Features

- High FSK Sensitivity: -106 dBm at 20 kBaud/-109.5 dBm at 2.4 kBaud (433.92 MHz)
- High ASK Sensitivity: -112.5 dBm at 10 kBaud/-116.5 dBm at 2.4 kBaud (433.92 MHz)
- Low Supply Current: 10.5 mA in RX and TX Mode (3 V/TX with 5 dBm)
- Data Rate 1 to 20 kBaud Manchester FSK, 1 to 10 kBaud Manchester ASK
- ASK/FSK Receiver Uses a Low-IF Architecture with High Selectivity, Blocking and Low Intermodulation (Typical Blocking 55 dB at ±750 kHz/61 dB at ±1.5 MHz and 70 dB at ±10 MHz, System I1dBCP = -30 dBm/System IIP3 = -20 dBm)
- 226 kHz IF Frequency with 30 dB Image Rejection and 170 kHz Usable IF Bandwidth
- Transmitter Uses Closed Loop Fractional-N Synthesizer for FSK Modulation with a High PLL Bandwidth and an Excellent Isolation between PLL and PA
- Tolerances of XTAL Compensated by Fractional-N Synthesizer with 800 Hz RF Resolution
- Integrated RX/TX-Switch, Single-ended RF Input and Output
- RSSI (Received Signal Strength Indicator)
- Communication to Microcontroller with SPI Interface Working at Maximum 500 kBit/s
- Configurable Self Polling and RX/TX Protocol Handling with FIFO-RAM Buffering of Received and Transmitted Data
- 5 Push Button Inputs and One Wake-up Input are Active in Power-down Mode
- Integrated XTAL Capacitors
- PA Efficiency: up to 38% (433 MHz/10 dBm/3 V)
- Low Inband Sensitivity Change of Typically ±1.8 dB within ±58 kHz Center Frequency Change in the Complete Temperature and Supply Voltage Range
- Supply Voltage Switch, Supply Voltage Regulator, Reset Generation, Clock/Interrupt Generation and Low Battery Indicator for Microcontroller
- Fully Integrated PLL with Low Phase Noise VCO and PLL Loop Filter
- Sophisticated Threshold Control and Quasi Peak Detector Circuit in the Data Slicer
- Power Management via Different Operation Modes
- 433.92 MHz, 868.3 MHz and 315 MHz without External VCO and PLL Components
- Inductive Supply with Voltage Regulator if Battery is Empty (AUX Mode)
- Efficient XTO Start-up Circuit (> -1.5 kΩ Worst Case Start Impedance)
- Changing of Modulation Type ASK/FSK and Data Rate without Component Changes
- Minimal External Circuitry Requirements for Complete System Solution
- Adjustable Output Power: 0 to 10 dBm Adjusted and Stabilized with External Resistor
- ESD Protection at all Pins (2 kV HBM, 200 V MM)
- Supply Voltage Range: 2.4 V to 3.6 V or 4.4 V to 6.6 V
- Temperature Range: -40°C to +105°C
- Small 7 × 7 mm QFN48 Package

Applications

- Automotive Keyless Entry and Passive Entry Go Systems
- Access Control Systems
- Remote Control Systems
- Alarm and Telemetry Systems
- Energy Metering
- Home Automation

Benefits

- No SAW Device Needed in Key Fob Designs to Meet Automotive Specifications
- Low System Cost Due to Very High System Integration Level
- Only One Crystal Needed in System
- Less Demanding Specification for the Microcontroller Due to Handling of Power-down Mode, Delivering of Clock, Reset, Low Battery Indication and Complete Handling of Receive/Transmit Protocol and Polling
- Single-ended Design with High Isolation of PLL/VCO from PA and the Power Supply Allows a Loop Antenna in the Key Fob to Surround the Whole Application

UHF ASK/FSK Transceiver

ATA5811
ATA5812

Preliminary

Rev. 4689B–RKE–04/04

General Description

The ATA5811/ATA5812 is a highly integrated UHF ASK/FSK single-channel half-duplex transceiver with low power consumption supplied in a small QFN48 package. The receive part is built as a fully integrated low-IF receiver, whereas direct PLL modulation with the fractional-N synthesizer is used for FSK transmission and switching of the power amplifier for ASK transmission.

The device supports data rates of 1 kBaud to 20 kBaud (FSK) and 1 kBaud to 10 kBaud (ASK) in Manchester, Bi-phase and other codes in transparent mode. The ATA5811 can be used in the 433 MHz to 435 MHz and the 868 MHz to 870 MHz band, the ATA5812 in the 314 MHz to 316 MHz band. The very high system integration level results in few numbers of external components needed.

Due to its blocking and selectivity performance, together with the additional 15 dB to 20 dB loss and the narrow bandwidth of a typical key fob loop antenna, a bulky blocking SAW is not needed in the key fob or sensor application. Additionally, the building blocks needed for a typical RKE and access control system on both sides, the base and the mobile stations, are fully integrated.

Its digital control logic with self polling and protocol generation enables a fast challenge response systems without using a high-performance microcontroller. Therefore, the ATA5811/ATA5812 contains a FIFO buffer RAM and can compose and receive the physical messages themselves. This provides more time for the microcontroller to carry out other functions such as calculating crypto algorithms, composing the logical messages and controlling other devices. Due to that, a standard 4-/8-bit microcontroller without special periphery and clocked with the CLK output of about 4.5 MHz is sufficient to control the communication link. This is especially valid for passive entry and access control systems, where within less than 100 ms several challenge response communications with arbitration of the communication partner have to be handled.

It is hence possible to design bi-directional RKE and access control systems with a fast challenge response crypto function with the same PCB board size and with the same current consumption as uni-directional RKE systems.

Figure 1. System Block Diagram

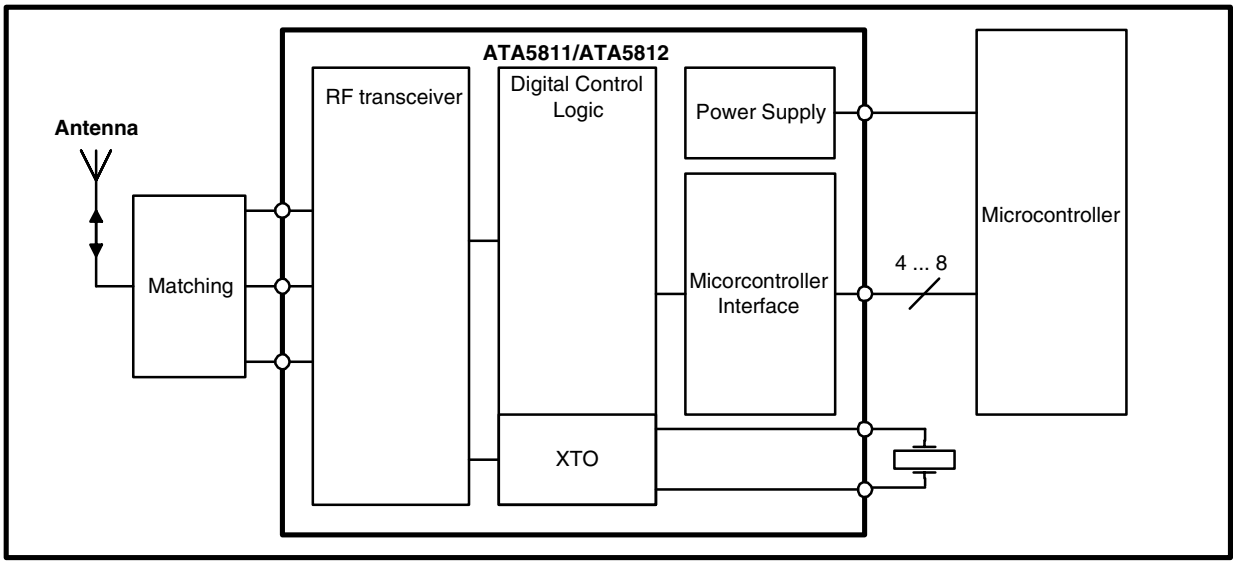
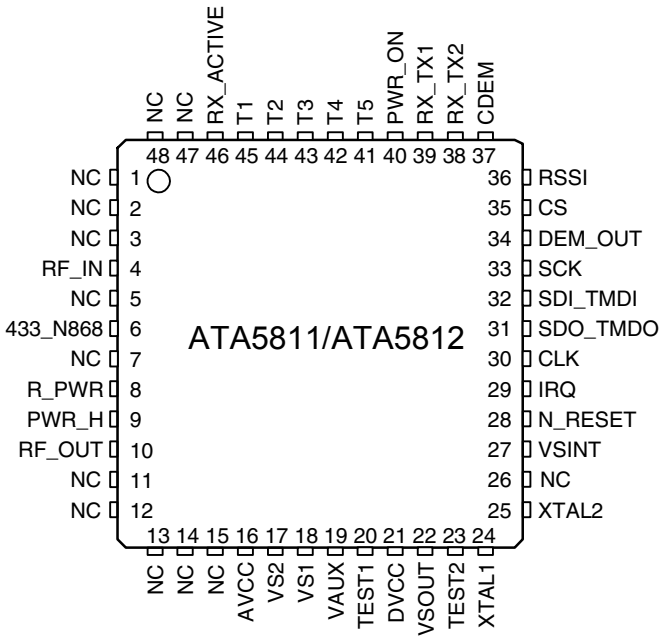


Figure 2. Pinning QFN48



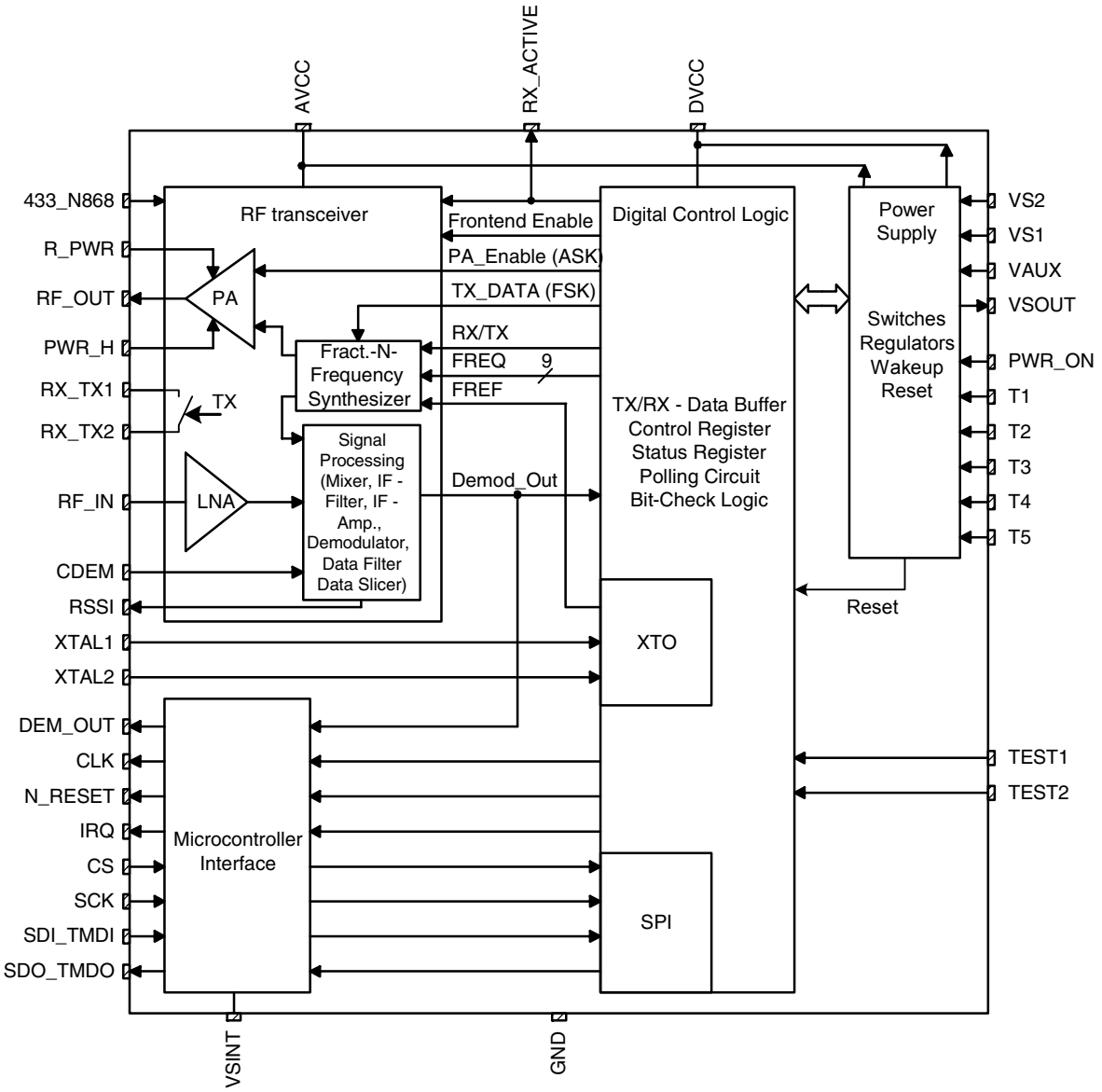
Pin Description

Pin	Symbol	Function
1	NC	Not connected
2	NC	Not connected
3	NC	Not connected
4	RF_IN	RF input
5	NC	Not connected
6	433_N868	Selects RF input/output frequency range
7	NC	Not connected
8	R_PWR	Resistor to adjust output power
9	PWR_H	Pin to select output power
10	RF_OUT	RF output
11	NC	Not connected
12	NC	Not connected
13	NC	Not connected
14	NC	Not connected
15	NC	Not connected
16	AVCC	Blocking of the analog voltage supply
17	VS2	Power supply input for voltage range 4.4 V to 6.6 V
18	VS1	Power supply input for voltage range 2.4 V to 3.6 V

Pin Description (Continued)

Pin	Symbol	Function
19	VAUX	Auxiliary supply voltage input
20	TEST1	Test input, at GND during operation
21	DVCC	Blocking of the digital voltage supply
22	VSOUT	Output voltage power supply for external devices
23	TEST2	Test input, at GND during operation
24	XTAL1	Reference crystal
25	XTAL2	Reference crystal
26	NC	Not connected
27	VSINT	Microcontroller Interface supply voltage
28	N_RESET	Output pin to reset a connected microcontroller
29	IRQ	Interrupt request
30	CLK	Output to clock a connected microcontroller
31	SDO_TMDO	Serial data out/transparent mode data out
32	SDI_TMDI	Serial data in/transparent mode data in
33	SCK	Serial clock
34	DEM_OUT	Demodulator open drain output signal
35	CS	Chip select for serial interface
36	RSSI	Output of the RSSI amplifier
37	CDEM	Capacitor to adjust the lower cut-off frequency data filter
38	RX_TX2	GND pin to decouple LNA in TX mode
39	RX_TX1	Switch pin to decouple LNA in TX mode
40	PWR_ON	Input to switch on the system (active high)
41	T1	Key input 1 (can also be used to switch on the system (active low)
42	T2	Key input 2 (can also be used to switch on the system (active low)
43	T3	Key input 3 (can also be used to switch on the system (active low)
44	T4	Key input 4 (can also be used to switch on the system (active low)
45	T5	Key input 5 (can also be used to switch on the system (active low)
46	RX_ACTIVE	Indicates RX operation mode
47	NC	Not connected
48	NC	Not connected
	GND	Ground/backplane

Figure 3. Block Diagram



Typical Key Fob or Sensor Application with 1 Battery

Figure 4. Typical RKE Key Fob or Sensor Application, 433.92 MHz, 1 Battery

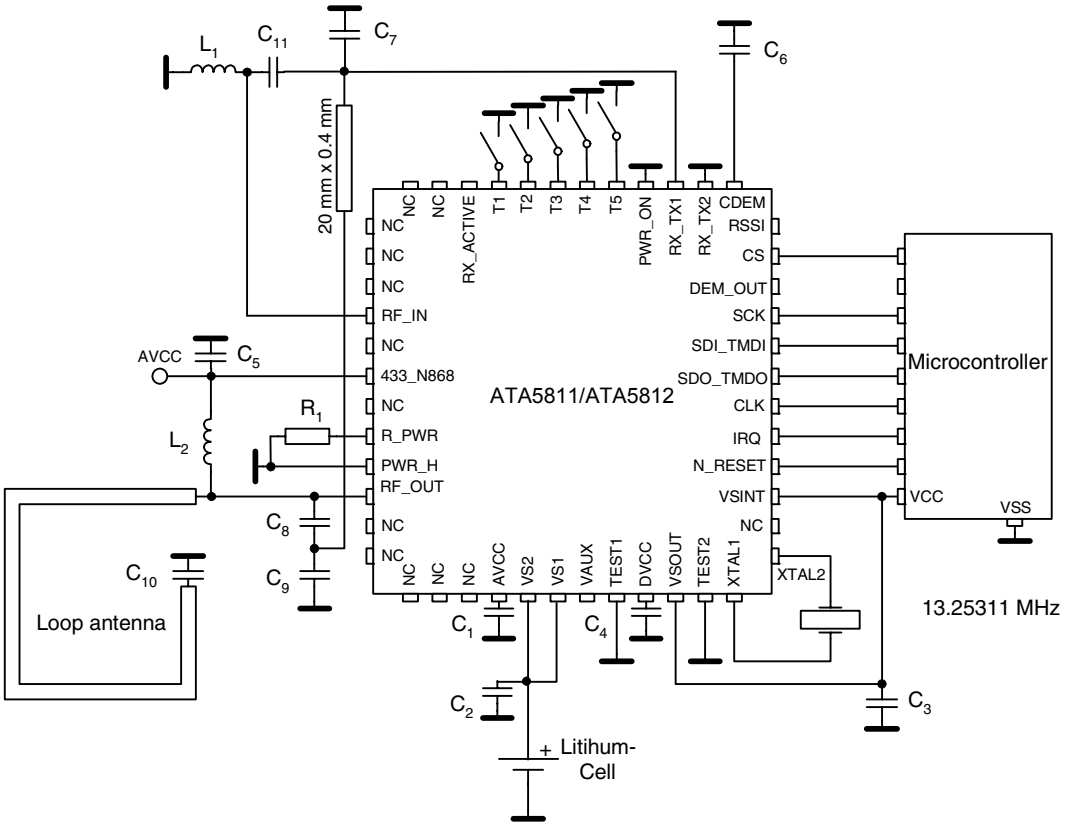


Figure 4 shows a typical 433.92 MHz RKE key fob or sensor application with one battery The external components are 11 capacitors, 1 resistor, 2 inductors and a crystal. C_1 to C_4 are 68 nF voltage supply blocking capacitors. C_5 is a 10 nF supply blocking capacitor. C_6 is a 15 nF fixed capacitor used for the internal quasi peak detector and for the highpass frequency of the data filter. C_7 to C_{11} are RF matching capacitors in the range of 1 pF to 33 pF. L1 is a matching inductor of about 5.6 nH to 56 nH. L_2 is a feed inductor of about 120 nH. A load capacitor of 9 pF for the crystal is integrated. R_1 is typically 22 kΩ and sets the output power to about 5.5 dBm. The loop antenna's quality factor is somewhat reduced by this application due to the quality factor of L_2 and the RX/TX switch. On the other hand, this lower quality factor is necessary to have a robust design with a bandwidth that is broad enough for production tolerances. Due to the single-ended and ground-referenced design, the loop antenna can be a free-form wire around the application as it is usually employed in RKE uni-directional systems. The ATA5811/ATA5812 provides sufficient isolation and robust pulling behavior of internal circuits from the supply voltage as well as an integrated VCO inductor to allow this. Since the efficiency of a loop antenna is proportional to the square of the surrounded area it is beneficial to have a large loop around the application board with a lower quality factor to relax the tolerance specification of the RF components and to get a high antenna efficiency in spite of their lower quality factor.

Typical Car or Sensor Base-station Application

Figure 5. Typical RKE Car or Sensor Base-station Application, 433.92 MHz

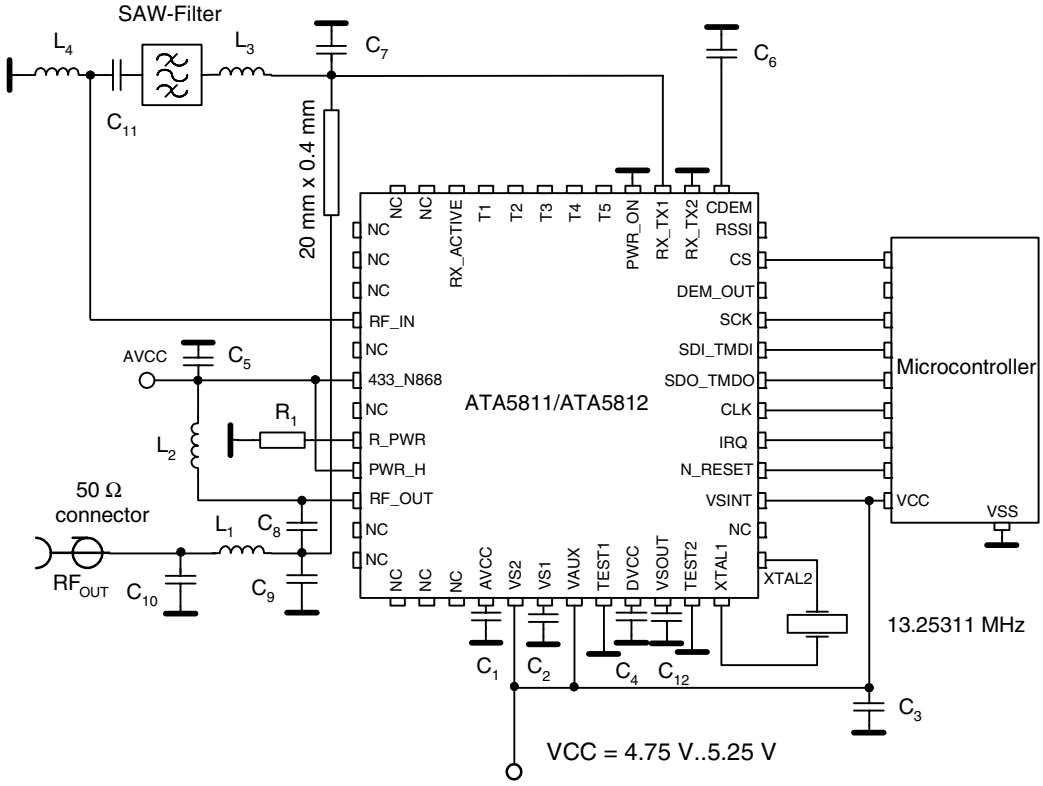


Figure 5 shows a typical 433.92 MHz V_{CC} = 4.75 V to 5.25 V RKE car or sensor base-station application. The external components are 12 capacitors, 1 resistor, 4 inductors, a SAW filter and a crystal. C_1 and C_3 to C_4 are 68 nF voltage supply blocking capacitors. C_2 and C_{12} are 2.2 µF supply blocking capacitors for the internal voltage regulators. C_5 is a 10 nF supply blocking capacitor. C_6 is a 15 nF fixed capacitor used for the internal quasi peak detector and for the highpass frequency of the data filter. C_7 to C_{11} are RF matching capacitors in the range of 1 pF to 33 pF. L_2 to L_4 are matching inductors of about 5.6 nH to 56 nH. A load capacitor for the crystal of 9 pF is integrated. R_1 is typically 22 kΩ and sets the output power at RF_OUT to about 10 dBm. Since a quarter wave or PCB antenna, which has high efficiency and wide band operation, is typically used here, it is recommended to use a SAW filter to achieve high sensitivity in case of powerful out-of-band blockers. L_1, C_{10} and C_9 together form a lowpass filter, which is needed to filter out the harmonics in the transmitted signal to meet regulations. An internally regulated voltage at pin VSOUT can be used in case the microcontroller only supports 3.3 V operation, a blocking capacitor with a value of C_{12} = 2.2 µF has to be connected to VSOUT in any case.

Typical Key Fob Application, 2 Batteries

Figure 6. Typical RKE Key Fob Application, 433.92 MHz, 2 Batteries

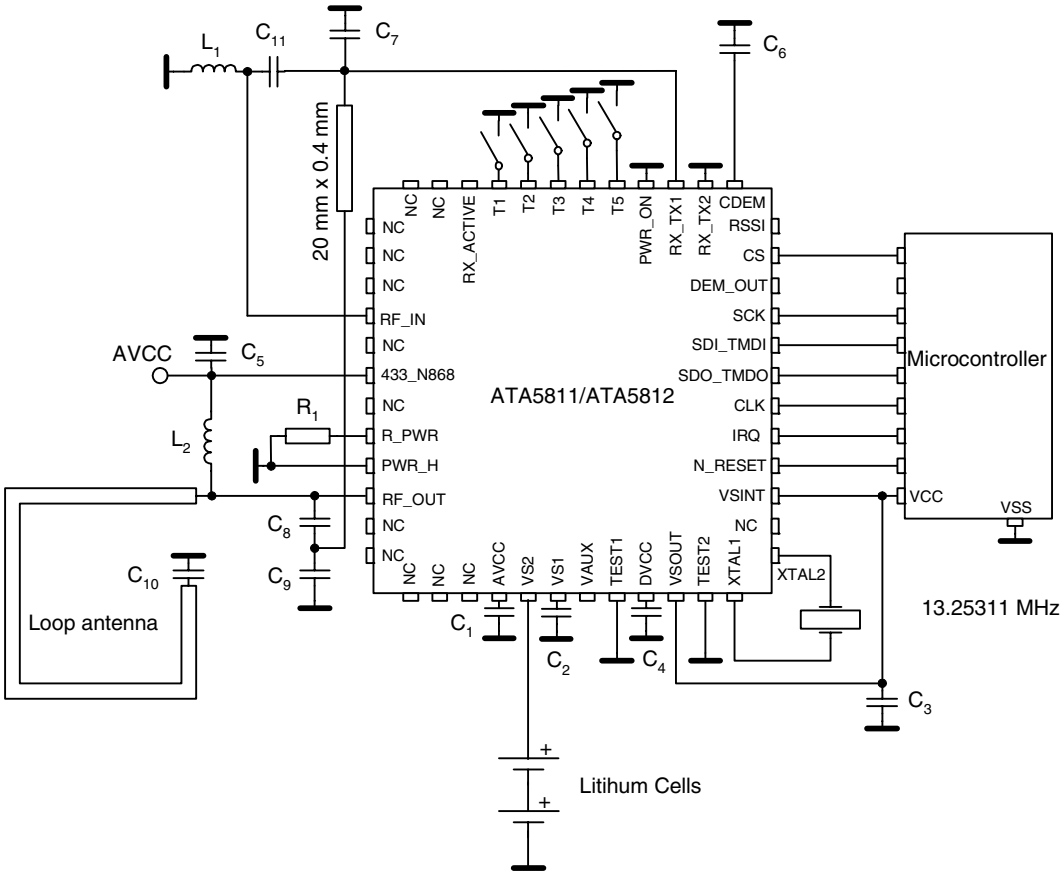


Figure 6 shows a typical 433.92 MHz 2-battery RKE key fob or sensor application. The external components are 11 capacitors, 1 resistor, 2 inductors and a crystal. C_1 and C_4 are 68 nF voltage supply blocking capacitors. C_2 and C_3 are 2.2 µF supply blocking capacitors for the internal voltage regulators. C_5 is a 10 nF supply blocking capacitor. C_6 is a 15 nF fixed capacitor used for the internal quasi peak detector and for the highpass frequency of the data filter. C_7 to C_{11} are RF matching capacitors in the range of 1 pF to 33 pF. L_1 is a matching inductor of about 5.6 nH to 56 nH. L_2 is a feed inductor of about 120 nH. A load capacitor for the crystal of 9 pF is integrated. R_1 is typically 22 kΩ and sets the output power to about 5.5 dBm.

RF Transceiver

According to Figure 3 on page 5, the RF transceiver consists of an LNA (Low-Noise Amplifier), PA (Power Amplifier), RX/TX switch, fractional-N frequency synthesizer and the signal processing part with mixer, IF filter, IF amplifier, FSK/ASK demodulator, data filter and data slicer.

In receive mode the LNA pre-amplifies the received signal which is converted down to 226 kHz, filtered and amplified before it is fed into an FSK/ASK demodulator, data filter and data slicer. The RSSI (Received Signal Strength Indicator) signal and the raw digital output signal of the demodulator are available at the pins RSSI and DEM_OUT. The demodulated data signal Demod_Out is fed to the digital control logic where it is evaluated and buffered as described in section “Digital Control Logic”.

In transmit mode the fractional-N frequency synthesizer generates the TX frequency which is fed to the PA. In ASK mode the PA is modulated by the signal PA_Enable. In FSK mode the PA is enabled and the signal TX_DATA (FSK) modulates the fractional-N frequency synthesizer. The frequency deviation is digitally controlled and internally fixed to about ±16 kHz (see Table 12 on page 24 for exact values). The transmit data can also be buffered as described in section “Digital Control Logic”. A lock detector within the synthesizer ensures that the transmission will only start if the synthesizer is locked.

The RX/TX switch can be used to combine the LNA input and the PA output to a single antenna with a minimum of losses.

Transparent modes without buffering of RX and TX data are also available to allow protocols and coding schemes other than the internal supported Manchester encoding.

Low-IF Receiver

The receive path consists of a fully integrated low-IF receiver. It fulfills the sensitivity, blocking, selectivity, supply voltage and supply current specification needed to manufacture an automotive key fob without the use of SAW blocking filter (see Figure 4 on page 6). The receiver can be connected to the roof antenna in the car when using an additional blocking SAW front-end filter as shown in Figure 5 on page 7.

At 433.92 MHz the receiver has a typical system noise figure of 7.0 dB, a system I1dBCP of -30 dBm and a system IIP3 of -20 dBm. There is no AGC or switching of the LNA needed, thus, a better blocking performance is achieved. This receiver uses an IF (Intermediate Frequency) of 226 kHz, the typical image rejection is 30 dB and the typical 3 dB IF filter bandwidth is 185 kHz (f_{IF} = 226 kHz ±92.5 kHz, f_{lo_IF} = 133.5 kHz and f_{hi_IF} = 318.5 kHz). The demodulator needs a signal to Gaussian noise ratio of 8 dB for 20 kBaud Manchester with ±16 kHz frequency deviation in FSK mode, thus, the resulting sensitivity at 433.92 MHz is typically -106 dBm at 20 kBaud Manchester.

Due to the low phase noise and spurious of the synthesizer in receive mode[1] together with the eighth order integrated IF filter the receiver has a better selectivity and blocking performance than more complex double superhet receivers but without external components and without numerous spurious receiving frequencies.

A low-IF architecture is also less sensitive to second-order intermodulation (IIP2) than direct conversion receivers where every pulse or AM-modulated signal (especially the signals from TDMA systems like GSM) demodulates to the receiving signal band at second-order non-linearities.

Note: 1. -120 dBC/Hz at ±1 MHz and -75 dBC at ±FREF at 433.92 MHz

Input Matching at RF_IN

The measured input impedances as well as the values of a parallel equivalent circuit of these impedances can be seen in Table 1. The highest sensitivity is achieved with power matching of these impedances to the source impedance of 50 Ω

Table 1. Measured Input Impedances of the RF_IN Pin

f_{RF}/MHz	Z(RF_IN)	R_p//C_p
315	(44-j233)Ω	1278 Ω//2.1 pF
433.92	(32-j169)Ω	925 Ω//2.1 pF
868.3	(21-j78)Ω	311 Ω//2.2 pF

The matching of the LNA Input to 50 Ω was done with the circuit according to Figure 7 and with the values given in Table 2. The reflection coefficients were always ≤10 dB. Note that value changes of C_1 and L_1 may be necessary to for compensate individual board layouts. The measured typical FSK and ASK Manchester code sensitivities with a Bit Error Rate (BER) of 10^{-3} are shown in Table 3 on page 11 and Table 4 on page 11. These measurements were done with inductors having a quality factor according to Table 2, resulting in estimated matching losses of 1.0 dB at 315 MHz, 1.2 dB at 433.92 MHz and 0.6 dB at 868.3 MHz. These losses can be estimated when calculating the parallel equivalent resistance of the inductor with $R_{loss} = 2 \times \pi \times f \times L \times Q_L$ and the matching loss with $10 \log(1+R_p/R_{loss})$.

With an ideal inductor, for example, the sensitivity at 433.92 MHz/FSK/20 kBaud/ ±16 kHz/Manchester can be improved from -106 dBm to -107.2 dBm. The sensitivity depends on the control logic which examines the incoming data stream. The examination limits must be programmed in control registers 5 and 6. The measurements in Table 3 on page 11 and Table 4 on page 11 are based on the values of registers 5 and 6 according to Table 39 on page 57.

Figure 7. Input Matching to 50 Ω

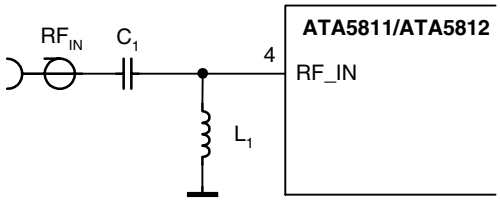


Table 2. Input Matching to 50 Ω

f_{RF}/MHz	C_1/pF	L_1/nH	Q_{L1}
315	2.2	56	43
433.92	1.8	27	40
868.3	1.2	6.8	58

Table 3. Measured Sensitivity FSK, ±16 kHz, Manchester, dBm, BER = 10^{-3}

RF Frequency	BR_Range_0 1.0 kBaud	BR_Range_0 2.4 kBaud	BR_Range_1 5.0 kBaud	BR_Range_2 10 kBaud	BR_Range_3 20 kBaud
315 MHz	-110.0 dBm	-110.5 dBm	-109.0 dBm	-108.0 dBm	-107.0 dBm
433.92 MHz	-109.0 dBm	-109.5 dBm	-108.0 dBm	-107.0 dBm	-106.0 dBm
868.3 MHz	-106.0 dBm	-106.5 dBm	-105.5 dBm	-104.0 dBm	-103.5 dBm

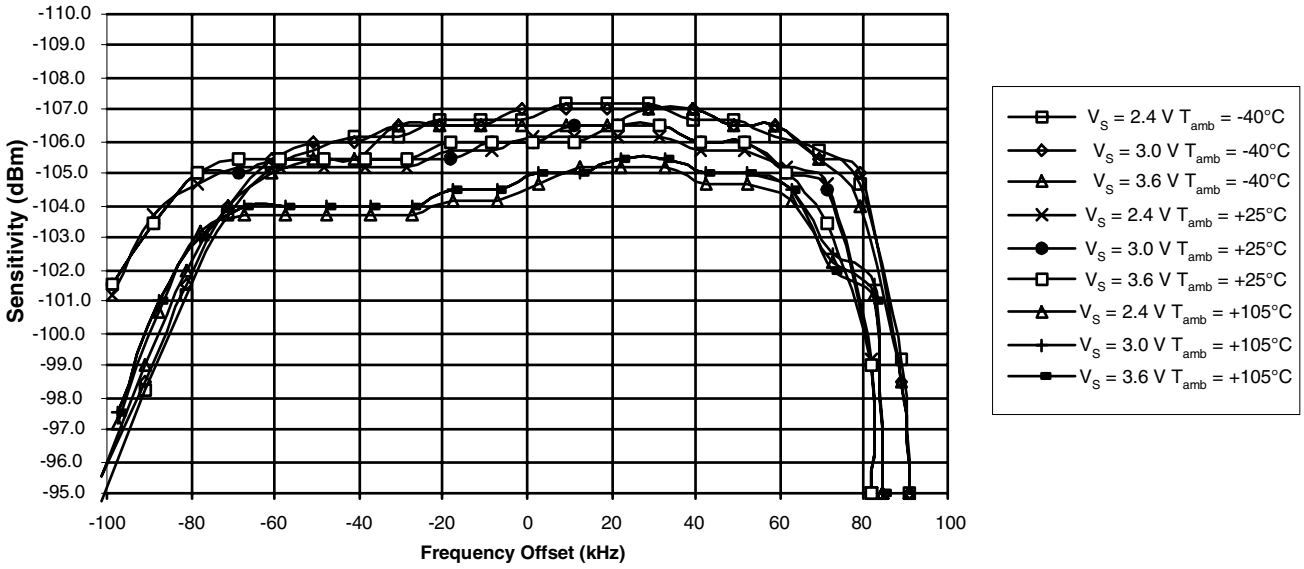
Table 4. Measured Sensitivity 100% ASK, Manchester, dBm, BER = 10^{-3}

RF Frequency	BR_Range_0 1.0 kBaud	BR_Range_0 2.4 kBaud	BR_Range_1 5.0 kBaud	BR_Range_2 10 kBaud
315 MHz	-117.0 dBm	-117.5 dBm	-115 dBm	-113.5 dBm
433.92 MHz	-116.0 dBm	-116.5 dBm	-114.0 dBm	-112.5 dBm
868.3 MHz	-112.5 dBm	-113.0 dBm	-111.5 dBm	-109.5 dBm

Sensitivity versus Supply Voltage, Temperature and Frequency Offset

To calculate the behavior of a transmission system it is important to know the reduction of the sensitivity due to several influences. The most important are frequency offset due to crystal oscillator (XTO) and crystal frequency (XTAL) errors, temperature and supply voltage dependency of the noise figure and IF filter bandwidth of the receiver. Figure 8 shows the typical sensitivity at 433.92 MHz/FSK/20kBaud/±16 kHz/Manchester versus the frequency offset between transmitter and receiver with T_{amb} = -40°C, +25°C and +105°C and supply voltage VS1 = VS2 = 2.4 V, 3.0 V and 3.6 V.

Figure 8. Measured Sensitivity 433.92 MHz/FSK/20 kBaud/±16 kHz/Manchester versus Frequency Offset, Temperature and Supply Voltage



As can be seen in Figure 8 on page 11 the supply voltage has almost no influence. The temperature has an influence of about +1.5/-0.7 dB and a frequency offset of ±65 kHz also influences by about ±1 dB. All these influences, combined with the sensitivity of a typical IC, are then within a range of -103.7 dBm and -107.3 dBm over temperature, supply voltage and frequency offset which is -105.5 dBm ±1.8dB. The integrated IF filter has an additional production tolerance of only ±7 kHz, hence, a frequency offset between the receiver and the transmitter of ±58 kHz can be accepted for XTAL and XTO tolerances.

Note: For the demodulator used in the ATA5811/ATA5812, the tolerable frequency offset does not change with the data frequency, hence, the value of ±58 kHz is valid for up to 1 kBaud.

This small sensitivity spread over supply voltage, frequency offset and temperature is very unusual in such a receiver. It is achieved by an internal, very fast and automatic frequency correction in the FSK demodulator after the IF filter, which leads to a higher system margin. This frequency correction tracks the input frequency very quickly, if however, the input frequency makes a larger step (e.g., if the system changes between different communication partners), the receiver has to be restarted. This can be done by switching back to Idle mode and then again to RX mode. For that purpose, an automatic mode is also available. This automatic mote switches to Idle mode and back into RX mode every time a bit error occurs (see section “Digital Control Logic”).

Frequency Accuracy of the Crystals

The XTO is an amplitude regulated Pierce oscillator with integrated load capacitors. The initial tolerances (due to the frequency tolerance of the XTAL, the integrated capacitors on XTAL1, XTAL2 and the XTO’s initial transconductance gm) can be compensated to a value within ±0.5 ppm by measuring the CLK output frequency and programming the control registers 2 and 3 (see Table 20 on page 35 and Table 23 on page 36). The XTO then has a remaining influence of less than ±2 ppm over temperature and supply voltage due to the bandgap controlled gm of the XTO.

The needed frequency stability of the used crystals over temperature and aging is hence ±58 kHz/433.92 MHz - 2 × ±2.5 ppm = ±128.66 ppm for 433.92 MHz and ±58 kHz/868.3 MHz - 2 × ±2.5 ppm = ±61.8 ppm for 868.3 MHz. Thus, the used crystals in receiver and transmitter each need to be better than ±64.33 ppm for 433.92 MHz and ±30.9 ppm for 868.3 MHz. In access control systems it may be advantageous to have a more tight tolerance at the base-station in order to relax the requirement for the key fob.

RX Supply Current versus Temperature and Supply Voltage

Table 5 shows the typical supply current at 433.92 MHz of the transceiver in RX mode versus supply voltage and temperature with VS = VS1 = VS2. As you can see the supply current at 2.4 V and -40°C is less than the typical one which helps because this is also the operation point where a lithium cell has the worst performance. The typical supply current at 315 MHz or 868.3 MHz in RX mode is about the same as for 433.92 MHz.

Table 5. Measured 433.92 MHz Receive Supply Current in FSK Mode

VS =	2.4 V	3.0 V	3.6 V
T_{amb} = -40°C	8.4 mA	8.8 mA	9.2 mA
T_{amb} = 25°C	9.9 mA	10.3 mA	10.8 mA
T_{amb} = 105°C	11.4 mA	11.9 mA	12.4 mA

Blocking, Selectivity

As can be seen in Figure 9 on page 13 and Figure 10 on page 13, the receiver can receive signals 3 dB higher than the sensitivity level in presence of very large blockers of -47 dBm/-34 dBm with small frequency offsets of ±1/ ±10 MHz.

Figure 9 shows narrow band blocking and Figure 10 wide band blocking characteristics. The measurements were done with a useful signal of 433.92 MHz/FSK/ 20 kBaud/±16 kHz/Manchester with a level of -106 dBm + 3 dB = -103 dBm which is 3 dB above the sensitivity level. The figures show how much a continuous wave signal can be larger than -103 dBm until the BER is higher than 10^{-3}. The measurements were done at the 50 Ω input according to Figure 7 on page 10. At 1 MHz, for example, the blocker can be 56 dB higher than -103 dBm which is -103 dBm + 56 dB = -47 dBm. These values, together with the good intermodulation performance, avoid the need for a SAW filter in the key fob application.

Figure 9. Narrow Band 3 dB Blocking Characteristic at 433.92 MHz

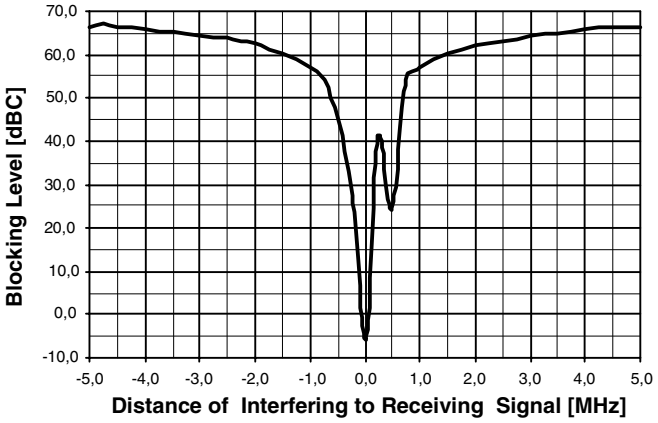


Figure 10. Wide Band 3 dB Blocking Characteristic at 433.92 MHz

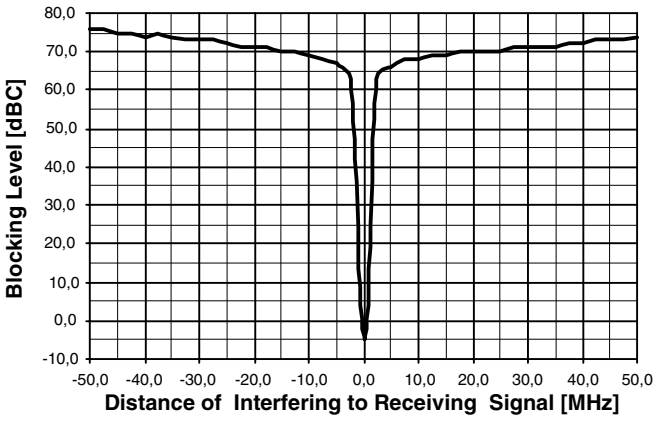


Figure 11 on page 14 shows the blocking measurement close to the received frequency to illustrate the selectivity and image rejection. This measurement was done 6 dB above the sensitivity level with a useful signal of 433.92 MHz/FSK/20kBaud/±16 kHz/ Manchester with a level of -106 dBm + 6 dB = -100 dBm. The figure shows to which extent a continuous wave signal can surpass -100 dBm until the BER is higher than 10^{-3}. For example, at 1 MHz the blocker can than be 59 dB higher than -100 dBm which is -100 dBm + 59 dB = -41 dBm.

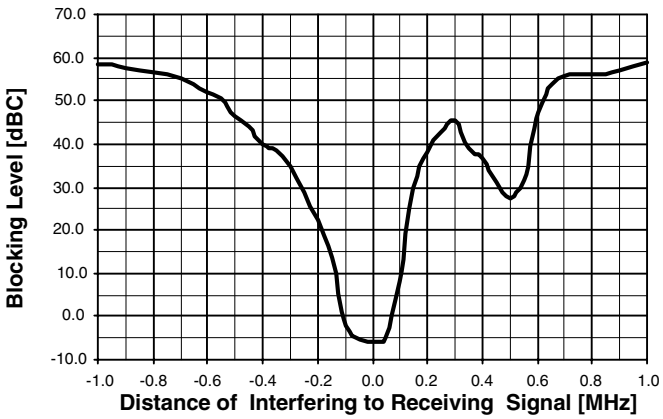
Table 6 shows the blocking performance measured relative to -100 dBm for some other frequencies. Note that sometimes the blocking is measured relative to the sensitivity level (dBS) instead of the carrier (dBC).

Table 6. Blocking 6 dB Above Sensitivity Level with BER < 10^{-3}

Frequency Offset	Blocker Level	Blocking
+0.75 MHz	-45 dBm	55 dBC/61 dBS
-0.75 MHz	-45 dBm	55 dBC/61 dBS
+1.5 MHz	-38 dBm	62 dBC/68 dBS
-1.5 MHz	-38 dBm	62 dBC/68 dBS
+10 MHz	-30 dBm	70 dBC/76 dBS
-10 MHz	-30 dBm	70 dBC/76 dBS

The ATA5811/ATA5812 can also receive FSK and ASK modulated signals if they are much higher than the I1dBCP. It can typically receive useful signals at 10 dBm. This is often referred to as the nonlinear dynamic range which is the maximum to minimum receiving signal which is 116 dB for 20 kBaud Manchester. This value is useful if two transceivers have to communicate and are very close to each other.

Figure 11. Close In 6 dB Blocking Characteristic and Image Response at 433.92 MHz



This high blocking performance makes it even possible for some applications using quarter wave whip antennas to use a simple LC band-pass filter instead of a SAW filter in the receiver. When designing such an LC filter take into account that the 3 dB blocking at 433.92 MHz/2 = 216.96 MHz is 43 dBC and at 433.92 MHz/3 = 144.64 MHz is 48 dBC and at 2 × (433.92 MHz + 226 kHz) + -226 kHz = 868.066 MHz/868.518 MHz is 56 dBC. And especially that at 3 × (433.92 MHz + 226 kHz)+226 kHz = 1302.664 MHz the receiver has its second LO harmonic receiving frequency with only 12 dBC blocking.

Inband Disturbers, Data Filter, Quasi Peak Detector, Data Slicer

If a disturbing signal falls into the received band or a blocker is not continuous wave the performance of a receiver strongly depends on the circuits after the IF filter. Hence the demodulator, data filter and data slicer are important in that case.

The data filter of the ATA5811/ATA5812 implies a quasi peak detector. This results in a good suppression of the above mentioned disturbers and exhibits a good carrier to Gaussian noise performance. The required useful signal to disturbing signal ratio to be received with a BER of 10^{-3} is less than 12 dB in ASK mode and less than 3 dB (BR_Range_0 ... BR_Range_2)/6 dB (BR_Range_3) in FSK mode. Due to the many different waveforms possible these numbers are measured for signal as well as for disturbers with peak amplitude values. Note that these values are worst case values and are valid for any type of modulation and modulating frequency of the disturbing signal as well as the receiving signal. For many combinations, lower carrier to disturbing signal ratios are needed.

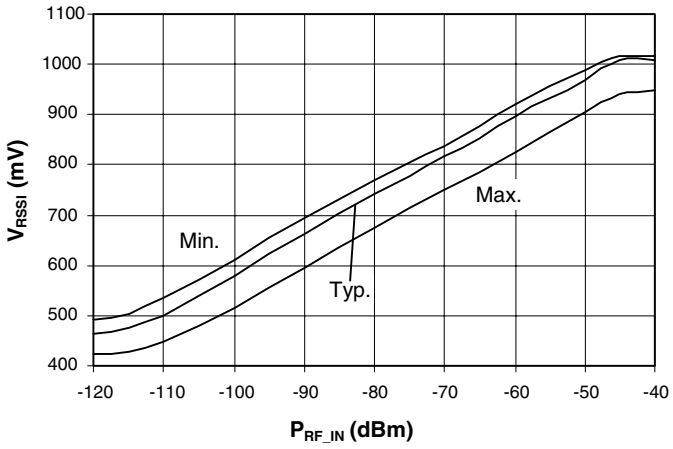
DEM_OUT Output

The internal raw output signal of the demodulator Demod_Out is available at pin DEM_OUT. DEM_OUT is an open drain output and must be connected to a pull-up resistor if it is used (typically 100 kΩ) otherwise no signal is present at that pin.

RSSI Output

The output voltage of the pin RSSI is an analog voltage, proportional to the input power level. Using the RSSI output signal, the signal strength of different transmitters can be distinguished. The usable dynamic range of the RSSI amplifier is 70 dB, the input power range $P(RF_{IN})$ is -115 dBm to -45 dBm and the gain is 8 mV/dB. Figure 12 shows the RSSI characteristic of a typical device at 433.92 MHz with VS1 = VS2 = 2, 4 V to 3, 6 V and T_{amb} = -40°C to +105°C with a matched input according to Table 2 on page 10 and Figure 7 on page 10. At 868.3 MHz about 2.7 dB more signal level and at 315 MHz about 1 dB less signal level is needed for the same RSSI results.

Figure 12. Typical RSSI Characteristic versus Temperature and Supply Voltage



Frequency Synthesizer

The synthesizer is a fully integrated fractional-N design with internal loop filters for receive and transmit mode. The XTO frequency f_{XTO} is the reference frequency FREF for the synthesizer. The bits FR0 to FR8 in control registers 2 and 3 (see Table 20 on page 35 and Table 23 on page 36) are used to adjust the deviation of f_{XTO}. In transmit mode, at 433.92 MHz, the carrier has a phase noise of -111 dBC/Hz at 1 MHz and spurious at FREF of -66 dBC with a high PLL loop bandwidth allowing the direct modulation of the carrier with 20 kBaud Manchester data. Due to the closed loop modulation any spurious caused by this modulation are effectively filtered out as can be seen in Figure 15 on page 17. In RX mode the synthesizer has a phase noise of -120 dBC/Hz at 1 MHz and spurious of -75 dBC.

The initial tolerances of the crystal oscillator due to crystal tolerances, internal capacitor tolerances and the parasitics of the board have to be compensated at manufacturing setup with control registers 2 and 3 as can be seen in Table 12 on page 24. The other control words for the synthesizer needed for ASK, FSK and receive/transmit switching are calculated internally. The RF (Radio Frequency) resolution is equal to the XTO frequency divided by 16384 which is 777.1 Hz at 315.0 MHz, 808.9 Hz at 433.92 MHz and 818.59 Hz at 868.3 MHz.

FSK/ASK Transmission

Due to the fast modulation capability of the synthesizer and the high resolution, the carrier can be internally FSK modulated which simplifies the application of the transceiver. The deviation of the transmitted signal is ±20 digital frequency steps of the synthesizer which is equal to ±15.54 kHz for 315 MHz, ±16.17 kHz for 433.92 MHz and ±16.37 kHz for 868.3 MHz.

Due to closed loop modulation with PLL filtering the modulated spectrum is very clean, meeting ETSI and CEPT regulations when using a simple LC filter for the power amplifier harmonics as it is shown in Figure 5 on page 7. In ASK mode the frequency is internally connected to the center of the FSK transmission and the power amplifier is switched on and off to perform the modulation. Figure 13 to Figure 15 on page 17 show the spectrum of the FSK modulation with pseudo random data with 20 kBaud/±16.17 kHz/Manchester and 5 dBm output power.

Figure 13. FSK-modulated TX Spectrum (20 kBaud/±16.17 kHz/Manchester Code)

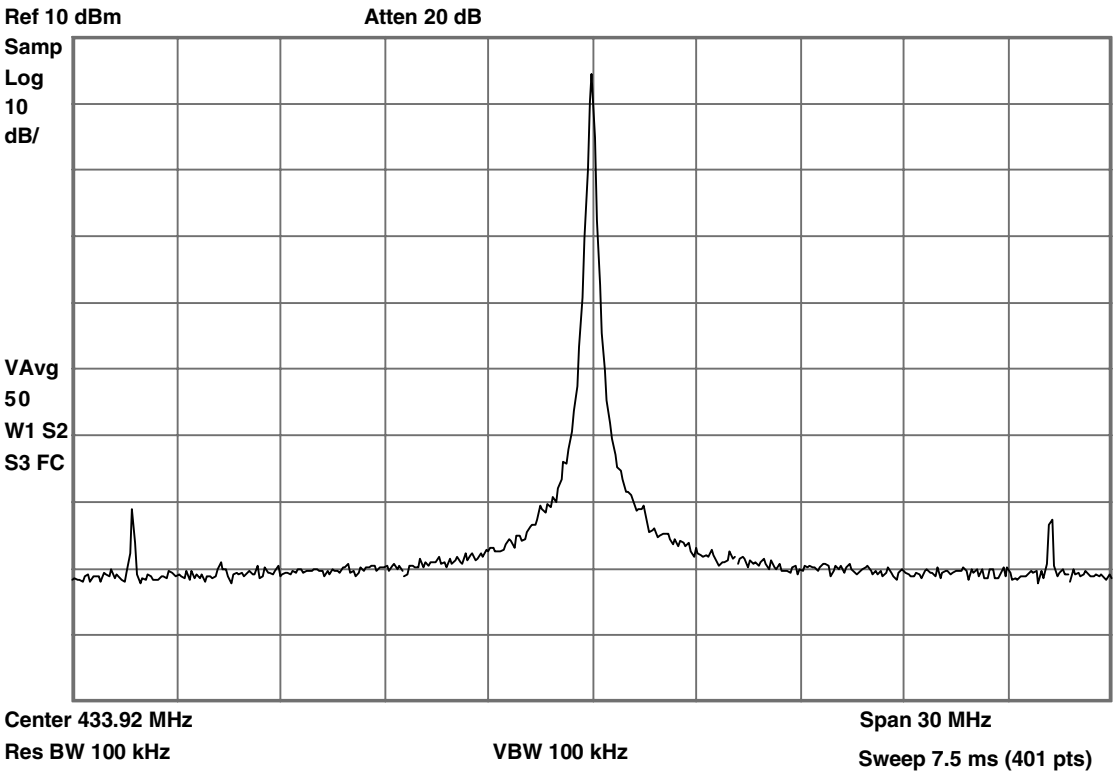


Figure 14. Unmodulated TX Spectrum f_{FSK_L}

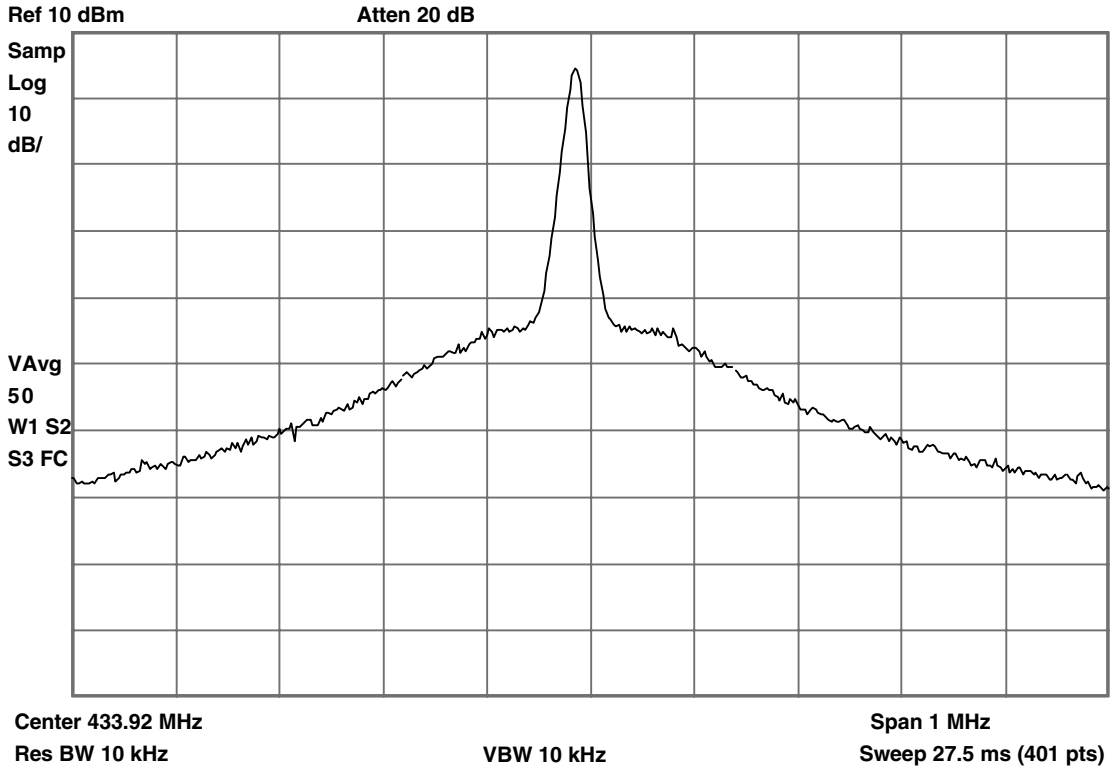
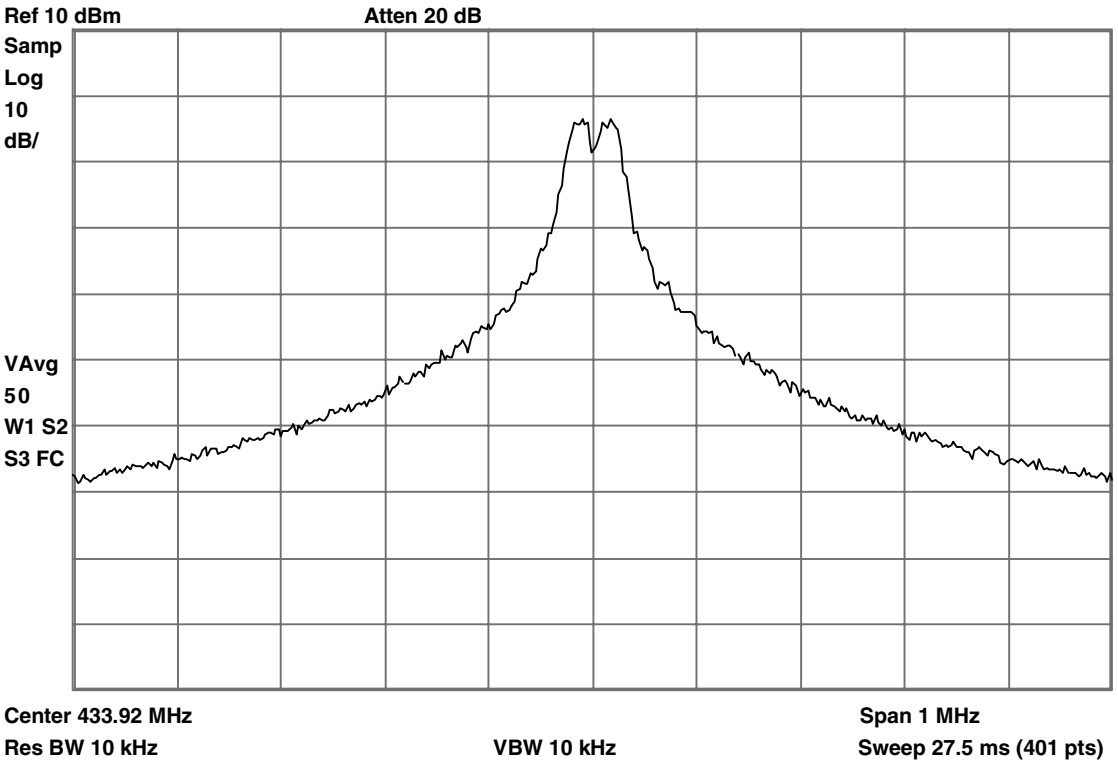


Figure 15. FSK-modulated TX Spectrum (20 kBaud/±16.17 kHz/Manchester Code)



Output Power Setting and PA Matching at RF_OUT

The Power Amplifier (PA) is a single-ended open collector stage which delivers a current pulse which is nearly independent of supply voltage, temperature and tolerances due to bandgap stabilization. Resistor R_1, see Figure 16 on page 19, sets a reference current which controls the current in the PA. A higher resistor value results in a lower reference current, a lower output power and a lower current consumption of the PA. The usable range of R_1 is 15 kΩ to 56 kΩ. Pin PWR_H switches the output power range between about 0 dBm to 5 dBm (PWR_H = GND) and 5 dBm to 10 dBm (PWR_H = AVCC) by multiplying this reference current with a factor 1 (PWR_H = GND) and 2.5 (PWR_H = AVCC) which corresponds to about 5 dB more output power.

If the PA is switched off in TX mode, the current consumption without output stage with VS1 = VS2 = 3 V, T_{amb} = 25°C is typically 6.5 mA for 868.3 MHz and 6.95 mA for 315 MHz and 433.92 MHz.

The maximum output power is achieved with optimum load resistances R_{Lopt} according to Table 7 on page 19 with compensation of the 1.0 pF output capacitance of the RF_OUT pin by absorbing it into the matching network consisting of L_1, C_1, C_3 as shown in Figure 16 on page 19. There must be also a low resistive DC path to AVCC to deliver the DC current of the power amplifier's last stage. The matching of the PA output was done with the circuit according to Figure 16 on page 19 with the values in Table 7 on page 19. Note that value changes of these elements may be necessary to compensate for individual board layouts.

Example:
According to Table 7 on page 19, with a frequency of 433.92 MHz and output power of 11 dBm the overall current consumption is typically 17.8 mA hence the PA needs 17.8 mA - 6.95 mA = 10.85 mA in this mode which corresponds to an overall power amplifier efficiency of the PA of $(10^{(11dBm/10)} \times 1\ mW)/(3\ V \times 10.85\ mA) \times 100\% = 38.6\%$ in this case.

Using a higher resistor in this example of $R_1 = 1.091 \times 22\ k\Omega = 24\ k\Omega$ results in 9.1% less current in the PA of 10.85 mA/1.091 = 9.95 mA and $10 \times \log(1.091) = 0.38$ dB less output power if using a new load resistance of $300\ \Omega \times 1.091 = 327\ \Omega$. The resulting output power is then 11 dBm - 0.38 dB = 10.6 dBm and the overall current consumption is 6.95 mA + 9.95 mA = 16.9 mA.

The values of Table 7 on page 19 were measured with standard multi-layer chip inductors with quality factors Q according to Table 7 on page 19. Looking to the 433.92 MHz/11 dBm case with the quality factor of $Q_{L1} = 43$ the loss in this inductor is estimated with the parallel equivalent resistance of the inductor $R_{loss} = 2 \times \pi \times f \times L \times Q_{L1}$ and the matching loss with 10 log $(1 + R_{Lopt}/R_{loss})$ which is equal to 0.32 dB losses in this inductor. Taking this into account the PA efficiency is then 42% instead of 38.6%.

Be aware that the high power mode (PWR_H = AVCC) can only be used with a supply voltage higher than 2.7 V, whereas the low power mode (PWR_H = GND) can be used down to 2.4 V as can be seen in the section “Electrical Characteristics”.

The supply blocking capacitor C_2 (10 nF) has to be placed close to the matching network because of the RF current flowing through it.

Figure 16. Power Setting and Output Matching

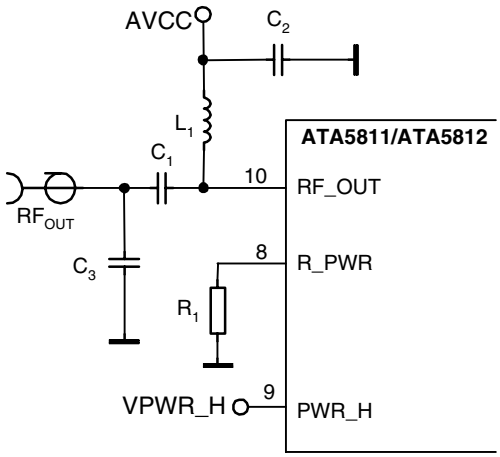


Table 7. Measured Output Power and Current Consumption with VS1 = VS2 = 3 V, T_{amb} = 25°C

Frequency (MHz)	TX Current (mA)	Output Power (dBm)	R1 (kΩ)	VPWR_H	R_{Lopt} (Ω)	L1 (nH)	Q_{L1}	C1 (pF)	C3 (pF)
315	8.5	0.4	56	GND	2500	82	28	1.5	0
315	10.5	5.7	27	GND	920	68	32	2.2	0
315	16.7	10.5	27	AVCC	350	56	35	3.9	0
433.92	8.6	0.1	56	GND	2300	56	40	0.75	0
433.92	11.2	6.2	22	GND	890	47	38	1.5	0
433.92	17.8	11	22	AVCC	300	33	43	2.7	0
868.3	9.3	-0.3	33	GND	1170	12	58	1.0	3.3
868.3	11.5	5.4	15	GND	471	15	54	1.0	0
868.3	16.3	9.5	22	AVCC	245	10	57	1.5	0

Output Power and TX Supply Current versus Supply Voltage and Temperature

Table 8 on page 20 shows the measurement of the output power for a typical device with VS1 = VS2 = VS in the 433.92 MHz and 6.2 dBm case versus temperature and supply voltage measured according to Figure 16 on page 19 with components according to Table 7. As opposed to the receiver sensitivity the supply voltage has here the major impact on output power variations because of the large signal behavior of a power amplifier. Thus, a two battery system with voltage regulator or a 5 V system shows much less variation than a 2.4 V to 3.6 V one battery system because the supply voltage is then well within 3.0 V and 3.6 V.

The reason is that the amplitude at the output RF_OUT with optimum load resistance is AVCC - 0.4 V and the power is proportional to (AVCC - 0.4 V)^2 if the load impedance is not changed. This means that the theoretical output power reduction if reducing the supply voltage from 3.0 V to 2.4 V is 10 log ((3 V - 0.4 V)^2/(2.4 V - 0.4 V)^2) = 2.2 dB. Table 8 on page 20 shows that principle behavior in the measurement. This is not the same case for higher voltages since here increasing the supply voltage from 3 V to 3.6 V should theoretical increase the power by 1.8 dB but only 0.8 dB in the measurement shows that the amplitude does not increase with the supply voltage because the load impedance is optimized for 3 V and the output amplitude stays more constant.

Table 8. Measured Output Power and Supply Current at 433.92 MHz, PWR_H = GND

VS =	2.4 V	3.0 V	3.6 V
T_{amb} = -40°C	10.19 mA 3.8 dBm	10.19 mA 5.5 dBm	10.78 mA 6.2 dBm
T_{amb} = +25°C	10.62 mA 4.6 dBm	11.19 mA 6.2 dBm	11.79 mA 7.1 dBm
T_{amb} = +105°C	11.4 mA 3.8 dBm	12.02 mA 5.4 dBm	12.73 mA 6.3 dBm

Table 9 shows the relative changes of the output power of a typical device compared to 3.0 V/25°C. As can be seen a temperature change to -40° as well as to +105° reduces the power by less than 1 dB due to the bandgap regulated output current. Measurements of all the cases in Table 7 on page 19 over temperature and supply voltage have shown about the same relative behavior as shown in Table 9

Table 9. Measurements of Typical Output Power Relative to 3 V/25°

VS =	2.4 V	3.0 V	3.6 V
T_{amb} = -40°C	-2.4 dB	-0.7 dB	0 dB
T_{amb} = +25°C	-1.6 dB	0 dB	+0.9 dB
T_{amb} = +105°C	-2.4 dB	-0.8 dB	+0.1 dB

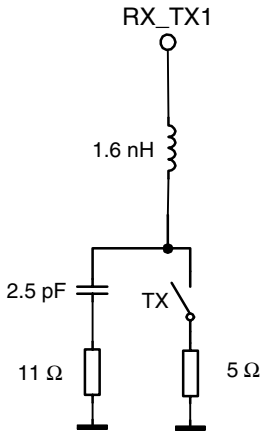
RX/TX Switch

The RX/TX switch decouples the LNA from the PA in TX mode, and directs the received power to the LNA in RX mode. To do this, it has a low impedance to GND in TX mode and a high impedance to GND in RX mode. To design a proper RX/TX decoupling a linear simulation tool for radio frequency design together with the measured device impedances of Table 1 on page 10, Table 7 on page 19, Table 10 and Table 11 on page 22 should be used, but the exact element values have to be found on board. Figure 17 on page 21 shows an approximate equivalent circuit of the switch. The principal switching operation is described here according to the application of Figure 4 on page 6. The application of Figure 5 on page 7 works similarly.

Table 10. Impedance of the RX/TX Switch RX_TX2 Shorted to GND

Frequency	Z(RX_TX1) TX Mode	Z(RX_TX1) RX Mode
315 MHz	(4.8 + j3.2) Ω	(11.3 - j214) Ω
433.92 MHz	(4.5 + j4.3) Ω	(10.3 - j153) Ω
868.3 MHz	(5 + j9) Ω	(8.9 - j73) Ω

Figure 17. Equivalent Circuit of the Switch



Matching Network in TX Mode

In TX mode the 20 mm long and 0.4 mm wide transmission line which is much shorter than λ/4 is approximately switched in parallel to the capacitor C_9 to GND. The antenna connection between C_8 and C_9 has an impedance of about 50 Ω locking from the transmission line into the loop antenna with pin RF_OUT, L_2, C_{10}, C_8 and C_9 connected (using a C_9 without the added 7.6 pF as discussed later). The transmission line can be approximated with a 16 nH inductor in series with a 1.5 Ω resistor, the closed switch can be approximated according to Table 10 on page 20 with the series connection of 1.6 nH and 5 Ω in this mode. To have a parallel resonant high impedance circuit with little RF power going into it looking from the loop antenna into the transmission line a capacitor of about 7.6 pF to GND is needed at the beginning of the transmission line (this capacitor is later absorbed into C_9 which is then higher as needed for 50 Ω transformation). To keep the 50 Ω impedance in RX mode at the end of this transmission line C_7 has to be also about 7.6 pF. This reduces the TX power by about 0.5 dB at 433.92 MHz compared to the case the where the LNA path is completely disconnected.

Matching Network in RX Mode

In RX mode the RF_OUT pin has a high impedance of about 7 kΩ in parallel with 1.0 pF at 433.92 MHz as can be seen in Table 11 on page 22. This together with the losses of the inductor L_2 with 120 nH and Q_{L2} = 25 gives about 3.7 kΩ loss impedance at RF_OUT. Since the optimum load impedance in TX mode for the power amplifier at RF_OUT is 890 Ω the loss associated with the inductor L_2 and the RF_OUT pin can be estimated to be $10 \times \log(1 + 890/3700) = 0.95$ dB compared to the optimum matched loop antenna without L_2 and RF_OUT. The switch represents, in this mode at 433.92 MHz, about an inductor of 1.6 nH in series with the parallel connection of 2.5 pF and 2.0 kΩ. Since the impedance level at pin RX_TX1 in RX mode is about 50 Ω this only negligiblably dampens the received signal by about 0.1 dB. When matching the LNA to the loop antenna the transmission line and the 7.6 pF part of C_9 has to be taken into account when choosing the values of C_{11} and L_1 so that the impedance seen from the loop antenna into the transmission line with the 7.6 pF capacitor connected is 50 Ω. Since the loop antenna in RX mode is loaded by the LNA input impedance the loaded Q of the loop antenna is lowered by about a factor of 2 in RX mode hence the antenna bandwidth is higher than in TX mode.

Table 11. Impedance RF_OUT Pin in RX Mode

Frequency	Z(RF_OUT)RX	$R_P//C_P$
315 MHz	36 Ω –j 502 Ω	7 kΩ / / 1.0 pF
433.92 MHz	19 Ω –j 366 Ω	7 kΩ / / 1.0 pF
868.3 MHz	2.8 Ω –j 141Ω	7 kΩ / / 1.3 pF

Note that if matching to 50 Ω, like in Figure 5 on page 7, a high Q wire wound inductor with a Q > 70 should be used for L_2 to minimize its contribution to RX losses which will otherwise be dominant. The RX and TX losses will be in the range of 1.0 dB there.

XTO

The XTO is an amplitude regulated Pierce oscillator type with integrated load capacitances (2 × 18 pF with a tolerance of ±17%) hence C_{Lmin} = 7.4 pF and C_{Lmax} = 10.6 pF. The XTO oscillation frequency f_{XTO} is the reference frequency FREF for the fractional-N synthesizer. When designing the system in terms of receiving and transmitting frequency offset the accuracy of the crystal and XTO have to be considered.

The synthesizer can adjust the local oscillator frequency for more than ±150 ppm at 433.92 MHz/315 MHz and up to ±118 ppm at 868.3 MHz of initial frequency error in f_{XTO}. This is done at nominal supply voltage and temperature with the control registers 2 and 3 (see Table 20 on page 35 and Table 23 on page 36). The remaining local oscillator tolerance at nominal supply voltage and temperature is then < ±0.5 ppm. A XTO frequency error of ±150 ppm/±118 ppm can hence be tolerated due to the crystal tolerance at 25°C and the tolerances of C_{L1} and C_{L2}. The XTO's gm has very low influence of less than ±2 ppm on the frequency at nominal supply voltage and temperature.

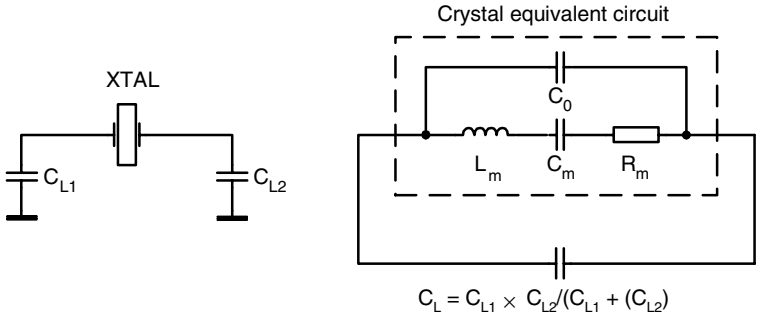
Over temperature and supply voltage, the XTO's additional pulling is only ±2 ppm if $C_m \leq 7$ fF. The XTAL versus temperature and its aging is then the main source of frequency error in the local oscillator.

The XTO frequency depends on XTAL properties and the load capacitances $C_{L1, 2}$ at pin XTAL1 and XTAL2. The pulling of f_{XTO} from the nominal f_{XTAL} is calculated using the following formula:

$$P = \frac{C_m}{2} \times \frac{C_{LN} - C_L}{(C_0 + C_{LN}) \times (C_0 + C_L)} \times 10^6 \text{ ppm.}$$

C_m is the crystal's motional, C_0 the shunt and C_{LN} the nominal load capacitance of the XTAL found in its data sheet. C_L is the total actual load capacitance of the crystal in the circuit and consists of C_{L1} and C_{L2} in series connection.

Figure 18. XTAL with Load Capacitance



With $C_m \leq 14$ fF, $C_0 \geq 1.5$ pF, $C_{LN} = 9$ pF and $C_L = 7.6$ pF to 10.6 pF the pulling amounts to P ≤±100 ppm and with $C_m \leq 7$ fF, $C_0 \geq 1.5$ pF, $C_{LN} = 9$ pF and $C_L = 7.4$ pF to 10.6 pF the pulling is P ≤±50 ppm.

Since typical crystals have less than ±50 ppm tolerance at 25° the compensation is not critical.

C_0 of the XTAL has to be lower than $C_{Lmin}/2 = 3.8$ pF for a Pierce oscillator type in order to not enter the steep region of pulling versus load capacitance where there is a risk of an unstable oscillation.

To ensure proper start-up behavior the small signal gain and thus the negative resistance provided by this XTO at start is very large, for example oscillation starts up even in worst case with a crystal series resistance of 1.5 kΩ at $C_0 \leq 2.2$ pF with this XTO. The negative resistance is approximately given by

$$\mathrm{Re}\{Z_{xtocore}\} = \mathrm{Re}\left\{\frac{Z_1 \times Z_3 + Z_2 \times Z_3 + Z_1 \times Z_2 \times Z_3 \times g_m}{Z_1 + Z_2 + Z_3 + Z_1 \times Z_2 \times g_m}\right\}$$

with Z_1, Z_2 as complex impedances at pin XTAL1 and XTAL2 hence
$Z1 = -j/(2 \times \pi \times f_{XTO} \times C_{L1}) + 5\ \Omega$ and $Z2 = -j/(2 \times \pi \times f_{XTO} \times C_{L2}) + 5\ \Omega$

Z_3 consists of crystals C_0 in parallel with an internal 110 kΩ resistor hence
$Z3 = -j/(2 \times \pi \times f_{XTO} \times C_0)$ /110 kΩ, gm is the internal transconductance between XTAL1 and XTAL2 with typically 19 ms at 25°C.

With $f_{XTO} = 13.5$ MHz, gm = 19 ms, CL = 9 pF, $C_0 = 2.2$ pF this results in a negative resistance of about 2 kΩ. The worst case for technological, supply voltage and temperature variations is then for $C_0 \leq 2.2$ pF always higher than 1.5 kΩ.

Due to the large gain at start the XTO is able to meet a very low start-up time. The oscillation start-up time can be estimated with the time constant τ.

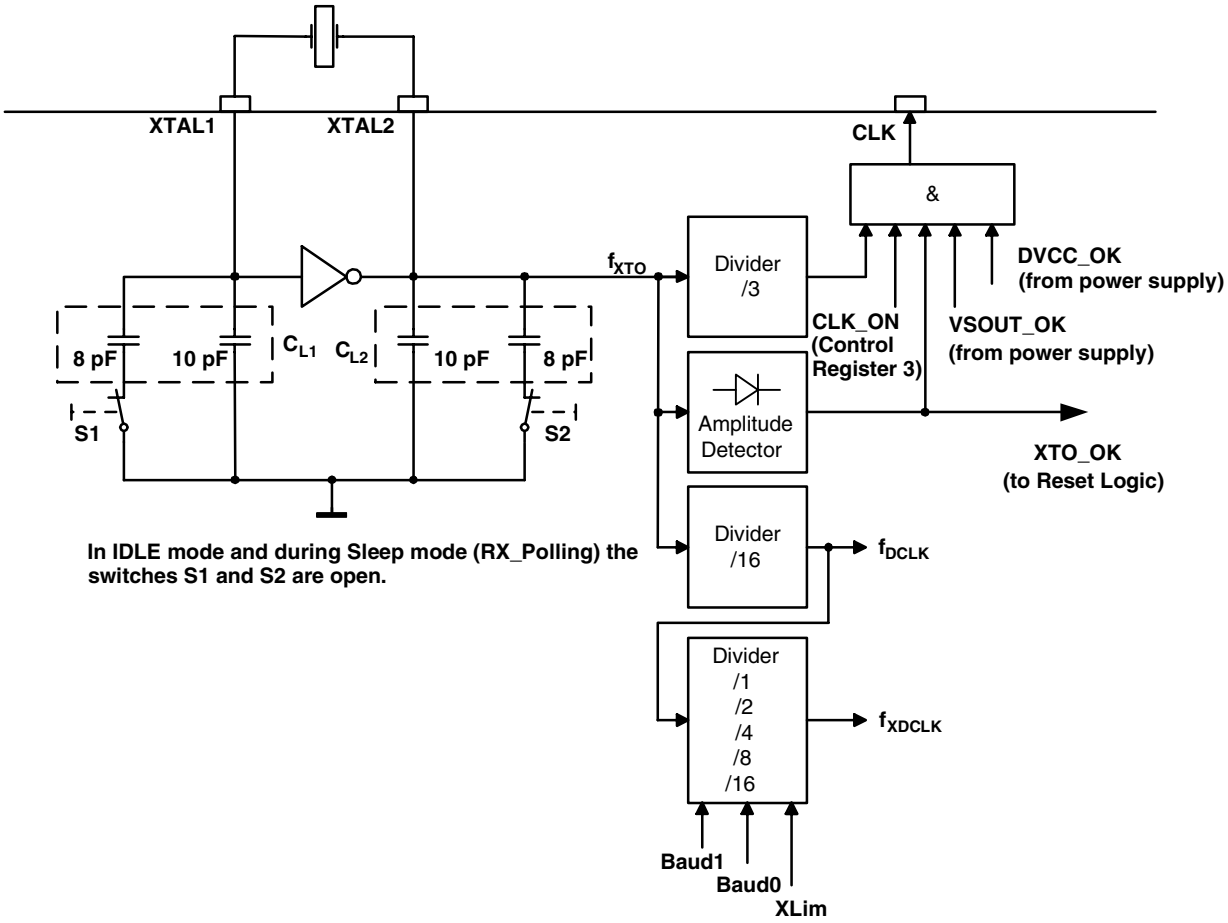
$$\tau = \frac{2}{4 \times \pi^2 \times f_m^2 \times C_m \times (\mathrm{Re}(Z_{xtocore}) + R_m)}$$

After 10 τ to 20 τ an amplitude detector detects the oscillation amplitude and sets XTO_OK to High if the amplitude is large enough, this sets N_RESET to High and activates the CLK output if CLK_ON in control register 3 is High (see Table 20 on page 35). Note that the necessary conditions of the VSOUT and DVCC voltage also have to be fulfilled (see Figure 19 on page 24 and Figure 21 on page 26).

To save current in Idle and sleep mode, the load capacitors partially are switched off in this modes with S1 and S2 seen in Figure 19 on page 24.

It is recommended to use a crystal with $C_m = 4.0$ fF to 7.0 fF, $C_{LN} = 9$ pF, $R_m < 120\ \Omega$ and $C_0 = 1.5$ pF to 2.2 pF.

Figure 19. XTO Block Diagram



To find the right values used in the control registers 2 and 3 (see Table 20 on page 35 and Table 23 on page 36) the relationship between f_{XTO} and the f_{RF} is shown in Table 12. To determine the right content the frequency at pin CLK as well as the output frequency at RF_OUT in ASK mode can be measured, than the FREQ value can be calculated according to Table 12 so that f_{RF} is exactly the desired radio frequency

Table 12. Calculation of f_{RF}

Frequency (MHz)	Pin 6 433_N868	CREG1 Bit(4) FS	f_{XTO} (MHz)	$f_{RF} = f_{TX_ASK} = f_{RX}$	$f_{TX_FSK_L}$	$f_{TX_FSK_H}$
433.92	AVCC	0	13.25311	$f_{XTO} \times \left(32,\ 5 + \frac{FREQ + 20{,}5}{16384}\right)$	f_{RF} - 16.17 kHz	f_{RF} + 16.17 kHz
868.3	GND	0	13.41191	$f_{XTO} \times \left(64,\ 5 + \frac{FREQ + 20{,}5}{16384}\right)$	f_{RF} - 16.37 kHz	f_{RF} + 16.37 kHz
315.0	AVCC	1	12.73193	$f_{XTO} \times \left(24,\ 5 + \frac{FREQ + 20{,}5}{16384}\right)$	f_{RF} - 15.54 kHz	f_{RF} + 15.54 kHz

The variable FREQ depends on FREQ2 and FREQ3, which are defined by the bits FR0 to FR8 in control register 2 and 3 and is calculated as follows:

FREQ = 3584 + FREQ2 + FREQ3

Only the range of FREQ = 3803 to 4053 of this register should be used because otherwise harmonics of f_{XTO} and f_{CLK} can cause interference with the received signals (FREQ_min = 3803, FREQ_max = 4053). The resulting tuning range is ±118 ppm at 868.3 MHz and more than ±150 ppm at 433.92 MHz or 315 MHz.

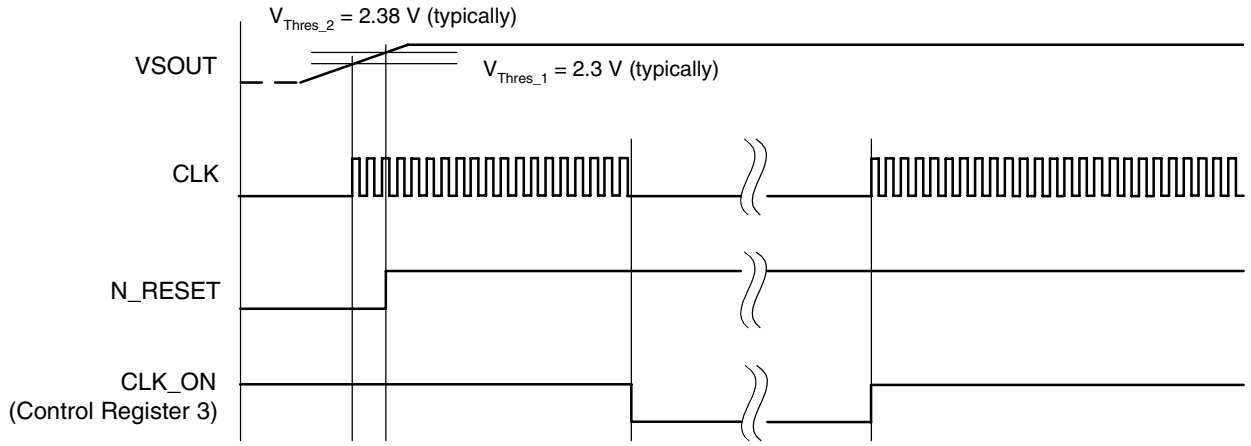
Pin CLK

Pin CLK is an output to clock a connected microcontroller. The clock frequency f_{CLK} is calculated as follows:

$$f_{CLK} = \frac{f_{XTO}}{3}$$

Because the enabling of pin CLK is asynchronous the first clock cycle may be incomplete. The signal at CLK output has a nominal 50% duty cycle.

Figure 20. Clock Timing



Basic Clock Cycle of the Digital Circuitry

The complete timing of the digital circuitry is derived from one clock. According to Figure 19 on page 24, this clock cycle T_{DCLK} is derived from the crystal oscillator (XTO) in combination with a divider.

$$f_{DCLK} = \frac{f_{XTO}}{16}$$

T_{DCLK} controls the following application relevant parameters:

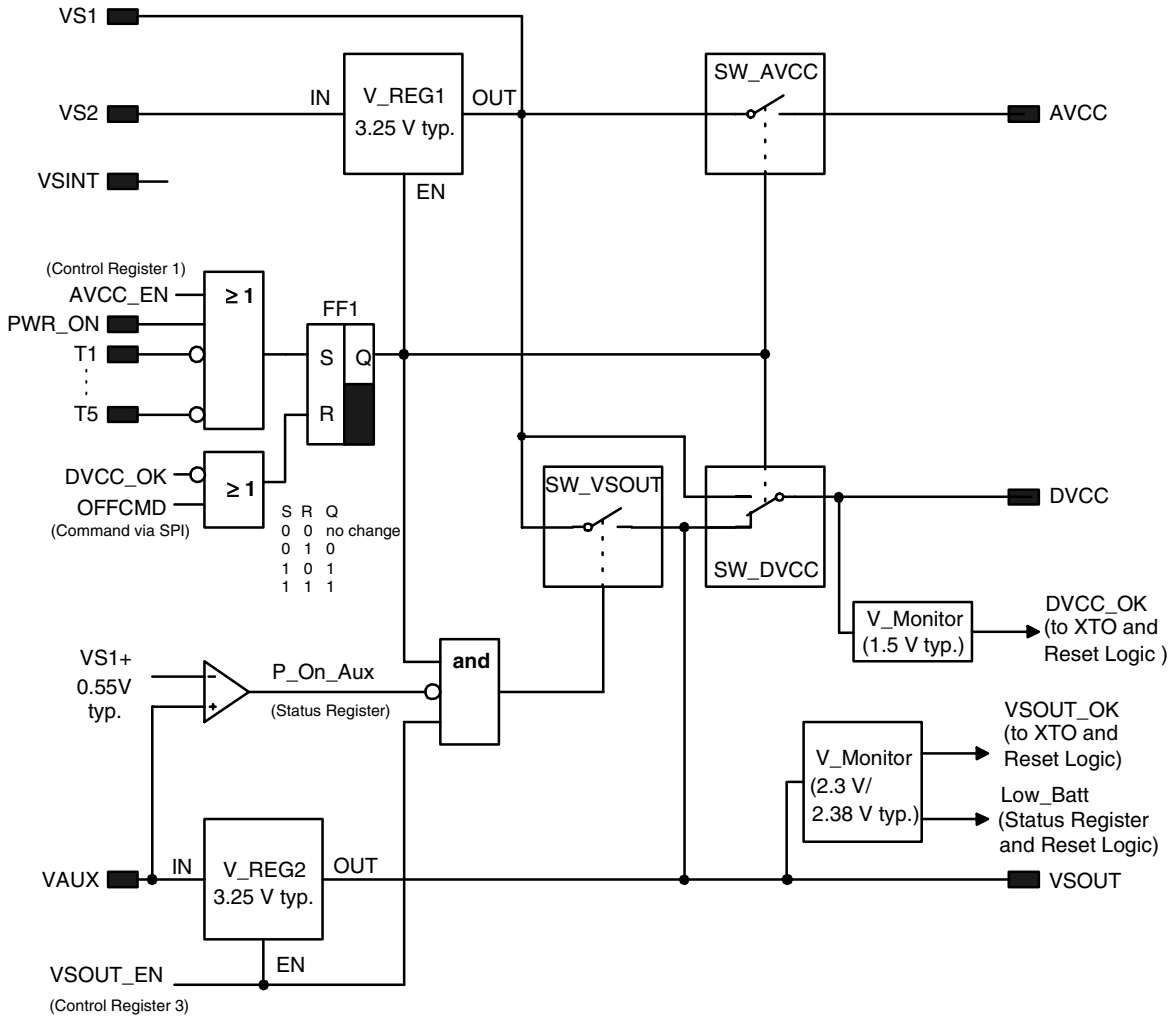
- Timing of the polling circuit including Bit-check
- TX baud rate

The clock cycle of the Bit-check and the TX baud rate depends on the selected baudrate range (BR_Range) which is defined in control register 6 (see Table 33 on page 38) and XLim which is defined in control register 4 (see Table 26 on page 36). This clock cycle T_{XDCLK} is defined by the following formulas for further reference:

BR_Range ⇒
BR_Range 0: $T_{XDCLK} = 8 \times T_{DCLK} \times XLim$
BR_Range 1: $T_{XDCLK} = 4 \times T_{DCLK} \times XLim$
BR_Range 2: $T_{XDCLK} = 2 \times T_{DCLK} \times XLim$
BR_Range 3: $T_{XDCLK} = 1 \times T_{DCLK} \times XLim$

Power Supply

Figure 21. Power Supply



The supply voltage range of the ATA5811/ATA5812 is 2.4 V to 3.6 V or 4.4 V to 6.6 V.

Pin VS1 is the supply voltage input for the range 2.4 V to 3.6 V and is used in battery applications using a single lithium 3 V cell. Pin VS2 is the voltage input for the range 4.4 V to 6.6 V (2 Battery Application and Car Applications) in this case the voltage regulator V_REG1 regulates VS1 to typically 3.25 V. If the voltage regulator is active a blocking capacitor of 2.2 µF has to be connected to VS1.

Pin VAUX is an input for an additional auxiliary voltage supply and can be connected e.g., to an inductive supply (see Figure 26 on page 32). This input can only be used together with a rectifier or as in the application of Figure 5 on page 7 and must otherwise be left open.

Pin VSINT is the voltage input for the Microcontoller_Interface and must be connected to the power supply of the microcontroller. The voltage range of V_{VSINT} is 2.4 V to 5.25 V (see Figure 25 on page 31 and Figure 26 on page 32).

AVCC is the internal operation voltage of the RF transceiver and is feed via the switch SW_AVCC by VS1. AVCC must be blocked with a 68 nF capacitor (see Figure 4 on page 6, Figure 5 on page 7 and Figure 6 on page 8).

DVCC is the internal operation voltage of the digital control logic and is feed via the switch SW_DVCC by VS1 or VSOUT. DVCC must be blocked on pin DVCC with 68 nF (see Figure 4 on page 6, Figure 5 on page 7 and Figure 6 on page 8).

Pin VSOUT is a power supply output voltage for external devices (e.g., microcontroller) and is fed via the switch SW_VSOUT by VS1 or via V_REG2 by the a auxiliary voltage supply VAUX. The voltage regulator V_REG2 regulates VSOUT to typically 3.25 V. If the voltage regulator is active a blocking capacitor of 2.2 µF has to be connected to VSOUT. VSOUT can be switched off by the VSOUT_EN bit in control register 3 and is then reactivated by conditions found in Figure 22 on page 28.

Pin N_RESET is set to low if the voltage V_{VSOUT} at pin VSOUT drops below 2.3 V (typically) and can be used as a reset signal for a connected microcontroller (see Figure 23 on page 30 and Figure 24 on page 31).

Pin PWR_ON is an input to switch on the transceiver (active high).

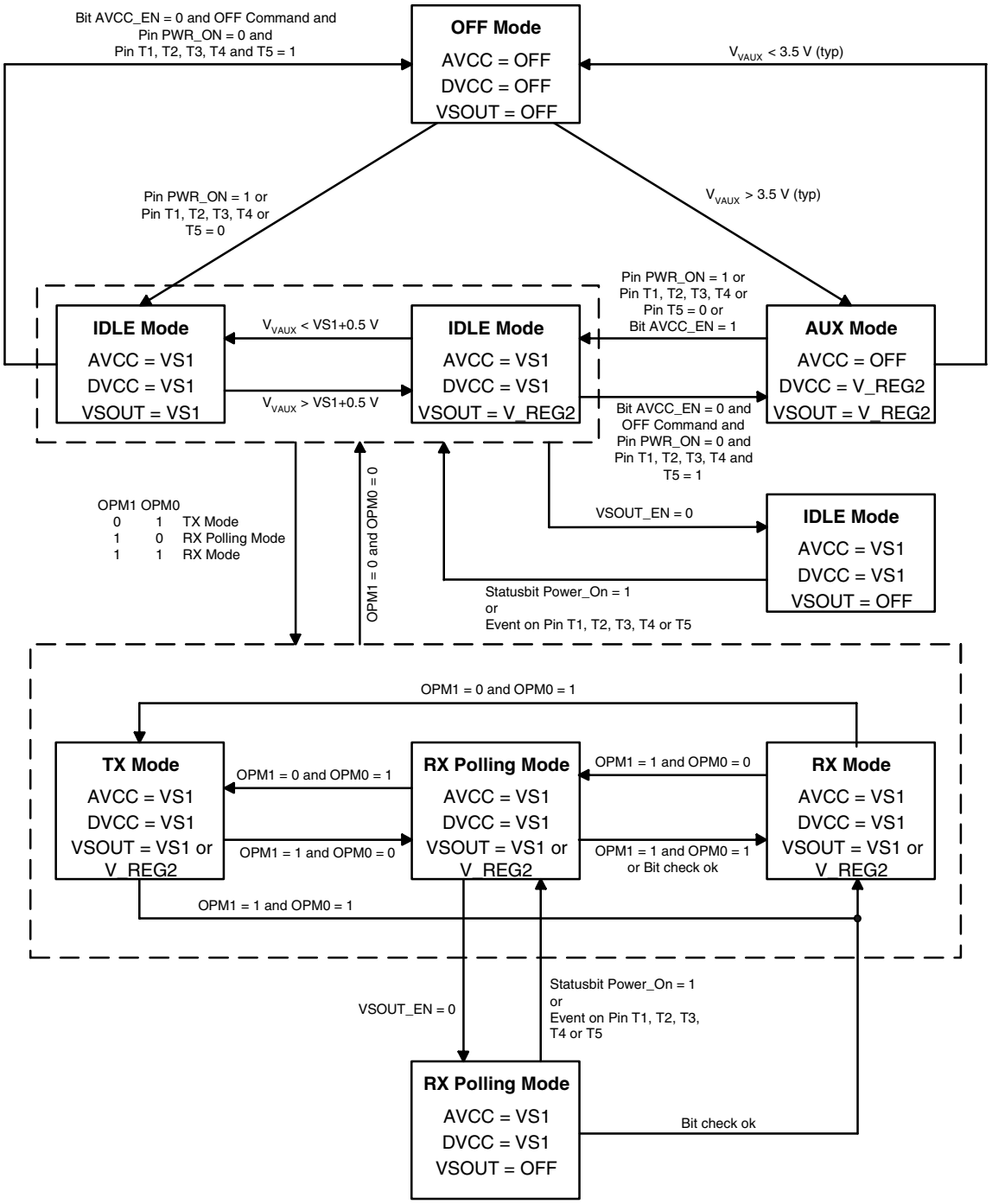
Pin T1 to T5 are inputs for push buttons and can also be used to switch on the transceiver (active low).

For current consumption reasons it is recommended to set T1 to T5 to GND or PWR_ON to VCC only temporarily. Otherwise an additional current flows.

There are two voltage monitors generating the following signals (see Figure 21 on page 26):

- DVCC_OK if DVCC > 1.5 V typically
- VSOUT_OK if VSOUT > V_{Thres1} (2.3 V typically)
- Low_Batt if VSOUT < V_{Thres2} (2.38 V typically)

Figure 22. Flow Chart Operation Modes



OFF Mode

After connecting the power supply (battery) to pin VS1 and/or VS2 and if the voltage on pin VAUX V_{VAUX} < 3.5 V (typically) the transceiver is in OFF mode. In OFF mode AVCC, DVCC and VSOUT are disabled, resulting in very low power consumption (I_{S_OFF} is typically 10 nA). In OFF mode the transceiver is not programmable via the 4-wire serial interface.

AUX Mode

The transceiver changes from OFF mode to AUX mode if the voltage at pin VAUX V_{VAUX} > 3.5 V (typically). In AUX mode DVCC and VSOUT are connected to the auxiliary power supply input (VAUX) via the voltage regulator V_REG2. In AUX mode the transceiver is programmable via the 4-wire serial interface, but no RX or TX operations are possible because AVCC = OFF.

The state transition OFF mode to AUX mode is indicated by an interrupt at pin IRQ and the status bit P_On_Aux = 1.

Idle Mode

In Idle mode AVCC and DVCC are connected to the battery voltage (VS1).

From OFF mode the transceiver changes to Idle mode if pin PWR_ON is set to 1 or pin T1, T2, T3, T4 or T5 is set to 0. This state transition is indicated by an interrupt at pin IRQ and the status bits Power_On = 1 or ST1, ST2, ST3, ST4 or ST5 = 1.

From AUX mode the transceiver changes to Idle mode by setting AVCC_EN = 1 in control register 1 via the 4-wire serial interface or if pin PWR_ON is set to 1 or pin T1, T2, T3, T4 or T5 is set to 0.

VSOUT is either connected to VS1 or to the auxiliary power supply (V_REG2).
If V_{VAUX} < VS1 + 0.5 V, VSOUT is connected to VS1. If $V_{VAUX} > V_{S1}$ + 0.5 V, VSOUT is connected to V_REG2 and the status bit P_On_Aux is set to 1.

In Idle mode the RF transceiver is disabled and the power consumption I_{S_IDLE} is about 230 µA (VSOUT OFF and CLK output OFF VS1 = VS2 = 3 V). The exact value of this current is strongly dependent on the application and the exact operation mode, therefore check the section “Electrical Characteristics” for the appropriate application case.

Via the 4-wire serial interface a connected microcontroller can program the required parameter and enable the TX, RX polling or RX mode.

The transceiver can be set back to OFF mode by an OFF command via the 4-wire serial interface (the bit AVCC_EN must be set to 0, the input level of pin PWR_ON must be 0 and pin T1, T2, T3, T4 and T5 = 1 before writing the OFF command).

Table 13. Control Register 1

OPM1	OPM0	Function
0	0	Idle mode

Reset Timing and Reset Logic

If the transceiver is switched on (OFF mode to Idle mode, OFF mode to AUX mode) DVCC and VSOUT are ramping up as illustrated in Figure 23 on page 30 (AVCC only ramps up if the transceiver is set to the Idle mode). The internal signal DVCC_RESET resets the digital control logic and sets the control register to default values.

A voltage monitor generates a low level at pin N_RESET until the voltage at pin VSOUT exceeds 2.38 V (typically) and the start-up time of the XTO has elapsed (amplitude detector, see Figure 19 on page 24). After the voltage at pin VSOUT exceeds 2.3 V (typically) and the start-up time of the XTO has elapsed the output clock at pin CLK is available. Because the enabling of pin CLK is asynchronous the first clock cycle may be incomplete.

The status bit Low_Batt is set to 1 if the voltage at pin VSOUT V_{VSOUT} drops below V_{Thres_2} (typically 2.38 V). Low_Batt is set to 0 if V_{VSOUT} exceeds V_{Thres_2} and the status register is read via the 4-wire serial interface or N_RESET is set to low.

If V_{VSOUT} drops below V_{Thres_1} (typically 2.3 V), N_RESET is set to low. If bit VSOUT_EN in control register 3 is 1, a DVCC_RESET is also generated. If V_{VSOUT} was prior disabled by the connected microcontroller by setting bit VSOUT_EN = 0, no DVCC_RESET is generated.

Note: If VSOUT < V_{Thres_1} (typically 2.3 V) the output of the pin CLK is low, the Microcontroller_Interface is disabled and the transceiver is not programmable via the 4-wire serial interface.

Figure 23. Reset Timing

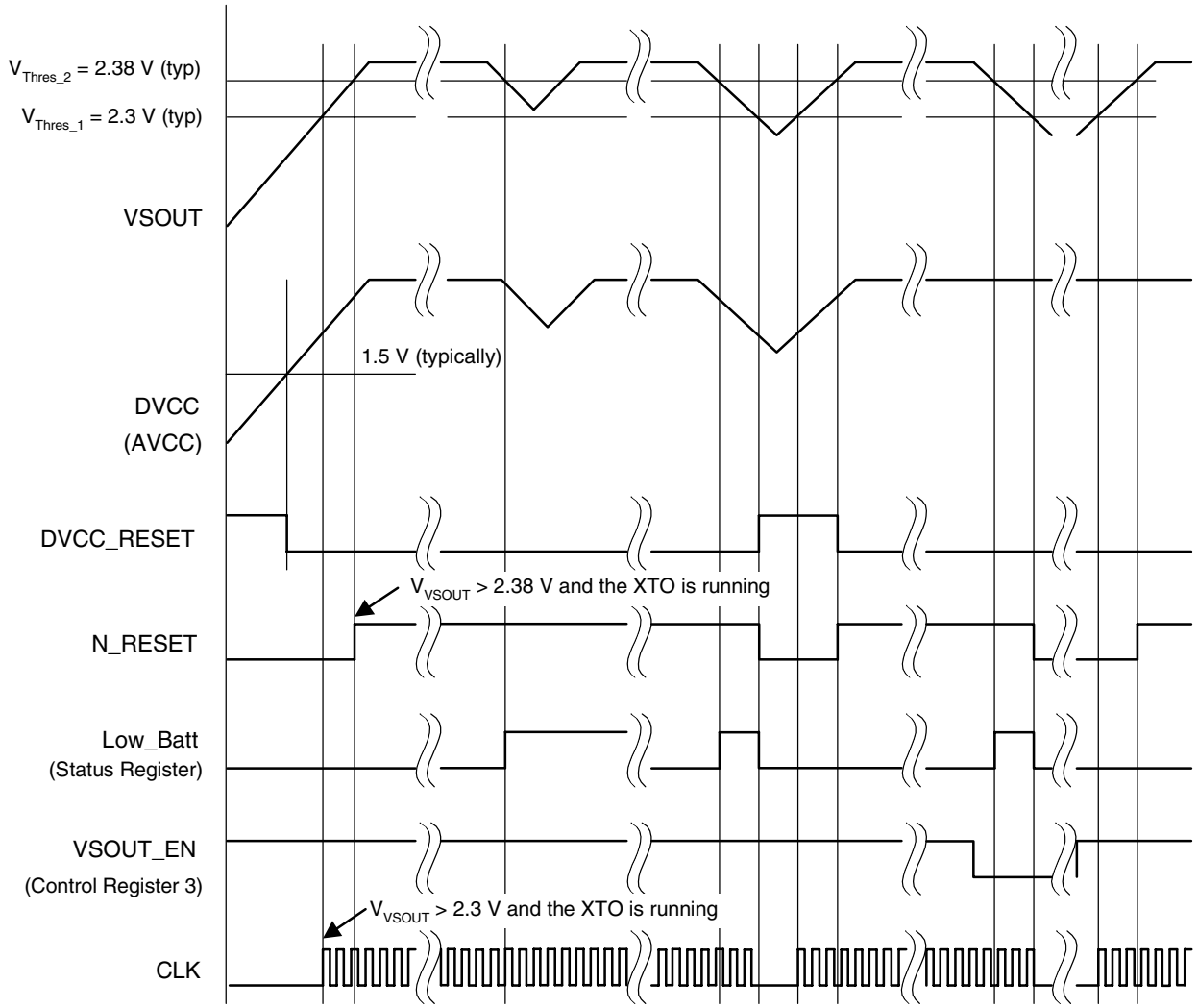
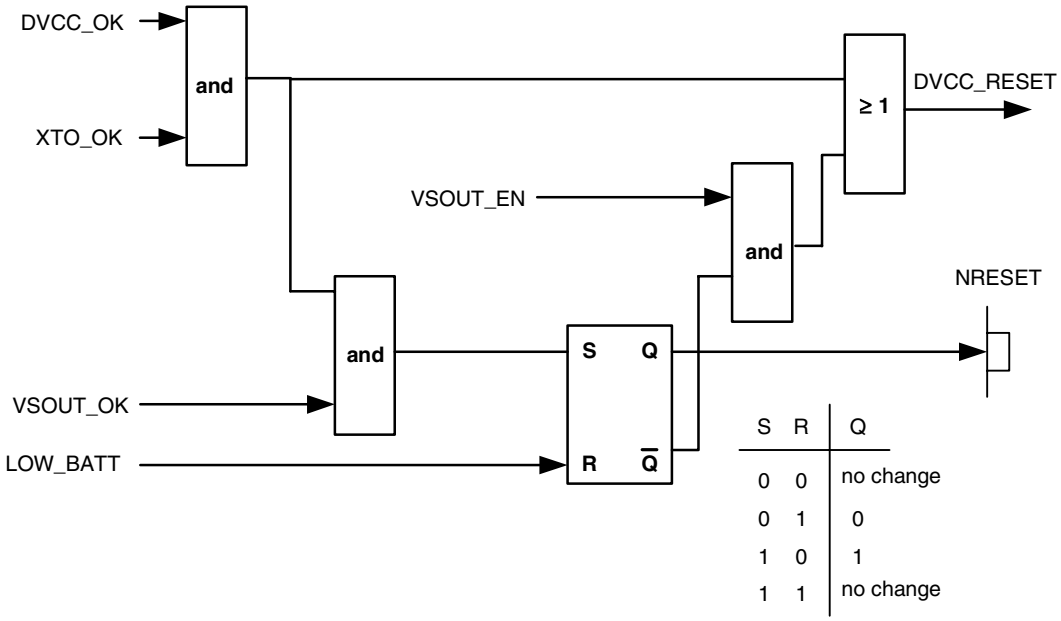


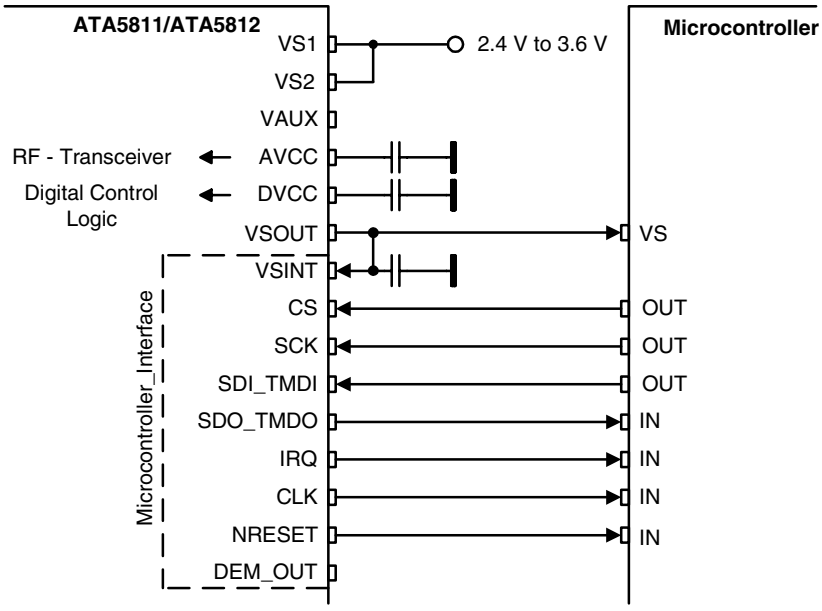
Figure 24. Reset Logic, SR Latch Generates the Hysteresis in the NRESET Signal



1-Battery Application

The supply voltage range is 2.4 V to 3.6 V and VAUX is not used.

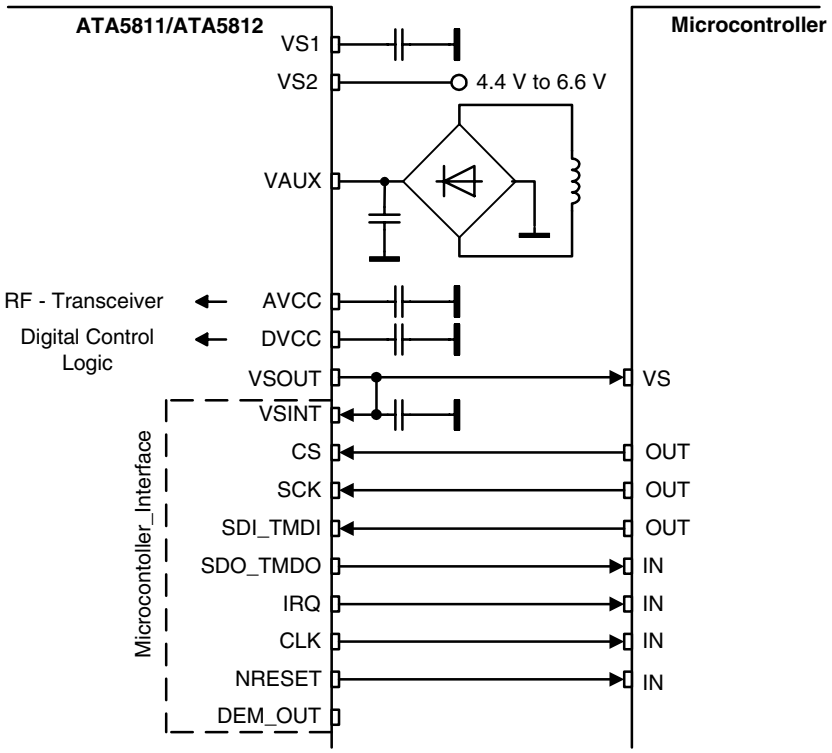
Figure 25. 1-Battery Application



2-Battery Application

The supply voltage range is 4.4 V to 6.6 V and VAUX is connected to an inductive supply.

Figure 26. 2-Battery Application with Inductive Emergency Supply



Microcontroller Interface

The microcontroller interface is a level converter which converts all internal digital signals which are referred to the DVCC voltage, into the voltage used by the microcontroller. Therefore, the pin VSINT has to be connected to the supply voltage of the microcontroller.

This makes it possible to use the internal voltage regulator/switch at pin VSOUT as in Figure 4 on page 6 and Figure 6 on page 8 or to connect the microcontroller and the pin VSINT directly to the supply voltage of the microcontroller as in Figure 5 on page 7.

Digital Control Logic

Register Structure

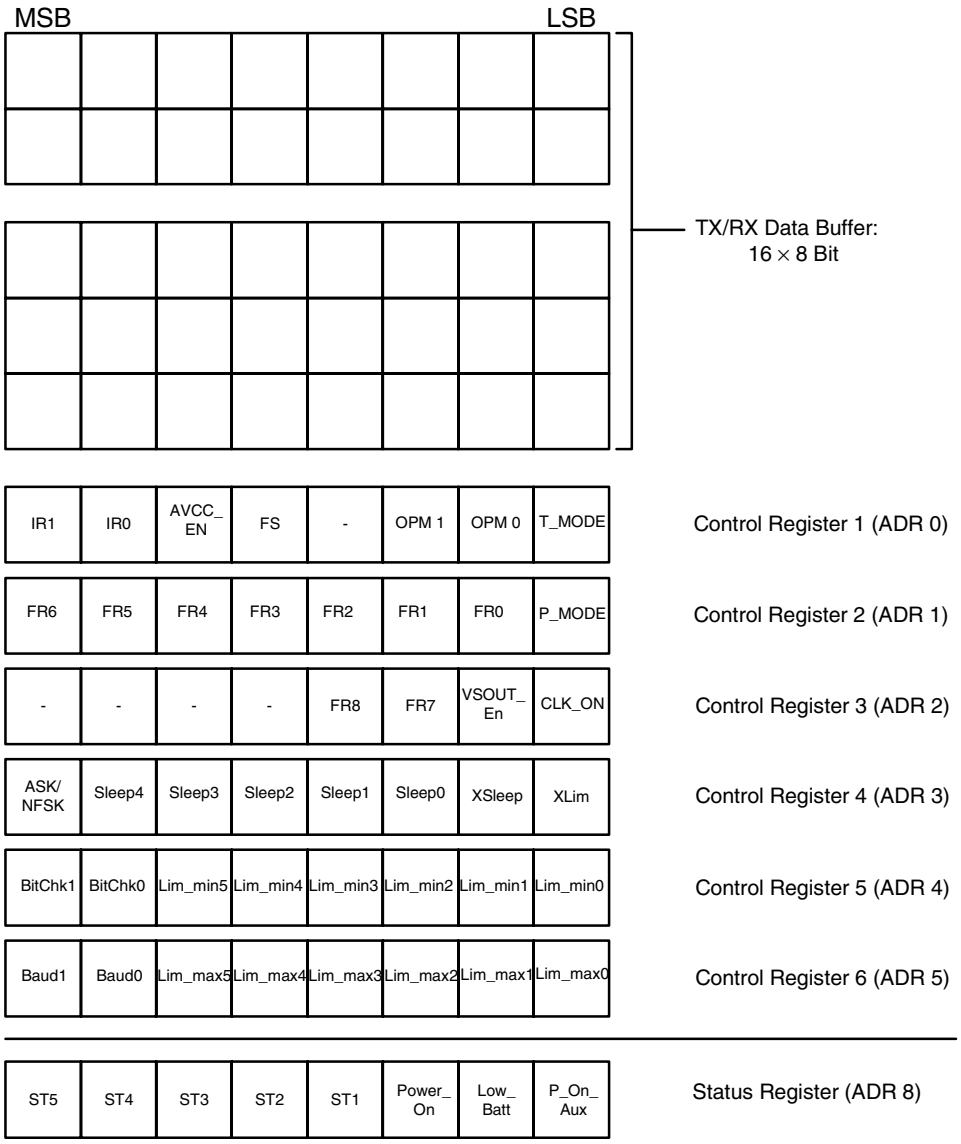
The configuration of the transceiver is stored in RAM cells. The RAM contains a 16 × 8-bit TX/RX data buffer and a 6 × 8-bit control register and is write and readable via a 4-wire serial interface (CS, SCK, SDI_TMDI, SDO_TMDO).

The 1 × 8-bit status register is not part of the RAM and is readable via the 4-wire serial interface.

The RAM and the status information is stored as long as the transceiver is in any active mode (DVCC = VS1 or DVCC = V_REG2) and gets lost if the transceiver is in OFF mode (DVCC = OFF).

After the transceiver is turned on via pin PWR_ON = High, T1 = Low, T2 = Low, T3 = Low, T4 = Low or T5 = Low or the voltage at pin VAUX V_{VAUX} > 3.5 V (typically) the control registers are in the default state.

Figure 27. Register Structure



TX/RX Data Buffer

The TX/RX data buffer is used to handle the data transfer during RX and TX operations.

Control Register

To use the transceiver in different applications it can be configured by a connected microcontroller via the 4-wire serial interface.

Control Register 1 (ADR 0)

Table 14. Control Register 1 (Function of Bit 7 and Bit 6 in RX Mode)

IR1	IR0	Function (RX Mode)
0	0	Pin IRQ is set to 1 if 4 received bytes are in the TX/RX data buffer or a receiving error occurred
0	1	Pin IRQ is set to 1 if 8 received bytes are in the TX/RX data buffer or a receiving error occurred
1	0	Pin IRQ is set to 1 if 12 received bytes are in the TX/RX data buffer or a receiving error occurred (default)
1	1	Pin IRQ is set to 1 if a receiving error occurred

Table 15. Control Register 1 (Function of Bit 7 and Bit 6 in TX Mode)

IR1	IR0	Function (TX Mode)
0	0	Pin IRQ is set to 1 if 4 bytes still are in the TX/RX data buffer or the TX data buffer is empty
0	1	Pin IRQ is set to 1 if 8 bytes still are in the TX/RX data buffer or the TX data buffer is empty
1	0	Pin IRQ is set to 1 if 12 bytes still are in the TX/RX data buffer or the TX data buffer is empty (default)
1	1	Pin IRQ is set to 1 if the TX data buffer is empty

Table 16. Control Register 1 (Function of Bit 5)

AVCC_EN	Function
0	(default)
1	Enables AVCC, if the ATA5811/ATA5812 is in AUX mode

Table 17. Control Register 1 (Function of Bit 4)

FS	Function
0	433/868 MHz
1	315 MHz

Table 18. Control Register 1 (Function of Bit 2 and Bit 1)

OPM1	OPM0	Function
0	0	Idle mode (default)
0	1	TX mode
1	0	RX polling mode
1	1	RX mode

Table 19. Control Register 1 (Function of Bit 0)

T_MODE	Function
0	TX and RX function via TX/RX data buffer (default)
1	Transparent mode, TX/RX data buffer disabled, TX modulation data stream via pin SDI_TMDI, RX modulation data stream via pin SDO_TMDO

Control Register 2 (ADR 1)

Table 20. Control Register 2 (Function of Bit 7, Bit 6, Bit 5, Bit 4, Bit 3, Bit 2 and Bit 1)

FR6	FR5	FR4	FR3	FR2	FR1	FR0	Function
0	0	0	0	0	0	0	FREQ2 = 0
0	0	0	0	0	0	1	FREQ2 = 1
.	.	.	.	.	.	.	
1	0	1	1	0	0	0	FREQ2 = 88 (default)
.	.	.	.	.	.	.	
1	1	1	1	1	1	1	FREQ2 = 127

Note: Tuning of f_{RF} LSB's (total 9 bits), frequency trimming, resolution of f_{RF} is $f_{XTO}/16384$ which is approximately 800 Hz (see section "XTO", Table 12 on page 24)

Table 21. Control Register 2 (Function of Bit 0 in RX Mode)

P_MODE	Function (RX Mode)
0	Pin IRQ is set to 1 if the Bit-check is successful (default)
1	No effect on pin IRQ if the Bit-check is successful

Table 22. Control Register 2 (Function of Bit 0 in TX Mode)

P_MODE	Function (TX Mode)
0	Manchester modulator on (default)
1	Manchester modulator off (NRZ mode)

Control Register 3 (ADR 2)

Table 23. Control Register 3 (Function of Bit 3 and Bit 2)

FR8	FR7	Function
0	0	FREQ3 = 0
0	1	FREQ3 = 128
1	0	FREQ3 = 256 (default)
1	1	FREQ3 = 384

Note: Tuning of f_{RF} MSB's

Table 24. Control Register 3 (Function of Bit 1)

VSOUT_EN	Function
0	Output voltage power supply for external devices off (pin VSOUT)
1	Output voltage power supply for external devices on (default)

Note: This bit is set to 1 if the Bit-check is ok (RX_Polling, RX mode), an event at pin T1, T2, T3, T4 or T5 occurs or the bit Power_On in the status register is 1.
Setting VSOUT_EN = 0 in AUX mode is not allowed

Table 25. Control Register 3 (Function of Bit 0)

CLK_ON	Function
0	Clock output off (pin CLK)
1	Clock output on (default)

Note: This bit is set to 1 if the Bit-check is ok (RX_Polling, RX mode), an event at pin T1, T2, T3, T4 or T5 occurs or the bit Power_On in the status register is 1.

Control Register 4 (ADR 3)

Table 26. Control Register 4 (Function of Bit 7)

ASK_NFSK	Function
0	FSK mode (default)
1	ASK mode

Table 27. Control Register 4 (Function of Bit 6, Bit 5, Bit 4, Bit 3 and Bit 2)

Sleep4	Sleep3	Sleep2	Sleep1	Sleep0	Function Sleep (T_{Sleep} = Sleep × 1024 × T_{DCLK} × X_{Sleep})
0	0	0	0	0	0
0	0	0	0	1	1
.	.	.	.	.	
0	1	0	1	0	10 (T_{Sleep} = 10 × 1024 × T_{DCLK} × X_{Sleep}) (default)
.	.	.	.	.	
1	1	1	1	1	31

Table 28. Control Register 4 (Function of Bit 1)

XSleep	Function
0	X_{Sleep} = 1; extended T_{Sleep} off (default)
1	X_{Sleep} = 8; extended T_{Sleep} on

Table 29. Control Register 4 (Function of Bit 0)

XLim	Function
0	X_{Lim} = 1; extended TLim_min, TLim_max off (default)
1	X_{Lim} = 2; extended T_{Lim_min}, T_{Lim_max} on

Control Register 5 (ADR 4)

Table 30. Control Register 5 (Function of Bit 7 and Bit 6)

BitChk1	BitChk0	Function
0	0	$N_{Bit\text{-}check}$ = 0 (0 bits checked during Bit-check)
0	1	$N_{Bit\text{-}check}$ = 3 (3 bits checked during Bit-check (default))
1	0	$N_{Bit\text{-}check}$ = 6 (6 bits checked during Bit-check)
1	1	$N_{Bit\text{-}check}$ = 9 (9 bits checked during Bit-check)

Table 31. Control Register 5 (Function of Bit 5, Bit 4, Bit 3, Bit 2, Bit 1 and Bit 0 in RX Mode)

Lim_min5	Lim_min4	Lim_min3	Lim_min2	Lim_min1	Lim_min0	Function (RX Mode) Lim_min (Lim_min < 10 are not applicable) (T_{Lim_min} = Lim_min × T_{XDCLK})
0	0	1	0	1	0	10
0	0	1	0	1	1	11
.	.	.	.	.	.	
0	1	0	0	0	0	16 (T_{Lim_min} = 16 × T_{XDCLK}) (default)
.	.	.	.	.	.	
1	1	1	1	1	1	63

Table 32. Control Register 5 (Function of Bit 5, Bit 4, Bit 3, Bit 2, Bit 1 and Bit 0 in TX Mode)

Lim_min5	Lim_min4	Lim_min3	Lim_min2	Lim_min1	Lim_min0	Function (TX Mode) Lim_min (Lim_min < 10 are not applicable) (TX_Baudrate = $1/((Lim_min + 1) \times T_{XDCLK} \times 2)$
0	0	1	0	1	0	10
0	0	1	0	1	1	11
.	.	.	.	.	.	
0	1	0	0	0	0	16 ($TX_Baudrate = 1/((16 + 1) \times T_{XDCLK} \times 2)$) (default)
.	.	.	.	.	.	
1	1	1	1	1	1	63

Control Register 6 (ADR 5)

Table 33. Control Register 6 (Function of Bit 7 and Bit 6)

Baud1	Baud0	Function
0	0	Baud-rate range 0 (B0) 1.0 kBaud to 2.5 kBaud; $T_{XDCLK} = 8 \times T_{DCLK} \times X_{Lim}$
0	1	Baud-rate range 1 (B1) 2.0 kBaud to 5.0 kBaud; $T_{XDCLK} = 4 \times T_{DCLK} \times X_{Lim}$
1	0	Baud-rate range 2 (B2) 4.0 kBaud to 10.0 kBaud; $T_{XDCLK} = 2 \times T_{DCLK} \times X_{Lim}$; (default)
1	1	Baud-rate range 3 (B3) 8.0 kBaud to 20.0 kBaud; $T_{XDCLK} = 1 \times T_{DCLK} \times X_{Lim}$, Note that the receiver is not working with >10 kBaud in ASK mode

Table 34. Control Register 6 (Function of Bit 5, Bit 4, Bit 3, Bit 2, Bit 1 and Bit 0)

Lim_max5	Lim_max4	Lim_max3	Lim_max2	Lim_max1	Lim_max0	Function Lim_max (Lim_max < 12 Are Not Applicable) ($T_{Lim_max} = (Lim_max - 1) \times T_{XDCLK}$)
0	0	1	1	0	0	12
0	0	1	1	0	1	13
.	.	.	.	.	.	
0	1	1	1	0	0	28 ($T_{Lim_max} = (28 - 1) \times T_{XDCLK}$) (default)
.	.	.	.	.	.	
1	1	1	1	1	1	63

Status Register

The status register indicates the current status of the transceiver and is readable via the 4-wire serial interface. Setting Power_On or P_On_Aux or an event on ST1, ST2, ST3, ST4 or ST5 is indicated by an IRQ.

Reading the status register resets the bits Power_On, Low_Batt, P_On_Aux and the IRQ

Status Register (ADR 8)

Table 35. Status Register

Status Bit	Function
ST5	Status of pin T5 Pin T5 = 0 →ST5 = 1 Pin T5 = 1 →ST5 = 0 (see Figure 29 on page 41)
ST4	Status of pin T4 Pin T4 = 0 →ST4 = 1 Pin T4 = 1 →ST4 = 0 (see Figure 29 on page 41)
ST3	Status of pin T3 Pin T3 = 0 →ST3 = 1 Pin T3 = 1 →ST3 = 0 (see Figure 29 on page 41)
ST2	Status of pin T2 Pin T2 = 0 →ST2 = 1 Pin T2 = 1 →ST2 = 0 (see Figure 29 on page 41)
ST1	Status of pin T1 Pin T1 = 0 →ST1 = 1 Pin T1 = 1 →ST1 = 0 (see Figure 29 on page 41)
Power_On	Indicates that the transceiver was woken up by pin PWR_ON (rising edge on pin PWR_ON). During Power_On = 1, the bits VSOUT_EN and CLK_ON in control register 3 are set to 1. (see Figure 30 on page 42)
Low_Batt	Indicates that output voltage on pin VSOUT is too low (V_{VSOUT} < 2.38 V typically) (see Figure 31 on page 43)
P_On_Aux	Indicates that the auxiliary supply voltage on pin VAUX is high enough to operate. State transition: a) OFF mode →AUX mode (see Figure 22 on page 28) b) Idle mode (VSOUT = VS1) →Idle mode (VSOUT = V_REG2) (see Figure 32 on page 44)

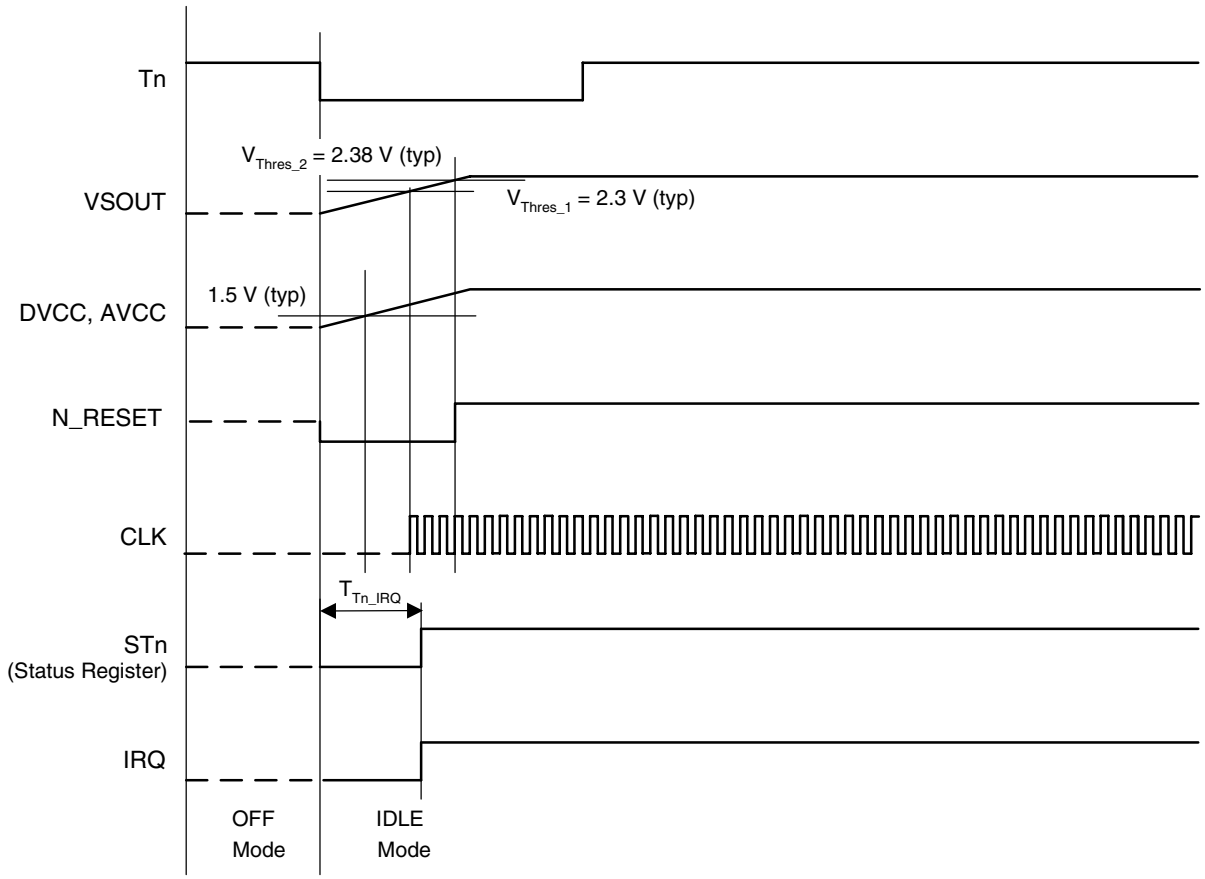
Pin Tn

To switch the transceiver from OFF to Idle mode, pin Tn must set to 0 (maximum $0.2 \times V_{VS2}$) for at least T_{Tn_IRQ} (see Figure 28). The transceiver recognize the negative edge, sets pin N_RESET to low and switches on DVCC, AVCC and the power supply for external devices VSOUT.

If V_{DVCC} exceeds 1.5 V (typically) and the XTO is settled, the digital control logic is active and sets the status bit STn to 1 and an interrupt is issued (T_{Tn_IRQ}).

After the voltage on pin VSOUT exceeds 2.3 V (typically) and the start-up time of the XTO is elapsed the output clock on pin CLK is available. Because the enabling of pin CLK is asynchronous the first clock cycle may be incomplete. N_RESET is set to high if V_{VSOUT} exceeds 2.38 V (typically) and the XTO is settled.

Figure 28. Timing Pin Tn, Status Bit STn



If the transceiver is in any active mode (Idle, AUX, TX, RX, RX_Polling), an integrated debounce logic is active. If there is an event on pin Tn a debounce counter is set to 0 (T = 0) and started. The status is updated, an interrupt is issued and the debounce counter is stopped after reaching the counter value $T = 8195 \times T_{DCLK}$.

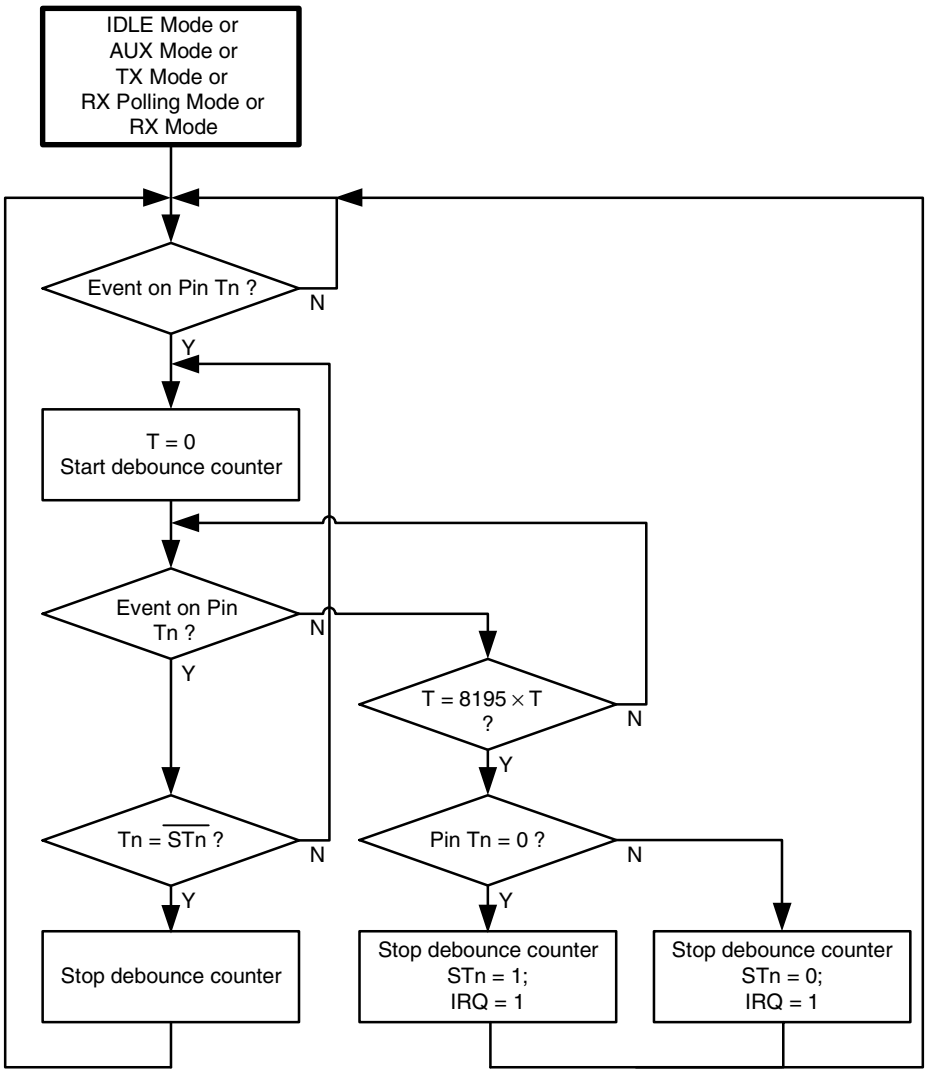
An event on the same key input before reaching $T = 8195 \times T_{DCLK}$ stops the debounce counter. An event on an other key input before reaching $T = 8195 \times T_{DCLK}$ resets and restarts the debounce counter.

While the debounce counter is running, the bits VSOUT_EN and CLK_ON in control register 3 are set to 1.

The interrupt is deleted after reading the status register or executes the command Delete_IRQ.

If a pin Tn is not used, it can be left open because of an internal pull-up resistor (typically 50 kΩ).

Figure 29. Timing Flow Pin Tn, Status Bit STn



Pin PWR_ON

To switch the transceiver from OFF to Idle mode, pin PWR_ON must set to 1 (minimum $0.8 \times V_{VS2}$) for at least T_{PWR_ON} (see Figure 30). The transceiver recognize the positive edge, sets pin N_RESET to low and switches on DVCC, AVCC and the power supply for external devices VSOUT.

If V_{DVCC} exceeds 1.5 V (typically) and the XTO is settled, the digital control logic is active and sets the status bit Power_On to 1 and an interrupt is issued ($T_{PWR_ON_IRQ_1}$).

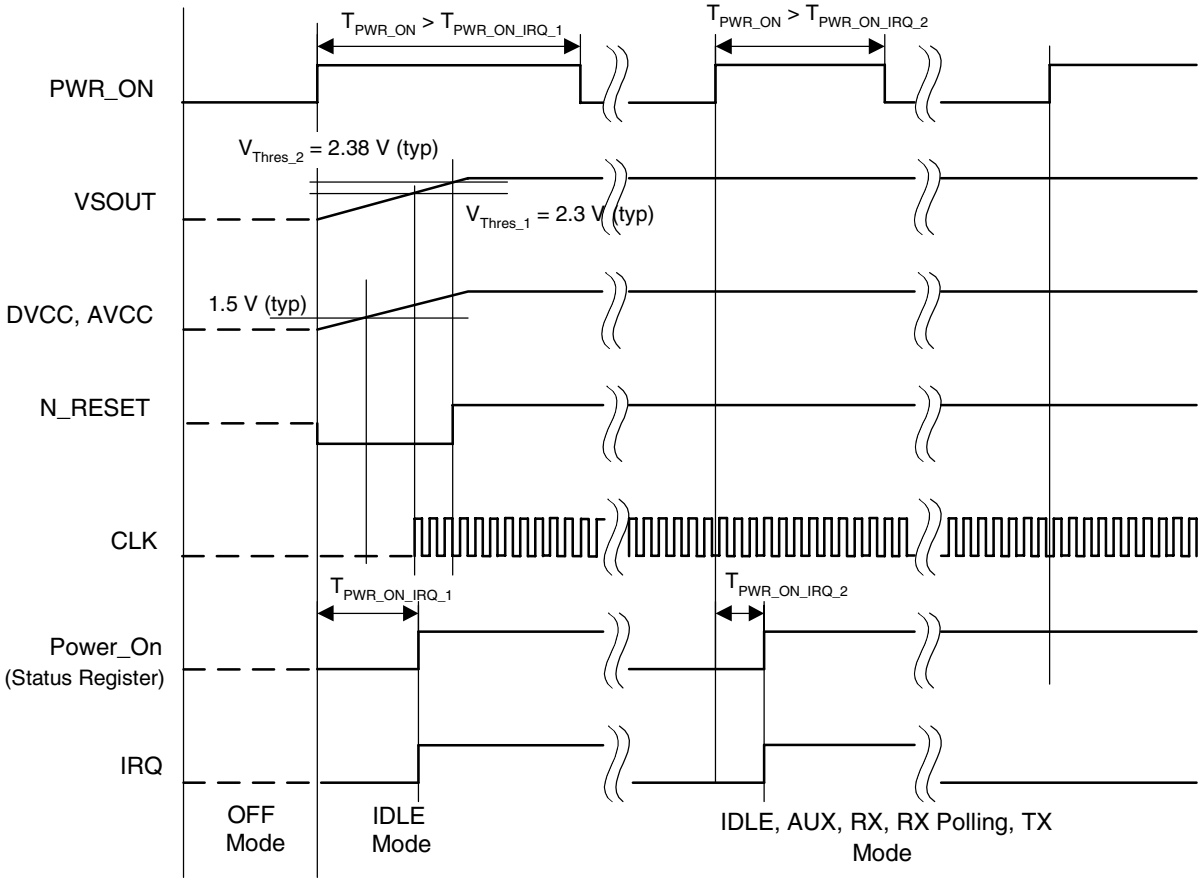
After the voltage on pin VSOUT exceeds 2.3 V (typically) and the start-up time of the XTO is elapsed the output clock on pin CLK is available. Because the enabling of pin CLK is asynchronous the first clock cycle may be incomplete. N_RESET is set to high if V_{VSOUT} exceeds 2.38 V (typically) and the XTO is settled.

If the transceiver is in any active mode (Idle, AUX, RX, RX_Polling, TX), a positive edge on pin PWR_ON sets Power_On to 1 (after $T_{PWR_ON_IRQ_2}$). The state transition Power_On 0 →1 generates an interrupt. If Power_On is still 1 during the positive edge on pin PWR_ON no interrupt is issued. Power_On and the interrupt is deleted after reading the status register.

During Power_On = 1, the bits VSOUT_EN and CLK_ON in control register 3 are set to 1.

Note: It is not possible to set the transceiver to OFF mode by setting pin PWR_ON to 0. If pin PWR_ON is not used, it must be connected to GND.

Figure 30. Timing Pin PWR_ON, Status Bit Power_On

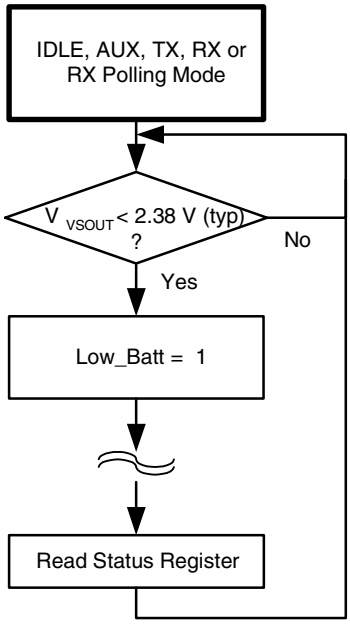


Low Battery Indicator

The status bit Low_Batt is set to 1 if the voltage on pin VSOUT V_{VSOUT} drops under 2.38 V (typically).

Low_Batt is set to 0 if V_{VSOUT} exceeds V_{Thres_2} and the status register is read via the 4-wire serial interface (see Figure 23 on page 30).

Figure 31. Timing Status Bit Low_Batt



Pin VAUX

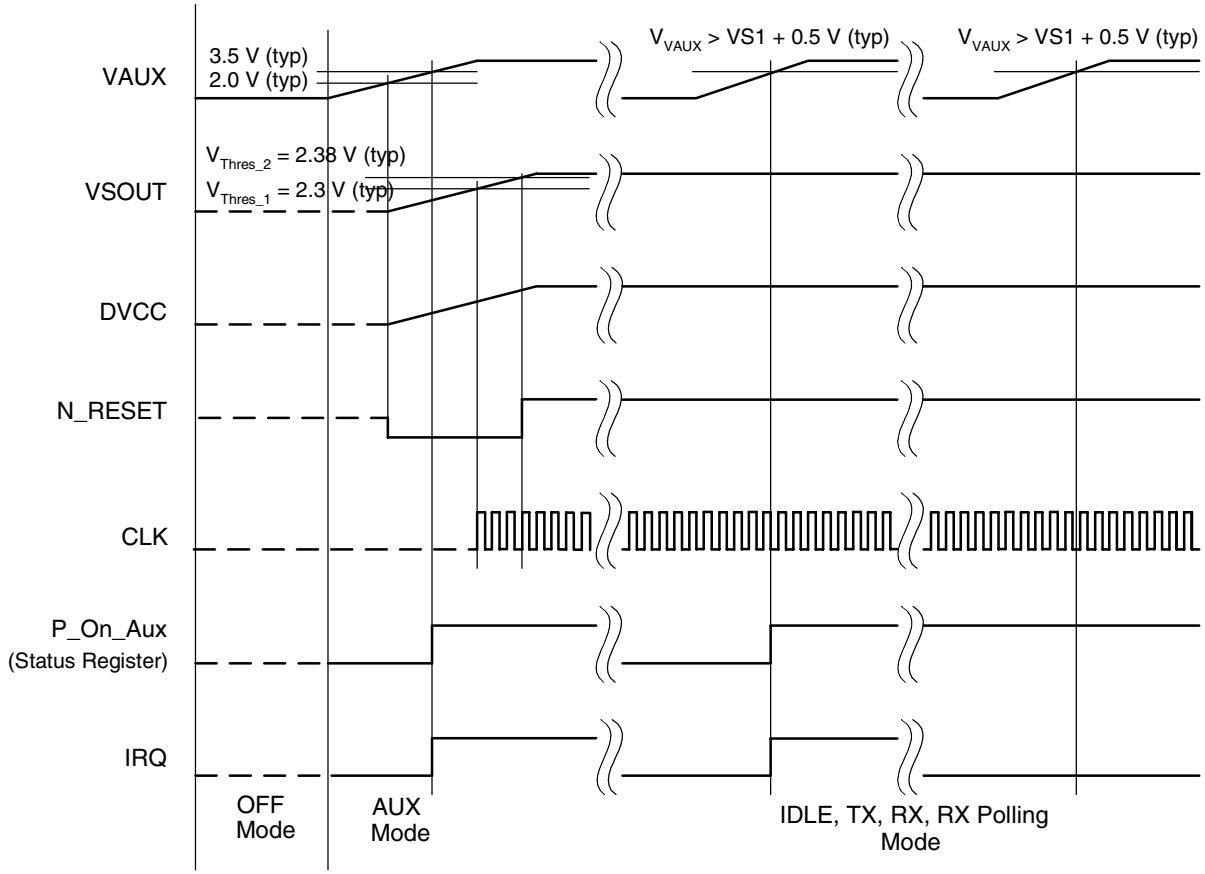
To switch the transceiver from OFF to AUX mode, the voltage on pin VAUX V_{VAUX} must exceed 3.5 V (typically) (see Figure 32). If V_{VAUX} exceeds 2 V (typically) pin N_RESET is set to low, DVCC and the power supply for external devices VSOUT are switched on.

If V_{VAUX} exceeds 3.5 V (typically) the status bit P_On_Aux is set to 1 and an interrupt is issued.

After the voltage on pin VSOUT exceeds 2.3 V (typically) and the start-up time of the XTO is elapsed the output clock on pin CLK is available. Because the enabling of pin CLK is asynchronous the first clock cycle may be incomplete. N_RESET is set to high if V_{VSOUT} exceeds 2.38 V (typically) and the XTO is settled.

If the transceiver is in any active mode (Idle, TX, RX, RX_Polling), a positive edge on pin VAUX and V_{VAUX} > VS1 + 0.5 V sets P_On_Aux to 1. The state transition P_On_Aux 0 → 1 generates an interrupt. If P_On_Aux is still 1 during the positive edge on pin VAUX no interrupt is issued. P_On_Aux and the interrupt is deleted after reading the status register.

Figure 32. Timing Pin VAUX, Status Bit P_On_Aux



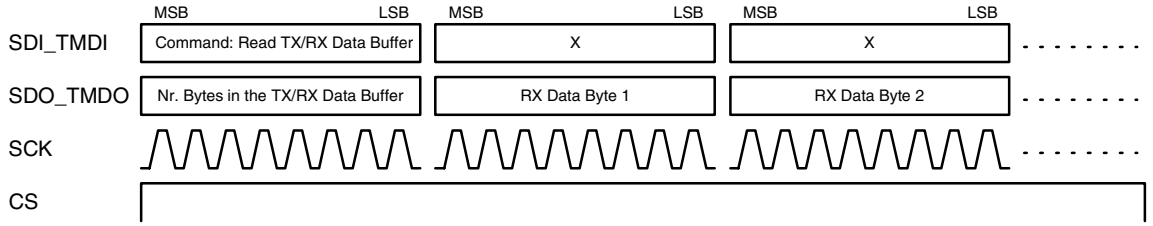
Transceiver Configuration

The configuration of the transceiver takes place via a 4-wire serial interface (CS, SCK, SDI_TMDI, SDO_TMDO) and is organized in 8-bit units. The configuration is initiated with a 8-bit command. While shifting the command into pin SDI_TMDI, the number of bytes in the TX/RX data buffer are available on pin SDO_TMDO. The read and write commands are followed by one or more 8-bit data units. Each 8-bit data transmission begins with the MSB. The serial interface is in reset state if the level on pin CS = Low.

Command: Read TX/RX Data Buffer

During a RX operation the user can read the received bytes in the TX/RX data buffer successively.

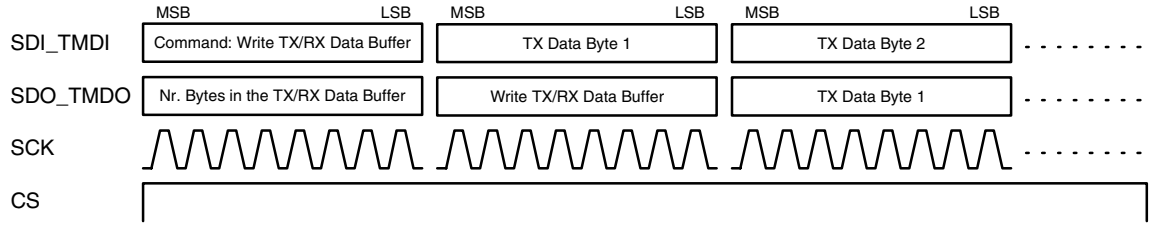
Figure 33. Read TX/RX Data Buffer



Command: Write TX/RX Data Buffer

During a TX operation the user can write the bytes in the TX/RX data buffer successively. An echo of the command and the TX data bytes are provided for the microcontroller on pin SDO_TMDO.

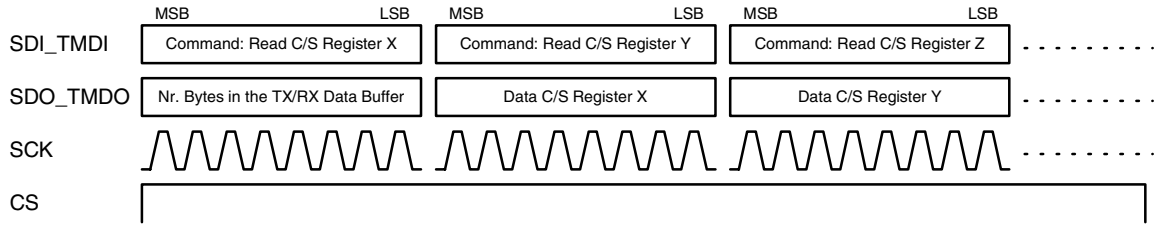
Figure 34. Write TX/RX Data Buffer



Command: Read Control/Status Register

The control and status registers can be read individually or successively.

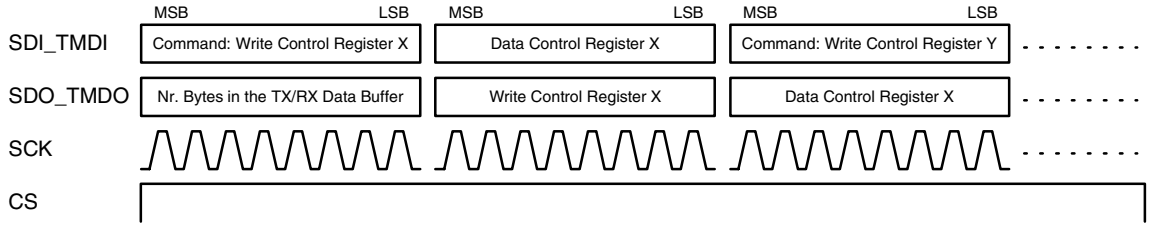
Figure 35. Read Control/Status Register



Command: Write Control Register

The control registers can be written individually or successively. An echo of the command and the data bytes are provided for the microcontroller on pin SDO_TMDO.

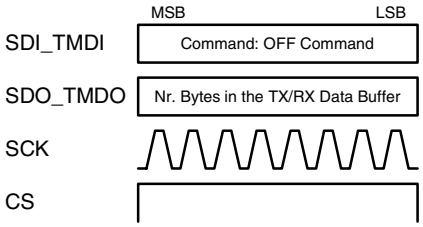
Figure 36. Write Control Register



Command: OFF Command

If AVCC_EN in control register 1 is 0, the input level on pin PWR_ON is low and on the key inputs Tn is high, the OFF command sets the transceiver in the OFF mode.

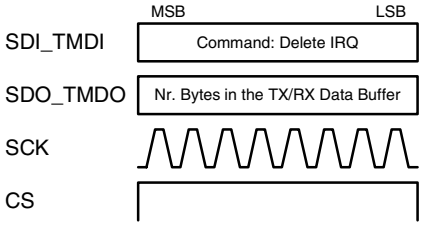
Figure 37. OFF Command



Command: Delete IRQ

The delete IRQ command sets pin IRQ to low.

Figure 38. Delete IRQ



Command Structure

The three most significant bits of the command (Bit 5 to Bit 7) indicates the command type. Bit 0 to Bit 4 describes the target address when reading or writing a control or status register. In all other commands Bit 0 to Bit 4 have no effect and should be set to 0 for compatibility reasons with future products.

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Table 36. Command Structure

Command	MSB Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	LSB Bit 0
Read TX/RX data buffer	0	0	0	x	x	x	x	x
Write TX/RX data buffer	0	0	1	x	x	x	x	x
Read control/status register	0	1	0	A4	A3	A2	A1	A0
Write control register	0	1	1	A4	A3	A2	A1	A0
OFF command	1	0	0	X	X	X	X	X
Delete IRQ	1	0	1	X	X	X	X	X
Not used	1	1	0	X	X	X	X	X
Not used	1	1	1	X	X	X	X	X

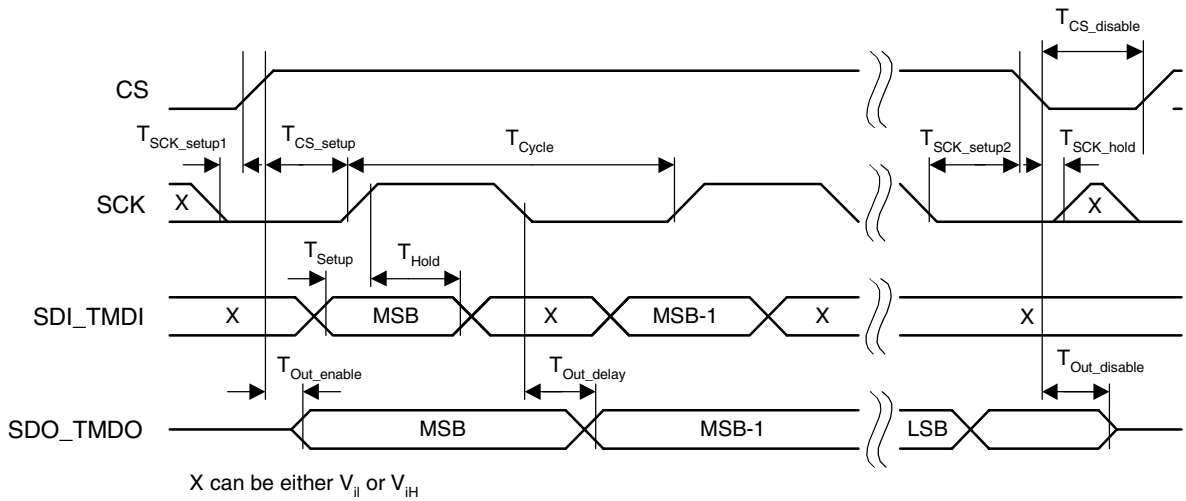
4-wire Serial Interface

The 4-wire serial interface consists of the Chip Select (CS), the Serial ClocK (SCK), the Serial Data Input (SDI_TMDI) and the Serial Data Output (SDO_TMDO). Data is transmitted/received bit by bit in synchronization with the serial clock.

Note: If the output level on pin N_RESET is low, no data communication with the microcontroller is possible.

When CS is low and the transparent mode is inactive (T_MODE = 0), SDO_TMDO is in a high-impedance state. When CS is low and the transparent mode is active (T_MODE = 1), the RX data stream is available on pin SDO_TMDO.

Figure 39. Serial Timing



Operation Modes

RX Operation

The transceiver is set to RX operation with the bits OPM0 and OPM1 in control register 1

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Table 37. Control Register 1

OPM1	OPM0	Function
1	0	RX polling mode
1	1	RX mode

The transceiver is designed to consume less than 1 mA in RX operation while being sensitive to signals from a corresponding transmitter. This is achieved via the polling circuit. This circuits enables the signal path periodically for a short time. During this time the Bit-check logic verifies the presence of a valid transmitter signal. Only if a valid signal is detected the transceiver remains active and transfers the data to the connected microcontroller. This transfer take place either via the TX/RX data buffer or via the pin SDO_TMDO. If there is no valid signal present the transceiver is in sleep mode most of the time resulting in low current consumption. This condition is called RX polling mode. A connected microcontroller can be disabled during this time.

All relevant parameters of the polling logic can be configured by the connected microcontroller. This flexibility enables the user to meet the specifications in terms of current consumption, system response time, data rate etc.

In RX mode the RF transceiver is enabled permanently and the Bit-check logic verifies the presence of a valid transmitter signal. If a valid signal is detected the transceiver transfers the data to the connected microcontroller. This transfer take place either via the TX/RX data buffer or via the pin SDO_TMDO.

RX Polling Mode

If the transceiver is in RX polling mode it stays in a continuous cycle of three different modes. In sleep mode the RF transceiver is disabled for the time period T_{Sleep} while consuming low current of $I_S = I_{IDLE_X}$. During the start-up period, $T_{Startup_PLL}$ and $T_{Startup_Sig_Proc}$, all signal processing circuits are enabled and settled. In the following Bit-check mode, the incoming data stream is analyzed bit by bit contra a valid transmitter signal. If no valid signal is present, the transceiver is set back to sleep mode after the period $T_{Bit\text{-}check}$. This period varies check by check as it is a statistical process. An average value for $T_{Bit\text{-}check}$ is given in the electrical characteristics. During $T_{Startup_PLL}$ the current consumption is $I_S = I_{RX_X}$. During $T_{Startup_Sig_Proc}$ and $T_{Bit\text{-}check}$ the current consumption is $I_S = I_{Startup_Sig_Proc_X}$. The condition of the transceiver is indicated on pin RX_ACTIVE (see Figure 40 on page 50 and Figure 41 on page 51). The average current consumption in RX polling mode I_P is different in 1 battery application, 2 battery application or car application. To calculate I_P the index X must be replaced by VS1, 2 in 1 battery application, VS2 in 2 battery application or VS2, VAUX in car application (see section “Electrical Characteristics”)

$$I_P = \frac{I_{IDLE_X} \times T_{Sleep} + I_{Startup_PLL_X} \times T_{Startup_PLL} + I_{RX_X} \times (T_{Startup_Sig_Proc} + T_{Bitcheck})}{T_{Sleep} + T_{Startup_PLL} + T_{Startup_Sig_Proc} + T_{Bit_check}}$$

To save current it is recommended CLK and V_{VSOUT} be disabled during RX polling mode. I_P does not include the current of the Microcontroller_Interface I_{VSINT} and the current of an external device connected to pin VSOUT (e.g., microcontroller). If CLK and/or VSOUT is enabled during RX polling mode the current consumption is calculated as follows:

$$I_{S_Poll} = I_P + I_{VSINT} + I_{EXT}$$

During T_{Sleep}, $T_{Startup_PLL}$ and $T_{Startup_Sig_Proc}$ the transceiver is not sensitive to a transmitter signal. To guarantee the reception of a transmitted command the transmitter must start the telegram with an adequate preburst. The required length of the preburst $T_{Preburst}$ depends on the polling parameters T_{Sleep}, $T_{Startup_PLL}$, $T_{Startup_Sig_Proc}$ and $T_{Bit\text{-}check}$. Thus, $T_{Bit\text{-}check}$ depends on the actual bit rate and the number of bits ($N_{Bit\text{-}check}$) to be tested

$$.T_{Preburst} \geq T_{Sleep} + T_{Startup_PLL} + T_{Startup_Sig_Proc} + T_{Bit_check}$$

Sleep Mode

The length of period T_{Sleep} is defined by the 5-bit word sleep in control register 4, the extension factor X_{Sleep} defined by the bit XSleep in control register 4 and the basic clock cycle T_{DCLK}. It is calculated to be:

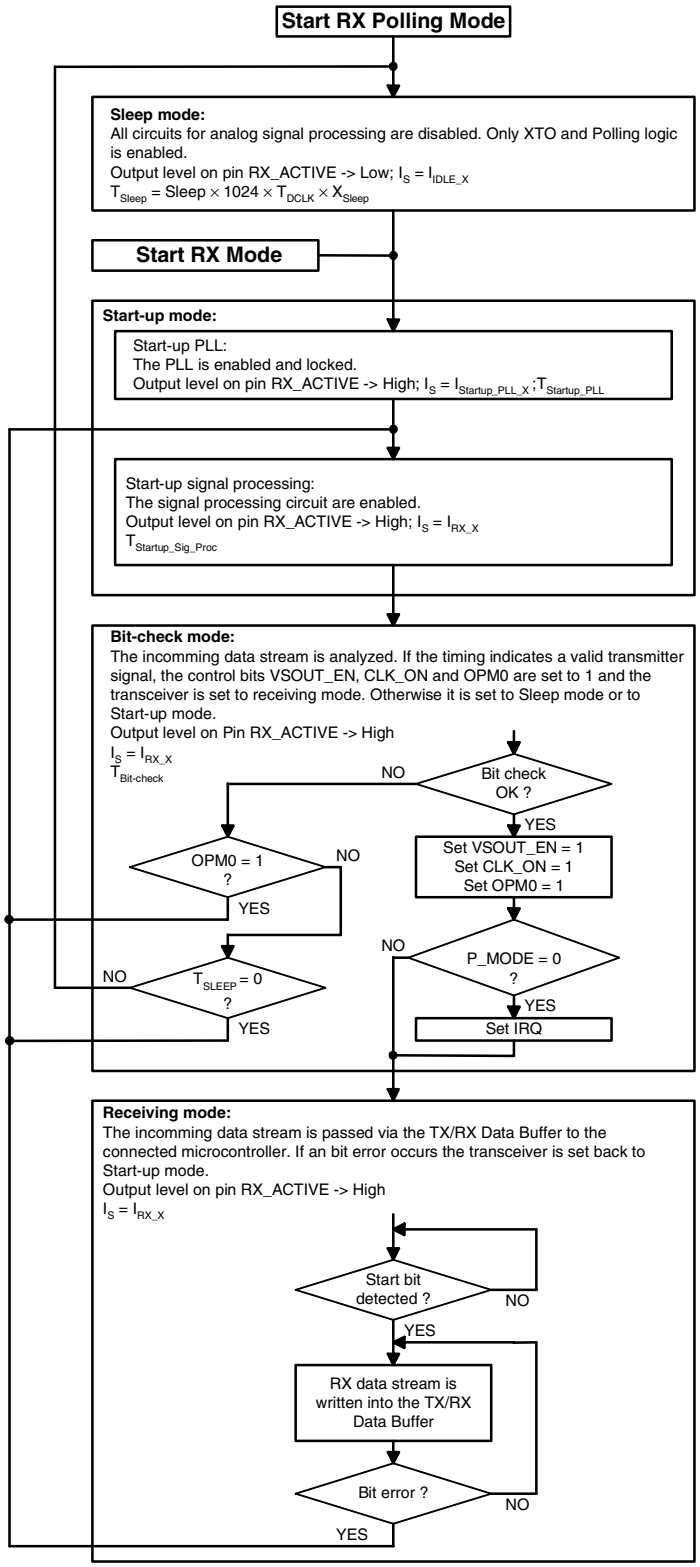
$$T_{Sleep} = Sleep \times 1024 \times T_{DCLK} \times X_{Sleep}$$

In US and European applications, the maximum value of T_{Sleep} is about 38 ms if X_{Sleep} is set to 1 (which is done by setting the bit XSleep in control register 4 to 0). The time resolution is about 1.2 ms in that case. The sleep time can be extended to about 300 ms by setting X_{Sleep} to 8 (which is done by setting XSleep in control register 4 to 1), the time resolution is then about 9.6 ms.

Start-up Mode

During $T_{Startup_PLL}$ the PLL is enabled and starts up. If the PLL is locked, the signal processing circuit starts up ($T_{Startup_Sig_Proc}$). After the start-up time all circuits are in stable condition and ready to receive.

Figure 40. Flow Chart Polling Mode/RX Mode (T_MODE = 1, Transparent Mode Inactive)



Sleep: Defined by bits Sleep0 to Sleep4 in Control Register 4

X_{Sleep}: Defined by bit XSleep in Control Register 4

T_{DCLK}: Basic clock cycle

$T_{Startup_PLL}$: $798.5 \times T_{DCLK}$ (typ)

$T_{Startup_Sig_Proc}$:

$882 \times T_{DCLK}$	(BR_Range 0)
$498 \times T_{DCLK}$	(BR_Range 1)
$306 \times T_{DCLK}$	(BR_Range 2)
$210 \times T_{DCLK}$	(BR_Range 3)

Is defined by the selected baud rate range and T_{DCLK}. The baud-rate range is defined by bit Baud0 and Baud1 in Control Register 6.

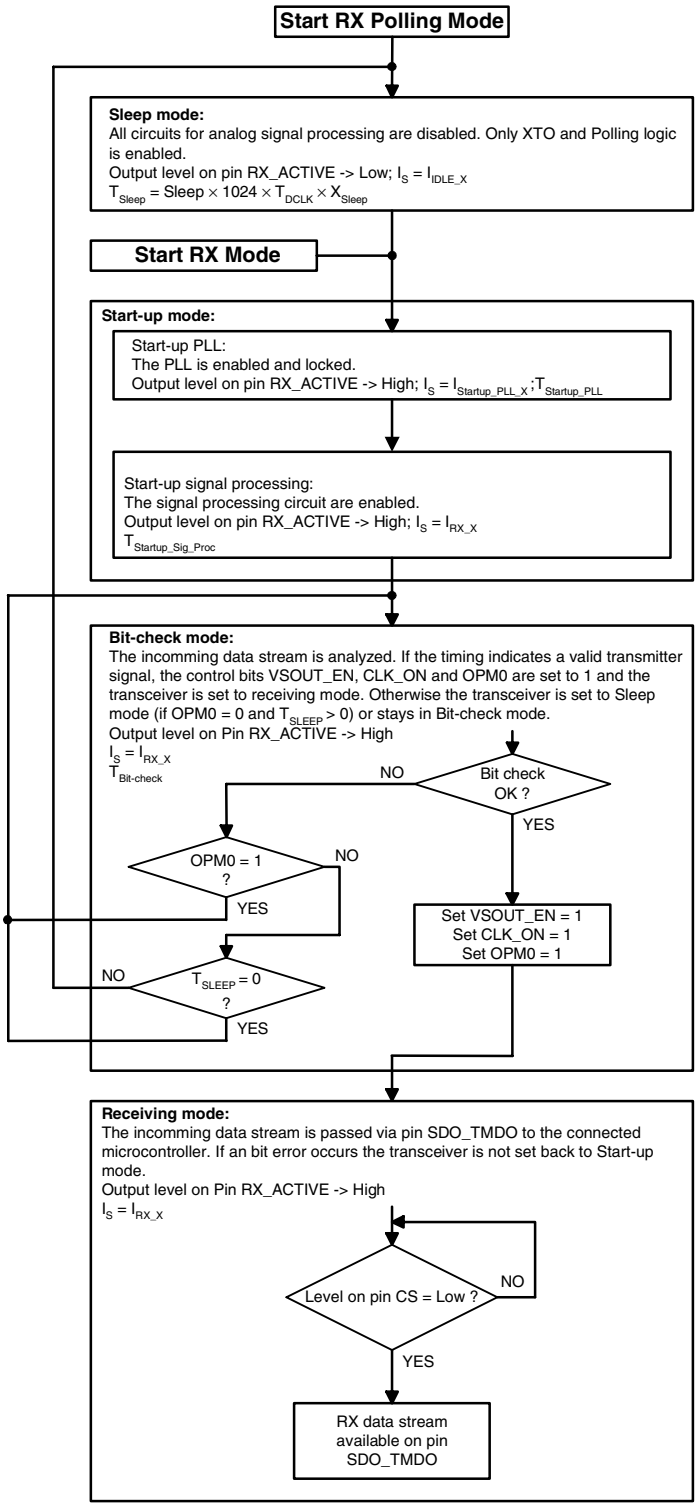
$T_{Bit\text{-}check}$: Depends on the result of the bit check.

If the bit check is ok, $T_{Bit\text{-}check}$ depends on the number of bits to be checked ($N_{Bit\text{-}check}$) and on the utilized data rate.

If the bit check fails, the average time period for that check depends on the selected baud-rate range and on T_{XDCLK}. The baud-rate range is defined by bit Baud0 and Baud1 in Control Register 6.

If the transceiver detects a bit error after a successful bit check and before the start bit is detected pin IRQ will be set to high (only if P_MODE=0) and the transceiver is set back to start-up mode.

Figure 41. Flow Chart Polling Mode/RX Mode (T_MODE = 1, Transparent Mode Active)



Sleep: Defined by bits Sleep0 to Sleep4 in Control Register 4

X_{Sleep}: Defined by bit XSleep in Control Register 4

T_{DCLK}: Basic clock cycle

$T_{Startup_PLL}$: $798.5 \times T_{DCLK}$ (typ)

$T_{Startup_Sig_Proc}$:

$882 \times T_{DCLK}$	(BR_Range 0)
$498 \times T_{DCLK}$	(BR_Range 1)
$306 \times T_{DCLK}$	(BR_Range 2)
$210 \times T_{DCLK}$	(BR_Range 3)

Is defined by the selected baud rate range and T_{DCLK}. The baud-rate range is defined by bit Baud0 and Baud1 in Control Register 6.

$T_{Bit\text{-}check}$: Depends on the result of the bit check.

If the bit check is ok, $T_{Bit\text{-}check}$ depends on the number of bits to be checked ($N_{Bit\text{-}check}$) and on the utilized data rate.

If the bit check fails, the average time period for that check depends on the selected baud-rate range and on T_{XDCLK}. The baud-rate range is defined by bit Baud0 and Baud1 in Control Register 6.

If in FSK mode the datastream is interrupted the FSK-Demodulator-PLL tends to lock out and is further not able to lock in, even there is a valid data stream available.
In this case the transceiver must be set back to IDLE mode.

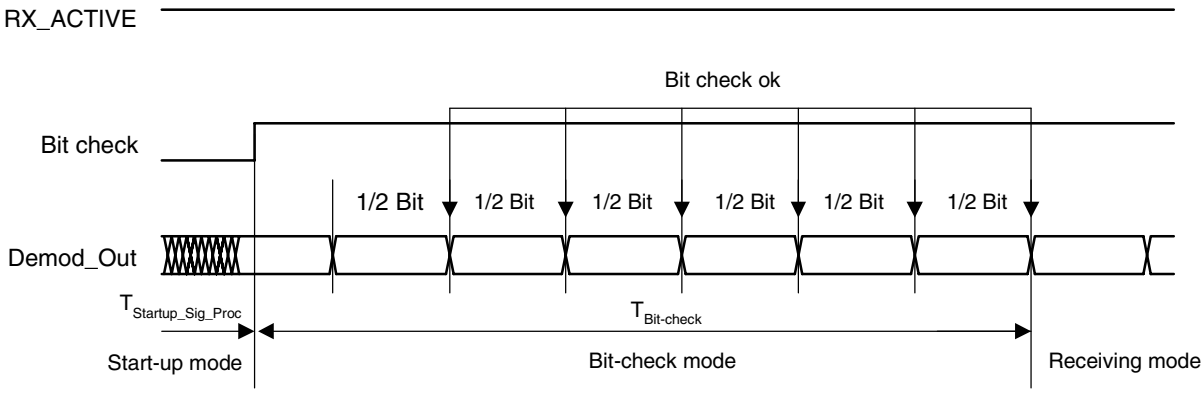
Bit-check Mode

In Bit-check mode the incoming data stream is examined to distinguish between a valid signal from a corresponding transmitter and signals due to noise. This is done by subsequent time frame checks where the distance between 2 signal edges are continuously compared to a programmable time window. The maximum count of this edge to edge test before the transceiver switches to receiving mode is also programmable.

Configuration the Bit-check

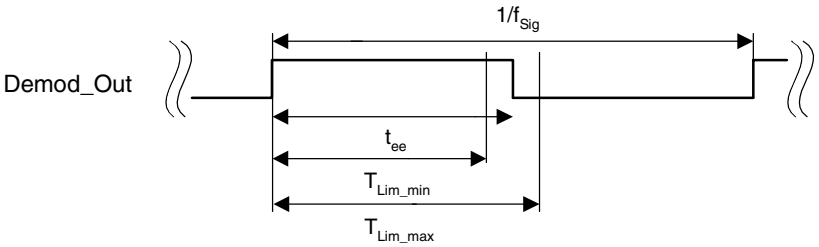
Assuming a modulation scheme that contains 2 edges per bit, two time frame checks are verifying one bit. This is valid for Manchester, Bi-phase and most other modulation schemes. The maximum count of bits to be checked can be set to 0, 3, 6 or 9 bits via the variable $N_{Bit\text{-}check}$ in control register 5. This implies 0, 6, 12 and 18 edge to edge checks respectively. If $N_{Bit\text{-}check}$ is set to a higher value, the transceiver is less likely to switch to receiving mode due to noise. In the presence of a valid transmitter signal, the Bit-check takes less time if $N_{Bit\text{-}check}$ is set to a lower value. In RX polling mode, the Bit-check time is not dependent on $N_{Bit\text{-}check}$. Figure 42 shows an example where 3 bits are tested successful.

Figure 42. Timing Diagram for Complete Successful Bit-check (Number of Checked Bits: 3)



According to Figure 43, the time window for the Bit-check is defined by two separate time limits. If the edge to edge time t_{ee} is in between the lower Bit-check limit T_{Lim_min} and the upper Bit-check limit T_{Lim_max}, the check will be continued. If t_{ee} is smaller than limit T_{Lim_min} or exceeds T_{Lim_max}, the Bit-check will be terminated and the transceiver switches to sleep mode.

Figure 43. Valid Time Window for Bit-check



For the best noise immunity it is recommended to use a low span between T_{Lim_min} and T_{Lim_max}. This is achieved using a fixed frequency at a 50% duty cycle for the transmitter preburst. A '11111...' or a '10101...' sequence in Manchester or Bi-phase is a good choice concerning that advice. A good compromise between sensitivity and susceptibility to noise regarding the expected edge to edge time t_{ee} is a time window of ±38%, to get the maximum sensitivity the time window should be ±50% and then $N_{Bit\text{-}check} \geq 6$. Using preburst patterns that contain various edge to edge time periods, the Bit-check limits must be programmed according to the required span.

The Bit-check limits are determined by means of the formula below:

$T_{Lim_min} = Lim_min \times T_{XDCLK}$
$T_{Lim_max} = (Lim_max - 1) \times T_{XDCLK}$

Lim_min is defined by the bits Lim_min 0 to Lim_min 5 in control register 5.

Lim_max is defined by the bits Lim_max 0 to Lim_max 5 in control register 6.

Using the above formulas, Lim_min and Lim_max can be determined according to the required T_{Lim_min}, T_{Lim_max} and T_{XDCLK}. The time resolution defining T_{Lim_min} and T_{Lim_max} is T_{XDCLK}. The minimum edge to edge time t_{ee} is defined according to the section “Receiving Mode”. The lower limit should be set to Lim_min ≥ 10. The maximum value of the upper limit is Lim_max = 63.

Figure 44, Figure 45 on page 54, and Figure 46 on page 54 illustrate the Bit-check for the Bit-check limits Lim_min = 14 and Lim_max = 24. The signal processing circuits are enabled during $T_{Startup_PLL}$ and $T_{Startup_Sig_Proc}$. The output of the ASK/FSK demodulator (Demod_Out) is undefined during that period. When the Bit-check becomes active, the Bit-check counter is clocked with the cycle T_{XDCLK}.

Figure 44 shows how the Bit-check proceeds if the Bit-check counter value CV_Lim is within the limits defined by Lim_min and Lim_max at the occurrence of a signal edge. In Figure 45 on page 54 the Bit-check fails as the value CV_Lim is lower than the limit Lim_min. The Bit-check also fails if CV_Lim reaches Lim_max. This is illustrated in Figure 46 on page 54.

Figure 44. Timing Diagram During Bit-check

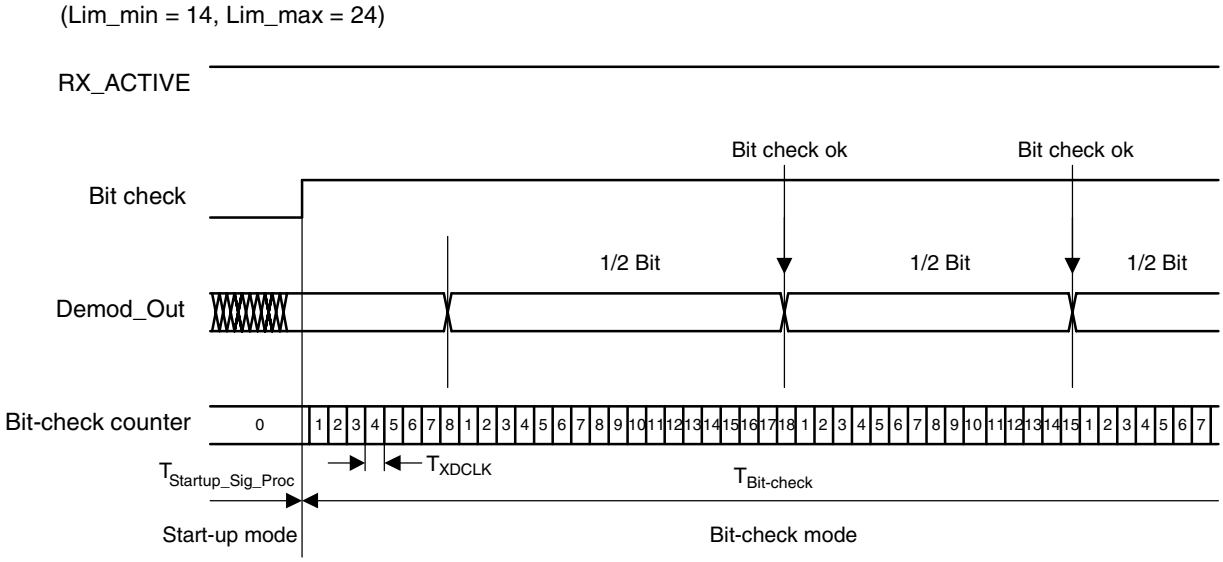


Figure 45. Timing Diagram for Failed Bit-check (Condition CV_Lim < Lim_min)

(Lim_min = 14, Lim_max = 24)

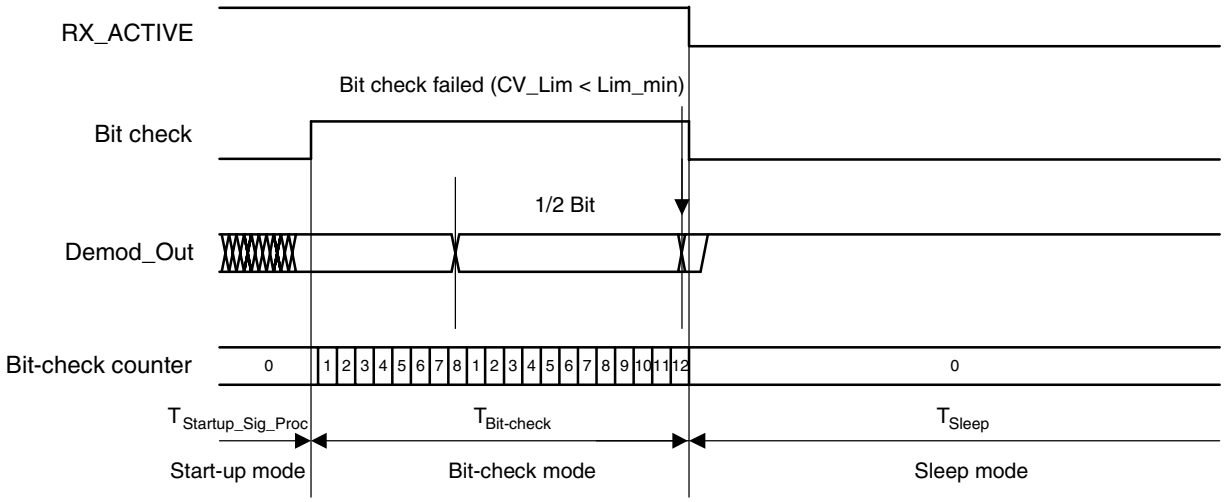
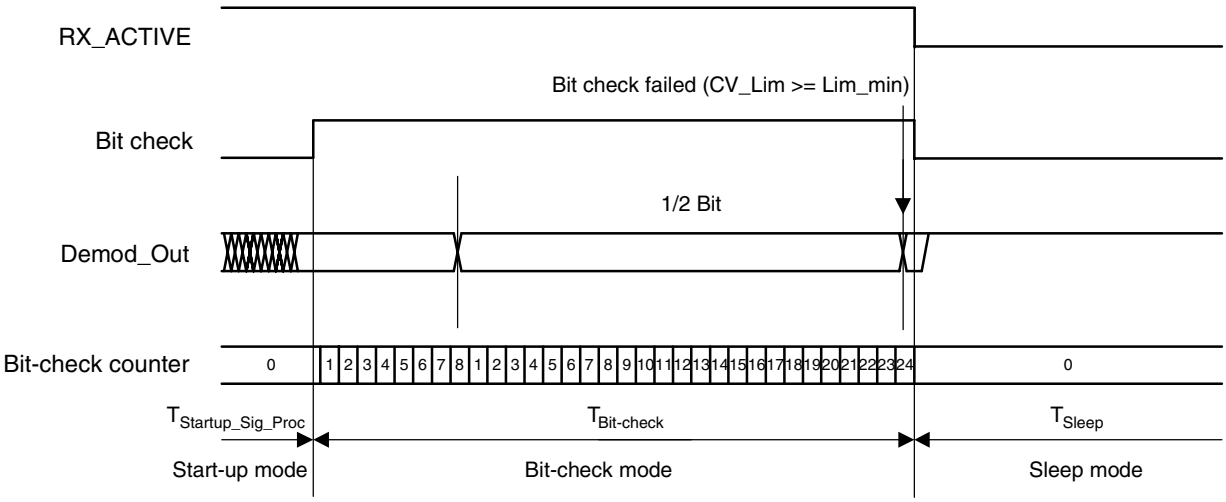


Figure 46. Timing Diagram for Failed Bit-check (Condition: CV_Lim ≥ Lim_max)

(Lim_min = 14, Lim_max = 24)



Duration of the Bit-check

If no transmitter is present during the Bit-check, the output of the ASK/FSK demodulator delivers random signals. The Bit-check is a statistical process and $T_{Bit\text{-}check}$ varies for each check. Therefore, an average value for $T_{Bit\text{-}check}$ is given in the electrical characteristics. $T_{Bit\text{-}check}$ depends on the selected baud rate range and on T_{XDCLK}. A higher baud-rate range causes a lower value for $T_{Bit\text{-}check}$ resulting in a lower current consumption in RX polling mode.

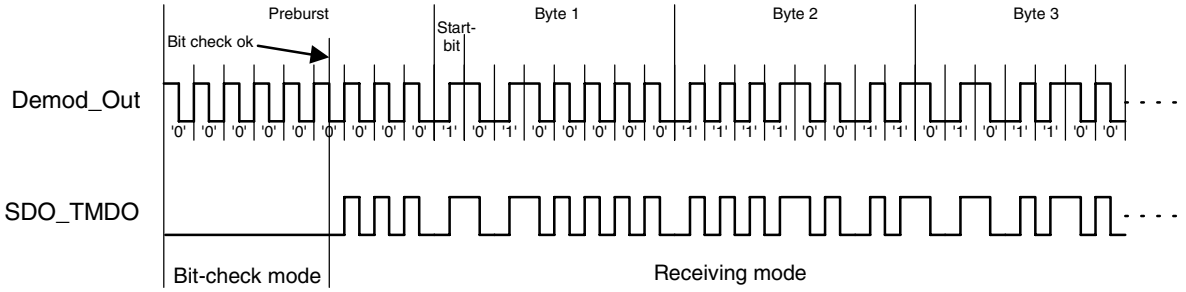
In the presence of a valid transmitter signal, $T_{Bit\text{-}check}$ is dependent on the frequency of that signal, f_{Sig}, and the count of the bits, $N_{Bit\text{-}check}$. A higher value for $N_{Bit\text{-}check}$ thereby results in a longer period for $T_{Bit\text{-}check}$ requiring a higher value for the transmitter preburst $T_{Preburst}$.

Receiving Mode

If the Bit-check was successful for all bits specified by $N_{Bit\text{-}check}$, the transceiver switches to receiving mode. To activate a connected microcontroller, the bits VSOUT_EN and CLK_ON in control register 3 are set to 1. An interrupt is issued at pin IRQ if the control bits T_MODE = 0 and P_MODE = 0.

If the transparent mode is active (T_MODE = 1) and the level on pin CS is low (no data transfer via the serial interface), the RX data stream is available on pin SDO_TMDO (Figure 47).

Figure 47. Receiving Mode (TMODE = 1)



If the transparent mode is inactive (T_MODE = 0), the received data stream is buffered in the TX/RX data buffer (see Figure 48 on page 56). The TX/RX data buffer is only usable for Manchester and Bi-phase coded signals. It is permanently possible to transfer the data from the data buffer via the 4-wire serial interface to a microcontroller (see Figure 33 on page 45).

Buffering of the data stream:

After a successful Bit-check, the transceiver switches from Bit-check mode to receiving mode. In receiving mode the TX/RX data buffer control logic is active and examines the incoming data stream. This is done, like in the Bit-check, by subsequent time frame checks where the distance between two edges is continuously compared to a programmable time window as illustrated in Figure 48 on page 56, only two distances between two edges in Manchester and Bi-phase coded signals are valid (T and 2T).

The limits for T are the same as used for the Bit-check. They can be programmed in control register 5 and 6 (Lim_min, Lim_max).

The limits for 2T are calculated as follows:

Lower limit of 2T:

$$Lim_min_2T = (Lim_min + Lim_max) - (Lim_max - Lim_min) / 2$$
$$T_{Lim_min_2T} = Lim_min_2T \times T_{XDCLK}$$

Upper limit of 2T:

$$Lim_max_2T = (Lim_min + Lim_max) + (Lim_max - Lim_min) / 2$$
$$T_{Lim_max_2T} = (Lim_max_2T - 1) \times T_{XDCLK}$$

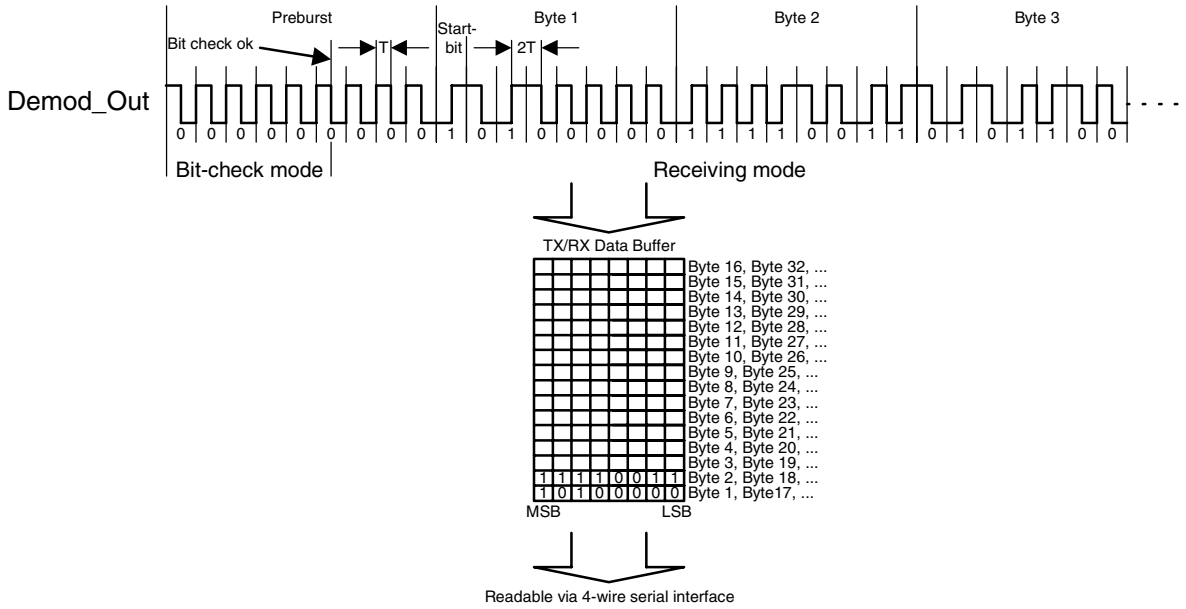
If the result of Lim_min_2T or Lim_max_2T is not an integer value, it will be round up.

If the TX/RX data buffer control logic detects the start bit, the data stream is written in the TX/RX data buffer byte by byte. The start bit is part of the first data byte and must be different from the bits of the preburst. If the preburst consists of a sequence of '00000...', the start bit must be a 1. If the preburst consists of a sequence of '11111...', the start bit must be a 0.

If the data stream consists of more than 16 bytes, a buffer overflow occurs and the TX/RX data buffer control logic overwrites the bytes already stored in the TX/RX data buffer. So it is very important to ensure that the data is read in time so that no buffer overflow occurs in that case (see Figure 33 on page 45). There is a counter that indicates the number of received bytes in the TX/RX data buffer (see section “Transceiver Configuration”). If a byte is transferred to the microcontroller, the counter is decremented, if a byte is received, the counter is incremented. The counter value is available via the 4-wire serial interface.

An interrupt is issued, if the counter while counting forwards reaches the value defined by the control bits IR0 and IR1 in control register 1.

Figure 48. Receiving Mode (TMODE = 0)



If the TX/RX data buffer control logic detects a bit error, an interrupt is issued and the transceiver is set back to the start-up mode (see Figure 40 on page 50, Figure 41 on page 51and Figure 49 on page 57).

Bit error: a) $t_{ee} < T_{Lim_min}$ or $T_{Lim_max} < t_{ee} < T_{Lim_min_2T}$ or $t_{ee} > T_{Lim_max_2T}$

b) Logical error (no edge detected in the bit center)

Note: The byte consisting of the bit error will not be stored in the TX/RX data buffer. Thus it is not available via the 4-wire serial interface.

Writing the control register 1, 4, 5 or 6 during receiving mode resets the TX/RX data buffer control logic and the counter which indicates the number of received bytes. If the bits OPM0 and OPM1 are still '1' after writing to a control register, the transceiver changes to the start-up mode (start-up signal processing).

Figure 49. Bit Error (TMODE = 0)

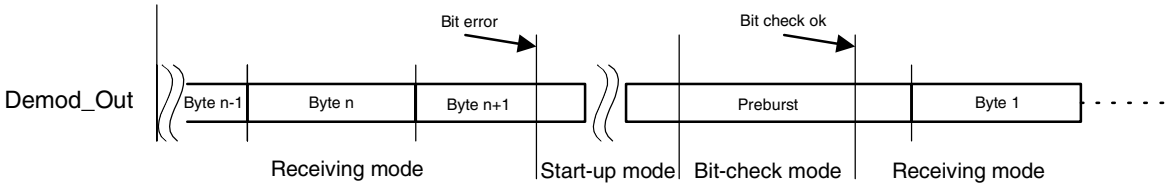


Table 38. RX Modulation Scheme

Mode	ASK/_NFSK	T_MODE	RF_{IN}	Bit in TX/RX Data Buffer	Level on Pin SD0_TMDO
RX	0	0	$f_{FSK_L} \rightarrow f_{FSK_H}$	1	Z
		0	$f_{FSK_H} \rightarrow f_{FSK_L}$	0	Z
		1	f_{FSK_H}	-	1
		1	f_{FSK_L}	-	0
	1	0	f_{ASK} off $\rightarrow f_{ASK}$ on	1	Z
		0	f_{ASK} on $\rightarrow f_{ASK}$ off	0	Z
		1	f_{ASK} on	-	1
		1	f_{ASK} off	-	0

Recommended Lim_min and Lim_max for Maximum Sensitivity

The sensitivity measurement in the section “Low-IF Receiver” in Table 3 on page 11 and Table 4 on page 11 have been done with the Lim_min and Lim_max values according to Table 39. These values are optimized for maximum sensitivity. Note that since these Limits are optimized for sensitivity the number of checked bit $N_{Bit\text{-}check}$ has to be at least 6 to prevent the circuit from waking up to a often in polling mode due to noise.

Table 39. Recommended Lim_min and Lim_max Values for Different Baud Rates

f_{RF} (f_{XTAL})/ MHz	1.0 kBaud BR_Range_0/XLim = 1	2.4 kBaud BR_Range_0/XLim = 0	5 kBaud BR_Range_1/XLim = 0	10 Kbaud BR_Range_2/XLim = 0	20 kBaud BR_Range_3/XLim = 0
315.0 (12.73193)	Lim_min = 13 (261 µs) Lim_max = 38 (744 µs)	Lim_min = 12 (121 µs) Lim_max = 34 (332 µs)	Lim_min = 11 (55 µs) Lim_max = 32 (156 µs)	Lim_min = 11 (28 µs) Lim_max = 32 (78 µs)	Lim_min = 11 (14 µs) Lim_max = 31 (38 µs)
433.92 (13.25311)	Lim_min = 13 (251 µs) Lim_max = 38 (715 µs)	Lim_min = 12 (116 µs) Lim_max = 34 (319 µs)	Lim_min = 11 (53 µs) Lim_max = 32 (150 µs)	Lim_min = 11 (27 µs) Lim_max = 32 (75 µs)	Lim_min = 11 (13 µs) Lim_max = 32 (37 µs)
868.3 (13.41191)	Lim_min = 13 (248 µs) Lim_max = 38 (706 µs)	Lim_min = 12 (115 µs) Lim_max = 34 (315 µs)	Lim_min = 11 (52 µs) Lim_max = 32 (148 µs)	Lim_min = 11 (26 µs) Lim_max = 32 (74 µs)	Lim_min = 11 (13 µs) Lim_max = 32 (37 µs)

TX Operation

The transceiver is set to TX operation by using the bits OPM0 and OPM1 in the control register 1.

Table 40. Control Register 1

OPM1	OPM0	Function
0	1	TX mode

Before activating TX mode, the TX parameters (baud rate, modulation scheme ...) must be selected as illustrated in Figure 50 on page 59. The baud rate depends on Baud 0 and Baud 1 in control register 6, Lim_min0 to Lim_min5 in control register 5 and XLIM in control register 4 (see section “Control Register”). The modulation is selected with ASK_/NFSK in control register 4. The FSK frequency deviation is fixed to about ±16 kHz. If P_Mode is set to 1, the Manchester modulator is disabled and pattern mode is active (NRZ, see Table 41 on page 61).

After the transceiver is set to TX mode the start-up mode is active and the PLL is enabled. If the PLL is locked, the TX mode is active.

If the transceiver is in start-up or TX mode, the TX/RX data buffer can be loaded via the 4-wire serial interface. After the first byte is in the buffer and the TX mode is active, the transceiver starts transmitting automatically (beginning with the MSB). While transmitting it is permanently possible to load new data in the TX/RX data buffer. To prevent a buffer overflow or interruptions during transmitting the user must ensure that data is loaded at the same speed as it is transmitted.

There is a counter that indicates the number of bytes to be transmitted (see section “Transceiver Configuration”). If a byte is loaded, the counter is incremented, if a byte is transmitted, the counter is decremented. The counter value is available via the 4-wire serial interface. An IRQ is issued, if the counter while counting backwards reaches the value defined by the control bits IR0 and IR1 in control register 1.

Note: Writing to the control register 1, 4, 5 or 6 during TX mode, resets the TX/RX data buffer and the counter which indicates the number of bytes to be transmitted.

If T_Mode in control register 1 is set to 1, the transceiver is in TX transparent mode. In this mode the TX/RX data buffer is disabled and the TX data stream must be applied on pin SDI_TMDI. Figure 50 on page 59 illustrates the flow chart of the TX transparent mode.

Figure 50. TX Operation (T_MODE = 0)

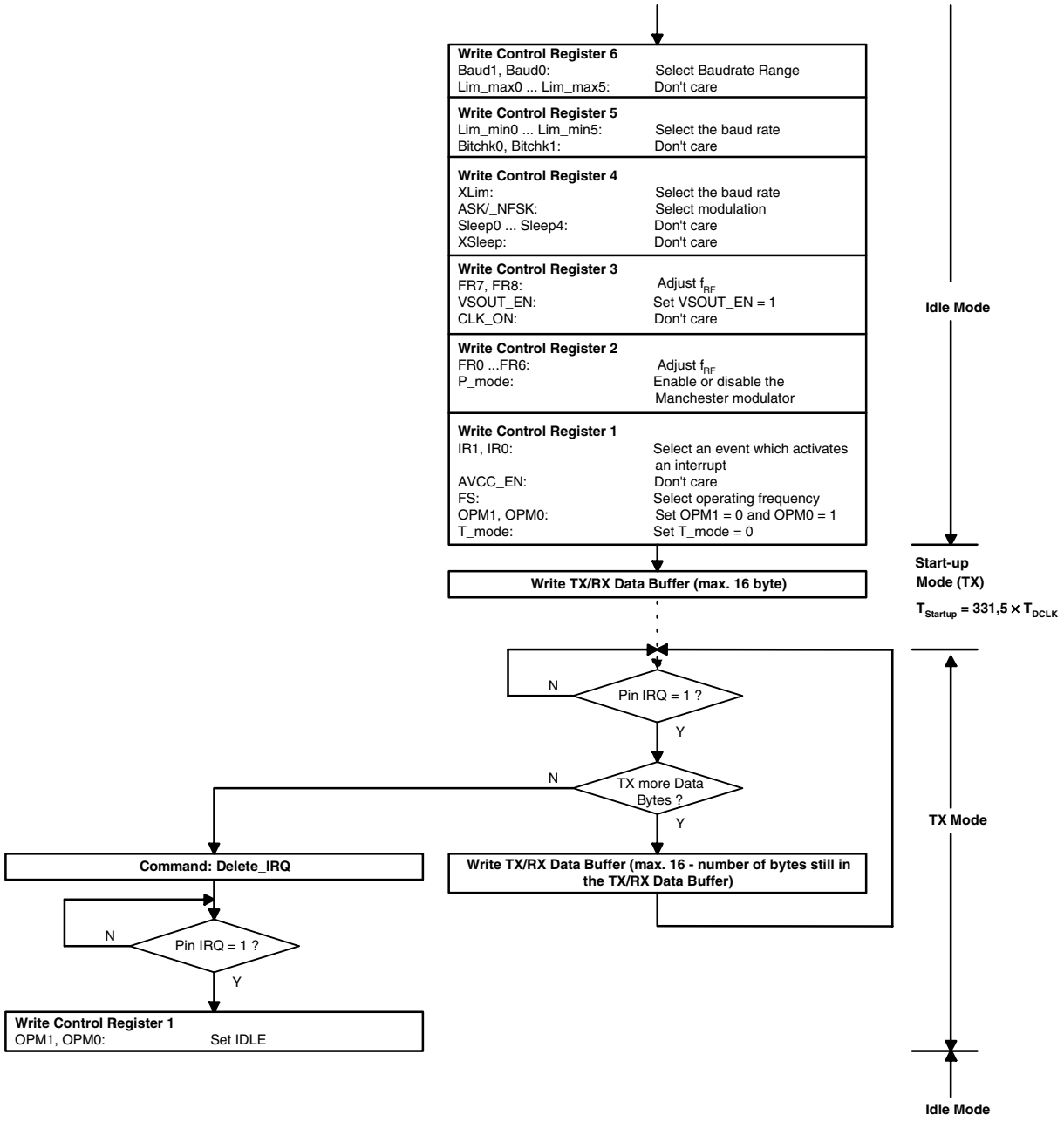


Figure 51. TX Transparent Mode (T_MODE = 1)

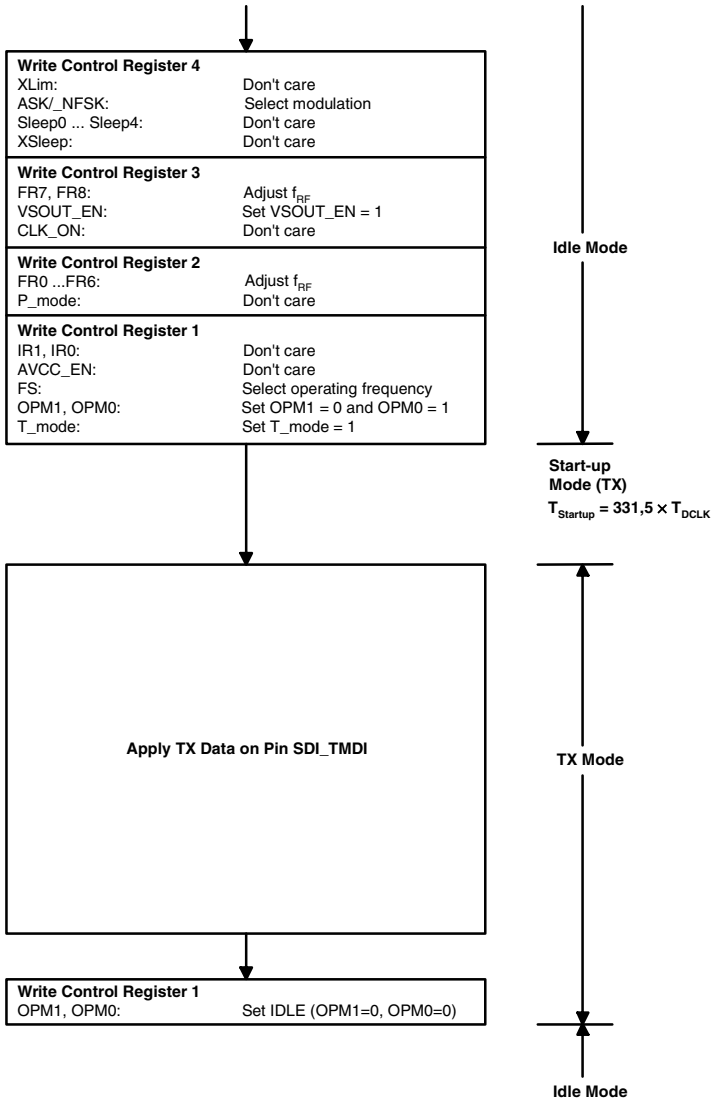


Table 41. TX Modulation Schemes

Mode	ASK/_NFSK	P_Mode	T_Mode	Bit in TX/RX Data Buffer	Level on Pin SDI_TMDI	RF_{OUT}
TX	0	0	0	1	X	$f_{FSK_L} \rightarrow f_{FSK_H}$
		0	0	0	X	$f_{FSK_H} \rightarrow f_{FSK_L}$
		1	0	1	X	f_{FSK_H}
		1	0	0	X	f_{FSK_L}
		X	1	X	1	f_{FSK_H}
		X	1	X	0	f_{FSK_L}
	1	0	0	1	X	f_{ASK} off $\rightarrow f_{ASK}$ on
		0	0	0	X	f_{ASK} on $\rightarrow f_{ASK}$ off
		1	0	1	X	f_{ASK} on
		1	0	0	X	f_{ASK} off
		X	1	X	1	f_{ASK} on
		X	1	X	0	f_{ASK} off

Interrupts

Via pin IRQ, the transceiver signals different operating conditions to a connected microcontroller. If a specific operating condition occurs, pin IRQ is set to high level.

If an interrupt occurs it is recommended to delete the interrupt be immediately deleted by reading the status register, thus the next possible interrupt doesn't get lost. If the Interrupt pin doesn't switch to low level by reading the status register the interrupt was triggered by the RX/TX data buffer. In this case read or write the RX/TX data buffer according to Table 42.

Table 42. Interrupt Handling

Operating Conditions Which Sets Pin IRQ to High Level	Operations Which Sets Pin IRQ to Low Level
Events in Status Register	
State transition of status bit STn (0 →1; 1 →0)	Read status register or Command Delete IRQ
Appearance of status bit Power_On (0 →1)	
Appearance of status bit P_On_Aux (0 →1)	
Events During TX Operation (T_MODE = 0)	
4, 8 or 12 Bytes are in the TX data buffer or the TX data buffer is empty (depends on IR0 and IR1 in control register 1).	Write TX data buffer or Write control register 1 or Write control register 4 or Write control register 5 or Write control register 6 or Command delete IRQ
Events During RX Operation (T_MODE = 0)	
4, 8 or 12 received bytes are in the RX data buffer or a receiving error is occurred (depends on IR0 and IR1 in control register 1).	Read RX data buffer or Write control register 1 or Write control register 4 or Write control register 5 or Write control register 6 or Command delete IRQ
Successful Bit-check (P_MODE = 0)	

Absolute Maximum Ratings

Stresses beyond those listed under “Absolute Maximum Ratings” may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Parameters	Symbol	Min.	Max.	Unit
Junction temperature	T_j		150	°C
Storage temperature	T_{stg}	-55	+125	°C
Ambient temperature	T_{amb}	-40	+105	°C
Supply voltage VS2	V_{MaxVS2}	-0.3	+7.2	V
Supply voltage VS1	V_{MaxVS1}	-0.3	+4	V
Supply voltage VAUX	$V_{MaxVAUX}$	-0.3	+7.2	V
Supply voltage VSINT	$V_{MaxVSINT}$	-0.3	+5.5	V
ESD (Human Body Model ESD S 5.1) every pin	HBM	-2	+2	kV
ESD (Machine Model JEDEC A115A) every pin	MM	-200	+200	V
Maximum input level, input matched to 50 Ω	P_{in_max}		10	dBm

Thermal Resistance

Parameters	Symbol	Value	Unit
Junction ambient	R_{thJA}	25	K/W

Electrical Characteristics: General

All parameters refer to GND and are valid for T_{amb} = -40°C to +105°C, $V_{VS1} = V_{VS2}$ = 2.4 V to 3.6 V (1-battery application), V_{VS2} = 4.4 V to 6.6 V (2-battery application) and $V_{VS2} = V_{VAUX}$ = 4.75 V to 5.25 V (car application). Typical values are given at $V_{VS1} = V_{VS2}$ = 3 V and T_{amb} = 25°C, f_{RF} = 433.92 MHz (1-battery application) unless otherwise specified. Details about current consumption, timing and digital pin properties can be found in the specific sections of the “Electrical Characteristics”.

No.	Parameters	Test Conditions	Pin(1)	Symbol	Min.	Typ.	Max.	Unit	Type*
1	**RX_TX_IDLE Mode**								
1.1	RF operating frequency range	ATA5811 V_{433_N868} = 0 V	4, 10	f_{RF}	867		870	MHz	A
		ATA5811 V_{433_N868} = AVCC	4, 10	f_{RF}	433		435	MHz	A
		ATA5812 V_{433_N868} = 0 V	4, 10	f_{RF}	313		316	MHz	A
1.2	Supply current OFF mode	$V_{VS1} = V_{VS2}$ = 3 V, V_{VSINT} = 0 V (1 battery) and V_{VS2} = 6 V (2 battery) OFF mode is not available if $V_{VS2} = V_{VAUX}$ = 5 V V_{VSINT} = 0 V (car)		I_{S_OFF}		<10		nA	A
1.3	Supply current Idle mode	V_{VSOUT} disabled, XTO running $V_{VS1} = V_{VS2}$ = 3 V (1 battery)		I_{S_IDLE}		220		µA	B
		V_{VS2} = 6 V (2 battery)		I_{S_IDLE}		310		µA	B
		$V_{VS2} = V_{VAUX}$ = 5 V (car)		I_{S_IDLE}		310		µA	B
1.4	System start-up time	From OFF mode to Idle mode including reset and XTO start-up (see Figure 30 on page 42) XTAL: C_m = 5 fF, C_0 = 1.8 pF, R_m =15 Ω		$T_{PWR_ON_IRQ_1}$		0.3		ms	C
1.5	RX start-up time	From Idle mode to receiving mode $N_{Bit\text{-}check}$ = 3 Baud rate = 20 kBaud, BR_Range_3 (see Figure 40 on page 50 , Figure 41 on page 51 and Figure 42 on page 52)		$T_{Startup_PLL}$ + $T_{Startup_Sig_Proc}$ + $T_{Bit\text{-}chek}$		1.39		ms	A
1.6	TX start-up time	From Idle mode to TX mode (see Figure 50 on page 59)		$T_{Startup}$		0.4		ms	A

*) Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter

Note: 1. Pin numbers in brackets mean they were measured with RF_IN matched to 50 Ω according to Figure 7 on page 10 with component values according to Table 2 on page 10 and RF_OUT matched to 50 Ω according to Figure 16 on page 19 with component values according to Table 7 on page 19.

Electrical Characteristics: General (Continued)

All parameters refer to GND and are valid for T_{amb} = -40°C to +105°C, V_{VS1} = V_{VS2} = 2.4 V to 3.6 V (1-battery application), V_{VS2} = 4.4 V to 6.6 V (2-battery application) and V_{VS2} = V_{VAUX} = 4.75 V to 5.25 V (car application). Typical values are given at V_{VS1} = V_{VS2} = 3 V and T_{amb} = 25°C, f_{RF} = 433.92 MHz (1-battery application) unless otherwise specified. Details about current consumption, timing and digital pin properties can be found in the specific sections of the “Electrical Characteristics”.

No.	Parameters	Test Conditions	Pin[1]	Symbol	Min.	Typ.	Max.	Unit	Type*
2	**Receiver/RX Mode**								
2.1	Supply current RX mode	f_{RF} = 433.92 MHz and f_{RF} = 315 MHz	17, 18	I_{S_RX}		10.5		mA	A
		f_{RF} = 868 MHz	17, 18	I_{S_RX}		10.3		mA	A
2.2	Supply current RX polling mode	T_{Sleep} = 49.45 ms X_{SLEEP} = 8, Sleep = 5 Baud rate = 20 kBaud FSK, V_{VSOUT} disabled	17, 18	I_P		444		µA	B
2.3	Input sensitivity FSK f_{RF} = 433.92 MHz	FSK deviation f_{DEV} = ±16 kHz limits according to Table 39 on page 57, BER = 10^{-3} T_{amb} = 25°C							
		Baud rate 20 kBaud	(4)	P_{REF_FSK}	-104.0	-106.0	-107.5	dBm	B
		Baud rate 2.4 kBaud	(4)	P_{REF_FSK}	-107.5	-109.5	-111.0	dBm	B
2.4	Input sensitivity ASK f_{RF} = 433.92 MHz	ASK 100%, level of carrier limits according to Table 39 on page 57, BER = 10^{-3} T_{amb} = 25°C							
		Baud rate 10 kBaud	(4)	P_{REF_ASK}	-110.5	-112.5	-114.0	dBm	B
		Baud rate 2.4 kBaud	(4)	P_{REF_ASK}	-114.5	-116.5	-118.0	dBm	B
2.5	Sensitivity change at f_{RF} = 315.0 MHz f_{RF} = 868.3 MHz compared to f_{RF} = 433.92 MHz	f_{RF} = 433.92 MHz to f_{RF} = 315.00 MHz f_{RF} = 433.92 MHz to f_{RF} = 868.00 MHz $P = P_{REF_ASK} + \Delta P_{REF1} + \Delta P_{REF2}$ $P = P_{REF_FSK} + \Delta P_{REF1} + \Delta P_{REF2}$	(4)	ΔP_{REF1}		-1.0 +2.7		dB	B
2.6	Maximum frequency offset in FSK mode	Maximum frequency difference of f_{RF} between receiver and transmitter in FSK mode (f_{RF} is the center frequency of the FSK signal with f_{DEV} = ±16 kHz)	(4)	Δf_{OFFSET}	-58		+58	kHz	B

*) Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter

Note: 1. Pin numbers in brackets mean they were measured with RF_IN matched to 50 Ω according to Figure 7 on page 10 with component values according to Table 2 on page 10 and RF_OUT matched to 50 Ω according to Figure 16 on page 19 with component values according to Table 7 on page 19.

Electrical Characteristics: General (Continued)

All parameters refer to GND and are valid for T_{amb} = -40°C to +105°C, $V_{VS1} = V_{VS2}$ = 2.4 V to 3.6 V (1-battery application), V_{VS2} = 4.4 V to 6.6 V (2-battery application) and $V_{VS2} = V_{VAUX}$ = 4.75 V to 5.25 V (car application). Typical values are given at $V_{VS1} = V_{VS2}$ = 3 V and T_{amb} = 25°C, f_{RF} = 433.92 MHz (1-battery application) unless otherwise specified. Details about current consumption, timing and digital pin properties can be found in the specific sections of the “Electrical Characteristics”.

No.	Parameters	Test Conditions	Pin(1)	Symbol	Min.	Typ.	Max.	Unit	Type*
2.7	Sensitivity change versus temperature, supply voltage and frequency offset	FSK f_{DEV} = ±16 kHz $\Delta f_{OFFSET} \leq$ ±58 kHz ASK 100% $\Delta f_{OFFSET} \leq$ 58 kHz $P = P_{REF_ASK} + \Delta P_{REF1} + \Delta P_{REF2}$ $P = P_{REF_FSK} + \Delta P_{REF1} + \Delta P_{REF2}$	(4)	ΔP_{REF2}	+4.5		-1.5		B
2.8	Supported FSK frequency deviation	With up to 2 dB loss of sensitivity. Note that the tolerable frequency offset is for f_{DEV} = ±22 kHz, 6 kHz lower than for f_{DEV} = ±16 kHz hence $\Delta f_{OFFSET} \leq$ ±52 kHz	(4)	f_{DEV}	±14	±16	±22	kHz	B
2.9	System noise figure	f_{RF} = 315 MHz	(4)	NF		6.0		dB	B
		f_{RF} = 433.92 MHz	(4)	NF		7.0		dB	B
		f_{RF} = 868.3 MHz	(4)	NF		9.7		dB	B
2.10	Intermediate frequency	f_{RF} = 868.3 MHz		f_{IF}		226		kHz	A
		f_{RF} = 433.92 MHz		f_{IF}		223		kHz	A
		f_{RF} = 315 MHz		f_{IF}		227		kHz	A
2.11	System bandwidth	This value is for information only! Note that for crystal and system frequency offset calculations, Δf_{OFFSET} must be used.	(4)	SBW		185		kHz	A
2.12	System outband 2nd-order input intercept point with respect to f_{IF}	Δf_{meas1} = 1,800 MHz Δf_{meas2} = 2,026 MHz $f_{IF} = \Delta f_{meas2} - \Delta f_{meas1}$	(4)	IIP2		+50		dBm	C
2.13	System outband 3rd-order input intercept point	Δf_{meas1} = 1.8 MHz Δf_{meas2} = 3.6 MHz f_{RF} = 315 MHz	(4)	IIP3		-22		dBm	C
		f_{RF} = 433.92 MHz	(4)	IIP3		-21		dBm	C
		f_{RF} = 868.3 MHz	(4)	IIP3		-17		dBm	C

*) Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter

Note: 1. Pin numbers in brackets mean they were measured with RF_IN matched to 50 Ω according to Figure 7 on page 10 with component values according to Table 2 on page 10 and RF_OUT matched to 50 Ω according to Figure 16 on page 19 with component values according to Table 7 on page 19.

Electrical Characteristics: General (Continued)

All parameters refer to GND and are valid for T_{amb} = -40°C to +105°C, $V_{VS1} = V_{VS2}$ = 2.4 V to 3.6 V (1-battery application), V_{VS2} = 4.4 V to 6.6 V (2-battery application) and $V_{VS2} = V_{VAUX}$ = 4.75 V to 5.25 V (car application). Typical values are given at $V_{VS1} = V_{VS2}$ = 3 V and T_{amb} = 25°C, f_{RF} = 433.92 MHz (1-battery application) unless otherwise specified. Details about current consumption, timing and digital pin properties can be found in the specific sections of the "Electrical Characteristics".

No.	Parameters	Test Conditions	Pin[1]	Symbol	Min.	Typ.	Max.	Unit	Type*
2.14	System outband input 1 dB compression point	Δf_{meas1} = 10 MHz f_{RF} = 315 MHz	(4)	I1dBCP		-31		dBm	C
		f_{RF} = 433.92 MHz	(4)	I1dBCP		-30		dBm	C
		f_{RF} = 868.3 MHz	(4)	I1dBCP		-27		dBm	C
2.15	LNA input impedance	f_{RF} = 315 MHz	4	Z_{in_LNA}		(44 – j233)		Ω	C
		f_{RF} = 433.92 MHz	4	Z_{in_LNA}		(32 – j169)		Ω	C
		f_{RF} = 868.3 MHz	4	Z_{in_LNA}		(21 – j78)		Ω	C
2.16	Maximum peak RF input level, ASK and FSK	BER < 10^{-3}, ASK: 100%	(4)	P_{IN_max}	-10	+10		dBm	C
		FSK: f_{DEV} = ±16 kHz	(4)	P_{IN_max}	-10	+10		dBm	C
2.17	LO spurious at LNA_IN	f < 1 GHz	(4)				-57	dBm	C
		f >1 GHz	(4)				-47	dBm	C
		f_{RF} = 315 MHz	(4)			-100		dBm	C
		f_{RF} = 433.92 MHz	(4)			-97		dBm	C
		f_{RF} = 868.3 MHz	(4)			-84		dBm	C
2.18	Image rejection	Within the complete image band	(4)		20	30		dB	A
2.19	Useful signal to interferer ratio	Peak level of useful signal to peak level of interferer for BER < 10^{-3} with any modulation scheme of interferer							
		FSK BR_Ranges 0, 1, 2	(4)	SNR_{FSK0-2}		2	3	dB	B
		FSK BR_Range_3	(4)	SNR_{FSK3}		4	6	dB	B
		ASK ($P_{RF} < P_{RFIN_High}$)	(4)	SNR_{ASK}		10	12	dB	B
2.20	RSSI output	Dynamic range	(4), 36	D_{RSSI}		70		dB	A
		Lower level of range f_{RF} = 315 MHz f_{RF} = 433.92 MHz f_{RF} = 868.3 MHz	(4), 36	P_{RFIN_Low}		 -116 -115 -112.3		 dBm dBm dBm	A
		Upper level of range f_{RF} = 315 MHz f_{RF} = 433.92 MHz f_{RF} = 868.3 MHz	(4), 36	P_{RFIN_High}		 -46 -45 -42.3		 dBm dBm dBm	A
		Gain	(4), 36		5.5	8.0	10.5	mV/dB	A
		Output voltage range	(4), 36	OV_{RSSI}	400		1100	mV	A
2.21	Output resistance RSSI pin	RX mode TX mode	36	R_{RSSI}	8 32	10 40	12.5 50	kΩ	C

*) Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter

Note: 1. Pin numbers in brackets mean they were measured with RF_IN matched to 50 Ω according to Figure 7 on page 10 with component values according to Table 2 on page 10 and RF_OUT matched to 50 Ω according to Figure 16 on page 19 with component values according to Table 7 on page 19.

Electrical Characteristics: General (Continued)

All parameters refer to GND and are valid for T_{amb} = -40°C to +105°C, $V_{VS1} = V_{VS2}$ = 2.4 V to 3.6 V (1-battery application), V_{VS2} = 4.4 V to 6.6 V (2-battery application) and $V_{VS2} = V_{VAUX}$ = 4.75 V to 5.25 V (car application). Typical values are given at $V_{VS1} = V_{VS2}$ = 3 V and T_{amb} = 25°C, f_{RF} = 433.92 MHz (1-battery application) unless otherwise specified. Details about current consumption, timing and digital pin properties can be found in the specific sections of the “Electrical Characteristics”.

No.	Parameters	Test Conditions	Pin(1)	Symbol	Min.	Typ.	Max.	Unit	Type*
2.22	Blocking	Sensitivity (BER = 10^{-3}) is reduced by 6 dB if a continuous wave blocking signal at ±Δf is ΔP_{Block} higher than the useful signal level (baud rate = 20 kBaud, FSK, f_{DEV} ±16kHz, Manchester code)							
		f_{RF} = 315 MHz Δf ± 0.75 MHz Δf ± 1.0 MHz Δf ± 1.5 MHz Δf ± 5 MHz Δf ± 10 MHz	(4)	ΔP_{Block}		56 60 63 69 71		dBC	C
		f_{RF} = 433.92 MHz Δf ± 0.75 MHz Δf ± 1.0 MHz Δf ± 1.5 MHz Δf ± 5 MHz Δf ± 10 MHz	(4)	ΔP_{Block}		55 59 62 68 70		dBC	C
		f_{RF} = 868.3 MHz Δf ± 0.75 MHz Δf ± 1.0 MHz Δf ± 1.5 MHz Δf ± 5 MHz Δf ± 10 MHz	(4)	ΔP_{Block}		50 53 57 67 69		dBC	C
2.23	CDEM	C_6 in Figure 4 on page 6, Figure 5 on page 7 and Figure 6 on page 8	37		-5%	15	+5%	nF	D
3	**Power Amplifier/TX Mode**								
3.1	Supply current TX mode power amplifier OFF	f_{RF} = 868.3 MHz		$I_{S_TX_PAOFF}$		6.50		mA	A
		f_{RF} = 433.92 MHz and f_{RF} = 315 MHz		$I_{S_TX_PAOFF}$		6.95		mA	A

*) Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter

Note: 1. Pin numbers in brackets mean they were measured with RF_IN matched to 50 Ω according to Figure 7 on page 10 with component values according to Table 2 on page 10 and RF_OUT matched to 50 Ω according to Figure 16 on page 19 with component values according to Table 7 on page 19.

Electrical Characteristics: General (Continued)

All parameters refer to GND and are valid for T_{amb} = -40°C to +105°C, $V_{VS1} = V_{VS2}$ = 2.4 V to 3.6 V (1-battery application), V_{VS2} = 4.4 V to 6.6 V (2-battery application) and $V_{VS2} = V_{VAUX}$ = 4.75 V to 5.25 V (car application). Typical values are given at $V_{VS1} = V_{VS2}$ = 3 V and T_{amb} = 25°C, f_{RF} = 433.92 MHz (1-battery application) unless otherwise specified. Details about current consumption, timing and digital pin properties can be found in the specific sections of the “Electrical Characteristics”.

No.	Parameters	Test Conditions	Pin(1)	Symbol	Min.	Typ.	Max.	Unit	Type*
3.2	Output power 1	$V_{VS1} = V_{VS2}$ = 3 V T_{amb} = 25°C V_{PWR_H} = 0 V f_{RF} = 315 MHz R_{R_PWR} = 56 kΩ R_{Lopt} = 2.5 kΩ f_{RF} = 433.92 MHz R_{R_PWR} = 56 kΩ R_{Lopt} = 2.3 kΩ f_{RF} = 868.3 MHz R_{R_PWR} = 30 kΩ R_{Lopt} = 1.3 kΩ RF_OUT matched to R_{Lopt} // j/(2 × π × f_{RF} × 1.0 pF)	(10)	P_{REF1}	-2.5	0	+2.5	dBm	B
3.3	Supply current TX mode power amplifier ON 1	PA on/0 dBm f_{RF} = 315 MHz	17, 18	$I_{S_TX_PAON1}$		8.5		mA	B
		f_{RF} = 433.92 MHz	17, 18	$I_{S_TX_PAON1}$		8.6		mA	B
		f_{RF} = 868.3 MHz	17, 18	$I_{S_TX_PAON1}$		9.6		mA	B
3.4	Output power 2	$V_{VS1} = V_{VS2}$ = 3 V T_{amb} = 25°C V_{PWR_H} = 0 V f_{RF} = 315 MHz R_{R_PWR} = 30 kΩ R_{Lopt} = 1.0 kΩ f_{RF} = 433.92 MHz R_{R_PWR} = 27 kΩ R_{Lopt} = 1.1 kΩ f_{RF} = 868.3 MHz R_{R_PWR} = 16 kΩ R_{Lopt} = 0.5 kΩ RF_OUT matched to R_{Lopt}// j/(2 × π × f_{RF} × 1.0 pF)	(10)	P_{REF2}	3.5	5.0	6.5	dBm	B

*) Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter

Note: 1. Pin numbers in brackets mean they were measured with RF_IN matched to 50 Ω according to Figure 7 on page 10 with component values according to Table 2 on page 10 and RF_OUT matched to 50 Ω according to Figure 16 on page 19 with component values according to Table 7 on page 19.

Electrical Characteristics: General (Continued)

All parameters refer to GND and are valid for T_{amb} = -40°C to +105°C, $V_{VS1} = V_{VS2}$ = 2.4 V to 3.6 V (1-battery application), V_{VS2} = 4.4 V to 6.6 V (2-battery application) and $V_{VS2} = V_{VAUX}$ = 4.75 V to 5.25 V (car application). Typical values are given at $V_{VS1} = V_{VS2}$ = 3 V and T_{amb} = 25°C, f_{RF} = 433.92 MHz (1-battery application) unless otherwise specified. Details about current consumption, timing and digital pin properties can be found in the specific sections of the “Electrical Characteristics”.

No.	Parameters	Test Conditions	Pin(1)	Symbol	Min.	Typ.	Max.	Unit	Type*
3.5	Supply current TX mode power amplifier ON 2	PA on/5 dBm f_{RF} = 315 MHz	17, 18	$I_{S_TX_PAON2}$		10.3		mA	B
		f_{RF} = 433.92 MHz	17, 18	$I_{S_TX_PAON2}$		10.5		mA	B
		f_{RF} = 868.3 MHz	17, 18	$I_{S_TX_PAON2}$		11.2		mA	B
3.6	Output power 3	$V_{VS1} = V_{VS2}$ = 3 V T_{amb} = 25°C V_{PWR_H} = AVCC f_{RF} = 315 MHz R_{R_PWR} = 30 kΩ R_{Lopt} = 0.38 kΩ f_{RF} = 433.92 MHz R_{R_PWR} = 27 kΩ R_{Lopt} = 0.36 kΩ f_{RF} = 868.3 MHz R_{R_PWR} = 20 kΩ R_{Lopt} = 0.22 kΩ RF_OUT matched to R_{Lopt}// j/(2 × π × f_{RF} × 1.0 pF)	(10)	P_{REF3}	8.5	10	11.5	dBm	B
3.7	Supply current TX mode power amplifier ON 3	PA on/10dBm f_{RF} = 315 MHz	17, 18	$I_{S_TX_PAON3}$		15.7		mA	B
		f_{RF} = 433.92 MHz	17, 18	$I_{S_TX_PAON3}$		15.8		mA	B
		f_{RF} = 868.3 MHz	17, 18	$I_{S_TX_PAON3}$		17.3		mA	B
3.8	Output power variation for full temperature and supply voltage range	T_{amb} = -40°C to +105°C $P_{out} = P_{REFX} + \Delta P_{REFX}$ X = 1, 2 or 3 $V_{VS1} = V_{VS2}$ = 3.0 V	(10)	ΔP_{REF}		-0.8	-1.5	dB	B
		$V_{VS1} = V_{VS2}$ = 2.4 V	(10)	ΔP_{REF}			-3.5	dB	B
		$V_{VS1} = V_{VS2}$ = 2.7 V	(10)	ΔP_{REF}			-2.5	dB	C

*) Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter

Note: 1. Pin numbers in brackets mean they were measured with RF_IN matched to 50 Ω according to Figure 7 on page 10 with component values according to Table 2 on page 10 and RF_OUT matched to 50 Ω according to Figure 16 on page 19 with component values according to Table 7 on page 19.

Electrical Characteristics: General (Continued)

All parameters refer to GND and are valid for T_{amb} = -40°C to +105°C, $V_{VS1} = V_{VS2}$ = 2.4 V to 3.6 V (1-battery application), V_{VS2} = 4.4 V to 6.6 V (2-battery application) and $V_{VS2} = V_{VAUX}$ = 4.75 V to 5.25 V (car application). Typical values are given at $V_{VS1} = V_{VS2}$ = 3 V and T_{amb} = 25°C, f_{RF} = 433.92 MHz (1-battery application) unless otherwise specified. Details about current consumption, timing and digital pin properties can be found in the specific sections of the "Electrical Characteristics".

No.	Parameters	Test Conditions	Pin(1)	Symbol	Min.	Typ.	Max.	Unit	Type*
3.9	Impedance RF_OUT in RX mode	f_{RF} = 315 MHz	10	$Z_{RF_OUT_RX}$		(36 –j502)		Ω	C
		f_{RF} = 433.92 MHz	10	$Z_{RF_OUT_RX}$		(19 –j366)		Ω	C
		f_{RF} = 868.3 MHz	10	$Z_{RF_OUT_RX}$		(2.8 –j141)		Ω	C
3.10	Noise floor power amplifier	at ±10 MHz/at 5 dBm f_{RF} = 868.3 MHz	(10)	L_{TX10M}		-125		dBC/Hz	C
		at f_{RF} = 433.92 MHz	(10)	L_{TX10M}		-126		dBC/Hz	C
		f_{RF} = 315 MHz	(10)	L_{TX10M}		-127		dBC/Hz	C
3.11	ASK modulation rate	This correspond to 10 kBaud Manchester coding and 20 kBaud NRZ coding		f_{Data_ASK}			10	kHz	C
4	**XTO**								
4.1	Pulling XTO due to XTO, C_{L1} and C_{L2} tolerances	Pulling at nominal temperature and supply voltage f_{XTAL} = resonant frequency of the XTAL $C_0 \geq$ 1.5 pF $R_m \leq$120 Ω $C_m \leq$7.0 fF $C_m \leq$14 fF	24, 25	Δf_{XTO1}	 -50 -100	 f_{XTAL}	 +50 +100	 ppm	A
4.2	Transconductance XTO at start	At start-up, after start-up the amplitude is regulated to V_{PPXTAL}	24, 25	$g_{m,\ XTO}$		19		ms	B
4.3	XTO start-up time	$C_0 \leq$2.2 pF C_m = 4.0 fF to 7.0 fF $R_m \leq$120 Ω	24, 25	$T_{PWR_ON_IRQ_1}$		300	800	µs	A
4.4	Maximum C_0 of XTAL	Required for stable operation with internal load capacitors	24, 25	C_{0max}			3.8	pF	D
4.5	Internal capacitors	C_{L1} and C_{L2}	24, 25	C_{L1}, C_{L2}	14.8	18 pF	21.2	pF	B
4.6	Pulling of radio frequency f_{RF} due to XTO, C_{L1} and C_{L2} versus temperature and supply changes	1.5 pF $\leq C_0 \leq$2.2 pF C_m = 4.0 fF to 7.0 fF $R_m \leq$120 Ω PLL adjusted with FREQ at nominal temperature and supply voltage	4, 10	Δf_{XTO2}	-2		+2	ppm	C

*) Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter

Note: 1. Pin numbers in brackets mean they were measured with RF_IN matched to 50 Ω according to Figure 7 on page 10 with component values according to Table 2 on page 10 and RF_OUT matched to 50 Ω according to Figure 16 on page 19 with component values according to Table 7 on page 19.

Electrical Characteristics: General (Continued)

All parameters refer to GND and are valid for T_{amb} = -40°C to +105°C, $V_{VS1} = V_{VS2}$ = 2.4 V to 3.6 V (1-battery application), V_{VS2} = 4.4 V to 6.6 V (2-battery application) and $V_{VS2} = V_{VAUX}$ = 4.75 V to 5.25 V (car application). Typical values are given at $V_{VS1} = V_{VS2}$ = 3 V and T_{amb} = 25°C, f_{RF} = 433.92 MHz (1-battery application) unless otherwise specified. Details about current consumption, timing and digital pin properties can be found in the specific sections of the “Electrical Characteristics”.

No.	Parameters	Test Conditions	Pin(1)	Symbol	Min.	Typ.	Max.	Unit	Type*
4.7	Amplitude XTAL after start-up	C_m = 5 fF, C_0 = 1.8 pF R_m =15 Ω							
		V(XTAL1, XTAL2) peak-to-peak value	24, 25	V_{PPXTAL}		700		mVpp	C
		V(XTAL1) peak-to-peak value	24, 25	V_{PPXTAL}		350		mVpp	C
4.8	Maximum series resistance R_m of XTAL at start-up	$C_0 \leq$2.2 pF, start-up may take longer under these conditions	24, 25	Z_{XTAL12_START}	-1,500	-2,000		Ω	B
4.9	Maximum series resistance R_m of XTAL after start-up	$C_0 \leq$2.2 pF C_m = 4.0 fF to 7.0 fF $R_m \leq$20 Ω	24, 25	R_{m_max}		15	120	Ω	B
4.10	Nominal XTAL load resonant frequency	FREQ = 3,928 f_{RF} = 868.3 MHz f_{RF} = 433.92 MHz f_{RF} = 315 MHz	24, 25	f_{XTAL}		13.41191 13.25311 12.73193		MHz MHz	D
4.11	External CLK frequency	FREQ = 3,928	30	f_{CLK}	$f_{CLK} = \frac{f_{XTO}}{3}$			MHz	D
		f_{RF} = 868.3 MHz CLK division ratio = 3 CLK has nominal 50% duty cycle	30	f_{CLK}		4.471		MHz	D
		f_{RF} = 433.92 MHz CLK division ratio = 3 CLK has nominal 50% duty cycle	30	f_{CLK}		4.41 8		MHz	D
		f_{RF} = 315 MHz CLK division ratio = 3 CLK has nominal 50% duty cycle	30	f_{CLK}		4.244		MHz	D
4.12	DC voltage after start-up	V_{DC}(XTAL1, XTAL2) XTO running (Idle mode, RX mode and TX mode)	24, 25	V_{DCXTO}	-150	-30			mV

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Note: 1. Pin numbers in brackets mean they were measured with RF_IN matched to 50 Ω according to Figure 7 on page 10 with component values according to Table 2 on page 10 and RF_OUT matched to 50 Ω according to Figure 16 on page 19 with component values according to Table 7 on page 19.

Electrical Characteristics: General (Continued)

All parameters refer to GND and are valid for T_{amb} = -40°C to +105°C, $V_{VS1} = V_{VS2}$ = 2.4 V to 3.6 V (1-battery application), V_{VS2} = 4.4 V to 6.6 V (2-battery application) and $V_{VS2} = V_{VAUX}$ = 4.75 V to 5.25 V (car application). Typical values are given at $V_{VS1} = V_{VS2}$ = 3 V and T_{amb} = 25°C, f_{RF} = 433.92 MHz (1-battery application) unless otherwise specified. Details about current consumption, timing and digital pin properties can be found in the specific sections of the “Electrical Characteristics”.

No.	Parameters	Test Conditions	Pin[1]	Symbol	Min.	Typ.	Max.	Unit	Type*
5	**Synthesizer**								
5.1	Spurious TX mode	At $\pm f_{CLK}$, CLK enabled f_{RF} = 315 MHz f_{RF} = 433.92 MHz f_{RF} = 868.3 MHz		SP_{TX}		-72 -68 -70		dBC	A
		at $\pm f_{XTO}$ f_{RF} = 315 MHz f_{RF} = 433.92 MHz f_{RF} = 868.3 MHz		SP_{TX}		-70 -66 -60		dBC	A
5.2	Spurious RX mode	At $\pm f_{CLK}$, CLK enabled f_{RF} = 315 MHz f_{RF} = 433.92 MHz f_{RF} = 868.3 MHz		SP_{RX}		< -75 < -75 < -75		dBC	A
		at $\pm f_{XTO}$ f_{RF} = 315 MHz f_{RF} = 433.92 MHz f_{RF} = 868.3 MHz		SP_{RX}		-75 -75 -68		dBC	A
5.3	In loop phase noise TX mode	Measured at 20 kHz distance to carrier f_{RF} = 315 MHz f_{RF} = 433.92 MHz f_{RF} = 868.3 MHz		L_{TX20k}		-85 -80 -75		dBC/Hz	A
5.4	Phase noise at 1M RX mode	f_{RF} = 315 MHz f_{RF} = 433.92 MHz f_{RF} = 868.3 MHz		L_{RX1M}		-121 -120 -113		dBC/Hz	A
5.5	Phase noise at 1M TX mode	f_{RF} = 315 MHz f_{RF} = 433.92 MHz f_{RF} = 868.3 MHz		L_{TX1M}		-113 -111 -107		dBC/Hz	A
5.6	Phase noise at 10M RX mode	Noise floor PLL		L_{RX10M}		-135		dBC/Hz	B
5.7	Loop bandwidth PLL TX mode	Frequency where the absolute value loop gain is equal to 1		f_{Loop_PLL}		70		kHz	B
5.8	Frequency deviation TX mode	f_{RF} = 315 MHz f_{RF} = 433.92 MHz f_{RF} = 868.3 MHz		f_{DEV_TX}		±15.54 ±16.17 ±16.37		kHz	D
5.9	Frequency resolution	f_{RF} = 315 MHz f_{RF} = 433.92 MHz f_{RF} = 868.3 MHz	4, 10	Δf_{Step_PLL}		777.1 808.9 818.6		Hz	D

*) Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter

Note: 1. Pin numbers in brackets mean they were measured with RF_IN matched to 50 Ω according to Figure 7 on page 10 with component values according to Table 2 on page 10 and RF_OUT matched to 50 Ω according to Figure 16 on page 19 with component values according to Table 7 on page 19.

Electrical Characteristics: General (Continued)

All parameters refer to GND and are valid for T_{amb} = -40°C to +105°C, $V_{VS1} = V_{VS2}$ = 2.4 V to 3.6 V (1-battery application), V_{VS2} = 4.4 V to 6.6 V (2-battery application) and $V_{VS2} = V_{VAUX}$ = 4.75 V to 5.25 V (car application). Typical values are given at $V_{VS1} = V_{VS2}$ = 3 V and T_{amb} = 25°C, f_{RF} = 433.92 MHz (1-battery application) unless otherwise specified. Details about current consumption, timing and digital pin properties can be found in the specific sections of the “Electrical Characteristics”.

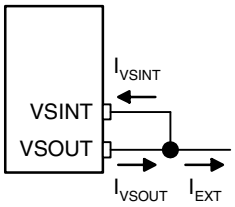
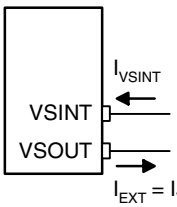
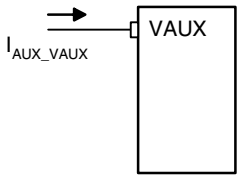
No.	Parameters	Test Conditions	Pin(1)	Symbol	Min.	Typ.	Max.	Unit	Type*
5.10	FSK modulation rate	This correspond to 20 kBaud Manchester coding and 40 kBaud NRZ coding		f_{Data_FSK}			20	kHz	B
6	**RX/TX Switch**								
6.1	Impedance RX mode	RX mode, pin 38 with short connection to GND, f_{RF} = 0 Hz (DC)	39	Z_{Switch_RX}		23000		Ω	A
		f_{RF} = 315 MHz	39	Z_{Switch_RX}		(11.3 – j214)		Ω	C
		f_{RF} = 433.92 MHz	39	Z_{Switch_RX}		(10.3 – j153)		Ω	C
		f_{RF} = 868.3 MHz	39	Z_{Switch_RX}		(8.9 – j73)		Ω	C
6.2	Impedance TX mode	TX mode, pin 38 with short connection to GND, f_{RF} = 0 Hz (DC)	39	Z_{Switch_TX}		5		Ω	A
		f_{RF} = 315 MHz f_{RF} = 433.92 MHz f_{RF} = 868.3 MHz	39	Z_{Switch_TX}		(4.8 + j3.2) (4.5 + j4.3) (5 + j9)		Ω	C C C
7	**Microcontroller Interface**								
7.1	Voltage range for microcontroller interface	I_{VSINT} < 10 µA if CLK is disabled and all interface pins are in stable condition and unloaded	27, 28, 29, 30, 31, 32, 33, 34, 35		2.4		5.25	V	A
7.2	CLK output rise and fall time	f_{CLK} < 4.5 MHz C_L = 10 pF C_L = Load capacitance on pin CLK 2.4 V ≤V_{VSINT} ≤5.25 V 20% to 80% V_{VSINT}	30	t_{rise} t_{fall}		20 20	30 30	ns ns	B

*) Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter

Note: 1. Pin numbers in brackets mean they were measured with RF_IN matched to 50 Ω according to Figure 7 on page 10 with component values according to Table 2 on page 10 and RF_OUT matched to 50 Ω according to Figure 16 on page 19 with component values according to Table 7 on page 19.

Electrical Characteristics: General (Continued)

All parameters refer to GND and are valid for T_{amb} = -40°C to +105°C, $V_{VS1} = V_{VS2}$ = 2.4 V to 3.6 V (1-battery application), V_{VS2} = 4.4 V to 6.6 V (2-battery application) and $V_{VS2} = V_{VAUX}$ = 4.75 V to 5.25 V (car application). Typical values are given at $V_{VS1} = V_{VS2}$ = 3 V and T_{amb} = 25°C, f_{RF} = 433.92 MHz (1-battery application) unless otherwise specified. Details about current consumption, timing and digital pin properties can be found in the specific sections of the “Electrical Characteristics”.

No.	Parameters	Test Conditions	Pin(1)	Symbol	Min.	Typ.	Max.	Unit	Type*
7.4	Current consumption of the microcontroller interface	CLK enabled V_{VSOUT} enabled CLK disabled V_{VSOUT} enabled V_{VSOUT} disabled C_L = Load capacitance on pin CLK (All interface pins, except pin CLK, are in stable condition and unloaded)	27	I_{VSINT}	$I_{VSINT} = \frac{(C_{CLK} + C_L) \times V_{VSINT} \times f_{XTO}}{3}$ < 10 µA < 10 µA				
7.5	Internal equivalent capacitance	Used for current calculation	30, 27	CCLK		8		pF	B
8	**Power Supply General Definitions and AUX Mode**								
8.1	Current consumption of an external device connected to pin VSOUT	 		I_{EXT}	$I_{EXT} = I_{VSOUT} - I_{VSINT}$ $I_{EXT} = I_{VSOUT}$				
8.2	AUX mode								

*) Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter

Note: 1. Pin numbers in brackets mean they were measured with RF_IN matched to 50 Ω according to Figure 7 on page 10 with component values according to Table 2 on page 10 and RF_OUT matched to 50 Ω according to Figure 16 on page 19 with component values according to Table 7 on page 19.

Electrical Characteristics: General (Continued)

All parameters refer to GND and are valid for T_{amb} = -40°C to +105°C, $V_{VS1} = V_{VS2}$ = 2.4 V to 3.6 V (1-battery application), V_{VS2} = 4.4 V to 6.6 V (2-battery application) and $V_{VS2} = V_{VAUX}$ = 4.75 V to 5.25 V (car application). Typical values are given at $V_{VS1} = V_{VS2}$ = 3 V and T_{amb} = 25°C, f_{RF} = 433.92 MHz (1-battery application) unless otherwise specified. Details about current consumption, timing and digital pin properties can be found in the specific sections of the “Electrical Characteristics”.

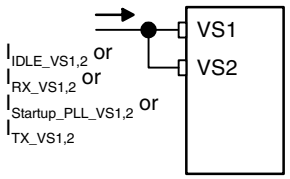
No.	Parameters	Test Conditions	Pin(1)	Symbol	Min.	Typ.	Max.	Unit	Type*
8.3	Power supply output voltage	AUX mode $V_{VAUX} \geq 4$ V $I_{VSOUT} \leq 13.5$ mA (3.25 V regulator mode, V_REG2, see Figure 21 on page 26)	22	V_{VSOUT}	2.7		3.5	V	A
8.4	Current in AUX mode on pin VAUX	I_{VSOUT} = 0 V_{VAUX} = 6 V V_{VAUX} = 4 to 7 V	19	I_{AUX_VAUX}		380	500 500	µA µA	B
8.5	Supply current AUX mode	CLK enabled V_{VSOUT} enabled CLK disabled V_{VSOUT} enabled	19, 22, 27	I_{S_AUX}	$I_{S_AUX} = I_{AUX_VAUX} + I_{VSINT} + I_{EXT}$ $I_{S_AUX} = I_{AUX_VAUX} + I_{EXT}$				
8.6	Supported voltage range VAUX		19	V_{VAUX}	4	6	7	V	

*) Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter

Note: 1. Pin numbers in brackets mean they were measured with RF_IN matched to 50 Ω according to Figure 7 on page 10 with component values according to Table 2 on page 10 and RF_OUT matched to 50 Ω according to Figure 16 on page 19 with component values according to Table 7 on page 19.

Electrical Characteristic: 1-Battery Application

All parameters refer to GND and are valid for T_{amb} = -40°C to +105°C, $V_{VS1} = V_{VS2}$ = 2.4 V to 3.6 V typical values at $V_{VS1} = V_{VS2}$ = 3 V and T_{amb} = 25°C. Application according to Figure 4 on page 6. f_{RF} = 315.0 MHz/433.92 MHz/868.3 MHz unless otherwise specified

No.	Parameters	Test Conditions	Pin	Symbol	Min.	Typ.	Max.	Unit	Type*
9	**1-Battery Application**								
9.1	Supported voltage range (every mode except high power TX mode)	1-battery application PWR_H = GND	17, 18	V_{VS1}, V_{VS2}	2.4		3.6	V	A
9.2	Supported voltage range (high power TX mode)	1-battery application PWR_H = AVCC	17, 18	V_{VS1}, V_{VS2}	2.7		3.6	V	A

*) Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter

Note: 1. The voltage of VAUX may rise up to 2 V. The current I_{VAUX} may not exceed 100 µA.

Electrical Characteristic: 1-Battery Application (Continued)

All parameters refer to GND and are valid for T_{amb} = -40°C to +105°C, V_{VS1} = V_{VS2} = 2.4 V to 3.6 V typical values at V_{VS1} = V_{VS2} = 3 V and T_{amb} = 25°C. Application according to Figure 4 on page 6. f_{RF} = 315.0 MHz/433.92 MHz/868.3 MHz unless otherwise specified

No.	Parameters	Test Conditions	Pin	Symbol	Min.	Typ.	Max.	Unit	Type*
9.3	Power supply output voltage	1-battery application V_{VS1} = V_{VS2} ≥ 2.6 V VAUX open (1) I_{VSOUT} ≤13.5 mA (no voltage regulator to stabilize V_{VSOUT}) V_{VS1} = V_{VS2} ≥ 2.425 V VAUX open (1) I_{VSOUT} ≤1.5 mA (no voltage regulator to stabilize V_{VSOUT})	22	V_{VSOUT}	2.4		V_{VS1}	V	B
9.4	Supply voltage for microcontroller interface		27	V_{VSINT}	2.4		5.25	V	A
9.5	Threshold hysteresis	V_{Thres_2} - V_{Thres_1}	22	ΔV_{Thres}	60	80	100	mV	B
9.6	Reset threshold voltage at pin VSOUT (N_RESET)		22	V_{Thres_1}	2.18	2.3	2.42	V	A
9.7	Reset threshold voltage at pin VSOUT (Low_Batt)		22	V_{Thres_2}	2.26	2.38	2.5	V	A
9.8	Supply current OFF mode	V_{VS1} = V_{VS2} ≤3.6 V V_{VSINT} = 0 V	17, 18, 22, 27	I_{S_OFF}		2	350	nA	A
9.9	Current in Idle mode on pin VS1 and VS2	V_{VS1} = V_{VS2} ≤3 V I_{VSOUT} = 0 CLK enabled V_{VSOUT} enabled CLK disabled V_{VSOUT} enabled V_{VSOUT} disabled	17, 18	$I_{IDLE_VS1,2}$		312 260 225	430 370 320	µA µA µA	A B B
9.10	Supply current Idle mode		17, 18, 22, 27	I_{S_IDLE}	$I_{S_IDLE} = I_{IDLE_VS1,2} + I_{VSINT} + I_{EXT}$				
9.11	Current in RX mode on pin VS1and VS2	V_{VS1} = V_{VS2} ≤3 V I_{VSOUT} = 0	17, 18	$I_{RX_VS1,2}$		10.5	14	mA	A
9.12	Supply current RX mode	CLK enabled V_{VSOUT} enabled	17, 18, 22, 27	I_{S_RX}	$I_{S_RX} = I_{RX_VS1,2} + I_{VSINT} + I_{EXT}$				

*) Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter

Note: 1. The voltage of VAUX may rise up to 2 V. The current I_{VAUX} may not exceed 100 µA.

Electrical Characteristic: 1-Battery Application (Continued)

All parameters refer to GND and are valid for T_{amb} = -40°C to +105°C, $V_{VS1} = V_{VS2}$ = 2.4 V to 3.6 V typical values at $V_{VS1} = V_{VS2}$ = 3 V and T_{amb} = 25°C. Application according to Figure 4 on page 6. f_{RF} = 315.0 MHz/433.92 MHz/868.3 MHz unless otherwise specified

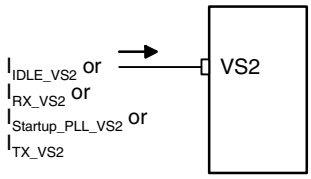
No.	Parameters	Test Conditions	Pin	Symbol	Min.	Typ.	Max.	Unit	Type*
9.13	Current during $T_{Startup_PLL}$ on pin VS1 and VS2	$V_{VS1} = V_{VS2} \leq 3$ V I_{VSOUT} = 0	17, 18	$I_{Startup_PLL_VS1,2}$		8.8	11.5	mA	C
9.14	Current in RX polling mode on pin VS1 and VS2	$I_P = \frac{I_{IDLE_VS1,2} \times T_{SLEEP} + I_{Startup_PLL_VS1,2} \times T_{Startup_PLL} + I_{RX_VS1,2} \times (T_{Startup_Sig_Proc} + T_{Bitcheck})}{T_{Sleep} + T_{Startup_PLL} + T_{Startup_Sig_Proc} + T_{Bitcheck}}$							
9.15	Supply current RX polling mode	CLK enabled V_{VSOUT} enabled CLK disabled V_{VSOUT} enabled V_{VSOUT} disabled	17, 18, 22, 27	I_{S_Poll}	$I_{S_Poll} = I_P + I_{VSINT} + I_{EXT}$ $I_{S_Poll} = I_P + I_{EXT}$ $I_{S_Poll} = I_P$				
9.16	Current in TX mode on pin VS1 and VS2	$V_{VS1} = V_{VS2} \leq 3$ V I_{VSOUT} = 0 Pout = 5 dBm/10 dBm 315 MHz/5 dBm 315 MHz/10 dBm 433.92 MHz/5 dBm 433.92 MHz/10 dBm 868.3 MHz/5 dBm 868.3 MHz/10 dBm	17, 18	$I_{TX_VS1_VS2}$		10.3 15.7 10.5 15.8 11.2 17.3	13.4 20.5 13.5 20.5 14.5 22.5	mA	B
9.17	Supply current TX mode	CLK enabled V_{VSOUT} enabled CLK disabled V_{VSOUT} enabled	17, 18, 22, 27	I_{S_TX}	$I_{S_TX} = I_{TX_VS1,2} + I_{VSINT} + I_{EXT}$ $I_{S_TX} = I_{TX_VS1,2} + I_{EXT}$				

*) Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter

Note: 1. The voltage of VAUX may rise up to 2 V. The current I_{VAUX} may not exceed 100 µA.

Electrical Characteristics: 2-Battery Application

All parameters refer to GND and are valid for T_{amb} = -40 °C to +105 °C, V_{VS2} = 4.4 V to 6.6 V typical values at V_{VS2} = 6V and T_{amb} = 25°C. Application according to Figure 6 on page 8. f_{RF} = 315.0MHz/433.92 MHz/868.3 MHz unless otherwise specified

No.	Parameters	Test Conditions	Pin	Symbol	Min.	Typ.	Max.	Unit	Type*
10	**2-Battery Application**								
10.1	Supported voltage range	2-battery application	17	V_{VS2}	4.4		6.6	V	A
10.2	Power supply output voltage	2 battery application $V_{VS2} \geq 4.4$ V VAUX open(1) $I_{VSOUT} \leq 13.5$ mA (3.3 V regulator mode, V_REG1, see Figure 21 on page 26)	22	V_{VSOUT}	3.0		3.5	V	A
10.3	Supply voltage for microcontroller interface		27	V_{VSINT}	2.4		5.25	V	A
10.4	Threshold hysteresis	V_{Thres_2} - V_{Thres_1}	22	ΔV_{Thres}	60	80	100	mV	B
10.5	Reset threshold voltage at pin VSOUT (N_RESET)		22	V_{Thres_1}	2.18	2.3	2.42	V	A
10.6	Reset threshold voltage at pin VSOUT (Low_Batt)		22	V_{Thres_2}	2.26	2.38	2.5	V	A
10.7	Supply current OFF mode	$V_{VS2} \leq 6.6$ V V_{VSINT} = 0 V	17, 22, 27	I_{S_OFF}		10	350	nA	A
10.8	Current in Idle mode on pin VS2	$V_{VS2} \leq 6$ V I_{VSOUT} = 0 CLK enabled V_{VSOUT} enabled CLK disabled V_{VSOUT} enabled V_{VSOUT} disabled	17	I_{IDLE_VS2}		410 348 309	560 490 430	µA µA µA	A B B
10.9	Supply current Idle mode		17, 22, 27	I_{S_IDLE}	$I_{S_IDLE} = I_{IDLE_VS2} + I_{VSINT} + I_{EXT}$				
10.10	Current in RX mode on pin VS2	I_{VSOUT} = 0	17	I_{RX_VS2}		10.8	14.5	mA	B

*) Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter

Note: 1. The voltage of VAUX may rise up to 2 V. The current I_{VAUX} may not exceed 100 µA.

Electrical Characteristics: 2-Battery Application (Continued)

All parameters refer to GND and are valid for T_{amb} = -40 °C to +105 °C, V_{VS2} = 4.4 V to 6.6 V typical values at V_{VS2} = 6V and T_{amb} = 25°C. Application according to Figure 6 on page 8. f_{RF} = 315.0MHz/433.92 MHz/868.3 MHz unless otherwise specified

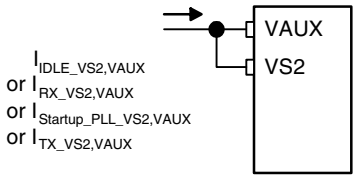
No.	Parameters	Test Conditions	Pin	Symbol	Min.	Typ.	Max.	Unit	Type*
10.11	Supply current RX mode	CLK enabled V_{VSOUT} enabled	17, 22, 27	I_{S_RX}	$I_{S_RX} = I_{RX_VS2} + I_{VSINT} + I_{EXT}$				
10.12	Current during $T_{Startup_PLL}$ on pin VS2	I_{VSOUT} = 0	17	$I_{Startup_PLL_VS2}$		9.1	12	mA	C
10.13	Current in RX polling mode on on pin VS2	$I_P = \frac{I_{IDLE_VS2} \times T_{SLEEP} + I_{Startup_PLL_VS2} \times T_{Startup_PLL} + I_{RX_VS2} \times (T_{Startup_Sig_Proc} + T_{Bitcheck})}{T_{Sleep} + T_{Startup_PLL} + T_{Startup_Sig_Proc} + T_{Bitcheck}}$							
10.14	Supply current RX polling mode	CLK enabled V_{VSOUT} enabled CLK disabled V_{VSOUT} enabled V_{VSOUT} disabled	17, 22, 27	I_{S_Poll}	$I_{S_Poll} = I_P + I_{VSINT} + I_{EXT}$ $I_{S_Poll} = I_P + I_{EXT}$ $I_{S_Poll} = I_P$				
10.15	Current in TX mode on pin VS2	I_{VSOUT} = 0 P_{out} = 5 dBm/10 dBm 315 MHz/5 dBm 315 MHz/10 dBm 433.92 MHz/5 dBm 433.92 MHz/10 dBm 868.3 MHz/5 dBm 868.3 MHz/10 dBm	17, 19	I_{TX_VS2}		10.7 16.2 10.9 16.3 11.6 17.8	13.9 21.0 14.0 21.0 15.0 23.0	mA	B
10.16	Supply current TX mode	CLK enabled V_{VSOUT} enabled CLK disabled V_{VSOUT} enabled	17, 22, 27	I_{S_TX}	$I_{S_TX} = I_{TX_VS2} + I_{VSINT} + I_{EXT}$ $I_{S_TX} = I_{TX_VS2} + I_{EXT}$				

*) Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter

Note: 1. The voltage of VAUX may rise up to 2 V. The current I_{VAUX} may not exceed 100 µA.

Electrical Characteristics: Car Application

All parameters refer to GND and are valid for T_{amb} = -40°C to +105°C, V_{VS2} = 4.75 V to 5.25 V. Typical values at V_{VS2} = 5 V and T_{amb} = 25°C. Application according to Figure 5 on page 7. f_{RF} = 315.0 MHz/433.92 MHz/868.3 MHz unless otherwise specified

No.	Parameters	Test Conditions	Pin	Symbol	Min.	Typ.	Max.	Unit	Type*
11	**Car Application**								
11.1	Supported voltage range	Car application	17, 19, 27	V_{VS2}, V_{AUX}	4.75		5.25	V	A
11.2	Power supply output voltage	Car application $V_{VS2} = V_{VAUX}$ $I_{VSOUT} \leq 13.5$ mA (3.25 V regulator mode, V_REG2, see Figure 21 on page 26)	22	V_{VSOUT}	3.0		3.5	V	A
11.3	Supply voltage for microcontroller-interface		27	V_{VSINT}	2.4		5.25	V	A
11.4	Threshold hysteresis	$V_{Thres_2} - V_{Thres_1}$	22	ΔV_{Thres}	60	80	100	mV	B
11.5	Reset threshold voltage at pin VSOUT (N_RESET)		22	V_{Thres_1}	2.18	2.3	2.42	V	A
11.6	Reset threshold voltage at pin VSOUT (Low_Batt)		22	V_{Thres_2}	2.26	2.38	2.5	V	A
11.7	Current in Idle mode on pin VS2 and VAUX	I_{VSOUT} = 0 CLK enabled V_{VSOUT} enabled	17, 19	$I_{IDLE_VS2_VAUX}$		444	580	µA	B
		CLK disabled V_{VSOUT} enabled				380	500	µA	B
		V_{VSOUT} disabled				310	400	µA	B
11.8	Supply current in Idle mode		17, 19, 22, 27	I_{S_IDLE}	$I_{S_IDLE} = I_{IDLE_VS2_VAUX} + I_{VSINT} + I_{EXT}$				
11.9	Current in RX mode on pin VS2 and VAUX	I_{VSOUT} = 0	17, 19	$I_{RX_VS2_VAUX}$		10.8	14.5	mA	B
11.10	Supply current in RX mode	CLK enabled V_{VSOUT} enabled	17, 19, 22, 27	I_{S_RX}	$I_{S_RX} = I_{RX_VS2_VAUX} + I_{VSINT} + I_{EXT}$				

*) Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter

Electrical Characteristics: Car Application (Continued)

All parameters refer to GND and are valid for T_{amb} = -40°C to +105°C, V_{VS2} = 4.75 V to 5.25 V. Typical values at V_{VS2} = 5 V and T_{amb} = 25°C. Application according to Figure 5 on page 7. f_{RF} = 315.0 MHz/433.92 MHz/868.3 MHz unless otherwise specified

No.	Parameters	Test Conditions	Pin	Symbol	Min.	Typ.	Max.	Unit	Type*
11.11	Current during $T_{Startup_PLL}$ on pin VS2 and VAUX	I_{VSOUT} = 0	17, 19	$I_{Startup_PLL_VS2_VAUX}$		9.1	12	mA	C
11.12	Current in RX_Polling_Mode on pin VS2 and VAUX $I_P = \frac{I_{IDLE_VS2,VAUX} \times T_{SLEEP} + I_{Startup_PLL_VS2,VAUX} \times T_{Startup_PLL} + I_{RX_VS2,VAUX} \times (T_{Startup_Sig_Proc} + T_{Bitcheck})}{T_{Sleep} + T_{Startup_PLL} + T_{Startup_Sig_Proc} + T_{Bitcheck}}$								
11.13	Supply current in RX polling mode	CLK enabled V_{VSOUT} enabled CLK disabled V_{VSOUT} enabled V_{VSOUT} disabled	17, 19, 22, 27	I_{S_Poll}	$I_{S_Poll} = I_P + I_{VSINT} + I_{EXT}$ $I_{S_Poll} = I_P + I_{EXT}$ $I_{S_Poll} = I_P$				
11.14	Current in TX mode on pin VS2 and VAUX	I_{VSOUT} = 0 P_{out} = 5dBm/10dBm 315 MHz/5dBm 315 MHz/10dBm 433.92 MHz/5dBm 433.92 MHz/10dBm 868.3 MHz/10dBm	17, 19	$I_{TX_VS2_VAUX}$		10.7 16.2 10.9 16.3 11.6 17.8	13.9 21.0 14.0 21.0 15.0 23.0	mA	B
11.15	Supply current in TX mode	CLK enabled V_{VSOUT} enabled CLK disabled V_{VSOUT} enabled	17, 19, 22, 27	I_{S_TX}	$I_{S_TX} = I_{TX_VS2_VAUX} + I_{VSINT} + I_{EXT}$ $I_{S_TX} = I_{TX_VS2_VAUX} + I_{EXT}$				

*) Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter

Digital Timing Characteristics

All parameters refer to GND and are valid for T_{amb} = -40°C to +105°C. $V_{VS1} = V_{S2}$ = 2.4 V to 3.6 V (1-battery application), V_{VS2} = 4.4 V to 6.6 V (2-battery application) and V_{VS2} = 4.75 V to 5.25 V (car application), typical values at $V_{VS1} = V_{VS2}$ = 3 V and T_{amb} = 25°C unless otherwise specified.

No.	Parameters	Test Conditions	Pin	Symbol	Min.	Typ.	Max.	Unit	Type*
12	**Basic Clock Cycle of the Digital Circuitry**								
12.1	Basic clock cycle			T_{DCLK}	$16/f_{XTO}$		$16/f_{XTO}$	µs	A
12.2	Extended basic clock cycle	XLIM = 0 BR_Range_0 BR_Range_1 BR_Range_2 BR_Range_3 XLIM = 1 BR_Range_0 BR_Range_1 BR_Range_2 BR_Range_3		T_{XDCLK}	 8 4 2 1 × T_{DCLK} 16 8 4 2 × T_{DCLK}		 8 4 2 1 × T_{DCLK} 16 8 4 2 × T_{DCLK}	µs	A
13	**RX Mode/RX Polling Mode**								
13.1	Sleep time	Sleep and XSleep are defined in control register 4		T_{Sleep}	Sleep × X_{Sleep} × 1024 × T_{DCLK}		Sleep × X_{Sleep} × 1024 × T_{DCLK}	ms	A
13.2	Start-up PLL RX mode	from Idle mode		$T_{Startup_PLL}$		798.5 × T_{DCLK}	798.5 × T_{DCLK}	µs	A
13.3	Start-up signal processing	BR_Range_0 BR_Range_1 BR_Range_2 BR_Range_3		$T_{Startup_Sig_Proc}$	882 498 306 210 × T_{DCLK}		882 498 306 210 × T_{DCLK}		A
13.4	Time for Bit-check	Average time during polling. No RF signal applied. $f_{Signal} = 1/(2 \times t_{ee})$ Signal data rate Manchester (Lim_min and Lim_max up to ±50% of t_{ee}, see Figure 43 on page 52) Bit-check time for a valid input signal f_{Sig} $N_{Bit\text{-}check}$ = 0 $N_{Bit\text{-}check}$ = 3 $N_{Bit\text{-}check}$ = 6 $N_{Bit\text{-}check}$ = 9		T_{Bit_check}	 $3/f_{Sig}$ $6/f_{Sig}$ $9/f_{Sig}$	$1/f_{Signal}$	 $3.5/f_{Sig}$ $6.5/f_{Sig}$ $9.5/f_{Sig}$	ms	C

*) Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter

Digital Timing Characteristics (Continued)

All parameters refer to GND and are valid for T_{amb} = -40°C to +105°C. $V_{VS1} = V_{S2}$ = 2.4 V to 3.6 V (1-battery application), V_{VS2} = 4.4 V to 6.6 V (2-battery application) and V_{VS2} = 4.75 V to 5.25 V (car application), typical values at $V_{VS1} = V_{VS2}$ = 3 V and T_{amb} = 25°C unless otherwise specified.

No.	Parameters	Test Conditions	Pin	Symbol	Min.	Typ.	Max.	Unit	Type*
13.5	Baud-rate range	BR_Range = BR_Range0 BR_Range1 BR_Range2 BR_Range3		BR_Range	 1.0 2.0 4.0 8.0		 2.5 5.0 10.0 20.0	kBaud	A
13.6	Minimum time period between edges at pin SDO_TMDO in RX transparent mode	XLIM = 0 BR_Range_0 BR_Range_1 BR_Range_2 BR_Range_3 XLIM = 1 BR_Range_0 BR_Range_1 BR_Range_2 BR_Range_3	31	T_{DATA_min}	$10 \times T_{XDCLK}$			µs	A
13.7	Edge-to-edge time period of the data signal for full sensitivity in RX mode	BR_Range_0 BR_Range_1 BR_Range_2 BR_Range_3		T_{DATA}	200 100 50 25		500 250 125 62.5	µs	B
14	**TX Mode**								
14.1	Start-up time	From Idle mode		$T_{Startup}$		$331.5 \times T_{DCLK}$	$331.5 \times T_{DCLK}$	µs	A
15	**Configuration of the Transceiver with 4-wire Serial Interface**								
15.1	CS set-up time to rising edge of SCK		33, 35	T_{CS_setup}	$1.5 \times T_{DCLK}$			µs	A
15.2	SCK cycle time		33	T_{Cycle}	2			µs	A
15.3	SDI_TMDI set-up time to rising edge of SCK		32, 33	T_{Setup}	250			ns	C
15.4	SDI_TMDI hold time from rising edge of SCK		32, 33	T_{Hold}	250			ns	C
15.5	SDO_TMDO enable time from rising edge of CS		31, 35	T_{Out_enable}			250	ns	C
15.6	SDO_TMDO output delay from falling edge of SCK	C_L = 10 pF	31, 35	T_{Out_delay}			250	ns	C
15.7	SDO_TMDO disable time from falling edge of CS		31, 33	$T_{Out_disable}$			250	ns	C
15.8	CS disable time period		35	$T_{CS_disable}$	$1.5 \times T_{DCLK}$			µs	A

*) Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter

Digital Timing Characteristics (Continued)

All parameters refer to GND and are valid for T_{amb} = -40°C to +105°C. $V_{VS1} = V_{S2}$ = 2.4 V to 3.6 V (1-battery application), V_{VS2} = 4.4 V to 6.6 V (2-battery application) and V_{VS2} = 4.75 V to 5.25 V (car application), typical values at $V_{VS1} = V_{VS2}$ = 3 V and T_{amb} = 25°C unless otherwise specified.

No.	Parameters	Test Conditions	Pin	Symbol	Min.	Typ.	Max.	Unit	Type*
15.9	Time period SCK low to CS high		33, 35	T_{SCK_setup1}	250			ns	C
15.10	Time period SCK low to CS low		33, 35	T_{SCK_setup2}	250			ns	C
15.11	Time period CS low to SCK high		33, 35	T_{SCK_hold}	250			ns	C
16	**Start Time Push Button Tn and PWR_ON** Timing of wake-up via PWR_ON or Tn								
16.1	PWR_ON high to positive edge on pin IRQ (see Figure 30 on page 42)	From OFF mode to Idle mode, applications according to Figure 4 on page 6, Figure 5 on page 7 and Figure 6 on page 8 XTAL: C_m = 4..7 fF (typ. 5 fF) C_0 < 2.2 pF (typ. 1.8 pF) R_m ≤120 Ω (typ. 15 Ω) 1-battery application $C_1 = C_2$ = 68 nF $C_3 = C_4$ = 68 nF C_5 = 10 nF 2-battery application $C_1 = C_4$ = 68 nF $C_2 = C_3$ = 2.2 µF C_5 = 10 nF Car application $C_1 = C_3 = C_4$ = 68 nF $C_2 = C_{12}$ = 2.2 µF C_5 = 10nF	29, 40	$T_{PWR_ON_IRQ_1}$		0.3 0.45 0.45	0.8 1.3 1.3	ms	B

*) Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter

Digital Timing Characteristics (Continued)

All parameters refer to GND and are valid for T_{amb} = -40°C to +105°C. $V_{VS1} = V_{S2}$ = 2.4 V to 3.6 V (1-battery application), V_{VS2} = 4.4 V to 6.6 V (2-battery application) and V_{VS2} = 4.75 V to 5.25 V (car application), typical values at $V_{VS1} = V_{VS2}$ = 3 V and T_{amb} = 25°C unless otherwise specified.

No.	Parameters	Test Conditions	Pin	Symbol	Min.	Typ.	Max.	Unit	Type*
16.2	PWR_ON high to positive edge on pin IRQ (see Figure 30 on page 42)	Every mode except OFF mode	29, 40	$T_{PWR_ON_IRQ_2}$			2 × T_{DCLK}	µs	A
16.3	Tn low to positive edge on pin IRQ (see Figure 28 on page 40)	From OFF mode to Idle mode, applications according to Figure 4 on page 6, Figure 5 on page 7 and Figure 6 on page 8 XTAL: C_m = 4..7 fF (typ 5 fF) C_0 < 2.2 pF (typ 1.8 pF) R_m ≤ 20 Ω (typ 15 Ω) 1-battery application $C_1 = C_2$ = 68 nF $C_3 = C_4$ = 68 nF C_5 = 10 nF 2-battery application $C_1 = C_4$ = 68 nF $C_2 = C_3$ = 2.2 µF C_5 = 10 nF Car application $C_1 = C_3 = C_4$ = 68 nF $C_2 = C_{12}$ = 2.2 µF C_5 = 10 nF	29, 41, 42, 43, 44, 45	T_{Tn_IRQ}		0.3 0.45 0.45	0.8 1.3 1.3	ms	B
16.4	Push button debounce time	Every mode except OFF mode	29, 41, 42, 43, 44, 45	$T_{Debounce}$	8195 × T_{DCLK}		8195 × T_{DCLK}	µs	A

*) Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter

Digital Port Characteristics

All parameter refer to GND and valid for T_{amb} = -40 °C to +105 °C, $V_{VS1} = V_{S2}$ = 2.4 V to 3.6 V (1 Battery Application) and V_{VS2} = 4.4 V to 6.6 V (2 Battery Application) and V_{VS2} = 4.75 V to 5.25 V (Car Application) typical values at $V_{VS1} = V_{VS2}$ = 3V and T_{amb} = 25°C unless otherwise specified

No.	Parameters	Test Conditions	Pin	Symbol	Min.	Typ.	Max.	Unit	Type*
17	**Digital Ports**								
17.1	CS input -Low level input voltage	V_{VSINT} = 2.4 V to 5.25 V	35	V_{Il}			0.2 × V_{VSINT}	V	A
	-High level input voltage	V_{VSINT} = 2.4 V to 5.25 V	35	V_{Ih}	0.8 × V_{VSINT}		V_{VSINT}	V	A
17.2	SCK input -Low level input voltage	V_{VSINT} = 2.4 V to 5.25 V	33	V_{Il}			0.2 × V_{VSINT}	V	A
	-High level input voltage	V_{VSINT} = 2.4 V to 5.25 V	33	V_{Ih}	0.8 × V_{VSINT}		V_{VSINT}	V	A
17.3	SDI_TMDI input -Low level input voltage	V_{VSINT} = 2.4 V to 5.25 V	32	V_{Il}			0.2 × V_{VSINT}	V	A
	-High level input voltage	V_{VSINT} = 2.4 V to 5.25 V	32	V_{Ih}	0.8 × V_{VSINT}		V_{VSINT}	V	A
17.4	TEST1 input	TEST1 input must always be directly connected to GND	20						D
17.5	TEST2 input	TEST2 input must always be direct connected to GND	23						D
17.6	PWR_ON input -Low level input voltage	Internal pull-down with series connection of 40 kΩ ±20% resistor and diode	40	V_{Il}			0.4	V	A
	-High level input voltage[1]	Internal pull-down with series connection of 40 kΩ ±20% resistor and diode	40	V_{Ih}	0.8 × V_{VS2}			V	A
17.7	Tn input -Low level input voltage	Internal pull-up resistor of 50 kΩ ±20%	41, 42, 43, 44, 45	V_{Il}			0.2 × V_{VS2}	V	A
	-High level input voltage[1]	Internal pull-up resistor of 50 kΩ ±20%	41, 42, 43, 44, 45	V_{Ih}	× V_{VS2} - 0.5 V			V	A
17.8	433_N868 input -Low level input voltage		6	V_{Il}			0.25	V	A
	-Input current low		6	I_{Il}			-5	µA	A
	-High level input voltage		6	V_{Ih}	1.7		AVCC	V	A
	-Input current high		6	I_{Ih}			1	µA	A

*) Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter

Note: 1. If a logic high level is applied to this pin a minimum serial impedance of 100 Ω must be ensured for proper operation over full temperature range.

Digital Port Characteristics (Continued)

All parameter refer to GND and valid for T_{amb} = -40 °C to +105 °C, V_{VS1} = V_{S2} = 2.4 V to 3.6 V (1 Battery Application) and V_{VS2} = 4.4 V to 6.6 V (2 Battery Application) and V_{VS2} = 4.75 V to 5.25 V (Car Application) typical values at V_{VS1} = V_{VS2} = 3V and T_{amb} = 25°C unless otherwise specified

No.	Parameters	Test Conditions	Pin	Symbol	Min.	Typ.	Max.	Unit	Type*
17.9	PWR_H input -Low level input voltage		9	V_{Il}			0.25	V	A
	-Input current low		9	I_{Il}			-5	µA	A
	-High level input voltage		9	V_{Ih}	1.7		AVCC	V	A
	-Input current high		9	I_{Ih}			1	µA	A
17.10	SDO_TMDO output -Saturation voltage low	V_{VSINT} = 2.4 V to 5.25 V I_{SDO_TMDO} = 250 µA	31	V_{ol}		0.15	0.4	V	B
	Saturation voltage high	V_{VSINT} = 2.4 V to 5.25 V I_{SDO_TMDO} = -250 µA	31	V_{oh}	V_{VSINT} - 0.4	V_{VSINT} - 0.15		V	B
17.11	IRQ output -Saturation voltage low	V_{VSINT} = 2.4 V to 5.25 V I_{IRQ} = 250 µA	29	V_{ol}		0.15	0.4	V	B
	Saturation voltage high	V_{VSINT} = 2.4 V to 5.25 V I_{IRQ} = -250 µA	29	V_{oh}	V_{VSINT} - 0.4	V_{VSINT} - 0.15		V	B
17.12	CLK output -Saturation voltage low	V_{VSINT} = 2.4 V to 5.25 V I_{CLK} = 100 µA internal series resistor of 1 kΩ for spurious reduction in PLL	30	V_{ol}		0.15	0.4	V	B
	Saturation voltage high	V_{VSINT} = 2.4 V to 5.25 V I_{CLK} = -100 µA internal series resistor of 1 kΩ for spurious reduction in PLL	30	V_{oh}	V_{VSINT} - 0.4	V_{VSINT} - 0.15		V	B
17.13	N_RESET output -Saturation voltage low	V_{VSINT} = 2.4 V to 5.25 V I_{N_RESET} = 250 µA	28	V_{ol}		0.15	0.4	V	B
	-Saturation voltage high	V_{VSINT} = 2.4 V to 5.25 V I_{N_RESET} = -250 µA	28	V_{oh}	V_{VSINT} - 0.4	V_{VSINT} - 0.15		V	B
17.14	RX_ACTIVE output -Saturation voltage high	V_{VSINT} = 2.4 V to 5.25 V I_{RX_ACTIVE} = -1.5 mA	46	V_{oh}	V_{AVCC} - 0.5V	V_{AVCC} - 0.15V		V	B
	-Saturation voltage low	V_{VSINT} = 2.4 V to 5.25 V I_{RX_ACTIVE} = 25 µA	46	V_{ol}		0.25	0.4	V	B
17.15	DEM_OUT output Saturation voltage low	Open drain output I_{DEM_OUT} = 250 µA	34	V_{ol}		0.15	0.4	V	B

*) Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter

Note: 1. If a logic high level is applied to this pin a minimum serial impedance of 100 Ω must be ensured for proper operation over full temperature range.

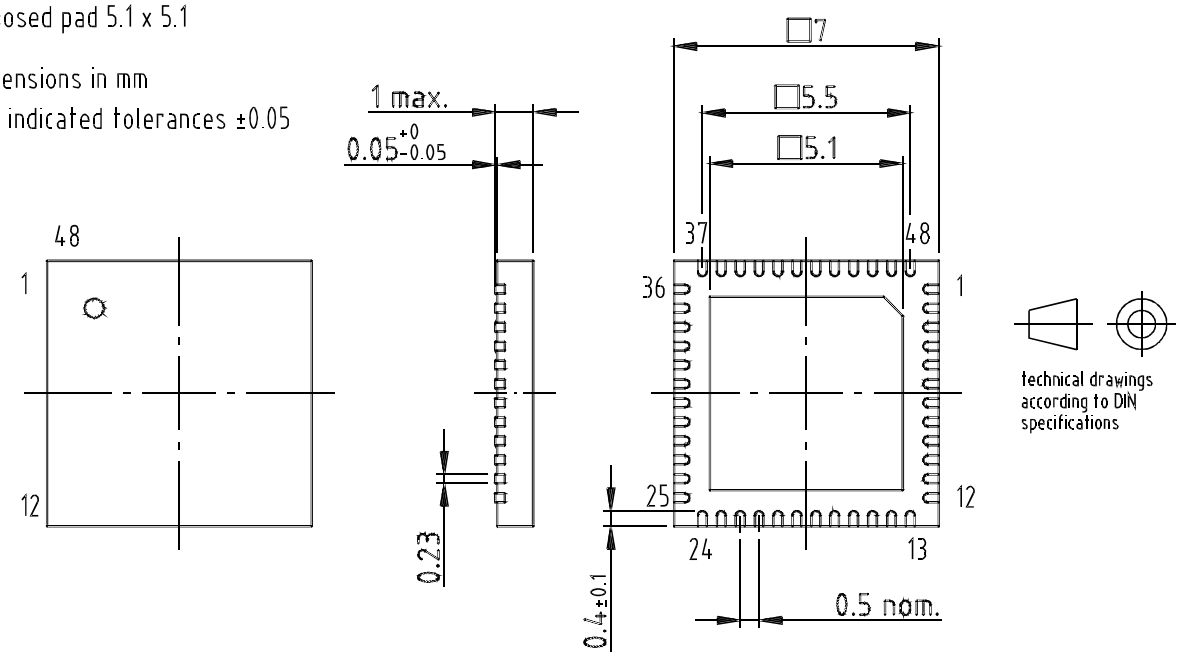
Ordering Information

Extended Type Number	Package	Remarks
ATA5811-PLQC	QFN48	7 mm × 7 mm
ATA5812-PLQC	QFN48	7 mm × 7 mm

Package Information

Package: QFN 48 7x7
Exposed pad 5.1 x 5.1

Dimensions in mm
Not indicated tolerances ±0.05



Drawing-No.: 6.543-5089.02-4
Issue: 1; 14.01.03

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