

FEATURES

- Fixed gain of 24.1 dB**
- Broadband operation from 30 MHz to 6 GHz**
- Input/output internally matched to 50 Ω**
- Integrated bias control circuit**
- OIP3 of 36.4 dBm at 900 MHz**
- P1dB of 18.1 dBm at 900 MHz**
- Noise figure of 2.9 dB at 900 MHz**
- Single 5 V power supply**
- Low quiescent current of 56 mA**
- Wide operating temperature range of -40°C to $+105^{\circ}\text{C}$**
- Thermally efficient SOT-89 package**
- ESD rating of ± 1.5 kV (Class 1C)**

GENERAL DESCRIPTION

The [ADL5545](#) is a single-ended RF/IF gain block amplifier that provides broadband operation from 30 MHz to 6 GHz. The [ADL5545](#) provides over 36 dBm of OIP3 using only 56 mA from a 5 V supply.

The [ADL5545](#) provides a gain of 24 dB, which is stable over frequency, temperature, power supply, and from device to device. The amplifier is offered in the industry-standard SOT-89 package and is internally matched to 50 Ω at the input and output, making the [ADL5545](#) very easy to implement in a wide variety of applications. The only external components required are the input/output ac coupling capacitors, power supply decoupling capacitors, and dc bias inductor.

FUNCTIONAL BLOCK DIAGRAM

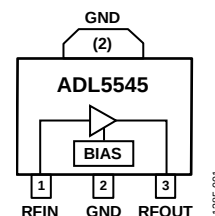


Figure 1.

The [ADL5545](#) is fabricated on an InGaP HBT process and has a high ESD rating of ± 1.5 kV (Class 1C). The [ADL5545](#) is also fully specified for operation across the wide temperature range of -40°C to $+105^{\circ}\text{C}$. A fully populated RoHS-compliant evaluation board is available.

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REVISION HISTORY

5/2020—Rev. A to Rev. B

Deleted Typical Scattering Parameters (S-Parameters) Section and Table 2; Renumbered Sequentially..... 5

11/2013—Rev. 0 to Rev. A

Added Figure 25, Renumbered Sequentially..... 12
 Changes to Figure 34 15
 Updated Outline Dimensions..... 18

4/2013—Revision 0: Initial Version

SPECIFICATIONS

$V_{POS} = 5\text{ V}$ and $T_A = 25^\circ\text{C}$, unless otherwise noted.

Table 1.

Parameter	Test Conditions/Comments	Min	Typ	Max	Unit
OVERALL FUNCTION					
Frequency Range		30		6000	MHz
FREQUENCY = 30 MHz					
Gain			22.6		dB
Output 1 dB Compression Point			12.0		dBm
Output Third-Order Intercept	$\Delta f = 1\text{ MHz}$, output power (P_{OUT}) = -3 dBm per tone		30.3		dBm
Noise Figure			3.8		dB
FREQUENCY = 140 MHz					
Gain			24.8		dB
vs. Frequency	$\pm 10\text{ MHz}$		± 0.03		dB
vs. Temperature	$-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$		± 0.6		dB
vs. Supply	4.75 V to 5.25 V		± 0.30		dB
Output 1 dB Compression Point			14.6		dBm
Output Third-Order Intercept	$\Delta f = 1\text{ MHz}$, output power (P_{OUT}) = -3 dBm per tone		31.5		dBm
Noise Figure			3.1		dB
FREQUENCY = 350 MHz					
Gain			24.6		dB
vs. Frequency	$\pm 10\text{ MHz}$		± 0.01		dB
vs. Temperature	$-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$		± 0.6		dB
vs. Supply	4.75 V to 5.25 V		± 0.30		dB
Output 1 dB Compression Point			16.1		dBm
Output Third-Order Intercept	$\Delta f = 1\text{ MHz}$, output power (P_{OUT}) = -3 dBm per tone		32.6		dBm
Noise Figure			3.3		dB
FREQUENCY = 700 MHz					
Gain		23.7	24.4	25.1	dB
vs. Frequency	$\pm 50\text{ MHz}$		± 0.05		dB
vs. Temperature	$-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$		± 0.5		dB
vs. Supply	4.75 V to 5.25 V		± 0.13		dB
Output 1 dB Compression Point			17.9		dBm
Output Third-Order Intercept	$\Delta f = 1\text{ MHz}$, output power (P_{OUT}) = -3 dBm per tone		38.8		dBm
Noise Figure			3.0		dB
FREQUENCY = 900 MHz					
Gain		23.4	24.1	24.8	dB
vs. Frequency	$\pm 50\text{ MHz}$		± 0.07		dB
vs. Temperature	$-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$		± 0.5		dB
vs. Supply	4.75 V to 5.25 V		± 0.13		dB
Output 1 dB Compression Point			18.1		dBm
Output Third-Order Intercept	$\Delta f = 1\text{ MHz}$, output power (P_{OUT}) = -3 dBm per tone		36.4		dBm
Noise Figure			2.9		dB
FREQUENCY = 1900 MHz					
Gain		21.5	22.2	22.9	dB
vs. Frequency	$\pm 50\text{ MHz}$		± 0.11		dB
vs. Temperature	$-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$		± 0.7		dB
vs. Supply	4.75 V to 5.25 V		± 0.17		dB
Output 1 dB Compression Point			16.6		dBm
Output Third-Order Intercept	$\Delta f = 1\text{ MHz}$, output power (P_{OUT}) = -3 dBm per tone		35.2		dBm
Noise Figure			3.4		dB

Parameter	Test Conditions/Comments	Min	Typ	Max	Unit
FREQUENCY = 2140 MHz					
Gain		21.1	21.8	22.4	dB
vs. Frequency	± 50 MHz		± 0.09		dB
vs. Temperature	$-40^{\circ}\text{C} \leq T_A \leq +85^{\circ}\text{C}$		± 0.7		dB
vs. Supply	4.75 V to 5.25 V		± 0.17		dB
Output 1 dB Compression Point			16.2		dBm
Output Third-Order Intercept	$\Delta f = 1$ MHz, output power (P_{OUT}) = -3 dBm per tone		35.7		dBm
Noise Figure			3.5		dB
FREQUENCY = 2600 MHz					
Gain		20.1	20.9	21.7	dB
vs. Frequency	± 50 MHz		± 0.09		dB
vs. Temperature	$-40^{\circ}\text{C} \leq T_A \leq +85^{\circ}\text{C}$		± 0.7		dB
vs. Supply	4.75 V to 5.25 V		± 0.16		dB
Output 1 dB Compression Point			15.7		dBm
Output Third-Order Intercept	$\Delta f = 1$ MHz, output power (P_{OUT}) = -3 dBm per tone		34.6		dBm
Noise Figure			3.6		dB
FREQUENCY = 3500 MHz					
Gain		19.0	19.7	20.4	dB
vs. Frequency	± 50 MHz		± 0.10		dB
vs. Temperature	$-40^{\circ}\text{C} \leq T_A \leq +85^{\circ}\text{C}$		± 0.8		dB
vs. Supply	4.75 V to 5.25 V		± 0.17		dB
Output 1 dB Compression Point			14.5		dBm
Output Third-Order Intercept	$\Delta f = 1$ MHz, output power (P_{OUT}) = -3 dBm per tone		33.7		dBm
Noise Figure			4.0		dB
FREQUENCY = 4000 MHz					
Gain		17.8	18.6	19.4	dB
vs. Frequency	± 50 MHz		± 0.14		dB
vs. Temperature	$-40^{\circ}\text{C} \leq T_A \leq +85^{\circ}\text{C}$		± 1.0		dB
vs. Supply	4.75 V to 5.25 V		± 0.19		dB
Output 1 dB Compression Point			13.1		dBm
Output Third-Order Intercept	$\Delta f = 1$ MHz, output power (P_{OUT}) = -3 dBm per tone		29.0		dBm
Noise Figure			4.6		dB
FREQUENCY = 5000 MHz					
Gain			16.8		dB
vs. Frequency	± 50 MHz		± 0.05		dB
vs. Temperature	$-40^{\circ}\text{C} \leq T_A \leq +85^{\circ}\text{C}$		± 1.2		dB
vs. Supply	4.75 V to 5.25 V		± 0.20		dB
Output 1 dB Compression Point			9.9		dBm
Output Third-Order Intercept	$\Delta f = 1$ MHz, output power (P_{OUT}) = -3 dBm per tone		31.8		dBm
Noise Figure			4.8		dB
FREQUENCY = 5800 MHz					
Gain			15.9		dB
vs. Frequency	± 50 MHz		± 0.20		dB
vs. Temperature	$-40^{\circ}\text{C} \leq T_A \leq +85^{\circ}\text{C}$		± 1.3		dB
vs. Supply	4.75 V to 5.25 V		± 0.20		dB
Output 1 dB Compression Point			9.4		dBm
Output Third-Order Intercept	$\Delta f = 1$ MHz, output power (P_{OUT}) = -3 dBm per tone		28.4		dBm
Noise Figure			5.2		dB
POWER INTERFACE					
Supply Voltage	V_{POS}	4.75	5	5.25	V
Supply Current			56	70	mA
vs. Temperature	$-40^{\circ}\text{C} \leq T_A \leq +85^{\circ}\text{C}$		-6		mA
Power Dissipation	$V_{\text{POS}} = 5$ V		280		mW

ABSOLUTE MAXIMUM RATINGS

Table 2.

Parameter	Rating
Supply Voltage, V_{POS}	6.5 V
Input Power (50 Ω Impedance)	18 dBm
Internal Power Dissipation (Pad Soldered to Ground)	400 mW
Maximum Junction Temperature	150°C
Operating Temperature Range	–40°C to +105°C
Storage Temperature Range	–65°C to +150°C

Stresses at or above those listed under Absolute Maximum Ratings may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

THERMAL RESISTANCE

Table 3 lists the junction-to-air thermal resistance (θ_{JA}) and the junction-to-case thermal resistance (θ_{JC}) for the ADL5545.

Table 3. Thermal Resistance

Package Type	θ_{JA} ¹	θ_{JC} ²	Unit
3-Lead SOT-89 (RK-3)	53	15	°C/W

¹ Measured on the ADL5545 evaluation board. For more information about board layout, see the Soldering Information and Recommended PCB Land Pattern section.

² Based on simulation with a standard JEDEC board per JESD51.

ESD CAUTION



ESD (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

PIN CONFIGURATION AND FUNCTION DESCRIPTIONS

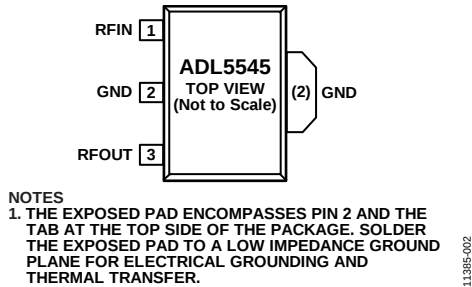


Figure 2. Pin Configuration

Table 4. Pin Function Descriptions

Pin No.	Mnemonic	Description
1	RFIN	RF Input. This pin requires a dc blocking capacitor.
2	GND	Ground. Connect this pin to a low impedance ground plane.
3	RFOUT	RF Output and Supply Voltage. DC bias is provided to this pin through an inductor that is connected to the external power supply. The RF path requires a dc blocking capacitor.
	EPAD	Exposed Pad. The exposed pad encompasses Pin 2 and the tab at the top side of the package. Solder the exposed pad to a low impedance ground plane for electrical grounding and thermal transfer.

TYPICAL PERFORMANCE CHARACTERISTICS

500 MHz TO 4 GHz FREQUENCY BAND

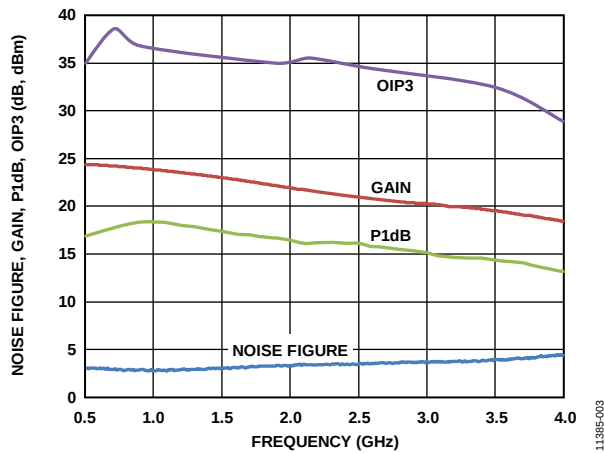


Figure 3. Noise Figure, Gain, P1dB, and OIP3 vs. Frequency

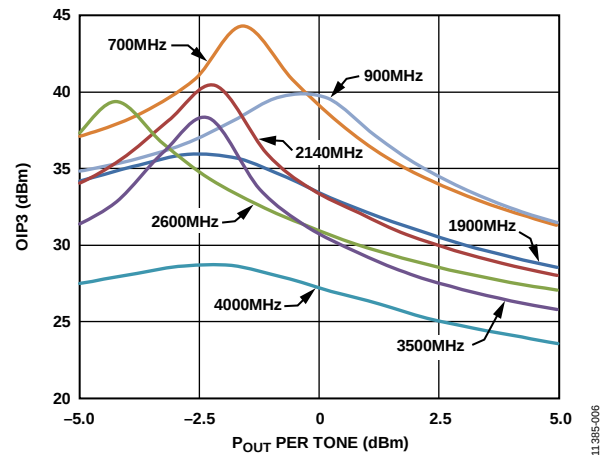
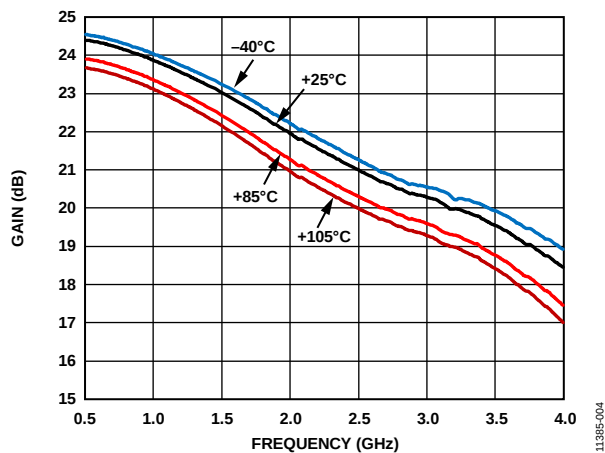
Figure 6. OIP3 vs. Output Power (P_{out}) and Frequency

Figure 4. Gain vs. Frequency and Temperature

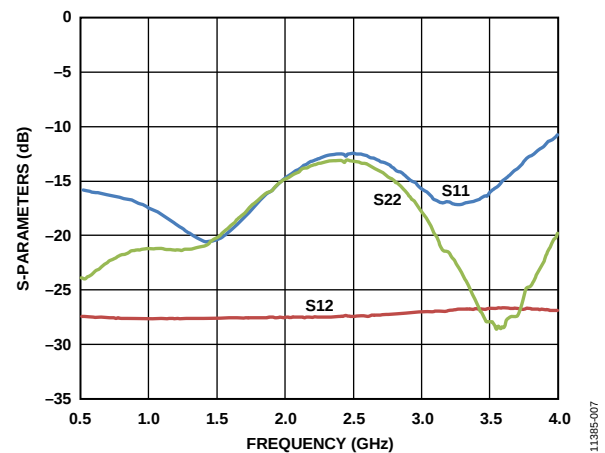
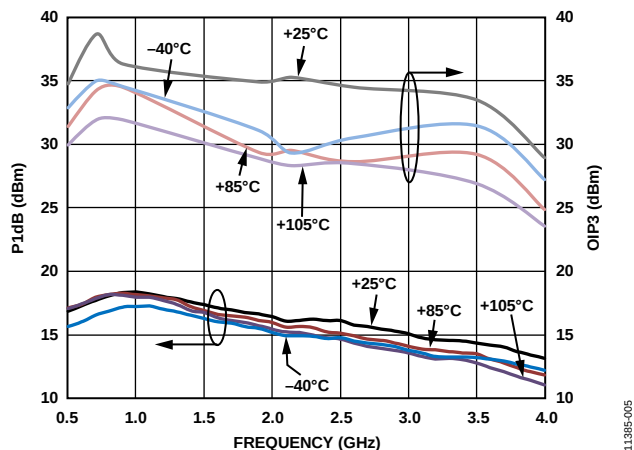
Figure 7. Input Return Loss (S_{11}), Output Return Loss (S_{22}), and Reverse Isolation (S_{12}) vs. Frequency

Figure 5. P1dB and OIP3 vs. Frequency and Temperature

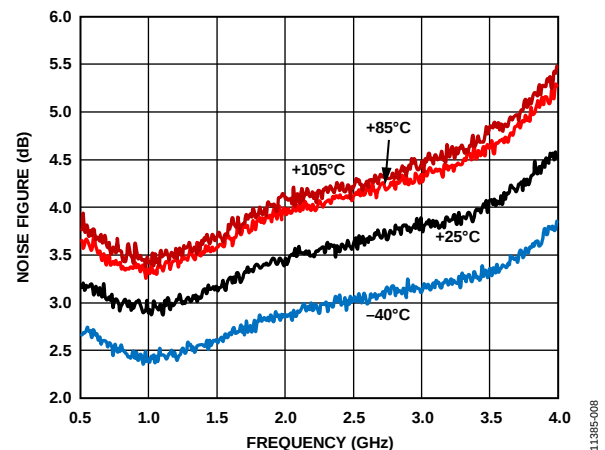


Figure 8. Noise Figure vs. Frequency and Temperature

100 MHz TO 500 MHz FREQUENCY BAND

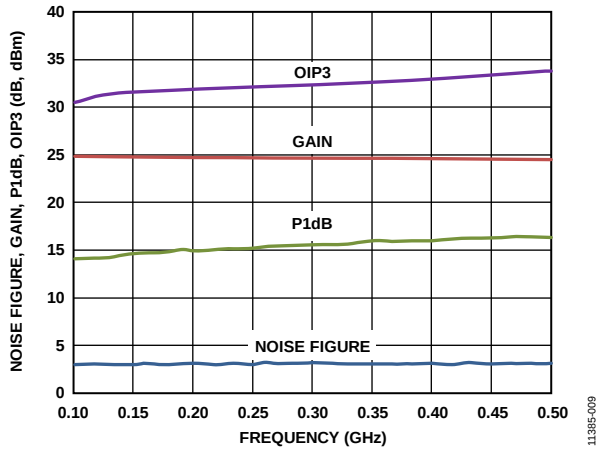


Figure 9. Noise Figure, Gain, P1dB, and OIP3 vs. Frequency, Low Frequency Configuration

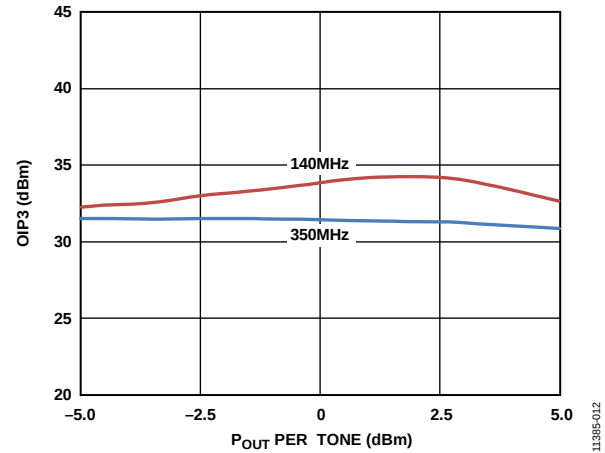


Figure 12. OIP3 vs. Output Power (P_{OUT}) and Frequency, Low Frequency Configuration

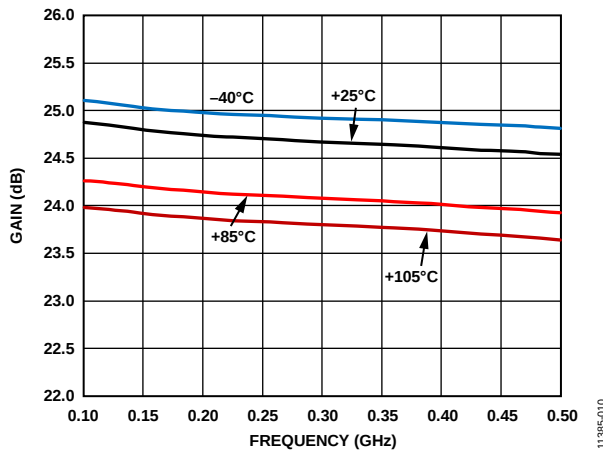


Figure 10. Gain vs. Frequency and Temperature, Low Frequency Configuration

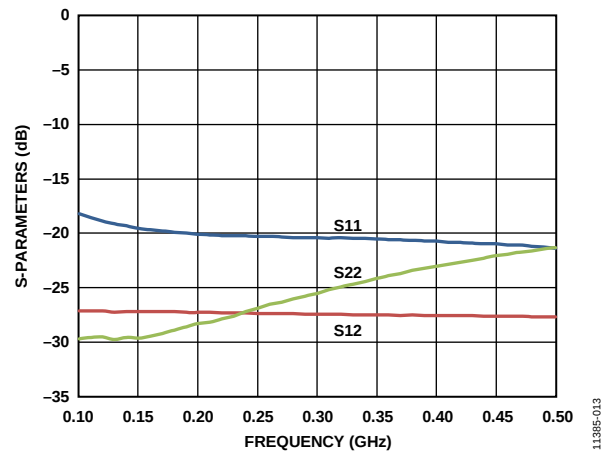


Figure 13. Input Return Loss (S_{11}), Output Return Loss (S_{22}), and Reverse Isolation (S_{12}) vs. Frequency, Low Frequency Configuration

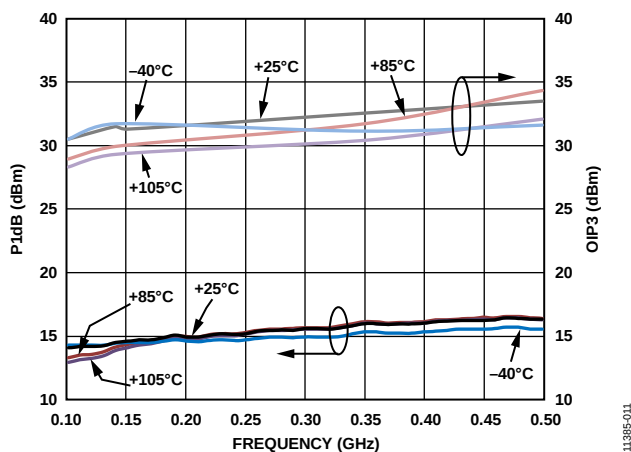


Figure 11. P1dB and OIP3 vs. Frequency and Temperature, Low Frequency Configuration

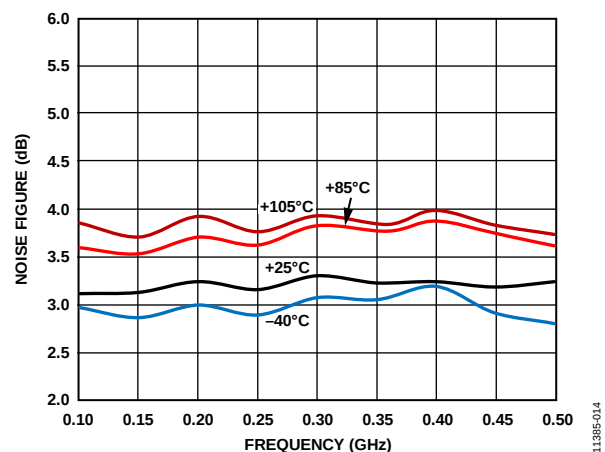


Figure 14. Noise Figure vs. Frequency and Temperature, Low Frequency Configuration

4 GHz TO 6 GHz FREQUENCY BAND

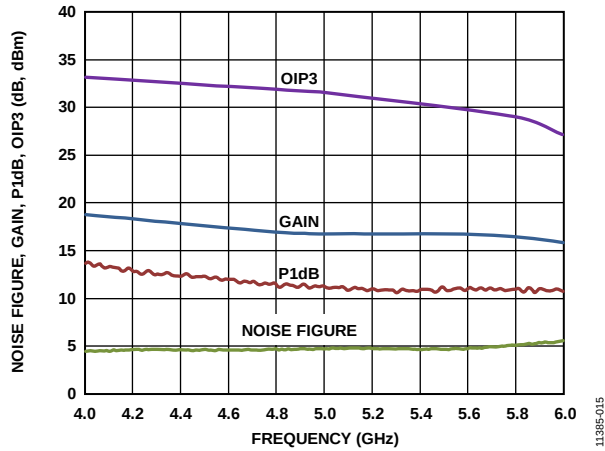


Figure 15. Noise Figure, Gain, P1dB, and OIP3 vs. Frequency, High Frequency Configuration

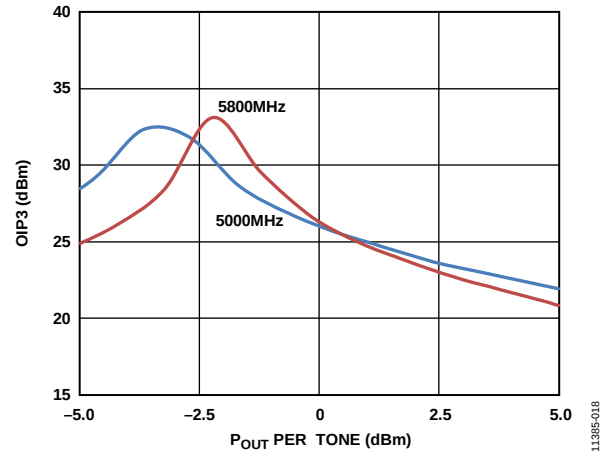


Figure 18. OIP3 vs. Output Power (P_{OUT}) and Frequency, High Frequency Configuration

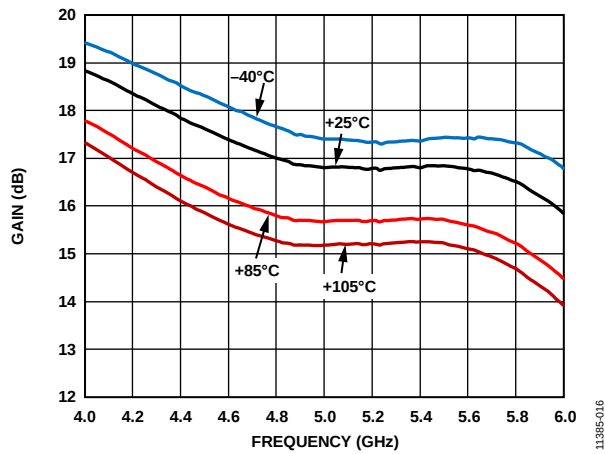


Figure 16. Gain vs. Frequency and Temperature, High Frequency Configuration

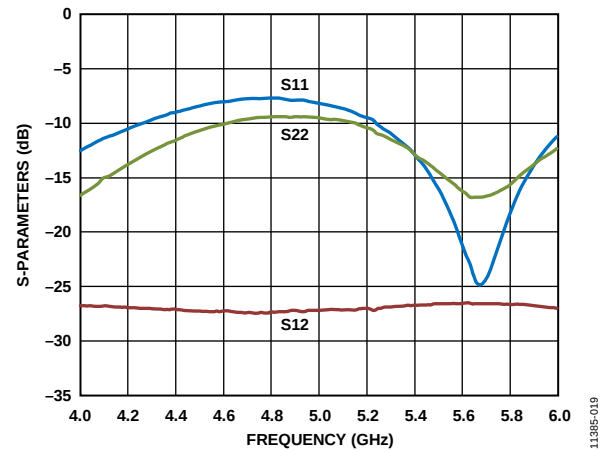


Figure 19. Input Return Loss (S11), Output Return Loss (S22), and Reverse Isolation (S12) vs. Frequency, High Frequency Configuration

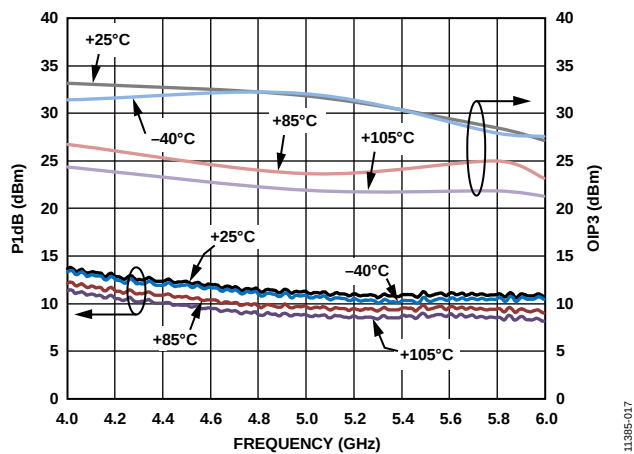


Figure 17. P1dB and OIP3 vs. Frequency and Temperature, High Frequency Configuration

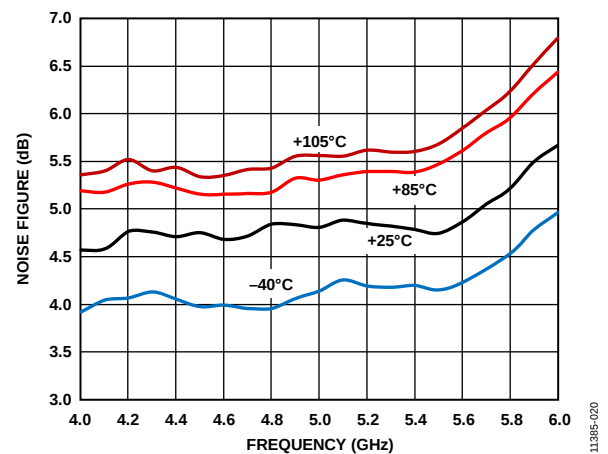


Figure 20. Noise Figure vs. Frequency and Temperature, High Frequency Configuration

GENERAL

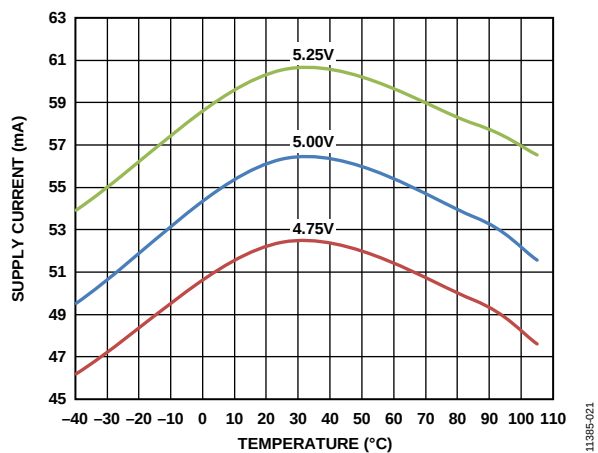


Figure 21. Supply Current vs. Temperature

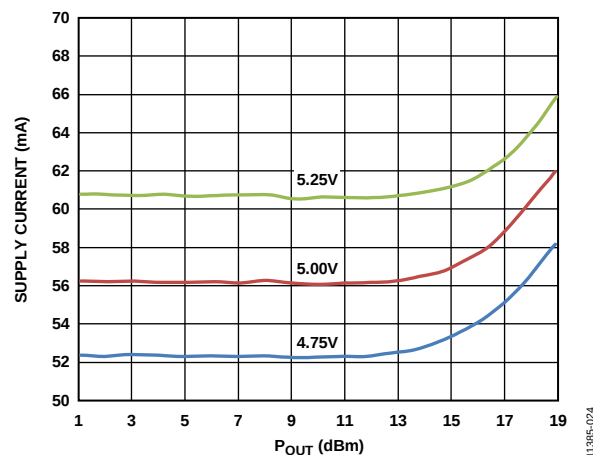
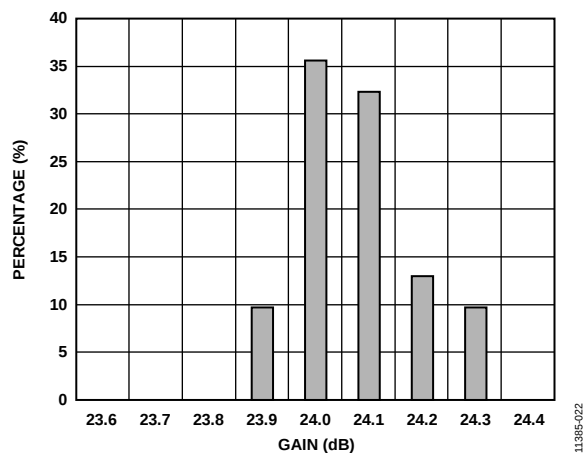
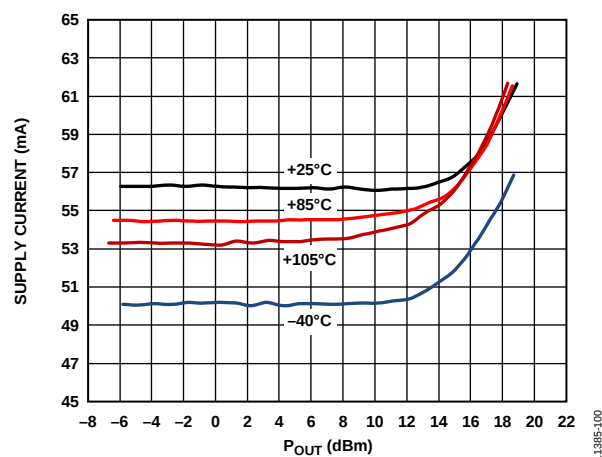
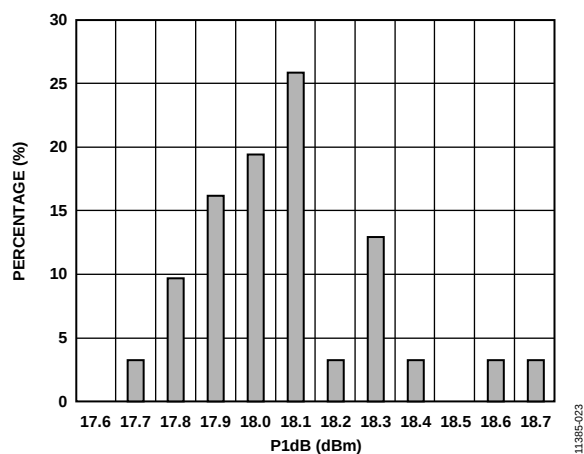
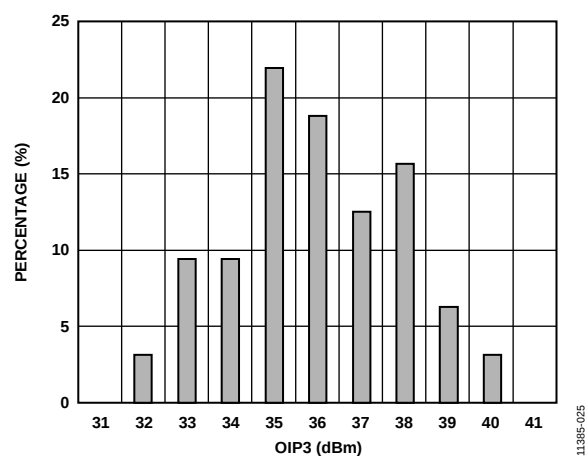
Figure 24. Supply Current vs. P_{OUT} at 900 MHz

Figure 22. Gain Distribution at 900 MHz

Figure 25. Supply Current vs. P_{OUT} and Temperature at 900 MHzFigure 23. P_{1dB} Distribution at 900 MHzFigure 26. $OIP3$ Distribution at 900 MHz, $P_{OUT} = -3$ dBm per Tone

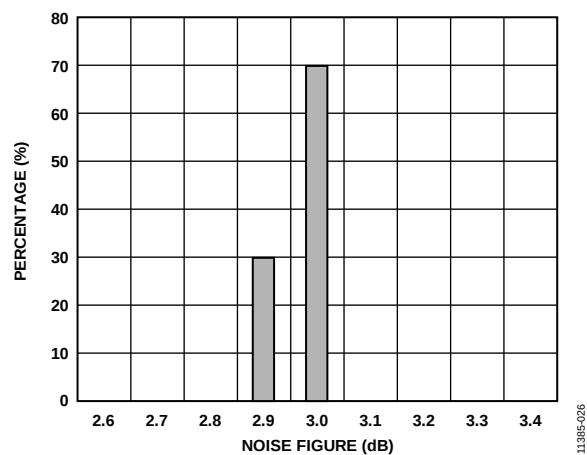


Figure 27. Noise Figure Distribution at 900 MHz

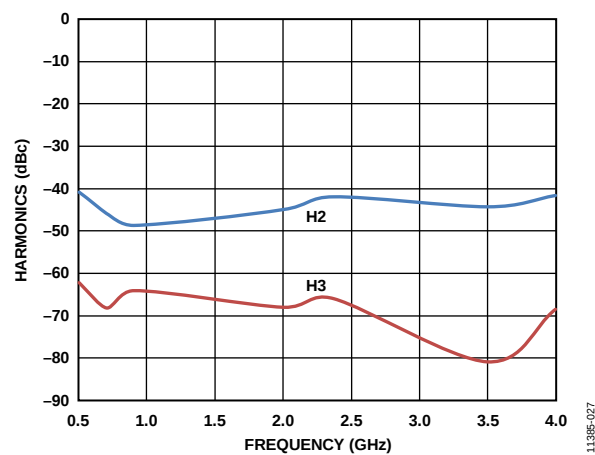


Figure 28. Single-Tone Harmonics vs. Frequency, $P_{OUT} = 0$ dBm

APPLICATIONS INFORMATION

BASIC CONNECTIONS

Figure 29 shows the basic connections for operating the ADL5545. The device supports operation from 30 MHz to 6 GHz. However, for optimal performance at lower and higher frequency bands, the board configuration must be adjusted. Table 5 lists the recommended board configuration to operate the device at various frequency bands.

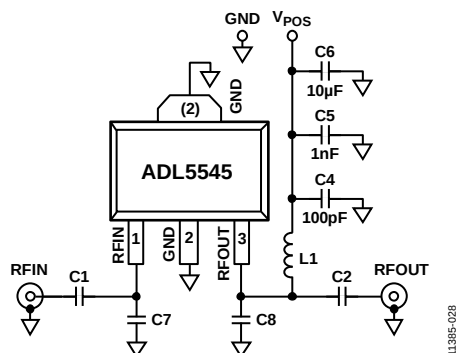


Figure 29. Basic Connections

A 5 V dc bias is supplied to the amplifier through the bias inductor connected to RFOUT (Pin 3). The bias voltage must be decoupled using 100 pF, 1 nF, and 10 µF power supply decoupling capacitors. The typical current consumption for the ADL5545 is 56 mA.

At low and high frequencies, the device exhibits improved performance with the suggested setup configuration listed in Table 5. Figure 30 to Figure 33 provide a comparison of the performance of the device at the 100 MHz to 500 MHz and 4 GHz to 6 GHz bands when driven with the optimal setup configuration and the default setup configuration.

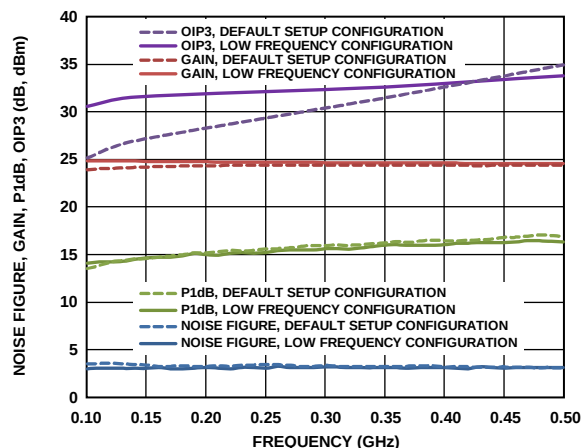


Figure 30. Noise Figure, Gain, P1dB, and OIP3 vs. Frequency, 100 MHz to 500 MHz, Comparison of Performance with the Optimized Settings and the Default Configuration

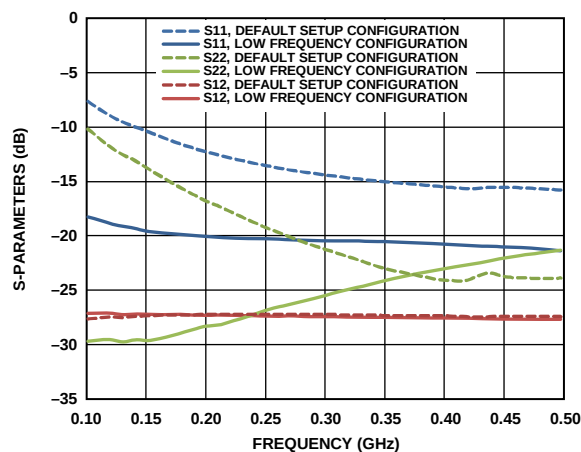


Figure 31. Return Loss and Reverse Isolation, 100 MHz to 500 MHz, Comparison of Performance with the Optimized Settings and the Default Configuration

Table 5. Recommended Components for Basic Connections

Frequency Band	AC Coupling Capacitors (0402)		DC Bias Inductor (0603HP)	High Frequency Matching Capacitors (0402)	
	C1	C2	L1	C7	C8
100 MHz to 500 MHz	100 nF	100 nF	1000 nH	Do not install	Do not install
500 MHz to 4 GHz (default)	100 pF	100 pF	100 nH	Do not install	Do not install
4 GHz to 6 GHz	100 pF	100 pF	12 nH	0.1 pF	0.1 pF

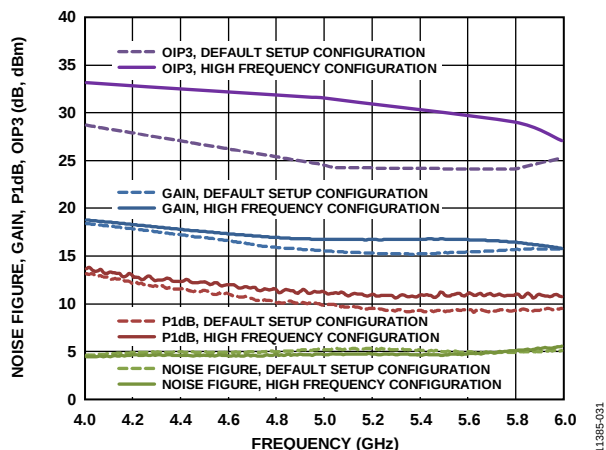


Figure 32. Noise Figure, Gain, P1dB, and OIP3 vs. Frequency, 4 GHz to 6 GHz, Comparison of Performance with the Optimized Settings and the Default Configuration

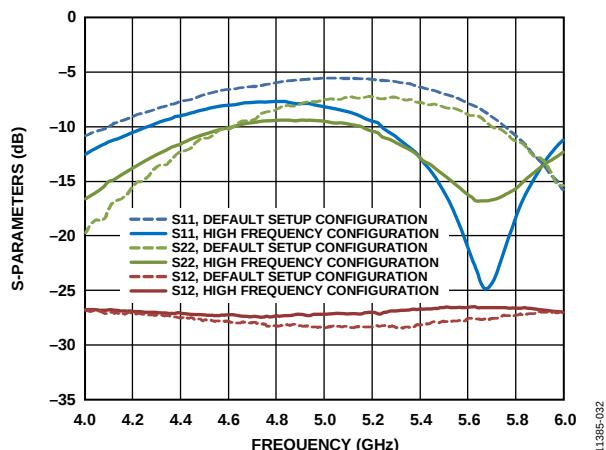


Figure 33. Return Loss and Reverse Isolation, 4 GHz to 6 GHz, Comparison of Performance with the Optimized Settings and the Default Configuration

SOLDERING INFORMATION AND RECOMMENDED PCB LAND PATTERN

Figure 34 shows the recommended land pattern for the ADL5545. To minimize thermal impedance, the exposed pad on the underside of the SOT-89 package is soldered to a ground plane, along with Pin 2. If multiple ground layers exist, stitch the layers together using vias.

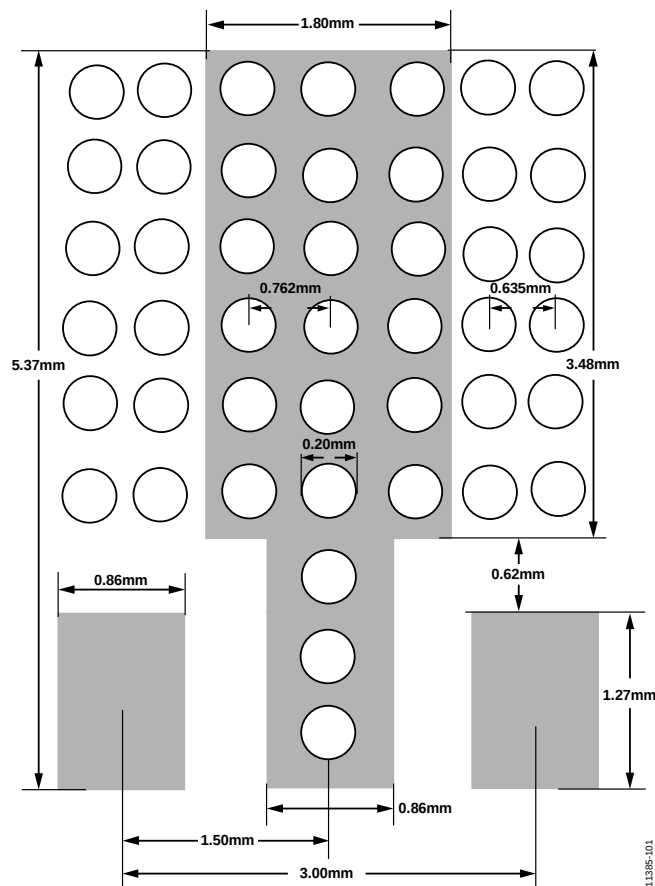


Figure 34. Recommended Land Pattern

The land pattern on the ADL5545 evaluation board provides a measured thermal resistance (θ_{JA}) of 53°C/W. To measure θ_{JA} , the temperature at the top of the SOT-89 package is found with an IR temperature gun. Thermal simulation suggests a junction temperature 10°C higher than the top-of-package temperature. With additional measurements of the ambient temperature and I/O power, θ_{JA} can be determined.

OPERATION DOWN TO 30 MHz

To operate the ADL5545 at frequencies below 100 MHz, a feedback network must be implemented between the input and output ports of the device to ensure stability. Figure 35 shows a sample configuration used to evaluate the device at frequencies below 100 MHz. Figure 36 to Figure 38 demonstrate the performance of the device in this configuration.

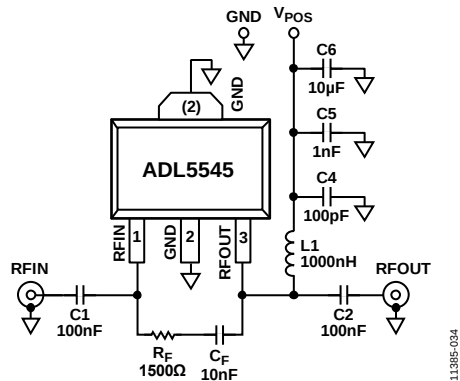


Figure 35. Setup for Low Frequency Operation Down to 30 MHz

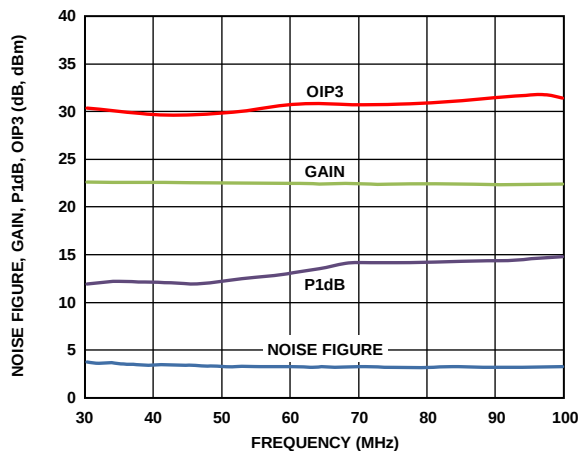


Figure 36. Noise Figure, Gain, P1dB, and OIP3 vs. Frequency, 30 MHz to 100 MHz

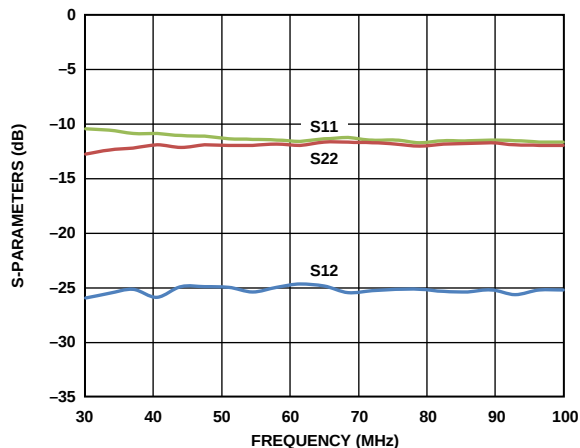


Figure 37. Return Loss and Reverse Isolation, 30 MHz to 100 MHz

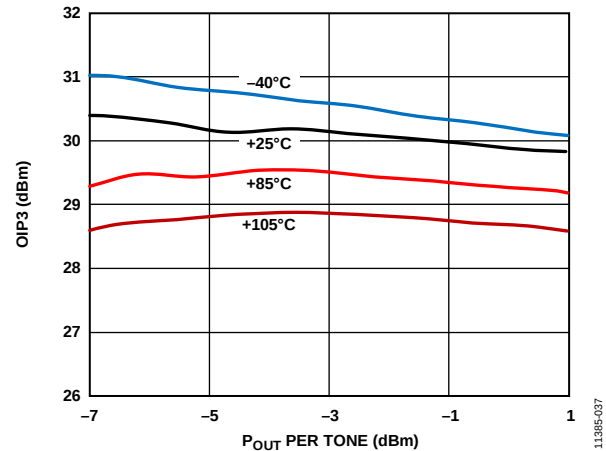


Figure 38. OIP3 vs. POUT at 30 MHz

W-CDMA ACPR PERFORMANCE

Figure 39 shows a plot of the adjacent channel power ratio (ACPR) vs. POUT for the ADL5545. The signal type used is a single wideband code division multiple access (W-CDMA) carrier (Test Model 1-64) at 2140 MHz. This signal is generated by a very low ACPR source. ACPR is measured at the output by a high dynamic range spectrum analyzer that incorporates an instrument noise-correction function.

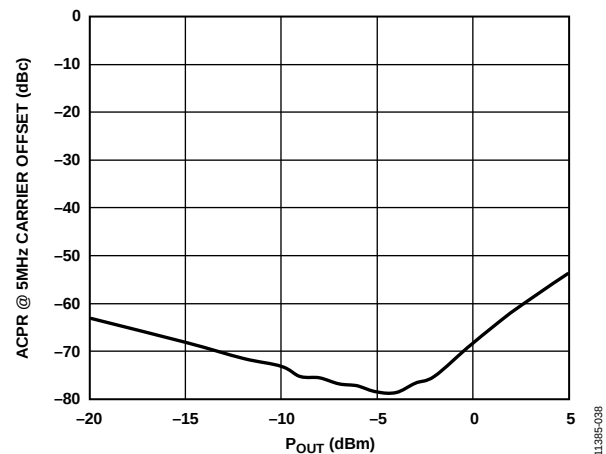


Figure 39. ACPR vs. POUT, Single W-CDMA Carrier (Test Model 1-64) at 2140 MHz

The ADL5545 achieves an ACPR of -79 dBc at an output power level of -4 dBm, at which point device noise and not distortion begins to dominate the power in the adjacent channels. At an output power level of 0 dBm, ACPR is still very low at -69 dBc.

EVALUATION BOARD

Figure 40 shows the ADL5545 evaluation board layout. Figure 41 shows the schematic for the evaluation board. The board is powered by a single 5 V supply. Table 6 lists the components used on the evaluation board. Power can be applied to the board through clip-on leads (V_{SUP} , GND).

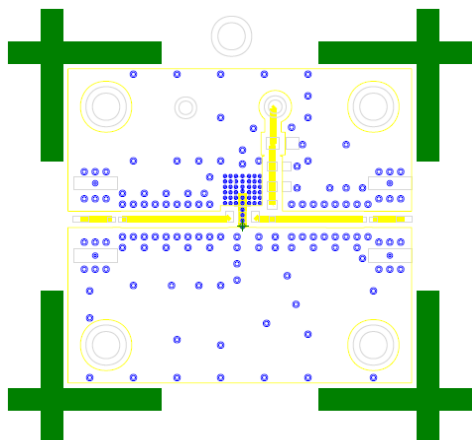


Figure 40. Evaluation Board Layout (Top)

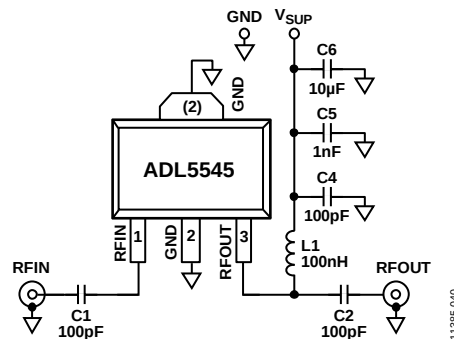


Figure 41. Evaluation Board Schematic

Table 6. Evaluation Board Configuration Options

Component	Function	Default Value
C1, C2	AC coupling capacitors	100 pF, 0402
L1	DC bias inductor	100 nH, 0603 (Coilcraft 0603HP or equivalent)
V_{SUP} and GND	Clip-on terminals for power supply	
C4, C5, C6	Power supply decoupling capacitors	C4: 100 pF, 0603 C5: 1 nF, 0603 C6: 10 μ F, 1206

OUTLINE DIMENSIONS

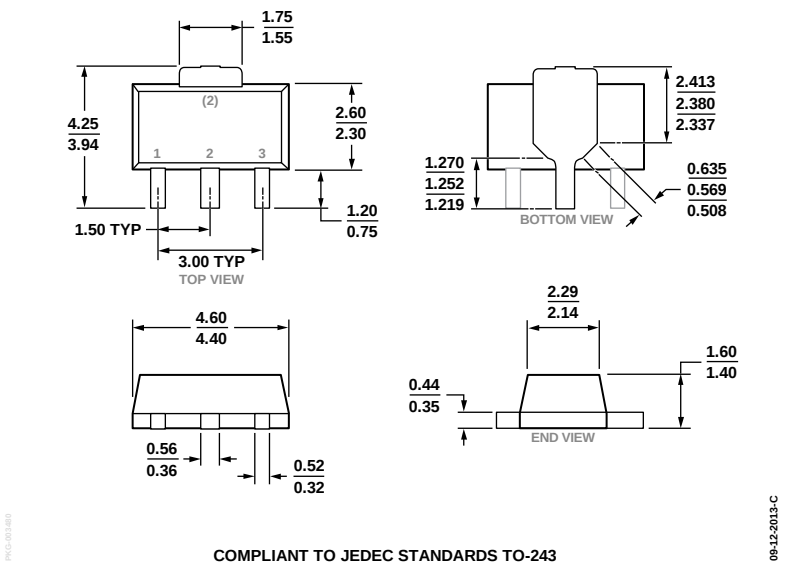


Figure 42. 3-Lead Small Outline Transistor Package [SOT-89]
(RK-3)
Dimensions shown in millimeters

ORDERING GUIDE

Model ¹	Temperature Range	Package Description	Package Option
ADL5545ARKZ-R7	−40°C to +105°C	3-Lead SOT-89, 7" Tape and Reel	RK-3
ADL5545-EVALZ	−40°C to +105°C	Evaluation Board	

¹ Z = RoHS Compliant Part.

NOTES

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