycle of the PWM output signal is varied using the module's PCA0CPLn capture/compare register. When the value in the low byte of the PCA counter/timer (PCA0L) is equal to the value in PCA0CPLn, the output on the CEXn pin will be set. When the count value in PCA0L overflows, the CEXn output will be reset (see Figure 23.8). Also, when the counter/timer low byte (PCA0L) overflows from 0xFF to 0x00, PCA0CPLn is reloaded automatically with the value stored in the module's capture/compare high byte (PCA0CPHn) without software intervention. Setting the ECOMn and PWMn bits in the PCA0CPMn register enables 8-Bit Pulse Width Modulator mode. The duty cycle for 8-Bit PWM Mode is given by Equation 23.2.

**Important Note About Capture/Compare Registers**: When writing a 16-bit value to the PCA0 Capture/Compare registers, the low byte should always be written first. Writing to PCA0CPLn clears the ECOMn bit to '0'; writing to PCA0CPHn sets ECOMn to '1'.

$$DutyCycle = \frac{(256 - PCA0CPHn)}{256}$$

**Equation 23.2. 8-Bit PWM Duty Cycle**

Using Equation 23.2, the largest duty cycle is 100% (PCA0CPHn = 0), and the smallest duty cycle is 0.39% (PCA0CPHn = 0xFF). A 0% duty cycle may be generated by clearing the ECOMn bit to '0'.

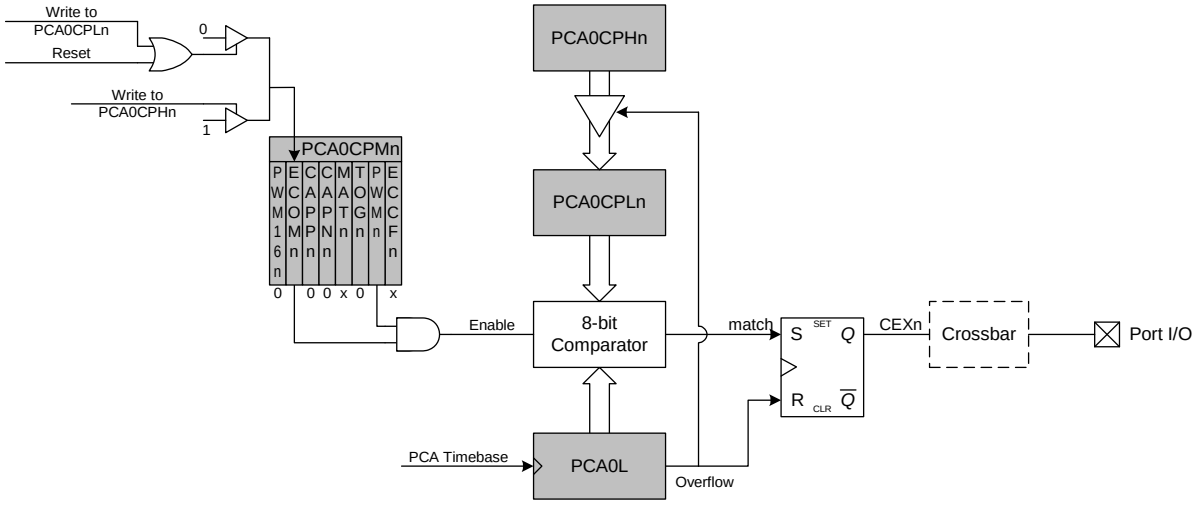



**Figure 23.8. PCA 8-Bit PWM Mode Diagram**

### 23.2.6. 16-Bit Pulse Width Modulator Mode

A PCA module may also be operated in 16-Bit PWM mode. In this mode, the 16-bit capture/compare module defines the number of PCA clocks for the low time of the PWM signal. When the PCA counter matches the module contents, the output on CEXn is asserted high; when the counter overflows, CEXn is asserted low. To output a varying duty cycle, new value writes should be synchronized with PCA CCFn match interrupts. 16-Bit PWM Mode is enabled by setting the ECOMn, PWMn, and PWM16n bits in the PCA0CPMn register. For a varying duty cycle, match interrupts should be enabled (ECCFn = 1 AND MATn = 1) to help synchronize the capture/compare register writes. The duty cycle for 16-Bit PWM Mode is given by Equation 23.3.

**Important Note About Capture/Compare Registers**: When writing a 16-bit value to the PCA0 Capture/Compare registers, the low byte should always be written first. Writing to PCA0CPLn clears the ECOMn bit to '0'; writing to PCA0CPHn sets ECOMn to '1'.

$$DutyCycle = \frac{(65536 - PCA0CPn)}{65536}$$

**Equation 23.3. 16-Bit PWM Duty Cycle**

Using Equation 23.3, the largest duty cycle is 100% (PCA0CPn = 0), and the smallest duty cycle is 0.0015% (PCA0CPn = 0xFFFF). A 0% duty cycle may be generated by clearing the ECOMn bit to '0'.

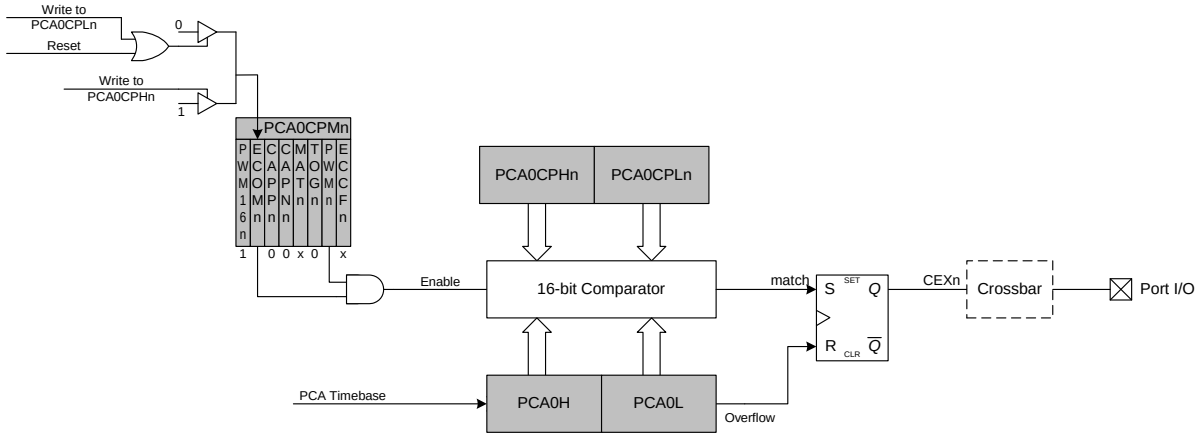



**Figure 23.9. PCA 16-Bit PWM Mode**

## 23.3. Watchdog Timer Mode

A programmable watchdog timer (WDT) function is available through the PCA Module 2. The WDT is used to generate a reset if the time between writes to the WDT update register (PCA0CPH2) exceed a specified limit. The WDT can be configured and enabled/disabled as needed by software.

With the WDTE bit set in the PCA0MD register, Module 2 operates as a watchdog timer (WDT). The Module 2 high byte is compared to the PCA counter high byte; the Module 2 low byte holds the offset to be used when WDT updates are performed. **The Watchdog Timer is enabled on reset. Writes to some PCA registers are restricted while the Watchdog Timer is enabled.**

### 23.3.1. Watchdog Timer Operation

While the WDT is enabled:

- PCA counter is forced on.
- Writes to PCA0L and PCA0H are not allowed.
- PCA clock source bits (CPS2–CPS0) are frozen.
- PCA Idle control bit (CIDL) is frozen.
- Module 2 is forced into software timer mode.
- Writes to the Module 2 mode register (PCA0CPM2) are disabled.

While the WDT is enabled, writes to the CR bit will not change the PCA counter state; the counter will run until the WDT is disabled. The PCA counter run control (CR) will read zero if the WDT is enabled but user software has not enabled the PCA counter. If a match occurs between PCA0CPH2 and PCA0H while the WDT is enabled, a reset will be generated. To prevent a WDT reset, the WDT may be updated with a write of any value to PCA0CPH2. Upon a PCA0CPH2 write, PCA0H plus the offset held in PCA0CPL2 is loaded into PCA0CPH2 (See Figure 23.10).

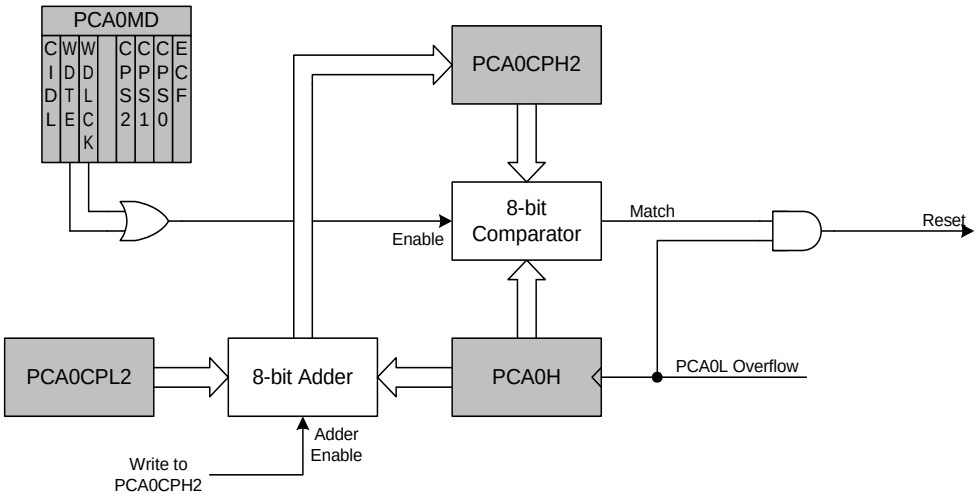



**Figure 23.10. PCA Module 2 with Watchdog Timer Enabled**

Note that the 8-bit offset held in PCA0CPH2 is compared to the upper byte of the 16-bit PCA counter. This offset value is the number of PCA0L overflows before a reset. Up to 256 PCA clocks may pass before the first PCA0L overflow occurs, depending on the value of the PCA0L when the update is performed. The total offset is then given (in PCA clocks) by Equation 23.4, where PCA0L is the value of the PCA0L register at the time of the update.

$$Offset = (256 \times PCA0CPL2) + (256 - PCA0L)$$

**Equation 23.4. Watchdog Timer Offset in PCA Clocks**

The WDT reset is generated when PCA0L overflows while there is a match between PCA0CPH2 and PCA0H. Software may force a WDT reset by writing a '1' to the CCF2 flag (PCA0CN.2) while the WDT is enabled.

### 23.3.2. Watchdog Timer Usage

To configure the WDT, perform the following tasks:

- Disable the WDT by writing a '0' to the WDTE bit.
- Select the desired PCA clock source (with the CPS2–CPS0 bits).
- Load PCA0CPL2 with the desired WDT update offset value.
- Configure the PCA Idle mode (set CIDL if the WDT should be suspended while the CPU is in Idle mode).
- Enable the WDT by setting the WDTE bit to '1'.
- Write a value to PCA0CPH2 to reload the WDT.

The PCA clock source and Idle mode select cannot be changed while the WDT is enabled. The watchdog timer is enabled by setting the WDTE or WDLCK bits in the PCA0MD register. When WDLCK is set, the WDT cannot be disabled until the next system reset. If WDLCK is not set, the WDT is disabled by clearing the WDTE bit.

The WDT is enabled following any reset. The PCA0 counter clock defaults to the system clock divided by 12, PCA0L defaults to 0x00, and PCA0CPL2 defaults to 0x00. Using Equation 23.4, this results in a WDT timeout interval of 256 PCA clock cycles, or 3072 system clock cycles. Table 23.3 lists some example timeout intervals for typical system clocks.

**Table 23.3. Watchdog Timer Timeout Intervals[1]**

| System Clock (Hz) | PCA0CPL2 | Timeout Interval (ms) |
|---|---|---|
| 24,500,000 | 255 | 32.1 |
| 24,500,000 | 128 | 16.2 |
| 24,500,000 | 32 | 4.1 |
| 18,432,000 | 255 | 42.7 |
| 18,432,000 | 128 | 21.5 |
| 18,432,000 | 32 | 5.5 |
| 11,059,200 | 255 | 71.1 |
| 11,059,200 | 128 | 35.8 |
| 11,059,200 | 32 | 9.2 |
| 3,062,500[2] | 255 | 257 |
| 3,062,500[2] | 128 | 129.5 |
| 3,062,500[2] | 32 | 33.1 |
| 32,000 | 255 | 24576 |
| 32,000 | 128 | 12384 |
| 32,000 | 32 | 3168 |

**Notes:**

1. Assumes SYSCLK / 12 as the PCA clock source, and a PCA0L value of 0x00 at the update time.
2. SYSCLK reset frequency = Internal oscillator frequency divided by 8.

## 23.4. Register Descriptions for PCA

Following are detailed descriptions of the special function registers related to the operation of the PCA.

### SFR Definition 23.1. PCA0CN: PCA Control

| R/W | R/W | R | R | R | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| CF | CR | — | — | — | CCF2 | CCF1 | CCF0 | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | Bit Addressable |

SFR Address: 0xD8

Bit7: CF: PCA Counter/Timer Overflow Flag.
Set by hardware when the PCA Counter/Timer overflows from 0xFFFF to 0x0000. When the Counter/Timer Overflow (CF) interrupt is enabled, setting this bit causes the CPU to vector to the PCA interrupt service routine. This bit is not automatically cleared by hardware and must be cleared by software.

Bit6: CR: PCA Counter/Timer Run Control.
This bit enables/disables the PCA Counter/Timer.
0: PCA Counter/Timer disabled.
1: PCA Counter/Timer enabled.

Bits5–3: UNUSED. Read = 000b, Write = don't care.

Bit2: CCF2: PCA Module 2 Capture/Compare Flag.
This bit is set by hardware when a match or capture occurs. When the CCF2 interrupt is enabled, setting this bit causes the CPU to vector to the PCA interrupt service routine. This bit is not automatically cleared by hardware and must be cleared by software.

Bit1: CCF1: PCA Module 1 Capture/Compare Flag.
This bit is set by hardware when a match or capture occurs. When the CCF1 interrupt is enabled, setting this bit causes the CPU to vector to the PCA interrupt service routine. This bit is not automatically cleared by hardware and must be cleared by software.

Bit0: CCF0: PCA Module 0 Capture/Compare Flag.
This bit is set by hardware when a match or capture occurs. When the CCF0 interrupt is enabled, setting this bit causes the CPU to vector to the PCA interrupt service routine. This bit is not automatically cleared by hardware and must be cleared by software.

## SFR Definition 23.2. PCA0MD: PCA Mode

| R/W | R/W | R/W | R | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| CIDL | WDTE | WDLCK | — | CPS2 | CPS1 | CPS0 | ECF | 01000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | Bit Addressable |

SFR Address: 0xD9

Bit7: CIDL: PCA Counter/Timer Idle Control.
Specifies PCA behavior when CPU is in Idle Mode.
0: PCA continues to function normally while the system controller is in Idle Mode.
1: PCA operation is suspended while the system controller is in Idle Mode.

Bit6: WDTE: Watchdog Timer Enable
If this bit is set, PCA Module 2 is used as the watchdog timer.
0: Watchdog Timer disabled.
1: PCA Module 2 enabled as Watchdog Timer.

Bit5: WDLCK: Watchdog Timer Lock
This bit locks/unlocks the Watchdog Timer Enable. When WDLCK is set, the Watchdog Timer may not be disabled until the next system reset.
0: Watchdog Timer Enable unlocked.
1: Watchdog Timer Enable locked.

Bit4: UNUSED. Read = 0b, Write = don't care.

Bits3–1: CPS2–CPS0: PCA Counter/Timer Pulse Select.
These bits select the timebase source for the PCA counter.

| CPS2 | CPS1 | CPS0 | Timebase |
|---|---|---|---|
| 0 | 0 | 0 | System clock divided by 12 |
| 0 | 0 | 1 | System clock divided by 4 |
| 0 | 1 | 0 | Timer 0 overflow |
| 0 | 1 | 1 | High-to-low transitions on ECI (max rate = system clock divided by 4) |
| 1 | 0 | 0 | System clock |
| 1 | 0 | 1 | External clock divided by 8* |
| 1 | 1 | 0 | Reserved |
| 1 | 1 | 1 | Reserved |
| ***Note:** External oscillator source divided by 8 is synchronized with the system clock. | | | |

Bit0: ECF: PCA Counter/Timer Overflow Interrupt Enable.
This bit sets the masking of the PCA Counter/Timer Overflow (CF) interrupt.
0: Disable the CF interrupt.
1: Enable a PCA Counter/Timer Overflow interrupt request when CF (PCA0CN.7) is set.

**Note: When the WDTE bit is set to '1', the PCA0MD register cannot be modified. To change the contents of the PCA0MD register, the Watchdog Timer must first be disabled.**

## SFR Definition 23.3. PCA0CPMn: PCA Capture/Compare Mode

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| PWM16n | ECOMn | CAPPn | CAPNn | MATn | TOGn | PWMn | ECCFn | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: PCA0CPM0: 0xDA, PCA0CPM1: 0xDB, PCA0CPM2: 0xDC

Bit7: PWM16n: 16-bit Pulse Width Modulation Enable.
This bit selects 16-bit mode when Pulse Width Modulation mode is enabled (PWMn = 1).
0: 8-bit PWM selected.
1: 16-bit PWM selected.

Bit6: ECOMn: Comparator Function Enable.
This bit enables/disables the comparator function for PCA module n.
0: Disabled.
1: Enabled.

Bit5: CAPPn: Capture Positive Function Enable.
This bit enables/disables the positive edge capture for PCA module n.
0: Disabled.
1: Enabled.

Bit4: CAPNn: Capture Negative Function Enable.
This bit enables/disables the negative edge capture for PCA module n.
0: Disabled.
1: Enabled.

Bit3: MATn: Match Function Enable.
This bit enables/disables the match function for PCA module n. When enabled, matches of the PCA counter with a module's capture/compare register cause the CCFn bit in PCA0MD register to be set to logic 1.
0: Disabled.
1: Enabled.

Bit2: TOGn: Toggle Function Enable.
This bit enables/disables the toggle function for PCA module n. When enabled, matches of the PCA counter with a module's capture/compare register cause the logic level on the CEXn pin to toggle. If the PWMn bit is also set to logic 1, the module operates in Frequency Output Mode.
0: Disabled.
1: Enabled.

Bit1: PWMn: Pulse Width Modulation Mode Enable.
This bit enables/disables the PWM function for PCA module n. When enabled, a pulse width modulated signal is output on the CEXn pin. 8-bit PWM is used if PWM16n is cleared; 16-bit mode is used if PWM16n is set to logic 1. If the TOGn bit is also set, the module operates in Frequency Output Mode.
0: Disabled.
1: Enabled.

Bit0: ECCFn: Capture/Compare Flag Interrupt Enable.
This bit sets the masking of the Capture/Compare Flag (CCFn) interrupt.
0: Disable CCFn interrupts.
1: Enable a Capture/Compare Flag interrupt request when CCFn is set.

## SFR Definition 23.4. PCA0L: PCA Counter/Timer Low Byte

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: 0xF9

Bits 7–0: PCA0L: PCA Counter/Timer Low Byte.
The PCA0L register holds the low byte (LSB) of the 16-bit PCA Counter/Timer.

## SFR Definition 23.5. PCA0H: PCA Counter/Timer High Byte

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | SFR Address: |

SFR Address: 0xFA

Bits 7–0: PCA0H: PCA Counter/Timer High Byte.
The PCA0H register holds the high byte (MSB) of the 16-bit PCA Counter/Timer.

### SFR Definition 23.6. PCA0CPLn: PCA Capture Module Low Byte

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: PCA0CPL0: 0xE9, PCA0CPL1: 0xEB, PCA0CPL2: 0xED

Bits7–0: PCA0CPLn: PCA Capture Module Low Byte.
The PCA0CPLn register holds the low byte (LSB) of the 16-bit capture module n.

### SFR Definition 23.7. PCA0CPHn: PCA Capture Module High Byte

| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | Reset Value |
|---|---|---|---|---|---|---|---|---|
| | | | | | | | | 00000000 |
| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 | |

SFR Address: PCA0CPH0: 0xEA, PCA0CPH1: 0xEB, PCA0CPH2: 0xEE

Bits7–0: PCA0CPHn: PCA Capture Module High Byte.
The PCA0CPHn register holds the high byte (MSB) of the 16-bit capture module n.

# 24. Revision Specific Behavior

This chapter describes a functional difference between C8051F35x “REV B” and “REV C” or later devices. The functionality of the VREF- pin differs between these revisions.

## 24.1. Revision Identification

The Lot ID Code on the top side of the device package can be used for decoding device revision information. On C8051F350/2 devices, the revision letter is the first letter of the Lot ID Code. On C8051F351/3 devices, the revision letter is the first of the Lot ID Code. Figure 24.1 shows how to find the revision on the top side of the device package.

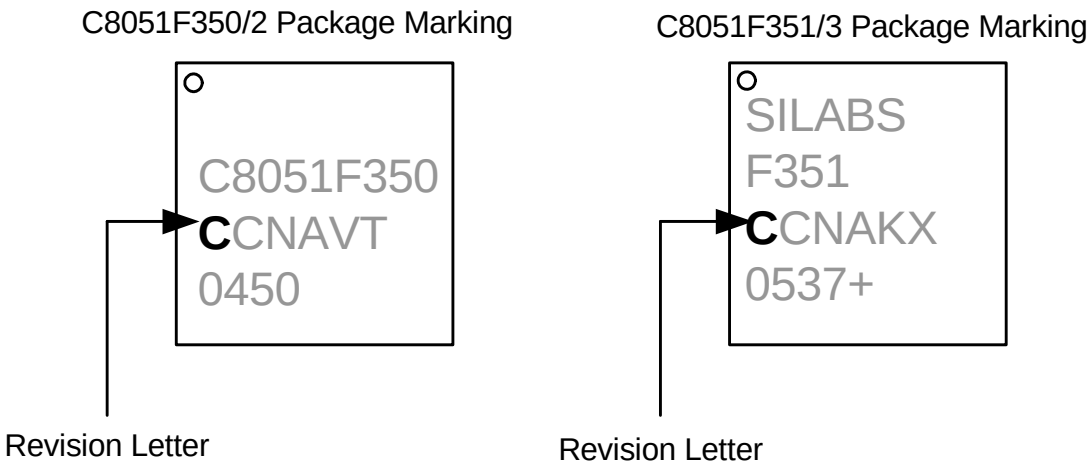



**Figure 24.1. Reading Package Marking**

## 24.1. VREF- pin

The required connection for the VREF- pin differs between the "REV B" and "REV C" and later devices. On "REV B" devices, when the internal voltage reference is enabled, the VREF- pin is internally connected to GND so the VREF- pin can be left unconnected externally.

On "REV C" and later devices, when the internal voltage reference is enabled, the VREF- pin is not internally connected to GND. The VREF- pin must be connected to GND externally for the voltage reference to operate properly.

## NOTES:

# 25. C2 Interface

C8051F350/1/2/3 devices include an on-chip Silicon Labs 2-Wire (C2) debug interface to allow Flash programming and in-system debugging with the production part installed in the end application. The C2 interface uses a clock signal (C2CK) and a bi-directional C2 data signal (C2D) to transfer information between the device and a host system. See the C2 Interface Specification for details on the C2 protocol.

## 25.1. C2 Interface Registers

The following describes the C2 registers necessary to allow Flash programming and in-system debugging with the production part installed through the C2 interface. All C2 registers are accessed through the C2 interface as described in the C2 Interface Specification.

### C2 Register Definition 25.1. C2ADD: C2 Address

Reset Value
00000000

| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 |
|---|---|---|---|---|---|---|---|
| | | | | | | | |

Bits7–0: The C2ADD register is accessed via the C2 interface to select the target Data register for C2 Data Read and Data Write commands.

| Address | Description |
|---|---|
| 0x00 | Selects the Device ID register for Data Read instructions |
| 0x01 | Selects the Revision ID register for Data Read instructions |
| 0x02 | Selects the C2 Flash Programming Control register for Data Read/Write instructions |
| 0xB4 | Selects the C2 Flash Programming Data register for Data Read/Write instructions |

### C2 Register Definition 25.2. DEVICEID: C2 Device ID

Reset Value
00001011

| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 |
|---|---|---|---|---|---|---|---|
| | | | | | | | |

This read-only register returns the 8-bit device ID: 0x0B (C8051F350/1/2/3).

### C2 Register Definition 25.3. REVID: C2 Revision ID

Reset Value
Variable

| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 |
|---|---|---|---|---|---|---|---|
| | | | | | | | |

This read-only register returns the 8-bit revision ID.

### C2 Register Definition 25.4. FPCTL: C2 Flash Programming Control

Reset Value
00000000

| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 |
|---|---|---|---|---|---|---|---|
| | | | | | | | |

Bits7–0 FPCTL: Flash Programming Control Register.
This register is used to enable Flash programming via the C2 interface. To enable C2 Flash programming, the following codes must be written in order: 0x02, 0x01. Note that once C2 Flash programming is enabled, a system reset must be issued to resume normal operation.

### C2 Register Definition 25.5. FPDAT: C2 Flash Programming Data

Reset Value
00000000

| Bit7 | Bit6 | Bit5 | Bit4 | Bit3 | Bit2 | Bit1 | Bit0 |
|---|---|---|---|---|---|---|---|
| | | | | | | | |

Bits7–0: FPDAT: C2 Flash Programming Data Register.
This register is used to pass Flash commands, addresses, and data during C2 Flash accesses. Valid commands are listed below.

| Code | Command |
|---|---|
| 0x06 | Flash Block Read |
| 0x07 | Flash Block Write |
| 0x08 | Flash Page Erase |
| 0x03 | Device Erase |

## 25.2. C2 Pin Sharing

The C2 protocol allows the C2 pins to be shared with user functions so that in-system debugging and Flash programming functions may be performed. This is possible because C2 communication is typically performed when the device is in the halt state, where all on-chip peripherals and user software are stalled. In this halted state, the C2 interface can safely ‘borrow’ the C2CK (/RST) and C2D (P2.0) pins. In most applications, external resistors are required to isolate C2 interface traffic from the user application. A typical isolation configuration is shown in Figure 25.1.

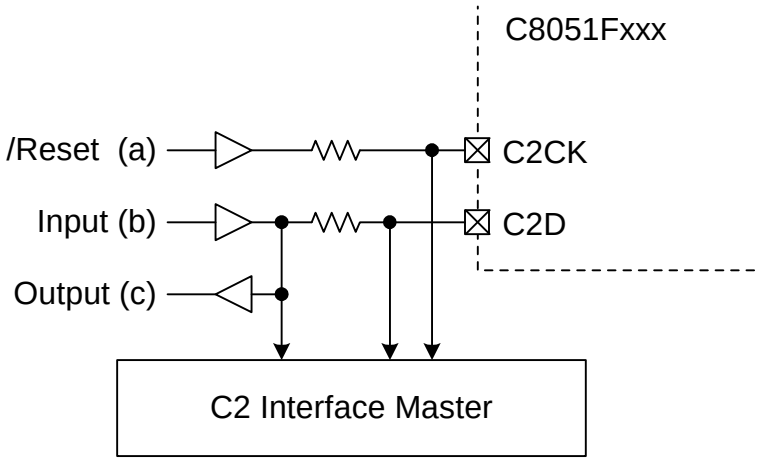



**Figure 25.1. Typical C2 Pin Sharing**

The configuration in Figure 25.1 assumes the following:

1. The user input (b) cannot change state while the target device is halted.
2. The /RST pin on the target device is used as an input only.

Additional resistors may be necessary depending on the specific application.

# DOCUMENT CHANGE LIST

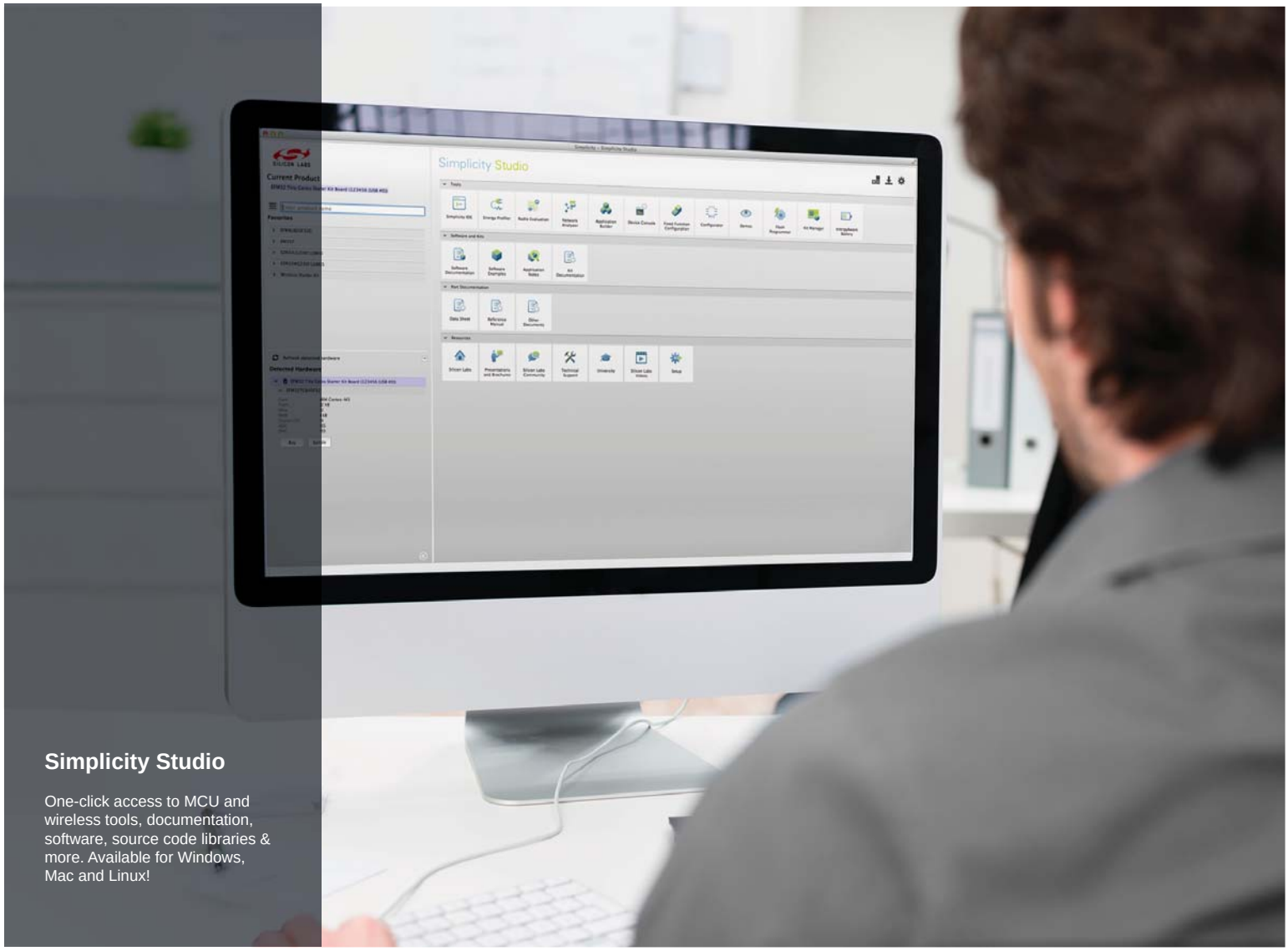
## Revision 0.4 to Revision 1.0

- Removed preliminary tag and updated various specifications.
- Updated package labeling and added "Lead-free (RoHS Compliant)" in Table 1.1, “Product Selection Guide,” on page 18.
- ADC chapter: Added Table 5.5, Table 5.6, Table 5.7, Table 5.8, and Table 5.9 on pages 63–65.
- Temperature Sensor chapter: Added Offset Error and Slope Error specifications to Table 8.1, “Temperature Sensor Electrical Characteristics,” on page 77.
- Reset Sources chapter: Table 14.1, “Reset Electrical Characteristics,” on page 120: Added $V_{DD}$ Ramp Time and changed “$V_{DD}$ POR Threshold” to “$V_{DD}$ Monitor Threshold.”
- Flash Memory chapter: Clarified descriptions of Flash security features.
- Oscillators chapter: Clarified external crystal initialization steps and added a specific 32.768 kHz crystal example.
- Oscillators chapter: Clarified external capacitor example.
- Port I/O chapter: Figure 18.3 and Figure 18.4, Crossbar Priority Decoder Tables: Changed PnSKIP[7:0] to PnSKIP[0:7] to match the Port I/O order.
- SMBus chapter: SFR Definition 19.1, SMB0CF register: Added a description of the behavior of Timer 3 in split mode if SMBTOE is set.
- PCA chapter: Updated Watchdog timer timeout intervals in Table 23.3 on page 221.
- C2 chapter: Removed references to “boundary scans.”

## Revision 1.0 to Revision 1.1

- Clarified text in Table 2.1, “Absolute Maximum Ratings,” on page 29.
- Updated Digital Supply Current values in Table 3.1, “Global DC Electrical Characteristics,” on page 30.
- Removed sentence in Section “5.2. Calibrating the ADC’ on page 44 that indicated the AD0CALC bit is cleared by clearing the AD0INT flag.
- Updated Table 5.3, “ADC0 Electrical Characteristics,” on page 61.
  - ADC input current
  - Burnout Current Source values
  - AV+ Supply Current values
- Added second note to SFR Definition 5.3.
- Updated Table 6.1, “IDAC Electrical Characteristics,” on page 72.
  - IDAC0 Gain-error Temp Co.
  - IDAC0 Power consumption values
- Clarified usage of VREF– pin in text and figure in 7. ‘Voltage Reference” on page 73.
- Updated Table 7.1, “Voltage Reference Electrical Characteristics,” on page 75.
  - VREF output voltage max and min
  - Power Specifications
- Corrected pins used by Comparator0 output in Section “9.1. Comparator0 Inputs and Outputs’ on page 83.
- Updated Comparator Power Consumptions values in Table 9.1, “Comparator Electrical Characteristics,” on page 85.
- Corrected maximum SMBus speed in Section “19. SMBus’ on page 151.
- Updated Table 19.4, “SMBus Status Decoding,” on page 167.
  - Slave Transmitter (0101 0XX)
  - Slave Receiver (0001 00X)
- Fixed Equation 23.4.
- Added last step to procedure described in Section “23.3.2. Watchdog Timer Usage’ on page 221.
- Changed Note 2 in Table 23.3, “Watchdog Timer Timeout Intervals1,” on page 221.
- Added Section “24. Revision Specific Behavior’ on page 227.

## NOTES:

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