AS5040

Data Sheet

10Bit 360° Programmable Magnetic Rotary Encoder

1 General Description

The AS5040 is a contactless magnetic rotary encoder for accurate angular measurement over a full turn of 360°. It is a system-on-chip, combining integrated Hall elements, analog front end and digital signal processing in a single device.

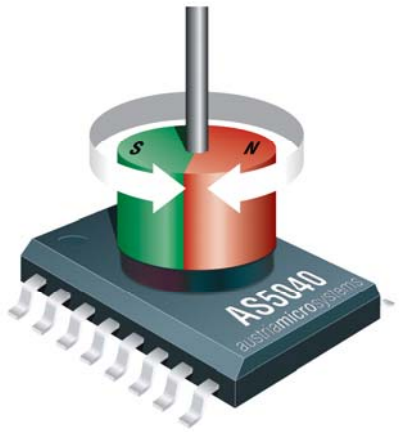
To measure the angle, only a simple two-pole magnet, rotating over the center of the chip, is required. The magnet may be placed above or below the IC.

The absolute angle measurement provides instant indication of the magnet's angular position with a resolution of 0.35° = 1024 positions per revolution. This digital data is available as a serial bit stream and as a PWM signal.

Furthermore, a user-programmable incremental output is available, making the chip suitable for replacement of various optical encoders.

An internal voltage regulator allows the AS5040 to operate at either 3.3 V or 5 V supplies.

Figure 1: Typical Arrangement of AS5040 and Magnet



Benefits

- Complete system-on-chip
- Flexible system solution provides absolute, PWM and incremental outputs simultaneously
- Ideal for applications in harsh environments due to contactless position sensing
- Tolerant to magnet misalignment and airgap variations
- No temperature compensation necessary
- No calibration required

2 Key Features

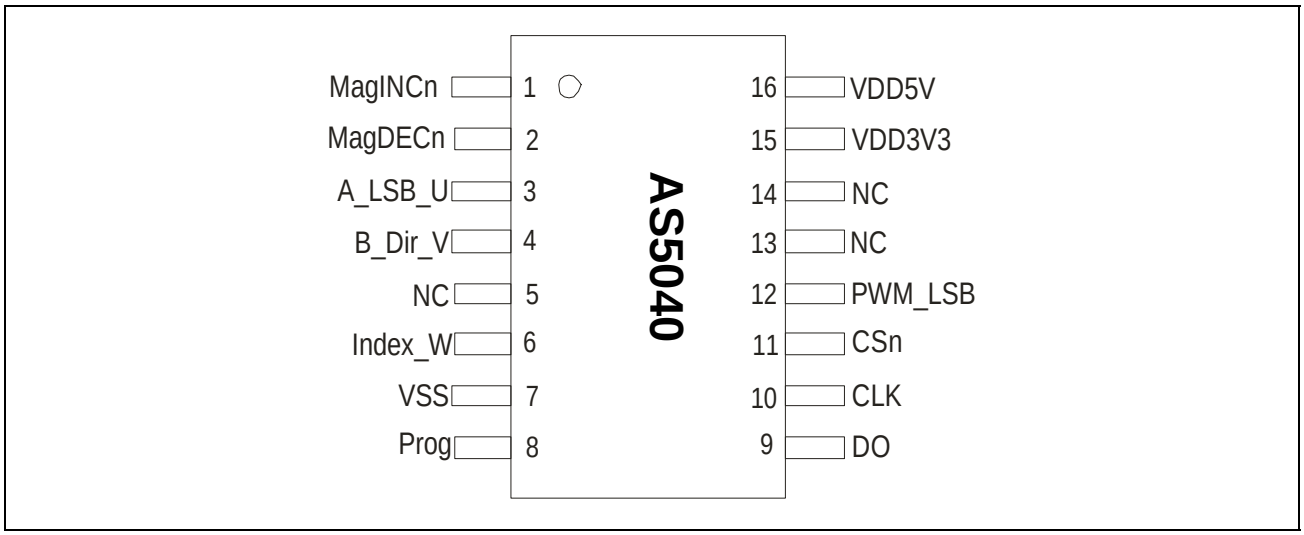
- Contactless high resolution rotational position encoding over a full turn of 360 degrees
- Two digital 10bit absolute outputs:
  - Serial interface and
  - Pulse width modulated (PWM) output
- Three incremental output modes:
  - Quadrature A/B and Index output signal
  - Step / Direction and Index output signal
  - 3-phase commutation for brushless DC motors
  - 10, 9, 8 or 7 bit user programmable resolution
- User programmable zero / index position
- Failure detection mode for magnet placement monitoring and loss of power supply
- Rotational speeds up to 30,000 rpm
- Push button functionality detects movement of magnet in Z-axis
- Serial read-out of multiple interconnected AS5040 devices using Daisy Chain mode
- Wide temperature range: - 40°C to + 125°C
- Fully automotive qualified to AEC-Q100, grade 1
- Small Pb-free package: SSOP 16 (5.3mm x 6.2mm)

3 Applications

- Industrial applications:
  - Contactless rotary position sensing
  - Robotics
  - Brushless DC motor commutation
  - Power tools
- Automotive applications:
  - Steering wheel position sensing
  - Gas pedal position sensing
  - Transmission gearbox encoder
  - Headlight position control
  - Power seat position indicator
- Office equipment: printers, scanners, copiers
- Replacement of optical encoders
- Front panel rotary switches
- Replacement of potentiometers

4 Pin Configuration

Figure 2: Pin Configuration SSOP16



Pin Description

Table 2 shows the description of each pin of the standard SSOP16 package (Shrink Small Outline Package, 16 leads, body size: 5.3mm x 6.2mmm; see Figure 2).

Pins 7, 15 and 16 are supply pins, pins 5, 13 and 14 are for internal use and must not be connected.

Pins 1 and 2 are the magnetic field change indicators, MagINCn and MagDECn (magnetic field strength increase or decrease through variation of the distance between the magnet and the device). These outputs can be used to detect the valid magnetic field range. Furthermore those indicators can also be used for contact-less push-button functionality.

Pins 3, 4 and 6 are the incremental pulse output pins. The functionality of these pins can be configured through programming the one-time programmable (OTP) register.

Table 1: Pin Assignment for the Different Incremental Output Modes

Output Mode	Pin 3	Pin 4	Pin 6	Pin 12
1.x: quadrature	A	B	Index	PWM
2.x:step/direction	LSB	Direction	Index	PWM
3.x: commutation	U	V	W	LSB

Mode 1.x: Quadrature A/B Output

Represents the default quadrature A/B signal mode.

Mode 2.x: Step / Direction Output

Configures pin 3 to deliver up to 512 pulses (up to 1024 state changes) per revolution. It is equivalent to the LSB (least significant bit) of the absolute position value. Pin 4 provides the information of the rotational direction.

Both modes (mode 1.x and mode 2.x) provide an index signal (1 pulse/revolution) with an adjustable width of one LSB or three LSB's.

Table 2 Pin Description SSOP16

Pin	Symbol	Type	Description
1	MagINCn	DO_OD	Magnet Field **Mag**nitude **INC**rease; active low, indicates a distance reduction between the magnet and the device surface.
2	MagDECn	DO_OD	Magnet Field **Mag**nitude **DEC**rease; active low, indicates a distance increase between the device and the magnet.
3	A_LSB_U	DO	*Mode1.x:* Quadrature **A** channel *Mode2.x:* **L**east **S**ignificant **B**it *Mode3.x:* **U** signal (phase1)
4	B_Dir_V	DO	*Mode1.x*: Quadrature **B** channel quarter period shift to channel A. *Mode2.x:* **Dir**ection of Rotation *Mode3.x:* **V** signal (phase2)
5	NC	-	Must be left unconnected
6	Index_W	DO	*Mode1.x and Mode2.x* : **Index** signal indicates the absolute zero position *Mode3.x*: **W** signal (phase3)
7	VSS	S	Negative Supply Voltage (GND)
8	Prog	DI_PD	OTP **Prog**ramming Input and Data Input for Daisy Chain mode. Internal pull-down resistor (~74kΩ). May be connected to VSS if programming is not used
9	DO	DO_T	**D**ata **O**utput of Synchronous Serial Interface
10	CLK	DI, ST	**Clock** Input of Synchronous Serial Interface; Schmitt-Trigger input
11	CSn	DI_PU, ST	**C**hip **S**elect, active low; Schmitt-Trigger input, internal pull-up resistor (~50kΩ) connect to VSS in incremental mode (see 0)
12	PWM_LSB	DO	**P**ulse **W**idth **M**odulation of approx. 1kHz; **LSB** in *Mode3.x*
13	NC	-	Must be left unconnected
14	NC	-	Must be left unconnected
15	VDD3V3	S	3V-Regulator Output (see Figure 19)
16	VDD5V	S	Positive Supply Voltage 5 V

DO_OD	digital output open drain	S	supply pin
DO	digital output	DI	digital input
DI_PD	digital input pull-down	DO_T	digital output /tri-state
DI_PU	digital input pull-up	ST	Schmitt-Trigger input

Mode 3.x: Brushless DC Motor Commutation Mode

In addition to the absolute encoder output over the SSI interface, this mode provides commutation signals for brushless DC motors with either one pole pair or two pole pair rotors. The commutation signals are usually provided by 3 discrete Hall switches, which are no longer required, as the AS5040 can fulfill two tasks in parallel: absolute encoder + BLDC motor commutation.

In this mode, pin 12 provides the LSB output instead of the PWM (Pulse-Width-Modulation) signal.

Pin 8 (Prog) is also used to program the different incremental interface modes, the incremental resolution and the zero position into the OTP (see page 18).

This pin is also used as digital input to shift serial data through the device in Daisy Chain configuration, (see page 11).

Pin 11 Chip Select (CSn; active low) selects a device within a network of AS5040 encoders and initiates serial data transfer. A logic high at CSn puts the data output pin (DO) to tri-state and terminates serial data transfer. This pin is also used for alignment mode (page 21) and programming mode (page 16).

Pin 12 allows a single wire output of the 10-bit absolute position value. The value is encoded into a pulse width modulated signal with 1µs pulse width per step (1µs to 1024µs over a full turn). By using an external low pass filter, the digital PWM signal is converted into an analog voltage, allowing a direct replacement of potentiometers.

5 Electrical Characteristics

Absolute Maximum Ratings (non operating)

Stresses beyond those listed under “Absolute Maximum Ratings“ may cause permanent damage to the device. These are stress ratings only. Functional operation of the device at these or any other conditions beyond those indicated under “Operating Conditions” is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Parameter	Symbol	Min	Max	Unit	Note
DC supply voltage at pin VDD5V	VDD5V	-0.3	7	V	
DC supply voltage at pin VDD3V3	VDD3V3	-0.3	5	V	
Input pin voltage	Vin	-0.3	VDD5V +0.3	V	Pins MagIncn, MagDecn, CLK, CSn,
		-0.3	7.5		Pin Prog
Input current (latchup immunity)	Iscr	-100	100	mA	Norm: JEDEC 78
Electrostatic discharge	ESD		± 2	kV	Norm: MIL 883 E method 3015
Storage temperature	Tstrg	-55	125	°C	Min – 67°F ; Max +257°F
Body temperature (Lead-free package)	TBody		260	°C	t=20 to 40s, Norm: IPC/JEDEC J-Std-020C Lead finish 100% Sn “matte tin”
Humidity non-condensing	H	5	85	%	

Operating Conditions

Parameter	Symbol	Min	Typ	Max	Unit	Note
Ambient temperature	T_{amb}	-40		125	°C	-40°F…+257°F
Supply current	I_{supp}		16	21	mA	
External supply voltage at pin VDD5V Internal regulator output voltage at pin VDD3V3	VDD5V VDD3V3	4.5 3.0	5.0 3.3	5.5 3.6	V V	5V operation
External supply voltage at pin VDD5V, VDD3V3	VDD5V VDD3V3	3.0 3.0	3.3 3.3	3.6 3.6	V V	3.3V operation (pins VDD5V and VDD3V3 connected)

DC Characteristics for Digital Inputs and Outputs

CMOS Schmitt-Trigger Inputs: CLK, CSn (CSn = internal Pull-up)

(operating conditions: T_{amb} = -40 to +125°C, VDD5V = 3.0-3.6V (3V operation) VDD5V = 4.5-5.5V (5V operation) unless otherwise noted)

Parameter	Symbol	Min	Max	Unit	Note
High level input voltage	V_{IH}	0.7 * VDD5V		V	Normal operation
Low level input voltage	V_{IL}		0.3 * VDD5V	V	
Schmitt Trigger hysteresis	V_{Ion}- V_{Ioff}	1		V	
Input leakage current	I_{LEAK}	-1	1	µA	CLK only
Pull-up low level input current	I_{iL}	-30	-100	µA	CSn only, VDD5V: 5.0V

CMOS / Program Input: Prog

(operating conditions: T_{amb} = -40 to +125°C, VDD5V = 3.0-3.6V (3V operation) VDD5V = 4.5-5.5V (5V operation) unless otherwise noted)

Parameter	Symbol	Min	Max	Unit	Note
High level input voltage	V_{IH}	0.7 * VDD5V	5	V	
High level input voltage	V_{PROG}	See “programming conditions”		V	During programming
Low level input voltage	V_{IL}		0.3 * VDD5V	V	
Pull-down high level input current	I_{iL}		100	µA	VDD5V: 5.5V

CMOS Output Open Drain: MagINCn, MagDECn

(operating conditions: T_{amb} = -40 to +125°C, VDD5V = 3.0-3.6V (3V operation) VDD5V = 4.5-5.5V (5V operation) unless otherwise noted)

Parameter	Symbol	Min	Max	Unit	Note
Low level output voltage	V_{OL}		VSS+0.4	V	
Output current	I_O		4 2	mA	VDD5V: 4.5V VDD5V: 3V
Open drain leakage current	I_{OZ}		1	µA	

CMOS Output: A, B, Index, PWM

(operating conditions: T_{amb} = -40 to +125°C, VDD5V = 3.0-3.6V (3V operation) VDD5V = 4.5-5.5V (5V operation) unless otherwise noted)

Parameter	Symbol	Min	Max	Unit	Note
High level output voltage	V_{OH}	VDD5V-0.5		V	
Low level output voltage	V_{OL}		VSS+0.4	V	
Output current	I_O		4 2	mA	VDD5V: 4.5V VDD5V: 3V

Tristate CMOS Output: DO

(operating conditions: T_{amb} = -40 to +125°C, VDD5V = 3.0-3.6V (3V operation) VDD5V = 4.5-5.5V (5V operation) unless otherwise noted)

Parameter	Symbol	Min	Max	Unit	Note
High level output voltage	V_{OH}	VDD5V –0.5		V	
Low level output voltage	V_{OL}		VSS+0.4	V	
Output current	I_O		4 2	mA	VDD5V: 4.5V VDD5V: 3V
Tri-state leakage current	I_{OZ}		1	µA	

Magnetic Input Specification

(operating conditions: T_{amb} = -40 to +125°C, VDD5V = 3.0-3.6V (3V operation) VDD5V = 4.5-5.5V (5V operation) unless otherwise noted)

Two-pole cylindrical diametrically magnetised source:

Parameter	Symbol	Min	Typ	Max	Unit	Note
Diameter	d_{mag}	4	6		mm	Recommended magnet: Ø 6mm x 2.5mm for cylindrical magnets
Thickness	t_{mag}	2.5			mm	
Magnetic input field amplitude	B_{pk}	45		75	mT	Required vertical component of the magnetic field strength on the die’s surface, measured along a concentric circle with a radius of 1.1mm

Parameter	Symbol	Min	Typ	Max	Unit	Note
Magnetic offset	B_{off}			± 10	mT	Constant magnetic stray field
Field non-linearity				5	%	Including offset gradient
Input frequency (rotational speed of magnet)	f_{mag_abs}			10	Hz	Absolute mode: 600 rpm @ readout of 1024 positions (see Table 6)
	f_{mag_inc}			500	Hz	Incremental mode: no missing pulses at rotational speeds of up to 30,000 rpm (see Table 6)
Displacement radius	Disp			0.25	mm	Max. X-Y offset between defined IC package center and magnet axis (see Figure 21)
				0.485		Max. X-Y offset between chip center and magnet axis.
Chip placement tolerance				±0.235	mm	Placement tolerance of chip within IC package (see Figure 23)
Recommended magnet material and temperature drift			-0.12		%/K	NdFeB (Neodymium Iron Boron)
			-0.035			SmCo (Samarium Cobalt)

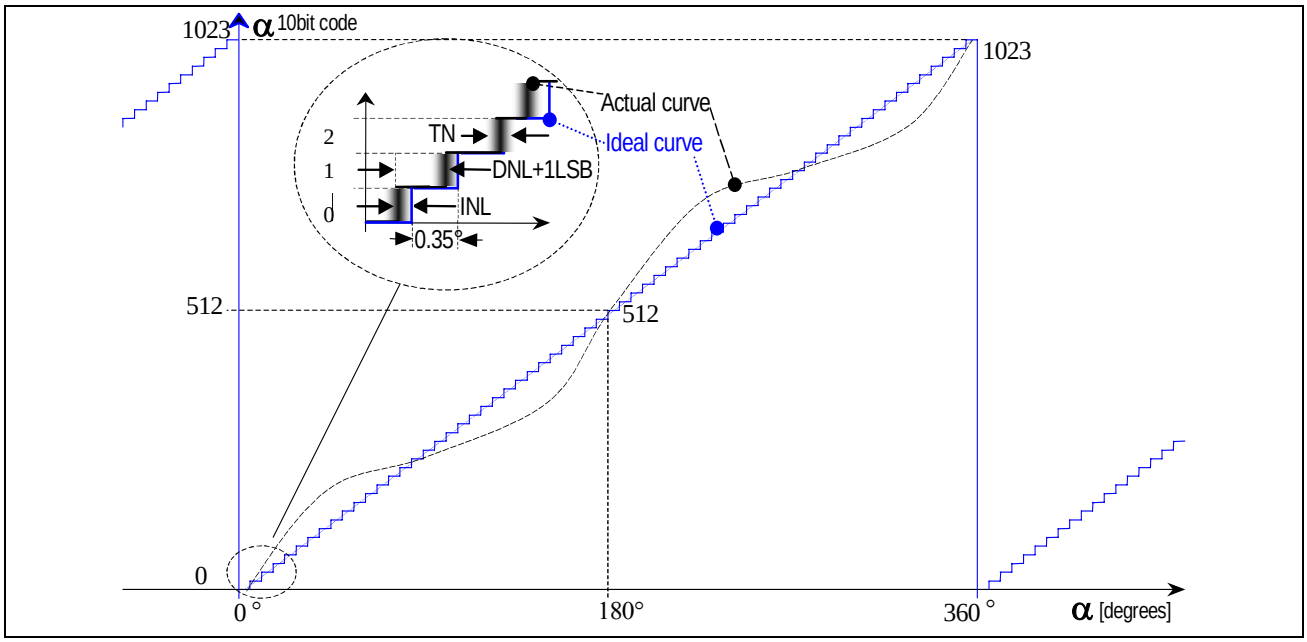
Electrical System Specifications

(operating conditions: T_{amb} = -40 to +125°C, VDD5V = 3.0-3.6V (3V operation) VDD5V = 4.5-5.5V (5V operation) unless otherwise noted)

Parameter	Symbol	Min	Typ	Max	Unit	Note
Resolution	RES			10	bit	0.352 deg
7 bit 8 bit 9 bit 10 bit	LSB		2.813 1.406 0.703 0.352		deg	Adjustable resolution only available for incremental output modes; Least significant bit, minimum step
Integral non-linearity (optimum)	INL_{opt}			± 0.5	deg	Maximum error with respect to the best line fit. Verified at optimum magnet placement, T_{amb} =25 °C.
Integral non-linearity (optimum)	INL_{temp}			± 0.9	deg	Maximum error with respect to the best line fit. Verified at optimum magnet placement , T_{amb} = -40 to +125°C
Integral non-linearity	INL			± 1.4	deg	Best line fit = (Err_{max} – Err_{min}) / 2 Over displacement tolerance with 6mm diameter magnet, T_{amb} = -40 to +125°C
Differential non-linearity	DNL			± 0.176	deg	10bit, no missing codes
Transition noise	TN			0.12	Deg RMS	RMS equivalent to 1 sigma
Hysteresis	Hyst		0.704		deg	Incremental modes only
Power-on reset thresholds On voltage; 300mV typ. hysteresis Off voltage; 300mV typ. hysteresis	 V_{on} V_{off}	 1.37 1.08	 2.2 1.9	 2.9 2.6	V	 DC supply voltage 3.3V (VDD3V3) DC supply voltage 3.3V (VDD3V3)
Power-up time	t_{PwrUp}			50	ms	Until offset compensation finished
System propagation delay absolute output	t_{delay}			48	µs	Includes delay of ADC and DSP

Parameter	Symbol	Min	Typ	Max	Unit	Note
System propagation delay incremental output				192	µs	Calculation over two samples
Sampling rate for absolute output	f_S	9.90	10.42	10.94	kHz	Internal sampling rate, T_{amb} = 25°C
		9.38	10.42	11.46		Internal sampling rate, T_{amb} = -40 to +125°C
Read-out frequency	CLK			1	MHz	Max. clock frequency to read out serial data

Figure 3: Integral and Differential Non-linearity Example (exaggerated curve)



Integral Non-Linearity (INL) is the maximum deviation between actual position and indicated position.

Differential Non-Linearity (DNL) is the maximum deviation of the step length from one position to the next.

Transition Noise (TN) is the repeatability of an indicated position.

Timing Characteristics

Synchronous Serial Interface (SSI)

(operating conditions: T_{amb} = -40 to +125°C, VDD5V = 3.0-3.6V (3V operation) VDD5V = 4.5-5.5V (5V operation) unless otherwise noted)

Parameter	Symbol	Min	Typ	Max	Unit	Note
Data output activated (logic high)	$t_{DO\ active}$			100	ns	Time between falling edge of CSn and data output activated
First data shifted to output register	$t_{CLK\ FE}$	500			ns	Time between falling edge of CSn and first falling edge of CLK
Start of data output	$T_{CLK\ /\ 2}$	500			ns	Rising edge of CLK shifts out one bit at a time
Data output valid	$t_{DO\ valid}$	357		413	ns	Time between rising edge of CLK and data output valid
Data output tristate	$t_{DO\ tristate}$			100	ns	After the last bit DO changes back to “tristate”
Pulse width of CSn	t_{CSn}	500			ns	CSn = high; To initiate read-out of next angular position

Parameter	Symbol	Min	Typ	Max	Unit	Note
Read-out frequency	f_{CLK}	>0		1	MHz	Clock frequency to read out serial data

Pulse Width Modulation Output

(operating conditions: T_{amb} = -40 to +125°C, VDD5V = 3.0-3.6V (3V operation) VDD5V = 4.5-5.5V (5V operation) unless otherwise noted)

Parameter	Symbol	Min	Typ	Max	Unit	Note
PWM frequency	f_{PWM}	0.927	0.976	1.024	KHz	Signal period = 1025µs ±5% at T_{amb} = 25°C
		0.878	0.976	1.074		=1025µs ±10% at T_{amb} = -40 to +125°C
Minimum pulse width	PW_{MIN}	0.90	1	1.10	µs	Position 0d; angle 0 degree
Maximum pulse width	PW_{MAX}	922	1024	1126	µs	Position 1023d; angle 359.65 degree

Incremental Outputs

(operating conditions: T_{amb} = -40 to +125°C, VDD5V = 3.0-3.6V (3V operation) VDD5V = 4.5-5.5V (5V operation) unless otherwise noted)

Parameter	Symbol	Min	Typ	Max	Unit	Note
Incremental outputs valid after power-up	$t_{Incremental\ outputs\ valid}$			500	ns	Time between first falling edge of CSn after power-up and valid incremental outputs
Directional indication valid	$t_{Dir\ valid}$			500	ns	Time between rising or falling edge of LSB output and valid directional indication

Programming Conditions

(operating conditions: T_{amb} = -40 to +125°C, VDD5V = 3.0-3.6V (3V operation) VDD5V = 4.5-5.5V (5V operation) unless otherwise noted)

Parameter	Symbol	Min	Typ	Max	Unit	Note
Programming enable time	$t_{Prog\ enable}$	2			µs	Time between rising edge at Prog pin and rising edge of CSn
Write data start	$t_{Data\ in}$	2			µs	
Write data valid	$t_{Data\ in\ valid}$	250			ns	Write data at the rising edge of CLK_{PROG}
Load programming data	$t_{Load\ PROG}$	3			µs	
Rise time of V_{PROG} before CLK_{PROG}	t_{PrgR}	0			µs	
Hold time of V_{PROG} after CLK_{PROG}	t_{PrgH}	0		5	µs	
Write data – programming CLK_{PROG}	CLK_{PROG}			250	kHz	
CLK pulse width	t_{PROG}	1.8	2	2.2	µs	During programming; 16 clock cycles
Hold time of Vprog after programming	$t_{PROG\ finished}$	2			µs	Programmed data is available after next power-on
Programming voltage	V_{PROG}	7.3	7.4	7.5	V	Must be switched off after zapping
Programming voltage off level	$V_{ProgOff}$	0		1	V	Line must be discharged to this level
Programming current	I_{PROG}			130	mA	During programming
Analog read CLK	CLK_{Aread}			100	kHz	Analog readback mode
Programmed zener voltage (log.1)	$V_{programmed}$			100	mV	V_{Ref}-V_{PROG} during analog readback mode (see Analog Readback Mode)
Unprogrammed zener voltage (log. 0)	$V_{unprogrammed}$	1			V	

6 Functional Description

The AS5040 is manufactured in a CMOS standard process and uses a spinning current Hall technology for sensing the magnetic field distribution across the surface of the chip.

The integrated Hall elements are placed around the center of the device and deliver a voltage representation of the magnetic field at the surface of the IC.

Through Sigma-Delta Analog / Digital Conversion and Digital Signal-Processing (DSP) algorithms, the AS5040 provides accurate high-resolution absolute angular

position information. For this purpose a Coordinate Rotation Digital Computer (CORDIC) calculates the angle and the magnitude of the Hall array signals.

The DSP is also used to provide digital information at the outputs MagINCn and MagDECn that indicate movements of the used magnet towards or away from the device's surface.

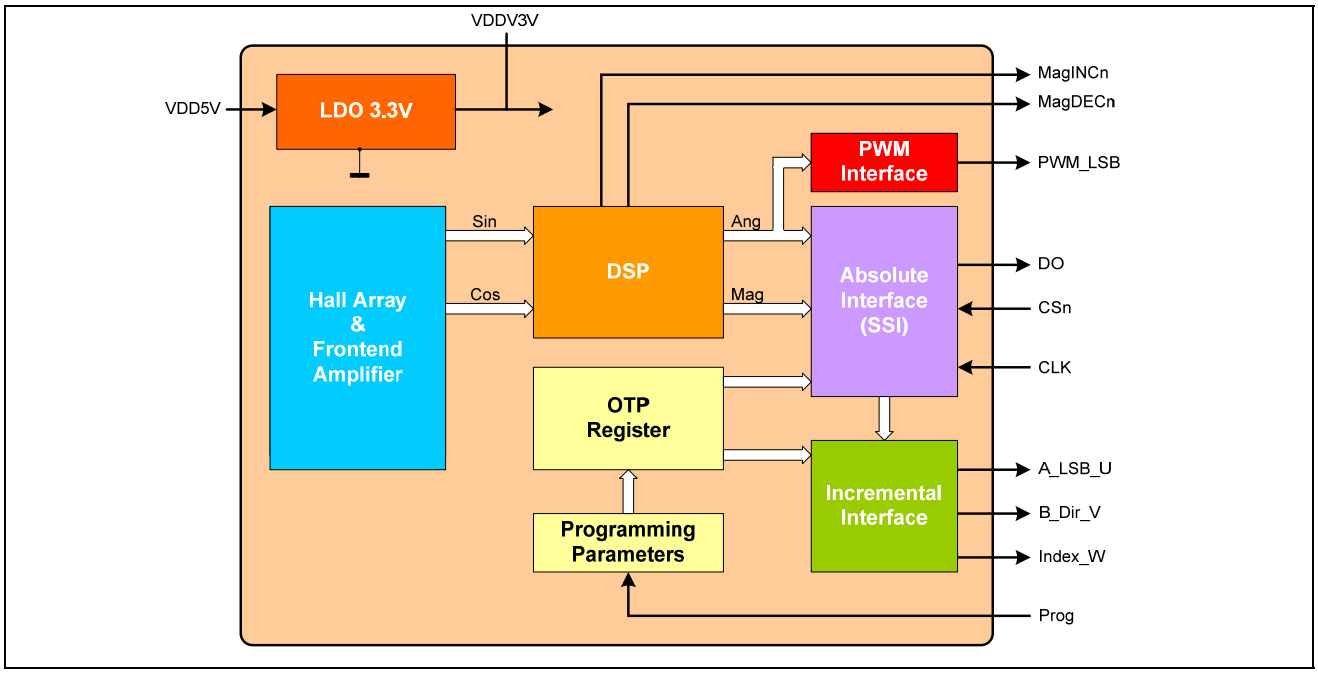
A small low cost diametrically magnetized (two-pole) standard magnet provides the angular position information (see Figure 20).

The AS5040 senses the orientation of the magnetic field and calculates a 10-bit binary code. This code can be accessed via a Synchronous Serial Interface (SSI). In addition, an absolute angular representation is given by a Pulse Width Modulated signal at pin 12 (PWM).

Besides the absolute angular position information the device simultaneously provides incremental output signals. The various incremental output modes can be selected by programming the OTP mode register bits (see page 19). As long as no programming voltage is applied to pin Prog, the new setting may be overwritten at any time and will be reset to default when power is turned off. To make the setting permanent, the OTP register must be programmed (see Figure 15). The default setting is a quadrature A/B mode including the Index signal with a pulse width of 1 LSB. The Index signal is logic high at the user programmable zero position.

The AS5040 is tolerant to magnet misalignment and magnetic stray fields due to differential measurement technique and Hall sensor conditioning circuitry.

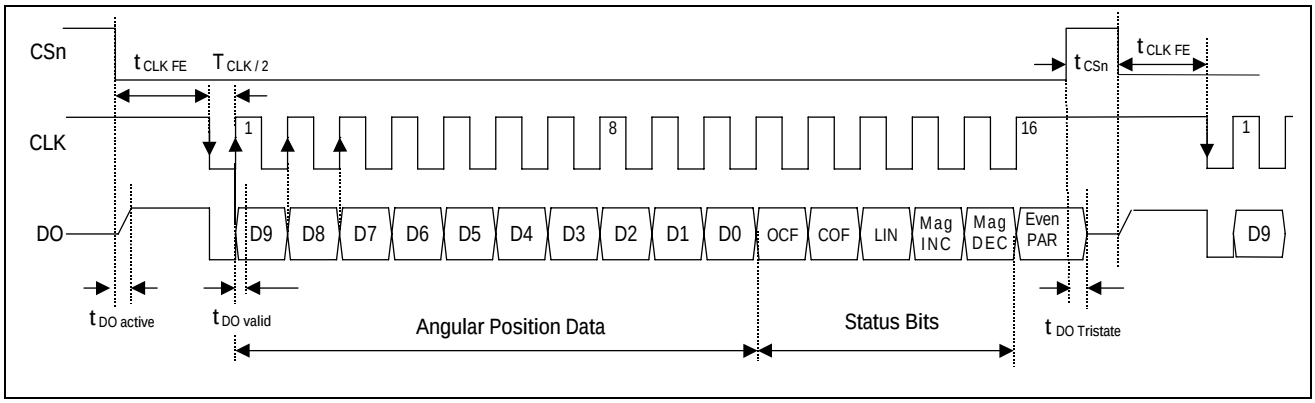
Figure 4: AS5040 Block Diagram



7 10-bit Absolute Angular Position Output

Synchronous Serial Interface (SSI)

Figure 5: Synchronous Serial Interface with Absolute Angular Position Data



If CSn changes to logic low, Data Out (DO) will change from high impedance (tri-state) to logic high and the read-out will be initiated.

- After a minimum time $t_{CLK\ FE,}$ data is latched into the output shift register with the first falling edge of CLK.
- Each subsequent rising CLK edge shifts out one bit of data.
- The serial word contains 16 bits, the first 10 bits are the angular information D[9:0], the subsequent 6 bits contain system information, about the validity of data such as OCF, COF, LIN, Parity and Magnetic Field status (increase/decrease) .
- A subsequent measurement is initiated by a log. “high” pulse at CSn with a minimum duration of $t_{CSn.}$

Data Content:

D9:D0 absolute angular position data (MSB is clocked out first)

OCF (**O**ffset **C**ompensation **F**inished), logic high indicates the finished Offset Compensation Algorithm. For fast startup, this bit may be polled by the external microcontroller. As soon as this bit is set, the AS5040 has completed the startup and the data is valid (see Table 4)

COF (**C**ordic **O**ver**f**low), logic high indicates an out of range error in the CORDIC part. When this bit is set, the data at D9:D0 is invalid. The absolute output maintains the last valid angular value.

This alarm may be resolved by bringing the magnet within the X-Y-Z tolerance limits.

LIN (**Lin**earity Alarm), logic high indicates that the input field generates a critical output linearity. When this bit is set, the data at D9:D0 may still be used, but can contain invalid data. This warning may be resolved by bringing the magnet within the X-Y-Z tolerance limits.

MagINCn, (**Mag**nitude **Inc**rease) becomes HIGH, when the magnet is pushed towards the IC, thus the magnetic field strength is increasing.

MagDECn, (**Mag**nitude **Dec**rease) becomes HIGH, when the magnet is pulled away from the IC, thus the magnetic field strength is decreasing.

Both signals HIGH indicate a magnetic field that is out of the allowed range (see Table 3).

Table 3: Magnetic Magnitude Variation Indicator

Mag INCn	Mag DECn	Description
0	0	No distance change; Magnetic input field OK (in range, 45..75mT)
0	1	Distance increase: Pull-function. This state is dynamic, it is only active while the magnet is moving away from the chip in Z-axis
1	0	Distance decrease: Push- function. This state is dynamic, it is only active while the magnet is moving towards the chip in Z.-axis.
1	1	Magnetic Input Field invalid – out of range: <45mT or >75mT (or missing magnet)

Note: Pins 1 and 2 (MagINCn, MagDECn) are open drain outputs and require external pull-up resistors. If the magnetic field is in range, both outputs are turned off.

The two pins may also be combined with a single pull-up resistor. In this case, the signal is high when the magnetic field is in range. It is low in all other cases (see Table 3).

Even Parity bit for transmission error detection of bits 1…15 (D9…D0, OCF, COF, LIN, MagINCn, MagDECn)

The absolute angular output is always set to a resolution of 10 bit. Placing the magnet above the chip, angular values increase in clockwise direction by default.

Data D9:D0 is valid, when the status bits have the following configurations:

Table 4: Status Bit Outputs

OCF	COF	LIN	Mag INCn	Mag DECn	Parity
1	0	0	0	0	even checksum of bits 1:15
			0	1	
			1	0	

The absolute angular position is sampled at a rate of 10kHz (0.1ms). This allows reading of all 1024 positions per 360 degrees within 0.1 seconds = 9.76Hz (~10Hz) without skipping any position. Multiplying 10Hz by 60, results the corresponding maximum rotational speed of 600 rpm.

Readout of every second angular position allows for rotational speeds of up to 1200rpm.

Consequently, increasing the rotational speed reduces the number of absolute angular positions per revolution (see Table 7). Regardless of the rotational speed or the number of positions to be read out, the absolute angular value is always given at the highest resolution of 10 bit.

The incremental outputs are not affected by rotational speed restrictions due to the implemented interpolator. The incremental output signals may be used for high-speed applications with rotational speeds of up to 30,000 rpm without missing pulses.

Daisy Chain Mode

The Daisy Chain mode allows connection of several AS5040's in series, while still keeping just one digital input for data transfer (see "Data IN" in Figure 6 below). This mode is accomplished by connecting the data output (DO; pin 9) to the data input (Prog; pin 8) of the subsequent device. An RC filter must be implemented between each PROG pin of device n and DO pin of device n+1, to prevent the encoders to enter the alignment mode, in case of ESD discharge, long cables, or not conform signal levels or shape. Using the values R=100R and C=1nF allow a max. CLK frequency of 1MHz on the whole chain. The serial data of all connected devices is read from the DO pin of the first device in the chain. The Prog pin of the last device in the chain should be connected to VSS. The length of the serial bit stream increases with every connected device, it is

n * (16+1) bits:

e.g. 34 bit for two devices, 51 bit for three devices, etc…

The last data bit of the first device (Parity) is followed by a logic low bit and the first data bit of the second device (D9), etc… (see Figure 7).

Programming Daisy Chained Devices

In Daisy Chain mode, the Prog pin is connected directly to the DO pin of the subsequent device in the chain (see Figure 6). During programming (see section 12), a programming voltage of 7.5V must be applied to pin Prog. This voltage level exceeds the limits for pin DO, so one of the following precautions must be made during programming:

- open the connection DO→Prog during programming or
- add a Schottky diode between DO and Prog (Anode = DO, Cathode = Prog)

Due to the parallel connection of CLK and CSn, all connected devices may be programmed simultaneously.

Figure 6: Daisy Chain Hardware Configuration

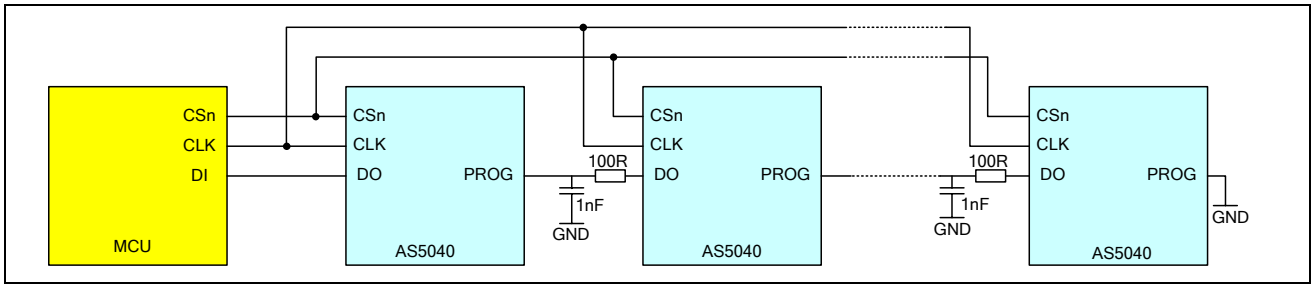
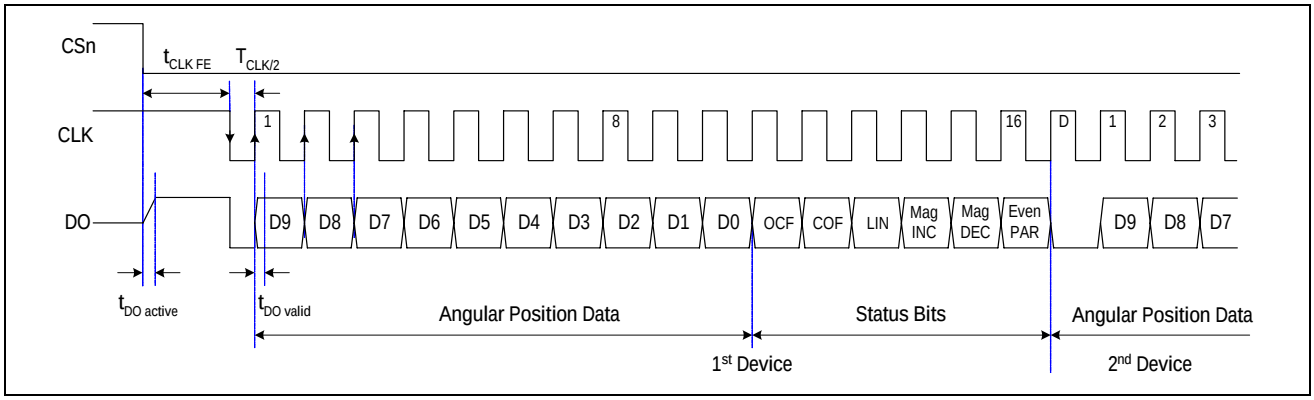


Figure 7: Daisy Chain Mode Data Transfer



8 Incremental Outputs

Three different incremental output modes are possible with quadrature A/B being the default mode.

Figure 8 shows the two-channel quadrature as well as the step / direction incremental signal (LSB) and the direction bit in clockwise (CW) and counter-clockwise (CCW) direction.

Quadrature A/B Output (Quad A/B Mode)

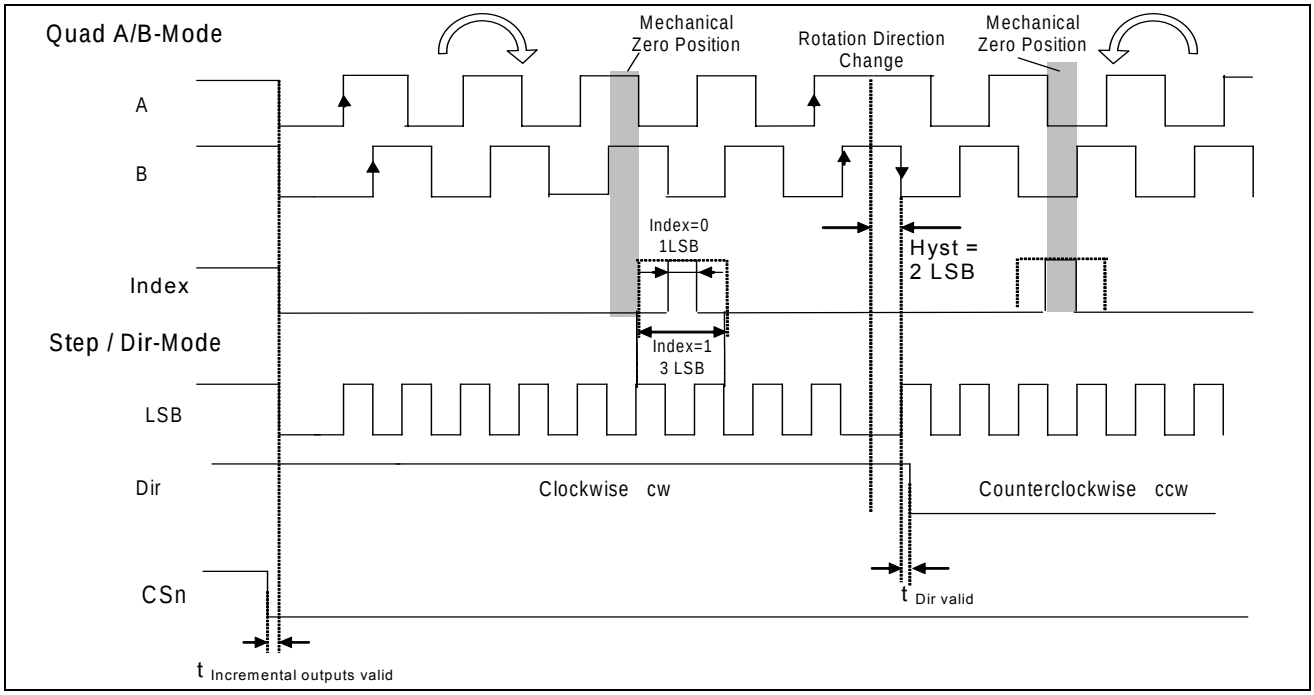
The phase shift between channel A and B indicates the direction of the magnet movement. Channel A leads channel B at a clockwise rotation of the magnet (top view) by 90 electrical degrees. Channel B leads channel A at a counter-clockwise rotation.

LSB Output (Step / Direction Mode)

Output LSB reflects the LSB (least significant bit) of the programmed incremental resolution (OTP Register Bit Div0, Div1). Output Dir provides information about the rotational direction of the magnet, which may be placed above or below the device (1=clockwise; 0=counter clockwise; top view). Dir is updated with every LSB change.

In both modes (quad A/B, step/direction) the resolution and the index output are user programmable. The index pulse indicates the zero position and is by default one angular step (1LSB) wide. However, it can be set to three LSBs by programming the Index-bit of the OTP register accordingly (see Table 6).

Figure 8: Incremental Output Modes



Incremental Power-up Lock Option

After power-up, the incremental outputs can optionally be locked or unlocked, depending on the status of the CSn pin:

CSn = low at power-up:

CSn has an internal pull-up resistor and must be externally pulled low ($R_{ext} \leq 5k\Omega$). If Csn is low at power-up, the incremental outputs (A, B, Index) will be high until the internal offset compensation is finished.
This unique state (A=B=Index = high) may be used as an indicator for the external controller to shorten the waiting time at power-up. Instead of waiting for the specified maximum power up-time (0), the controller can start requesting data from the AS5040 as soon as the state (A=B=Index = high) is cleared.

CSn = high or open at power-up:

In this mode, the incremental outputs (A, B, Index) will remain at logic high state, until CSn goes low or a low pulse is applied at CSn. This mode allows intentional disabling of the incremental outputs until for example the system microcontroller is ready to receive data.

Incremental Output Hysteresis

To avoid flickering incremental outputs at a stationary magnet position, a hysteresis is introduced.

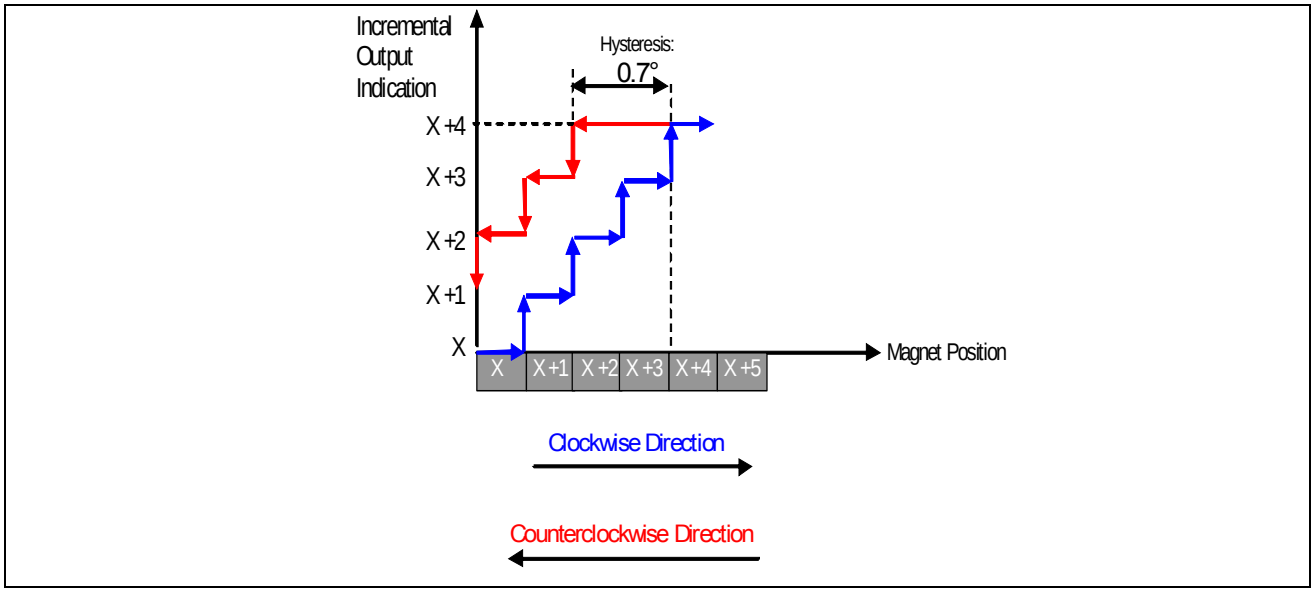
In case of a rotational direction change, the incremental outputs have a hysteresis of 2 LSB.

Regardless of the programmed incremental resolution, the hysteresis of 2 LSB always corresponds to the highest resolution of 10 bit. In absolute terms, the hysteresis is set to 0.704 degrees for all resolutions.

For constant rotational directions, every magnet position change is indicated at the incremental outputs (see Figure 9). If for example the magnet turns clockwise from position „x+3“ to „x+4“, the incremental output would also indicate this position accordingly.

A change of the magnet's rotational direction back to position „x+3“ means, that the incremental output still remains unchanged for the duration of 2 LSB, until position „x+2“ is reached. Following this direction, the incremental outputs will again be updated with every change of the magnet position.

Figure 9: Hysteresis Window for Incremental Outputs



9 Pulse Width Modulation (PWM) Output

The AS5040 provides a pulse width modulated output (PWM), whose duty cycle is proportional to the measured angle:

$$Position = \frac{t_{on} \cdot 1025}{(t_{on} + t_{off})} - 1$$

The PWM frequency is internally trimmed to an accuracy of ±5% (±10% over full temperature range). This tolerance can be cancelled by measuring the complete duty cycle as shown above.

Figure 10: PWM Output Signal

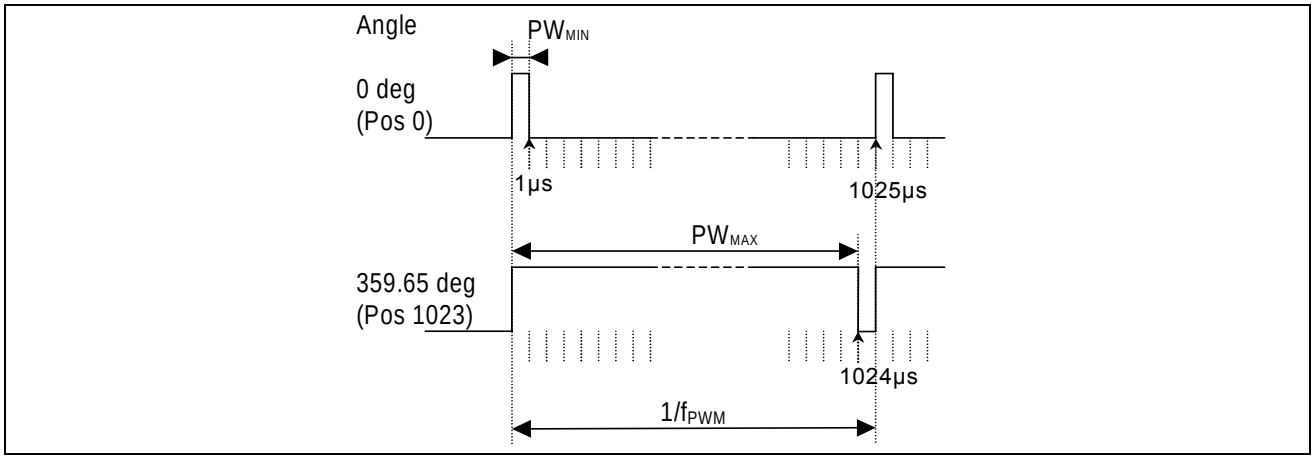


Table 5: PWM Signal Parameters

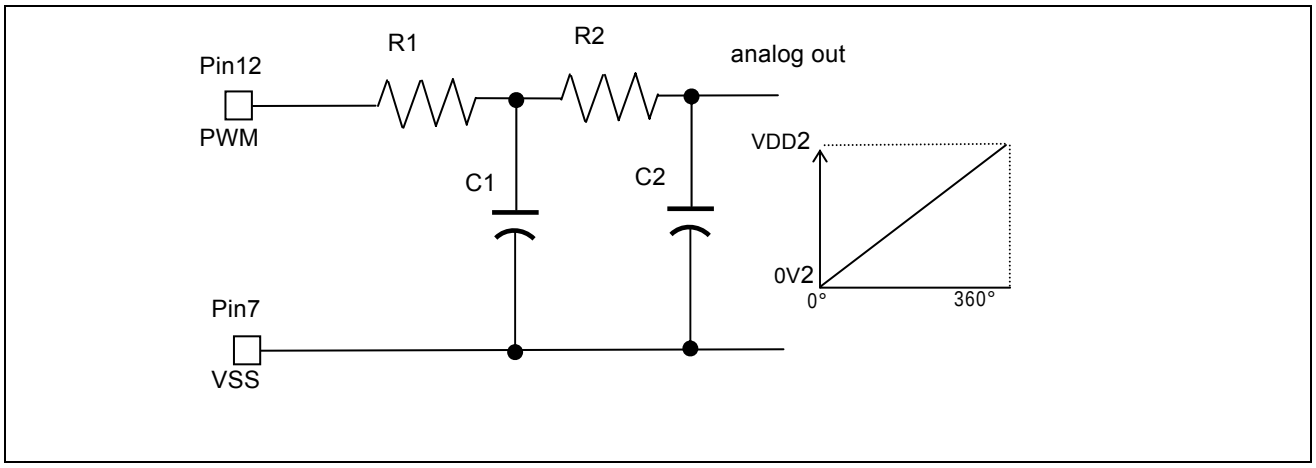
Parameter	Symbol	Typ	Unit	Note
PWM frequency	f_{PWM}	0.9756	kHz	Signal period: 1025µs
MIN pulse width	PW_{MIN}	1	µs	- Position 0d - Angle 0 deg
MAX pulse width	PW_{MAX}	1024	µs	- Position 1023d - Angle 359,65 deg

10 Analog Output

An analog output may be generated by averaging the PWM signal, using an external active or passive lowpass filter. The analog output voltage is proportional to the angle: 0°= 0V; 360° = VDD5V.

Using this method, the AS5040 can be used as direct replacement of potentiometers.

Figure 11: Simple Passive 2nd Order Lowpass Filter



R1,R2 ≥ 4k7 C1,C2 ≥ 1µF / 6V

R1 should be ≥4k7 to avoid loading of the PWM output. Larger values of Rx and Cx will provide better filtering and less ripple, but will also slow down the response time.

11 Brushless DC Motor Commutation Mode

Brushless DC motors require angular information for stator commutation. The AS5040 provides U-V-W commutation signals for one and two pole pair motors. In addition to the three-phase output signals, the step (LSB) output at pin 12 allows high accuracy speed measurement. Two resolutions (9 or 10 bit) can be selected by programming Div0 according to Table 6.

Mode 3.0 (3.1) is used for brush-less DC motors with one-pole pair rotors. The three phases (U, V, W) are 120 degrees apart, each phase is 180 degrees on and 180 degrees off.

Mode 3.2 (3.3) is used for motors with two pole pairs requiring a higher pulse count to ensure a proper current commutation. In this case the pulse width is 256 positions, equal to 90 degrees.

The precise physical angle at which the U, V and W signals change state (“Angle” in Figure 12 and Figure 13) is calculated by multiplying each transition position by the angular value of 1 count:

Angle [deg] = Position x (360 degree / 1024)

Figure 12: U, V and V-signals for BLDC Motor Commutation (Div1=0, Div0=0)

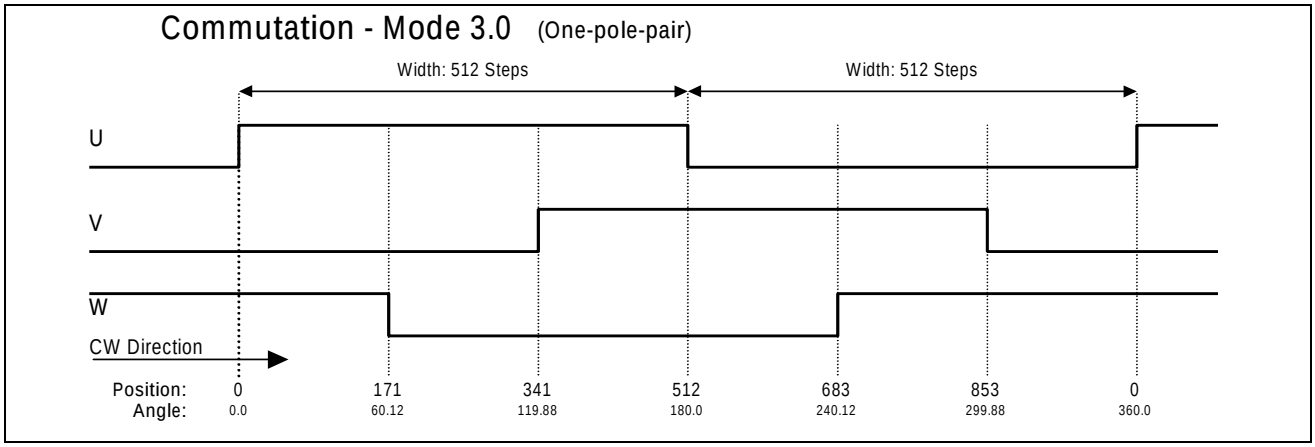
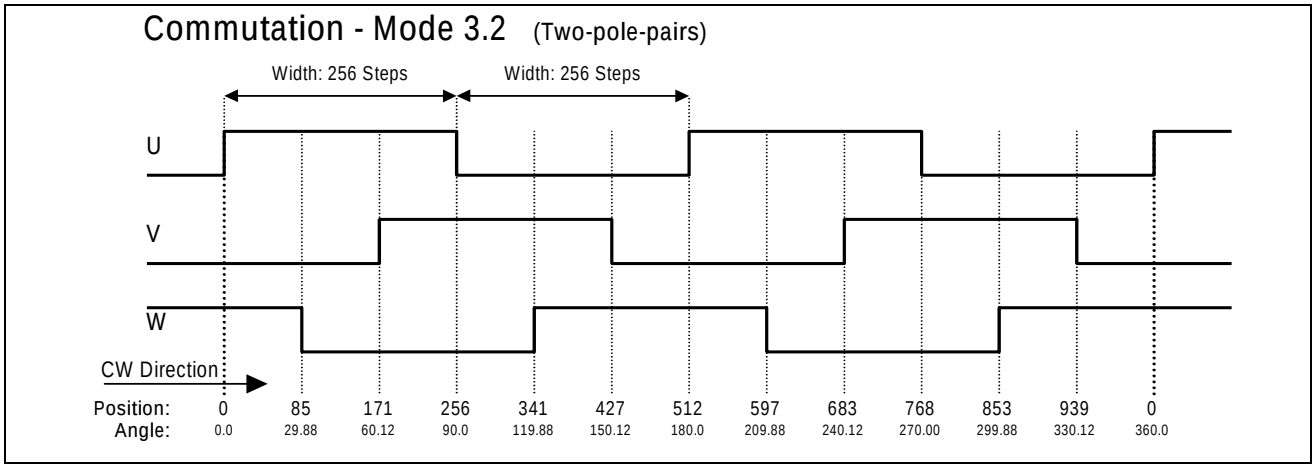


Figure 13: U, V and W-signals for 2-pole BLDC Motor Commutation (Div1=1; Div0=0)



12 Programming the AS5040

After power-on, programming the AS5040 is enabled with the rising edge of CSn with Prog = high and CLK = low. 16 bit configuration data must be serially shifted into the OTP register via the Prog-pin. The first “CCW” bit is followed by the zero position data (MSB first) and the incremental mode setting as shown in Table 6. Data must be valid at the rising edge of CLK (see Figure 14).

After writing data into the OTP register it can be permanently programmed by rising the Prog pin to the programming voltage V_{PROG}. 16 CLK pulses (t_{PROG}) must be applied to program the fuses (Figure 15). To exit the programming mode, the chip must be reset by a power-on-reset. The programmed data is available after the next power-up.

Note: During the programming process, the transitions in the programming current may cause high voltage spikes generated by the inductance of the connection cable. To avoid these spikes and possible damage to the IC, the connection wires, especially the signals Prog and VSS must be kept as short as possible. The maximum wire length between the V_{PROG} switching transistor and pin Prog (see Figure 16) should not exceed 50mm (2 inches). To suppress eventual voltage spikes, a 10nF ceramic capacitor should be connected close to pins Prog and VSS. This capacitor is only required for programming, it is not required for normal operation.

The clock timing t_{clk} must be selected at a proper rate to ensure that the signal Prog is stable at the rising edge of CLK (see Figure 14). Additionally, the programming supply voltage should be buffered with a 10µF capacitor mounted close to the switching transistor. This capacitor aids in providing peak currents during programming.

The specified programming voltage at pin Prog is 7.3 – 7.5V (see section 0). To compensate for the voltage drop across the V_{PROG} switching transistor, the applied programming voltage may be set slightly higher (7.5 - 8.0V, see Figure 16).

OTP Register Contents:

CCW Counter Clockwise Bit
ccw=0 – angular value increases in clockwise direction
ccw=1 – angular value increases in counterclockwise direction

Z [9:0] Programmable Zero / Index Position

Indx Index Pulse Width Selection: 1LSB / 3LSB

Div1,Div0 Divider Setting of Incremental Output

Md1, Md0 Incremental Output Mode Selection

OTP Default Setting

The AS5040 can also be operated without programming. The default, un-programmed setting is shown in Table 6 (Mode 0.0):

CCW: 0 = clockwise operation
Z9 to Z0: 00 = no programmed zero position
Indx: 0 = Index bit width = 1LSB
Div0,Div1 : 00 = incremental resolution = 10bit
Md0, MD1: 00 = incremental mode = quadrature

Figure 14: Programming Access – Write Data (section of Figure 15)

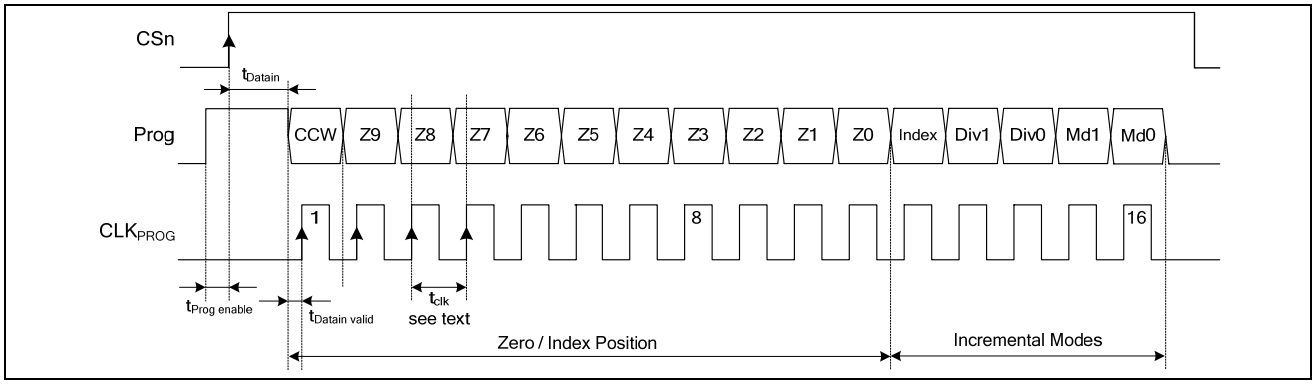


Figure 15: Complete Programming Sequence

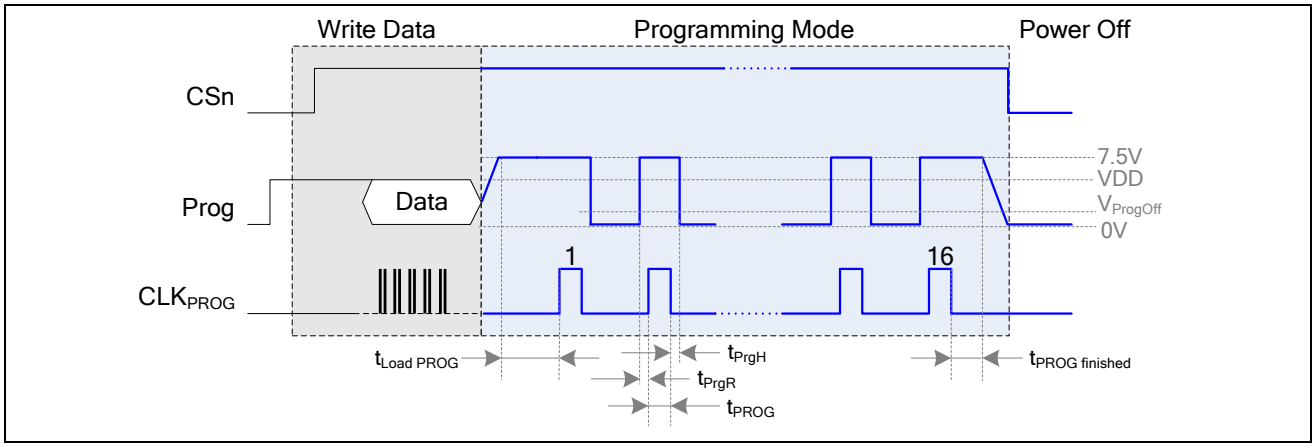
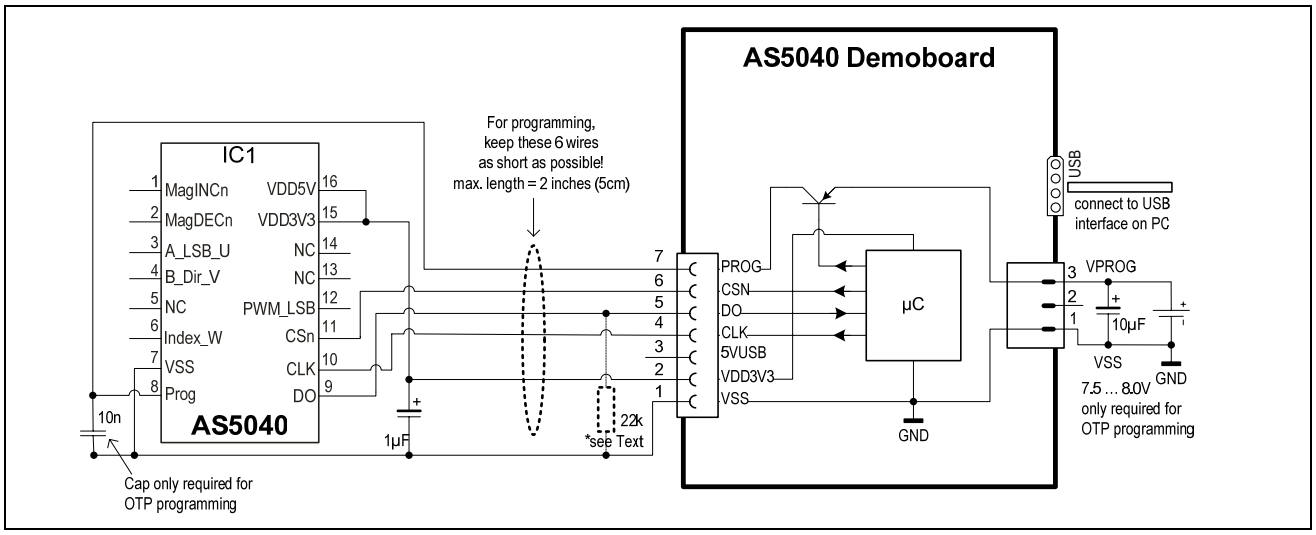


Figure 16: OTP Programming Connection of AS5040 (shown with AS5040 demoboard)



Incremental Mode Programming

Three different incremental output modes are available.

Mode: Md1=0 / Md0=1 sets the AS5040 in quadrature mode.

Mode: Md1=1 / Md0=0 sets the AS5040 in step / direction mode (see **Table 1**)

In both modes, the incremental resolution may be reduced from 10 bit down to 9, 8 or 7 bit using the divider OTP bits Div1 and Div0. (see Table 6 below).

Mode: Md1=1 / Md0=1 sets the AS5040 in brushless DC motor commutation mode with an additional LSB incremental signal at pin 12 (PWM_LSB).

To allow programming of all bits, the default factory setting is all bits = 0. This mode is equal to mode 1:0 (quadrature A/B, 1LSB index width, 256ppr).

The absolute angular output value, by default, increases with clockwise rotation of the magnet (top view).

Setting the CCW-bit (see Figure 14) allows reversing the indicated direction, e.g. when the magnet is placed underneath the IC:

CCW = 0 – angular value increases clockwise;

CCW = 1 – angular value increases counterclockwise.

By default, the zero / index position pulse is one LSB wide. It can be increased to a three LSB wide pulse by setting the Index-bit of the OTP register.

Further programming options (commutation modes) are available for brushless DC motor-control.

Md1 = Md0 = 1 changes the incremental output pins 3, 4 and 6 to a 3-phase commutation signal. Div1 defines the number of pulses per revolution for either a two-pole (Div1=0) or four-pole (Div1=1) rotor.

In addition, the LSB is available at pin 12 (the LSB signal replaces the PWM signal), which allows for high rotational speed measurement of up to 30,000 rpm.

Table 6: One Time Programmable (OTP) Register Options

Mode	OTP-Mode-Register-Bit					Pin #				Pulses per Revolution	Incremental Resolution
	Md1	Md0	Div1	Div0	Index	3	4	6	12	ppr	bit
Default (Mode0.0)	0	0	0*	0*	0*	A	B	1LSB	PWM 10 bit	2x256	10
quadAB-Mode1.0	0	1	0	0	0			1LSB			
quadAB-Mode1.1	0	1	0	0	1			**3LSBs**			
quadAB-Mode1.2	0	1	0	1	0			1LSB		2x128	9
quadAB-Mode1.3	0	1	0	1	1			**3LSBs**			
quadAB-Mode1.4	0	1	1	0	0			1LSB		2x64	8
quadAB-Mode1.5	0	1	1	0	1			**3LSBs**			
quadAB-Mode1.6	0	1	1	1	0			1LSB		2x32	7
quadAB-Mode1.7	0	1	1	1	1			**3LSBs**			
Step/Dir-Mode2.0	1	0	0	0	0	LSB	Dir	1LSB	PWM 10 bit	512	10
Step/Dir-Mode2.1	1	0	0	0	1			**3LSBs**			
Step/Dir -Mode2.2	1	0	0	1	0			1LSB		256	9
Step/Dir -Mode2.3	1	0	0	1	1			**3LSBs**			
Step/Dir -Mode2.4	1	0	1	0	0			1LSB		128	8
Step/Dir -Mode2.5	1	0	1	0	1			**3LSBs**			
Step/Dir -Mode2.6	1	0	1	1	0			1LSB		64	7
Step/Dir -Mode2.7	1	0	1	1	1			**3LSBs**			
Commutation-Mode3.0	1	1	0	0	0	U(0°)	V(120°)	W(240°)	LSB	3 x 1	10
Commutation-Mode3.1	1	1	0	1	0						9
Commutation-Mode3.2	1	1	1	0	0	U' (0°, 180°)	V' (60°, 240°)	W' (120°, 300°)	LSB	2 x 3	10
Commutation-Mode3.3	1	1	1	1	0						9

Note: Div1, Div0 and Index cannot be programmed in Mode 0:0

Zero Position Programming

Zero position programming is an OTP option that simplifies assembly of a system, as the magnet does not need to be manually adjusted to the mechanical zero position. Once the assembly is completed, the mechanical and electrical zero positions can be matched by software. Any position within a full turn can be defined as the permanent new zero/index position.

For zero position programming, the magnet is turned to the mechanical zero position (e.g. the “off”-position of a rotary switch) and the actual angular value is read.

This value is written into the OTP register bits Z9:Z0 (see Figure 14) and programmed as described in section 12.

This new absolute zero position is also the new Index pulse position for incremental output modes.

Note: The zero position value may also be modified before programming, e.g. to program an electrical zero position that is 180° (half turn) from the mechanical zero position, just add 512 to the value read at the mechanical zero position and program the new value into the OTP register.

Repeated OTP Programming

Although a single AS5040 OTP register bit can be programmed only once (from 0 to 1), it is possible to program other, unprogrammed bits in subsequent programming cycles. However, a bit that has already been programmed should not be programmed twice. Therefore it is recommended that bits that are already programmed are set to “0” during a programming cycle.

Non-permanent Programming

It is also possible to re-configure the AS5040 in a non-permanent way by overwriting the OTP register.

This procedure is essentially a “Write Data” sequence (see Figure 14) without a subsequent OTP programming cycle.

The “Write Data” sequence may be applied at any time during normal operation. This configuration remains set while the power supply voltage is above the power-on reset level (see 0).

See Application Note AN5000-20 for further information.

Analog Readback Mode

Non-volatile programming (OTP) uses on-chip zener diodes, which become permanently low resistive when subjected to a specified reverse current.

The quality of the programming process depends on the amount of current that is applied during the programming process (up to 130mA). This current must be provided by an external voltage source. If this voltage source cannot provide adequate power, the zener diodes may not be programmed properly.

In order to verify the quality of the programmed bits, an analog level can be read for each zener diode, giving an indication whether this particular bit was properly programmed or not.

To put the AS5040 in analog readback mode, a digital sequence must be applied to pins CSn, Prog and CLK as shown in Figure 17. The digital level for this pin depends on the supply configuration (3.3V or 5V; see section 14).

The second rising edge on CSn (OutpEN) changes pin Prog to a digital output and the log. high signal at pin Prog must be removed to avoid collision of outputs (grey area in Figure 17).

The following falling slope of CSn changes pin Prog to an analog output, providing a reference voltage V_{ref}, that must be saved as a reference for the calculation of the subsequent programmed and unprogrammed OTP bits. Following this step, each rising slope of CLK outputs one bit of data in the reverse order as during programming. (see Figure 17: Md0-MD1-Div0,Div1-Indx-Z0…Z9, ccw)

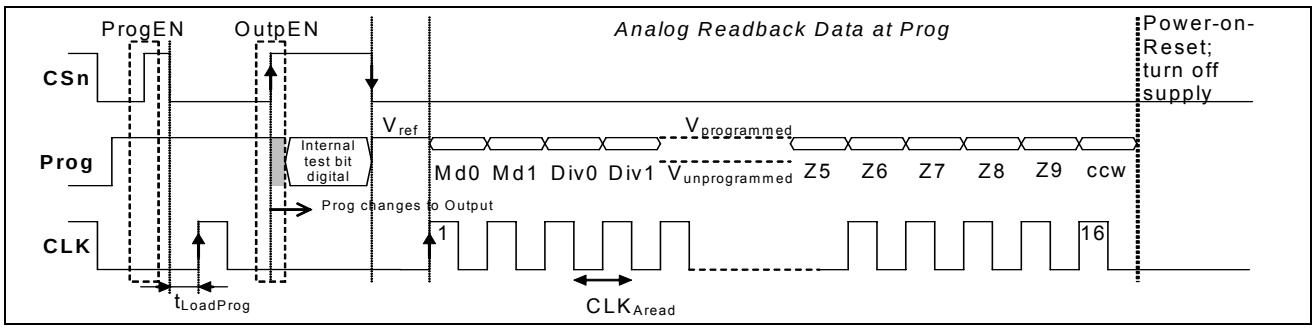
During analog readback, the capacitor at pin Prog (see Figure 16) should be removed to allow a fast readout rate. . If the capacitor is not removed the analog voltage will take longer to stabilize due to the additional capacitance.

The measured analog voltage for each bit must be subtracted from the previously measured V_{ref}, and the resulting value gives an indication on the quality of the programmed bit: a reading of <100mV indicates a properly programmed bit and a reading of >1V indicates a properly unprogrammed bit.

A reading between 100mV and 1V indicates a faulty bit, which may result in an undefined digital value, when the OTP is read at power-up.

Following the 16th clock (after reading bit “ccw”), the chip must be reset by disconnecting the power supply.

Figure 17: OTP Register Analog Read



13 Alignment Mode

The alignment mode simplifies centering the magnet over the chip to gain maximum accuracy and XY-alignment tolerance.

This electrical centering method allows a wider XY-alignment tolerance (0.485mm radius) than mechanical centering (0.25mm radius) as it eliminates the placement tolerance of the die within the IC package (+/- 0.235mm).

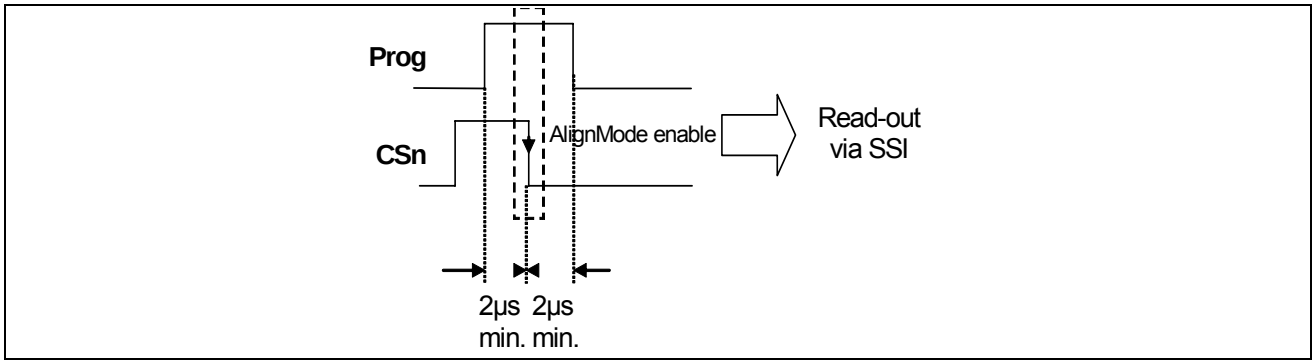
Alignment mode can be enabled with the falling edge of CSn while Prog = logic high (Figure 18). The Data bits D9-D0 of the SSI change to a 10-bit displacement amplitude output. A high value indicates large X or Y displacement, but also higher absolute magnetic field strength. The magnet is properly aligned, when the difference between highest and lowest value over one full turn is at a minimum.

Under normal conditions, a properly aligned magnet will result in a reading of less than 32 over a full turn. The MagINCn and MagDECn indicators will be = 1 when the alignment mode reading is < 32. At the same time, both hardware pins MagINCn (#1) and MagDECn (#2) will be pulled to VSS. A properly aligned magnet will therefore produce a MagINCn = MagDECn = 1 signal throughout a full 360° turn of the magnet.

Stronger magnets or short gaps between magnet and IC may show values larger than 32. These magnets are still properly aligned as long as the difference between highest and lowest value over one full turn is at a minimum.

The alignment mode can be reset to normal operation mode by a power-on-reset (disconnect / re-connect power supply).

Figure 18: Enabling the Alignment Mode



14 3.3V / 5V Operation

The AS5040 operates either at 3.3V ±10% or at 5V ±10%. This is made possible by an internal 3.3V Low-Dropout (LDO) voltage regulator. The internal supply voltage is always taken from the output of the LDO, meaning that the internal blocks are always operating at 3.3V.

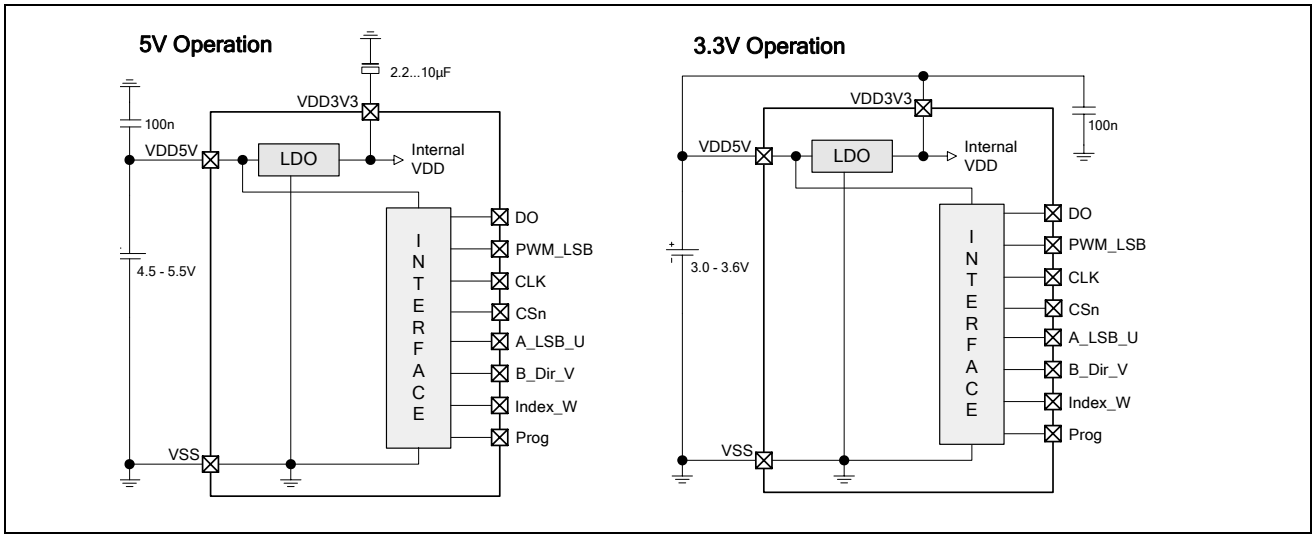
For 3.3V operation, the LDO must be bypassed by connecting VDD3V3 with VDD5V (see Figure 19).

For 5V operation, the 5V supply is connected to pin VDD5V, while VDD3V3 (LDO output) must be buffered by a 2.2...10µF capacitor, which is supposed to be placed close to the supply pin (see Figure 19).

The VDD3V3 output is intended for internal use only It must not be loaded with an external load.

The output voltage of the digital interface I/O's corresponds to the voltage at pin VDD5V, as the I/O buffers are supplied from this pin (see Figure 19).

Figure 19: Connections for 5V / 3.3V Supply Voltages



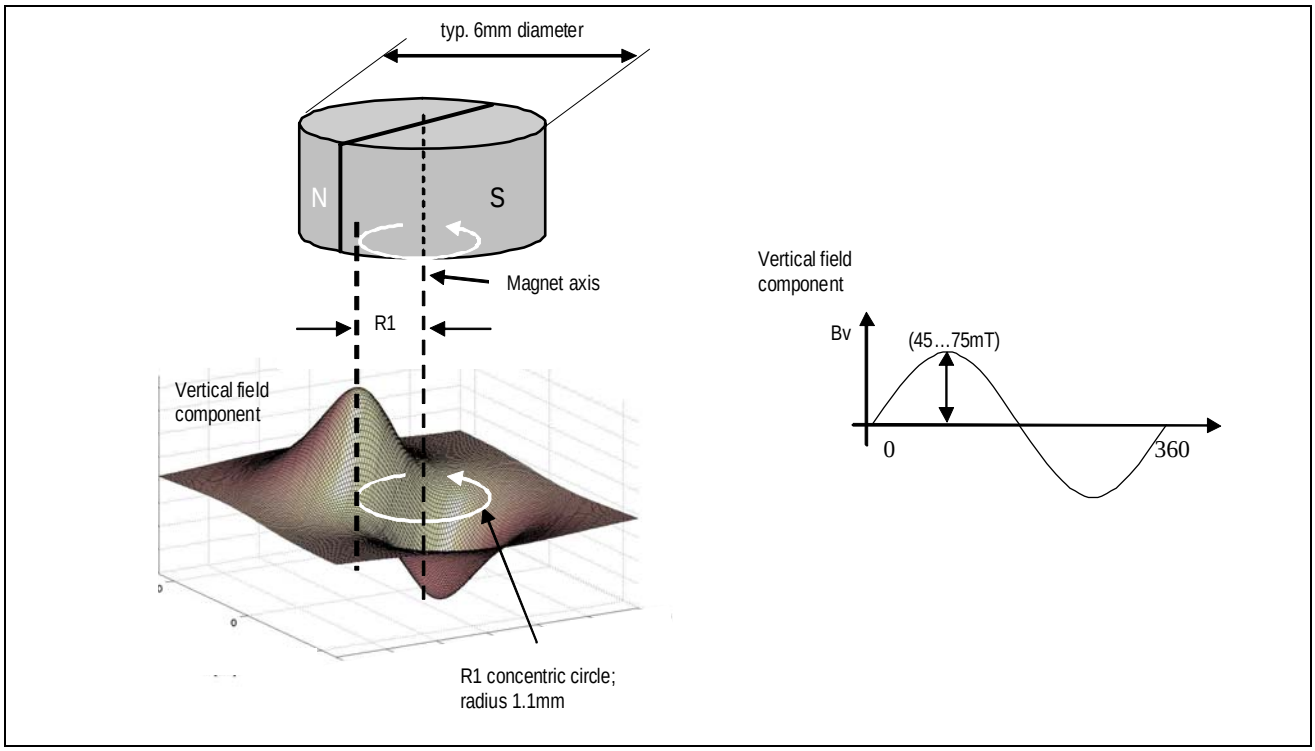
A buffer capacitor of 100nF is recommended in both cases close to pin VDD5V. Note that pin VDD3V3 must always be buffered by a capacitor. It must not be left floating, as this may cause an instable internal 3.3V supply voltage which may lead to larger than normal jitter of the measured angle.

15 Choosing the Proper Magnet

Typically the magnet should be 6mm in diameter and ≥2.5mm in height. Magnetic materials such as rare earth AlNiCo, SmCo5 or NdFeB are recommended.

The magnet's field strength perpendicular to the die surface should be verified using a gauss-meter. The magnetic field B_v at a given distance, along a concentric circle with a radius of 1.1mm (R1), should be in the range of ±45mT…±75mT. (see Figure 20).

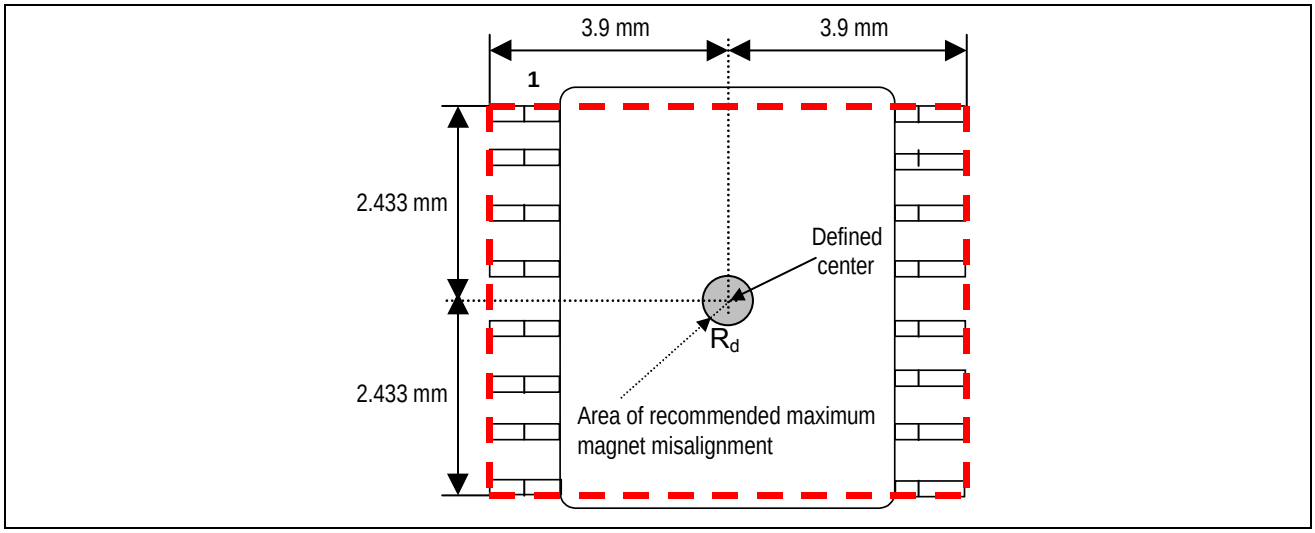
Figure 20: Typical Magnet and Magnetic Field Distribution



Physical Placement of the Magnet

The best linearity can be achieved by placing the center of the magnet exactly over the defined center of the IC package as shown in Figure 21:

Figure 21: Defined IC Center and Magnet Displacement Radius



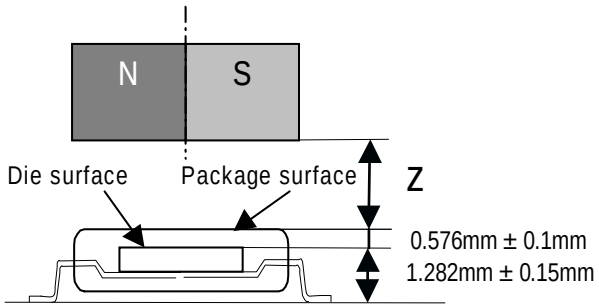
Magnet Placement

The magnet's center axis should be aligned within a displacement radius R_d of 0.25mm from the defined center of the IC with reference to the edge of pin #1 (see Figure 21). This radius includes the placement tolerance of the chip within the SSOP-16 package (+/- 0.235mm). The displacement radius R_d is 0.485mm with reference to the center of the chip (see section 13: Alignment Mode)

The vertical distance should be chosen such that the magnetic field on the die surface is within the specified limits (see Figure 20). The typical distance “z” between the magnet and the package surface is 0.5mm to 1.8mm with the recommended magnet (6mm x 2.5mm). Larger gaps are possible, as long as the required magnetic field strength stays within the defined limits.

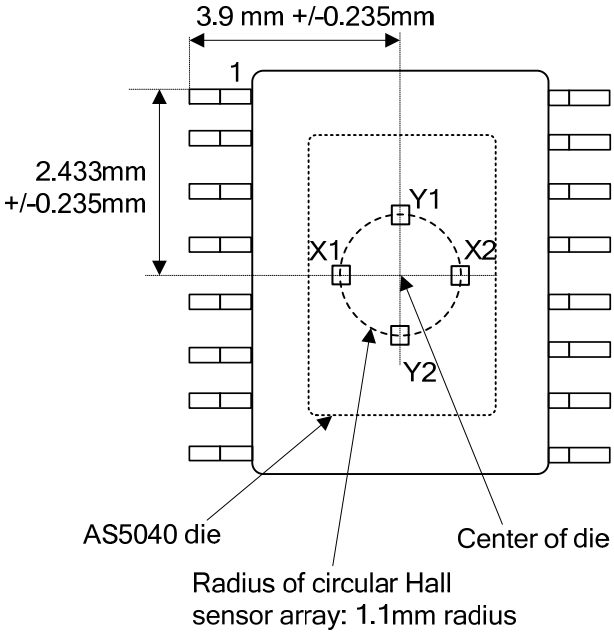
A magnetic field outside the specified range may still produce usable results, but the out-of-range condition will be indicated by MagINCn (pin 1) and MagDECn (pin 2), see Table 3.

Figure 22: Vertical Placement of the Magnet



16 Simulation Modelling

Figure 23: Arrangement of Hall Sensor Array on Chip (principle)



With reference to Figure 23, a diametrically magnetized permanent magnet is placed above or below the surface of the AS5040. The chip uses an array of Hall sensors to sample the vertical vector of a magnetic field distributed across the device package surface. The area of magnetic sensitivity is a circular locus of 1.1mm radius with respect to the center of the die. The Hall sensors in the area of magnetic sensitivity are grouped and configured such that orthogonally related components of the magnetic fields are sampled differentially.

The differential signal Y1-Y2 will give a sine vector of the magnetic field. The differential signal X1-X2 will give an orthogonally related cosine vector of the magnetic field.

The angular displacement (Θ) of the magnetic source with reference to the Hall sensor array may then be modelled by:

$$\Theta = \arctan\frac{(Y1 - Y2)}{(X1 - X2)} \pm 0.5°$$

The ±0.5° angular error assumes a magnet optimally aligned over the center of the die and is a result of gain mismatch errors of the AS5040. Placement tolerances of the die within the package are ±0.235mm in X and Y direction, using a reference point of the edge of pin #1 (Figure 23).

In order to neglect the influence of external disturbing magnetic fields, a robust differential sampling and ratiometric calculation algorithm has been implemented. The differential sampling of the sine and cosine vectors removes any common mode error due to DC components introduced by the magnetic source itself or external disturbing magnetic fields. A ratiometric division of the sine and cosine vectors removes the need for an accurate absolute magnitude of the magnetic field and thus accurate Z-axis alignment of the magnetic source.

The recommended differential input range of the magnetic field strength ($B_{(X1-X2)}$,$B_{(Y1-Y2)}$) is ±75mT at the surface of the die. In addition to this range, an additional offset of ±5mT, caused by unwanted external stray fields is allowed.

The chip will continue to operate, but with degraded output linearity, if the signal field strength is outside the recommended range. Too strong magnetic fields will introduce errors due to saturation effects in the internal preamplifiers. Too weak magnetic fields will introduce errors due to noise becoming more dominant.

17 Failure Diagnostics

The AS5040 also offers several diagnostic and failure detection features:

Magnetic Field Strength Diagnosis

By software: the MagINCn and MagDECn status bits will both be high when the magnetic field is out of range.

By hardware: Pins #1 (MagINCn) and #2 (MagDECn) are open-drain outputs and will both be turned on (= low with external pull-up resistor) when the magnetic field is out of range. If only one of the outputs is low, the magnet is either moving towards the chip (MagINCn) or away from the chip (MagDECn).

Power Supply Failure Detection

By software: If the power supply to the AS5040 is interrupted, the digital data read by the SSI will be all “0”s. Data is only valid, when bit OCF is high, hence a data stream with all “0”s is invalid. To ensure adequate low levels in the failure case, a pull-down resistor (~10kΩ) should be added between pin DO and VSS at the receiving side.

By hardware: The MagINCn and MagDECn pins are open drain outputs and require external pull-up resistors. In normal operation, these pins are high ohmic and the outputs are high (see Table 3). In a failure case, either when the magnetic field is out of range or the power supply is missing, these outputs will become low. To ensure adequate low levels in case of a broken power supply to the AS5040, the pull-up resistors (>10kΩ) from each pin must be connected to the positive supply at pin 16 (VDD5V).

By hardware: PWM output: The PWM output is a constant stream of pulses with 1kHz repetition frequency. In case of power loss, these pulses are missing.

By hardware: Incremental outputs: In normal operation, pins A(#3), B(#4) and Index (#6) will never be high at the same time, as Index is only high when A=B=low. However, after a power-on-reset, if VDD is powered up or restarts after a power supply interruption, all three outputs will remain in high state until pin CSn is pulled low. If CSn is already tied to VSS during power-up, the incremental outputs will all be high until the internal offset compensation is finished (within t_{PwrUp}).

18 Angular Output Tolerances

Accuracy

Accuracy is defined as the error between measured angle and actual angle. It is influenced by several factors:

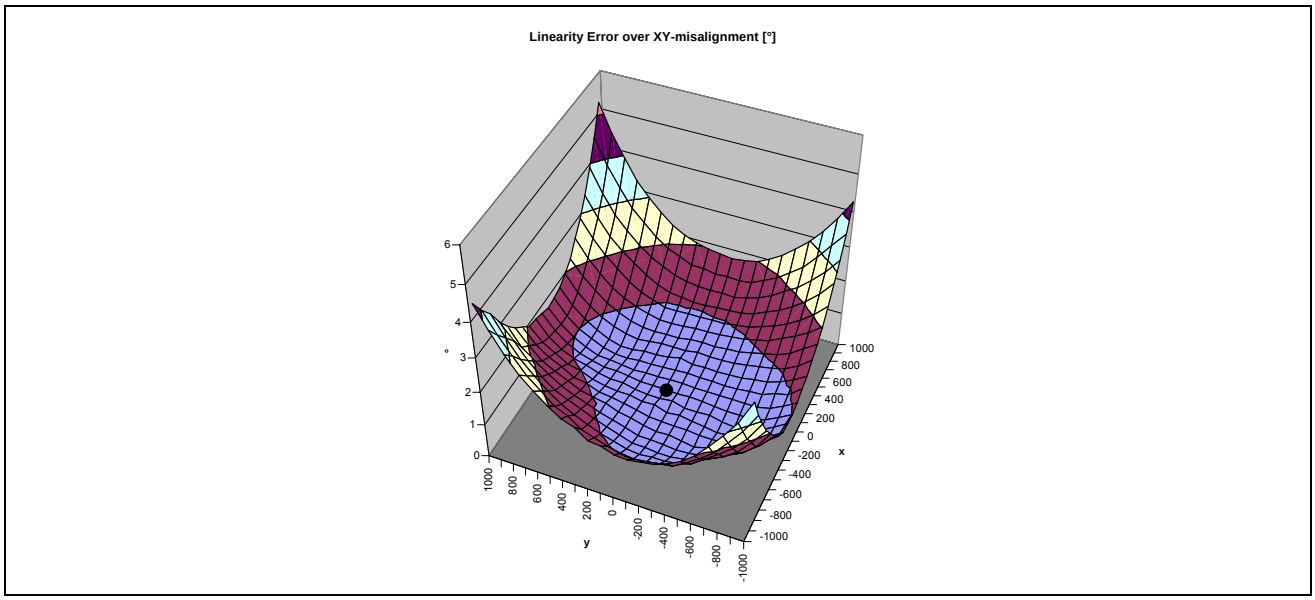
- the non-linearity of the analog-digital converters,
- internal gain and mismatch errors,
- non-linearity due to misalignment of the magnet

As a sum of all these errors, the accuracy with centered magnet = $(Err_{max} – Err_{min})/2$ is specified as better than ±0.5 degrees @ 25°C (see Figure 25).

Misalignment of the magnet further reduces the accuracy. Figure 24 shows an example of a 3D-graph displaying non-linearity over XY-misalignment. The center of the square XY-area corresponds to a centered magnet (see dot in the center of the graph). The X- and Y- axis extends to a misalignment of ±1mm in both directions. The total misalignment area of the graph covers a square of 2x2 mm (79x79mil) with a step size of 100µm.

For each misalignment step, the measurement as shown in Figure 25 is repeated and the accuracy $(Err_{max} – Err_{min})/2$ (e.g. 0.25° in Figure 25) is entered as the Z-axis in the 3D-graph.

Figure 24: Example of Linearity Error over XY Misalignment

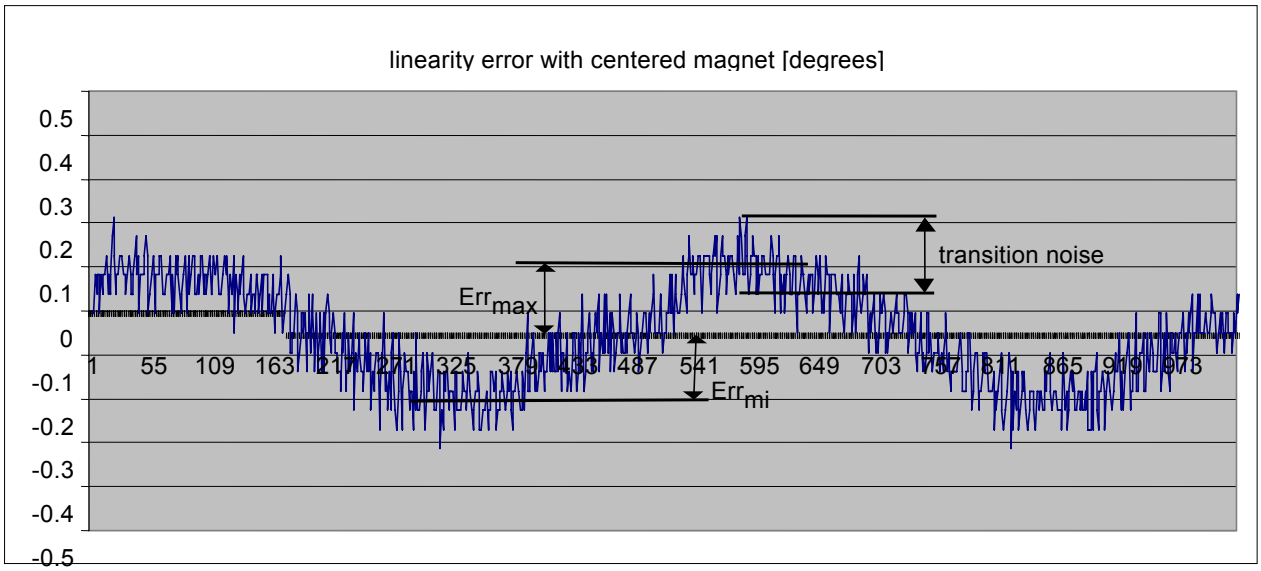


The maximum non-linearity error on this example is better than ±1 degree (inner circle) over a misalignment radius of ~0.7mm. For volume production, the placement tolerance of the IC within the package (±0.235mm) must also be taken into account.

The total nonlinearity error over process tolerances, temperature and a misalignment circle radius of 0.25mm is specified better than ±1.4 degrees.

The magnet used for this measurement was a cylindrical NdFeB (Bomatec® BMN-35H) magnet with 6mm diameter and 2.5mm in height.

Figure 25: Example of Linearity Error over 360°



Transition Noise

Transition noise is defined as the jitter in the transition between two steps.

Due to the nature of the measurement principle (Hall sensors + Preamplifier + ADC), there is always a certain degree of noise involved.

This transition noise voltage results in an angular transition noise at the outputs. It is specified as 0.12 degrees rms (1 sigma)[*1].

This is the repeatability of an indicated angle at a given mechanical position.

The transition noise has different implications on the type of output that is used:

- Absolute output; SSI interface:
  The transition noise of the absolute output can be reduced by the user by applying an averaging of readings. An averaging of 4 readings will reduce the transition noise by 6dB or 50%, e.g. from 0.12°rms to 0.06°rms (1 sigma).
- PWM interface:
  If the PWM interface is used as an analog output by adding a low pass filter, the transition noise can be reduced by lowering the cutoff frequency of the filter.
  If the PWM interface is used as a digital interface with a counter at the receiving side, the transition noise may again be reduced by averaging of readings.
- Incremental mode:
  In incremental mode, the transition noise influences the period, width and phase shift of the output signals A, B and Index. However, the algorithm used to generate the incremental outputs guarantees no missing or additional pulses even at high speeds (up to 30,000 rpm and higher)

*[*1]: statistically, 1 sigma represents 68.27% of readings,*
3 sigma represents 99.73% of readings.

High Speed Operation

Sampling Rate

The AS5040 samples the angular value at a rate of 10.42k samples per second. Consequently, the incremental, as well as the absolute outputs are updated each 96µs. At a stationary position of the magnet, this sampling rate creates no additional error.

Absolute Mode with Serial Communication:

With the given sampling rate of 10.4 kHz, the number of samples (n) per turn for a magnet rotating at high speed can be calculated by:

$$n = \frac{60}{rpm \cdot 96\mu s}$$

In practice, there is no upper speed limit. The only restriction is that there will be fewer samples per revolution as the speed increases.

Regardless of the rotational speed, the absolute angular value is always sampled at the highest resolution of 10 bit. Likewise, for a given number of samples per revolution (n), the maximum speed can be calculated by:

$$rpm = \frac{60}{n \cdot 96\mu s}$$

In absolute mode with serial communication, 610 rpm is the maximum speed, where 1024 readings per revolution can be obtained.

In incremental mode, the maximum error caused by the sampling rate of the ADCs is 0/+96µs. It has a peak of 1LSB = 0.35° at 610 rpm. At higher speeds this error is reduced again due to interpolation and the output delay remains at 192µs as the DSP requires two sampling periods (2x96µs) to synthesize and redistribute any missing pulses.

Absolute Mode with PWM:

The principle is the same as with the serial communication. The PWM output is refreshed with a rate of 1.025ms, the number of samples (n) per turn for a magnet rotating at high speed can be calculated by:

$$n = \frac{60}{rpm \cdot 1.025ms}$$

In absolute mode with PWM output, 57 rpm is the maximum speed, where 1024 readings per revolution can be obtained.

Incremental Mode:

Incremental encoders are usually required to produce no missing pulses up to several thousand rpm's.

Therefore, the AS5040 has a built-in interpolator, which ensures that there are no missing pulses at the incremental outputs for rotational speeds of up to 30,000 rpm, even at the highest resolution of 10 bits (512 pulses per revolution).

Table 7: Speed Performance

Absolute Output Mode	Incremental Output Mode
610rpm = 1024 samples / turn	no missing pulses @ 10 bit resolution (512ppr): max. speed = 30,000 rpm
1220rpm = 512 samples / turn	
2441rpm = 256 samples / turn	
etc…	

Propagation Delays

The propagation delay is the delay between the time that a sample is taken until it is converted and available as angular data. This delay is 48µs for the absolute interface and 192µs for the incremental interface.

Using the SSI interface for absolute data transmission, an additional delay must be considered, caused by the asynchronous sampling (t= 0…1/f_s) and the time it takes the external control unit to read and process the data.

Angular Error Caused by Propagation Delay

A rotating magnet will therefore cause an angular error caused by the output delay. This error increases linearly with speed:

$$e_{sampling} = rpm * 6 * prop.delay$$

where: $e_{sampling}$ = angular error [°]
rpm = rotating speed [rpm]
$prop.delay$ = propagation delay [seconds]

Note: since the propagation delay is known, it can be automatically compensated by the control unit that is processing the data from the AS5040, thus reducing the angular error caused by speed.

Internal Timing Tolerance

The AS5040 does not require an external ceramic resonator or quartz. All internal clock timings for the AS5040 are generated by an on-chip RC oscillator. This oscillator is factory trimmed to ±5% accuracy at room temperature (±10% over full temperature range). This tolerance influences the ADC sampling rate and the pulse width of the PWM output:

- Absolute output; SSI interface:
  A new angular value is updated every 100µs (typ.)
- Incremental outputs:
  the incremental outputs are updated every
  100µs (typ.)
- PWM output:
  A new angular value is updated every 100µs (typ.).
  The PWM pulse timings T_{on} and T_{off} also have the same tolerance as the internal oscillator.
  If only the PWM pulse width T_{on} is used to measure the angle, the resulting value also has this timing tolerance.
  However, this tolerance can be cancelled by measuring both T_{on} and T_{off} and calculating the angle from the duty cycle (see section 9):

$$Position = \frac{t_{on} \cdot 1025}{(t_{on} + t_{off})} - 1$$

Temperature

Magnetic Temperature Coefficient

One of the major benefits of the AS5040 compared to linear Hall sensors is that it is much less sensitive to temperature. While linear Hall sensors require a compensation of the magnet's temperature coefficient, the AS5040 automatically compensates for the varying magnetic field strength over temperature. The magnet's temperature drift does not need to be considered, as the AS5040 operates with magnetic field strengths from ±45…±75mT.

Example:
A NdFeB magnet has a field strength of 75mT @ –40°C and a temperature coefficient of -0.12% per Kelvin.
The temperature change is from –40° to +125° = 165K.
The magnetic field change is: 165 x -0.12% = -19.8%, which corresponds to 75mT at –40°C and 60mT at 125°C.
The AS5040 can compensate for this temperature related field strength change automatically, no user adjustment is required.

Accuracy over Temperature

The influence of temperature in the absolute accuracy is very low. While the accuracy is ≤ ±0.5° at room temperature, it may increase to ≤±0.9° due to increasing noise at high temperatures.

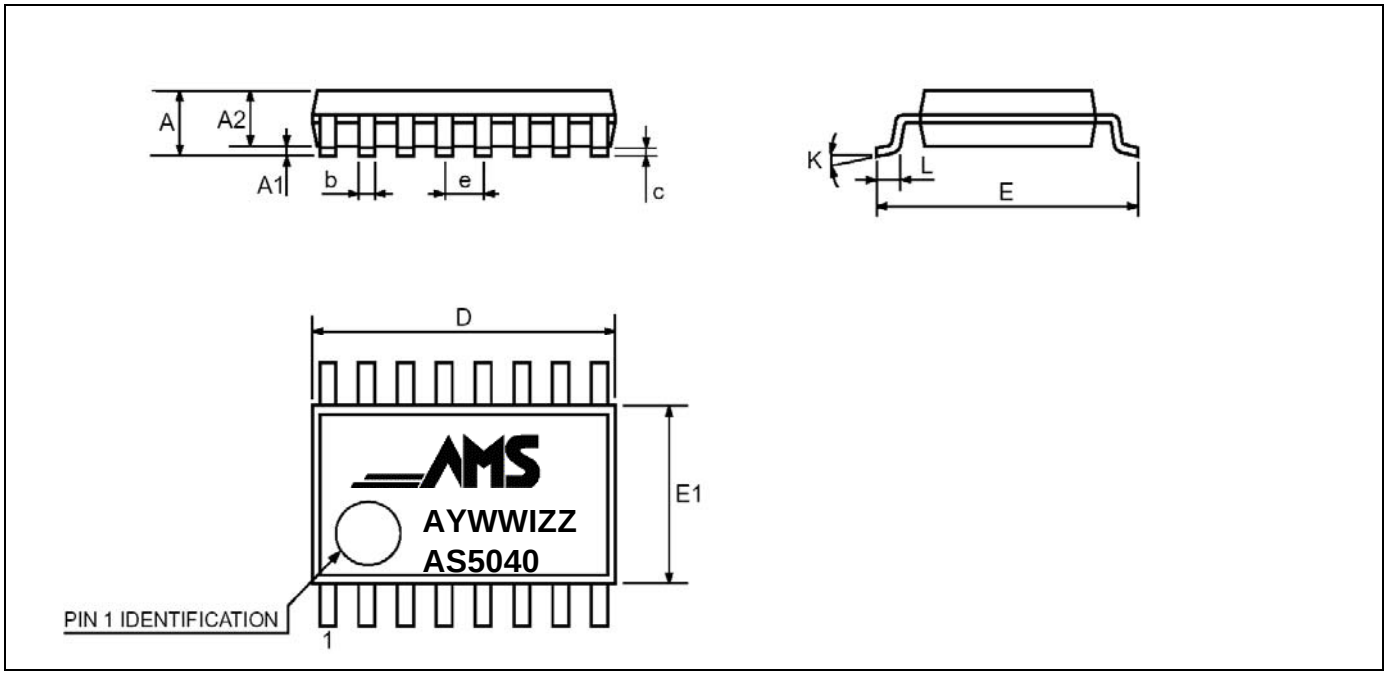
Timing Tolerance over Temperature

The internal RC oscillator is factory trimmed to ±5%. Over temperature, this tolerance may increase to ±10%. Generally, the timing tolerance has no influence in the accuracy or resolution of the system, as it is used mainly for internal clock generation.

The only concern to the user is the width of the PWM output pulse, which relates directly to the timing tolerance of the internal oscillator. This influence however can be cancelled by measuring the complete PWM duty cycle (see Internal Timing Tolerance).

19 Package Drawings and Markings

Figure 26: 16-Lead Shrink Small Outline Package SSOP-16



Dimensions						
Symbol	mm			inch		
	Min	Typ	Max	Min	Typ	Max
A	1.73	1.86	1.99	.068	.073	.078
A1	0.05	0.13	0.21	.002	.005	.008
A2	1.68	1.73	1.78	.066	.068	.070
b	0.25	0.315	0.38	.010	.012	.015
c	0.09	-	0.20	.004	-	.008
D	6.07	6.20	6.33	.239	.244	.249
E	7.65	7.8	7.9	.301	.307	.311
E1	5.2	5.3	5.38	.205	.209	.212
e	0.65			.0256		
K	0°	-	8°	0°	-	8°
L	0.63	0.75	0.95	.025	.030	.037

Marking: AYWWIZZ

A: Pb-free Identifier
Y: Last Digit of Manufacturing Year
WW: Manufacturing Week
I: Plant Identifier
ZZ: Traceability Code

JEDEC Package Outline Standard:
MO - 150 AC

Thermal Resistance $R_{th(j-a)}$:
typ. 151 K/W in still air, soldered on PCB

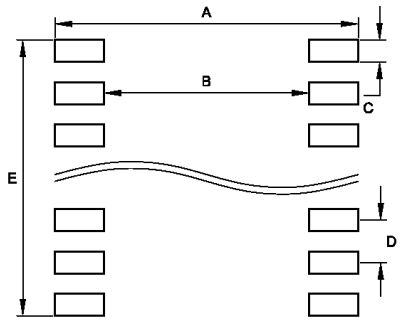
IC's marked with a white dot or the letters "ES" denote Engineering Samples

20 Packing Options

Delivery: Tape and Reel (1 reel = 2000 devices)
Tubes (1 box = 100 tubes à 77 devices)

Order # AS5040ASSU for delivery in tubes
Order # AS5040ASST for delivery in tape and reel

21 Recommended PCB Footprint



Recommended Footprint Data		
	mm	inch
A	9.02	0.355
B	6.16	0.242
C	0.46	0.018
D	0.65	0.025
E	5.01	0.197

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